

PART NUMBER: KX132-1211 Rev. 1.0

31-Jul-2019

### **Product Description**

The KX132-1211 is a tri-axis ±2g, ±4g, ±8g, or ±16g silicon micromachined accelerometer featuring a user-configurable 3-stage Advanced Data Path (ADP) consisting of a low-pass filter, low-pass/high-pass filter, and RMS calculation engine. The KX132-1211 accelerometer also features an advanced Wake-Up and Back-to-Sleep detection with a high-resolution threshold capability configurable down to 3.9 mg, 512-byte buffer that continues to record data even when being read, as well as embedded engines for orientation, Directional-Tap<sup>™</sup>/Double-Tap<sup>™</sup>, and Free fall detection. The sense element is fabricated using Kionix's proprietary plasma micromachining process technology. Acceleration sensing is based on the principle of a differential capacitance arising from acceleration-induced motion of the sense element, which further utilizes common mode



cancellation to decrease errors from process variation, temperature, and environmental stress. The sense element is hermetically sealed at the wafer level by bonding a second silicon lid wafer to the device wafer. A separate ASIC device packaged with the sense element provides signal conditioning and intelligent user-programmable application algorithms. The KX132-1211 accelerometers offer lower noise and improved linearity over of the entire temperature range. The accelerometer is delivered in a 2 x 2 x 0.9 mm LGA 12-pin plastic package operating from a 1.7V - 3.6V (VDD) / 1.2V - 3.6V (IO\_VDD) DC supplies. Internal voltage regulators are used to maintain constant internal operating voltages over the range of input supply voltages. This results in stable operating characteristics even if the supply voltage changes. I<sup>2</sup>C or SPI digital protocol is supported to configure the chip, read acceleration outputs, and check for updates to the orientation, Directional-Tap<sup>TM</sup>/Double-Tap<sup>TM</sup> detection, Free fall detection, and activity monitoring algorithms. Two configurable interrupt pins are also available to show the output of the embedded detection algorithms.

### **Features**

- Operating temperature range from -40°C to +105°C
- Small footprint: 2 x 2 x 0.9 mm LGA 12-pin package
- User-configurable g-range up to ±16g and Output Data Rate up to 25600Hz
- A user-configurable 3-stage Advanced Data Path (ADP) consisting of a low-pass filter, low-pass/high-pass filter, and RMS calculation engine.
- High resolution Wake-Up / Back-to-Sleep functions with threshold configurable down to 3.9 mg
- User accessible manufacturer and part ID registers
- Self-test Function

- Integrated Free fall, Directional-Tap<sup>™</sup> / Double-Tap<sup>™</sup>, and Device-orientation algorithms
- Improved ODR accuracy in Low Power mode over temperature
- Embedded 512-byte FIFO buffer continues to record data even when being read
- User-selectable Low Power or High-Performance modes
- Internal voltage regulator
- Digital I<sup>2</sup>C up to 3.4MHz and Digital SPI up to 10MHz
- Excellent temperature performance with high shock survivability
- RoHS / REACH compliant



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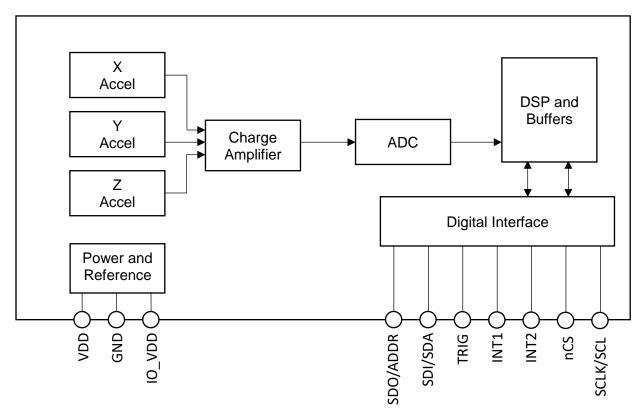
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### **Functional Diagram**





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### **Product Specifications**

### Mechanical

(specifications are for operation at 2.5V and T = 25C unless stated otherwise)

	rameters	Units	Min	Typical	Max
Operating Temperatur	e Range	°C	-40	-	+105
Zero-g Offset		mg		±25	±90
Zero-g Offset Variation	n from RT over Temp.	mg/ºC		0.25	
	GSEL1=0, GSEL0=0 (±2g)		15401	16384	17367
Concitivity 1 (16 hit)	GSEL1=0, GSEL0=1 (±4g)	oounto/a	7700	8192	8684
Sensitivity <sup>1</sup> (16 bit)	GSEL1=1, GSEL0=0 (±8g)	counts/g	3850	4096	4342
	GSEL1=1, GSEL0=1 (±16g)		1925	2048	2171
	GSEL1=0, GSEL0=0 (±2g)			64	
Sensitivity	GSEL1=0, GSEL0=1 (±4g)	oounto/a		32	
(Buffer 8-bit mode) <sup>1,2</sup>	GSEL1=1, GSEL0=0 (±8g)	counts/g		16	
	GSEL1=1, GSEL0=1 (±16g)			8	
Sensitivity Variation fro	om RT over Temperature	%/°C		0.01 (xy) 0.03 (z)	
Positive Self-Test Outp	out change on Activation <sup>3</sup>	g	0.25 (xy) 0.2 (z)	0.5	0.75
Mechanical Signal Bar	ndwidth (-3dB) <sup>4</sup>	Hz		4200 (xy) 2900 (z)	
Non-Linearity		% of FS		0.5	
Cross Axis Sensitivity		%		2	
Noise <sup>5</sup>	RMS	mg		0.7	
INOISE	Density	μg/√Hz		130	

Table 1: Mechanical Specifications

### Notes:

- 1. Resolution and acceleration ranges are user selectable via I<sup>2</sup>C or SPI
- 2. Sensitivity is proportional to BRES in BUF\_CNTL2.
- 3. Requires changing of STPOL bit in INC1 register to 1 prior to performing self-test
- 4. Signal bandwidth varies with Output Data Rate (ODR), and Low Pass Filter setting. Measured with ODR = 25600Hz, LPRO = 1 settings.
- Noise varies with ODR, power mode, and the Average Filter Control (AVC) settings.
   Measured with RES = 1, ODR = 50Hz, IIR\_BYPASS = 0, LPRO = 1 settings.



