User's Manual

RTKA2108502H00000BU

User's Manual: Evaluation Board

Industrial Analog and Power

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1. Overview

The RTKA2108502H00000BU evaluation board is a 8-layer FR4 board with 2oz. copper on all layers. The board measures 4.7in x 4.8in. The evaluation board includes placeholders for the pin-strap resistor population that allow for the adjustment of the output voltage, switching frequency, soft-start/stop timing, input UVLO, ASCR (Gain and Residual) parameters, and the device PMBus address. Also, fault limits can be programmed or changed easily with the PMBus compliant serial bus interface.

The evaluation board makes use of the RAA210850, a 50A step-down DC/DC power supply module with an integrated digital PWM controller, dual-phase synchronous power switches, inductors, and passives. Only input, output capacitors and minimal passives are required to finish the design. The board can output a continuous current at 50A without airflow or heatsink. The RAA210850 uses a ChargeMode™ control (ASCR) architecture that responds to a transient load within a single switching cycle.

Provided with the evaluation kit, the ZLUSBEVAL3Z (a USB to PMBus adapter) connects the evaluation board with a PC to activate the PMBus communication interface. The PMBus command set is accessed by using the PowerNavigator™ evaluation software from a PC running Microsoft Windows.

The RTKA2108502H00000BU can operate in Pin-Strap mode without needing the ZLUSBEVAL3Z adapter or PMBus communication.

1.1 Key Features

- V_{IN} range of 4.5V to 14V, V_{OUT} adjustable from 0.6V to 5V
- Programmable V_{OUT}, input and output UVP/OVP, OTP/UTP, soft-start/stop, and external synchronization
- Monitor: V_{IN}, V_{OUT}, I_{OUT}, temperature, duty cycle, switching frequency, and faults
- ChargeMode control tunable with PMBus
- Mechanical switch for enable and a power-good LED indicator

1.2 Specifications

The board has default configurations for the following operating conditions:

- V_{IN} = 4.5V to 12V
- V_{OUT} = 1.2V
- I_{MAX} = 50A
- f_{SW} = 421kHz
- Peak efficiency: >91% at 70% load
- ASCR gain = 200, ASCR residual = 90
- On/off delay = 5ms, On/off ramp time = 5ms

1.3 Ordering Information

Part Number	Description
RTKA2108502H00000BU	RAA210850 Kit (Evaluation Board, ZLUSBEVAL3Z Adapter, USB Cable)

1.4 Related Literature

For a full list of related documents, visit our website.

• RAA210850 product page

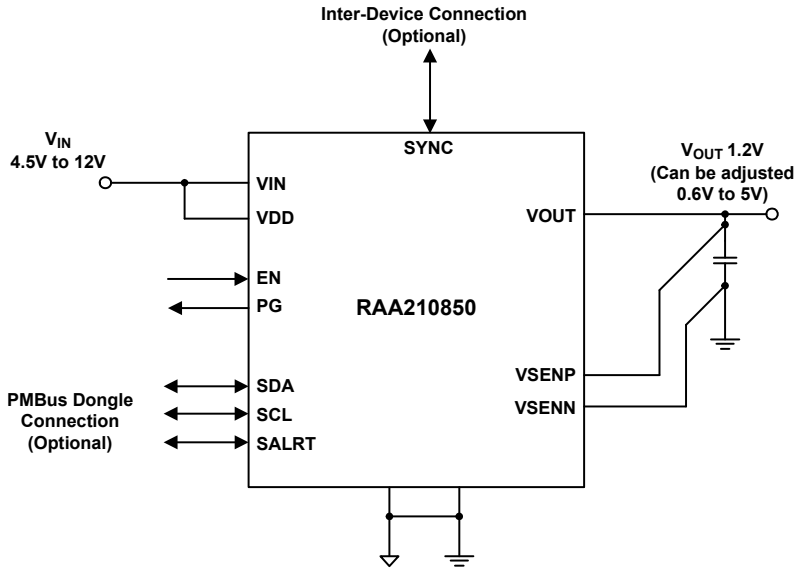

Figure 1. Block Diagram

1.5 Recommended Equipment

- DC power supply with minimum 15V/25A sourcing capacity
- Electronic load capable of sinking current up to 50A
- Digital Multimeters (DMMs)
- Oscilloscope with higher than 100MHz bandwidth

2. Functional Description

The RTKA2108502H00000BU provides all circuitry required to evaluate the features of the RAA210850. A majority of the features of the RAA210850 are available on the RTKA2108502H00000BU evaluation board, such as compensation-free ChargeMode control and soft-start delay and ramp times. For interleave functionality evaluation, the board can connect to another Renesas digital module evaluation board.

Figures 5 and 6 show the board images of the RTKA2108502H00000BU evaluation board.

2.1 Operating Range

By default, the RTKA2108502H00000BU is configured to operate at V_{OUT} = 1.2V, f_{SW} = 421kHz. V_{IN} ranges from 4.5V to 12V. The board can also support a wider operating range to meet the requirements of specific applications. The V_{OUT} can be adjusted from 0.6V to 5V. Load current range is 0A to 50A. Note that airflow across the board may be needed for continuous operation at 50A at elevated ambient temperatures. The f_{SW} and output voltage can also be tuned. However, to ensure sufficient stability margins, switching frequency and output capacitors can only be selected using the "RAA210850 Design Guide Matrix and Output Voltage Response" table in the RAA210850 datasheet. If input voltage is less than 5V, tie the VDRV/VCC test point directly to V_{IN} or to a separate 5V power supply for normal operation and best efficiency.

If external synchronization is used, connect the SYNC test point to the external clock. Note that the external clock signal should be active before the module is enabled.

2.2 Quick Start Guide

2.2.1 Pin-Strap Option

RTKA2108502H00000BU can be configured in pin-strap mode with standard 1% 0603 resistors. The PMBus interface is not required to evaluate the RAA210850 in pin-strap mode. The output voltage (VOUT), switching frequency (f_{SW}), soft-start/stop delay and ramp time, input undervoltage protection (UVLO) threshold, Interleave, ASCR gain and residual, and the device PMBus address can be changed by populating recommended resistors at placeholders provided in the evaluation board. By default, the evaluation board operates in Pin-Strap mode and regulates at V_{OUT} = 1.2V, f_{SW} = 421kHz, soft-start/stop delay time = 5ms, soft-start/stop ramp time = 5ms, UVLO = 4.2V, ASCR gain = 200, ASCR residual = 90, and PMBus address = 28h. Complete the following steps to evaluate the RAA210870 in Pin-Strap mode.

Complete the following steps to evaluate the RAA210850 in Pin-Strap mode.

(1) Set the ENABLE switch to "DISABLE".

(2) Connect the load to the VOUT lug connectors (J_7-J_8 and J_9-J_{10}).

(3) Connect the power supply to the VIN connectors (J_5 and J_6). Make sure the power supply is not enabled when making the connection.

(4) Turn on the power supply.

(5) Set the ENABLE switch to "ENABLE".

(6) Measure 1.2V V_{OUT} at the probe point labeled "VOUT REGULATION MONITOR" (J_{11}).

(7) Observe the switching frequency of 421kHz at the probe points labeled PHASE1 (TP_{10}) and PHASE2 (TP_{11}).

(8) To measure the module efficiency, connect the multimeter voltage probes at probe points labeled VIN (TP_1), GND (TP_2), and VOUT (TP_{12}).

(9) To change the V_{OUT}, disconnect the board from the setup and populate a 1% standard 0603 resistor at the RVSET_CRS placeholder location on bottom layer. If additional output voltage fine tuning is required then populate a resistor at the RVSET_FINE placeholder. Refer to the "Output Voltage Selection" section in the RAA210850 datasheet to determine the correct selection of resistors for programming the output voltage. By default, VOUT_MAX is set to 110% of V_{OUT} set by pin-strap resistor.

(10) To change the switching frequency, disconnect the board from the setup and populate a 1% standard 0603 resistor at the RFSET placeholder on the bottom layer. For recommended resistor values, refer to the "Switching Frequency and PLL" section in the RAA210850 datasheet.

(11) To change the soft-start/stop delay and ramp time, disconnect the board from the setup and populate a 1% standard 0603 resistor at the bottom layer R_6 placeholder location. For the recommended resistor values, refer to the "Soft-Start/Stop Delay, and Ramp Times" section in the RAA210850 datasheet.

(12) To change the UVLO, disconnect the board from the setup and populate a 1% standard 0603 resistor on the bottom layer R_6 placeholder. Notice that the UVLO programming shares the same pin with the soft-start/stop programming. For the recommended resistor values, refer to the "Input Undervoltage Lockout (UVLO)" section in the RAA210850 datasheet.

(13) To change the ASCR gain and residual, disconnect the board from the setup and populate a 1% standard 0603 resistor on the bottom layer R_7 placeholder location. For the recommended resistor values, refer to the "Loop Compensation" and the design guide matrix in the RAA210850 datasheet.

