

General Description

DA9063L is a high current system PMIC suitable for dual- and quad-core processors used in smartphones, tablets, ultra-books, and other handheld and automotive applications that require up to 5 A core processor supply.

DA9063L contains six DC-DC buck converters designed for small external 1 µH inductors capable of supplying in total up to 12 A continuous output (0.3 V to 3.3 V). The buck converters do not require external Schottky diodes. They dynamically optimize their efficiency depending on the load current using an Automatic Sleep mode. The bucks incorporate pin and s/w controlled Dynamic Voltage Control (DVC) to support processor load adaptive adjustment of the supply voltage. One buck can also be used in a DDR memory termination mode.

Five SmartMirror™ programmable LDO regulators are incorporated, rated up to 300 mA. All support remote capacitor placement and can support operation from a low 1.5 V/1.8 V input voltage: this allows the linear regulators to be cascaded with a suitable buck supply to improve overall system efficiency.

Current limited switches provide support for external peripherals such as external accessory or memory cards.

There are four distinct operating modes consuming $< 20 \,\mu\text{A}$. A system monitor watchdog can be enabled in ACTIVE mode.

The DA9063L provides an OTP start-up sequencing engine that offers autonomous hardware system start-up or software controlled start-up and configurable power modes. The on key detects the button press time and offers configurable key lock and application shutdown functions. Up to 16 freely configurable GPIO pins can perform system functions, including: keypad supervision, application wakeup, and timing controlled external regulator, power switch, or other IC enable.

An integrated 9-channel ADC includes advanced voltage monitoring, internal temperature supervision, three general-purpose channels with programmable high/low thresholds, an integrated current source for resistive measurements, and system voltage monitoring with a programmable low-voltage warning. The ADC has 8-bit resolution in AUTO mode and 10-bit resolution in manual conversion mode.

Three RGB-LED driver pins are provided with PWM control.

LDO8 can be configured as a 6-bit, PWM-controlled, vibration motor driver with automatic battery voltage correction.



Key Features

- Input voltage 2.8 V to 5.5 V
- 6 buck converters, 0.3 V to 3.34 V up to 2.5 A
 - □ Dual-phase mode up to 5 A
 - □ Merged mode up to 3 A
 - □ DDR VTT supply mode
 - DVS support
- 3 MHz switching frequency allows use of low profile 1 µH inductors
- Integrated power switches
- 5 LDO regulators, 0.9 V to 3.6 V up to 300 mA
 - □ 1 low-noise
 - □ 1 with DVS
 - □ 4 with current-limited switch option
- I²C serial interface for system communication and control

- System supply monitoring
- Junction temperature monitoring
- Watchdog timer
- Up to 16 flexible GPIO pins for enhanced wakeup and peripheral control
- RGB-LED driver (PWM) with autonomous flashing
- PWM vibration motor driver
- 10-bit ADC with nine channels and configurable alarm thresholds
- Two package variants:
 - □ 100 VFBGA, 8.0 mm x 8.0 mm x 1.0 mm, 0.8 mm pitch, 0.30 mm balls
 - 100 TFBGA, 8.0 mm x 8.0 mm x 1.2 mm,
 0.8 mm pitch, 0.45 mm balls
- Programmable power manager
- Automotive AEC-Q100 Grade 2 available (DA9063L-A)

Applications

- Portable industrial and medical devices
- TV dongle
- Supply for single-, dual- and quad-core application processors, for example the i.MXTM families
- eReaders
- Power supply for FPGAs
- Automotive infotainment/dashboard
- IoT devices



Contents

Ge	eneral	Descrip	tion					
Κe	y Feat	ures		2				
Αŗ	plicati	ions		2				
Cc	ntents	3		3				
1	Refe	rences.		7				
2	Block Diagram7							
3		•						
4			verview					
5	•		naracteristics					
O	5.1		ite Maximum Ratings					
	5.2		nmended Operating Conditions					
	5.2	5.2.1						
	- 0		Power Derating Curves					
	5.3		Current Consumption					
	5.4	•	I/O Characteristics					
	5.5		dog					
	5.6		Vire Control Bus					
	5.7		oltage Regulators					
		5.7.1	LDO3	_				
		5.7.2	LD07					
		5.7.3	LD08					
		5.7.4	LDO9	30				
		5.7.5	LDO11	32				
		5.7.6	LDOCORE	34				
	5.8	DC/DC	Buck Converters	35				
		5.8.1	BUCKCORE1 and BUCKCORE2					
		5.8.2	BUCKPRO	39				
		5.8.3	BUCKMEM	44				
		5.8.4	BUCKIO	47				
		5.8.5	BUCKPERI	49				
		5.8.6	Typical Characteristics	51				
	5.9	Genera	al Purpose ADC	54				
	5.10	Interna	al Oscillator	55				
	5.11	POR, F	Reference Generation, and Voltage Supervision	55				
	5.12		al Supervision					
6	Func		Description					
	6.1		Manager IO Ports					
		6.1.1	On/Off Port (nONKEY)					
		6.1.2	Wakeup Port (CHG_WAKE)					
_	41		, ,					
υā	Datasheet		Revision 2.4	11-Feb-2022				



	6.1.3	Hardware Reset (nOFF, nSHUTDOWN, nONKEY, GPIO14, GPIO15, WATCHDOG)	59
	6.1.4	Reset Output (nRESET)	
	6.1.5	System Enable (SYS_EN)	
	6.1.6	Power Enable (PWR_EN)	
	6.1.7	Power1 Enable (PWR1_EN)	
	6.1.8	GP_FB1, General Purpose Signal 1 (EXT_WAKEUP/READY)	
	6.1.9	GP_FB2, General Purpose Signal 2 (PWR_OK/KEEP_ACT)	
	6.1.10	, , , , , , , , , , , , , , , , , , , ,	
	6.1.11	Supply Rail Fault (nVDD_FAULT)	
	6.1.12	, ,	
	6.1.13		
6.2	Operat	ting Modes	
	6.2.1	ACTIVE Mode	63
	6.2.2	POWERDOWN Mode	63
	6.2.3	RESET Mode	64
	6.2.4	DELIVERY Mode	65
	6.2.5	NO-POWER Mode	65
	6.2.6	Power Commander Mode	65
6.3	Start-U	Jp from NO-POWER Mode	67
	6.3.1	Power-On-Reset (nPOR)	67
6.4	Exiting	Reset Mode and Application Wakeup	67
	6.4.1	VDD_START	67
	6.4.2	Wake-Up Events	68
6.5	Power	Supply Sequencer	70
	6.5.1	Powering Up	70
	6.5.2	Power-Up Timing	73
	6.5.3	Programmable Slot Delays	74
	6.5.4	Powering Down	74
	6.5.5	User Programmable Delay	75
6.6	System	n Monitor (Watchdog)	77
6.7	GPIO E	Extender	77
6.8	Contro	l Interface	81
6.9	Voltage	e Regulators	85
	6.9.1	Regulators Controlled by Software	85
	6.9.2	Regulators Controlled by Hardware	86
	6.9.3	Power Sequencer Control of LDOs	86
	6.9.4	Dynamic Voltage Control	87
	6.9.5	Pull-Down Resistor	87
	6.9.6	Bypass Mode and Current Limit	88
	6.9.7	LDO Supply from Buck Converter	88
	6.9.8	LDO Sleep Mode for Reduced IouT	88
Datashe	eet	Revision 2.4	1-Feb-2022



		6.9.9	Vibration Motor Driver	89				
		6.9.10	Core Regulator LDOCORE	89				
	6.10	DC/DC E	Buck Converters	90				
		6.10.1	BUCKCORE1, BUCKCORE2, and BUCKPRO	92				
		6.10.2	BUCKPRO in DDR Memory Bus Termination Mode	93				
		6.10.3	BUCKMEM and BUCKIO in Merged Mode	94				
	6.11	General	Purpose ADC	95				
		6.11.1	ADC Overview	95				
		6.11.2	ADC Input MUX					
		6.11.3	Manual Conversion Mode	96				
		6.11.4	Automatic Measurements Scheduler	96				
		6.11.5	Fixed Threshold Comparator	100				
	6.12	-	ole Frequency Internal Oscillator					
	6.13		ce Voltage Generation: VREF, VLNREF					
	6.14	Thermal	Supervision	101				
	6.15	Main Sys	stem Rail Voltage Supervision	101				
7	Regis	ster Map.		103				
8	Defin	itions		116				
	8.1							
	8.2	Regulato	or Parameter - Dropout Voltage	117				
	117							
	8.4	Regulato	or Parameter - Line Regulation	117				
	8.5	Regulato	or Parameter - Load Regulation	118				
9	Packa	age Infor	mation	119				
	9.1	Package Outlines						
	9.2	2 Package Marking						
	9.3 Moisture Sensitivity Level							
10	Orde	ring Infor	rmation	122				
		•	Ordering Information					
11			ıformation					
	11.1		or Selection					
	11.2	•						
	11.3	Resistor						
12	Lavo	ut Guidel	lines	126				
	12.1	•						
12.2 LDOs and Switched Mode Supplies								
			Connection, Land Pad, and Stencil Design					
αA			ster Descriptions					
4-1	A.1	•	Page Control					
	A.2	•	r Page 0					
Dat	ashee	et	Revision 2.4	11-Feb-2022				

DA9063L



	A.2.1	Power Manager Control and Monitoring	127
	A.2.2	GPIO Control	136
	A.2.3	Regulator Control	143
	A.2.4	GPADC	151
	A.2.5	ADC Results	
A.3	Registe	r Page 1	154
	A.3.1	Power Sequencer	154
	A.3.2	Regulator Settings	
	A.3.3	High Power GPO PWM	184
	A.3.4	GPADC Thresholds	185
A.4	Registe	r Page 2	188
	A.4.1	OTP	188
	A.4.2	Customer Trim and Configuration	189
A.5	Registe	r Page 3	202



1 References

Dialog Semiconductor technical documentation is available on the web site: Power Management ICs (PMICs).

- AN-PM-068, Application Note, VBBAT Current in RTC or DELIVERY Modes, Dialog Semiconductor.
- [2] AN-PM-024, Application Note, DA9063 Voltage Monitoring, Dialog Semiconductor.
- [3] AN-PM-010, Application Note, PCB Layout Guidelines, Dialog Semiconductor.

2 Block Diagram

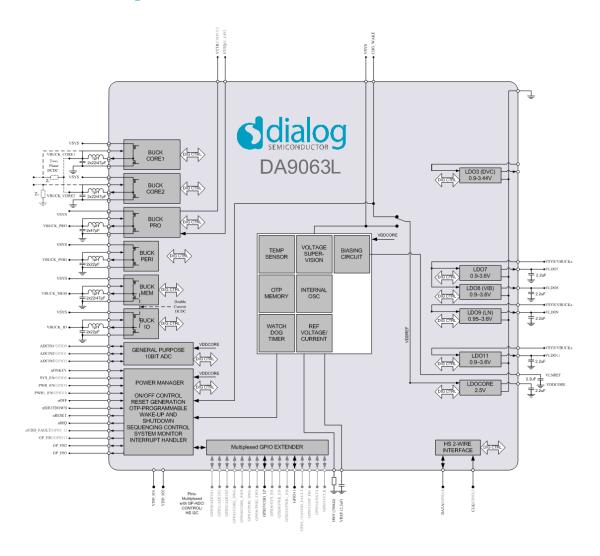


Figure 1: Block Diagram



3 Pinout

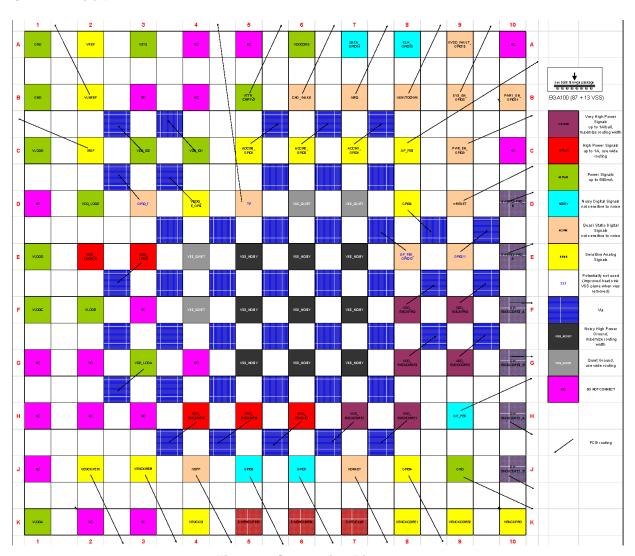


Figure 2: Connection Diagram

Table 1: Pin Type Definition

Pin Type	Description	Pin Type	Description
DI	Digital input	Al	Analog input
DO	Digital output	AO	Analog output
DIO	Digital input/output	AIO	Analog input/output
PWR	Power supply	GND	Ground connection



Table 2: Pin Description

Pin	Pin name	Alternate Function	Type (Table 1)	Description	
Power	· Manager				
A9	nVDD_FAULT	GPIO12	DO/DIO	Indication for low supply voltage / GPIO12 / VDD_MON controlled GPO	
B6	CHG_WAKE		DI/PWR	trigger a start-up and temporary supply voltage for PMIC (VBUS_PROT in case of a inserted supply until charger Buck provides power to Vsys).	
	IDO		5.0	Connect to GND if not used.	
B7	nIRQ		DO	IRQ line for host	
B8	nSHUTDOWN		DI	Active-low input from switch or host to initiate shutdown	
B9	SYS_EN	GPIO8	DI/DIO	Hardware enable of power domain SYSTEM/GPIO8	
B10	PWR1_EN	GPIO10	DI/DIO	Hardware enable of power domain POWER1/GPIO10 with high power output / input for power sequencer WAIT ID	
C3	VDD_IO2		PWR	Alternate supply I/O voltage	
C4	VDD_IO1		PWR	First supply I/O voltage rail	
C8	GP_FB2		DO/DI	PWR_OK status indicator: all supervised regulators are in-range / HW input for watchdog supervision / dual-phase BUCKCORE voltage sense at output capacitor	
C9	PWR_EN	GPIO9	DI/DIO	Hardware enable of power domain power / sequencer controlled GPO	
D3	GPIO7		DIO	Sequencer controlled GPO	
D5	TP		DIO	Test pin: enables power commander boot mode and supply pin for OTP fusing voltage	
D9	nRESET		DO	Active low reset for host	
E8	GP_FB1	GPIO13	DO/DIO	Status indication for host of a valid wakeup event (EXT_WAKEUP) / indicator for ongoing power mode transition (READY) / GPIO13, regulator HW control	
E9	GPIO11		DIO	GPIO11 with high power output and blinking feature	
H9	GP_FB3		DI	nVIB_BREAK control signal for vibration motor driver (LDO8)	
J4	nOFF		DI	Active-low input from error indication line to initiate fast emergency shutdown	
J7	nONKEY		DI	On/off key with optional long press shutdown	
	1		1	1	



Pin	Pin name	Alternate Function	Type (Table 1)	Description	
2-Wire	Interface				
A4					
A5				Connect to ground	
A7	DATA	GPIO14	DIO	HS 2-WIRE data / GPIO14 (optional reset if long press in parallel with GPI15) with high power output and blinking feature	
A8	CLK	GPIO15	DI	HS 2-WIRE clock / GPIO15 (optional reset if long press in parallel with GPI14) with high power output and blinking feature	
В3				Connect to ground	
В4				Connect to ground	
Voltag	e Regulators				
А3	VSYS		PWR	Supply voltage for PMIC and input for voltage supervision (decouple with 1.0 µF)	
A6	VDDCORE		AO	Regulated supply for internal circuitry (2.2 V/2.5 V) (decouple with 2.2 µF)	
C1	VLDO11		AO	Output voltage from LDO11	
D1					
D2	VDD_LDO11		PWR	Supply voltage for LDO11	
E1	VLDO9		AO	Output voltage from LDO9	
E2	VDD_LDO7_8		PWR	Supply voltage for LDO7 and LDO8	
E3	VDD_LDO9		PWR	Supply voltage for LDO9	
F1	VLDO8		AO	Output voltage from LDO8	
F2	VLDO7		AO	Output voltage from LDO7	
F3				Connect to ground	
G1					
G2				Connect to ground	
G3	VDD_LDO3		PWR	Supply voltage for LDO3	
G4				Connect to ground	
H1					
H2				Connect to ground	
НЗ				Connect to ground	
J1					
K1	VLDO3		AO	Output voltage from LDO3	



Pin	Pin name	Alternate Function	Type (Table 1)	Description
K2				
K3				



Pin	Pin name	Alternate Function	Type (Table 1)	Description	
DC/DC	Buck Converters				
A1				Connect to ground	
A2	VREF		AIO	Filter node for internal reference voltage (decouple with 0.22 µF)	
A10					
B1				Connect to ground	
B2	VLNREF			Filter node for LN (low noise) (decouple with 0.22 µF)	
B5	VTTR	CMP1V2	AO/DO	Memory bus termination reference voltage (50 % of VDDQ), COMP1V2 controlled GPO	
C2	IREF		АО	Connection for bias setting (configure with high precision 200 k Ω resistor)	
C5	ADCIN2	GPIO1	AI/DIO	Connection to GPADC channel 2 with 1.2 V HW comparator IRQ/GPIO1, regulator HW control	
C6	ADCIN3	GPIO2	AI/DIO	Connection to GPADC channel 3/GPIO2, regulator HW control	
C7	ADCIN1	GPIO0	AI/DIO	Connection to GPADC auto channel 1 with threshold IRQ and resistor measurement option/GPIO0	
C10					
D4	VDDQ	E_GPI2	AI/DO	BUCKPRO target voltage sense port / state of E_GPI2 controlled GPO	
D8	PERI_SWS	GPIO6	AI/DIO	BUCKPERI sense node/GPIO6	
D10	SWBUCKPRO_B		AO	Switching node for BUCKPRO (full-current)	
E10	SWBUCKPRO_A		AO	Switching node for BUCKPRO (half-current)	
F8, F9	VDD_BUCKPRO		PWR	Supply voltage for buck To be connected to VSYS	
F10	SWBUCKCORE2_A		AO	Switching node for BUCKCORE2 (half-current)	
G8, G9	VDD_BUCKCORE2		PWR	Supply voltage for buck To be connected to VSYS	
G10	SWBUCKCORE2_B		AO	Switching node for BUCKCORE2 (full-current)	
H4	VDD_BUCKPERI		PWR	Supply voltage for buck To be connected to VSYS	
H5	VDD_BUCKMEM		PWR	Supply voltage for buck To be connected to VSYS	



Pin	Pin name	Alternate Function	Type (Table 1)	Description	
H6	VDD_BUCKIO		PWR	Supply voltage for buck To be connected to VSYS	
H7, H8	VDD_BUCKCORE1		PWR	Supply voltage for buck To be connected to VSYS	
H10	SWBUCKCORE1_A		AO	Switching node for BUCKCORE1 (half-current)	
J2	VBUCKPERI		Al	Sense node for BUCKPERI	
J3	VBUCKMEM		Al	Sense node for DC/DC BUCKMEM	
J5	GPIO5		DIO	GPIO5	
J6	GPIO3		DIO	GPIO3	
J8	CORE_SWS	GPIO4	AI/DIO	BUCKCORE sense node from output capacitor of dual-phase BUCKCORE1/2 connection of internal switch to the output of GPIO4	
J10	SWBUCKCORE1_B		AO	Switching node for BUCKCORE1 (full-current)	
K4	VBUCKIO		Al	Sense node for BUCKIO	
K5	SWBUCKPERI		AO	Switching node for BUCKPERI	
K6	SWBUCKMEM		AO	Switching node for BUCKMEM	
K7	SWBUCKIO		AO	Switching node for BUCKIO To be connected to SWBUCKMEM for buck merge	
K8	VBUCKCORE1		Al	Sense node for BUCKCORE1	
K9	VBUCKCORE2		Al	Sense node for BUCKCORE2	
K10	VBUCKPRO		Al	Sense node for BUCKPRO	
J9				Connect to ground	
vss	1	П	П		
D6 to D7, E4, F4	GND		GND	VSS_LDO, VSS_ADC, VSS_CORE, VSUB	
E5 to E7, F5 to F7, G5 to G7	GND		GND	VSS_BUCKCORE1_A, VSS_BUCKCORE1_B, VSS_BUCKCORE2_A, VSS_BUCKCORE2_B, VSS_BUCK_PRO_A, VSS_BUCK_PRO_B, VSS_BUCK_IO, VSS_BUCK_MEM, VSS_BUCK_PERI	



4 Regulator Overview

Table 3: Regulators

Regulator	Supplied Pins	Supplied Voltage (V)	Supplied Max. Current (mA)	External Component	Notes
BUCKCORE1	VBUCKCORE1	0.3 to 1.57	1250/ 2500 (full-current mode) Note 1	1.0 μH/ 44 μF / 88 μF	GPIO and host interface- controlled DVC with variable slew rate (10 mV in [0.5, 1.0, 2.0, 4.0] µs) 10 mV steps < 0.7 V PFM mode only 2500 mA in full-current mode (double pass device and current limit) Provides dual-phase buck with up to 5 A if combined with BUCKCORE2
BUCKCORE2	VBUCKCORE2	0.3 to 1.57	1250/ 2500 (full-current mode) Note 1	1.0 μH/ 44/88 μF	 GPIO and host interface-controlled DVC with variable slew rate (10 mV in [0.5/1.0/2.0/4.0] μs) 10 mV steps < 0.7 V PFM mode only 2500 mA in full-current mode (double pass device and current limit) Provides dual-phase Buck if combined with BUCKCORE1
BUCKPRO	VBUCKPRO	0.53 to 1.80	1250/ 2500 (full-current mode) Note 1	1.0 μH/ 44/88 μF	 GPIO and host interface-controlled DVC with variable slew rate, (10 mV in [0.5/1.0/2.0/4.0] μs) 10 mV steps and VTT regulator mode < 0.7 V PFM mode only 2500 mA in full-current mode (double pass device and current limit)
BUCKMEM	VBUCKMEM	0.8 to 3.34	1500 Note 1	1.0 μH/ 44 μF	 GPIO and host interface-controlled DVC with variable slew rate (10 mV in [0.5/1.0/2.0/4.0] µs) 20 mV steps Can be merged with BUCK_IO towards single



Regulator	Supplied Pins	Supplied Voltage (V)	Supplied Max. Current (mA)	External Component	Notes
					buck with up to 3 A output current
BUCKIO	VBUCKIO	0.8 to 3.34	1500 Note 1	1.0 μH/ 44 μF	 GPIO and host interface-controlled DVC with variable slew rate (10 mV in [0.5/1.0/2.0/4.0] µs) 20 mV steps, can be merged with BUCK_MEM
BUCKPERI	VBUCKPERI	0.8 to 3.34	1500 Note 1	1.0 μH/ 44 μF	 GPIO and host interface- controlled DVC with variable slew rate (10 mV in [0.5/1.0/2.0/4.0] µs) 20 mV steps
LDO3	VLDO3	0.9 to 3.44	200	2.2 μF	 Bypass mode GPIO and host interface-controlled DVC with variable slew rate (10 mV in [0.5/1.0/2.0/4.0] µs) 20 mV steps
LDO7	VLDO7	0.9 to 3.6	200	2.2 μF	Bypass mode50 mV stepsCommon supply with LDO8
LDO8	VLDO8	0.9 to 3.6	200	2.2 μF	 Bypass and switching vibration motor driver mode 50 mV steps Common supply with LDO7
LDO9	VLDO9	0.95 to 3.6	200	2.2 µF	Low noise50 mV stepsOTP trimmed
LDO11	VLDO11	0.9 to 3.6	300	2.2 μF	Bypass mode 50 mV steps
LDOCORE	Internal PMIC supply	2.5 ± 2 % accuracy	4	2.2 μF	Internal LDO OTP trimmed

Note 1 For short durations to meet peak current requirements I_{OUT} can be operated at up to 20 % higher than the specified maximum operating condition. The part should not be operated in this mode for extended periods and is not guaranteed for continuous operation.



5 Electrical Characteristics

5.1 Absolute Maximum Ratings

Table 4 lists the absolute maximum ratings of the device. Exceeding these ratings may cause permanent damage to the device. Device functionality is only guaranteed under the conditions listed in Sections 5.1 and 5.2. Operating the device in conditions exceeding those listed in Table 5 but compliant with the absolute maximum ratings listed in Table 4, for extended periods of time may affect device reliability.

Table 4: Absolute Maximum Ratings

Parameter	Description	Conditions	Min	Тур	Max	Unit
T _{STG}	Storage temperature		-65		+150	°C
TJ	Junction temperature	Note 1	-40		+150	°C
V _{SYS} , V _{CHG_WAKE}			-0.3		6.0	٧
V_{TP}	Supply voltage	Note 3	-0.3		8.0	V
All other pins		Note 2	-0.3		V _{DDREF} + 0.3	٧
Vesd_HBM	ESD protection - Human Body Model (HBM)		2000			V
V _{ESD_CDM}	ESD protection -	Corner pins	750			
	Charged Device Model (CDM)	All other pins	500			V

- **Note 1** See Section 5.12 and Section 6.14.
- Note 2 Maximum V_{DDREF} = 6 V. An internal node V_{DDREF} is defined as the higher rail of CHG_WAKE and VSYS.
- **Note 3** Voltage on TP pin should be 0 V except during in-circuit programming.



5.2 Recommended Operating Conditions

All voltages are referenced to VSS unless otherwise stated. Currents flowing into DA9063L are deemed positive; currents flowing out are deemed negative. All parameters are valid over the recommended temperature range and power supply range unless otherwise stated. Please note that the power dissipation must be limited to avoid overheating of DA9063L.

Table 5: Recommended Operating Conditions

Parameter	Description	Conditions	Min	Тур	Max	Unit
TJ	Junction temperature		-40		+125	°C
V _{SYS} , V _{CHG_WAKE}	Supply voltage		0		5.5	V
V _{DD_IO1/2}	Supply voltage IO	Note 1	1.2		3.6	V
V _{TP}	TP programming supply	Note 5	7.25	7.5	7.75	V
Гоит	BUCKCORE1, BUCKCORE2, BUCKPRO	Note 2			2500	mA
Гоит	BUCKMEM, BUCKIO, BUCKPERI	Note 2			1500	mA
R _{θ JA}	Thermal resistance	100 VFBGA package Note 3		27.7		°C/W
Kθ_JA	junction to ambient	100 TFBGA package Note 4		17.2		°C/W
R_{θ_JB}	Thermal resistance junction to Board	100 TFBGA package Note 4		11.1		°C/W
R _{θ_} Jc	Thermal resistance junction to case	100 TFBGA package Note 4		13.29		°C/W
	Maximum power	100 VFBGA Derating factor above T _A = 70 °C: 36.1 mW/°C (1/θ _{JA})	2000			mW
P _D dissipation, 5.2.1	dissipation, see Section 5.2.1	100 TFBGA Derating factor above T _A = 70 °C: 58.1 mW/°C (1/θ _{JA})	3770			mW

- Note 1 V_{DDIO1/2} must not exceed V_{DDREF}.
- **Note 2** For short durations to meet peak current requirements I_{OUT} can be operated at up to 20 % higher than the specified maximum operating condition. The part should not be operated in this mode for extended periods and is not guaranteed for continuous operation.
- Note 3 Obtained from package thermal simulations, JEDEC 2S2P four layer board (76.2 mm x 114 mm x 1.6 mm), 70 μm (2 oz) copper thickness power planes, 35 μm (1 oz) copper thickness signal layer traces, natural convection (still air), see Section 9.1.
- Note 4 Obtained from package thermal simulations, JEDEC 2S2P four layer board (101.6 mm x 114.3 mm x 1.6 mm), 70 μm (2 oz) copper thickness power planes, 35 μm (1 oz) copper thickness signal layer traces, natural convection (still air), see Section 9.1.



Note 5 Voltage on TP pin should be 0 V except during in-circuit programming.

5.2.1 Power Derating Curves

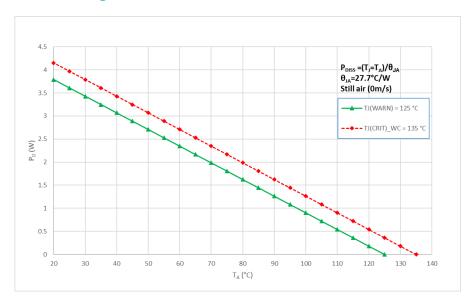


Figure 3: 100 VFBGA Power Derating Curve

Table 6: Typical Temperatures

	T _A = 70 °C	T _A = 85 °C	T _A = 105 °C
T _{J_WARN}	$P_D = 1.99 \text{ W}$	P _D = 1.44 W	$P_D = 0.72 \text{ W}$
T _{J_CRIT_WC}	$P_D = 2.35 \text{ W}$	P _D = 1.81 W	P _D = 1.083 W



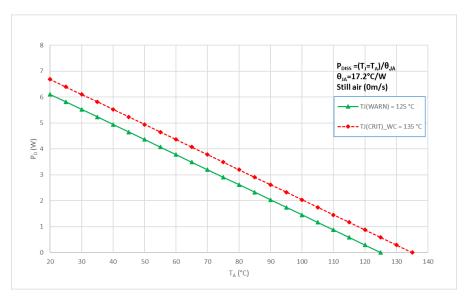


Figure 4:100 TFBGA Power Derating Curve

Table 7: Typical Temperatures

	T _A = 70 °C	T _A = 85 °C	T _A = 105 °C
T _{J_WARN}	$P_D = 3.19 \text{ W}$	P _D = 2.32 W	P _D = 1.16 W
T _{J_CRIT}	$P_D = 3.77 \text{ W}$	P _D = 2.9 W	P _D = 1.74 W



5.3 Typical Current Consumption

Table 8: Typical Current Consumption

Operating Mode	Conditions (Note 1)	Battery	Unit
NO-POWER mode (POR)	2.4 V > V _{DDREF} > 1.5 V	16	μΑ
DELIVERY mode	V _{DDREF} > 1.5 V	1.8	μA
RESET mode	V _{DDREF} > 2.2 V, supplies off (except LDOCORE), pulsed mode	18	μΑ
LOW-POWER mode	V _{SYS} > V _{DD_FAULT_LOWER} , supplies off (except LDOCORE), all blocks in POWERDOWN mode, pulsed mode with limited parametric compliance	12	μA
POWERDOWN mode (Hibernate)	V _{SYS} > V _{DD_FAULT_LOWER} , supplies off (except LDOCORE), all blocks in POWERDOWN mode	28	μΑ
POWERDOWN mode (Standby)	BUCKCORE, LDOCORE, and LD03, 8, 11 enabled, GPIO unit on (Note 2)	60	μΑ
ACTIVE mode	All supplies, GPIO, and GPADC off	200	μA
ACTIVE mode	All supplies, GPIO, and GPADC on	320	μΑ

Note 1 nONKEY/CHG_WAKE/V_{DDREF} detection circuit is enabled in all modes.