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### **Electrical**

(specifications are for operation at 2.5V and T = 25C unless stated otherwise)

	Parameters	Units	Min	Typical	Max
Supply Voltage (VDD)	Operating	V	1.7	2.5	3.6
I/O Pads Supply	SPI, I <sup>2</sup> C (Fast/Standard mode)	V	1.2		3.6
Voltage (IO_VDD)	I <sup>2</sup> C (High Speed mode)	V	1.7		3.6
	Operating (High Performance with Wake-up Detection) ODR=400Hz			148	
Current Consumption (Accelerometer Only)	Topolating (Low power with Warte up detection)			0.53	
	Operating (Low Power with Wake-up detection plus Advanced Data Path) ODR=0.781Hz <sup>7</sup>			0.67	
Standby Current Cons	umption	μΑ		0.5	
Output Low Voltage	(IO_VDD < 2V)	V	-	-	0.2 * IO_VDD
(VoL) <sup>2</sup>	(IO_VDD ≥ 2V)	V	-	-	0.4
Output High Voltage (\	/он)	<b>V</b>	0.8 * IO_VDD	-	-
Input Low Voltage (VIL)		V	-	-	0.2 * IO_VDD
Input High Voltage (V⊪	4)	٧	0.8 * IO_VDD	-	-
Start Up Time <sup>3</sup>		ms	2		1300
Power Up Time <sup>4</sup>		ms		20	50
I <sup>2</sup> C Communication Ra	ate	MHz			3.4
I <sup>2</sup> C Slave Address (7-b	oit)			0x1E / 0x1F	
WHO_AM_I register va	alue			0x3D	
Output Data Rate (OD	R) <sup>5</sup>	Hz	0.781	50	25600
Output Signal Bandwic	dth (-3dB) <sup>6</sup>	Hz		ODR/9 or ODR/2	

Table 2: Electrical Specifications

### Notes:

- 1. Current varies with Output Data Rate (ODR) as shown in Figure 2, types and number of enabled digital engines, the average filter control settings, and VDD. Measured with OWUF<2:0> = 0, OSA<3:0> = 0, AVC<2:0> = 1, LPSTPSEL = 1.
- 2. For I<sup>2</sup>C communication, this assumes a minimum 1.5k $\Omega$  pull-up resistor on SCL and SDA pins.
- 3. Start up time is from PC1 set to valid outputs. Time varies with ODR, Power Mode, and FSTUP bit setting (see Figure 1).
- 4. Power up time is from VDD valid to device boot completion.
- 5. Typical values. ODR is user-selectable via I<sup>2</sup>C or SPI. See ODCNTL register for details.
- 6. Refers to accelerometer's raw output data. Additional bandwidth control is available using the Advanced Data Path (ADP) engine.
- 7. Measured with RMS\_AVC<2:0> = 1, OADP<3:0> = 0, LPSTPSEL = 0.



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### **Start Up Time Profile**

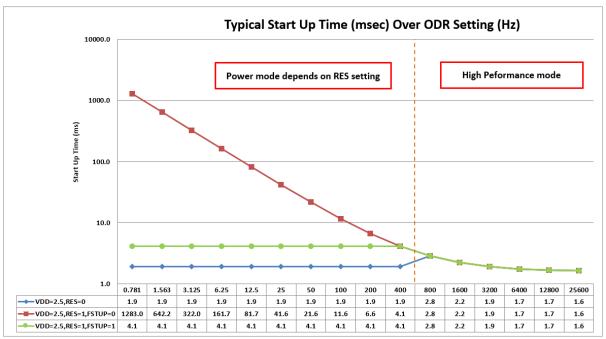


Figure 1: Start Up Time Diagram

#### **Current Profile**

	Typical Current (μA)										
ODR (Hz)	High Performance with Wake-up	High Performance with Wake-up & ADP	Low Power with Wake-up	Low Power with Wake-up & ADP							
Standby	0.50	0.50	0.50	0.50							
0.781	148	148	0.53	0.67							
1.563	148	148	0.60	0.74							
3.125	148	148	0.72	0.83							
6.25	148	148	0.97	1.14							
12.5	148	148	1.5	1.7							
25	148	148	2.4	2.7							
50	148	148	4.4	4.9							
100	148	148	8.5	9.2							
200	148	150	16	18							
400	148	150	31	34							
800	148	152	148	152							
1600	149	154	149	154							
3200	150	160	150	160							
6400	152	172	152	172							
12800	155	194	155	194							
25600	162	239	162	239							