2.2.2 PMBus Option

The features of the RTKA2108502H00000BU can be evaluated using the provided ZLUSBEVAL3Z dongle and PowerNavigator evaluation software. Follow these steps to evaluate the RAA210850 with PMBus option.

(1) Install PowerNavigator from the Renesas website.

(2) Set the ENABE switch to "DISABLE".

(3) Connect the load to the VOUT lug connectors (J_7-J_8 and J_9-J_{10}).

(4) Connect the power supply to the VIN connectors (J_5 and J_6). Make sure the power supply is not enabled when making the connection.

(5) Turn on the power supply.

(6) Connect the ZLUSBEVAL3Z dongle (USB to PMBus adapter) to the RTKA2108502H00000BU board with the 6-pin male connector labeled "PMBus DONGLE IN".

(7) Connect the supplied USB cable from the computer USB to the ZLUSBEVAL3Z dongle.

(8) Launch PowerNavigator.

(9) By factory default, the RAA210850 device on the board operates in Pin-Strap mode, but the user can modify the operating parameters with PowerNavigator. The default pin-strap configurations are overwritten if PowerNavigator is used.

(10) Set the ENABLE switch to "ENABLE". Alternatively, the PMBus ON_OFF_CONFIG and OPERATION commands may be used from the PowerNavigator™ software for enabling the module.

(11) Monitor and configure the RTKA2108502H00000BU board using the PMBus commands in PowerNavigator.

(12) PowerNavigator tutorial videos are available at the Renesas website.

(13) For evaluating Interleave functionality for multiple Renesas digital power products using a single ZLUBEVAL3Z dongle, RAA210850 can be daisy chained with other digital power evaluation boards. The PMBus address can be changed by placing a 1% standard 0603 resistor at the R4 placeholder location on the bottom layer. Refer to the "SMBus Module Address Selection" section in the RAA210850 datasheet for recommended values.

3. Evaluation Board Information

If the input voltage is less than 5.3V, tie the VCC test point directly to the VIN, or for operational efficiency, tie the VCC test point to a separate 5V power supply. For external synchronization, connect the SYNC test point to the external clock.

Note: The external clock signal should be active before the module is enabled.

3.1 External Clock Synchronization

The RAA210850 can synchronize to an external clock. External clock synchronization allows the user to operate multiple converters at the same switching frequency and can lead to improved EMI characteristics. The RTKA2108502H00000BU evaluation board can assess this functionality. A function generator is required. Complete the following steps to operate RAA210850 with an external clock frequency of 593kHz:

(1) Set the Enable switch to "DISABLE" position.

(2) Solder a 17.8kΩ resistor at RSET on the CFG pin (R_{13}).

(3) Program the function generator to output a continuous square pulse waveform of 593kHz. Program the pulse width to be at least 150ns.

(4) Ensure that the clock signal is stable by monitoring the waveform on an oscilloscope.

(5) After verifying clock stability, connect the output cables from the function generator to test points TP_3 (labeled "SYNC").

(6) Turn the output of the function generator ON.

(7) Enable the module by setting the Enable switch to "ENABLE".

(8) Observe the switching frequency at test points TP_{10} "PHASE1" and TP_{11} "PHASE2".

(9) The module synchronizes to the 593kHz external clock from the function generator.

(10) Ensure that the module is always disabled before changing the frequency of the external clock.

(11) Loss of Sync fault is generated when the external clock is lost.

3.2 Bode Plot Measurement

Assessing the stability of the converter is an important step in the design process. Bode plots can be a useful and reliable tool to identify the loop response of the converter. Phase and gain margins give an insight into the stability of the system. Bandwidth can indicate how quickly the converter responds to disturbances in input voltage or load transients. Correctly measuring the loop response is critical for designing stable converter systems.

A network analyzer is required to perform the frequency response measurements on the RTKA2108502H00000BU evaluation board. Complete the following steps to evaluate the converter loop response for RAA210850 on the RTKA2108502H00000BU evaluation board.

(1) Break the feedback loop by removing the R_{10} resistor in remote sense path, connected between VSENP and VOUT.

(2) Solder a 20Ω resistor in its place. The value of the resistor could be in the range 10Ω to 50Ω.

(3) Solder a twisted wire pair to the 20Ω resistor. Ensure that the wires are short in length. A small twisted pair works well by minimizing noise pickup, which is important for a good measurement.

(4) Enable the converter.

(5) Use a Network Analyzer to inject a small AC signal (~20mV) across the 20Ω resistor as shown in Figure 2 on page 7.

(6) Measure the amplitudes of the signals at points A and B as shown in Figure 2.

(7) Sweep the frequency using the Network Analyzer to observe the bandwidth and the phase and gain margins.

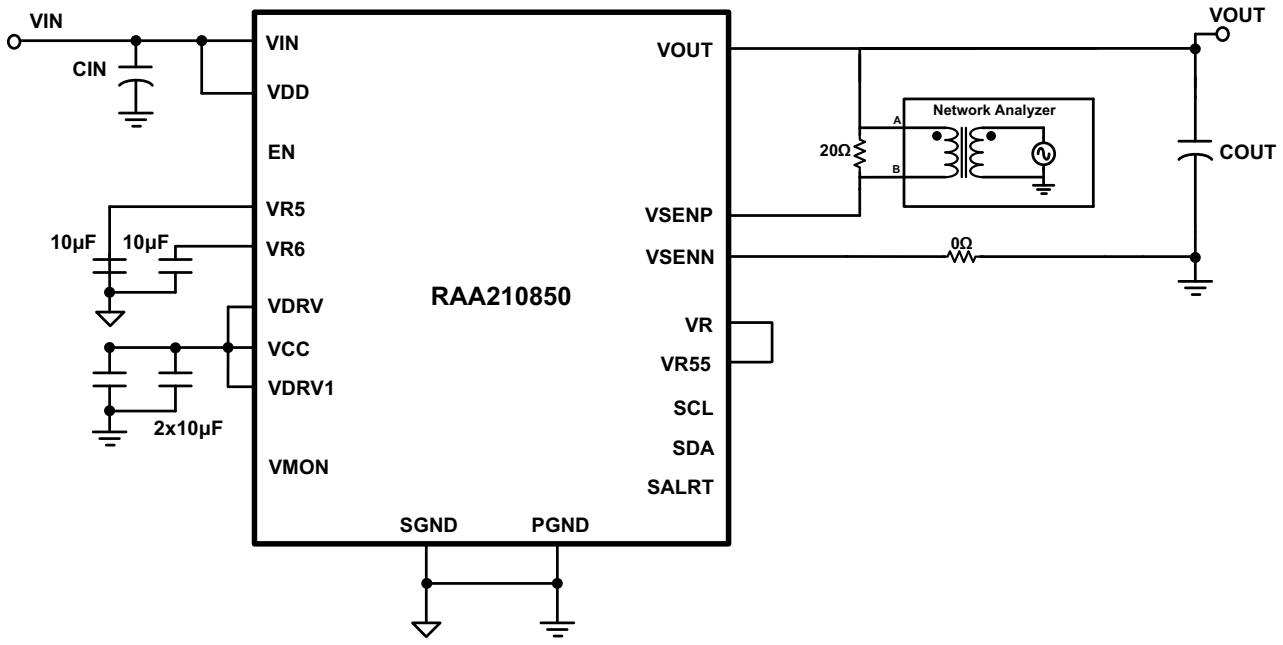

Figure 2. Schematic for Evaluating a Converter Loop Response

Figure 3 shows an example bode plot generated by the Network Analyzer for RAA210850 at 12V input, 1.2V output, and 50A load at 727kHz with 8x100µF ceramic + 2x470µF POS output capacitors. The ASCR gain is set at 320, and the ASCR residual is set at 90. The plot shows a crossover frequency of 70kHz with a phase margin of 64º. A 10dB gain margin is observed at 194kHz.

Refer to the RAA210850 Design Guide Matrix and Output Voltage Response Table in the RAA210850 datasheet for detailed design guidelines. The guidelines include a selection of input/output capacitors and different ASCR gain and residual values.