5.4 Digital I/O Characteristics

Table 9: Digital I/O Electrical Characteristics, $T_J = -40$ °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
	GPI0 to GPI15, CLK, DATA, nOFF, nSHUTDOWN	VDDCORE mode	1.0		Vsys	
ViH	SYS_EN, PWR_EN, PWR1_EN Input High Voltage	VDD_IO2 mode	0.7*VDD_IO2		V _{SYS}	V
VIL	GPI0 to GPI15, CLK, DATA, nOFF, nSHUTDOWN SYS EN, PWR EN,	VDDCORE mode	-0.3		0.4	>
- TE	PWR1_EN Input Low Voltage	VDD_IO2 mode	-0.3		0.3*VDD_IO2	•
V _{IH}	nONKEY, CHG_WAKE Input High Voltage		1.0		Vsys	V
V _{IL}	nONKEY, CHG_WAKE Input Low Voltage		-0.3		0.4	V
Vон	GPO0 to GPO15, nVDD_FAULT, nRESET, nIRQ, E_GPI_2, COMP1V2	I _{OUT} = 1 mA VDD_IO1 mode ≥ 1.5 V	0.8*VDD_IO1			V

Note 2 Regulators are running in sleep mode.



Parameter	Description	Conditions	Min	Тур	Max	Unit
	Output High Voltage	I _{OUT} = 1 mA VDD_IO2 mode ≥ 1.5 V	0.8*VDD_IO2			
Vон	GPO1, 3 to 6, 10, and 12 to 15, DATA, nRESET, nIRQ, PWR_OK (open drain mode) Output High Voltage	Open drain			VDDREF	V
VoL	GPO0 to GPO15, nVDD_FAULT, nRESET (Note 1), nIRQ (Note 1), PWR_OK, E_GPI_2, COMP1V2, Output Low Voltage	louτ = 1 mA			0.3	V
C _{IN}	CLK, DATA Input capacitance				10	pF
	Sink current capability GPO 10, 11, 14, 15	V _{GPIO} = 0.5 V Note 2		11		mA
	Source current capability GPO 10, 11,14,15	V _{GPIO} = 0.8*VDD_IO1/2 Note 2		-4		mA
	Sink current capability GPO 0 to 9, 12, 13	V _{GPIO} = 0.3 V		1		mA
	Source current capability GPO 0 to 9, 12, 13	V _{GPIO} = 0.8*VDD_IO1/2 Note 3		-1		mA
	GPI pull-down resistor		50	100	250	kΩ
		VDD_IO1/2 = 1.5 V	60	180	310	
	GPO pull-up resistor Note 4	VDD_IO1/2 = 1.8 V	45	120	190	kΩ
		VDD_IO1/2 = 3.3 V	20	40	60	

Note 1 Electrical characteristics are guaranteed down to $V_{DDREF} = 2.0 \text{ V}$ (VPOR_LOWER). For lower voltages the port continues operating with reduced performance.

Note 2 At low V_{DDREF} values and high temperatures, the sink current capability is reduced.

Note 3 For VDD_IO1/2 < 1.5 V the source current capability is reduced.

Note 4 V(PAD) = 0 V.



5.5 Watchdog

Table 10: Watchdog, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
T _{WD_MIN} Minimum Watchd	Minimum Watahdag tima	External 32 kHz oscillator			0.11	S
	wilnimum watchdog time	Internal 25 kHz oscillator			0.2	s
т	WD_MAX Maximum Watchdog time	External 32 kHz oscillator	2			s
I WD_MAX		Internal 25 kHz oscillator	2.5			

5.6 HS 2-Wire Control Bus

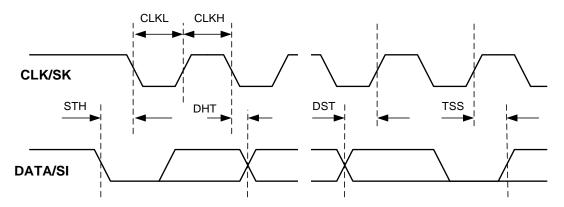


Figure 5: 2-Wire Bus Timing

Table 11: HS 2-Wire Control Bus Electrical Characteristics, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
	Bus free time STOP to START		0.5			μs
Св	Bus line capacitive load				150	pF
Standard/Fast/	/Fast+ Mode					
	CLK clock frequency	Note 1	0		1000	kHz
	Bus free time STOP to START		0.5			μs
	Start condition set-up time		0.26			μs
STH	Start condition hold time		0.26			μs
CLKL	CLK low time		0.5			μs
CLKH	CLK high time		0.26			μs
	2-WIRE CLK and DATA rise time	(input requirement)			1000	ns
	2-WIRE CLK and DATA fall time	(input requirement)			300	ns
DST	Data set-up time		50			ns



Parameter	Description	Conditions	Min	Тур	Max	Unit
DHT	Data hold-time		0			ns
	Data valid time				0.45	μs
	Data valid time acknowledge				0.45	μs
TSS	Stop condition set-up time		0.26			μs
High Speed Me	ode					
	CLK clock frequency	Requires VDDIO ≥ 1.8 V	0		3400	kHz
		Note 1				
	Start condition set-up time		160			ns
STH	Start condition hold time		160			ns
CLKL	CLK low time		160			ns
CLKH	CLK high time		60			ns
	2-WIRE CLKH and SDAH rise/fall time	Input requirement			160	ns
DST	Data set-up time		10			ns
DHT	Data hold-time		0			ns
TSS	Stop condition set-up time		160			ns

Note 1 Minimum clock frequency is 10 kHz if 2WIRE_TO is enabled

5.7 LDO Voltage Regulators

5.7.1 LDO3

Table 12: LDO3, $T_J = -40$ °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
		V _{DD} = V _{SYS}	2.8			
V _{DD}	Input voltage	If supplied from buck	(1.5) Note 1		5.5	V
V _{LDO}	Output voltage		0.9	Note 2	3.44	V
V _{LDO_ACC}	Output accuracy Note 3	$V_{DD} = V_{SYS} = 2.8 V$ to 5.5 V $I_{OUT} = 200$ mA Including static line/load regulation	-3		+3	%
Соит	Stabilization capacitor	Including voltage and temperature coefficient at configured VLDO3	-55%	2.2	+35%	μF
RCOUT_ESR	Output capacitor ESR	f > 1 MHz including wiring parasitics	0		300	mΩ
IOUT_MAX	Output current	V _{DD} = V _{SYS} = 2.8 V to 5.5 V	200			mA



Parameter	Description	Conditions	Min	Тур	Max	Unit
ISHORT	Short circuit current			400		mA
I _{SLEEP}	Maximum forced sleep mode current	V _{DD} ≥ 1.8 V	20			mA
VDROPOUT	Dropout voltage	V _{DD} = V _{SYS} > 2.8 V I _{OUT} = 100 mA (V _{DD} = 1.5 V, I _{OUT} = I _{MAX} /3)		100	150	mV
V _{S_LINE}	Static line regulation	$V_{DD} = V_{SYS} = 3.0 \text{ V to } 5.5 \text{ V}$ $I_{OUT} = 200 \text{ mA}$		1	5	mV
Vs_LOAD	Static load regulation	I _{OUT} = 1 mA to 200 mA		5	10	mV
V _{TR_} LINE	Line transient response	$V_{DD} = V_{SYS} = 3.0 \text{ V to } 3.6 \text{ V}$ $I_{OUT} = 100 \text{ mA}$ $tr = tf = 10 \mu\text{s}$		5	10	mV
V_{TR_LOAD}	Load transient response	$V_{DD} = 3.6 \text{ V}$ $I_{OUT} = 1 \text{ mA to } 200 \text{ mA}$ $tr = tf = 1 \mu\text{s}$		30	50	mV
PSRR Note 4	PSRR	$f = 10 \text{ Hz to } 10 \text{ kHz, RT}$ $V_{DD} = 3.6 \text{ V, } I_{OUT} = I_{MAX}/2$ $V_{DD} - V_{LDO} \ge 0.6 \text{ V}$	50	60		dB
N	Output noise	V _{DD} = V _{SYS} = 3.6 V, V _{LDO} = 2.8 V I _{OUT} = 5 mA to I _{MAX} f = 10 Hz to 100 kHz RT		70		μV rms
I _{Q_ON}	Quiescent current in ON mode	Note 5		9 + 0.45% of I _{OUT}		μА
IQ_SLEEP	Quiescent current forced sleep mode			1.5 + 1.0% of І _{оит}		μА
I _{Q_OFF}	Quiescent current OFF mode			1		μA
ton	Turn-on time	10 to 90%			300	μs
		SLEEP mode			390	
toff	Turn off time	90 % to 10 % Pull-down resistor enabled			1	ms
R _{OFF}	Pull-down resistance in OFF mode	Can be disabled via LDO3_PD_DIS		100		Ω
Bypass Mod	le					
Ron	Bypass on-resistance	V _{DD} > 2.2 V		0.5	0.7	0
TXUN	Dypass on lesistance	V _{DD} > 1.8 V			1.0	Ω



Parameter	Description	Conditions	Min	Тур	Max	Unit
I _{LIM}	Current limit in Bypass mode			380		mA
IQ_BYPASS	Quiescent current in Bypass mode			50	100	μΑ

- Note 1 Max output current is 30 % when the input voltage is 1.5 V.
- Note 2 Programmable in 20 mV voltage steps.
- **Note 3** Accuracy values not applicable if bypass mode has been configured.
- Note 4 Measured at point of load.
- Note 5 Internal regulator current flowing to ground.



5.7.2 LDO7

Table 13: LDO7, $T_J = -40$ °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
		V _{DD} = V _{SYS}	2.8			
V _{DD}	Input voltage	If supplied from buck	1.5 Note 1		5.5	V
V _{LDO}	Output voltage		0.9	Note 2	3.6	V
	_	V _{DD} = V _{SYS} = 2.8 V to 5.5 V				
V _{LDO_ACC}	Output accuracy Note 3	I _{OUT} = 200 mA	-3		+3	%
NO	Note 5	Including static line/load regulation				
Соит	Stabilization capacitor	Including voltage and temperature coefficient at configured VLDO7	-55 %	2.2	+35 %	μF
R _{COUT_ESR}	ESR of capacitor	f > 1 MHz Including track impedance	0		300	mΩ
IOUT_MAX	Maximum output current	$V_{DD} = V_{SYS} = 2.8 \text{ V to}$ 5.5 V	200			mA
Ishort	Short circuit current			400		mA
ISLEEP	Maximum forced sleep mode current	V _{DD} ≥ 1.8 V	30			mA
Vdropout	Dropout voltage	$V_{DD} = V_{SYS} > 2.8 \text{ V}$ $I_{OUT} = 200 \text{ mA}$ $(V_{DD} = 1.5 \text{ V}, I_{OUT} = I_{MAX}/3)$		100	150	mV
Vs_LINE	Static line regulation	V _{DD} = V _{SYS} = 3.0V to 5.5 V I _{OUT} = 200 mA		1	5	mV
V _{S_LOAD}	Static load regulation	I _{OUT} = 1 mA to 200 mA		5	10	mV
VTR_LINE	Line transient response	$V_{DD} = V_{SYS} = 3.0 \text{ V to}$ 3.6 V $I_{OUT} = 100 \text{ mA}$ $tr = tf = 10 \mu\text{s}$		5	10	mV
V_{TR_LOAD}	Load transient response	$V_{DD} = 3.6 \text{ V}$ $I_{OUT} = 1 \text{ mA to } 200 \text{ mA}$ $tr = tf = 1 \mu s$		30	50	mV
PSRR Note 4	PSRR	$f = 10 \text{ Hz to } 10 \text{ kHz, RT}$ $V_{DD} = 3.6 \text{ V, lout} =$ $I_{MAX}/2$ $V_{DD} - V_{LDO} \ge 0.6 \text{ V}$	50	60		dB



Parameter	Description	Conditions	Min	Тур	Max	Unit
N	Output noise	$\begin{split} V_{DD} &= V_{sys} = 3.6 \text{ V}, \\ V_{LDO} &= 2.8 \text{ V} \\ I_{OUT} &= 5 \text{ mA to I}_{MAX} \\ f &= 10 \text{ Hz to } 100 \text{ kHz}, \\ RT \end{split}$		70		μV rms
I _{Q_ON}	Quiescent current in ON mode	Note 5		9+0.4% of I _{ОUТ}		μA
I _{Q_SLEEP}	Quiescent current forced sleep mode			1.5+1.0% of I _{OUT}		μA
IQ_OFF	Quiescent current OFF mode			1		μA
4	Turn-on time	10 % to 90 %			300	
ton		SLEEP mode			390	μs
toff	Turn off time	90 % to 10 % Pull-down resistor enabled			1	ms
R _{OFF}	Pull-down resistance in OFF mode	Can be disabled via LDO7_PD_DIS		100		Ω
Bypass Mo	de					
Ron	Pynasa an registance	V _{DD} > 2.2 V		0.5	0.7	Ω
NON	Bypass on-resistance	V _{DD} > 1.8 V			1.0	12
Ішм	Current limit in Bypass mode			280		mA
I _{Q_BYPASS}	Quiescent current			50	100	μA

- Note 1 Max output current is 30 % when the input voltage is 1.5 V.
- Note 2 Programmable in 50 mV voltage steps.
- Note 3 Accuracy values not applicable if bypass mode has been configured.
- Note 4 Measured at point of load.
- Note 5 Internal regulator current flowing to ground.



5.7.3 LDO8

Table 14: LDO8, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
		V _{DD} = V _{SYS}	2.8			
V _{DD}	Input voltage	If supplied from buck	1.5 Note 1		5.5	V
V _{LDO}	Output voltage		0.9	Note 2	3.6	V
V	Output accuracy	V _{DD} = V _{SYS} = 2.8 V to 5.5 V	2		. 2	0/
V _{LDO_ACC}	Note 3	I _{OUT} = 200 mA Including static line/load regulation	-3		+3	%
Соит	Stabilization capacitor	Including voltage and temperature coefficient at configured VLDO8	-55 %	2.2	+35 %	μF
R _{COUT_ESR}	ESR of capacitor	f > 1 MHz Including track impedance	0		300	mΩ
Iout_max	Maximum output current	$V_{DD} = V_{SYS} = 2.8 \text{ V to}$ 5.5 V	200			mA
ISHORT	Short circuit current			400		mA
ISLEEP	Maximum forced sleep mode current	V _{DD} ≥ 1.8 V	30			mA
Vdropout	Dropout voltage	$V_{DD} = V_{SYS} > 2.8 \text{ V}$ $I_{OUT} = 200 \text{ mA}$ $(V_{DD} = 1.5 \text{ V}, I_{OUT} = I_{MAX}/3)$		100	150	mV
Vs_line	Static line regulation	V _{DD} = V _{SYS} = 3.0 V to 5.5 V I _{OUT} = 200 mA		1	5	mV
V _{S_LOAD}	Static load regulation	I _{OUT} = 1 mA to 200 mA		5	10	mV
VTR_LINE	Line transient response	$V_{DD} = V_{SYS} = 3.0 \text{ V to}$ 3.6 V $I_{OUT} = 100 \text{ mA}$ $tr = tf = 10 \mu\text{s}$		5	10	mV
V _{TR_LOAD}	Load transient response	$V_{DD} = 3.6 \text{ V}$ $I_{OUT} = 1 \text{ mA to } 200 \text{ mA}$ $tr = tf = 1 \mu\text{s}$		30	50	mV
PSRR Note 4	PSRR	$f = 10 \text{ Hz to } 10 \text{ kHz, RT}$ $V_{DD} = 3.6 \text{ V, lout} =$ $I_{MAX}/2$ $V_{DD} - V_{LDO} \ge 0.6 \text{ V}$	50	60		dB



Parameter	Description	Conditions	Min	Тур	Max	Unit
N	Output noise	V _{DD} = V _{SYS} = 3.6 V, V _{LDO} = 2.8 V I _{OUT} = 5 mA to I _{MAX} f = 10 Hz to 100 kHz, RT		70		μV rms
IQ_ON	Quiescent current in ON mode	Note 5		9 + 0.4 % of I _{OUT}		μA
IQ_SLEEP	Quiescent current forced sleep mode			1.5 + 1.0 % of I _{OUT}		μA
IQ_OFF	Quiescent current OFF mode			1		μA
ton	Towns on the s	10 % to 90 %			300	
	Turn-on time	SLEEP mode			390	μs
toff	Turn off time	90 % to 10 % Pull-down resistor enabled			1	ms
R _{OFF}	Pull-down resistance in OFF mode	Can be disabled via LDO8_PD_DIS		100		Ω
Vibration M	otor Driver Mode					
VIB_SET	Output voltage (average)	6-bit resolution	0		3	V
I _{MAX}	Maximum output current				300	mA
ISHORT	Short circuit current			400		mA
R _{LOAD}	Load resistance		8	10	10000	Ω
RLOAD	Load impedance			200		μH
Ron	Pull-up resistor			0.5		Ω
R _{OFF}	Pull-down resistor			5		Ω
Bypass Mo	de					
D	Dunasa an registance	V _{DD} > 2.2 V		0.5	0.7	
Ron	Bypass on-resistance	V _{DD} > 1.8 V			1.0	Ω
ILIM	Current limit in Bypass mode			400		mA
IQ_BYPASS	Quiescent current in Bypass mode			50	100	μA

- Note 1 Max output current is 30 % when the input voltage is 1.5 V.
- Note 2 Programmable in 50 mV voltage steps.
- Note 3 Accuracy values not applicable if bypass mode has been configured.
- Note 4 Measured at point of load.
- Note 5 Internal regulator current flowing to ground.



5.7.4 LDO9

Table 15: LDO9, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
		V _{DD} = V _{SYS}	2.8			
V_{DD}	Input voltage	If supplied from buck	1.5 Note 1		5.5	V
V _{LDO}	Output voltage		0.95	Note 2	3.6	V
VLDO_ACC	Output accuracy	V _{DD} = V _{SYS} = 2.8 V to 5.5 V I _{OUT} = 200 mA Including static line/load regulation Note 3	-3		+3	%
Соит	Stabilization capacitor	Including voltage and temperature coefficient at configured VLDO9	-55 %	2.2	+35 %	μF
R _{COUT_ESR}	ESR of capacitor	f > 1 MHz	0		300	mΩ
I _{OUT_MAX}	Maximum output current	V _{DD} = V _{SYS} = 2.8 V to 5.5 V	200			mA
Ishort	Short circuit current			400		mA
I _{SLEEP}	Maximum forced sleep mode current	V _{DD} ≥ 1.8 V	30			mA
VDROPOUT	Dropout voltage	$V_{DD} = V_{SYS} > 2.8 \text{ V}$ $I_{OUT} = 200 \text{ mA}$ $(V_{DD} = 1.5 \text{ V, } I_{OUT} = I_{MAX}/3)$		100	150	mV
Vs_LINE	Static line regulation	V _{DD} = V _{SYS} = 3.0 V to 5.5 V I _{OUT} = 200 mA		1	5	mV
V _{S_LOAD}	Static load regulation	I _{OUT} = 1 mA to 200 mA		5	10	mV
VTR_LINE	Line transient response	$V_{DD} = V_{SYS} = 3.0 \text{ V to}$ 3.6 V $I_{OUT} = 100 \text{ mA}$ $I_{t} = 10 \text{ ms}$		5	10	mV
Vtr_load	Load transient response	$V_{DD} = 3.6 \text{ V}$ $I_{OUT} = 1 \text{ mA to } 200 \text{ mA}$ $tr = tf = 1 \mu\text{s}$		30	50	mV
PSRR Note 4	PSRR	$f = 10 \text{ Hz to } 10 \text{ kHz, RT}$ $V_{DD} = 3.6 \text{ V, } I_{OUT} = I_{MAX}/2$ $V_{DD} - V_{LDO} \ge 0.6 \text{ V}$	60	70		dB



Parameter	Description	Conditions	Min	Тур	Max	Unit
N	Output noise	$V_{DD} = V_{sys} = 3.6 \text{ V}, \\ V_{LDO} = 2.8 \text{ V} \\ I_{OUT} = 5 \text{ mA to } I_{MAX} \\ f = 10 \text{ Hz to } 100 \text{ kHz, RT}$		35		μV rms
IQ_ON	Quiescent current in ON mode	Note 5		9+0.4% of I _{OUT}		μΑ
I _{Q_SLEEP}	Quiescent current forced sleep mode			2+1.0% of I _{ОUТ}		μΑ
IQ_OFF	Quiescent current OFF mode			1		μA
_	Turns on time o	10 % to 90 %			200	
ton	Turn-on time	SLEEP mode			260	μs
toff	Turn off time	90 % to 10 % Pull-down resistor enabled			1	ms
Roff	Pull-down resistance in OFF mode	Can be disabled via LDO9_PD_DIS		100		Ω

- Note 1 Max output current is 30 % when the input voltage is 1.5 V.
- Note 2 Programmable in 50 mV voltage steps.
- Note 3 At trimmed output voltage.
- Note 4 Measured at point of load.
- Note 5 Internal regulator current flowing to ground.



5.7.5 LDO11

Table 16: LDO11, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
		V _{DD} = V _{SYS}	2.8			
V_{DD}	Input voltage	If supplied from buck	1.5 Note 1		5.5	V
V _{LDO}	Output voltage		0.9	Note 2	3.6	V
VLDO_ACC	Output accuracy Note 3	V _{DD} = V _{SYS} = 2.8V to 5.5 V I _{OUT} = 200 mA Including static line/load regulation	-3		+3	%
Соит	Stabilization capacitor	Including voltage and temperature coefficient at configured VLDO11	-55 %	2.2	+35 %	μF
R _{COUT_ESR}	ESR of capacitor	f > 1 MHz Including track impedance	0		300	mΩ
IOUT_MAX	Maximum output current	V _{DD} = V _{SYS} = 2.8 V to 5.5 V	300			mA
Ishort	Short circuit current			600		mA
ISLEEP	Maximum sleep mode current	V _{DD} ≥ 1.8 V	30			mA
VDROPOUT	Dropout voltage	$V_{DD} = V_{SYS} > 2.8 \text{ V}$ $I_{OUT} = 300 \text{ mA}$ $(V_{DD} = 1.5 \text{ V}, I_{OUT} = I_{MAX}/3)$		100	150	mV
Vs_LINE	Static line regulation	V _{DD} = V _{SYS} = 2.8 V to 5.5 V V _{LDO} = 1.86 V I _{OUT} = 300 mA		2	15	mV
V _{S_LOAD}	Static load regulation	I _{OUT} = 1 mA to 300 mA		5	20	mV
VTR_LINE	Line transient response	$V_{DD} = V_{SYS} = 2.8 \text{ V to}$ 5.5 V $V_{LDO} = 1.86 \text{ V}$ $I_{OUT} = 300 \text{ mA}$ $tr = tf = 10 \mu\text{s}$		5	10	mV
V _{TR_LOAD}	Load transient response	$V_{DD} = 3.6 \text{ V}$ $I_{OUT} = 1 \text{ mA to } 300 \text{ mA}$ $tr = tf = 1 \mu\text{s}$		30	50	mV
PSRR Note 4	PSRR	$ f = 10 \text{ Hz to } 10 \text{ kHz, RT} \\ V_{DD} = 3.6 \text{ V, } I_{OUT} = I_{MAX}/2 \\ V_{DD} - V_{LDO} \ge 0.6 \text{ V} $	60	70		dB



Parameter	Description	Conditions	Min	Тур	Max	Unit
N	Output noise	$\begin{split} V_{DD} &= V_{SYS} = 3.6 \text{ V}, \\ V_{LDO} &= 2.8 \text{ V} \\ I_{OUT} &= 5 \text{ mA to } I_{MAX} \\ f &= 10 \text{ Hz to } 100 \text{ kHz, RT} \end{split}$		70		μV rms
I _{Q_ON}	Quiescent current in ON mode	Note 5		9 + 0.45 % of l _{OUT}		μA
I _{Q_SLEEP}	Quiescent current forced sleep mode			2 + 0.7 % of lout		μA
IQ_OFF	Quiescent current OFF mode			1		μA
	Turn-on time	10 % to 90 %			200	
ton		SLEEP mode			260	μs
toff	Turn off time	90 % to 10 % Pull-down resistor enabled			1	ms
Roff	Pull-down resistance in OFF mode	Can be disabled via LDO11_PD_DIS		100		Ω
Bypass Mod	e					
Davi	Pungas on registance	V _{DD} > 2.2 V		0.3	0.7	Ω
Ron	Bypass on-resistance	V _{DD} > 1.8 V			1.0	12
Ішм	Current limit in Bypass mode			270		mA
I _{Q_BYPASS}	Quiescent current in Bypass mode			50	100	μA

- Note 1 Max output current is 30 % when the input voltage is 1.5 V.
- Note 2 Programmable in 50 mV voltage steps.
- Note 3 Accuracy values not applicable if bypass mode has been configured.
- Note 4 Measured at point of load.
- Note 5 Internal regulator current flowing to ground.



5.7.6 LDOCORE

Table 17: LDOCORE, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
VDDCORE	Output voltage	Note 1	2.45	2.5	2.55	V
		RESET mode		2.2		V
Соит	Stabilization capacitor	Including voltage and temperature coefficient	- 55 %	2.2	+35 %	μF
RCOUT_ESR	Output capacitor ESR	f > 1 MHz Including wiring parasitics	0		300	mΩ
VDROPOUT	Dropout voltage	Note 2		50	100	mV

Note 1 Setting VDD_FAULT_LOWER ≥ 2.65 V avoids LDOCORE dropout, see Section 5.11.

Note 2 The LDOCORE supply, VSYS or CHG_WAKE, must be maintained above VDDCORE + VDROPOUT.

Note
LDOCORE is only used to supply internal circuits.



5.8 DC/DC Buck Converters

5.8.1 BUCKCORE1 and BUCKCORE2

Table 18: BUCKCORE1 and BUCKCORE2, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
V _{DD}	Input voltage	V _{DD} = V _{SYS}	2.8		5.5	V
		Including voltage and temperature coefficient			Note 1	
Соит	Output capacitor	Half-current mode	-50 %	2 x 22	+30 %	μF
		Full-current mode	-50 %	2 x 47	+30 %	
	Output capacitor ESR	Including wiring parasitics f > 100 kHz				
RCOUT_ESR		Half-current mode Coυτ = 2 * 22 μF		15	50	mΩ
		Full-current mode Cout = 2 * 47 µF		7.5	25	
L _{виск}	Inductor value	Including current and temperature dependence	0.7	1.0	1.3	μΗ
LESR	Inductor resistance			55	100	mΩ
PWM Mode						
Vвиск	Output voltage	Programmable in 10 mV steps Note 2	0.3		1.57	>
		Excluding static line/load regulation and voltage ripple T _A = 25 °C V _{DD} = 4.2 V V _{BUCK} = 1.03 V	-1		+1	
VBUCK_ACC	Output voltage accuracy	Excluding static line/load regulation and voltage ripple TA = -40 °C to +85 °C VDD = 4.2 V VBUCK = 1.03 V	-1.5		+1.5	%
		Including static line/load regulation and voltage ripple IOUT = IMAX VBUCK = 1.03 V LBUCK, LESR = Typ	-2		+2	



Parameter	Description	Conditions	Min	Тур	Max	Unit
		Including static line/load regulation and voltage ripple lout = IMAX Note 3	-3		+3	
V _{TR_LD}	Load regulation transient	$V_{DD} = 3.6 \text{ V}$ $V_{BUCK} = 1.2 \text{ V}$ $I_{OUT} = 200 \text{ mA} / 0.8 * I_{MAX}$ $dI/dt = 3 \text{ A/}\mu\text{s}$ $L_{BUCK} = 1.0 \mu\text{H}$ Note 4	-4		+4	%
VTR_LINE	Line regulation transient	$V_{DD} = 3.0 \text{ V to } 3.6 \text{ V}$ $I_{OUT} = 500 \text{ mA}$ $tr = tf = 10 \mu\text{s}$		0.2	3	mV
	Parasitic track resistance	From output capacitor to sense connection at point-of-load		10		mΩ
	Parasitic track inductance	From output capacitor to sense connection at point-of-load		5		nH
Z _{FB}	Feedback Comparator input impedance		500			kΩ
I _{MAX}	Output current	Half-current mode	1250			- mA
		Full-current mode	2500			
I _{LIM} Note 5	Current limit (programmable)	BCORE1_ILIM=0000	-20 %	500	+20 %	mA
		BCORE1_ILIM=1111	-20 %	2000	+20 %	mA
IQ_OFF	Quiescent current in OFF mode				1	μA
I _{Q_ON}	Quiescent current in PWM mode	Half-current mode I _{OUT} = 0 mA		9.0		mA
		Full-current mode		11.0		
f	Switching frequency Note 6	OSC_FRQ = 0000	2.85	3	3.15	MHz
D	Switching duty cycle		10.5		84	%
ton	Turn on time	VBUCK = 1.15 V BUCK_SLOWSTART = disabled SLEW_RATE = 10 mV/1 µs BUCK <x>_ILIM = 1500</x>		0.37	1.2	ms



Parameter	Description	Conditions	Min	Тур	Max	Unit
R _{PD}	Output pull-down resistor	V _{BUCK} = 0.5 V Can be disabled via BCORE1_PD_DIS		80	200	Ω
D	DMOS ON registance	Half-current mode Including pin and routing Vsys = 3.6 V		160		
R _{PMOS}	PMOS ON resistance	Full-current mode Including pin and routing Vsys = 3.6 V		80		mΩ
	NIMOS ON registeres	Half-current mode Including pin and routing V _{SYS} = 3.6 V		60		mΩ
R _{NMOS}	NMOS ON resistance	Full-current mode Including pin and routing Vsys = 3.6 V		30		
η	Efficiency Note 7	$V_{DD} = 3.6 \text{ V}$ $V_{BUCK} = 1.2 \text{ V}$ $I_{OUT} = 0.1 \text{ to } 0.7 \text{ * } I_{MAX}$		84		%
PFM Mode						
Vвиск	Output voltage	Programmable in 10 mV steps	0.3		1.57	V
I _{AUTO_THR}	Typical automatic mode switching current			260		mA
lout_pfm	Output current			300		mA
I _{LIM_PFM}	Current limit			600		mA
		I _{OUT} = 0				
I _{Q_PFM}	Quiescent current	Forced PFM mode		27	32	μA
		AUTO mode		35	42	
f	Frequency of operation		0		3	MHz
η	Efficiency Note 7	V _{DD} = 3.6 V V _{BUCK} = 1.2 V I _{OUT} = 10 mA		86		%

- Note 1 The purpose of the specified maximum Cout value is to ensure that the specified turn-on time, ton, can be met. Additional output rail capacitors can be added to reduce voltage ripple or to improve transient load regulation. However, this will compromise ton performance, create larger in-rush currents and will affect DVC slew rates.
- Note 2 If BUCK<x>_MODE = 10 (synchronous) then the buck operates in PFM mode when $V_{BUCK} < 0.7 \text{ V}$. For complete control of the buck mode (PWM versus PFM) use BUCK<x>_MODE = 00.
- Note 3 Minimum tolerance 35 mV.
- Note 4 Measured at Cout, depends on parasitics of PCB and external components when remote sensing.