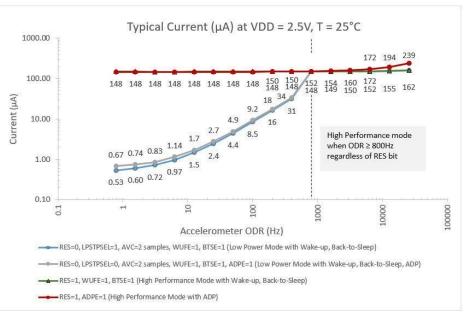


Figure 2: Current as a function of Output Data Rate (ODR) and Power Mode Settings



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### **Power-On Procedure**

Proper functioning of power-on reset (POR) is dependent on the specific VDD,  $VDD_{LOW}$ ,  $T_{VDD}$  (rise time), and  $T_{VDD\_OFF}$  profile of individual applications. It is recommended to minimize  $VDD_{LOW}$ , and  $T_{VDD}$ , and maximize  $T_{VDD\_OFF}$ . It is also advised that the VDD ramp up time  $T_{VDD}$  be monotonic. Note that the outputs will not be stable until VDD has reached its final value.

To assure proper POR, the application should be evaluated over the customer specified range of VDD, VDD<sub>LOW</sub>,  $T_{VDD}$ ,  $T_{VDD\_OFF}$  and temperature as POR performance can vary depending on these parameters.

Please refer to Technical Note **TN027 Power-On Procedure** for more information.



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#### **Environmental**

Paran	neters	Units	Min	Typical	Max
Supply Voltage (VDD)	Absolute Limits	V	-0.3	-	3.60
Operating Temperatur	e Range	°C	-40	-	+105
Storage Temperature	Range	°C	-55	ı	+150
Mech. Shock (powered	d and unpowered)	g		-	5000 for 0.5ms 10000 for 0.2ms
ESD	НВМ	V	-	-	2000

Table 3: Environmental Specifications



Caution: ESD Sensitive and Mechanical Shock Sensitive Component, improper handling can cause permanent damage to the device.







This product is in conformance with RoHS directive, REACH regulation, and is Halogen-Free. For the current certificate of compliance, visit www.kionix.com website.

### Handling, Mounting, Soldering

For package handling, mounting, and soldering guidelines, see <u>TN007 Package Handling, Mounting, and Soldering Guidelines</u> technical note.

#### Floor Life

Factory floor life exposure of the KX132-1211 reels removed from the moisture barrier bag should not exceed a maximum of 168 hours at 30C/60%RH. If this floor life is exceeded, the parts should be dried per the IPC/JEDEC J-STD-033D standard (or latest revision).



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### **Terminology**

g

A unit of acceleration equal to the acceleration of gravity at the earth's surface.

$$1g = 9.8 \frac{m}{s^2}$$

One thousandth of a g (0.0098 m/s<sup>2</sup>) is referred to as 1 milli-g (1 mg).

### Sensitivity

The sensitivity of an accelerometer is the change in output per unit of input acceleration at nominal VDD and temperature. The term is essentially the gain of the sensor expressed in counts per g (counts/g) or LSB's per g (LSB/g). Occasionally, sensitivity is expressed as a resolution, i.e. milli-g per LSB (mg/LSB) or milli-g per count (mg/count). Sensitivity for a given axis is determined by measurements of the formula:

$$Sensitivity = \frac{\left(Output@+1g - Output@-1g\right)}{2g}$$

The sensitivity tolerance describes the range of sensitivities that can be expected from a large population of sensors at room temperature and over life. When the temperature deviates from room temperature (25°C), the sensitivity will vary by the amount shown in Table 1.

### Zero-g offset

Zero-g offset or 0-g offset describes the actual output of the accelerometer when no acceleration is applied. Ideally, the output would always be in the middle of the dynamic range of the sensor (content of the OUTX, OUTY, OUTZ registers = 00, expressed as a 2's complement number). However, because of mismatches in the sensor, calibration errors, and mechanical stress, the output can deviate from 00. This deviation from the ideal value is called 0-g offset. The zero-g offset tolerance describes the range of 0-g offsets of a population of sensors over the operating temperature range.

#### Self-test

Self-test allows a functional test of the sensor without applying a physical acceleration to it. When activated, an electrostatic force is applied to the sensor, simulating an input acceleration. The sensor outputs respond accordingly. If the output signals change within the amplitude specified in Table 1, then the sensor is working properly and the parameters of the interface chip are within the defined specifications.



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### **Functionality**

#### Sense element

The sense element is fabricated using Kionix's proprietary plasma micromachining process technology. This process technology allows Kionix to create mechanical silicon structures, which are essentially mass-spring systems that move in the direction of the applied acceleration. Acceleration sensing is based on the principle of a differential capacitance arising from the acceleration-induced motion. Capacitive plates on the moving mass move relative to fixed capacitive plates anchored to the substrate. The sense element is hermetically sealed at the wafer level by bonding a second silicon lid wafer to the device using a glass frit.

#### **ASIC** interface

A separate ASIC device packaged with the sense element provides all of the signal conditioning and communication with the sensor. The complete measurement chain is composed by a low-noise capacitance to voltage amplifier, which converts the differential capacitance of the MEMS sensor into an analog voltage that is sent through an analog-to-digital converter. The acceleration data may be accessed through the I<sup>2</sup>C or SPI digital communications provided by the ASIC. In addition, the ASIC contains all of the logic to allow the user to choose data rates, g-ranges, filter settings, and interrupt logic.