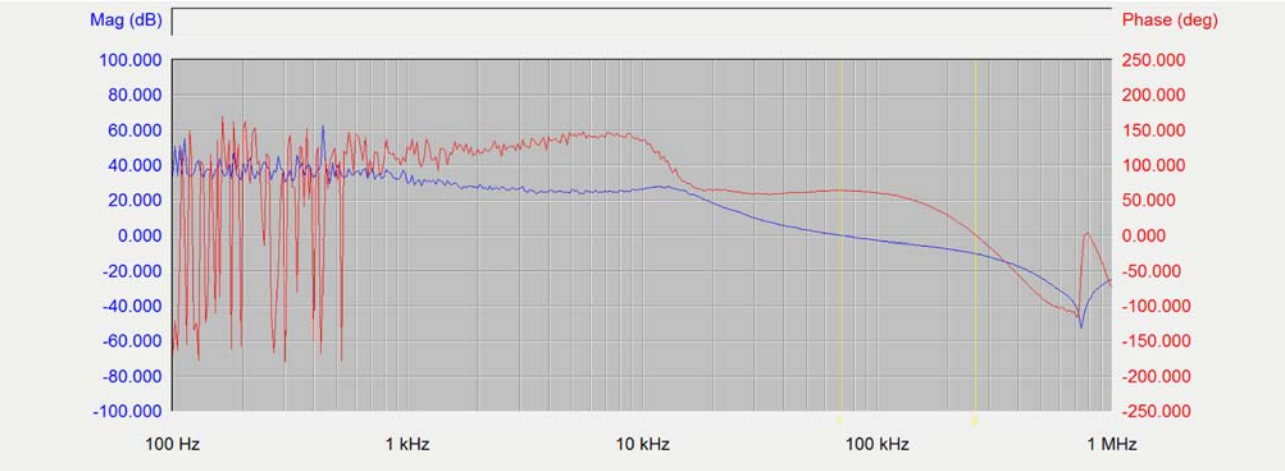

Figure 3. Generated Bode Plot by the Network Analyzer for RAA210850

3.3 Interleave Operation

When multiple point-of-load converters share a common DC input supply, it is desirable to adjust the clock phase offset of each device, so that not all devices start to switch simultaneously. Setting each converter to start its switching cycle at a different point in time can dramatically reduce input capacitance requirements and efficiency losses. Because the peak current drawn from the input supply is effectively spread out over a period of time, the

peak current drawn at any given moment is reduced. In fact, the power losses proportional to the I^2_{RMS} are reduced dramatically.

To enable phase spreading in a multi-module operation, all converters must be synchronized to the same switching clock. The phase offset of each device may be set to any value between 0° and 360° in 22.5° increments by choosing the device SMBus address from the “Interleave” table in the RAA210850 datasheet. The lower 4 bits of the SMBus address set the value of INTERLEAVE command. Complete the following steps to implement interleave functionality for a two module operation:

(1) Choose SA (SMBus Address) for Module 1 and Module 2 from the “Interleave” table in the RAA210850 datasheet based on the desired phase difference. Populate the corresponding RSET for SA (R4) according to the “SMBus Address Resistor Selection” table in the RAA210850 datasheet.

For example, when Module 1 has SA = 28h (INTERLEAVE =8, 180° phase shift from the rising edge of the external clock) and Module 2 has SA = 24h (INTERLEAVE = 4, 90° phase shift from the rising edge of the external clock), the net phase difference between Module 1 and Module 2 will be 180 - 90 = 90°.

(2) Populate RSET on CFG pin for both boards to sync to an external clock source of a particular switching frequency based on the “External Frequency Sync Settings” table in the RAA210850 datasheet.

(3) Connect the power supply to the VIN connectors (VIN/GND) on both the boards.

(4) Connect the ZLUSBEVAL3Z dongle to the 6-pin male connector labeled "PMBus DONGLE In" to one of the boards.

(5) Daisy chain the second board to the first board by connecting "PMBus DONGLE Out" of first board to the "PMBus DONGLE In" of the second board.

(6) Provide an external clock on the SYNC pins of the two boards from a function generator. External clock frequency from the function generator should be within ±10% of the listed options shown the “External Frequency Sync Settings” table in the RAA210850 datasheet. Incoming clock signal must be stable before the enable pin is asserted. The external clock signal must not vary more than 10% from its initial value and should have a minimum pulse width of 150ns.

(7) Turn the input power supply ON. Next, set the ENABLE switch to the "ENABLE" position.

(8) Monitor the switch node at the probe point labeled PHASE1 (TP_{10}) on the two boards using an oscilloscope to verify the phase spread set.

(9) This functionality can also be verified using the INTERLEAVE command in PowerNavigator.

Note: Every module is assigned a unique Rail ID based on the SA setting. These settings can be observed in the Power Map window of PowerNavigator.

3.4 V_{OUT} Transient Response Check

The RTKA2108502H00000BU board has a built-in transient load test circuit (Figure 4). A 100A N-Channel MOSFET (Manufacturer PN: BSC010NE2LSI) is connected across VOUT and PGND next to the remote voltage sensing location (CVSEN). The 10mΩ current-sense resistor R_{54} is used for monitoring the drain-to-source current of the MOSFET. For a transient load test, inject the gate drive pulse signal at J_{16}. The load current can be monitored through J_{15}. When the gate turn-on signal is applied, the MOSFET operates in the saturation region (not the linear region). To avoid MOSFET overheating, both the pulse width and duty cycle of the gate signal must be reduced (recommended duty cycle should be less than 2%). The amplitude of the gate driver pulse voltage can be adjusted to obtain the desired step size for a transient load current.

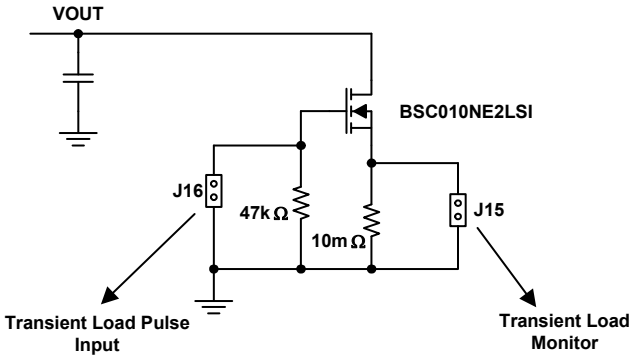

Figure 4. Schematic for Transient Load Measurement

4. PCB Layout Guidelines

The RTKA2108502H00000BU board layout has been optimized for electrical performance, low loss, and excellent thermal performance.

The following are the key features of the RTKA2108502H00000BU layout:

- Large PGND planes and a separate SGND plane, where the SGND plane is connected to PGND on the second layer with a single point connection. Multiple vias are used for small pins such as J_{16}, H_{16}, K_{16}, M_5, M_{14}, M_{17}, and N_5 to connect to inner SGND or PGND layer.
- Ceramic capacitors between VIN and PGND, VOUT and PGND, and bypass capacitors between VDD, VDRV and the ground plane are placed close to the module to minimize high frequency noise. Some output ceramic capacitors are placed close to the VOUT pads in the direction of the load current path to create a low impedance path for the high frequency inductor ripple current.
- Large copper areas are used for power path (VIN, PGND, VOUT) to minimize conduction loss and thermal stress. Multiple vias are used to connect the power planes in different layers.
- Remote sensing traces are connected from the regulation point to VSENP and VSENN pins. The two traces are placed in parallel, to achieve tight output voltage regulation. The regulation point is on the right side of the board in between the VOUT power lugs and the PGND power lugs.
- Multiple vias are used to connect PAD14 and 16 (SW_1 and SW_2) to inner layers for better thermal performance. The inner layer SW1 and SW2 traces are limited in area and are surrounded by PGND planes to avoid noise coupling. Caution was taken that no sensitive traces, such as the remote sensing traces, were placed close to these noisy planes.
- SWD1 (L_3) and SWD2 (P_{11}) pins are connected to SW1 and SW2 pads respectively with short loop wires of 40mil width. The wire width should be at least 20mils.

4.1 Thermal Considerations and Current Derating

The board layout is critical for safely operating the module and delivering maximum allowed power. The board layout requires a proper design to maximize thermal performance in order to function in high temperature environments and carry large currents. A proper design requires a user to select enough trace width, copper weight, and the proper connectors.

The RTKA2108502H00000BU evaluation board is designed for running 50A at room temperature without additional cooling systems. However, if the output voltage increases or the board operates at elevated temperatures, the available current is derated. Refer to the derated current curves in the RAA210850 datasheet to determine the maximum output current the evaluation board can safely supply. Θ_{JA} is measured by inserting a thermocouple inside the module to measure peak junction temperature.