- Note 5 Current limit values are doubled in full-current mode.
- Note 6 Generated from internal 6 MHz oscillator and can be adjusted by ± 10 % via control OSC_FRQ.
- Note 7 Depends on external components and PCB routing.



5.8.2 BUCKPRO

Table 19: BUCKPRO, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
V _{DD}	Input voltage	V _{DD} = V _{SYS}	2.8		5.5	V
Соит	Output capacitor	Including voltage and temperature coefficient Half-current mode		2 * 22	Note 1	μF
			-50 %	2 * 47	+30 %	-
		Full-current mode Including wiring parasitics f > 100 kHz	,~	2 47	,~	
RCOUT_ESR	Output capacitor ESR	Half-current mode Coυτ = 2 * 22 μF		15	50	mΩ
		Full-current mode Cout = 2 * 47 µF		7.5	25	
Lвиск	Inductor value	Including current and temperature dependence	0.7	1.0	1.3	μH
LESR	Inductor resistance			55	100	mΩ
PWM Mode						
Vвиск	Output voltage	Programmable in 10 mV steps Note 2	0.53		1.8	V
		Including static line/load regulation and voltage ripple lout = IMAX Note 3	-3		+3	
RCOUT_ESR	Output voltage accuracy	Excluding static line/load regulation and voltage ripple TA = 25 °C VBUCK > 1 V VDD = 5 V	-1		+1	%
V _{TR_LD}	Transient load regulation	$V_{DD} = 3.6 \text{ V}$ $V_{BUCK} = 1.2 \text{ V}$ $I_{OUT} = 200 \text{ mA / } 0.8 \text{ *}$ I_{MAX} $dI/dt = 3 \text{ A/}\mu\text{s}$ $L_{BUCK} = 1.0 \mu\text{H}$ Note 4		20	50	mV



Parameter	Description	Conditions	Min	Тур	Max	Unit
V_{TR_LINE}	Transient line regulation	$V_{DD} = 3.0 \text{ V to } 3.6 \text{ V}$ $I_{OUT} = 500 \text{ mA}$ $tr = tf = 10 \mu\text{s}$		0.2	3	mV
Luny	Output current	Half-current mode	1250			mA
I _{MAX}	Output current	Full-current mode	2500			IIIA
I _{LIM}	Current limit (programmable)	BPRO_ILIM=0000	-20 %	500	+20 %	mA
ILIM	Current limit (programmable)	BPRO_ILIM=1111	-20 %	2000	+20 %	mA
lq_off	Quiescent current in OFF mode				1	μΑ
I _{Q_ON}	Quiescent current in PWM	Half-current mode		9.0		mA
IQ_ON	mode	Full-current mode		11.0		ША
f	Switching frequency Note 5	OSC_FRQ = 0000	2.85	3	3.15	MHz
D	Switching duty cycle		10.6		84	%
R _{PD}	Output pull-down resistor	@ Vout = 0.5 V Can be disabled via BPRO_PD_DIS		80	200	Ω
		Half-current mode Including pin and routing V _{SYS} = 3.6 V		160		mΩ
R _{PMOS}	PMOS ON resistance	Full-current mode Including pin and routing Vsys = 3.6 V		80		
		Half-current mode Including pin and routing Vsys = 3.6 V		60		mΩ
R _{NMOS}	NMOS ON resistance	Full-current mode Including pin and routing Vsys = 3.6 V		30		
η	Efficiency Note 6	V _{DD} = 3.6 V V _{BUCK} = 1.2 V I _{OUT} = 0.1 mA to 0.7 * I _{MAX}		84		%
PFM Mode						



Parameter	Description	Conditions	Min	Тур	Max	Unit
V _{BUCK}	Output voltage	Programmable in 10 mV steps	0.53		1.8	V
I _{AUTO_THR}	Typical automatic mode switching current			260		mA
IOUT_PFM	Output current			300		mA
I _{LIM}	Current limit			600		mA
		I _{OUT} = 0 mA				
I _{Q_PFM}	Quiescent current	Forced PFM mode		22	25	μA
		AUTO mode		30	35	
f	Frequency of operation		0		3	MHz
η	Efficiency Note 6	V _{DD} = 3.6 V V _{BUCK} = 1.2 V I _{OUT} = 10 mA		86		%
VTT Mode						
V _{DD}	Input voltage	V _{DD} = V _{SYS}	2.8		5.5	V
Соит	Output capacitor	Including voltage and temperature coefficient	-50 %	2 * 47	+30 %	μF
RCOUT_ESR	Output capacitor ESR	ESR of C _{OUT} @ f > 100 kHz + track impedance		7.5	25	mΩ
Lвиск	Inductor value	Including current and temperature dependence		0.25		μH
L _{ESR}	Inductor resistance			55	100	mΩ
V _{BUCK}	Output voltage	VBUCK = VDDQ/2	0.675		1.3	V
VBUCK_ACC	Output voltage accuracy	Relative to VTTR Including static line and load regulation	- 3		4	%
V _{TR_LD}	Transient load regulation	Half-current mode $V_{DD}=3.6\ V$ $V_{BUCK}=0.7\ V$ $I_{OUT}=10\ mA\ to\ 1\ A$ $I_{OUT}=-750\ mA\ to\ -10$ mA $L_{BUCK}=0.24\ \mu H$ $dI/dt=3\ A/\mu s$		20	40	mV
		Full-current mode $V_{DD} = 3.6 \text{ V}$ $V_{BUCK} = 0.75 \text{ V}$ $I_{OUT} = 10 \text{ mA to } 1.4 \text{ A}$ $I_{OUT} = -10 \text{ mA to } -1.4 \text{ A}$		20	40	



Parameter	Description	Conditions	Min	Тур	Max	Unit
		L _{BUCK} = 0.24 μH dl/dt = 3 A/μs				
		Full-current mode $V_{DD} = 3.6 \text{ V}$ $V_{BUCK} = 0.7 \text{ V}$ $I_{OUT} = 10 \text{ mA to } 1.1 \text{ A}$ $I_{OUT} = -10 \text{ mA to } -1.1 \text{ A}$ $L_{BUCK} = 0.24 \mu\text{H}$ $dI/dt = 3 \text{ A}/\mu\text{s}$		20	40	
		Full-current mode $V_{DD} = 3.6 \text{ V}$ $V_{BUCK} = 0.675 \text{ V}$ $I_{OUT} = 10 \text{ mA to } 900 \text{ mA}$ $I_{OUT} = -10 \text{ mA to } -900 \text{ mA}$ $L_{BUCK} = 0.24 \mu\text{H}$ $dI/dt = 3 \text{ A}/\mu\text{s}$		20	40	
		Half-current mode Vвиск = 0.7 V	-550		1250	mA
		Full-current mode V _{BUCK} = 0.75	- 1400		2500	
IMAX	Maximum output current	Full-current mode VBUCK = 0.7 V	- 1100		2500	mA
		Full-current mode VBUCK = 0.675 V	-900		2500	
ton	Turn-on time	VBUCK = 0.75 V BUCK_SLOWSTART = disabled SLEW_RATE = 10 mV/1 µs BUCK4_ILIM = 1500 mA		0.33	1.2	ms
VTTR Buffer						
V _{DDQ}	Feedback voltage	V _{DD} = V _{SYS}	1.35		2.6	V
V_{VTTR}	Output voltage	Iout = 0 mA to Ivttr	0.675	VDDQ/2	1.3	V
Vvttr_acc	Voltage accuracy	V _{VTTR_ACC} related to VDDQ input voltage	-1	VDDQ/2	+1	%
Соит	Output capacitor	Including voltage and temperature coefficient	-50 %	0.1	+30 %	μF



Parameter	Description	Conditions	Min	Тур	Max	Unit
Іоит	Sink/source current		-10		10	mA

- Note 1 The purpose of the specified maximum C_{OUT} value is to ensure that the specified turn-on time, t_{ON}, can be met. Additional output rail capacitors can be added to reduce voltage ripple or to improve transient load regulation. However, this will compromise t_{ON} performance, create larger in-rush currents and will affect DVC slew rates.
- Note 2 If BUCK<x>_MODE = 10 (synchronous) then the buck operates in PFM mode when $V_{BUCK} < 0.7 \text{ V}$. For complete control of the buck mode (PWM versus PFM) use BUCK<x>_MODE = 00.
- Note 3 Minimum tolerance 35 mV.
- Note 4 Measured at C_{OUT}, depends on parasitics of PCB and external components when remote sensing.
- Note 5 Generated from internal 6 MHz oscillator and can be adjusted by ± 10 % via control OSC_FRQ.
- Note 6 Depends on external components and PCB routing.



5.8.3 BUCKMEM

Table 20: BUCKMEM, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
V_{DD}	Input voltage	V _{DD} = V _{SYS}	2.8		5.5	V
Соит	Output capacitor	Including voltage and temperature coefficient	-50	2 * 22	Note 1	μF
		Merged mode	%	2 * 47	+30 %	•
Rcout_esr	Output capacitor ESR	Including wiring parasitics f > 100 kHz				
T COOT_LON	Catput supusitor Lort	C _{OUT} = 2 * 22 µF		15	50	0
		C _{OUT} = 2 * 47 µF		7.5	25	mΩ
L _{BUCK}	Inductor value	Including current and temperature dependence	0.7	1.0	1.3	μH
Lesr	Inductor resistance			55	100	mΩ
Vвиск	Output voltage	Programmable in 20 mV steps	0.8		3.34 Note 7	V
		Including static line/load regulation and voltage ripple IOUT = IMAX Note 2	-3		+3	
VBUCK_ACC	Output voltage accuracy	Excluding static line/load regulation and voltage ripple TA = 25 °C VDD = 5 V VBUCK > 1 V	-2		+2	%
V _{TR_LD}	Transient load regulation	V _{DD} = 3.6 V V _{BUCK} = 1.2 V I _{OUT} = 200 mA / 0.8 * I _{MAX} dI/dt = 3 A/μs Note 3	-4		+4	%
VTR_LINE	Transient line regulation	$V_{DD} = 3.0 \text{ V to } 3.6 \text{ V}$ $I_{OUT} = 500 \text{ mA}$ $tr = tf = 10 \mu\text{s}$		0.2	3	mV
I _{MAX}	Output current		1500			mA
I _{LIM} Note 4	Current limit (programmable)	BMEM_ILIM = 0000	-20 %	1500	+20 %	mA



Parameter	Description	Conditions	Min	Тур	Max	Unit
		BMEM_ILIM = 1111	-20 %	3000	+20 %	mA
I _{Q_OFF}	Quiescent current in OFF mode				1	μΑ
IQ_ON	Quiescent current in PWM mode	I _{OUT} = 0 mA		9		mA
f	Switching frequency Note 5	OSC_FRQ = 0000	2.85	3	3.15	MHz
D	Switching duty cycle		14.5		100	%
	Output pull-down resistor	VBUCK = 0.5 V Disabled via BMEM_PD_DIS		80	200	Ω
R _{PMOS}	PMOS ON resistance	Including pin and routing V _{SYS} = 3.6 V		160		mΩ
R _{NMOS}	NMOS ON resistance	Including pin and routing Vsys = 3.6 V		60		mΩ
η	Efficiency Note 6	V _{DD} = 3.6 V V _{BUCK} = 1.2 V I _{OUT} = 0.1 mA to 0.7 * I _{MAX}		83		%
PFM Mode						
Vвиск	Output voltage	Programmable in 20 mV steps	0.8		3.34 Note 7	V
	Typical automatic mode switching current			260		mA
I _{MAX}	Output current			300		mA
I _{LIM}	Current limit			600		mA
		I _{OUT} = 0 mA				
I _{Q_PFM}	Quiescent current	Forced PFM mode		22	25	μΑ
		AUTO mode		30	35	μ/ (
f	Frequency of operation		0		3	MHz
η	Efficiency Note 6	V _{DD} = 3.6 V V _{BUCK} = 1.2 V I _{OUT} = 10 mA		86		%

- Note 1 The purpose of the specified maximum C_{OUT} value is to ensure that the specified turn-on time, t_{ON} , can be met. Additional output rail capacitors can be added to reduce voltage ripple or to improve transient load regulation. However, this will compromise t_{ON} performance, create larger in-rush currents and will affect DVC slew rates.
- Note 2 Minimum tolerance 35 mV.
- Note 3 Measured at Cout, depends on parasitics of PCB and external components when remote sensing.



- Note 4 The current limits are automatically doubled when BUCKMEM is merged with BUCKIO.
- Note 5 Generated from internal 6 MHz oscillator and can be adjusted by ± 10 % via control OSC_FRQ.
- Note 6 Depends on external components and PCB routing.
- Note 7 Maximum V_{DD} 0.7 V.



5.8.4 BUCKIO

Table 21: BUCKIO, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
V_{DD}	Input voltage	V _{DD} = V _{SYS}	2.8		5.5	V
Соит	Output capacitor	Including voltage and temperature coefficient	-50%	2x22	Note 1 +30%	μF
RCOUT_ESR	Output capacitor ESR	f > 100 kHz All caps + track impedance		15	50	mΩ
L _{виск}	Inductor value	Including current and temperature dependence	0.7	1.0	1.3	μН
L _{ESR}	Inductor resistance			55	100	mΩ
PWM Mode						
Vвиск	Output voltage	Programmable in 20 mV steps Note 2	0.8		3.34 Note 3	٧
		Including static line/load regulation and voltage ripple @ lout = lmax	-3	Note 4	+3	
VBUCK_ACC	Output voltage accuracy	$T_{A} = 25 \text{ °C}$ $I_{OUT} = 0$ $V_{OUT} > 1 \text{ V}$ $V_{DD} = 5 \text{ V}$	-2		+2	%
Vtr_ld	Load regulation transient	I _{OUT} = 200 mA/0.8 * I _{MAX} dI/dt = 3 A/µs	-4	Note 5	+4	%
V _{TR_LINE}	Line regulation transient	$V_{DD} = 3.0 \text{ V to } 3.6 \text{ V}$ $I_{OUT} = 500 \text{ mA}$ $tr = tf = 10 \mu\text{s}$		0.2	3	mV
I _{MAX}	Output current		1500			mA
Lan	Current limit	BIO_ILIM=0000	-20 %	1500	20 %	mA
I _{LIM}	(programmable)	BIO_ILIM=1111	-20 %	3000	20 %	mA
I _{Q_OFF}	Quiescent current in OFF mode				1	μΑ
IQ_ON	Quiescent current in PWM mode			9		mA
f	Switching frequency		2.85	3	3.15	MHz
D	Switching duty cycle		14.5		100	%
	Output pull-down resistor	Vout = 0.5 V Can be disabled via BIO_PD_DIS		80	200	Ω



Parameter	Description	Conditions	Min	Тур	Max	Unit
R _{PMOS}	PMOS ON resistance	Including pin and routing Vsys = 3.6 V		160		mΩ
R _{NMOS}	NMOS ON resistance	Including pin and routing V _{SYS} = 3.6 V		60		mΩ
η	Efficiency Note 6	V _{DD} = 3.6 V V _{BUCK} = 1.8 V I _{OUT} = 0.1 to 0.7 * I _{MAX}		87		%
	PFM Mode					
	Typical automatic mode switching current			260		mA
I _{MAX}	Output current			300		mA
I _{LIM}	Current limit			600		mA
		I _{OUT} = 0 mA				
IQ_PFM	Quiescent current	Forced PFM mode		22	25	μA
		AUTO mode		30	35	
f	Frequency of operation		0		3	MHz
η	Efficiency Note 6	V _{DD} = 3.6 V V _{BUCK} = 1.8 V I _{OUT} = 10 mA		90		%

- Note 1 The purpose of the specified maximum C_{OUT} value is to ensure that the specified turn-on time, t_{ON}, can be met. Additional output rail capacitors can be added to reduce voltage ripple or to improve transient load regulation. However, this will compromise t_{ON} performance, create larger in-rush currents and will affect DVC slew rates.
- Note 2 If BUCK<x>_MODE = 10 (synchronous) then the buck operates in PFM mode when V_{BUCK} < 0.7 V. For complete control of the buck mode (PWM versus PFM) use BUCK<x>_MODE = 00.
- Note 3 Maximum V_{DD} 0.7 V.
- Note 4 Minimum tolerance 35 mV.
- Note 5 Measured at C_{OUT}, depends on parasitics of PCB and external components when remote sensing.
- Note 6 Depends on external components and PCB routing.



5.8.5 BUCKPERI

Table 22: BUCKPERI, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
V _{DD}	Input voltage	V _{DD} = V _{SYS}	2.8		5.5	V
Соит	Output capacitor	Including voltage and temperature coefficient	-50 %	2 x 22	Note 1 +30 %	μF
RCOUT_ESR	Output capacitor ESR	f > 100 kHz All caps + track impedance		15	50	mΩ
L _{BUCK}	Inductor value	Including current and temperature dependence	0.7	1.0	1.3	μH
LESR	Inductor resistance			55	100	mΩ
Vвиск	Output voltage	Programmable in 20 mV steps Note 2	0.8		3.34 Note 3	٧
		Including static line/load regulation and voltage ripple IOUT = IMAX	-3	Note 4	+3	
VBUCK_ACC	Output voltage accuracy	T _A = 25 °C I _{OUT} = 0 V _{OUT} > 1 V V _{DD} = 5 V	-2 +2	%		
V_{TR_LD}	Load regulation transient	I _{OUT} = 200 mA/0.8 * I _{MAX} dI/dt = 3 A/μs	-4	Note 5	+4	%
VTR_LINE	Line regulation transient	$V_{DD} = 3.0 \text{ V to } 3.6 \text{ V}$ $I_{OUT} = 500 \text{ mA}$ $tr = tf = 10 \mu\text{s}$		0.2	3	mV
I _{MAX}	Output current		1500			mA
	Current limit	BPERI_ILIM = 0000	-20%	1500	+20%	mA
ILIM	(programmable)	BPERI_ILIM = 1111	-20%	3000	+20%	mA
I _{Q_OFF}	Quiescent current in OFF mode			1	μΑ	
I _{Q_ON}	Quiescent current in PWM mode			9		mA
f	Switching frequency		2.85	3	3.15	MHz
D	Switching duty cycle		14.5		100	%
	Output pull-down resistor	Vout = 0.5 V Can be disabled via BPERI_PD_DIS		80	200	Ω
R _{PMOS}	PMOS ON resistance	Including pin and routing V _{SYS} = 3.6 V		160		mΩ



Parameter	Description	Conditions	Min	Тур	Max	Unit
R _{NMOS}	NMOS ON resistance	Including pin and routing Vsys = 3.6 V		60		mΩ
η	Efficiency Note 6	$V_{DD} = 3.6 \text{ V}$ $V_{BUCK} = 2.86 \text{ V}$ $I_{OUT} = 0.1 \text{ to } 0.7 \text{ * } I_{MAX}$		91		%
	PFM Mode					
	Typical automatic mode switching current			260		mA
I _{MAX}	Output current			300		mA
ILIM	Current limit			600		mA
		I _{OUT} = 0 mA				
I _{Q_PFM}	Quiescent current	Forced PFM mode		22	25	μΑ
		AUTO mode		30	35	
f	Frequency of operation		0		3	MHz
η	Efficiency Note 6	V _{DD} = 3.6 V V _{BUCK} = 2.86 V I _{OUT} = 10 mA		93		%

- Note 1 The purpose of the specified maximum C_{OUT} value is to ensure that the specified turn-on time, t_{ON}, can be met. Additional output rail capacitors can be added to reduce voltage ripple or to improve transient load regulation. However, this will compromise t_{ON} performance, create larger in-rush currents and will affect DVC slew rates.
- Note 2 If BUCK<x>_MODE = 10 (synchronous) then the buck operates in PFM mode when $V_{BUCK} < 0.7 \text{ V}$. For complete control of the buck mode (PWM versus PFM) use BUCK<x>_MODE = 00.
- Note 3 Maximum V_{DD} 0.7 V.
- Note 4 Minimum tolerance 35 mV.
- Note 5 Measured at Cout, depends on parasitics of PCB and external components when remote sensing.
- Note 6 Depends on external components and PCB routing.



5.8.6 Typical Characteristics

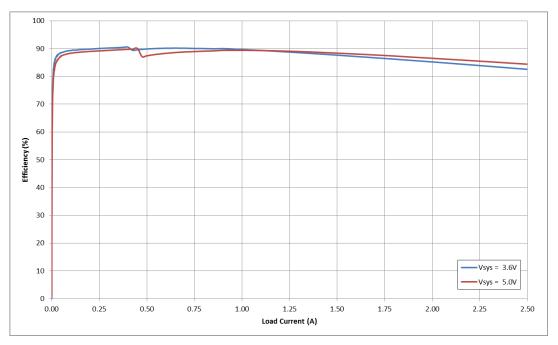


Figure 6: BUCKCORE1 Efficiency in AUTO Mode, Vout = 1.2 V

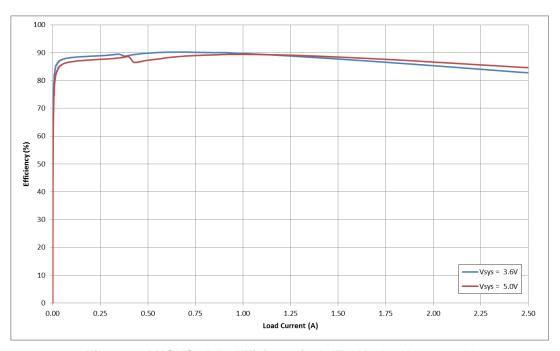


Figure 7: BUCKCORE2 Efficiency in AUTO Mode, V_{OUT} = 1.2 V



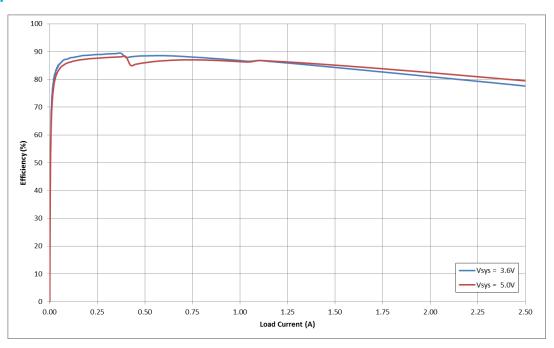


Figure 8: BUCKPRO Efficiency in AUTO Mode, V_{OUT} = 1.2 V

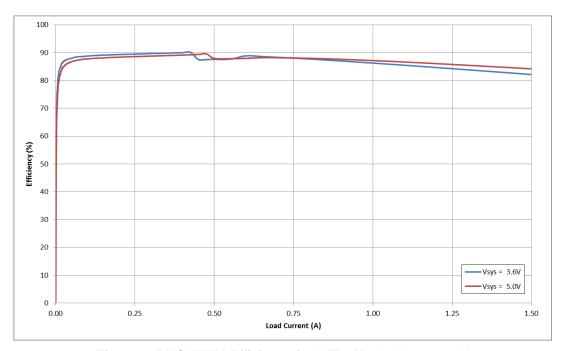


Figure 9: BUCKMEM Efficiency in AUTO Mode, Vout = 1.2 V



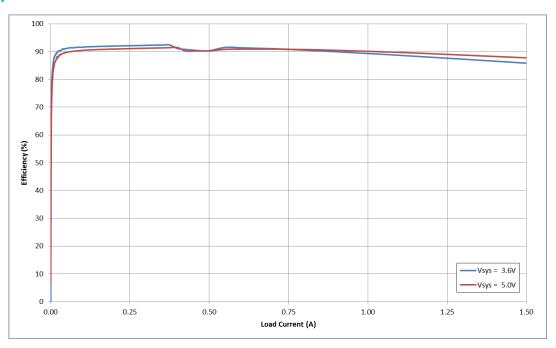


Figure 10: BUCKIO Efficiency in AUTO Mode, Vout = 1.8 V

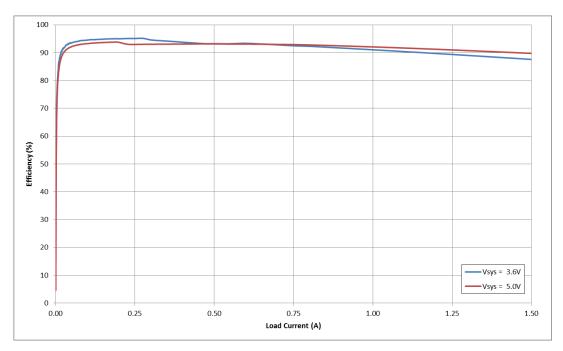


Figure 11: BUCKPERI Efficiency in AUTO Mode, Vout = 2.86 V



5.9 General Purpose ADC

Table 23: General Purpose ADC, T_J = -40 °C to +125 °C

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V_{DD}	ADC reference voltage	V _{DD} = VDDCORE	2.45	2.5	2.55	V
	Off current				1	μΑ
	ADC resolution			10		bit
	ADC integral non linearity			± 2		LSB
	ADC differential non linearity			± 0.8		LSB
	ADC absolute accuracy	Note 4		13	15	mV
R _{SRC} Note 1	Maximum source impedance				120	kΩ
CIN	Input capacitance	Total input capacitance		10.5		pF
	V _{SYS} voltage range Channel 0	V _{SYS} minus VDDCORE V _{SYS} = 3.125*(ADC/255)+2.5 (Auto) V _{SYS} = 3.125*(ADC/1023)+2.5 (Man)	2.5		5.5	V
	ADCIN1 to 3 voltage range Channel 1 to 3	V _{IN} = (ADC*2.5)/255 (Auto) V _{IN} = (ADC*2.5)/1023 (Man)	0		2.5	V
	Internal temperature Sensor voltage range Channel 4	T _J = -0.398 * ADC +330	0		0.833	V
	V _{BBAT} voltage range Channel 5	V _{BBAT} = (ADC*5)/1023	0		5.0	V
	Regulator monitor voltage range Channel 8 to 10	V _{REG} = (ADC*5)/255	0		5.0	V
	Inter channel isolation	Note 2		60		dB
	ADCIN1,2 current source Note 3		-3%	1 to 40	3%	μΑ
	COMP1V2 comparator level Channel 2			1.2		٧

Note 1 R_{SRC} is the impedance of the external source the ADC is sampling.

Note 2 80 dB for channel A2 (ADC_IN2).

Note 3 Variance guaranteed for 10 μ A to 40 μ A and up to 2 V output voltage.

Note 4 Excludes reference voltage variation.



5.10 Internal Oscillator

Table 24: Internal Oscillator, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
	Internal oscillator frequency	After trimming	5.7	6.0	6.3	MHz

5.11 POR, Reference Generation, and Voltage Supervision

Table 25: POR, Reference Generation and Voltage/Temperature Supervision, T_J = -40 °C to +125 °C

Parameter	Description	Conditions	Min	Тур	Max	Unit
Vpor_lower	Deep discharge lockout lower threshold			2.0		V
VPOR_UPPER	Deep discharge lockout upper threshold			2.3		V
V _{DD_FAULT_LOWER} Note 1	Under-voltage lower threshold		2.5	2.8	3.25	V
Vdd_fault_lower_acc	Under-voltage lower threshold accuracy			±2		%
VDD_FAULT_UPPER Note 2	Under-voltage upper threshold			V _{DD_FAULT_LOWER} + VDD_HST_ADJ		V
V _{REF}	Reference voltage		-1.25%	1.2	+1%	V
	VREF decoupling capacitor			0.22		μF
	VLNREF decoupling capacitor			0.22		μF
I _{REF}	Reference current resistor		-1%	200	+1%	kΩ

Note 1 During production V_{DD_FAULT_LOWER} voltage is configured via OTP over the range 2.5 V to 3.25 V in 50 mV steps.

5.12 Thermal Supervision

Parameter	Description	Conditions	Min	Тур	Max	Unit
TEMP_WARN	Thermal warning	Note 1	120	125	130	ů

Note 2 During production the hysteresis between VDD_FAULT_LOWER and VDD_FAULT_UPPER is configured via OTP over the range 100 mV to 450 mV in 50 mV steps, the hysteresis can be further changed through control VDD_HYST_ADJ.



TEMP_CRIT	Thermal shutdown	Note 1	135	140	145	°C
TEMP_POR	Thermal POR threshold	Note 1	145	150	155	ů

Note 1 Thermal thresholds are non-overlapping.