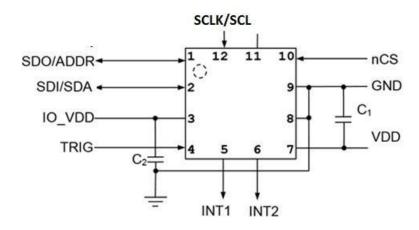
### **Factory calibration**

Kionix trims the offset and sensitivity of each accelerometer by adjusting gain (sensitivity) and 0-g offset trim codes stored in nonvolatile memory (OTP). Additionally, all functional register default values are also programmed into the nonvolatile memory. Every time the device is turned on or a software reset command is issued, the trimming parameters and default register values are downloaded into the volatile registers to be used during active operation. This allows the device to function without further calibration.



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### **Application Schematic**



### **Pin Description**

Pin	Name	Description
1	SDO/ADDR	Serial Data Out pin during 4-wire SPI communication and part of the device address during I2C communication. Do not leave floating.
2	SDI/SDA	SPI Data input / I2C Serial Data
3	IO_VDD	The power supply input for the digital communication bus. Optionally decouple this pin to ground with a 0.1uF ceramic capacitor.
4	TRIG	Trigger pin for FIFO buffer control. Connect to GND when not using external trigger option.
5	INT1	Physical Interrupt 1 (Push-Pull). The pin is in High-Z state during POR and is driven LOW following POR. Leave floating if not used.
6	INT2	Physical Interrupt 2 (Push-Pull). The pin is in High-Z state during POR and is driven LOW following POR. Leave floating if not used.
7	VDD	The power supply input. Decouple this pin to ground with a 0.1uF ceramic capacitor.
8	GND	Ground
9	GND	Ground
10	nCS	Chip Select (active LOW) for SPI communication. Connect to IO_VDD for I2C communication. Do not leave floating.
11	NC	Not Internally Connected. Can be connected to VDD, IO_VDD, GND or leave floating.
12	SCLK/SCL	SPI and I2C Serial Clock

Table 4: Pin Description

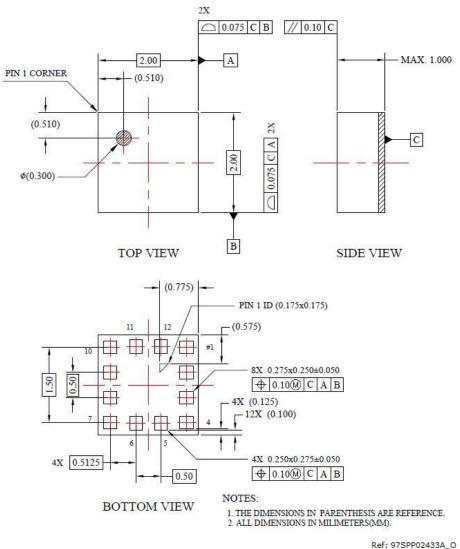


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### **Package Dimensions and Orientation**

#### **Dimensions**

2 x 2 x 0.9 mm LGA 12-pin

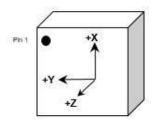


All dimensions and tolerances conform to ASME Y14.5M-1994



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### Orientation



When device is accelerated in +X, +Y or +Z direction, the corresponding output will increase.

### Static X/Y/Z Output Response versus Orientation to Earth's surface (1g):

GSEL1=0, GSEL0=0 (±2g)

Position	1		2		3		4		5		6	
Diagram						Top Bottom		ı	Bottom Top			
Resolution (bits)	16	8	16	8	16	8	16	8	16	8	16	8
X (counts)	+16384	+64	0	0	-16384	-64	0	0	0	0	0	0
Y (counts)	0	0	-16384	-64	0	0	+16384	+64	0	0	0	0
Z (counts)	0	0	0	0	0	0	0	0	+16384	+64	-16384	-64
X-Polarity	+		0		-		0		0		0	
Y-Polarity	0		-		0		+		0		0	
Z-Polarity	0		0		0		0		+		-	



Earth's Surface



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### Static X/Y/Z Output Response versus Orientation to Earth's surface (1g):

GSEL1=0, GSEL0=1 (±4g)

Position	1		2		3		4		5		6	
Diagram							Top Bottom		Bottom Top			
Resolution (bits)	16	8	16	8	16	8	16	8	16	8	16	8
X (counts)	+8192	+32	0	0	-8192	-32	0	0	0	0	0	0
Y (counts)	0	0	-8192	-32	0	0	+8192	+32	0	0	0	0
Z (counts)	0	0	0	0	0	0	0	0	+8192	+32	-8192	-32
X-Polarity	+		0		-		0		0		0	
Y-Polarity	0		-		0		+		0		0	
Z-Polarity	0		0		0		0		+		-	

(1g)

Earth's Surface

### Static X/Y/Z Output Response versus Orientation to Earth's surface (1g):

GSEL1=1, GSEL0=0 (±8g)

Position	1		2		3	3		4			6	
Diagram									Top Bottom		Bottom Top	
Resolution (bits)	16	8	16	8	16	8	16	8	16	8	16	8
X (counts)	+4096	+16	0	0	-4096	-16	0	0	0	0	0	0
Y (counts)	0	0	-4096	-16	0	0	+4096	+16	0	0	0	0
Z (counts)	0	0	0	0	0	0	0	0	+4096	+16	-4096	-16
X-Polarity	+		0		-		0		0		0	
Y-Polarity	0		0		0		+ 0		0		0	
Z-Polarity			U		U	L	(1g)		+		-	

Earth's Surface



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### Static X/Y/Z Output Response versus Orientation to Earth's surface (1g):

GSEL1=1, GSEL0=1 (±16g)

3 16 8 0	8	16	8	16	8	Top Botton		Botton Top	n   
		16	8	16	Q	40			
8 0	0		l l		٥	16	8	16	8
	1	-2048	-8	0	0	0	0	0	0
-2048	-8	0	0	+2048	+8	0	0	0	0
0	0	0	0	0	0	+2048	+8	-2048	-8
0	)	-		0		0		0	
-		0		+		0		0	
0	)	0		0		+		-	
		-	- 0	- 0	- 0 + 0 0 0	- 0 + 0 0 0	- 0 + 0	- 0 + 0 0 0 0 +	- 0 + 0 0 0 0 0 + -

Earth's Surface



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### **Digital Interface**

The Kionix KX132-1211 digital accelerometer can communicate via the I<sup>2</sup>C and SPI digital serial interface protocols. This allows for easy system integration by eliminating analog-to-digital converter requirements and by providing direct communication with system micro-controllers.