4.2 RTKA2108502H00000BU Evaluation Board

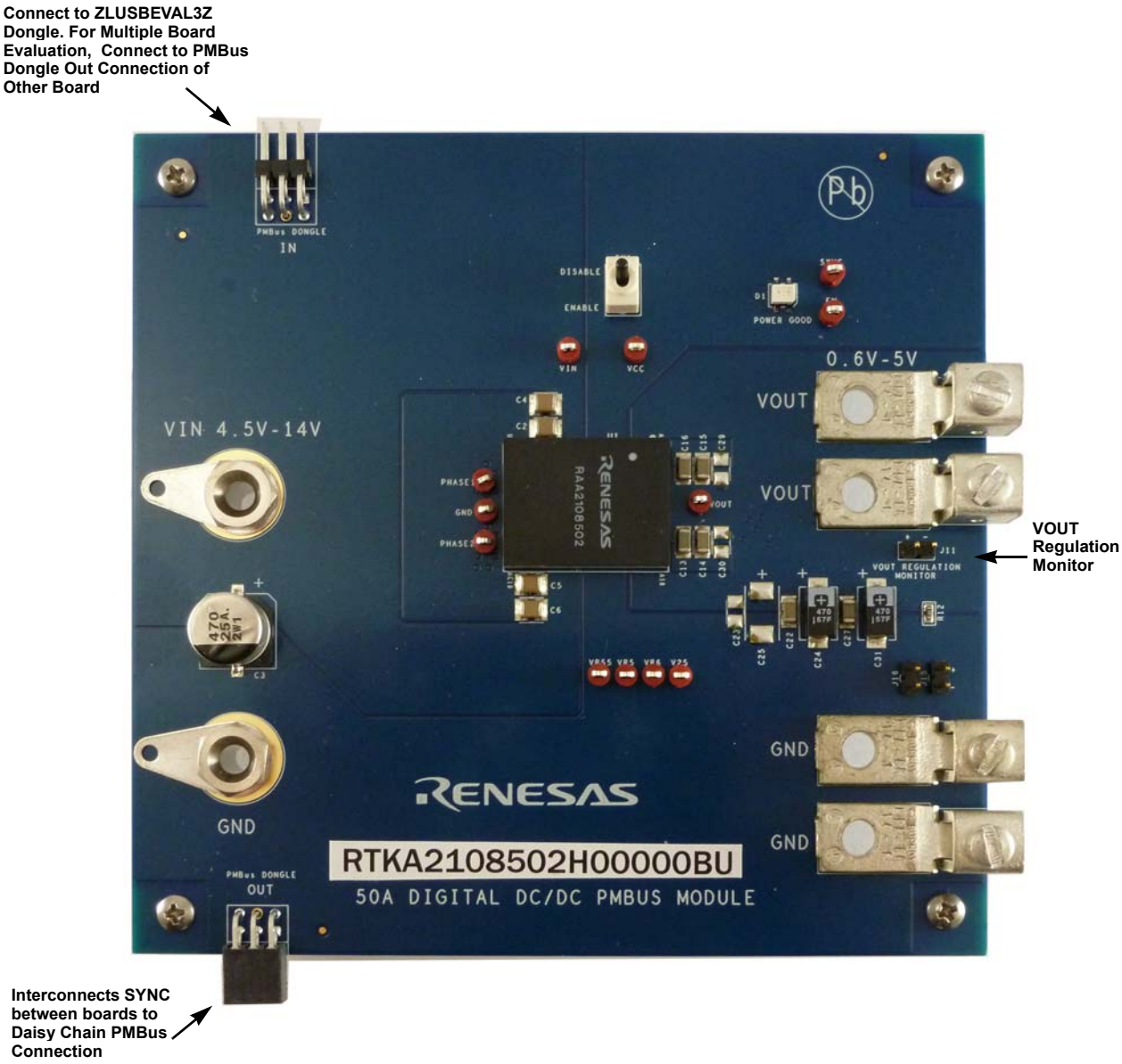

Figure 5. RTKA2108502H00000BU Evaluation Board (Top)

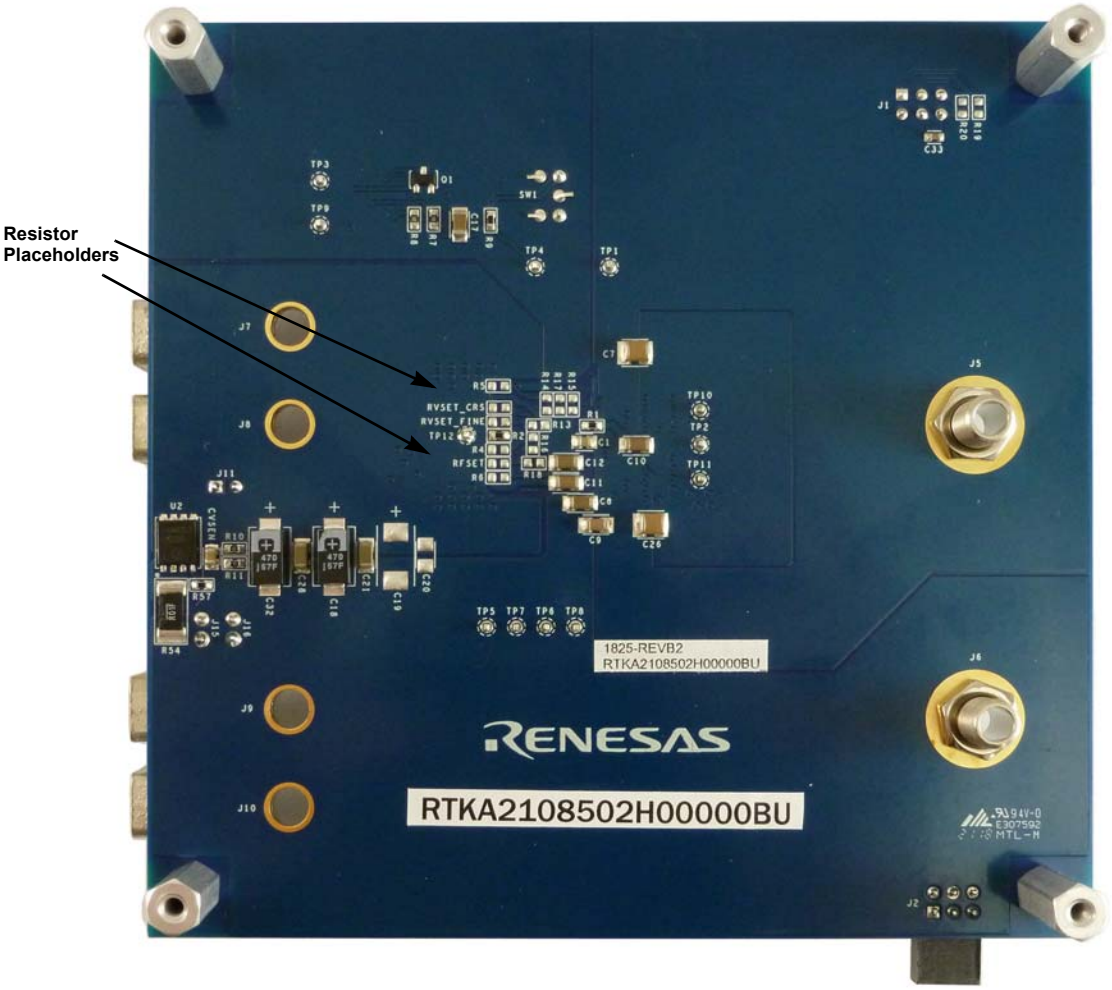

Figure 6. RTKA2108502H00000BU Evaluation Board (Bottom)