6 Functional Description

The DA9063L provides separate power domains for the host processor, memory, and peripherals to enable a flexible low-power system design. Multiple low-power modes permit varying combinations of peripherals to be powered off to conserve battery power. Other system components, such as DRAM and FLASH memory, RF transceivers, audio codec, and companion chips, are supplied from optimized regulators designed for dedicated power requirements. The DA9063L power supplies can be programmed to default voltages via OTP and provide system-configuration flexibility by selecting the power-up sequence of the regulators and switching converters.

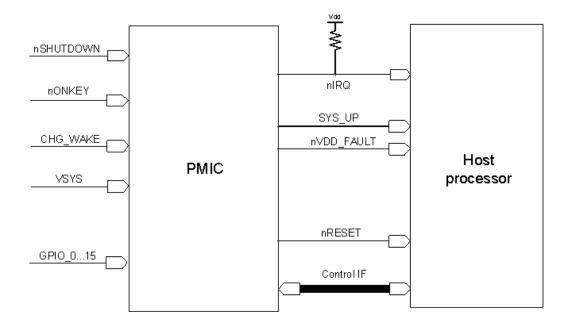


Figure 12: Control Ports and Interface



6.1 Power Manager IO Ports

The power manager input ports are supplied from either the internal rail VDDCORE or VDD_IO2, selected via PM_I_V. The output ports are supplied from VDD_IO1 or VDD_IO2, selected via PM_O_V (nVDD_FAULT, GP_FB1 via GPIO controls). During the initial start-up sequence all power manager IO ports (with the exception of nRESET and nIRQ) are in a high impedance (tri-state) mode until they are configured from OTP prior to reaching POWERDOWN mode. Output ports are push-pull except for nRESET and nIRQ, which can also be configured as open-drain via PM_O_TYPE.

6.1.1 On/Off Port (nONKEY)

The nONKEY signal is a wakeup interrupt/event intended to power-on the application supplied by DA9063L. The level of the debounced signal is provided by status flag nONKEY (asserted at low level). The nONKEY unit is always enabled so that the application can be powered-on when the GPIO extender is disabled. The IRQ assertion and wakeup event can be suppressed via the interrupt mask M nONKEY.

nONKEY provides four modes of operation selected by field nONKEY_PIN in register CONFIG_I.

Table 26: nONKEY_PIN Settings

nONKEY_PIN	Description
00	An E_nONKEY event is generated when the debounced signal from port nONKEY goes low (asserting edge). If not masked, an interrupt is signaled to the host via nIRQ (with wakeup during POWERDOWN mode).
01	If (after powering up from POWERDOWN mode) the debounced signal from port nONKEY is low after an asserting edge for less than the key-press time (selected by KEY_DELAY, default 2 s), an E_nONKEY event is generated at the releasing edge. If the signal is low for longer than selected by KEY_DELAY, the DA9063L asserts the event E_nONKEY plus control nONKEY_LOCK when it reaches the selected key-press time.
10	If (after powering up from POWERDOWN mode) the debounced signal from port nONKEY is low after an asserting edge for less than the key-press time (selected by KEY_DELAY, default 2 s), an E_nONKEY event is generated at the releasing edge. If the signal is low for longer than selected by KEY_DELAY, the DA9063L asserts the event E_nONKEY plus control nONKEY_LOCK when it reaches the selected key-press time and powers down by clearing control SYSTEM_EN.
11	If (after powering up from POWERDOWN mode) the debounced signal from port nONKEY is low after an asserting edge for less than the key-press time (selected by KEY_DELAY, default 2 s), an E_nONKEY event is generated at the releasing edge. Control SYSTEM_EN is cleared and STANDBY asserted, which triggers a partial power down from a short press. If the signal is low for longer than selected by KEY_DELAY, the DA9063L asserts the event E_nONKEY plus control nONKEY_LOCK and clear SYSTEM_EN plus STANDBY when it reaches the selected pressing time (powers down to full POWERDOWN from a long press).

For nONKEY_PIN settings other than '00', the wakeup is not suppressed by an asserted M_nONKEY. With an asserted nONKEY_LOCK, the wakeup is only executed if the debounced nONKEY signal is asserted for more than the key-press time (selected by KEY_DELAY, default 2 s). This behaves similarly to a keypad lock since any short (unintended) pressing of nONKEY does not wake the application. If the application also has wakeup from a short nONKEY press, the host has to clear nONKEY_LOCK before entering POWERDOWN mode. In mode '10' when nONKEY key press is longer than the time selected by KEY_DELAY, SYSTEM_EN is re-asserted in mode '11'. SYSTEM_EN is re-asserted from any consecutive pressing of nONKEY. nONKEY_LOCK is



automatically cleared by the DA9063L when powering up from POWERDOWN mode. POWERDOWN mode is described in Section 6.2.2.

Note

During DELIVERY mode, the functionality of nONKEY is restricted to a termination of this mode. To enable this feature, the pull-up resistor of nONKEY has to be connected to V_{SYS}. Asserting nONKEY stops the DELIVERY mode and triggers a start-up of the DA9063L.

6.1.2 Wakeup Port (CHG_WAKE)

The CHG_WAKE signal is a rising edge sensitive, wakeup interrupt/event intended to wake the DA9063L from an event on the companion charger (for example, supply insertion). The CHG_WAKE port is always enabled so that the application can be powered-on with a disabled GPIO extender. The IRQ assertion and wakeup event can be suppressed via the interrupt mask M_WAKE. During DELIVERY mode, asserting CHG_WAKE terminates this mode.

6.1.3 Hardware Reset (nOFF, nSHUTDOWN, nONKEY, GPIO14, GPIO15, WATCHDOG)

The DA9063L nOFF port is an active-low input (no debouncing) typically initiated by an asserted error detection line. It asserts nSHUTDOWN in the fault register. The sequencer asserts port nRESET, and all domains and supplies of the DA9063L except LDOCORE are disabled in a fast emergency shutdown.

The DA9063L nSHUTDOWN port is an active-low input typically asserted from a host processor (or a push button switch). It asserts nSHUTDOWN in the fault register. The sequencer asserts port nRESET and then powers down all domains in reverse sequencer order down to slot 0 and all supplies of the DA9063L except LDOCORE are disabled. HOST_SD_MODE determines if normal power sequence timing or a fast shutdown is implemented.

The DA9063L includes a third hardware reset trigger that follows the debounced nONKEY signal after being asserted for a period greater than KEY_DELAY + SHUT_DELAY. The same can be achieved by a long parallel connection of GPI14 and GPI15 to ground. The long nONKEY shutdown and GPI014/15 shutdown are enabled by the power manager control register bits nONKEY_SD and GPI14_15_SD.

If the hardware reset was initiated by a (debounced) press of nONKEY (or GPIO14 and GPIO15 together) longer than SD_DELAY, the DA9063L initially only asserts control bit KEY_RESET in the fault register and signals a non-maskable interrupt allowing the host to clear the armed reset sequence within 1 s. If the host does not clear KEY_RESET then a shutdown to RESET mode is executed. KEY_SD_MODE determines if normal power sequence timing or a fast shutdown is implemented.

The DA9063L then waits for a valid wakeup event (for example, a key press) or starts the power sequencer automatically if AUTO_BOOT is configured.

If the WATCHDOG has been disabled, this hardware reset can be used to turn off the application in the event of a software lock-up without removing the battery. This type of reset should only be used for severe hardware or software problems as it will completely reset the processor and could result in data loss.



6.1.4 Reset Output (nRESET)

The nRESET signal is an active-low output signal from DA9063L to the host processor that can either be push-pull or open drain (selected via PM_O_TYPE), which tells the host to enter the reset state. nRESET is always asserted at the beginning of a DA9063L cold start from NO-POWER and DELIVERY modes. It is asserted in ACTIVE mode before the DA9063L starts powering down to RESET mode (triggered from user, host, or an error condition detected by the DA9063L). nRESET may also be asserted (depending on nRES_MODE setting) as a soft reset before the sequencer starts powering down without progressing to RESET mode.

An assertion of nRESET from voltage supervised regulators being out-of-range can be enabled via control MON RES (minimum assertion time 1 ms).

After being asserted, nRESET remains low until the reset timer has been started from the selected trigger signal and expires. The reset release timer trigger signal can be selected via RESET_EVENT to be EXT_WAKEUP, SYS_UP, PWR_UP, or leaving PMIC RESET state. The expiry time can be configured via RESET_TIMER from 1 ms to 1 s.

6.1.5 System Enable (SYS EN)

SYS_EN is an input signal from the host processor to the DA9063L that enables the regulators in domain SYSTEM. The feature is enabled using GPIO8_PIN and configured as active-low or -high by GPIO8_TYPE. It asserts SYSTEM_EN and simultaneously generates an IRQ. It also triggers a wakeup event in POWERDOWN mode if enabled via GPIO8_WEN. De-asserting SYS_EN (changing from active to passive state) clears control SYSTEM_EN which triggers a power down sequence into hibernate/standby mode (without IRQ assertion or wakeup event trigger). By setting nRES_MODE, the port SYS_EN can be used as a soft reset input with the assertion of nRESET before powering down. With the exception of supplies that have the xxxx_CONF control bit asserted, all regulators in power domains POWER1, POWER, and SYSTEM are sequentially disabled in reverse order. Regulators with the <x>_CONF bit set remain on but change the active voltage control registers from V<x> A to V<x> B (if V<x> B is not already selected).

The control register bit SYSTEM_EN can also be used to power down domain SYSTEM by a software command. It can be read and changed via the control interfaces and can be initialized from OTP when leaving POWERDOWN mode. The DA9063L will not process any changes on port SYS_EN or register control SYSTEM_EN until the sequencer has stopped processing IDs.

6.1.6 Power Enable (PWR_EN)

PWR_EN is an input signal from the host processor to the DA9063L. The input signal can be configured as active-high or -low via GPIO9_TYPE, and to trigger a wakeup event from POWERDOWN mode if configured via GPIO9_WEN. Initialization, IRQ assertion and the direct control via register bit POWER_EN is similar to the function of SYS_EN in domain SYSTEM as described in Section 6.1.5. To ensure correct sequencing, SYSTEM_EN (SYS_EN) must be active before asserting PWR_EN/POWER_EN. When de-asserting PWR_EN/POWER_EN, the sequencer sequentially powers down POWER1 and POWER domains.

6.1.7 Power1 Enable (PWR1_EN)

PWR1_EN is an input signal from a host to the DA9063L. The input signal can be configured as active-high or -low via GPIO10_TYPE, and to trigger a wakeup event in POWERDOWN mode if enabled via GPIO10_WEN. Initialization, IRQ assertion and the direct control via register bit POWER1_EN is similar to the function of SYS_EN in domain SYSTEM as described in Section 6.1.5. POWER1 is a general purpose power domain.



6.1.8 GP_FB1, General Purpose Signal 1 (EXT_WAKEUP/READY)

This port supports two different modes selected by the control PM_FB1_PIN.

Table 27: PM_FB1_PIN Settings

PM_FB1_PIN	Description
0	EXT_WAKEUP. This output signal to the host processor indicates a valid wakeup event during POWERDOWN mode. External signals that can trigger wakeup events are debounced before the EXT_WAKEUP signal is asserted. EXT_WAKEUP is released when register control SYSTEM_EN is asserted (minimum pulse duration = 500 µs).
1	READY. The READY signal indicates on-going DVC or power sequencer activities. The READY signal is asserted (typically active-low) from the DA9063L power sequencer when the processing of IDs commences, and is released when the target power state (final sequencer slot) has been reached. READY is also asserted during DVC voltage transitions.

The active level is configured via the control GPIO13_MODE. The logical threshold voltage is selected by GPIO13_TYPE.

6.1.9 **GP_FB2**, **General Purpose Signal 2** (PWR_OK/KEEP_ACT)

The GP_FB2 port supports two different modes selected by the control PM_FB2_PIN.

Table 28: PM_FB2_PIN Settings

PM_FB2_PIN	Description
0	PWR_OK. In this mode the port is a regulator status indicator. The port is an open drain output asserted if none of the selected regulators are out-of-range. The regulator monitoring via ADC must be enabled and all regulators to be monitored must have supervision enabled with the selected persistence, and mask bit M_REG_UVOV must be asserted. In case at least one of the supervised regulators is out-of-range or regulator monitoring is disabled, the PWR_OK signal is low.
1	KEEP_ACT. If enabled, every assertion of the port (rising to active level edge sensitive) sets the watchdog trigger, similar to writing to bit WATCHDOG via the power manager bus. The host has to release KEEP_ACT before the next assertion during continuous watchdog supervision (if enabled). The minimum assertion and de-assertion cycle time is 150 μs.

The output active level (and driver type) can be configured via GP_FB2_TYPE.

Alternatively, with BCORE_MERGE = 1, FB in register BCORE1_CFG set to 0b000 and MERGE_SENSE = 0, the GP_FB2 pin becomes a voltage feedback signal for BUCKCORE.

6.1.10 GP_FB3, General Purpose Signal 3 (nVIB_BRAKE)

The GP_FB3 port supports two different modes selected by the control PM_FB3_PIN.

Table 29: PM_FB3_PIN Settings

PM_FB3_PIN	Description		
0	nVIB_BRAKE disabled.		
1	nVIB_BRAKE enabled. If LDO8 is configured as a vibrator motor driver, GP_FB3 can be configured to provide an external brake signal. The vibrator motor can be started or stopped by a change in the level on the nVIB_BRAKE signal. If the port is not used as a		



brake command, the vibration motor runs continuously at the speed configured by
VIB_SET.

GP FB3 TYPE defines the active level.

6.1.11 Supply Rail Fault (nVDD_FAULT)

nVDD_FAULT is a signal to the host processor to indicate a supply voltage (V_{SYS}) low status. Asserting nVDD_FAULT indicates that the main supply input voltage is low ($V_{SYS} < VDD_FAULT_UPPER$) and therefore informs the host processor that the power will shut down soon. The event control E_VDD_WARN is asserted and the nIRQ line is asserted (if not masked). During POWER_DOWN mode a wakeup is generated. After that the processor may operate for a limited time from the remaining battery capacity or the processor may enter a standby mode. As long as V_{SYS} does not recover, the host can re-enable the nIRQ line by asserting M_VDD_WARN or clearing E_VDD_WARN. The DA9063L starts a fault power down sequence. If V_{SYS} drops below VDD_FAULT_LOWER, the DA9063L enters RESET mode. The VDD_FAULT_LOWER threshold and the hysteresis on VDD_FAULT_UPPER are OTP configurable.

The nVDDFAULT port can alternatively be controlled by the state of the debounced V_{SYS} monitor inside the ADC (selected via GPIO12_PIN). The signal is asserted when the ADC detects three consecutive results below the configurable threshold VSYS_MON (it becomes passive after three consecutive results above VSYS_MON). This provides a variable power good signal to trigger boot activities on external ICs.

The active level/debounce, wakeup, and IO supply voltage can be selected via the controls GPIO12_MODE, GPIO12_WEN and GPIO12_TYPE, respectively.

6.1.12 Interrupt Request (nIRQ)

The nIRQ is an output signal that can either be push-pull or open drain (selected via PM_O_TYPE). If an active high IRQ signal is required, it can be achieved by asserting control IRQ_TYPE (recommended for push-pull mode). This port indicates that an interrupt-causing event has occurred and that event/status information is available in the EVENT and STATUS registers. Events are triggered by a status change at the monitored signals. When an event bit is set, the nIRQ signal is asserted (unless this interrupt is masked by a bit in the IRQ mask register). The nIRQ is not released until all event registers with asserted bits have been read and cleared. New events that occur during reading an event register are held until the event register is cleared, ensuring that the host processor does not miss them. The same happens to all events occurring while the sequencer processes time slots (that is, the generation of interrupts is delayed).

6.1.13 IO Supply Voltage (VDD IO1 and VDD IO2)

VDD_IO1 and VDD_IO2 are two independent IO supply rail inputs of the DA9063L that can be individually assigned to the power manager interfaces (see control bit GPI_V), power manager IOs (see control bits PM_O_V, PM_I_V) and GPIOs (bits GPIOx_TYPE). The rail assignment determines the IO voltage levels and logical thresholds, see Section 5.4. The selection of the supply rail for GPIOs is also partially used for their alternate functions, see Table 2. As an example, GPIO13_TYPE determines the supply rail when this pin is configured as the GP_FB1 output.



6.2 Operating Modes

6.2.1 ACTIVE Mode

A running application is typically in ACTIVE mode. The DA9063L transitions to ACTIVE mode after the host processor performs at least one initial 'alive' watchdog write (or alternatively an initial assertion of the KEEP_ACT port) inside the target time window. If the WATCHDOG function is disabled by setting TWDSCALE to zero, the DA9063L transitions to ACTIVE mode when all of the sequencer IDs in the POWER domain are complete.

In ACTIVE mode, the PMIC core functions as LDOCORE and internal oscillator are running. Typically additional features are enabled, such as the GPADC. The DA9063L can send interrupt requests to the host via a dedicated interrupt port (nIRQ) and status information can be read from the host processor via the power manager interface. Temperature and voltages inside and outside the DA9063L can be monitored and fault conditions can be flagged to the host processor.

6.2.2 POWERDOWN Mode

The DA9063L is in POWERDOWN mode when the power domain SYSTEM is disabled (even partially). This can be achieved when progressing from NO-POWER/DELIVERY mode or by returning from ACTIVE mode. A return from ACTIVE mode is initiated by low power mode instructions from the host (for example, releasing signal SYS_EN or clearing register bit SYSTEM_EN), from the user by asserting nONKEY (if nONKEY_PIN='1x') or as an interim state during a shutdown to RESET mode.

During POWERDOWN mode LDOCORE, VREF reference voltage, the nONKEY pin, CHG_WAKE port, and the calendar counter are active. Dedicated power supplies can be kept enabled during POWERDOWN mode if their xxx_CONF bits are asserted (supply voltage settings are taken from the respective Vxxx_B registers).

GPIO ports, the GPADC, and the control interfaces also remain active in POWERDOWN mode if not configured otherwise via register PD_DIS. Disabling these blocks during POWERDOWN mode reduces quiescent current, especially if all blocks that require an oscillator clock are disabled (HS2WIRE_DIS, GPADC_PAUSE, GPI_DIS, PMCONT_DIS). If required, the application supervision by the WATCHDOG timer can be continued in POWERDOWN mode via WATCHDOG_PD. If the host will not communicate with the DA9063L during POWERDOWN mode, then the control interfaces may also be temporarily disabled (see controls HS2WIRE_DIS).

If the sequencer pointer has stopped at position PART_DOWN (inside domain SYSTEM) it results in a partial power down. When on the way down the sequencer pointer reaches position 0, relevant regulators with corresponding position 0 IDs that have cleared control Bxxx_CONF/LDOxx_CONF/xxx_SW_CONF are disabled, otherwise the regulator voltages change to the values defined in VBxxx_B/VLDOxx_B when control DEF_SUPPLY is asserted. When DEF_SUPPLY is released, slot 0 is not processed by the sequencer, hence regulators with an ID pointing to slot 0 remain unchanged. Following the next wakeup event Vxxx_A voltage levels and the sequencer power domain controls/timers are set to their default OTP values if OTPREAD_EN is asserted.

When nONKEY_SD is asserted and the key is continuously pressed for longer than the time selected by KEY_DELAY + SHUT_DELAY, it asserts KEY_RESET to indicate that the transition to RESET mode was triggered by a long nONKEY, see Section 6.1.1.

When the device is in POWERDOWN or RESET mode, asserting ECO_MODE enables low power. This is achieved internally by using a pulsed mode for VDDCORE and reference voltage generation. This maintains basic functionality but full parametric compliance is no longer guaranteed (as it affects



ADC precision, buck performance, LDO voltage resolution, and so on). The pulsed mode is driven from a (free-running) low-power on-chip oscillator.

6.2.3 RESET Mode

The DA9063L is in RESET mode when a complete application shutdown is required. The RESET mode can be triggered by the user, a host processor or by an action on the DA9063L, as outlined below:

- By the user:
 - from a long press of nONKEY (interruptible by host)
 - o from a long parallel assertion of GPIO14 and GPIO15 (interruptible by host)
- By pressing a reset switch connected to port nSHUTDOWN (non-interruptible)
- Forced from the host processor (non-interruptible) by:
 - asserting port nSHUTDOWN (falling edge)
 - writing to register bit SHUTDOWN
- By an error condition that forces a RESET mode (non-interruptible):
 - no WATCHDOG write (KEEP_ACT signal assertion) from the host inside the watchdog time window (if watchdog was enabled)
 - an under-voltage detected at V_{SYS} (V_{SYS} < VDD_FAULT_LOWER)
 - o an internal die over-temperature
- Forced by the error detection line (non-interruptible):
 - by asserting port nOFF (falling edge)

The controls INT_SD_MODE, HOST_SD_MODE, and KEY_SD_MODE can be used to individually configure the shutdown sequences from an internal fault, host or user trigger. In each case, the sequence can be configured to implement either the reverse timing of the power-up sequence or an immediate transition into RESET mode, skipping any delay from the sequencer or dummy slot timers. Asserting nOFF always triggers a fast emergency shutdown. To allow the host to determine the reason for the reset, the source is recorded in FAULT_LOG (as either the KEY_RESET or nSHUT_DOWN bit). The host processor clears FAULT_LOG by writing asserted bits with a 1.

Note

- KEY_SD_MODE = 1 enables a full POR following a long press of ONKEY or a long assertion of GPIO14 and 15.
- In the case of an aborted OTP read, the DA9063L enters RESET mode without asserting any bits in FAULT_LOG.

A shutdown to RESET mode begins with the DA9063L asserting the nRESET port. Then domain SYSTEM is completely powered down (sequencer position 0) at which time the device has reached RESET mode: this is a low current consumption state. The only circuits in RESET mode remaining active are LDOCORE (at a reduced level of 2.2 V), the control interfaces and GPIOs, the VREF reference, and the comparators for over-temperature and $V_{\rm SYS}$ level. Regulators and blocks are automatically disabled to avoid draining the battery. During the DA9063L RESET mode, the host processor can be held in a RESET state via port nRESET.



When entering RESET mode, all user and system events are cleared. When leaving RESET mode, the complete DA9063L register configuration is reloaded from OTP (with the exception of AUTO_BOOT in case of a VDD_START fault).

Note

FAULT_LOG, GP_ID_10 to GP_ID_19 and other non-OTP loaded registers remain unchanged when leaving RESET mode.

nRESET is always asserted low after a cold start from NO-POWER or DELIVERY mode and can also be asserted (depending on configuration of nRES_MODE) before the sequencer starts to power down towards POWERDOWN mode.

Some reset conditions such as shutdown via register write, watchdog error, or over-temperature automatically expire (that is, are automatically cleared by the device as it shuts down). Other RESET triggers such as via port nOFF or nSHUTDOWN need to be released before the DA9063L can move from RESET to POWERDOWN mode. In the case that the application requires regulators to discharge in advance of a consecutive power-up sequence, a minimum duration of the RESET mode can be selected via RESET_DUR.

If the reset was initiated by user action from a long nONKEY key-press (or GPI14 and GPI15), bit KEY_RESET is set and the nIRQ port asserted. After 1 s the shutdown sequence is started, unless this is inhibited by the host clearing KEY_RESET within this 1 s period (by writing a 1 to the related bit in register FAULT_LOG). When the RESET condition has been removed, the DA9063L requires the presence of a good supply (V_{SYS} > VDD_FAULT_UPPER and able to provide enough power) before it can start-up again and move into POWERDOWN mode.

6.2.4 DELIVERY Mode

The DELIVERY mode provides the lowest possible quiescent current, allowing connected precharged batteries to maintain charge prior to the end-user starting the device for the first time. During DELIVERY mode, only the nONKEY, CHG WAKE, and the VDDREF detection circuitry is enabled.

6.2.5 NO-POWER Mode

The DA9063L enters NO-POWER mode when VDDCORE drops below the VPOR_LOWER threshold. As long as VDDCORE stays below the VPOR_UPPER threshold, an internal power-on-reset (nPOR) signal remains asserted. In this mode, only the VDDCORE threshold comparator is active. This comparator simply checks for a condition that allows the DA9063L to turn on again. When a good supply is subsequently available again on V_{DDREF} (> 2.4 V), VDDCORE is able to rise above VPOR_UPPER and the DA9063L leaves NO-POWER mode.

6.2.6 Power Commander Mode

This is a special mode for evaluation and configuration development. In Power Commander mode, the DA9063L is configured to load the control register default values from the HS 2-WIRE interface, instead of from the OTP cells, so that un-programmed DA9063L samples will power up, allowing evaluation and verification of a proposed user configuration.

Power Commander mode is enabled by connecting TP to a 3.3 V to 5.0 V voltage.

Note

In Power Commander mode, GPI14 and 15 are configured for HS 2-WIRE interface operation (with VDDCORE as the supply) and GPO12 is configured as an output for nVDD_FAULT. Any register writes or



OTP loads which can change this configuration are ignored until DA9063L has exited from Power Commander mode.

After leaving the POR state, the DA9063L informs the system that it is waiting for a programming sequence by driving nVDD_FAULT low. The software running on the PC monitors nVDD_FAULT and responds by downloading the values into the configuration registers within DA9063L. nVDD_FAULT is automatically released after the download is complete.

There are two programming sequences performed in Power Commander mode. The first takes place between RESET and POWERDOWN mode and the second between POWERDOWN and SYSTEM mode.

Note

To correctly configure DA9063L, addresses 0x0A to 0x36, 0x82 to 0xCF, and 0x104 to 0x12E should be programmed during the first sequence. Registers 0x0E, 0x82, and 0xA3 to 0xB3 should be programmed during the second sequence.

When the first programming sequence is complete, DA9063L will be in POWERDOWN mode. Progression from this mode is determined by the values programmed for SYS_EN and AUTO_BOOT. If DA9063L has been directed to progress from POWERDOWN mode then it drives pin nVDD_FAULT low for a second time to request that the SW performs the second programming sequence.

Once the second programming sequence is complete, the progress of the power-up sequence is controlled by the values loaded during the programming sequence.

The programmed configuration can be identified by reading the fuse register CONFIG ID.

Note

During Power Commander mode, the fault detection status bit VDD_FAULT and the level at the related pin nVDD-FAULT do not match and do not indicate a low voltage level at VDDOUT. An enabled shutdown from a 5 s assertion of GPIO14/15 will be ignored during POWER Commander mode. Any nIRQ and event assertion when accessing the HS 2-WIRE interface (E_GPI14) is suppressed in this mode.



6.3 Start-Up from NO-POWER Mode

6.3.1 Power-On-Reset (nPOR)

The DA9063L generates an internal power-on-reset nPOR (active low) following the initial connection of a supply to V_{DDREF}.

While the VDDCORE voltage is below the threshold VPOR_UPPER, the internal signal nPOR is driven low and the DA9063L will not start-up. This is NO-POWER mode. When the VDDCORE voltage rises above VPOR_UPPER, the following occur:

- The nPOR is driven high (flagged by the POR bit being set in register FAULT_LOG).
- The oscillator is enabled.
- The VREF reference is enabled.
- The complete OTP block is read and stored in the register bank.
- The DA9063L progresses into POWERDOWN mode.

From POWERDOWN mode, the DA9063L continues through the power-up sequence if either:

- the power domain SYSTEM was enabled by the input port, SYS_EN, or,
- the power domain SYSTEM was enabled in OTP settings and AUTO_BOOT was enabled.

With AUTO_BOOT disabled and the power domain SYSTEM enabled in OTP settings, a non-suppressed wakeup event allows the DA9063L to continue through the power-up sequence.

6.4 Exiting Reset Mode and Application Wakeup

Application startup requires a sufficient V_{SYS} supply, as determined by the VDD_START feature, and a wake-up event.

6.4.1 VDD START

The DA9063L VDD_START feature prevents endless looping of false-starts in a weakly-powered system. VDD_START limits the number of automatic restart attempts to three. With reference to Figure 13:

- Figure 13(a): the PMIC starts from NO-POWER mode.
- Figure 13(b): First startup attempt. The PMIC transitions to the ACTIVE state when V_{SYS} rises above V_{DD FAULT UPPER}.
- Figure 13(c): If, during the first startup attempt, the V_{SYS} supply drops below V_{DD_FAULT_LOWER} for more than 100 ms, the PMIC shuts down to the SHUTDOWN state. V_{DD_FAULT_UPPER} is increased automatically by 250 mV. However, the increased V_{DD_FAULT_UPPER} is limited to a maximum of 3.70 V.
- Figure 13(d): Second startup attempt. If V_{SYS} rises above the new V_{DD_FAULT_UPPER} threshold, then the PMIC will power-up to the ACTIVE state.
- Figure 13(e): If, during the second startup attempt, the V_{SYS} supply drops below VDD_FAULT_LOWER for more than 100 ms, the PMIC shuts down to the SHUTDOWN state. V_{DD_FAULT_UPPER} is increased automatically by a further 250 mV (500 mV above its original setting). However, the increased V_{DD_FAULT_UPPER} is limited to a maximum of 3.70 V.
- Figure 13(f): Third startup attempt. If V_{SYS} rises above the new V_{DD_FAULT_UPPER} threshold, then the PMIC will power-up to the ACTIVE state.



• Figure 13(g): If, during the third startup attempt, the V_{SYS} supply drops below V_{DD_FAULT_LOWER} for more than 100 ms, the PMIC shuts down. From then on, AUTO_BOOT and wake-up from non-user events are temporarily disabled. 'User events' are defined in Table 30.

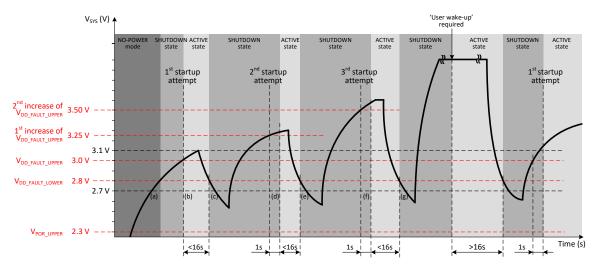


Figure 13: VDD_START Faults

NOTE

- During an attempt to restart, the entry into the ACTIVE state is delayed by 1 s.
- The ACTIVE state also includes POWER1 mode. The PMIC has the same VDD_START behavior from the POWERDOWN and SYSTEM modes as from the ACTIVE state.
- AUTO_BOOT and wake-up from non-user events are re-enabled after the application has successfully
 powered up to the ACTIVE state for more than 16 s. This also resets the start-up threshold to
 VDD_FAULT_UPPER + 0 mV.