The serial interface terms and descriptions as indicated in Table 5 below will be observed throughout this document.

Term	Description
Transmitter	The device that transmits data to the bus.
Receiver	The device that receives data from the bus.
Master	The device that initiates a transfer, generates clock signals, and terminates a transfer.
Slave	The device addressed by the Master.

Table 5: Serial Interface Terminologies

### I<sup>2</sup>C Serial Interface

As previously mentioned, the KX132-1211 can communicate on an I<sup>2</sup>C bus. I<sup>2</sup>C is primarily used for synchronous serial communication between a Master device and one or more Slave devices. The Master, typically a micro controller, provides the serial clock signal and addresses Slave devices on the bus. The KX132-1211 always operates as a Slave device during standard Master-Slave I<sup>2</sup>C operation.

I<sup>2</sup>C is a two-wire serial interface that contains a Serial Clock (SCL) line and a Serial Data (SDA) line. SCL is a serial clock that is provided by the Master, but can be held LOW by any Slave device, putting the Master into a wait condition. SDA is a bi-directional line used to transmit and receive data to and from the interface. Data is transmitted MSB (Most Significant Bit) first in 8-bit per byte format, and the number of bytes transmitted per transfer is unlimited. The I<sup>2</sup>C bus is considered free when both lines are HIGH.

The I<sup>2</sup>C interface is compliant with high-speed mode, fast mode and standard mode I<sup>2</sup>C protocols.



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### I<sup>2</sup>C Operation

Transactions on the I<sup>2</sup>C bus begin after the Master transmits a start condition (S), which is defined as a HIGH-to-LOW transition on the data line while the SCL line is held HIGH. The bus is considered busy after this condition. The next byte of data transmitted after the start condition contains the Slave Address (SAD) in the seven MSBs (Most Significant Bits), and the LSB (Least Significant Bit) tells whether the Master will be receiving data '1' from the Slave or transmitting data '0' to the Slave. When a Slave Address is sent, each device on the bus compares the seven MSBs with its internally stored address. If they match, the device considers itself addressed by the Master. The KX132-1211 Slave Address is comprised of a user programmable part, a factory programmable part, and a fixed part, which allows for connection of multiple accelerometers to the same I<sup>2</sup>C bus. The Slave Address associated with the KX132-1211 is 00111YX, where the user programmable bit X, is determined by the assignment of ADDR (pin 1) to GND or IO\_VDD. Also, the factory programmable bit Y is set at the factory. For KX132-1211, the factory programmable bit Y is fixed to 1 (contact your Kionix sales representative for list of available devices). Table 6 lists possible I<sup>2</sup>C addresses for KX132-1211. As a result, up to four accelerometers can be implemented on a shared I<sup>2</sup>C bus as shown in Figure 3 (e.g. two KX132-1211 accelerometers and two other accelerometers with factory programmable bit Y set to 0).

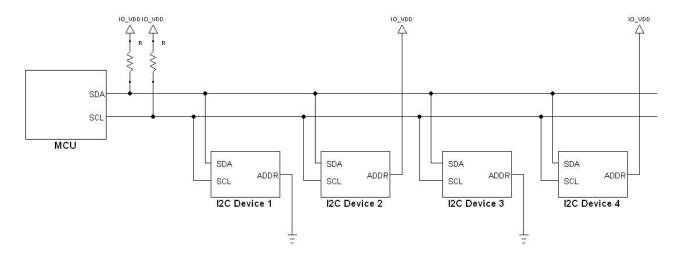
									Υ	X	
Description	ADDR pin	7-bit Address	Address	<7>	<6>	<5>	<4>	<3>	<2>	<1>	<0>
I2C Wr	GND	0x1E	0x3C	0	0	1	1	1	1	0	0
I2C Rd	GND	0x1E	0x3D	0	0	1	1	1	1	0	1
I2C Wr	IO_VDD	0x1F	0x3E	0	0	1	1	1	1	1	0
I2C Rd	IO VDD	0x1F	0x3F	0	0	1	1	1	1	1	1

Table 6: I<sup>2</sup>C Slave Addresses for KX132-1211

It is mandatory that receiving devices acknowledge (ACK) each transaction. Therefore, the transmitter must release the SDA line during this ACK pulse. The receiver then pulls the data line LOW so that it remains stable LOW during the HIGH period of the ACK clock pulse. A receiver that has been addressed, whether it is Master or Slave, is obliged to generate an ACK after each byte of data has been received. To conclude a transaction, the Master must transmit a stop condition (P) by transitioning the SDA line from LOW to HIGH while SCL is HIGH. The I<sup>2</sup>C bus is now free. Note that if the KX132-1211 is accessed through I<sup>2</sup>C protocol before the startup is finished a NACK signal is sent.