4.3 RTKA2108502H00000BU Circuit Schematic

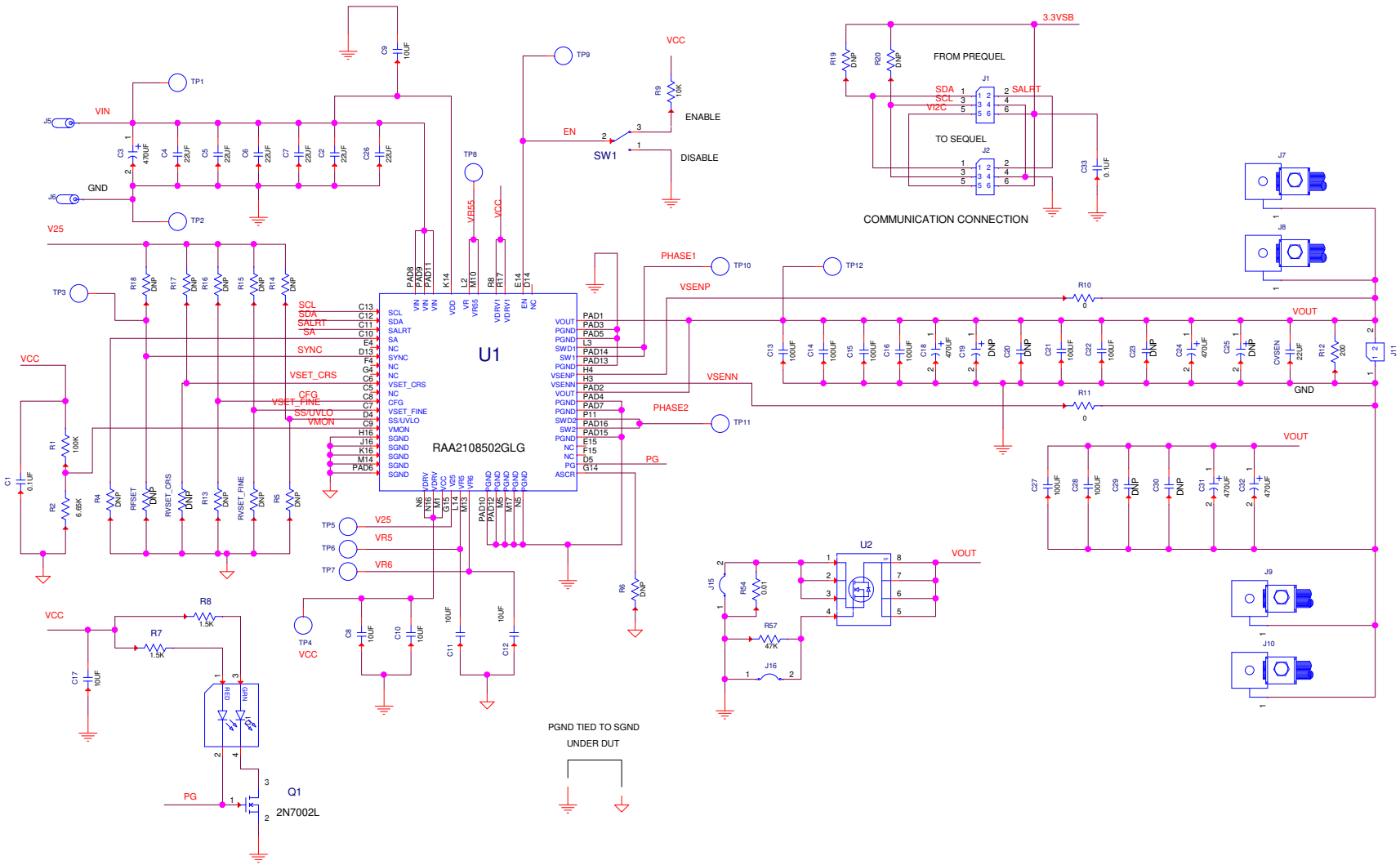

Figure 7. Schematic

4.4 Bill of Materials

Qty	Reference Designator	Description	Manufacturer	Manufacturer Part
1		PCB, RTKA2108502H00000BU, REVB, RoHS	Multilayer PCB (Hi Tech Circuit)	RTKA2108502H00000BURBPCB
2	C19, C25	DO NOT POPULATE_PLACE HOLDER		
4	C18, C24, C31, C32	CAP TANT POLY 470µF, 6.3V, 20%, 10mΩ, 2917, RoHS	Panasonic	6TPF470MAH
1	CVSEN	CAP CER 22µF, 6.3V, 20%, X5R, 0805, RoHS	TDK	C2012X5R0J226M
1	C3	CAP ALUM 470µF, 25V, 20%, SMD, RoHS	Panasonic	EEE1EA471P
8	C13-C16, C21, C22, C27, C28	CAP CER 100µF, 6.3V, 20%, X5R, 1206, RoHS	Murata	GRM31CR60J107ME39L
4	C20, C23, C29, C30	DO NOT POPULATE_PLACE HOLDER		
5	C8, C10-C12, C17	CAP CER 10µF, 10V, 10%, X7R, 1206, RoHS	Mµrata	GRM31CR71A106KA01L
6	C2, C4-C7, C26	CAP CER 22µF, 16V, 10%, X7R, 1210, RoHS	Murata	GRM32ER71C226KE18L
1	C33	CAP CER 0.1µF, 25V, 10%, X7R, 0603, RoHS	Kemet	C0603X104K3RACTU
1	C1	CAP CER 0.1µF, 25V, 10%, X7R, 0805, RoHS	Kemet	C0805X104K3RACTU
1	C9	CAP CER 10µF, 25V, 10%, X7R, 1206, RoHS	Kemet	C1206X106K3RAC7800
14	R4-R6,R13-R20,RVSET_FINE, RVSET_CRS, RFSET	DO NOT POPULATE_PLACE HOLDER		
2	R10,R11	RES SMD 0Ω JUMPER, 1/10W, 0603, RoHS	Panasonic	ERJ-3GEY0R00V
1	R9	RES SMD 10kΩ, 1%, 1/10W, 0603, RoHS	Panasonic	ERJ-3EKF1002V
1	R1	RES SMD 100kΩ, 1%, 1/10W, 0603, RoHS	Panasonic	ERJ-3EKF1003V
2	R7,R8	RES SMD 1.5kΩ, 1%, 1/10W, 0603, RoHS	Panasonic	ERJ-3EKF1501V
1	R12	RES SMD 200Ω, 1%, 1/10W, 0603, RoHS	Panasonic	ERJ-3EKF2000V
1	R57	RES SMD 47kΩ, 1%, 1/10W, 0603, RoHS	Panasonic	ERJ-3EKF4702V
1	R2	RES SMD 6.65kΩ, 1%, 1/10W, 0603, RoHS	Panasonic	ERJ-3EKF6651V
1	R54	RES 0.01Ω, 1%, 2W, 2512, RoHS	Vishay	WSL2512R0100FEA
1	U1	50A Digital DC/DC PMBus Power Module, RoHS	Renesas	RAA2108502GLG
1	U2	MOSFET N-CH 25V, 39A, TDSON-8, RoHS	Infineon	BSC010NE2LS
1	D1	LED GREEN/RED CLEAR AXIAL SMD, RoHS	Lumex Opto	SSL-LXA3025IGC
1	Q1	MOSFET N-CH 60V, 0.115A, SOT-23, RoHS	ON Semiconductor	2N7002L
1	SW1	SWITCH TOGGLE SPDT, 0.4VA, 20V, RoHS	C&K	GT13MCBE
12	TP1-TP12	TEST POINT PC COMPACT 0.063"D RED, RoHS	Keystone Electronics	5005
1	J2	CONN-SOCKET STRIP, TH, 2x3, 2.54mm, TIN, R/A, RoHS	Samtec	SSQ-103-02-T-D-RA
1	J1	CONN-SOCKET STRIP, TH, 2x3, 2.54mm, TIN, R/A, RoHS	Samtec	TSW-103-08-T-D-RA
3	J11, J15, J16	CONN-HEADER, 2x3, BRKAWY, 2.54mm, TIN, R/A, RoHS	Samtec	TSW-102-07-F-S
4	J7-J10	HDWARE, MTG, CABLE TERMINAL, 6-14AWG, LUG&SCREW, RoHS	Burndy	KPA8CTP

(Continued)

Qty	Reference Designator	Description	Manufacturer	Manufacturer Part
2	J5, J6	Conn Jack Banana Unins Panel Mount, RoHS	Cinch Connectivity Solutions Johnson	108-0740-001