Until the host has been booted, an OTP-enabled flashing LED may be driven from GPIO11, 14, or 15 to indicate to the user that the device is supplied with (insufficient) power. When CHG_WAKE is connected to a charger, the VDDSTART-triggered LED flashing continues as long as an external supply is charging the battery. The flashing LED can be configured via controls RESET_BLINKING, BLINK_DUR, and BLINK_FRQ. After the application is running, the blinking LED can be stopped via a host register write.

6.4.2 Wake-Up Events

DA9063L offers two types of wakeup event, user events and system events (see Table 30). Non-suppressed user events (for example, nONKEY, CHG_WAKE or from GPIOs) are always processed and trigger a wakeup.

Wakeup events can be individually suppressed by setting the related nIRQ mask bit. When nONKEY_LOCK is asserted a wakeup requires the debounced signal from nONKEY to be low for a time longer than the configured KEY_DELAY. It is not recommended to mask system events, instead disable the unwanted event sources (for example, GPIs, GPADC, 1.2 V comparator). The wakeup from GPIOs (or selected alternate features that use a shared GPI event) has to be enabled via GPI<</p>



After a valid wakeup condition is detected, a subset of the OTP configuration is read and the values are used to reconfigure the regulator voltage registers Vxxx_A, the power domain enable settings (if not suppressed via SYSTEM_EN_RD) and the sequencer timer.

DA9063L then asserts the EXT_WAKEUP signal towards the host processors and configures regulators with an ID pointing at slot 0 to their target state. If the power domains are not pre-enabled from OTP settings, the host processor must control further application start-up (via the power domain enable ports, SYS_EN, PWR_EN and PWR1_EN). Alternatively the DA9063L continues powering-up the OTP-enabled domains via the power domain sequencer, but the power sequencer will not start to enable the system supplies unless SYSTEM EN is asserted.

Progression to ACTIVE mode requires assertion of POWER_EN from the host via port PWR_EN, a register write, or enabled in OTP. After starting the WATCHDOG timer the host processor must assert the WATCHDOG bit within the configured time window. If this does not happen, the statemachine terminates ACTIVE mode and returns to RESET mode.

Table 30: Wakeup Events

Signal : Event	Wakeup	User Event	System Event	IRQ
V _{SYS} monitor : E_VDD_MON	Х		Х	Х
VDD_FAULT pre-warning : E_VDD_WARN	Х		Х	Х
Voltage comparator flipped : E_COMP1V2	Х		Х	Х
Pressed On key : E_nONKEY	Х	Х		Х
Wakeup from companion charger : E_WAKE	Х	Х		Х
LDO over current detect : E_LDO_LIM	Х		Х	Х
Regulator voltage out-of-range : E_REG_UVOV	Х		Х	Х
Critical junction temperature : E_TEMP	Х		Х	Х
Power sequencing ready : E_SEQ_RDY			Х	Х
Voltage ramping ready : E_DVC_RDY			Х	Х
Manual ADC result ready : E_ADC_RDY			Х	Х
GPIOs passive to active transition : E_GPIx	Х	Х		Х
ADC 1, 2, 3 threshold : via GPI0, 1, 2	Х		Х	Х
SYS_EN, PWR_EN, PWR1_EN (passive to active transition) : via GPIO8, 9, 10	Х		Х	Х
HS 2-WIRE interface : via GPIO14	Х		Х	Х



6.5 Power Supply Sequencer

The DA9063L power supplies are enabled with a sequencer that contains a programmable step timer, a programmable ID array of slot pointers, and four predefined pointers (SYSTEM_END, POWER_END, MAX_COUNT, and PART_DOWN), as illustrated by Figure 14. The sequencer is able to control up to 32 IDs (6 bucks, 5 LDOs, 5 external FET/IC controls, a Wait ID (GPI10), and an ID to activate power down settings), which can be grouped to three power domains.

The power domains have configurable size and their borders are described by the location pointers SYSTEM_END, POWER_END, and MAX_COUNT.

The lowest level power domain SYSTEM starts at step 1 and ends at the step that is described by the location pointer SYSTEM_END. The second level domain POWER starts at the successive step and ends at POWER_END. The third level domain POWER1 starts at the consecutive step and ends at MAX_COUNT. The values of pointer SYSTEM_END, POWER_END, and MAX_COUNT are predefined in OTP registers and should be configured as SYSTEM_END < POWER_END < MAX_COUNT.

The domain system can be thought of as the minimum set of supplies required to enable the core of the target system.

If the control OTPREAD_EN is enabled, the regulator voltages, sequence domain enables (if not suppressed via control SYSTEM_EN_RD), and the sequence timer are reset to their OTP values during the transition from power down to system.

The second level domain POWER includes supplies that are required on top to trigger the application and set the DA9063L into ACTIVE mode. POWER1 can be understood as one of the POWER domains that can be used for further sequenced control of supply blocks during ACTIVE mode (for example, for a sub-application like WLAN or a baseband chipset).

Note

It is recommended that the system is configured to reach ACTIVE mode before running applications.

6.5.1 Powering Up

All buck converters and LDOs of DA9063L have a unique sequencer ID. The power-up sequence is defined by an OTP register bank that contains a series of supplies (and other features), each of which point to a selected sequencer time slot. Several supplies can point to the same time slot which is therefore enabled in parallel by the sequencer. Time slots that have no IDs pointing at them are dummy steps that do nothing but insert a configurable time delay (marked in Figure 14 as D). Supplies/IDs that do not point to a sequencer time slot between 1 and MAX_COUNT are not enabled by the power sequencer but can be controlled individually by the host (via the power manager interface).

During power-up, the sequencer starts at slot 0. If DEF_SUPPLY is asserted, it checks all regulators for an ID pointing to slot 0. Cleared LDOxx_AUTO/ BUCKxxx_AUTO/xxx_SW_AUTO bits are configured by setting the related control Bxxx_CONF/LDOxx_CONF/xxx_SW_CONF, otherwise the regulator is enabled. To minimize inrush currents, it is recommended to enable no more than a single default regulator via DEF_SUPPLY. During power-up, the regulator output voltage is taken from the VBxxx_A/VLDOxx_A registers. During power-down, regulators with a cleared control in Bxxx_CONF/LDOxx_CONF/ xxx_SW_CONF are disabled, otherwise the regulator voltage is changed to VBxxx_B/VLDOxx_B when entering slot 0. When DEF_SUPPLY is released, slot 0 is not processed by the sequencer (regulators with an ID pointing at slot 0 remain unchanged).



The progression of the sequencer to slot 1 is dependent on certain conditions:

- If AUTO_BOOT and SYSTEM_EN are both asserted (via port, by register write or in OTP), the sequencer asserts the READY signal (if GP_FB1 is so configured) and then continues by processing slot 1.
- If AUTO BOOT is not asserted, the sequencer remains in a holding start state, waiting for either:
 - the assertion of SYSTEM EN, or,
 - o any other wakeup event if SYSTEM_EN is already enabled.

All supplies (and other sequenced features) that are pointing at slot 1 are then processed. This is similar to the processing of slot 0 with the exception that DEF_SUPPLY has no effect on slots apart from slot 0. From slot 1, the sequencer progresses until it reaches the position of pointer SYSTEM_END. At this point, all IDs of the first power domain SYSTEM are enabled and, if POWER_EN is not asserted, the DA9063L releases the READY signal (in combination with optional assertion of E_SEQ_RDY).

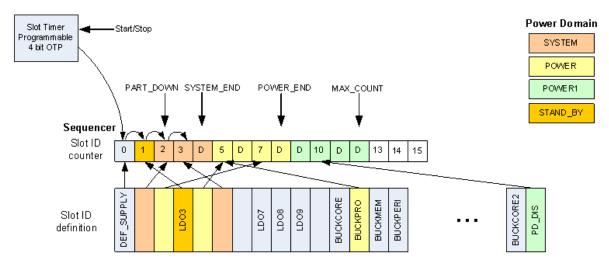


Figure 14: Assignment of Actions to Sequencer Slot IDs

Table 31: Power Sequencer Controlled Actions

Action	Sequencer Time Slot
Control LDO3	LDO3_STEP
Control LDO7	LDO7_STEP
Control LDO8	LDO8_STEP
Control LDO9	LDO9_STEP
Control LDO11	LDO11_STEP
Control BUCKCORE1	BUCKCORE1_STEP
Control BUCKCORE2	BUCKCORE2_STEP
Control BUCKPRO	BUCKPRO_STEP
Control BUCKIO	BUCK_IO_STEP
Control BUCKMEM	BUCKMEM_STEP



Action	Sequencer Time Slot
Control BUCKPERI	BUCKPERI_STEP
Assert/Release GPIO2	GP_RISE1_STEP
Release/Assert GPIO2	GP_FALL1_STEP
Assert/Release GPIO7	GP_RISE2_STEP
Release/Assert GPIO7	GP_FALL2_STEP
Assert/Release GPIO8	GP_RISE3_STEP
Release/Assert GPIO8	GP_FALL3_STEP
Assert/Release GPIO9	GP_RISE4_STEP
Release/Assert GPIO9	GP_FALL4_STEP
Assert/Release GPIO11	GP_RISE5_STEP
Release/Assert GPIO11	GP_FALL5_STEP
Wait for active state at GPI 10	WAIT_STEP
PD_DIS	PD_DIS_STEP

On completion of domain SYSTEM, the sequencer waits for POWER_EN to be asserted (via the PWR_EN port, a register write or in OTP). When POWER_EN is asserted, the signal READY is asserted (if not already asserted) and regulators/IDs of domain POWER are enabled sequentially. The sequencer stops at the position of pointer POWER_END. At this point it also: releases the READY signal (if POWER1_EN is not asserted); optionally asserts E_SEQ_RDY; enables the initial WATCHDOG timer and waits for the first associated alive feedback from the host processor. After this, the start-up of the DA9063L progresses into ACTIVE mode.

A third power domain, POWER1, can be enabled via POWER1_EN (asserted by PWR1_EN port, register write or in OTP). It enables all consecutive IDs until the position of pointer MAX_COUNT has been reached. The READY signal is asserted as long as IDs are processed (if enabled) and E SEQ RDY is asserted when reaching MAX COUNT.



6.5.2 Power-Up Timing

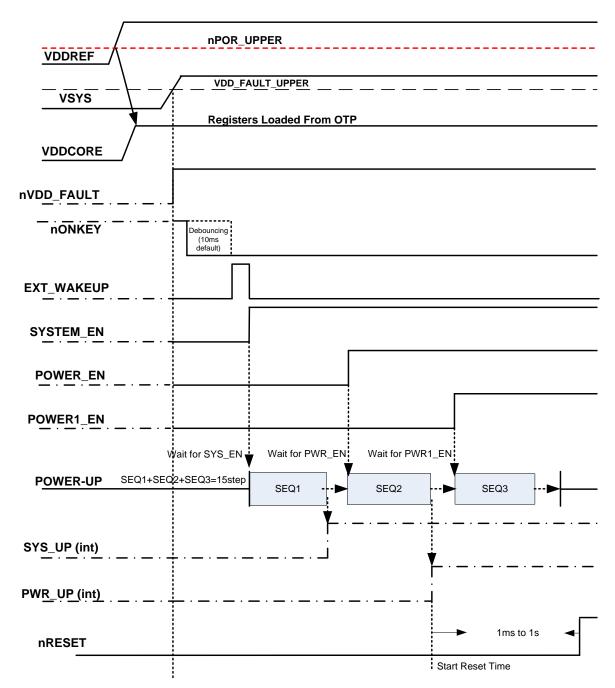


Figure 15: Power-Up Timing



6.5.3 Programmable Slot Delays

The delay between the slots of a sequence is controlled via the programmable value of SEQ_TIME in register SEQ_TIMER. This has a default delay of 128 µs per slot (min. 32 µs, max. 8 ms). The delay time between individual supplies can be extended by leaving a consecutive slot(s) with no IDs pointing to it: these are dummy slots. The dummy slots have an independent delay configured by SEQ_DUMMY. These delay times, in register SEQ_TIMER, are (re-)loaded from OTP every time domain SYSTEM begins to power-up.

6.5.4 Powering Down

When the DA9063L is powering down, the sequencer disables the supplies in reverse order and timing, asserts READY during sequencing, and triggers E_SEQ_RDY on reaching the target sequencer slot. Supplies that are configured to stay on (LDO<x>_CONF, B<x>_CONF, xxx_SW_CONF bit is set) are not disabled and are configured with the voltage setting from register VB<x>_B/VLDO<x>_B when the related time slot/ID is processed. The state of the regulators that are enabled for GPI control will not be changed by the sequencer when processing the related ID. This also applies for the selection of the related V<x>_A or V<x>_B voltage control register in case a regulator is enabled for GPI voltage selection.

If powering down is initiated by clearing POWER1_EN, the sequencer stops controlling IDs before the domain pointer POWER_END is reached. If POWER_EN is cleared, the domain POWER1 is powered down followed by POWER before the sequencer reaches pointer SYSTEM_END. These modes are used to temporarily disable optional features of a running application for reduced power (sleep mode).

If SYSTEM_EN is cleared the sequencer processes all IDs lower than the pointer position down to slot 0. The sequencer can be forced to stop the intended power down sequence prior to maturity at pointer position PART_DOWN via an asserted control STANDBY (PART_DOWN has to point into domain SYSTEM). In these cases the power sequencer has reached the application's POWERDOWN mode (hibernate/standby), which enables the option to reset regulator settings for the consecutive power-up sequence from OTP (enabled by OTPREAD_EN).

Wakeup events are enabled when the sequencer reaches slot 0 or pointer PART_DOWN (ignored outside of POWERDOWN mode). The assertion of nIRQ from events during POWERDOWN mode may be delayed until ACTIVE mode is reached the next time if configured by nIRQ_MODE. During processing slot 0, all supplies pointing into this step with a cleared control Bxxx_CONF/LDOxx_CONF/xxx_SW_CONF are disabled, otherwise the regulator voltage is changed to VBxxx_B/VLDOxx_B (if bit DEF_SUPPLY is asserted). Asserting control register bit SHUTDOWN first powers down to slot 0 and then forces the DA9063L into RESET mode. Autonomous features such as the Auto-ADC measurement can be disabled temporarily for POWERDOWN mode via register PD_DIS. The timing for processing PD_DIS can be defined by selecting a step inside the sequence. Features asserted in PD_DIS are (re-)enabled when PD_DIS is processed during a power-up sequence.

Control nRES_MODE enables the assertion of nRESET before executing a power-down sequence and starting the reset timer during the consecutive powering up. This is also true for partial POWERDOWN mode, when the sequencer powers down to pointer position PART_DOWN. The reset timer starts to run from the selected RESET_EVENT and releases the nRESET port after the reset timer expires.



6.5.5 User Programmable Delay

A conditional mode transition can be achieved using ID WAIT_STEP. If pointing into the power sequence the progress of an initiated mode transition can be synchronized, for example with the state of a host. This is indicated by toggling the signal at GPI10 to its configured active state. A safety timeout of 500 ms can be selected in TIME_OUT to trigger a power-down to RESET mode (including the assertion of WAIT_SHUT inside register FAULT_LOG) if E_GPI10 is not asserted in time. The ID WAIT_STEP provides an alternate timer mode, selected by WAIT_MODE and configured by WAIT_TIME, which provides a delay timer for a selected sequencer step. To enable symmetric sequence behavior, ID WAIT_STEP should not share a sequencer slot with other IDs. In the case of a shutdown sequence to RESET mode any waiting/delay at ID WAIT_STEP is skipped.



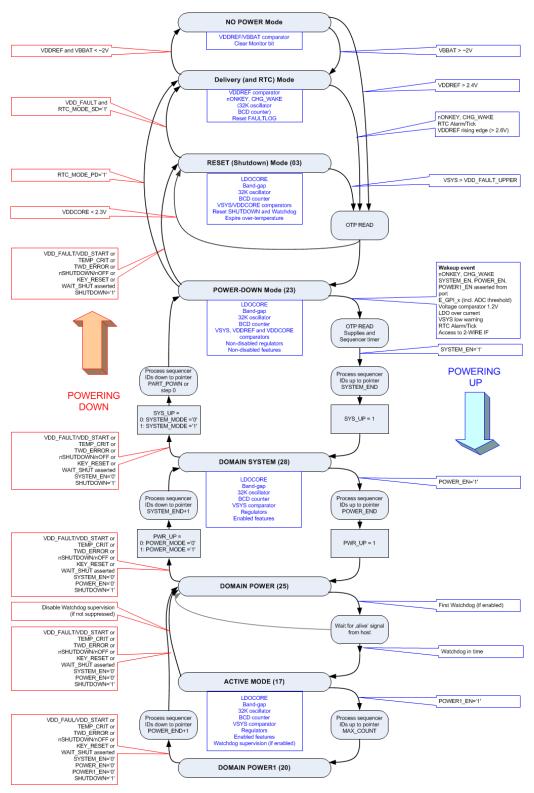


Figure 16: Power Mode Transitions



6.6 System Monitor (Watchdog)

After powering up domain POWER, the DA9063L can initiate a watchdog monitor function. The host processor must write a 1 within a configured two_max time into control WATCHDOG, thereby indicating that the host is alive. If the host does not write 1 to this watchdog bit within the two_max time, the DA9063L asserts TWD_ERROR in the FAULT_LOG register and powers down to RESET mode.

After this first write, the host must continue to write to this watchdog bit within the configured time or DA9063L powers down as described above. The time window has a minimum time twD_MIN fixed at 256 ms and a maximum time twD_MAX, nominally 2.048 s. The twD_MAX value can be extended by multiplying the nominal twD_MAX by the value of register bits TWDSCALE. TWDSCALE is used to extend the twD MAX time by x1, x2, x4, x8, x16, x32, or x64.

Once in the ACTIVE state, the DA9063L continues to monitor the system unless it is disabled by setting TWDSCALE to zero. When powering down from ACTIVE mode, the watchdog monitor is stopped unless it enters POWERDOWN mode via WATCHDOG PD.

If the WATCHDOG register bit is set to a 1 within the time window, the watchdog monitor resets the timer, sets the watchdog bit back to zero (this bit is always read as zero) and waits for the next watchdog signal. The watchdog trigger can also be asserted from the host by asserting KEEP_ACT in hardware. This mode is selected with control PM_FB2_PIN and removes the above requirement for the periodic setting of the watchdog bit.

The watchdog feature can be disabled by setting TWDSCALE to zero.

6.7 GPIO Extender

The DA9063L includes a GPIO extender that provides up to 16 V_{DDREF} -tolerant general purpose input/output ports, each controlled via registers from the host, see Table 32 and Figure 17.

The GPIO ports are pin-shared with ports from GPADC, HS 2-WIRE-interface and signals from the power manager. Configuration settings and events from GPIx ports are also shared with alternative features. For example, if GPIO1_PIN is configured to be ADCIN2, exceeding the configured ADC thresholds triggers a GPI1 event that generates a maskable GPI1 interrupt. The GPI active High/Low setting from the GPIOx_TYPE register and the selection of pull-up resistor is also applicable to the alternative port functions selected via GPIOx_PIN (for example, SYS_EN, PWR_EN and PWR1_EN). This is also true for GPIOx_WEN, which is used to enable triggering of a wakeup event (ADCIN1, ADCIN2, ADCIN3, SYS_EN, PWR_EN, PWR1_EN, HS 2-WIRE interface). When GPI ports are enabled (including being enabled by changing the setting of GPIOx_PIN), the GPI status bits are set to their non-active state. This ensures that any signals that are already active are detected and immediately generate any appropriate events.

In ACTIVE and POWERDOWN mode, the GPIO extender can continuously monitor the level of ports that are selected as general purpose inputs. GPIs are supplied from the internal rail VDDCORE or VDD_IO2 (selected via GPI_V) and can be configured to trigger events in active-high or active-low mode. The input signals can optionally be debounced (configurable via control DEBOUNCING, 10 ms default) and the resulting signal level is reflected by the status register GPIx. When the status has changed to its configured active state (edge sensitive) the assigned event register is set and the nIRQ signal is asserted (unless this nIRQ is masked). GPIs can be individually configured to generate a system wakeup via GPIxx_WEN.

If enabled via regulator controls LDOx_GPI/Bxxx_GPI, the ports GPI1, GPI2, and GPI13 can be used to enable/disable regulators (that is, controlling LDOx_EN/Bxxx_EN). The GPI active level is selected via the related GPIxx_TYPE control. GPI ports that are selected for this hardware control of one or more regulators do not generate events (nIRQ). GPI1, 2, and 13 can alternatively be selected to



toggle the VLDOx_SEL/VBxxxx_SEL. Apart from changing the regulator output voltage, this feature also allows hardware control of regulator mode (sync/sleep mode) via selection of the settings contained in xxxx_SL_A and xxxx_SL_B (but only for those bucks configured with Bxxxx_MODE = 00). When a regulator is controlled via GPI, its enable and voltage register selection are no longer controlled by the power sequencer (processing the related ID only affects non-GPI controlled functionality). However, these settings can still be changed via register writes from the control interface.

Events on GPI10 can be used to control the progress of the power sequencer. Processing ID WAIT STEP causes the sequencer to wait until GPI10 changes into its active state.

Note

Supplies directly enabled/disabled from GPI1, 2, or 13 have to be excluded from the power sequencer control (IDs of these supplies should point into a slot higher than MAX_COUNT)

If defined as an output, GPO0, 1, 3 to 6, 10 to 11, and 13 to 15 can be configured to be open-drain instead of push-pull. The supply rail can be individually selected from either VDD_IO1 or VDD_IO2. By disabling the internal 120 k Ω pull-up resistor when in open-drain mode, the GPO can also be supplied from an external rail (see registers CONFIG_K and CONFIG_L). The GPO output state reflects the respective register bit GPIOx MODE.

When configured as outputs, GPO 2, 7, 8, 9, and 11 can be controlled by the DA9063L power sequencer. Five pairs of level asserting and level releasing IDs (GP_RISE1_STEP/GP_FALL1_STEP to GP_RISE5_STEP/GP_FALL5_STEP) may be assigned individually to slots of the power sequencer, which trigger the configured level transition on the GPOs when processing the related ID during powering up (see Table 31 for assignments). The configured level change is inverted when processing the IDs during powering down. These are intended for use as enable signals either for external regulators or other devices in the system.

When the GPIO unit is off (POR), all ports are configured as open drain output with high level (pass device switched off, high impedance state). When leaving POR, the pull-up or pull-down resistors are configured from registers CONFIG_K and CONFIG_L. When the GPIO unit is temporarily disabled by the power sequencer (via GPI_DIS or PMCONT_DIS) level transitions on inputs are no longer detected and I/O drivers keep their configuration and programmed levels.

GPO12 can be driven by the state of VDD_MON to provide an active high 'Power good' signal (selected via GPIO12 PIN).

GPO10, 11, 14, and 15 are extended power GPO ports, where the maximum sink current is 11 mA and the maximum source current is 4 mA. This enables driving LEDs. The output ports GPO11, GPO14, and GPO15 can be toggled with a configurable periodic pulse configured via BLINK_FRQ and BLINK_DUR and include an optional PWM control. The generated PWM signals have a duty cycle from 0 % to 100 % with a repetition frequency of 21 kHz and 95 steps (using one 2 MHz clock for each step). The duty cycle is set by the controls GPO11_PWM, GPO14_PWM, and GPO15_PWM, with any value larger than 0 enabling the PWM mode of operation. The PWM control can also be made to dim the brightness between its current value and a new value at a rate of 32 ms per step. Selection of this mode is set by GPO11_DIM, GPO14_DIM, and GPO15_DIM. When set to zero the PWM ratio immediately changes. This creates a common anode tricolor LED brightness control. Flashing is driven from the crystal oscillator when control CRYSTAL has been asserted; otherwise an auxiliary on-chip oscillator is used.

LEDs are recommended to be low-side driven (using the GPIOs in sink mode) which is configured by setting GPIOx_MODE = 1.



Table 32: GPIO Overview

GPI O	Alternate Port	Alternate Port Shared Resources	GPI Wakeup	Remark
0	ADCIN1	E_GPI0, M_GPI0, GPI00_MODE	х	Auto measure ADC
1	ADCIN2	E_GPI1, M_GPI1, GPIO1_MODE	Regulator control x (in other modes)	Auto measure ADC/1.2V comparator, HW control of regulator
2	ADCIN3	E_GPI2, M_GPI2, GPIO2_MODE	Regulator control x (in other modes)	Auto measure ADC, HW control of regulator/ power sequencer controlled GPO
3			x	
4	CORE_SWS		Х	Buck remote feedback voltage sense
5			x	
6	PERI_SWS		х	Buck remote feedback voltage sense
7			X	Power sequencer controlled GPO
8	SYS_EN	E_GPI8, M_GPI8, GPIO8_TYPE, GPIO8_WEN, GPIO8_MODE	х	Power sequencer controlled GPO
9	PWR_EN	E_GPI9, M_GPI9, GPIO9_TYPE, GPIO9_WEN, GPIO9_MODE	х	Power sequencer controlled GPO
10	PWR1_EN	E_GPI10, M_GPI10, GPIO10_TYPE, GPIO10_WEN, GPIO10_MODE	х	High power GPO, input signal for ID WAIT
11			х	High power GPO (LED flashing/PWM), Power Sequencer controlled GPO
12	nVDD_FAULT	GPIO12_TYPE, GPIO12_WEN, GPIO12_MODE	x	VDD_MON state controlled GPO (POWER_GOOD)
13	GP_FB1	GPIO13_TYPE, GPIO13_MODE	Regulator control x (in other modes)	HW control of regulator
14	DATA	E_GPI14, M_GPI14, GPIO14_TYPE, GPIO14_MODE	х	High power GPO (LED flashing/PWM), Reset via long assertion in parallel with GPI15, 2nd 2-WIRE or DVC Control Interface
15	CLK		х	High power GPO (LED flashing/PWM), Reset via long assertion in parallel with



GPI O	Alternate Port	Alternate Port Shared Resources	GPI Wakeup	Remark	
				GPI14, 2nd 2-WIRE or DVC Control Interface	

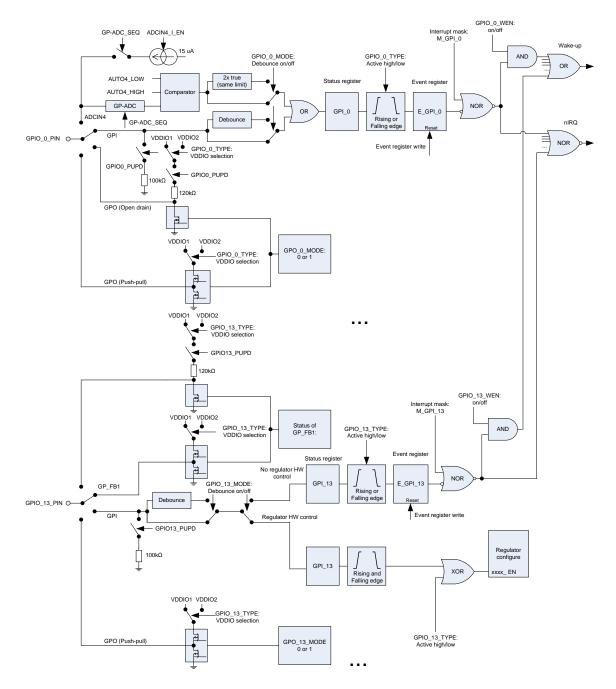


Figure 17: GPIO Principal Block Diagram (Example Paths)



6.8 Control Interface

The DA9063L is register controlled by the host software. Data is shifted into or out from DA9063L under the control of the host processor that also provides the serial clock. The interface is usually only configured once from OTP values, which are loaded during the initial start-up. The interface configuration can be changed by the host. However, care must be taken that changes are not made while the interface is active. If enabled, IF_RESET forces a reset of all control interfaces when port nSHUTDOWN is asserted.

The HS 2-WIRE interface is the serial control bus. It consists of DATA (data line) and CLK (clock line) and can be used as an independent control interface for data transactions between the DA9063L and a second host processor. The DA9063L HS 2-WIRE interface has a configurable 8-bit write address (default 0xB4) and a configurable read address (default 0xB5). For details of configurable addresses see control IF_BASE_ADDR in Section A.4.2 The interface is enabled if DATA was selected via configuration control GPIO14_PIN. The bus lines have to be pulled high by external pull-up resistors (2 k Ω to 20 k Ω). GPIO14_TYPE defines the supply rail of the interface (used for input logic levels and the internal pull-up resistors). The controls GPIO15_PIN and GPIO15_WEN are disabled when enabling the interface via GPIO14_PIN.

When the interface receives a read or write command that includes a matching slave address, the DA9063L can trigger the assertion of nIRQ and an optional wakeup event (enabled via GPIO14_WEN). If the nIRQ assertion from interface access is enabled (E_GPI14), it should be masked as long as the HS 2-WIRE is in use. This nIRQ cannot be cleared via the HS 2-WIRE interface because every interface access triggers a re-assertion.

High speed mode at 3.4 MHz can be enabled either via master code or continuously via HS_IF_HSM, but it does not support slope control for minimum tfDA specification.

The 2-wire interface is open-drain, supporting multiple devices on a single line. The bus lines must be pulled high by external pull-up resistors ($2 \text{ k}\Omega$ to $20 \text{ k}\Omega$). The attached devices only drive the bus lines low by connecting them to ground. As a result, two devices cannot conflict if they drive the bus simultaneously. In standard/fast mode, the highest frequency of the bus is 400 kHz. The exact frequency can be determined by the application and does not have any relation to the DA9063L internal clock signals. The DA9063L follows the host clock speed within the described limitations and does not initiate any clock arbitration or slow down. Control TWOWIRE_TO enables an automatic interface RESET that is triggered when the clock signal ceases to toggle for >35 ms (compatible with SMBus T_{TIMEOUT}).