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I <sup>2</sup> C Device	Part Number	ADDR Pin	Slave Address	Bit Y (Bit 1 in 7-bit address)
1	KX132-1211	GND	0x1E	Factory Set to 1
2	KX132-1211	IO_VDD	0x1F	Factory Set to 1
3	*KXMMM	GND	0x1C	Factory Set to 0
4	*KXMMM	IO_VDD	0x1D	Factory Set to 0

<sup>\*</sup> KXMMM – contact Kionix sales representative for list of compatible devices

Figure 3: Multiple KX132-1211 Accelerometers on a Shared I<sup>2</sup>C Bus



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### Writing to an 8-bit Register

Upon power up, the Master must write to the KX132's control registers to set its operational mode. Therefore, when writing to a control register on the I<sup>2</sup>C bus, as shown Sequence 1 on the following page, the following protocol must be observed: After a start condition, SAD+W transmission, and the KX132-1211 ACK has been returned, an 8-bit Register Address (RA) command is transmitted by the Master. This command is telling the KX132-1211 to which 8-bit register the Master will be writing the data. Since this is I<sup>2</sup>C mode, the LSB of the RA command should always be zero (0). The KX132-1211 acknowledges the RA and the Master transmits the data to be stored in the 8-bit register. The KX132-1211 acknowledges that it has received the data and the Master transmits a stop condition (P) to end the data transfer. The data sent to the KX132-1211 is now stored in the appropriate register. The KX132-1211 automatically increments the received RA commands and, therefore, multiple bytes of data can be written to sequential registers after each Slave ACK as shown in Sequence 2 on the following page.

Note\*\* If a STOP condition is sent on the least significant bit of write data or the following master acknowledge cycle, the last write operation is not guaranteed and it may alter the content of the affected registers

### Reading from an 8-bit Register

When reading data from a KX132-1211 8-bit register on the I<sup>2</sup>C bus, as shown in Sequence 3 on the next page, the following protocol must be observed: The Master first transmits a start condition (S) and the appropriate Slave Address (SAD) with the LSB set at '0' to write. The KX132-1211 acknowledges and the Master transmits the 8-bit RA of the register it wants to read. The KX132-1211 again acknowledges, and the Master transmits a repeated start condition (Sr). After the repeated start condition, the Master addresses the KX132-1211 with a '1' in the LSB (SAD+R) to read from the previously selected register. The Slave then acknowledges and transmits the data from the requested register. The Master does not acknowledge (NACK) it received the transmitted data, but transmits a stop condition to end the data transfer. Note that the KX132-1211 automatically increments through its sequential registers, allowing data to be read from multiple registers following a single SAD+R command as shown below in Sequence 4 on the following page. Reading data from a buffer read register is a special case because if register address (RA) is set to buffer read register (BUF\_READ) in Sequence 4, the register auto-increment feature is automatically disabled. Instead, the Read Pointer will increment to the next data in the buffer, thus allowing reading multiple bytes of data from the buffer using a single SAD+R command.

Note\*\* Accelerometer's output data should be read in a single transaction using the auto-increment feature to prevent output data from being updated prior to intended completion of the read transaction.



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### **Data Transfer Sequences**

The following information illustrates the variety of data transfers that can occur on the I<sup>2</sup>C bus and how the Master and Slave interact during these transfers. Table 7 defines the I<sup>2</sup>C terms used during the data transfers.

Term	Definition
S	Start Condition
Sr	Repeated Start Condition
SAD	Slave Address
W	Write Bit
R	Read Bit
ACK	Acknowledge
NACK	Not Acknowledge
RA	Register Address
Data	Transmitted/Received Data
Р	Stop Condition

Table 7: I<sup>2</sup>C Terms

### Sequence 1: The Master is writing one byte to the Slave

Master	S	SAD + W		RA		DATA		Р
Slave			ACK		ACK		ACK	

### Sequence 2: The Master is writing multiple bytes to the Slave

	Master	S	SAD + W		RA		DATA		DATA		Р
ĺ	Slave			ACK		ACK		ACK		ACK	

### Sequence 3: The Master is receiving one byte of data from the Slave

Master	S	SAD + W		RA		Sr	SAD + R			NACK	Р
Slave			ACK		ACK			ACK	DATA		

### Sequence 4: The Master is receiving multiple bytes of data from the Slave

Master	S	SAD + W		RA		Sr	SAD + R			ACK		NACK	Р
Slave			ACK		ACK			ACK	DATA		DATA		



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### **HS-mode**

To enter the 3.4MHz high speed mode of communication, the device must receive the following sequence of conditions from the master: a Start condition followed by a Master code (00001XXX) and a Master Non-acknowledge. Once recognized, the device switches to HS-mode communication. Read/write data transfers then proceed as described in the sequences above. Devices return to the FS-mode after a STOP occurrence on the bus.

Sequence 5: HS-mode data transfer of the Master writing multiple bytes to the Slave

Speed		FS-mode	!				HS-mc	de				FS-mode
Master	S	M-code	NACK	Sr	SAD + W		RA		DATA		Р	
Slave						ACK		ACK		ACK		

n bytes + ack.

Sequence 6: HS-mode data transfer of the Master receiving multiple bytes of data from the Slave

Speed		FS-mode	)	HS-mode							
Master	S	M-code	NACK	Sr	SAD + W		RA				
Slave						ACK		ACK			

Speed				FS-mode					
Master	Sr	SAD + R					NACK	Р	
Slave			ACK	DATA	ACK	DATA			

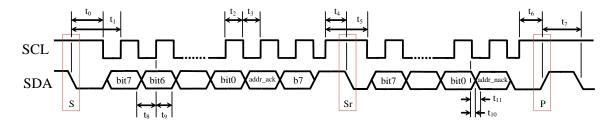
(n-1) bytes + ack.