4.5 Board Layout

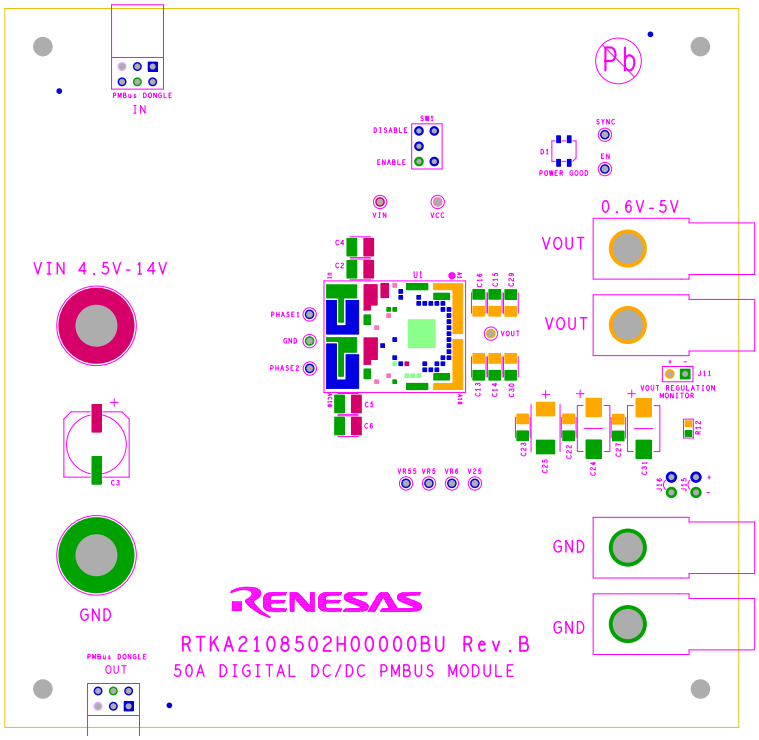

Figure 8. Top Layer Silk Screen

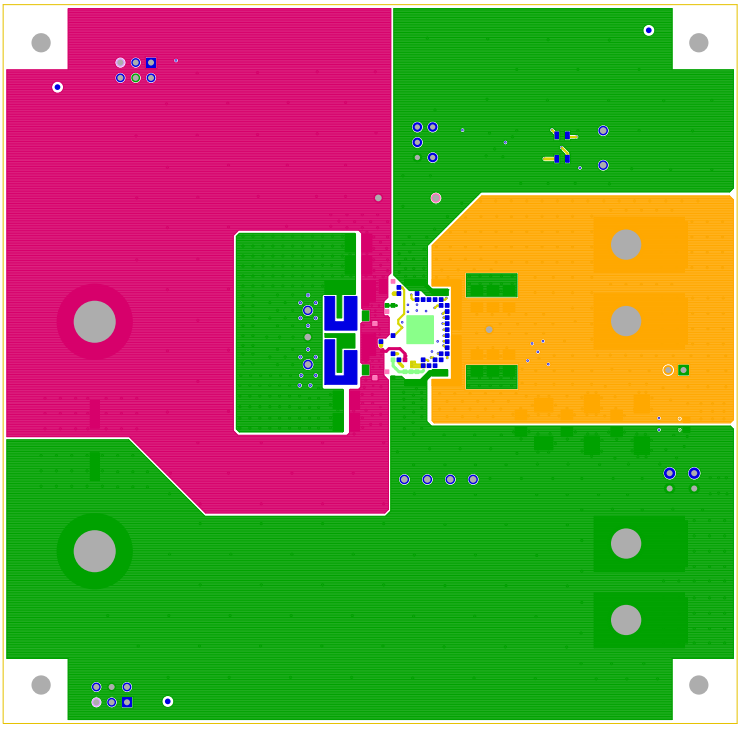

Figure 9. Top Layer Component Side

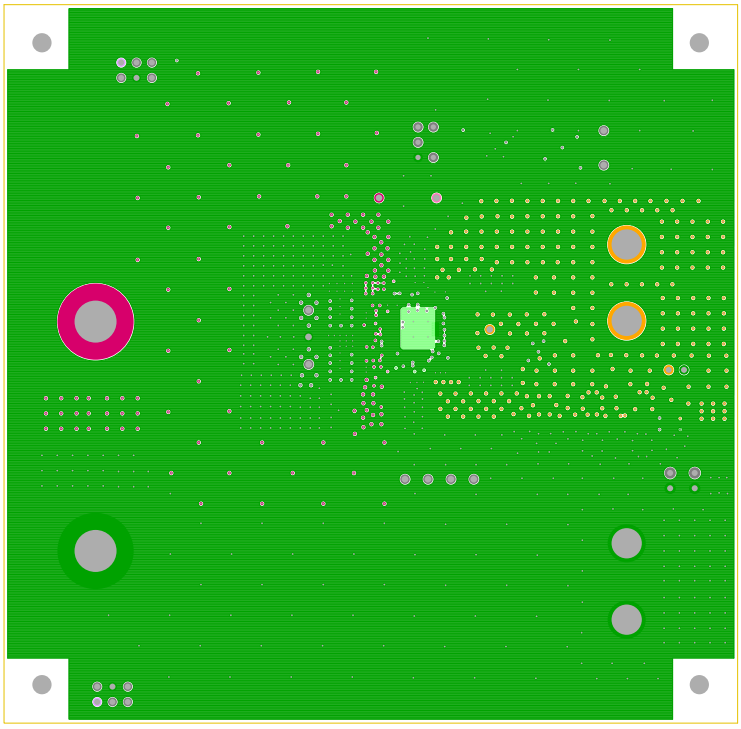

Figure 10. Layer 2

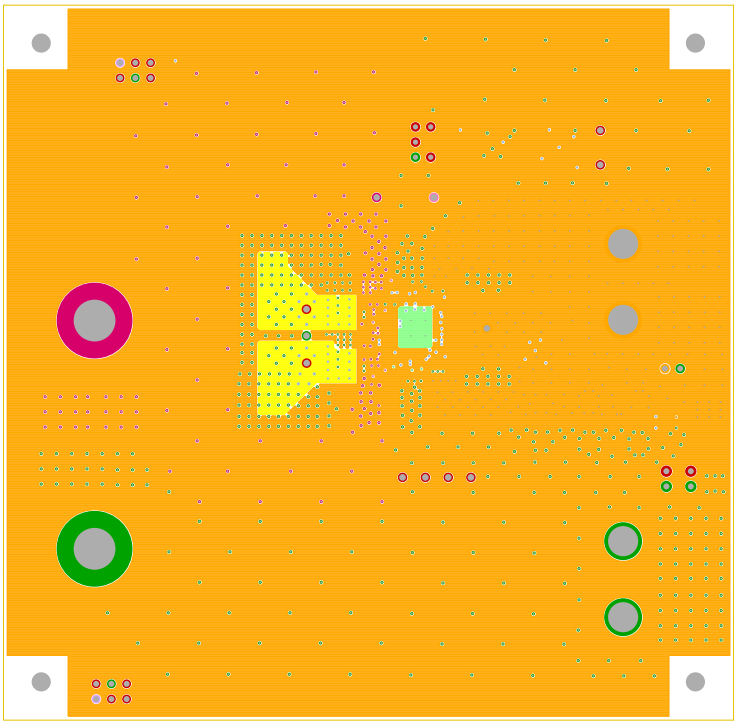

Figure 11. Layer 3

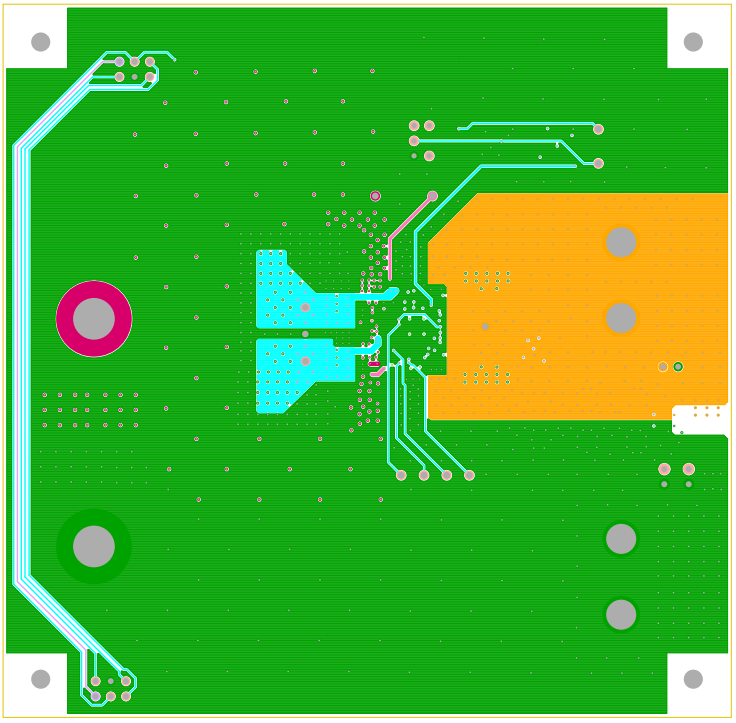

Figure 12. Layer 4

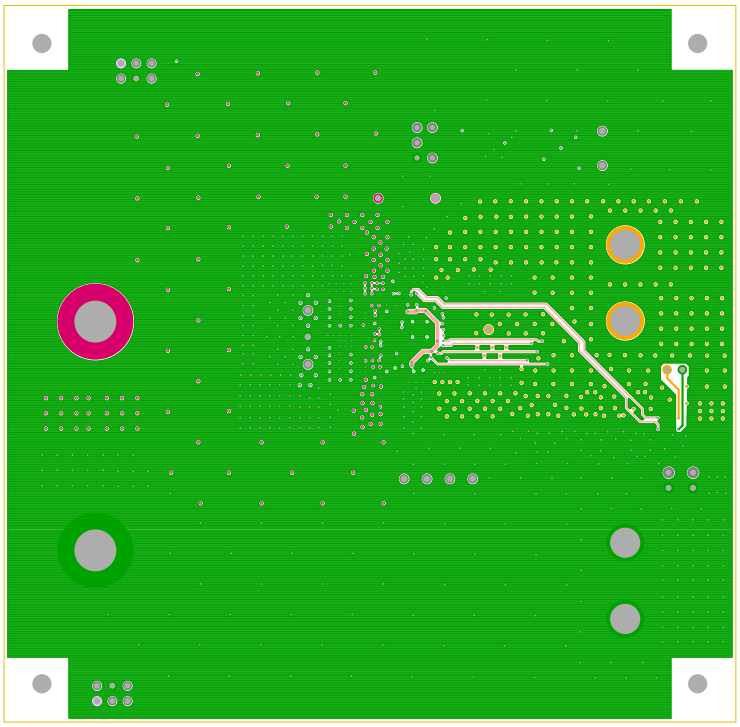

Figure 13. Layer 5

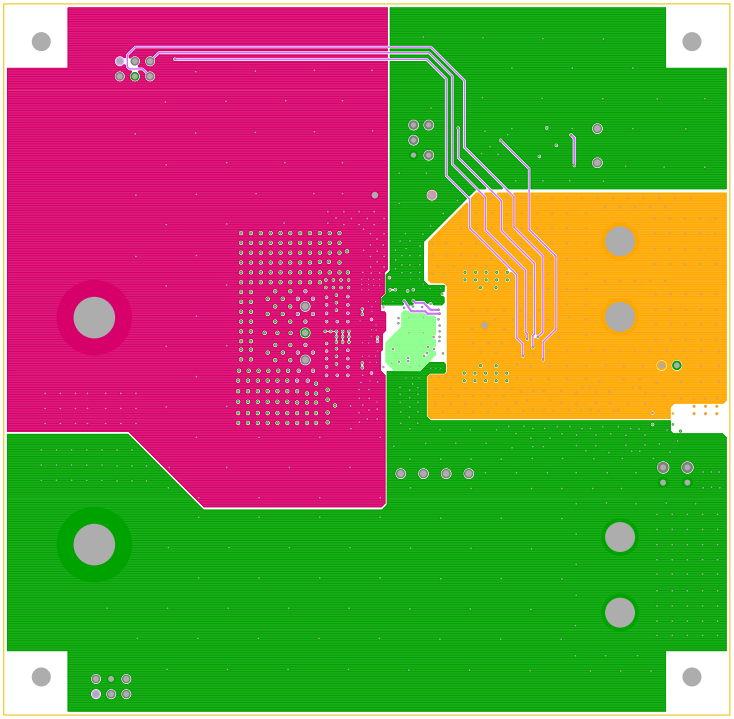

Figure 14. Layer 6

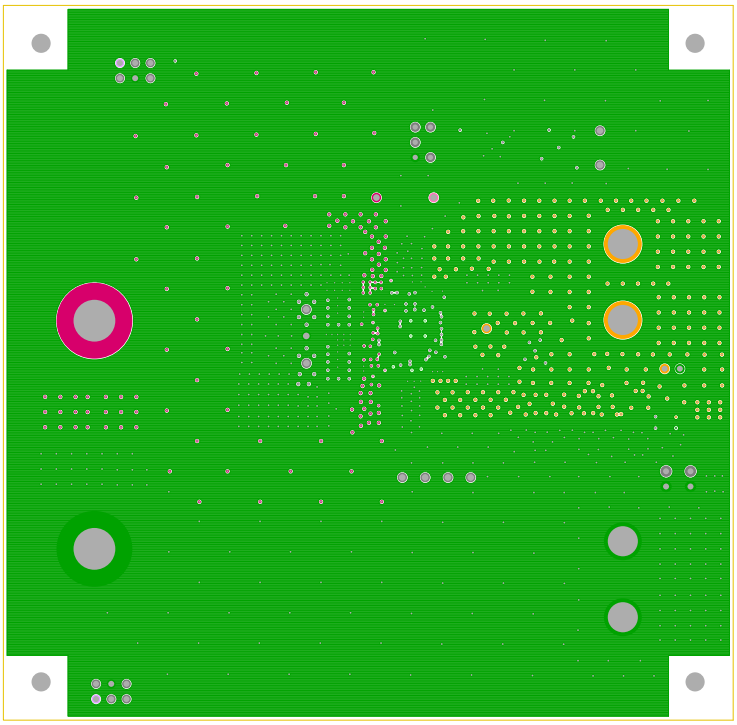

Figure 15. Layer 7

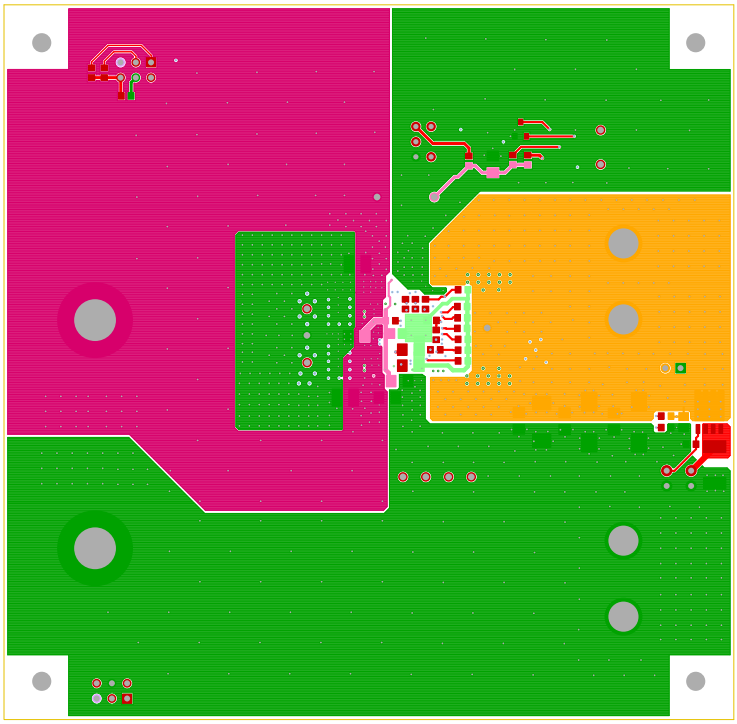