The interface supports operation compatible with Standard, Fast, Fast-Plus and High Speed modes of the I²C-bus specification Rev 03 (UM10204_3). Bus clear, in the case of the DATA signal being stuck low, is achieved after receiving 9 clock pulses. Operation in High Speed mode at 3.4 MHz requires a minimum interface supply voltage of 1.8 V and a mode change in order to enable spike suppression and slope control characteristics compatible with the I²C-bus specification. The high speed mode can be enabled on a transfer-by-transfer basis by sending the master code (0000 1XXX) at the beginning of the transfer. The DA9063L does not make use of clock stretching and delivers read data without additional delay up to 3.4 MHz.

Communication on the 2-wire bus always takes place between two devices, one acting as the master and the other as the slave. The DA9063L only operates as a slave. The 2-wire interface has direct access to two pages of the DA9063L register map (up to 256 addresses). The register at address zero on each page is used as a page control register (with the 2-wire bus ignoring the LSB of control REG_PAGE). Writing to the page control register changes the active page for all subsequent read/write operations unless an automatic return to page 0 was selected by asserting control REVERT. Unless REVERT was asserted after modifying the active page, a read-back of the page



control register is recommended to ensure that future data exchange is accessing the intended registers.

The DA9063L offers an alternative method to access register pages 2 and 3. These pages can be accessed directly by incrementing the device address by one (default read address 0xB2; write address 0xB3). This removes the need to write to the page register before access to pages 2 and 3, thus reducing the traffic on the 2-wire bus.

6.8.1.1 Details of the 2-wire Control Bus Protocol

All data is transmitted across the 2-bus in groups of 8 bits. To send a bit, the DATA line is driven at the intended state while the CLK is LOW (a low on DATA indicates a zero bit). Once the DATA has settled, the CLK line is brought high and then low. This pulse on CLK clocks the DATA bit into the receiver's shift register, see Figure 18.

A two byte serial protocol is used containing one byte for address and one byte data. Data and address transfer is MSB transmitted first for both read and write operations. Transmission begins with the START condition from the master while the bus is idle. It is initiated by a high-to-low transition on the DATA line while the CLK is in the high state (a STOP condition is indicated by a low-to-high transition on the DATA line while the CLK is in the high state).



Figure 18: Timing of START and STOP Condition

The 2-WIRE bus is monitored by the DA9063L for a valid slave address when the interface is enabled. It responds immediately when it receives its own slave address. This 'Acknowledge' is done by pulling the DATA line low during the following clock cycle (see the white blocks marked A in Figure 19 to Figure 23).

The protocol for a register write from master to slave consists of a start condition, a slave address with read/write bit and the 8-bit register address followed by 8 bits of data terminated by a STOP condition (all bytes responded by DA9063L with Acknowledge), as illustrated in Figure 19.

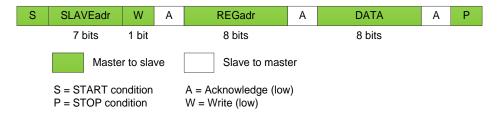


Figure 19: Byte Write (DATA Line)

When the host reads data from a register, it first has to write access the DA9063L with the target register address and then read access the DA9063L with a Repeated START or alternatively a



second START condition. After receiving the data, the host sends Not Acknowledge and terminates the transmission with a STOP condition (Figure 20).

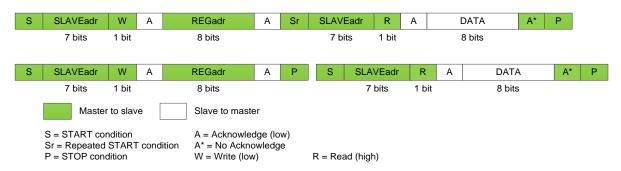


Figure 20: Examples of Byte Read (DATA Line)

Consecutive (page) read mode is initiated from the master by sending an Acknowledge instead of Not Acknowledge after receipt of the data word. The 2-wirecontrol block then increments the address pointer to the next 2-wireaddress and sends the data to the master. This enables an unlimited read of data bytes until the master sends a Not Acknowledge directly after the receipt of data, followed by a subsequent STOP condition. If a non-existent 2-wire address is read then the DA9063L returns code zero (Figure 21).



Figure 21: Examples of Page Read (DATA Line)

The slave address after the Repeated START condition must be the same as the previous slave address.

For enhanced data transfer efficiency, the DA9063L supports two write modes: Page Write mode and Repeated Write mode.

Page Write mode is used where the host has multiple bytes of data to be written to consecutive register addresses. It is selected by setting the WRITE MODE control to 0. For Page Write mode the master sends a device address followed by a register address then multiple data bytes. The 2-wire interface automatically increments the register address pointer after each data byte is received. The slave acknowledges each received byte of data until the master sends the STOP condition (Figure 22).



Figure 22: Page Write (DATA Line)



Repeated Write mode is used where the host has multiple bytes of data to be sent to non-consecutive registers. It is selected by setting the WRITE MODE control to 1. For Repeated Write mode the master sends a device address followed by multiple address-data pairs. The slave acknowledges each received byte until the master sends the STOP condition (Figure 23).



Figure 23: Repeated Write (DATA Line)



6.9 Voltage Regulators

Three types of low drop-out regulators (LDOs) are integrated on the DA9063L: for sensitive analog rails (for example, RF transceiver supply), the low noise regulators offer high PSRR across a wide frequency range; the LDOs provide an optimized PSRR and noise performance with lowest quiescent current. Quiescent current has been optimized for the always-on type regulators.

The regulators employ Dialog Semiconductor's Smart Mirror™ dynamic biasing that guarantees PSRR to be maintained across the full current range. Quiescent current consumption is dynamically adjusted to the load, which improves efficiency at light load conditions. Furthermore, Dialog Semiconductor's Smart Mirror™ technology allows the capacitor to be placed close to the load.

Note

When placing an LDO capacitor remotely from the DA9063L, the voltage drop (= load current * parasitic PCB impedance) needs to be considered when configuring the LDO output voltage.

Table 33: Regulator Control

Regulator	Туре	V _{out} Steps (mV)	Mode	Output Voltage (V)	Supplied Max. Current (mA)	Current Limit LDO/ Bypass (mA)	V _{OUT} Control	Notes
LDO3	Standard	20	Bypass	0.9 to 3.44	200	400/200	DVC variable slew rate	
LDO7	Standard	50	Bypass	0.9 to 3.6	200	400/300	V _{оит} programmable	
LDO8	Standard	50	Bypass	0.9 to 3.6	200	400/300	V _{OUT} programmable	Switching vibration motor driver, common supply with LDO7
LDO9	Low noise	50		0.95 to 3.6	200	400	V _{оит} programmable	
LDO11	Standard	50	Bypass	0.9 to 3.6	300	400/300	V _{out} programmable	
LDOCOR E	Always-on			2.5 ±2% accuracy	4		V _{OUT} non- programmable	Internal LDO

6.9.1 Regulators Controlled by Software

The regulators can be programmed via the power manager interface. All regulators can be enabled or disabled by a write command to the enable bit LDOxx_EN. Each LDO has two voltage registers for output voltage A and B. The appropriate values are stored in the registers VLDOxx_A and VLDOxx_B. The specific output voltage is selected with the bit VLDOxx_SEL. Changes to this control result in immediate output voltage changes on non-DVC regulators and ramped voltage transitions on DVC-enabled regulators. The output voltage can also be changed by directly re-programming the voltage control register. The sequencer also uses these registers and may write to them: their contents can therefore be found to differ from previous write commands.



For security reasons, the re-programming of registers that may cause damage when being incorrectly programmed (for example, voltage settings) can be disabled with control V_LOCK. This disables write access to registers with an address higher than 0x7F.

6.9.2 Regulators Controlled by Hardware

All regulators can be enabled or disabled under hardware control using GPIO1, 2, or 13. The GPIO port used is defined in register LDOxx_GPI. The output voltages can be switched by the GPIO port between the A and B voltage. The specific GPIO port is defined in register VLDOxx_GPI. After detecting a rising or falling edge at the GPI, the DA9063L configures the related regulators with the status of GPI1, GPI2, or GPI13 (the event bit E_GPI1, E_GPI2, or E_GPI13 is automatically cleared). A parallel write access to the regulator control registers is delayed and later overrides the hardware configuration. The sequencer does not affect regulators controlled via GPIOs.

6.9.3 Power Sequencer Control of LDOs

The power sequencer can control the LDOs. The specific time slot of each LDO is defined with bit LDOx_STEP in register bank starting at address 0x83. The sequencer enables and disables each LDO individually depending on the setting of each LDO's bit LDOx_CONF and LDOx_AUTO. To limit the inrush current, it is recommended to enable a maximum of one regulator (including bucks) per time slot.

If the control OTPREAD_EN is set, the regulator control registers are reloaded from OTP before leaving POWERDOWN mode. During power-up, the sequencer always takes output voltage A (defined in register VLDOxx A). Therefore, it also clears all VLDOxx SEL bits.

When powering down, the sequencer disables all LDOs, but the LDOs can be configured to remain on by setting bit LDOxx_CONF. In this case the output voltage B is always selected (value programmed in register VLDOxx_B). The related bit VLDOxx_SEL is set by the sequencer accordingly.

Table 34: LDO Power Sequence Voltage

LDO Output Voltage During Power-Up Sequencing										
LDOx_CONF	LDOx_AUTO	LDO	Output Voltage							
-	1	gets enabled	Α							
1 0		gets enabled	Α							
0		disabled	-							
LDO Output Voltage Du	ring Power-Down Sequenc	ing								
LDOx_CONF	LDOx_AUTO	LDO	Output Voltage							
1	-	remains enabled	В							
0		gets disabled	-							

The bit DEF_SUPPLY defines the sequencer action for time slot 0. If Bit DEF_SUPPLY is set, all LDOs configured to time slot 0 are enabled or disabled during power-up according to Table 34. If bit DEF_SUPPLY is not set, the LDOs configured to time slot 0 are disabled.

Note

When control bit LDOxx_SL_B is asserted, the LDO enters a forced sleep mode with the lowest quiescent current, but with a reduced maximum output current. The maximum current is reduced because a smaller



output driver is used (a partial pass device). Asserting LDOxx_SL_A results in the same forced sleep mode for an LDO when using the type A voltage register. Before wakeup from POWERDOWN mode (processing time slots from domain SYSTEM), the sequencer can configure all regulators with default voltage values from OTP: this allows any previously altered VLDOxx_A and LDOxx_SL_A settings to be reset.

Entering RESET mode automatically disables all regulators.

6.9.4 Dynamic Voltage Control

LDO3 includes DVC:

- The output voltage can be programmed in 20 mV steps.
- If the feedback signal GP_FB1 is configured to be READY (by asserting PM_FB1_PIN), this port
 is asserted while slewing and asserts E_DVC_RDY after all voltage and buck regulators have
 completed slewing.

DVC voltage transitions are handled by the following registers:

- Output voltage setting registers VLDO3_A/VLDO3_B.
 When writing into a selected voltage control register the output voltage is immediately ramped to the new value. When writing into the non-selected voltage register the ramping is delayed until this register is selected by toggling VLDO3_SEL.
- The voltage selection registers VLDO3_SEL activate a pre-configured transition to the alternate output voltage. These controls have been grouped together in registers DVC_1 and DVC_2 to better enable synchronized ramping of supply voltages.
- The DVC slew rate for all DVC-enabled regulators can be configured as 10 mV per (0.5, 1.0, 2.0, or 4.0) μs via control SLEW_RATE. Under light load conditions (< 10 mA), the slew rate is less than the programmed value when the output is close to the start and end of the slope This is especially the case for the fastest slew rate settings. The negative slew rate is load dependent and might be lower than the one mentioned above.</p>

6.9.5 Pull-Down Resistor

All LDOs have a pull-down resistor at the output when they are disabled. The pull-down resistor can be disabled with bit LDOxx_PD_DIS, and is required when LDOs are used in parallel with another supply. Otherwise the output is pulled to GND.

If an over-voltage occurs (LDO3: $V_{OUT} > 109$ % of nominal V_{OUT} , LDO7 to 9, and 11: $V_{OUT} > 106$ % of nominal V_{OUT}), the voltage regulators enable an internal load to discharge the output back to its configured voltage. This can be disabled via LDOxx_PD_DIS.



6.9.6 Bypass Mode and Current Limit

All LDOs feature a current limiting function. For LDOs with a bypass mode (LDO3, 7, 8, and 11), an over-current is indicated with an interrupt. When at least one of these LDOs reaches the current limit for more than 10 ms, an interrupt is raised to the host (during POWERDOWN mode a wakeup sequence is initiated) and the event bit E_LDO_LIM is set. The interrupt IRQ can be suppressed via the mask bit M_LDO_LIM.

If the current limit condition persists for more than 200 ms (indicating a probable short circuit condition), the related LDO is disabled and its LDOx_EN bit is de-asserted. The LDO remains disabled until a new enable occurs (via hardware or software activation). The automatic shutdown of the LDO can be disabled via bit LDO_SD. The host processor can distinguish if the IRQ is related to a temporary over-current or to a permanent shutdown by polling the related bit LDOx_ILIM or checking LDO3 EN, LDO7 EN, LDO8 EN, and LDO11 EN.

If the current limit is hit for more than 10 ms but less than 200 ms, the IRQ is generated but the related LDO is not disabled. If the current limit is hit for more than 200 ms and the involved LDO is shut down, the LDO<x>_EN bit is de-asserted. If the over-current spike has stopped before the host is able to read the xx LIM bits, the LDO that has been in current limit cannot be evaluated.

Changing from LDO to bypass mode and back triggers a change of the output voltage with some over/undershoot during the transition phase.

The LDO accuracy specification is not applicable if the LDO has been configured for bypass mode.

For LDOs operated in bypass mode it is recommended to configure the voltage in the VLDO<x>_A register to the expected output voltage of the LDO. When using the LDO above the normal operating range, it is recommended to set the VLDO<x>_A to 0x40 for LDO 3 and to 0x20 for LDO 7, 8, and 11.

6.9.7 LDO Supply from Buck Converter

The LDOs can optionally be supplied from a buck output (VDD < 2.8 V). In this mode some specification parameters change:

- at VDD = 1.8 V, the dropout voltage at I_{max} increases by 70 %
- for a supply voltage less than 1.8 V, the LDO dropout voltage is valid only for 1/3 of the standard I_{max} and the current capability decreases with the provided supply voltage.

LDO11 may be supplied from a rail higher than VSYS/CHG_WAKE (for example, the output of a 5 V boost) as long as $V_{DD} < 5.5 \text{ V}$.

6.9.8 LDO Sleep Mode for Reduced I_{OUT}

If the required output current is < 10 % of I_{MAX}, the quiescent power can be reduced by setting an LDO into sleep mode. In this mode, the output driver current capability is reduced to 10 % of I_{MAX}. Sleep mode can be set independently for the output voltage A and B by setting bit LDOxx_SL_A or bit LDOxx_SL_B. During LDO sleep mode, the over-current limit of the LDOs with a bypass function (LDO3, 7, 8, and 11) is reduced to 50 %. As a benefit of Dialog Semiconductor's Smart Mirror™ technology, sleep mode is typically not required because the quiescent current taken by the regulator is automatically minimized when operating at low current demands.



6.9.9 Vibration Motor Driver

LDO8 provides a third mode dedicated to drive vibration motors selected via bit LDO8_MODE. In this mode, the voltage regulation circuitry is disabled and no external stabilization capacitor is needed. In comparison to LDO mode, the PWM control is more efficient and allows an instant on and off for the vibrator signal.

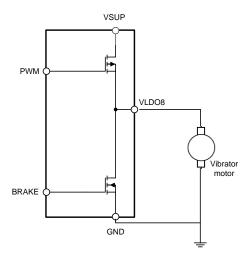


Figure 24: Vibration Motor Driver

The vibrator motor driver is a half-bridge PWM-controlled motor driver, with an automatic battery supply correction of the PWM duty cycle (Figure 24). The PWM base frequency can be selected by PWM_CLK to be either 1.0 MHz or 2.0 MHz (resulting in a PWM repetition rate of 15.6 kHz or 31.25 kHz). The vibration motor speed is determined by the effective output voltage which is set via control VIB_SET (6 bits giving 64 programmable speeds). Setting the output voltage to 0 turns on a braking NMOS transistor to stop the vibration motor immediately.

The motor can also be stopped and started by a level on port nVIB_BRAKE, if enabled via bit PM_FB3_PIN.

The PWM duty cycle is corrected automatically before it is enabled and after each breaking period. It is also automatically corrected every 10 s when it is running for longer periods. These corrections are done via autonomous V_{SYS} measurement via the internal GPADC (overrides control setting of AUTO_VSYS_EN). The duty cycle, D, is given by D = VIB_SET / V_{SYS} .

Note

The half-bridge driver transistors have an internal current limit of approximately 400 mA

6.9.10 Core Regulator LDOCORE

The LDOCORE is a 2.5 V supply dedicated for the internal logic of DA9063L. It is used for running the state machine, GPIO pins with comparators, bias, reference, GPADC, OTP, and power manager registers. It is supplied from internal rail VDDREF, powered from either CHG_WAKE or VSYS. When LDOCORE is supplied in RESET mode, its output voltage is temporarily reduced to 2.2 V. In general, LDOCORE is an always-on supply (remaining enabled during RESET mode), but for lowest dissipation power LDOCORE can also be disabled when progressing towards DELIVERY mode.



6.10 DC/DC Buck Converters

DA9063L includes six DC/DC buck converters with DVC.

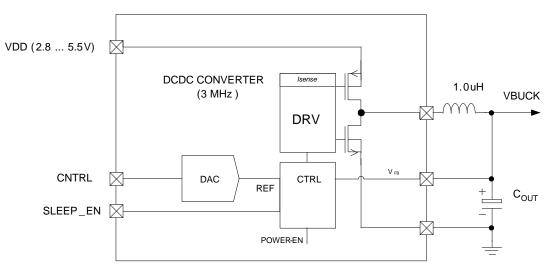


Figure 25: DC-DC Buck Converter

The converters are high efficiency synchronous step-down regulators, operating at a high frequency (3 MHz), supplying individual output voltages with \pm 3 % accuracy. The default output voltage is loaded from OTP and can be set in 10 mV steps. To limit in-rush current from V_{SYS}, the buck converters perform a soft-start for up to 3 ms, when enabled via control SOFT_START. During this 3 ms period, the output current of the buck is limited.

The DVC controller allows the following features:

- The buck converter output voltage is programmable over the power manager bus in 10 mV steps.
- If the feedback port GP_FB1 is configured as READY, this port is asserted while slewing and asserts E_DVC_RDY after all voltage and buck regulators have stopped slewing.
- Output voltages below 0.7 V are only supported in Pulse Frequency Modulation (PFM) mode.
 During a voltage reduction below 0.7 V, the slew rate control ends at 0.7 V and the buck mode is automatically changed to PFM mode.

The DVC control is handled by the following registers:

- Output voltage setting register VBxxxx_A/VBxxxx_B.
 When writing to the voltage control register that is in use by an enabled buck, the output immediately ramps to the new setting. When writing to the voltage control register that is not in use, the ramping is delayed until this register is selected by toggling VBxxxx_SEL.
- The voltage register selection VBxxxx_SEL.
 This activates a pre-configured transition to the alternate output voltage. These controls are grouped into registers DVC_1 and DVC_2 to better enable synchronized ramping of supply voltages.
- The DVC slew rate is programmable as at 10 mV per (4, 2, 1, or 0.5) µs via control SLEW_RATE.
 During PFM mode, the negative slew rate is load-dependent and might be lower than the programmed rate.



The supply current during PWM (synchronous) operation is in the order of 3.5 mA (quiescent current and charge/discharge current) and drops to <1 μ A in shutdown. Switching frequency is chosen to be high enough (3 MHz) to allow the use of a small 1.0 μ H inductor.

The operating mode of the buck converter is selected via the buck control register bits B<x>_MODE. The buck converter can be forced to operate in either PWM or PFM mode. Additionally, the buck converter has an automatic mode where it switches between PWM and PFM modes depending on the load current.

The switching converters can be enabled/disabled/configured via the power manager and HS 2-WIRE interface. Writing to Bxxxx_EN/VBxxxx_SEL unconditionally configures the regulator to the selected mode (enabled/disabled). Reading Bxxxx_EN/VBxxxx_SEL provides the actual state, which may differ from a previous write (in the case where the regulator state is changed from GPIO or power sequencer control). All bucks can be controlled with an ID from the power sequencer. If enabled in DEF_SUPPLY, supplies can be configured to default settings when the sequencer passes slot 0.

To limit the inrush current, it is recommended to select individual regulators (including LDOs) only with xxxx_DEF settings.

When powering up, the power sequencer clears VBxxxx_SEL for a buck when it has an ID pointing to the time slot being processed. This forces the regulator to ramp the output voltage to the value programmed inside the related register VBxxxx_A.

When powering down (for example, to POWERDOWN mode), sequencer-controlled supplies are usually disabled but can be configured to remain on by setting Bxxxx_CONF. In the latter case, the sequencer sets VBxxxx_SEL so that the regulator output voltage is ramped to the value programmed inside the related register VBxxxx_B. Disabled bucks can switch off their pull-down resistor. Before wakeup from POWERDOWN mode (processing time slots from domain SYSTEM), the sequencer can configure the bucks with default voltage values from OTP and reset any changed VBxxxx_A settings.

All buck converters provide an optional hardware enable/disable via GPIO1, 2, and 13. A regulator that has to be enabled/disabled from a GPI port selects this feature via its control Bxxxx_GPI. A change of the output voltage from the state of a GPI is enabled via control VBxxxx_GPI. After detecting a rising or falling edge at the GPI, the DA9063L configures the enabled regulators with the status of GPI1, GPI2, or GPI13 (the event bit E_GPI1, E_GPI2 or E_GPI13 is automatically cleared). A parallel write access to the regulator control registers is delayed and later overrides the HW configuration. The sequencer does not change regulator settings enabled for GPI control. Powering down to RESET mode automatically disables all buck converters. When the output of a buck converter is combined with a parallel low power LDO, its pull-down resistor needs to be disabled via Bxxxx PD DIS. Otherwise its output is discharged to GND when being disabled.

To allow DVC transitions under load, the buck current limit should be configured at least 40% higher than the required maximum continuous output current. See Table 35 as a guide to determining this limit.

Table 35: Selection of Buck Current Limit from Coil Parameters

Min. ISAT (mA)	Frequency (MHz)	Buck Current Limit (mA)	Max. Output Current (mA)	
3800	3	3400	2400	
3100	3	2800	2000	
2400	3	2100	1500	



Min. ISAT (mA)	Frequency (MHz)	Buck Current Limit (mA)	Max. Output Current (mA)	
1700	3	1700	1200	

To ensure correct regulation, the buck converters require the supply voltage to be 0.7 V higher than the output voltage. As this is not always possible at higher output voltage settings, the converters BUCKMEM, BUCKIO, and BUCKPERI provide a follower mode where the electrical characteristics of the DC-DC converter no longer apply, but instead the PMOS output driver is fully-on and the output voltage simply follows the dropping input voltage. There will be a voltage drop between the buck VDD supply and the output which results from the on-resistance of the buck PMOS driver and the coil, with the voltage drop magnitude being depending on load current. Bucks running in follower mode will temporarily stop switching and by that process will generate PWM mode 3 MHz subharmonics.

6.10.1 BUCKCORE1, BUCKCORE2, and BUCKPRO

BUCKCORE1, BUCKCORE2, and BUCKPRO include a full-current (previously overdrive) mode, individually enabled via control BCORE1_OD, BCORE2_OD, and BPRO_OD.

In full-current mode:

- The maximum current capability is 2500 mA
- The selected current limits are automatically doubled
- The quiescent current increases due to the increased switching losses

For full-current mode, the application requires two 47 µF output capacitors and an appropriate inductor that can sustain higher currents without heating up or suffering from inductance degradation.

BUCKCORE1 and 2 can also be merged as a dual-phase BUCKCORE with up to 5000 mA maximum output current. If enabled in OTP via BCORE_MERGE, the register controls of BUCKCORE2 (except BCORE2_PD_DIS) are automatically disabled and the output from both coils must be routed together. The feedback signal for both phases is taken from the sense node switch matrix of BUCKCORE1 (the VBUCKCORE2 pin may be left floating if the internal pull-down resistor is enabled by setting BCORE2_PD_DIS = 0). With BCORE1_FB programmed in OTP to 0b000, a differential remote sensing at the point-of-load can be enabled, using VBUCKCORE2 as a GND sense port. In this mode, the BUCKCORE output capacitor voltage has to be routed to port CORE_SWS or GP_FB_2 (selected via control MERGE_SENSE). Depending on the settings of BCORE1_OD, the dual-phase buck provides a maximum 2500 mA or 5000 mA, requiring two or four 47 µF output capacitors, respectively.

BUCKCORE2 always runs on the inverted clock (anti-phase) of BUCKCORE1. The switching node output of both phases must be connected symmetrically on the PCB (with matched routing inductances and resistances).



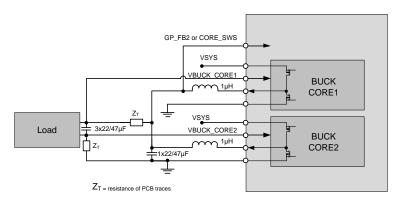


Figure 26: BUCKCORE1 and BUCKCORE2 in Dual-Phase Remote Sense Mode

6.10.2 BUCKPRO in DDR Memory Bus Termination Mode

If enabled via BPRO_VTT_EN, BUCKPRO offers an alternative mode to provide VTT bus termination for DDR memory. In this mode, its output voltage tracks 50 % of the VDDQ sense port voltage (Figure 27). In this mode, BUCKPRO must be set to sync mode either by the host or by OTP configuration. If enabled via BPRO_VTTR_EN, a second VTTR output provides the same voltage for a DDR VTTR reference rail, buffered with ± 10 mA source/sink capability (requires 0.1 µF stabilization capacitor). With BPRO_VTTR_EN being asserted in combination with BPRO_VTT_EN released, the DA9063L provides a VTTR reference buffer with BUCKPRO running in a normal output voltage control mode. If memory termination is not required (BPRO_VTTR_EN = 0), port VDDQ provides the state of event E GPI2 and port VTTR provides the state of the 1.2 V comparator.

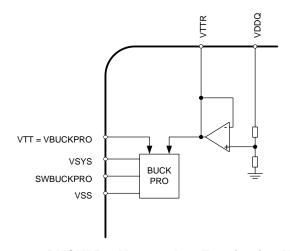


Figure 27: BUCKPRO Memory Bus Termination Mode



6.10.3 BUCKMEM and BUCKIO in Merged Mode

The converter BUCKMEM can be merged with BUCKIO via control BUCK_MERGE to form a single DC-DC converter with a maximum output current of 3000 mA (Figure 28). The routing of the switcher output pins to the common inductor must be symmetrical. The VBUCKIO feedback pin may be left floating in merged mode if its internal pull-down resistor is enabled by setting BIO_PD_DIS = 0. The inductor (1.0 μ H) and the output capacitor have to be selected according to the increased output current configuration controls of BUCKIO are disabled by asserting the bit BUCK_MERGE; the selected current limits of BUCKMEM are automatically doubled.

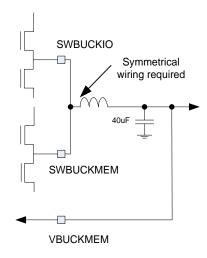


Figure 28: BUCKMEM Merged With BUCKIO



6.11 General Purpose ADC

6.11.1 ADC Overview

The Analog to Digital Converter (ADC) uses a sample and hold successive approximation switched capacitor architecture. It is supplied from VDDCORE (2.5 V). Configured via control ADC_MODE, it can be used either in high-speed mode with measurement sequences repeated every 1 ms or in economy mode with sequences repeated every 10 ms.

6.11.2 ADC Input MUX

The DA9063L provides an ADC with 10-bit resolution and track and hold circuitry combined with an analog input multiplexer (Figure 29). The analog input multiplexer allows conversion of up to nine different inputs. The track and hold circuit ensures stable input voltages at the input of the ADC during the conversion.

The ADC is used to measure the following inputs:

- Channel 0: VSYS_RES measurement of the system VDD (2.5 V to 5.5 V)
- Channel 1: ADCIN1_RES high impedance input (0 to 2.5 V)
- Channel 2: ADCIN2 RES high impedance input (0 to 2.5 V)
- Channel 3: ADCIN3_RES high impedance input (0 to 2.5 V)
- Channel 4: T_J measurement of internal temperature sensor
- Channel 5: Reserved
- Channel 8: MON_A8_RES group 1 internal regulators voltage (0 to 5.0 V)
- Channel 9: MON A9 RES group 2 internal regulators voltage (0 to 5.0 V)
- Channel 10: MON_A10_RES group 3 internal regulators voltage (0 to 5.0 V)

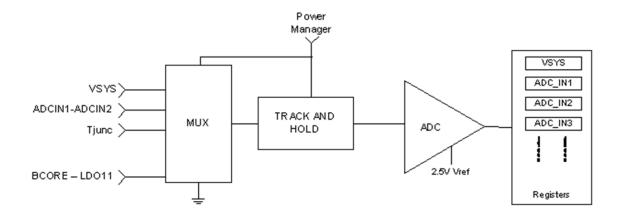


Figure 29: ADC Block Diagram

The MUX selects from and isolates the nine inputs, and presents the channel to be measured to the ADC input. When selected, an input amplifier on the V_{SYS} channel subtracts the VDDCORE reference voltage and scales the signal to the correct value for the ADC.



6.11.3 Manual Conversion Mode

Manual measurements by the ADC are initiated by an ADC_MAN bit register write. The ADC powers up, one conversion is done on the channel specified by ADC_MUX and the 10-bit result is stored in the ADC_RES_H and ADC_RES_L registers. After the conversion is completed, the ADC powers down again, ADC_MAN bit is reset and an IRQ event flag is set (E_ADC_RDY). The generation of this IRQ can be masked by the IRQ mask M_ADC_RDY.