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### I<sup>2</sup>C Timing Diagram



Number	Description	MIN	MAX	Units
t <sub>0</sub>	SDA LOW to SCL LOW transition (Start event)	50	-	ns
t <sub>1</sub>	SDA LOW to first SCL rising edge	100	-	ns
<b>t</b> <sub>2</sub>	SCL pulse width: HIGH	100	-	ns
t <sub>3</sub>	SCL pulse width: LOW	100	-	ns
t <sub>4</sub>	SCL HIGH before SDA falling edge (Start Repeated)	50	-	ns
<b>t</b> <sub>5</sub>	SCL pulse width: HIGH during a S/Sr/P event	100	-	ns
<b>t</b> 6	SCL HIGH before SDA rising edge (Stop)	50	-	ns
t <sub>7</sub>	SDA pulse width: HIGH	25	-	ns
t <sub>8</sub>	SDA valid to SCL rising edge	50	-	ns
t <sub>9</sub>	SCL rising edge to SDA invalid	50	-	ns
t <sub>10</sub>	SCL falling edge to SDA valid (when slave is transmitting)	-	100	ns
t <sub>11</sub>	SCL falling edge to SDA invalid (when slave is	0	-	ns
Note	Recommended I <sup>2</sup> C CLK	2.5	-	μS

Table 8: I<sup>2</sup>C Timing (Fast Mode)



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### **SPI Communications**

#### 4-Wire SPI Interface

The KX132-1211 also utilizes an integrated 4-Wire Serial Peripheral Interface (SPI) for digital communication. The SPI interface is primarily used for synchronous serial communication between one Master device and one or more Slave devices. The Master, typically a micro controller, provides the SPI clock signal (SCLK) and determines the state of Chip Select (nCS). The KX132-1211 always operates as a Slave device during standard Master-Slave SPI operation.

4-wire SPI is a synchronous serial interface that uses two control and two data lines. With respect to the Master, the Serial Clock output (SCLK), the Data Output (SDI or MOSI) and the Data Input (SDO or MISO) are shared among the Slave devices. The Master generates an independent Chip Select (nCS) for each Slave device that goes LOW at the start of transmission and goes back HIGH at the end. The Slave Data Output (SDO) line, remains in a high-impedance (hi-z) state when the device is not selected, so it does not interfere with any active devices. This allows multiple Slave devices to share a master SPI port as shown in Figure 4 below.

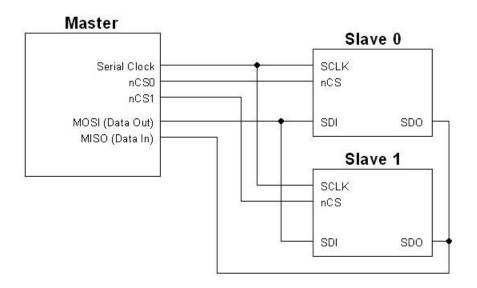


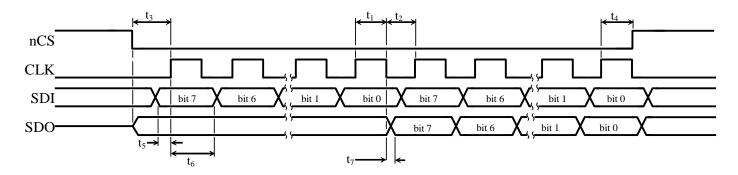
Figure 4. 4-wire SPI Connections



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### 4-Wire SPI Timing Diagram



Number	Description	MIN	MAX	Units
t <sub>1</sub>	CLK pulse width: HIGH	45		ns
t <sub>2</sub>	CLK pulse width: LOW	45		ns
t <sub>3</sub>	nCS LOW to first CLK rising edge	20		ns
t <sub>4</sub>	nCS LOW after the final CLK rising edge to nCS HIGH	20		ns
<b>t</b> 5	SDI valid to CLK rising edge	10		ns
t <sub>6</sub>	CLK rising edge to SDI invalid	10		ns
t <sub>7</sub>	CLK falling edge to SDO valid		35	ns

Table 9: 4-Wire SPI Timing

### Notes

- 1. t<sub>7</sub> is only present during reads.
- 2. Timings are for VDD of 1.8V to 3.6V with  $1k\Omega$  pull-up resistor and maximum 20pF load capacitor on SDO.



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### 4-Wire Read and Write Registers

The registers embedded in the KX132-1211 accelerometer have 8-bit addresses. Upon power up, the Master must write to the accelerometer's control registers to set its operational mode. On the falling edge of nCS, a 2-byte command is written to the appropriate control register. The first byte initiates the write to the appropriate register, and is followed by the user-defined, data byte. The MSB (Most Significant Bit) of the register address byte will indicate "0" when writing to the register and "1" when reading from the register. This operation occurs over 16 clock cycles. All commands are sent MSB first. **The host must return nCS HIGH for at least one clock cycle before the next data request.** However, when data is being read from a buffer read register (BUF\_READ), the nCS signal can remain LOW until the buffer is read. Figure 5 below shows the timing diagram for carrying out an 8-bit register write operation.