Figure 16. Bottom Layer Solder Side

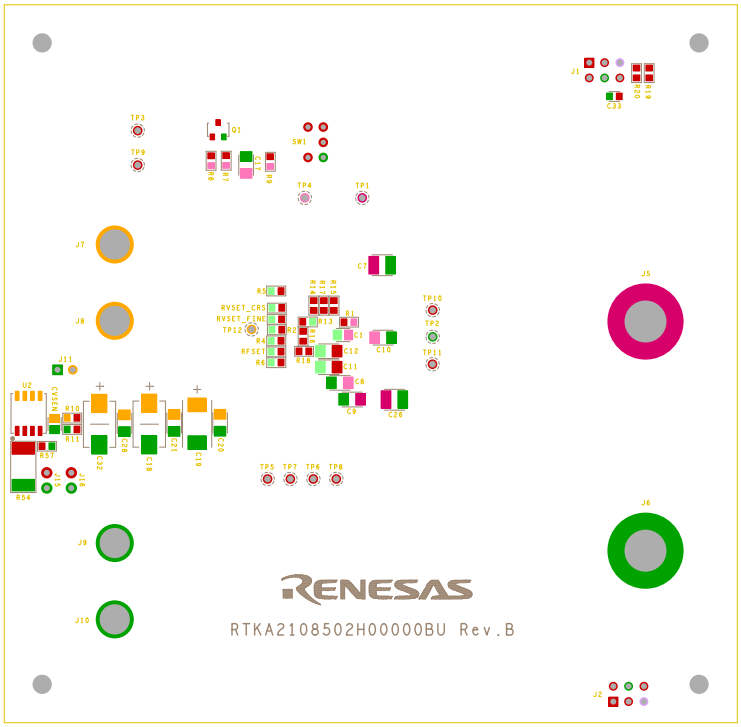

Figure 17. Bottom Silk Screen

5. Typical Performance Curves

The following data was acquired using a RTKA2108502H00000BU evaluation board.

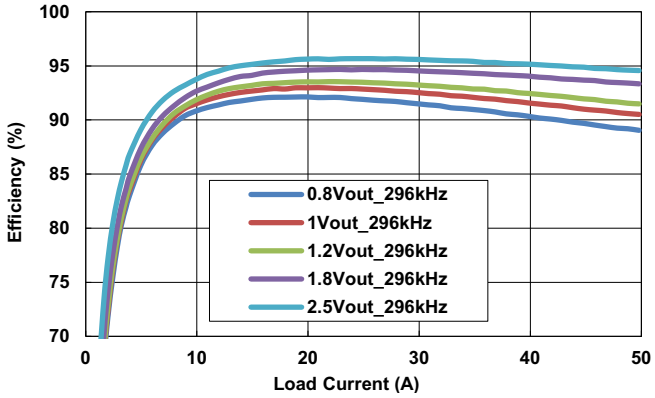

Figure 18. Efficiency vs Output Current at V_{IN} = 5V for Various Output Voltages

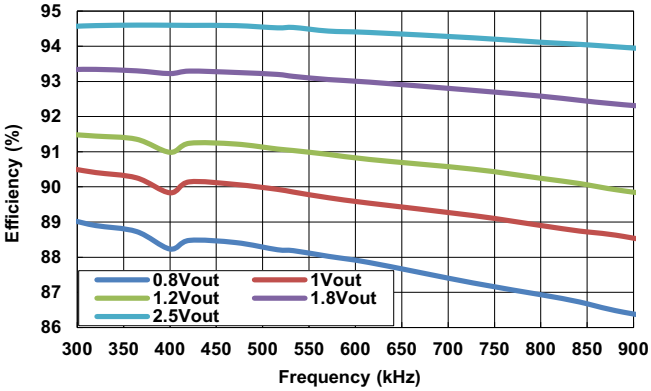

Figure 19. Efficiency vs Output Current at V_{IN} = 5V, I_{OUT} = 50A for Various Output Voltages

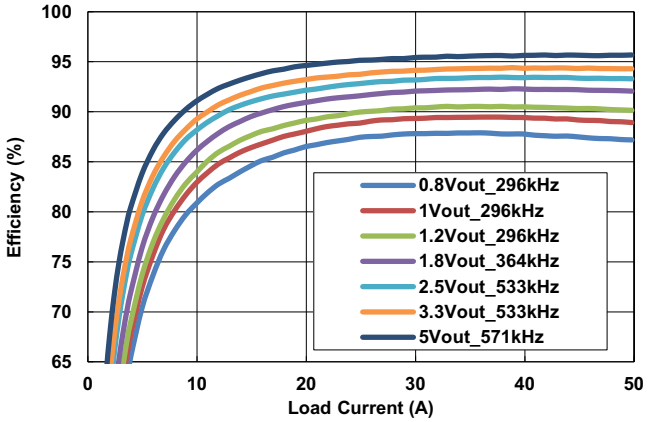

Figure 20. Efficiency vs Output Current at V_{IN} = 12V for Various Output Voltages