6.11.4 Automatic Measurements Scheduler

The automatic measurement scheduler allows monitoring of the system voltage V_{SYS} , the auxiliary channels ADCIN1 to 3 and the output voltage supervision of embedded regulators. The results are automatically compared with upper and lower thresholds set by power manager registers to give an nIRQ event if a measurement is outside these levels. All measurements are handled by the scheduler system detailed below.

The scheduler performs a sequence of 10 slots continually repeated according to the configured mode. A slot requires 100 μ s. The pattern of measurements over the 10 slots depends upon the enabled automatic measurements. Additional manual measurement opportunities are available in slots where automatic measurements have been disabled by control bits in ADC_CONT. Automatic measurements only store the eight MSBs of the ADC measurement.

Figure 30 shows (with typical configurations) how the different measurements are scheduled.

	Example sequence of AUTO-ADC measurements										
Slot N <u>o</u>	0	1	2	3	4	5	6	7	8	9	
	A0	A8	М	A1	A9	М	A2	A10	М	А3	

Each Slot allows 1 automatic or manual measurement to be made

A0 - Automatic measurement of VSYS (mux channel 0)

A1 - Automatic measurement of ADCIN1 (mux channel 1)

A2 - Automatic measurement of ADCIN2 (mux channel 2)

A3 - Automatic measurement of ADCIN3 (mux channel 3)

A8 - Automatic measurement of multiplexed regulator output voltages (mux channel 8)

A9 - Automatic measurement of multiplexed regulator output voltages (mux channel 9)

A10 - Automatic measurement of multiplexed regulator output voltages (mux channel 10)

M indicates time slots when a Manual measurement can be made

Figure 30: ADC Sequence



6.11.4.1 A0: V_{SYS} Voltage nIRQ Measurement Mode

 V_{SYS} is measured, stored in $VSYS_RES$ and compared with the $VSYS_MON$ threshold. If the result of the comparison is different to its previous state (being either lower or higher) for three consecutive readings, an E_VDD_MON event is generated. Glitches of a duration less than three consecutive measurements do not update the state; events are triggered at rising and falling edges of the state signal. This multiple reading debounces the V_{SYS} voltage before issuing an nIRQ. After the nIRQ assertion, the automatic measurement of channel V_{SYS} is paused for reading. The host must clear the associated event flag to re-enable the supervision of V_{SYS} . The event-causing value is kept in the result register.

If selected via GPIO12_PIN, the debounced comparator state can be indicated via the GPO12 port, representing a power good signal that can be used, for example, to trigger boot activities on external ICs. If no action is taken to restore the V_{SYS} voltage (that is, discharging of the battery continues), the host may consider switching off optional always-on blocks to save energy later on. V_{SYS} measurements are enabled via control AUTO_VSYS_EN.

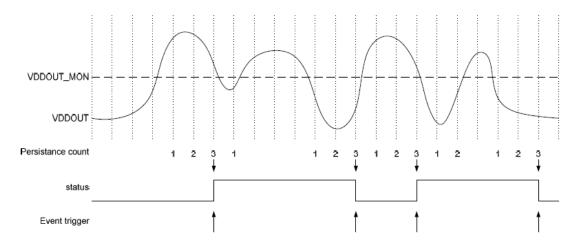


Figure 31: V_{SYS} Monitor Persistence Behavior



6.11.4.2 A1, A2, A3: Automatic Measurement and High/Low Threshold Warning nIRQ Mode

The automatic measurement result of channel ADC_IN1 is stored in the ADCIN1_RES register. If a reading of A1 is less than AUTO1_LOW or greater than AUTO1_HIGH, then the event flag E_GPI0 is set. If nIRQ is asserted, the automatic measurement of channel ADC_IN1 is paused until the host has cleared the associated event flag (the event-causing value is kept in the result register). The assertion of nIRQ can be masked by IRQ mask M_GPI0, which also disables the pausing of automatic measurements. If debouncing is selected via ADCIN1_DEB the event is only asserted if two consecutive measurements override the same threshold. The automatic measurement is enabled by register AUTO_AD1_EN. In addition, it is possible to use ADCIN_1 with a 1 μ A to 40 μ A current source that allows automatic measurement of a resistor value (programmed via ADCIN1_CUR). The current source is enabled by AD1_ISRC_EN. During automatic measurements the enabled current source is dynamically switched off at the end of the conversion and switched on one slot prior to the next ADCIN_1 measurement (to enable minimum current consumption, and allow any external capacitance voltage to settle); otherwise its status is static.

A similar functionality is available at ADC_IN2 and ADC_IN3. ADC_IN2 provides notification to the processor via a fixed voltage comparator (also available when ADC is powered down) but the ADCIN2 current source is static (no dynamic switch-off at the end of automatic conversion). The input selection switch of ADC_IN2 provides an enhanced isolation (80 dB typ.) between the externally-connected circuit and the internal ADC block (for example, allowing the DC supervision of noise sensitive audio lines).

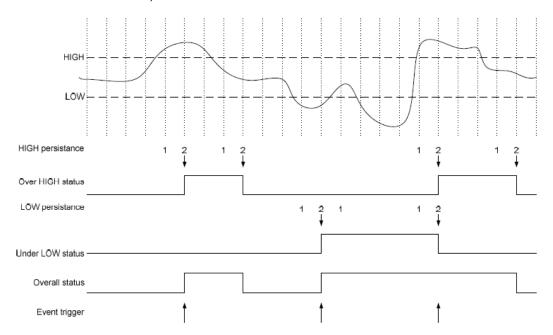


Figure 32: A1, A2, A3 Persistence Behavior



6.11.4.3 A8, A9, A10: Automatic Regulator Monitor with Under- or Over-Voltage Shutdown

DA9063L provides the capability to monitor the output voltage of internal regulators. In case of a catastrophic failure, the related regulator is disabled. This feature is enabled using control MON_MODE. Each internal regulator is assigned to a GPADC input channel and can be individually enabled for monitoring via controls BCORE1_MON_EN to LDO11_MON_EN. For example, if ADC channel A8 is connected to an enabled regulator and it is also enabled for monitoring, the ADC automatically measures the regulator output voltage every 1 ms (10 ms in ADC economy mode). Simultaneously, two relative thresholds are calculated from the regulator nominal output voltage (supporting DVC). If the regulator is out of range, the voltage measurement is stored inside MON_A8_RES, the regulator ID is recorded inside the index MON_A8_IDX and the event flag E_REG_UVOV is set. To secure a stable output voltage, the monitoring is delayed after regulators are switched on and when changing their voltage level to a new target (that is, during DVC slewing, after writing into the active regulator voltage register, or when the active voltage control register is changing between Vxxx_A and Vxxx_B). This delay is programmable using control UVOV_DELAY.

If debouncing is selected via MON_DEB, an event is only created if two consecutive measurements exceed the same threshold. The feature is also available for channels A9 and A10.

ADC Channel	Regulator	Enable		
A8	BUCKCORE1 BUCKCORE2 BUCKPRO LDO3 LDO11	BCORE1_MON_EN BCORE2_MON_EN BPRO_MON_EN LDO3_MON_EN LDO11_MON_EN		
A9	BUCKIO BUCKMEM BUCKPERI	BIO_MON_EN BMEM_MON_EN BPERI_MON_EN		
A10	LDO7 LDO8 LDO9	LDO7_MON_EN LDO8_MON_EN LDO9_MON_EN		

Table 36: Assignment of Regulators for Voltage Monitoring

If more than one regulator is assigned to A8, A9, or A10, the regulator measurements are sequentially multiplexed on this channel (elongates the measurement period of 1 ms or 10 ms by each additional assigned regulator). In the case of an E_REG_UVOV event, the regulator monitoring continues and shuts down rails facing catastrophic failure (voltage and IDX controls contain information for determining the cause of the most recently detected under- or over-voltage). When clearing E_REG_UVOV via a host write, the related regulator index controls are reset. In the unlikely case in which several regulators from a sequenced ADC channel have triggered an under- or over-voltage condition before the host was able to read the related controls, all monitored supplies may be checked for their actual state (enabled or shutdown) to detect if further supplies have also been shut down in addition to the one captured by the index control.

The assertion of nIRQ can be masked by IRQ mask M_REG_UVOV. With a masked nIRQ from regulator supervision, the setting of MON_RES determines whether an out-of-range detection disables the related regulator or triggers the assertion of port nRESET. In the latter case, nRESET is asserted for 1 ms and continues to be asserted until the regulator returns to being in range (regulators that are not enabled but are selected for monitoring do not trigger the assertion of nRESET). With a masked nIRQ, GP_FB2 can be configured to flag PWR_OK, indicating that all monitored regulator voltages are in-range. Selecting disabled regulators for monitoring suppresses the assertion of PWR_OK.



The regulator monitor unit provides alternative modes selected via MON MODE. In measurementonly mode, the event flag E REG UVOV is set for every automatic measurement result being available on either A8, A9, or A10 (with a maximum of one regulator per channel being enabled for measurements). When enabled, an auto-measurement on A8, A9, or A10 allows the host to get its actual output voltage from registers MON_A8_RES to MON_A10_RES. When multiple ADC channels are enabled for automatic regulator voltage measurements, the burst measurement mode may reduce the control interface traffic and number of nIRQ assertions. When the ADC time slot of A10 has finished, the event flag E REG UVOV is set (independent of A10 being enabled for automeasurements). The host can then read the measurement results stored inside MON A8 RES -MON A10 RES as a block. During measurement modes, there are no threshold comparisons or regulator shutdowns. If the nIRQ line was asserted, automatic measurements on channels A8 (A9 and A10) are paused until the host has cleared the associated event flag (the event-causing value is kept in the result registers). During measurement modes, E REG UVOV is cleared by writing the read value into register EVENT_B. To sequentially measure other regulators on ADC input channels A8, A9, and A10, the host has to set the monitor enable for the next measurement slot to the required regulator before clearing the event. The assertion of nIRQ can be masked by IRQ mask M REG UVOV, which also disables the pausing of automatic measurements. For further information about voltage monitoring, please see DA9063 Voltage Monitoring [2].

Note

Voltage monitoring function cannot be used for any LDO in bypass mode.

6.11.4.4 A4: Manual Measurement T_J

The 10-bit result of manual measurements is stored in the registers ADC_RES_L and ADC_RES_H. Channel 4 (T_J) is used to measure the output of the internal temperature sensor (generated from a proportional to absolute temperature (PTAT) current using a bandgap reference circuit). The ADC measurement result and the T_OFFSET value can be used by the host to calculate the internal junction temperature, defined by the following formula:

$$T_J$$
 [°C] = -0.398 * (ADC - T_OFFSET) + 330

Manual measurements on A8 to A10 are possible, but require disabling the automatic measurements on these channels and also ensuring that only one regulator is connected to each of these ADC channels.

NOTE

The T_OFFSET value is stored in the T_OFFSET register at address 0x104 during manufacture.

6.11.5 Fixed Threshold Comparator

A comparator with a threshold of 1.2 V is connected to the input of ADC channel 2. The comparator is asserted when the input voltage exceeds or drops below 1.2 V for at least 10 ms (debouncing). After being enabled via COMP1V2_EN, the status flag COMP1V2 indicates the actual state and a maskable interrupt request E_COMP1V2 is generated at the falling and rising edge state transitions. The comparator may be disabled via COMP1V2_EN when auto-measurements with high resolution are executed on ADCIN2.



6.12 Adjustable Frequency Internal Oscillator

An internal oscillator provides a nominal 6.0 MHz clock that is divided down to 3.0 MHz for the buck converters. It is divided down to 2.0 MHz to control the digital core, timers, PWM units, charge pump and ADC. The frequency of the internal oscillator is adjusted during the initial start-up sequence of the DA9063L to within 5 % of the nominal 6.0 MHz. It can be adjusted further within ±10 % via register control OSC_FRQ. The tolerance of this frequency affects most absolute timer values and PWM repetition rates (for example, LED and vibrator mode drivers) of the DA9063L.

6.13 Reference Voltage Generation: VREF, VLNREF

The DA9063L includes a temperature-independent voltage reference circuit which is derived from an internal band-gap reference and OTP-trimmed buffer amplifier. The output voltage on VREF is trimmed to 1.2 V and the reference is decoupled by an external capacitor on the VREF pin. A lower voltage instance of VREF is provided at VLNREF (0.9 V) and used for the LDOs. These pins must not be loaded. The IREF pin provides the internally used accurate current bias and requires an external 200 k Ω precision resistor.

6.14 Thermal Supervision

The application must ensure that the DA9063L junction temperature does not exceed 125 °C. To protect the DA9063L from damage due to excessive power dissipation, the internal temperature is continuously monitored. Whenever the junction temperature is higher than TEMP_WARN = 125 °C, an E_TEMP event is asserted and an IRQ is generated for the host. If this occurs during POWERDOWN mode, a wakeup is triggered.

The host may then check the exact junction temperature by a manual measurement on GPADC channel 4. An 8-bit OTP register (T_OFFSET) can be used to store its offset at a known temperature (for example 50 °C) to improve the absolute accuracy, which should then be ± 7 °C of the measured silicon die junction temperature. This T_OFFSET can be used by the host to calculate the absolute die temperature.

The absolute die junction temperature can be calculated by the host using the result from the ADC channel 4 measurement result and the T_OFFSET trim values.

When the junction temperature exceeds 125 °C, it is recommended to shut down optional functions of the application allowing the DA9063L to cool. When the junction temperature increases further, exceeding TEMP_CRIT = 140 °C, the fault flag TEMP_CRIT is asserted in the FAULT_LOG register and the DA9063L immediately shuts down to RESET mode. The fault condition remains as long as the junction temperature is higher than TEMP_WARN. The TEMP_CRIT flag can be evaluated by the application after the next power up. Whenever the junction temperature exceeds TEMP_POR = 150 °C, a POR to the digital core is immediately asserted and this stops all functions of the DA9063L. This is necessary to prevent the possibility of permanent device damage.

6.15 Main System Rail Voltage Supervision

The supervision of the system supply V_{SYS} is performed by two comparators. One monitors the under-voltage level VDD_FAULT_LOWER (fault condition indicator); the other, VDD_FAULT_UPPER, indicates a good (valid) battery/external supply. The high-to-low transition of the VDD_FAULT_UPPER signal is also used as a low system supply warning indicator, informing the host via event E_VDD_WARN (or the assertion of port nVDDFAULT) that the system rail may soon drop below the under-voltage threshold.

DA9063L



System PMIC for Mobile and Automotive Applications

The VDD_FAULT_LOWER threshold is OTP-configurable and can be set via the VDD_FAULT_ADJ control bits from 2.5 V to 3.25 V in steps of 50 mV. The VDD_FAULT_UPPER level is also OTP-configurable and can be set via the VDD_HYST_ADJ from 100 to 450 mV higher than the programmed VDD_FAULT_LOWER threshold.



7 Register Map

This section provides an overview of the registers. A description of each register is provided in 11.

Hex	Register	7	6	5	4	3	2	1	0	
					PAGE 0					
00	PAGE_CON	Revert	WRITE_MODE		Reserved			REG_PAGE		
Systen	n Control and Ev	ent Registers (S	(SMON)							
01	STATUS_A		Rese	rved		COMP1V2	DVC_BUSY	WAKE	nONKEY	
02	STATUS_B	GPI7	GPI6	GPI5	GPI4	GPI3	GPI2	GPI1	GPI0	
03	STATUS_C	GPI15	GPI14	GPI13	GPI12	GPI11	GPI10	GPI9	GPI8	
04	STATUS_D	LDO11_LIM	LDO8_LIM	LDO7_LIM	Reserved	LDO3_LIM		Reserved		
05	FAULT_LOG	WAIT_SHUT	nSHUTDOWN	KEY_RESET	TEMP_CRIT	VDD_START	VDD_FAULT	POR	TWD_ERROR	
06	EVENT_A	EVENTS_D	EVENTS_C	EVENTS_B	E_SEQ_RDY	E_ADC_RDY	E_TICK	E_ALARM	E_nONKEY	
07	EVENT_B	E_VDD_WARN	E_VDD_MON	E_DVC_RDY	E_REG_UVOV	E_LDO_LIM	E_COMP_1V2	E_TEMP	E_WAKE	
08	EVENT_C	E_GPI7	E_GPI6	E_GPI5	E_GPI4	E_GPI3	E_GPI2	E_GPI1	E_GPI0	
09	EVENT_D	E_GPI15	E_GPI14	E_GPI13	E_GPI12	E_GPI11	E_GPI10	E_GPI9	E_GPI8	
0A	IRQ_MASK_A		Reserved		M_SEQ_RDY	M_ADC_RDY	M_TICK	M_ALARM	M_nONKEY	
0B	IRQ_MASK_B	M_VDD_WARN	M_VDD_MON	M_DVC_RDY	M_REG_UVOV	M_LDO_LIM	M_COMP_1V2	M_TEMP	M_WAKE	
0C	IRQ_MASK_C	M_GPI7	M_GPI6	M_GPI5	M_GPI4	M_GPI3	M_GPI2	M_GPI1	M_GPI0	
0D	IRQ_MASK_D	M_GPI15	M_GPI14	M_GPI13	M_GPI12	M_GPI11	M_GPI10	M_GPI9	M_GPI8	
0E	CONTROL_A	CP_EN	M_POWER1_EN	M_POWER_EN	M_SYSTEM_E N	STANDBY	POWER1_EN	POWER_EN	SYSTEM_EN	
0F	CONTROL_B	BUCK_SLOWS TART	Reserved		nONKEY_LOCK	nRES_MODE	RESET_BLINKIN G	WATCHDOG_DIS	CHG_SEL	
10	CONTROL_C	DEF_SUPPLY	SLEW_RATE		OTPREAD_EN	AUTO_BOOT		DEBOUNCING		



Hex	Register	7	6	5	4	3	2	1	0	
11	CONTROL_D	BLIN	K_DUR		BLINK_FRQ			TWDSCALE		
12	CONTROL_E	V_LOCK	PM_FB3_PIN	PM_FB2_PIN	PM_FB1_PIN	ECO_MODE	Reserved	RTC_MODE_SD	RTC_MODE_PD	
13	CONTROL_F	Reserved	WAKE_UP	SHUTDOWN	WATCHDOG					
14	PD_DIS	PMCONT_DIS	OUT32K_PAUSE	BBAT_DIS	Reserved	HS2IF_DIS	PMIF_DIS	GPADC_PAUSE	GPI_DIS	
GPIO (Control Registers	(GPIO)								
15	GPIO0-1	GPIO1_WEN	GPIO1_TYPE	GPIO1	_P IN	GPIO0_WEN	GPIO0_TYPE	GPIO	0_P IN	
16	GPIO2-3	GPIO3_WEN	GPIO3_TYPE	GPIO3	B_P IN	GPIO2_WEN	GPIO2_TYPE	GP IO	2_P IN	
17	GPIO4-5	GPIO5_WEN	GPIO5_TYPE	GP IO	5_P IN	GPIO4_WEN	GPIO4_TYPE	GP IO	4_P IN	
18	GPIO6-7	GPIO7_WEN	GPIO7_TYPE	GPIO7	_P IN	GPIO6_WEN	GPIO6_TYPE	GPIO	GPIO6_P IN	
19	GPIO8-9	GPIO9_WEN	GPIO9_TYPE	GPIO9	GPIO9_PIN		GPIO8_TYPE	GPIO8_P IN		
1A	GPIO10-11	GPIO11_WEN	GPIO11_TYPE	GPIO1	1_P IN	GPIO10_WEN	GPIO10_TYPE	GPIO10_P IN		
1B	GPIO12-13	GPIO13_WEN	GPIO13_TYPE	GPIO1:	3_P IN	GPIO12_WEN	GPIO12_TYPE	GPIO1	2_P IN	
1C	GPIO14-15	GPIO15_WEN	GPIO15_TYPE	GPIO1	5_P IN	GPIO14_WEN	GPIO14_TYPE	GP IO14_P IN		
1D	GPIO_MODE0-7	GPIO7_MODE	GPIO6_MODE	GPIO5_MODE	GPIO4_MODE	GPIO3_MODE	GPIO2_MODE	GPIO1_MODE	GPIO0_MODE	
1E	GPIO_MODE8- 15	GPIO15_MODE	GPIO14_MODE	GPIO13_MODE	GPIO12_MODE	GPIO11_MODE	GPIO10_MODE	GPIO9_MODE	GPIO8_MODE	
1F	SWITCH_CONT	CP_EN_MODE	CORE_SW_INT		SWIT	CH_SR		CORE_S	SW_GP I	
Regula	ator Control Regi	sters (REG)								
20	BCORE2_CONT	Reserved	VBCOR	E2_GPI	Reserved	BCORE2_CONF	BCOR	E2_GPI	BCORE2_EN	
21	BCORE1_CONT	CORE_SW_CO	VBCORE1_GPI		CORE_SW_EN	BCORE1_CONF	BCOR	E1_GPI	BCORE1_EN	
22	BPRO_CONT	Reserved	VBPR	O_GPI	Reserved	BPRO_CONF	BPRO	D_GPI	BPRO_EN	
23	BMEM_CONT	Reserved	VBMEN	M_GPI	Reserved	BMEM_CONF	BME	M_GPI	BMEM_EN	
24	BIO_CONT	Reserved	VBIO	_GPI	Reserved	BIO_CONF	BIO	_GPI	BIO_EN	



Hex	Register	7	6	5	4	3	2	1	0	
25	BPERI_CONT	PERI_SW_CON F	BPER	I_GP I	PERI_SW_EN	BPERI_CONF	PERI_S	SW_GPI	BPERI_EN	
26	RESERVED	Reserved	Reserved		Reserved	Reserved	Reserved		Reserved	
27	RESERVED	Reserved	Rese	erved	Reserved	Reserved	Res	erved	Reserved	
28	LDO3_CONT	LDO3_CONF	VLDO	3_GPI	Reserved	LDO3_PD_DIS	LDO:	3_GPI	LDO3_EN	
29	RESERVED	Reserved	Rese	erved	Reserved	Reserved	Res	erved	Reserved	
2A	RESERVED	Reserved	Rese	erved	Reserved	Reserved	Res	erved	Reserved	
2B	RESERVED	Reserved	Rese	erved	Reserved	Reserved	Res	erved	Reserved	
2C	LDO7_CONT	LDO7_CONF	VLDO	7_GPI	VLDO7_SEL	LDO7_PD_DIS	LDO ⁻	7_GPI	LDO7_EN	
2D	LDO8_CONT	LDO8_CONF	VLDO	8_GPI	VLDO8_SEL	LDO8_PD_DIS	LDO8_GPI		LDO8_EN	
2E	LDO9_CONT	LDO9_CONF	VLDO	9_GPI	VLDO9_SEL	LDO9_PD_DIS	LDO9_GPI		LDO9_EN	
2F	RESERVED	Reserved	Rese	erved	Reserved	Reserved	Reserved		Reserved	
30	LDO11_CONT	LDO11_CONF	VLDO1	1_GPI	VLDO11_SEL	LDO11_PD_DIS	LDO1	1_GPI	LDO11_EN	
31	VIB	Res	erved			VIE	S_SET			
32	DVC_1	VLDO3_SEL	VLDO2_SEL	VLDO1_SEL	VBPERI_SEL	VBMEM_SEL	VBPRO_SEL	VBCORE2_SEL	VBCORE1_SEL	
33	DVC_2	VLDO4_SEL			Res	served			VBIO_SEL	
GP-AD	C Control Regis	ters (GPADC)								
34	ADC_MAN	Reserved	Reserved	ADC_MODE	ADC_MAN		ADC.	_M UX		
35	ADC_CONT	COMP1V2_EN	AD3_ISRC_EN	AD2_ISRC_EN	AD1_ISRC_EN	AUTO_AD3_EN	AUTO_AD2_EN	AUTO_AD1_EN	AUTO_VSYS_EN	
36	VSYS_MON				VSY	'S_MON				
37	ADC_RES_L	ADC_R	RES_LSB			Res	served			
38	ADC_RES_H		ADC_RES_H							
39	VSYS_RES				VS	YS_RES				



Hex	Register	7	6	5	4	3	2	1	0	
3A	ADCIN1_RES				ADC	IN1_RES				
3B	ADCIN2_RES				ADC	CIN2_RES				
3C	ADCIN3_RES				ADC	CIN3_RES				
3D	MON_A8_RES				МОМ	I_A8_RES				
3E	MON_A9_RES				МОМ	I_A9_RES				
3F	MON_A10_RES				MON	_A10_RES				
Reserve	ed									
40	RESERVED	Reserved	Reserved			Re	served			
41	RESERVED	Res	erved			Re	served			
42	RESERVED		Reserved		Reserved					
43	RESERVED		Reserved		Reserved					
44	RESERVED		Rese	erved	Reserved					
45	RESERVED	Reserved	Reserved			Re	served			
46	RESERVED	Res	erved			Re	served			
47	RESERVED	Res	erved		Reserved					
48	RESERVED		Reserved		Reserved					
49	RESERVED		Reserved				Reserved			
4A	RESERVED	Res	erved	Reserved	Reserved		Res	erved		
4B	RESERVED	Reserved	Reserved			Re	served			
Reserve	ed									
4C	RESERVED		Reserved							
4D	RESERVED		Reserved							
4E	RESERVED				Re	eserved				



Hex	Register	7	6	5	4	3	2	1	0
4F	RESERVED	Reserved							



Hex	Register	7	6	5	4	3	2	1	0			
	PAGE 1											
80	PAGE_CON	Revert	WRITE_MODE		Reserved		REG_PAGE					
Seque	Sequencer Control Registers (SEQ)											
81	SEQ		NXT_SEC	Q_START		SEQ_POINTER						
82	SEQ_TIMER	SEQ_DUMM				SEQ_TIME						
83	RESERVED	Reserved				Reserved						
84	ID_4_3		Rese	rved		LDO3_STEP						
85	ID_6_5		Rese	rved		Reserved						
86	ID_8_7	LDO8_STEP				LDO7_STEP						
87	ID_10_9	Reserved				LDO9_STEP						
88	ID_12_11	PD_DIS_STEP				LDO11_STEP						
89	ID_14_13	BUCKCORE2_STEP				BUCKCORE1_STEP						
8A	ID_16_15	BUCKIO_STEP				BUCKPRO_STEP						
8B	ID_18_17	BUCKPERI_STEP				BUCKMEM_STEP						
8C	RESERVED		Rese	rved		Reserved						
8D	ID_22_21		GP_FALI	_1_STEP		GP_RISE1_STEP						
8E	ID_24_23	GP_FALL2_STEP				GP_RISE2_STEP						
8F	ID_26_25	GP_FALL3_STEP				GP_RISE3_STEP						
90	ID_28_27	GP_FALL4_STEP				GP_RISE4_STEP						
91	ID_30_29	GP_FALL5_STEP				GP_RISE5_STEP						
92	ID_32_31		EN32K	_STEP		WAIT_STEP						
93	RESERVED		Rese	rved		Reserved						
94	RESERVED		Rese	rved		Reserved						



Hex	Register	7	6	5	4	3	2	1	0		
95	SEQ_A		POWEI	R_END			SYSTE	EM_END			
96	SEQ_B		PART_	DOWN		MAX_COUNT					
97	WAIT	WAI	T_DIR	TIME_OUT	WAIT_MODE	WAIT	WAIT_TIME				
98	RESERVED	Reserved	Reserved	Reserved	Reserved	Reserved		Reserved			
99	RESET	RESET	_EVENT			RESE	Γ_TIMER				
Regula	ator Setting Regis	sters (REG)									
9A	BUCK_ILIM_A		BMEN	I_ILIM			BIO	_ILIM			
9B	BUCK_ILIM_B		BPER	I_ILIM			BPR	O_ILIM			
9C	BUCK_ILIM_C		BCORE	2_ILIM			BCOR	E1_ILIM			
9D	BCORE2_CONF	BCORE	E2_MODE	BCORE2_PD_DI S	Res	served BCORE2_FB					
9E	BCORE1_CONF	BCORE	1_MODE	BCORE1_PD_DI S	Reserved			BCORE1_FB			
9F	BPRO_CONF	BPRO	_MODE	BPRO_PD_DIS	BPRO_VTT_EN	BPRO_VTTR_EN		BPRO_FB			
A0	BIO_CONF	BIO_	MODE	BIO_PD_DIS	Res	erved		BIO_FB			
A1	BMEM_CONF	BMEN	1_MODE	BMEM_PD_DIS	Res	erved		BMEM_FB			
A2	BPERI_CONF	BPER	I_MODE	BPERI_PD_DIS	Res	erved		BPERI_FB			
А3	VBCORE2_A	BCORE2_SL_A				VBCORE2_A					
A4	VBCORE1_A	BCORE1_SL_A				VBCORE1_A					
A5	VBPRO_A	BCPRO_SL_A		VBPRO_A							
A6	VBMEM_A	BMEM_SL_A		VBMEM_A							
A7	VBIO_A	BIO_SL_A		VBIO_A							
A8	VBPERI_A	BPERI_SL_A				VBPERI_A					
A9	Reserved	Reserved	Reserved								



Hex	Register	7	6	5	4	3	2	1	0			
AA	Reserved	Reserved	Reserved	Reserved								
AB	VLDO3_A	LDO3_SL_A		VLDO3_A								
AC	RESERVED	Reserved		Reserved								
AD	RESERVED	Reserved	Reserved	Reserved								
AE	RESERVED	Reserved	Reserved			Res	served					
AF	VLDO7_A	LDO7_SL_A	Reserved			VLD	007_A					
В0	VLDO8_A	LDO8_SL_A	Reserved			VLD	OO8_A					
B1	VLDO9_A	LDO9_SL_A	Reserved			VLD	OO9_A					
B2	Reserved	Reserved	Reserved			Res	served					
В3	VLDO11_A	LDO11_SL_A	Reserved			VLD	O11_A					
B4	VBCORE2_B	BCORE2_SL_B		VBCORE2_B								
B5	VBCORE1_B	BCORE1_SL_B				VBCORE1_B						
В6	VBPRO_B	BCPRO_SL_B				VBPRO_B						
В7	VBMEM_B	BMEM_SL_B				VBMEM_B						
B8	VBIO_B	BIO_SL_B				VBIO_B						
В9	VBPERI_B	BPERI_SL_B				VBPERI_B						
BA	RESERVED	Reserved	Reserved			Res	served					
BB	RESERVED	Reserved	Reserved			Res	served					
ВС	VLDO3_B	LDO3_SL_B		VLDO3_B								
BD	RESERVED	Reserved		Reserved								
BE	RESERVED	Reserved	Reserved	Reserved								
BF	RESERVED	Reserved	Reserved	Reserved								
C0	VLDO7_B	LDO7_SL_B	Reserved	VLDO7_B								