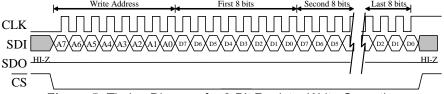


Figure 5: Timing Diagram for 8-Bit Register Write Operation

In order to read an 8-bit register, an 8-bit register address must be written to the accelerometer to initiate the read. The MSB of this register address byte will indicate "0" when writing to the register and "1" when reading from the register. Upon receiving the address, the accelerometer returns the 8-bit data stored in the addressed register. This operation also occurs over 16 clock cycles. All returned data is sent MSB first, and the host must return nCS HIGH for at least one clock cycle before the next data request. Figure 6 shows the timing diagram for an 8-bit register read operation.

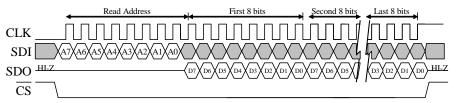


Figure 6: Timing Diagram for 8-Bit Register Read Operation



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#### 3-Wire SPI Interface

The KX132-1211 also utilizes an integrated 3-Wire Serial Peripheral Interface (SPI) for digital communication. 3-wire SPI is a synchronous serial interface that uses two control lines and one data line. With respect to the Master, the Serial Clock output (SCLK), the Data Output/Input (SDI) are shared among the Slave devices. The Master generates an independent Chip Select (nCS) for each Slave device that goes LOW at the start of transmission and goes back HIGH at the end. This allows multiple Slave devices to share a master SPI port as shown in Figure 7 below.

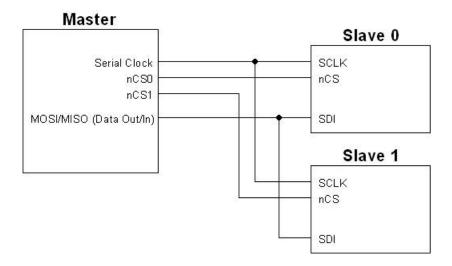


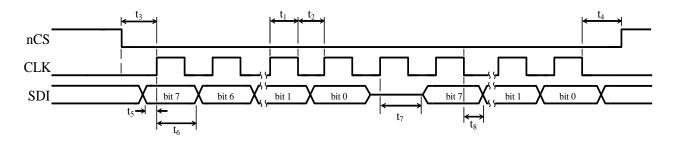
Figure 7: KX132-1211 3-wire SPI Connections



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### 3-Wire SPI Timing Diagram



Number	Description		MAX	Units
<b>t</b> 1	CLK pulse width: HIGH		-	ns
t <sub>2</sub>	CLK pulse width: LOW	45	-	ns
t <sub>3</sub>	nCS LOW to first CLK rising edge	20	-	ns
t <sub>4</sub>	nCS LOW after the final CLK falling edge to nCS HIGH	20	-	ns
<b>t</b> 5	SDI valid to CLK rising edge	10	-	ns
t <sub>6</sub>	CLK rising edge to SDI input invalid	10	-	ns
t <sub>7</sub>	CLK extra clock cycle rising edge to SDI output becomes	1	-	ns
t <sub>8</sub>	CLK falling edge to SDI output becomes valid	-	35	ns

Table 10: 3-Wire SPI Timing

#### Notes

- 1. t<sub>7</sub> and t<sub>8</sub> are only present during reads
- 2. Timings are for VDD of 1.8V to 3.6V with  $1k\Omega$  pull-up resistor and maximum 20pF load capacitor on SDI.
- 3. The SDO/ADDR pin is configured in a high-impedance input-state, and must be externally tied to GND or IO\_VDD



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### 3-Wire Read and Write Registers

The registers embedded in the KX132-1211 accelerometer have 8-bit addresses. Upon power up, the Master must write to the accelerometer's control registers to set its operational mode. On the falling edge of nCS, a 2-byte command is written to the appropriate control register. The first byte initiates the write to the appropriate register, and is followed by the user-defined, data byte. The MSB (Most Significant Bit) of the register address byte will indicate "0" when writing to the register and "1" when reading from the register. A read operation occurs over 17 clock cycles and a write operation occurs over 16 clock cycles. All commands are sent MSB first. **The host must return nCS HIGH for at least one clock cycle before the next data request**. However, when data is being read from a buffer read register (BUF\_READ), the nCS signal can remain LOW until the buffer is read. Figure 8 below shows the timing diagram for carrying out an 8-bit register write operation.

NOTE\*\* If a STOP condition is sent on the least significant bit of write data or the following master acknowledge cycle, the last write operation is not guaranteed and it would cause unexpected register write.

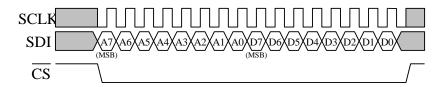


Figure 8: Timing Diagram for 8-Bit Register Write Operation

In order to read an 8-bit register, an 8-bit register address must be written to the accelerometer to initiate the read. The MSB of this register address byte will indicate "0" when writing to the register and "1" when reading from the register. Upon receiving the address, the accelerometer returns the 8-bit data stored in the addressed register. For 3-wire read operations, one extra clock cycle between the address byte and the data output byte is required. Therefore, this operation occurs over 17 clock cycles. All returned data is sent MSB first, and the host must return nCS HIGH for at least one clock cycle before the next data request. Figure 9 shows the timing diagram for an 8-bit register read operation.

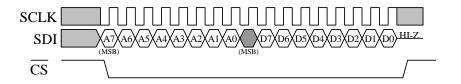


Figure 9: Timing Diagram for 8-Bit Register Read Operation



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### **Revision History**

Revision	Description	Date
1.0	Production Release.	31-Jul-2019

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### **Appendix**

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