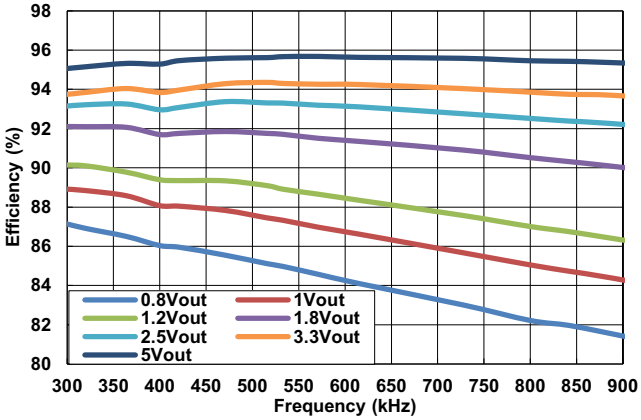

Figure 21. Efficiency vs Output Current at V_{IN} = 12V, I_{OUT} = 50A for Various Output Voltages

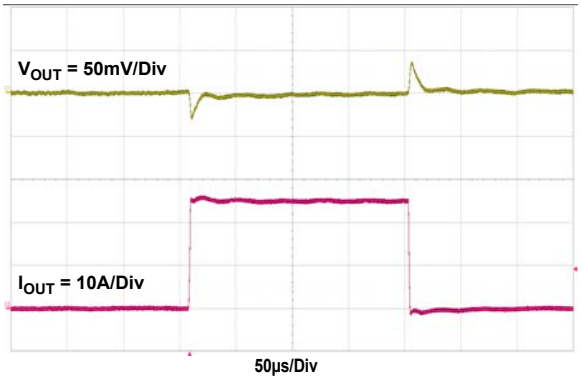

Figure 22. 5V_{IN} to 0.8V_{OUT} Transient response, f_{SW} = 615kHz, C_{OUT} = 10x100µF Ceramic + 3x470µF POSCAP, ACSR Residual = 90, ASCR Gain = 450

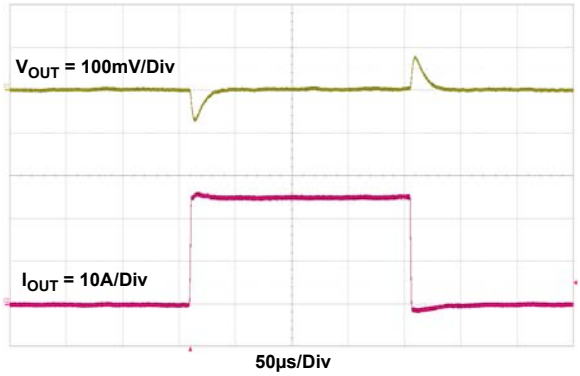

Figure 23. 12V_{IN} to 1.8V_{OUT} Transient Response, f_{SW} = 471kHz, C_{OUT} = 12x100µF Ceramic + 1x470µF POSCAP, ACSR Residual = 100, ASCR Gain = 160

The following data was acquired using a RTKA2108502H00000BU evaluation board. **(Continued)**

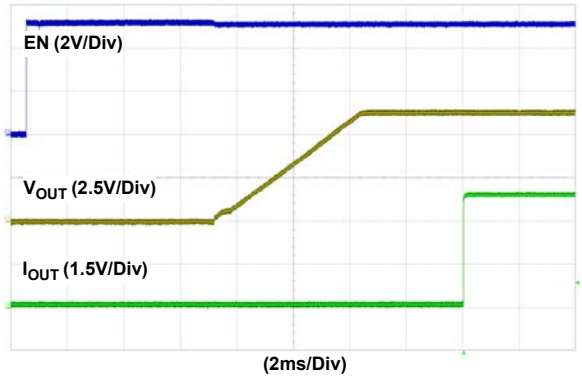

Figure 24. Soft-Start at V_{IN} = 12V, V_{OUT} = 5V, TON_DELAY = 5ms, TON_RISE = 5ms, POWER_GOOD_DELAY = 3ms

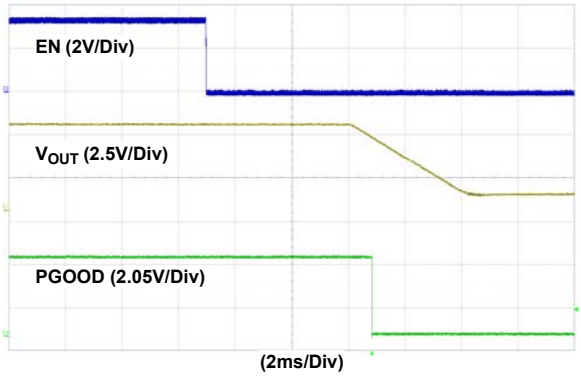

Figure 25. Soft-Stop at V_{IN} = 12V, V_{OUT} = 5V, TOFF_DELAY = 5ms, TOFF_FALL = 5ms, POWER_GOOD_DELAY = 3ms

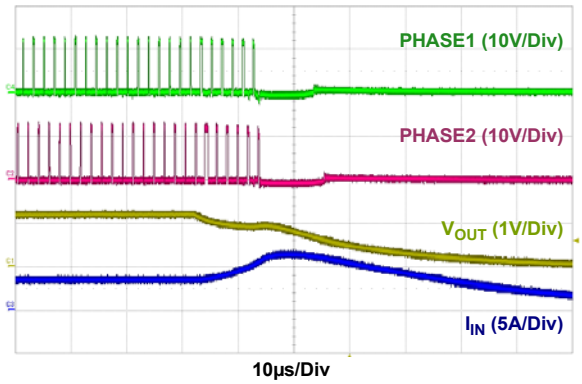

Figure 26. Output Short Circuit Protection at V_{IN} = 12V, V_{OUT} = 1.2V, f_{SW} = 533kHz

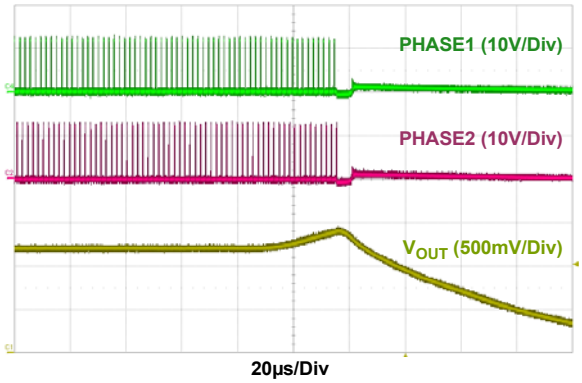

Figure 27. Output Overvoltage Protection at V_{IN} = 12V, V_{OUT} = 1.2V, f_{SW} = 533kHz, VOUT_OV_FAULT_LIMIT = 1.38V

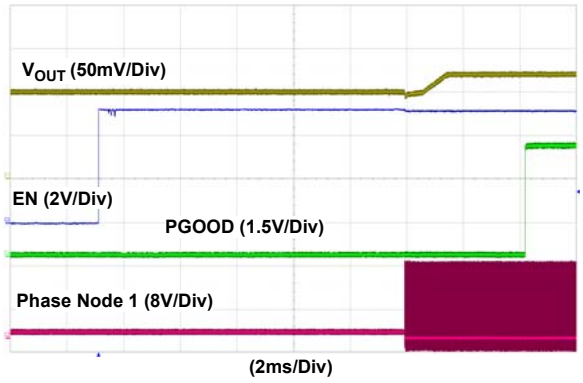

Figure 28. Pre-Bias Startup at V_{IN} = 12V, V_{OUT} = 1.2V, Pre-Bias Voltage = 1V, f_{SW} = 421kHz

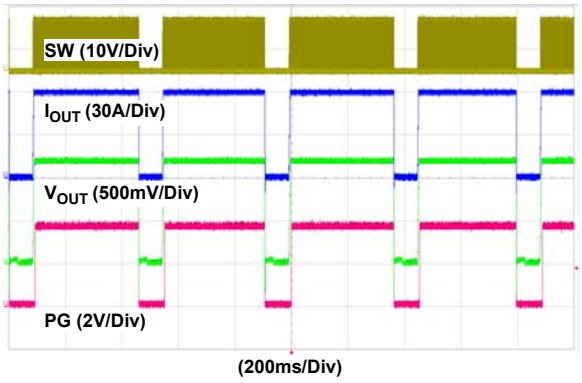

Figure 29. OCP 70ms Fault Retry at V_{IN} = 12V, V_{OUT} = 1.2V, f_{SW} = 421kHz, Programmed I_{OUT} OC Fault Limit = 60A

6. Revision History

Rev.	Date	Description
0.00	Sept 11, 2018	Initial release

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