Hex	Register	7	6	5	4	3	2	1	0			
C1	VLDO8_B	LDO8_SL_B	Reserved	VLDO8_B								
C2	VLDO9_B	LDO9_SL_B	Reserved	VLDO9_B								
C3	RESERVED	Reserved	Reserved	Reserved								
C4	VLDO11_B	LDO11_SL_B	Reserved			VLD	O11_B					
Reserve	Reserved											
C5	RESERVED		Rese	Reserved Reserved								
				Reserved								
C6	RESERVED	Reserved		Reserved								
C7	RESERVED	Reserved				Reserved						
C8	RESERVED	Reserved				Reserved						
GP-AD	C Threshold Reg	gisters (GPADC)										
C9	ADC_CONT	ADCIN3_DEB	ADCIN2_DEB	ADCIN1_DEB	ADCIN3_CUR	ADCIN:	2_CUR	ADCIN	I1_CUR			
CA	AUTO1_HIGH				AUT	O1_HIGH						
СВ	AUTO1_LOW				AUT	O1_LOW						
CC	AUTO2_HIGH				AUT	O2_HIGH						
CD	AUTO2_LOW		AUTO2_LOW									
CE	AUTO3_HIGH		AUTO3_HIGH									
CF	AUTO3_LOW				AUT	O3_LOW						



Hex	Register	7	6	5	4	3	2	1	0	
					PAGE 2					
100	PAGE_CON	REVERT	WRITE_MODE	Reserved			REG_PAGE			
ОТР										
101	OTP_CONT	GP_WRITE_DI S	OTP_CONF_LOC K	OTP_APPS_LOC K	OTP_GP_LOCK	PC_DONE	OTP_APPS_RD	OTP_GP_RD	OTP_TIM	
102	OTP_ADDR				ОТІ	P_ADDR				
103	OTP_DATA				OTI	P_DATA				
Custor	mer Trim and Co	nfiguration Regis	sters							
104	T_OFFSET				T_0	OFFSET				
105	INTERFACE		IF_BASE	_ADDR		R/W_POL	СРНА	CPOL	nCS_POL	
106	CONFIG_A	IF_TYPE	PM_IF_HSM	PM_IF_FMP	PM_IF_V	IRQ_TYPE	PM_O_TYPE	PM_O_V	PM_I_V	
107	CONFIG_B	CHG_CLK_MO DE	VDD_HYST_ADJ				VDD_FA	ULT_ADJ		
108	CONFIG_C	BPERI_CLK_IN V	BIO_CLK_INV	BMEM_CLK_INV	BPRO_CLK_IN V	BCORE1_CLK_IN V	BUCK_ACTV_DI SCHG	LDO1_	_DO1_TRACK	
109	CONFIG_D	GP_FB3_TYPE	GP_FB2_TYPE	FORCE_RESET	HS_IF_HSM	HS_IF_FMP	SYSTEM_EN_RD	nIRQ_MODE	GPI_V	
10A	CONFIG_E	PERI_SW_AUT O	CORE_SW_AUT O	BPERI_AUTO	BIO_AUTO	BMEM_AUTO	BPRO_AUTO	BCORE2_AUTO	BCORE1_AUTO	
10B	CONFIG_F	LDO11_BYP	LDO8_BYP	LDO7_BYP	Reserved	LDO3_BYP	LDO11_AUTO	Reserved	LDO9_AUTO	
10C	CONFIG_G	LDO8_AUTO	LDO7_AUTO	Reserved	Reserved	Reserved	LDO3_AUTO	Reserved	Reserved	
10D	CONFIG_H	BUCK_MERGE	BCORE1_OD	BCORE2_OD	BPRO_OD	BCORE_MERGE	MERGE_SENSE	LDO8_MODE	PWM_CLK	
10E	CONFIG_I	LDO_SD	INT_SD_MODE	HOST_SD_MOD E	KEY_SD_MOD E	GPI14_15_SD	nONKEY_SD	D nONKEY_PIN		
10F	CONFIG_J	IF_RESET	TWOWIRE_TO	RESET_DI	JRATION	SHUT_	DELAY	ELAY KEY_DELAY		
110	CONFIG_K	GPI7_PUPD	GPI6_PUPD	GPI5_PUPD	GPI4_PUPD	GPI3_PUPD	GPIO2_PUPD	GPIO1_PUPD	GPIO0_PUPD	



Hex	Register	7	6	5	4	3	2	1	0				
111	CONFIG_L	GPIO15_PUPD	GPIO14_PUPD	GPIO13_PUPD	GPIO12_PUPD	GPIO11_PUPD	GPIO10_PUPD	GPIO9_PUPD	GPIO8_PUPD				
112	CONFIG_M	OSC	_FREQ		Res	served							
113	CONFIG_N	Reserved											
114	MON_REG_1	UVOV	_DELAY	MON_N	MODE	MON_DEB	MON_RES	MON_	THRES				
115	MON_REG_2	LDO8_MON_E N	LDO7_MON_EN	Reserved	Reserved	Reserved	LDO3_MON_EN	Reserved	Reserved				
116	MON_REG_3			Reserved			LDO11_MON_EN	Reserved	LDO9_MON_EN				
117	MON_REG_4	Res	erved	BPERI_MON_EN	BMEM_MON_E N	BIO_MON_EN	BPRO_MON_EN	BCORE2_MON_E N	BCORE1_MON_E N				
118	RESERVED				Re	served							
119	RESERVED		Reserved										
11A	RESERVED		Reserved										
11B	RESERVED				Re	served							
11C	RESERVED				Re	served							
11D	RESERVED				Re	served							
11E	MON_REG_5	Reserved		MONA9_IDX		Reserved		MONA8_IDX					
11F	MON_REG_6			Reserved				MONA10_IDX					
120	TRIM_CLDR				TRII	M_CLDR							
Genera	al Purpose Regis	ters (GP)	s (GP)										
121	GP_ID_0		GP_0										
122	GP_ID_1		GP_1										
123	GP_ID_2	GP_2											
124	GP_ID_3		GP_3										
125	GP_ID_4				(GP_4							



Hex	Register	7	6	5	4	3	2	1	0					
126	GP_ID_5					GP_5								
127	GP_ID_6		GP_6											
128	GP_ID_7		GP_7											
129	GP_ID_8				(GP_8								
12A	GP_ID_9				(GP_9								
12B	GP_ID_10				C	SP_10								
12C	GP_ID_11				C	SP_11								
12D	RESERVED				RES	SERVED								
12E	RESERVED				RES	SERVED								
12F	RESERVED				RES	SERVED								
130	RESERVED				RES	SERVED								
131	RESERVED				RES	SERVED								
132	RESERVED				RES	SERVED								
133	RESERVED				RES	SERVED								
134	RESERVED				RES	SERVED								
Interna	al Debug Registe	rs												
135	RESERVED				Re	eserved								
136	RESERVED		Reserved											
137	RESERVED		Reserved											
138	RESERVED		Reserved											
139	RESERVED	Reserved												
13A	RESERVED		Reserved											
13B	RESERVED				Re	eserved								



Hex	Register	7	6	5	4	3	2	1	0				
13C	RESERVED		Reserved										
13D	RESERVED		Reserved										
13E	RESERVED				Re	eserved							
					PAGE 3								
180	PAGE_CON	REVERT WRITE_MODE Reserved REG_PAGE											
Chip lo	dentification Reg	isters											
181	DEVICE_ID				DE	VICE_ID							
182	VARIANT_ID		MF	RC			V	RC					
183	CUSTOMER_ID		CUST_ID										
184	CONFIG_ID		CONFIG_REV										
1A8	PMIC_STATUS	PC_DONE Reserved Status											

Most register bits are reset to defaults (zero in most cases) when powering up from RESET mode. An exception is for example FAULT_LOG that is not loaded from OTP. Register fields shown in BOLD are loaded from OTP.



8 Definitions

8.1 Power Dissipation and Thermal Design

When designing with the DA9063L, consideration must be given to power dissipation as the level of integration of the device can result in high power when all functions are operating with high battery voltages. Exceeding the package power dissipation capabilities results in the internal thermal sensor shutting down the device until it has sufficiently cooled.

The package includes a thermal management paddle to improve heat spreading into the PCB.

For Linear Regulators:

Linear regulators operating with a high current and high differential voltage between input and output dissipate the following power:

$$P_{diss} = (V_{in} - V_{out}) \times I_{out}$$

Example: a regulator supplying 150 mA at 2.8 V from a fully-charged lithium battery (VDD=4.1 V):

$$P_{diss} = (4.1V - 2.8V) \times 0.15A = 195mW$$

For Switching Regulators:

$$P_{out} = P_{in} \times Efficiency$$

Therefore,

$$\begin{split} P_{diss} &= P_{in} - P_{out} \\ P_{diss} &= \frac{P_{out}}{Efficiency} - P_{out} \\ P_{diss} &= P_{out} \times \left(\frac{1}{Efficiency} - 1\right) \\ P_{diss} &= I_{out} \times V_{out} \times \left(\frac{1}{Efficiency} - 1\right) \end{split}$$

Example: an 85 % efficient buck converter supplying 1.2 V at 400 mA:

$$P_{diss} = 1.2V \times 0.4A \times \left(\frac{1}{0.85} - 1\right) = 85mW$$

As the DA9063L has multiple regulators, each supply must be separately considered and their powers summed to give the total device dissipation (current drawn from the reference and control circuitry can be considered negligible in these calculations).



8.2 Regulator Parameter - Dropout Voltage

In the DA9063L, a regulator's dropout voltage is defined as the minimum voltage differential between the input and output voltages whilst regulation still takes place. Within the regulator, voltage control takes place across a PMOS pass transistor and, when entering the dropout condition, the transistor is fully turned on and therefore cannot provide any further voltage control. When the transistor is fully turned on, the output voltage tracks the input voltage and regulation ceases. As the DA9063L is a CMOS device and uses a PMOS pass transistor, the dropout voltage is directly related to the onresistance of the device. In the device, the pass transistors are sized to provide the optimum balance between required performance and silicon area. By employing a 0.25 µm process, Dialog Semiconductor is able to achieve very small pass transistor sizes for superior performance.

$$V_{dropout} = V_{in} - V_{out} = R_{dson} \times I_{out}$$

When defining dropout voltage, it is specified in relation to a minimum acceptable change in output voltage. For example, all Dialog Semiconductor regulators have dropout voltage defined as the point at which the output voltage drops 10 mV below the output voltage at the minimum guaranteed operating voltage. The worst case conditions for dropout are high temperature (highest on-resistance for the internal pass device) and maximum current load.

8.3 Regulator Parameter - Power Supply Rejection

Power supply rejection (PSRR) is especially important in the supplies to the RF and audio parts of the telephone. In a TDMA system such as GSM, the 217 Hz transmit burst from the power amplifier results in significant current pulses being drawn from the battery. These can peak at up to 2 A before reaching a steady state of 1.4 A (see below). Due to the battery having a finite internal resistance (typically 0.5 Ω), these current peaks induce ripple on the battery voltage of up to 500 mV. Since the supplies to the audio and RF are derived from this supply, it is essential that this ripple is removed otherwise it would show as a 217 Hz tone in the audio and could also affect the transmit signal. Power supply rejection should always be specified under worst case conditions – when the battery is at its minimum operating voltage and when there is minimum headroom available due to dropout.

8.4 Regulator Parameter - Line Regulation

Static line regulation is a measurement that indicates a change in the regulator output voltage, ΔV reg (regulator operating with a constant load current), in response to a change in the input voltage, ΔV in. Transient line regulation is a measurement of the peak change, ΔV reg, in regulated voltage seen when the line input voltage changes.

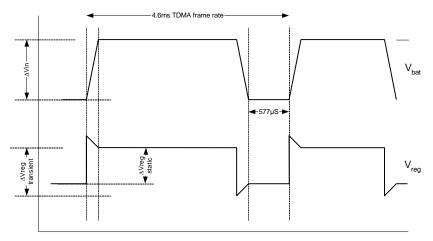


Figure 33: Line Regulation



8.5 Regulator Parameter - Load Regulation

Static load regulation is a measurement that indicates a change in the regulator output voltage, ΔV reg, in response to a change in the regulator loading, Δl oad, whilst the regulator input voltage remains constant. Transient load regulation is a measurement of the peak change in regulated voltage, ΔV reg, seen when the regulator load changes.

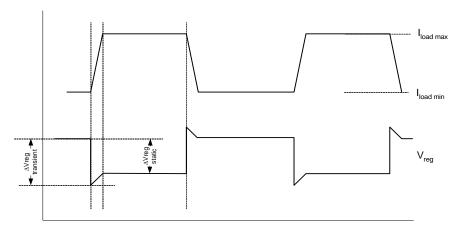


Figure 34: Load Regulation

Please contact Dialog Semiconductor for latest application information on the DA9063L and other power management devices.



9 Package Information

9.1 Package Outlines

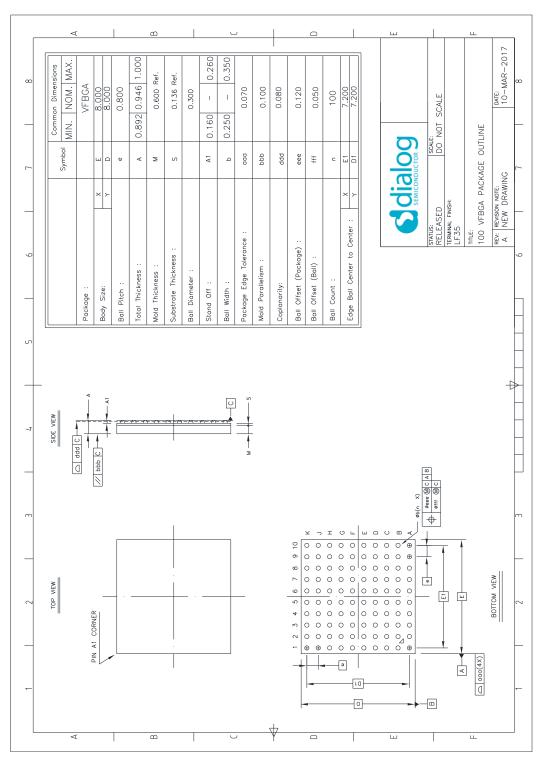


Figure 35: Package Outline Drawing 100 VFBGA 0.3 mm Ball Diameter



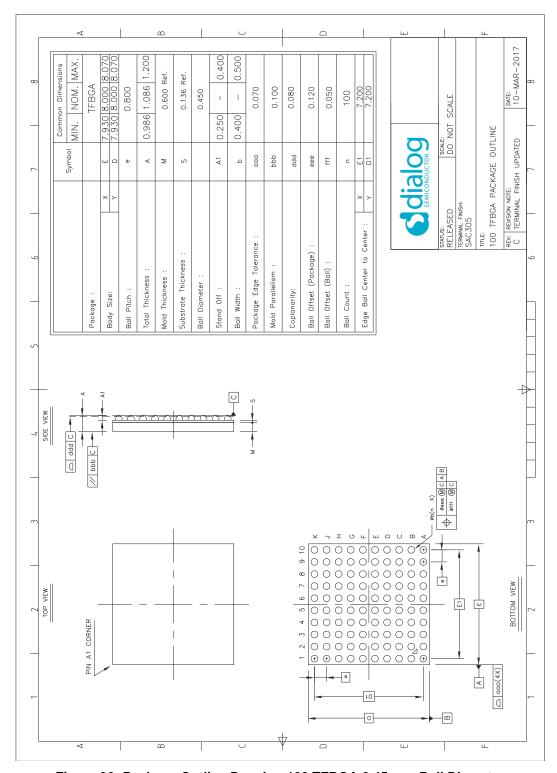


Figure 36: Package Outline Drawing 100 TFBGA 0.45 mm Ball Diameter



9.2 Package Marking

				Pa	ckag	је Ма	arkir	ıg				
A1 Corner >			M	larking	Conte	Format						
1st	•				Orientation							
2nd	dialog e1 Logo											
3rd	D	Α	9	0	6	3	L		Part No.			
4th	x	x	V	V	-	A	T		OTP/Silicon Version/Option			
5th	у	У	W	w	Z	Z	Z	z	Date Code			
Date Code Fo	Date Code Format: yy = Year, ww = Week, zzzz = Traceability											
xx identifies th	xx identifies the OTP Variant, vv may be used to show the silicon version.											
-A and -AT op	tionally in	dicate	the Au	ıtomot	ive an	d Auto	motive	e high	temp test options.			

9.3 Moisture Sensitivity Level

The Moisture Sensitivity Level (MSL) is an indicator for the maximum allowable time period (floor lifetime) in which a moisture sensitive plastic device, once removed from the dry bag, can be exposed to an environment with a specified maximum temperature and a maximum relative humidity before the solder reflow process. The MSL classification is defined in Table 37.

For detailed information on MSL levels refer to the IPC/JEDEC standard J-STD-020, which can be downloaded from http://www.jedec.org.

The VFBGA are TFBGA packages are qualified for MSL 3.

Table 37: MSL Classification

MSL Level	Floor Lifetime	Conditions
MSL 4	72 hours	30 °C / 60 % RH
MSL 3	168 hours	30 °C / 60 % RH
MSL 2A	4 weeks	30 °C / 60 % RH
MSL 2	1 year	30 °C / 60 % RH
MSL 1	Unlimited	30 °C / 85 % RH



10 Ordering Information

The ordering number consists of the part number followed by a suffix indicating the packing method. For details and availability, please consult Dialog Semiconductor's website or your local sales representative.

Table 38: Ordering Information

Part Number	Package	Shipment Form	Pack Quantity
Consumer: 0.30 mm ba	all diameter, RT production testing		
DA9063L-xxHK1 Note 2	100 VFBGA, 8.0 mm x 8.0 mm x 1.0 mm, 0.8 mm pitch, Pb-free/green	Tray	260
DA9063L-xxHK2 Note 2	100 VFBGA, 8.0 mm x 8.0 mm x 1.0 mm, 0.8 mm pitch, Pb-free/green	T&R	2500
Consumer / Industrial:	0.45 mm ball diameter, RT production testing	ng	
DA9063L-xxHO1	100 TFBGA, 8.0 mm x 8.0 mm x 1.2 mm, 0.8 mm pitch, Pb-free/green	Tray	260
DA9063L-xxHO2	100 TFBGA, 8.0 mm x 8.0 mm x 1.2 mm, 0.8 mm pitch, Pb-free/green	T&R	2500
Automotive AEC-Q100	Grade 2: 0.45 mm ball diameter, RT produc	tion testing Note 1	
DA9063L-xxHO1-A	100 TFBGA, 8.0 mm x 8.0 mm x 1.2 mm, 0.8 mm pitch, Pb-free/green	Tray	260
DA9063L-xxHO2-A	100 TFBGA, 8.0 mm x 8.0 mm x 1.2 mm, 0.8 mm pitch, Pb-free/green	T&R	2500
Automotive AEC-Q100	Grade 2: 0.45 mm ball diameter, HT produc	tion testing Note 1	
DA9063L-xxHO1-AT	100 TFBGA, 8.0 mm x 8.0 mm x 1.2 mm, 0.8 mm pitch, Pb-free/green	Tray	260
DA9063L-xxHO2-AT	100 TFBGA, 8.0 mm x 8.0 mm x 1.2 mm, 0.8 mm pitch, Pb-free/green	T&R	2500

Note 1 Operating temperature is defined in Section 5.2. Automotive qualification temperature is defined by the Automotive Electronics Council AEC-Q100 specification.

10.1 Variants Ordering Information

DA9063L supports delivery of variants indicated by xx in the Part number above, where xx is replaced with the actual variant number. Please contact your local Dialog Semiconductor office or representative to discuss requirements.

All Dialog automotive power management products that are AEC-Q100 qualified have an -A or -AT suffix for dual temperature production test.

Note 2 The 0.30 mm ball diameter version is not recommended for new designs.



11 Application Information

The following recommended components are examples selected from requirements of a typical application. The electrical characteristics (for example, supported voltage/current range) have to be cross-checked and component types may need to be adapted from the individual needs of the target circuitry.

11.1 Capacitor Selection

Ceramic capacitors are used as bypass capacitors at all VDD and output rails. When selecting a capacitor, especially for types with high capacitance at smallest physical dimension, the DC bias characteristic has to be taken into account. On the V_{SYS} main supply rail a minimum distributed capacitance of 40 μ F (actual capacitance after voltage and temperature derating) is required. For example, a typical design might use:

- 22 μF within 1.5 mm of each BUCKCORE1, BUCKCORE2 and BUCKPRO supply pin.
- 10 μ F within 1.5 mm of each BUCKPERI, BUCKIO and BUCKMEM supply pin or 1 μ F x 22 μ F if all are attached to a PCB power/split plane.
- 2 μF x 1 μF shared by all VDD_LDOx pins if they are all close together, for example, all attached
 to a power/split plane.
- 1 μF close to the VSYS pin.
- Buck output capacitors should be close to the buck inductors.

The amount of decoupling required will be dependent on the specific application.

Table 39: Recommended Capacitor Types

Application	Value	Tol. (%)	Size	Height (mm)	Temp. Char.	Rated Voltage (V)	Type (Murata/Samsung)
VDDCORE, VLDO3, VLDO7, VLDO8, VLDO9, VLDO11	2.2 µF	± 20	0402	0.55	X5R ±15 %	6.3	GRM155R60J225ME95
VDUOKDED	22 uE	± 20	0805	0.95	X5R ±15 %	6.3	GRM219R60J226M***
VBUCKPER, VBUCKIO,	22 µF	± 20	0402	0.5	X5R ±15 %	4.0	CL05A226MR5NZNC
VBUCKMEM, VSYS	47 µF	± 20	0805	0.95	X5R ±15 %	4.0	GRM219R60G476M***
V313	41 µr	± 20	0603	0.8	X5R ±15 %	4.0	CL10A476MR8NZN
VBUCKCORE	22 µF	± 20	0603	1.0	X5R ±15 %	6.3	GRM188R60J226MEA0 Note 1
1 and 2, VBUCKPRO		± 20	0402	0.5	X5R ±15 %	4.0	CL05A226MR5NZNC
(using full- current mode)	47 μF	± 20	0805	0.95	X5R ±15 %	4.0	GRM219R60G476M***61
carrent mode)	47 μΓ	± 20	0603	0.8	X5R ±15 %	4.0	CL10A476MR8NZN
VSYS	1.0 µF	± 10	0402	0.5	X5R ±15 %	10	GRM155R61A105KE15
VBBAT	470 nF	± 10	0402	0.5	X5R ±15 %	50	GRM155R61A474KE11
VREF, VLNREF	0.22 µF	± 10	0402	0.5	X/R ±15 %	16	GRM155R71C224KA12



Application	Value	Tol. (%)	Size	Height (mm)	Temp. Char.	Rated Voltage (V)	Type (Murata/Samsung)
VDDCORE	2.2 µF	± 20	0402	0.5	X5R ±15 %	6.3	GRM155R60J225ME95
XIN, XOUT	12 pF	± 5	0402	0.5	C0G	50	GRM1555C1H120JA01D
V_CP	47 nF	± 10	0402	0.5	X7R ±15 %	10	GRM155R71A473KA01

Note 1 For output voltages > 1.4 V Murata GRM219R60J226M*** is recommended.

Table 40: Recommended Automotive Capacitor Types

Application	Value	Tol. (%)	Size	Height (mm)	Temp. Char.	Rated Voltage (V)	Type (Murata/Samsung)
VDDCORE, VLDO3, VLDO7, VLDO8. VLDO9, VLDO11	2.2 µF	± 20	0402	1.25	X7R	10	GCM21BR71A225MA37
VBUCKPER,	22 µF	± 20	1206	1.6	X7R	6.3	GCM31CR70J226ME23
VBUCKIO, VBUCKMEM, VSYS	47 μF	± 10	1206	1.6	X7R	6.3	GCM32ER70J476KE19
VBUCKCORE 1 and 2,	22 µF	± 20	1206	1.6	X7R	6.3	GCM31CR70J226ME23
VBUCKPRO (using full- current mode)	47 μF	± 10	1206	1.6	X7R	6.3	GCM32ER70J476KE19
VSYS	10 µF	± 10	0805	1.25	X7R	10	GCM21BR71A106ME22
VSTS	1.0 µF	± 10	0805	1.25	X7R	16	GCM21BR71C105KA58
VBBAT	470 nF	± 10	0402	0.5	X7R	10	GCM155C71A474KE36 D
VREF, VLNREF	0.22 μF	± 10	0402	0.5	X7R	16	GCM155R71C224ME02
VDDCORE	2.2 µF	± 20	0402	1.25	X7R	10	GCM21BR71A225MA37
XIN, XOUT	12 pF	± 5	0402	0.5	COG	50	GRM1555C1H120JA01 D
V_CP	47 nF	± 10	0402	0.5	X7R	10	GRM155R71A473KA01

11.2 Inductor Selection

Inductors should be selected based upon the following parameters:

- Rated maximum current: Usually a coil provides two current limits: ISAT of an Inductor specifies
 the current required to cause a reduction in the Inductance by a specified amount, typically 30 %,
 IRMS of an Inductor specifies the current required to affect a temperature rise of a maximum
 specified amount.
- DC resistance: Critical to converter efficiency at high current and should therefore be minimized.



- ESR at the buck switching frequency: Critical to converter efficiency in PFM mode and should therefore be minimized.
- Inductance: Given by converter electrical characteristics; 1.0 μH for all DA9063L switched-mode step-down converters.

Table 41: Recommended Inductor Types

Application	Value (µH)	Tol. (%)	ISAT (A)	IRMS (A)	DCR (Typ.) (mΩ)	Size (mm)	Туре
		±30	2.7	2.3	55	2.0x1.6x1.0	Toko 1285AS-H-1R0N
BUCKPERI, BUCKMEM, BUCKIO, BUCKCORE1,	1.0	±20	2.65	2.45	60	2.0x1.6x1.0	Tayo Yuden MAKK2016T1R0M (Reference)
BUCKCORE2		±20	3.9	3.1	50	2.0x1.6x1.0	TDK TFM201610ALM- 1R0MTAA
		±30	3.4	3.0	60	2.5x2.0x1.0	Toko 1269AS-H-1R0N
		±20	3.6	3.1	45	2.5x2.0x1.2	Tayo Yuden MAMK2520T1R0M
DUGUDDO		±20	3.8	3.5	45	2.5x2.0x1.2	Toko 1239AS-H-1R0N (Reference)
BUCKPRO, BUCKCORE1 and 2 using		±30	3.9	3.1	48	3.2x2.5x1.0	Toko 1276AS-H-1R0N
full-current mode or merged BUCKMEM/	1.0	±20	4.5	3.6	38	2.5x2.0x1.0	TDK TFM252010ALM- 1R0MTAA
BUCKIO		±20	4.2	3.7	35	2.5x2.0x1.2	TDK TFM252012ALMA- 1R0MTAA (AECQ200)
		±20	3.35	2.5	52	3.0x3.0x1.2	Cyntec PST031B-1R0MS
		±20	8.8	12.0	8.2	4.0x4.0x2.1	Coilcraft XGL4020-102ME (AEC- Q200)

11.3 Resistors

Table 42: Recommended Resistor Types

Application	Value	Size	Tolerance	P _{MAX}	Туре
IREF bias current reference	200 kΩ	0402	±1%	100 mW	Panasonic ERJ2RKF2003x



12 Layout Guidelines

12.1 General Recommendations

- Appropriate trace width and quantity of vias should be used for all power supply paths.
 Too high trace resistances can prevent the system from achieving the best performance, for
 - example, the efficiency and the current ratings of switch-mode converters and charger might be degraded. Furthermore, the PCB may be exposed to thermal hot spots, which can lead to critical overheating due to the positive temperature coefficient of copper.
 - Special care must be taken with the DA9063L pad connections. The traces of the outer row should be connected with the same width as the pads and should become wider as soon as possible. For supply pins in the second row, connection to an inner board layer is recommended (depending on the maximum current two or more vias might be required).
- A common ground plane should be used, which allows proper electrical and thermal performance. Noise sensitive references such as the VREF/VLNREF capacitors and IREF resistor should be referred to a silent ground which is connected at a star point underneath or close to the DA9063L main ground connection.
- Generally, all power tracks with discontinuous/high currents should be kept as short as possible.
- Noise sensitive analog signals such as feedback lines or crystal connections should be kept away from traces carrying pulsed analog or digital signals. This can be achieved by separation (distance) or shielding with quiet signals or ground traces.

12.2 LDOs and Switched Mode Supplies

- The placement of the distributed capacitors on the V_{SYS} rail must ensure that all VDD inputs –
 and especially to the V_{SYS} pin, the buck converters and LDOs are connected to a bypass
 capacitor close to the pads. It is recommended placing at least two 1 μF capacitors close to the
 LDO supply pads and at least one 10 μF close to the buck VDD rail.
 - Using a local power plane underneath the chip for V_{SYS} might be considered.
- Transient current loops in the area of the switched mode converters should be minimised.
- The common references (IREF resistor, VREF/VLNREF capacitors) should be placed close to the DA9063L and cross coupling to any noisy digital or analog trace must be avoided.
- Output capacitors of the LDOs can be placed close to the input pins of the supplied devices (remote from the DA9063L).
- Care must be taken with trace routing to ensure that no current is carried on feedback lines of the buck output voltages V_{BUCK}.
- The inductor placement is less critical since parasitic inductances have negligible effect.

12.3 Thermal Connection, Land Pad, and Stencil Design

- The DA9063L provides a central ground area of balls, which are soldered to the PCB's central
 ground pad. This PCB ground pad must be connected with as many vias and as direct as
 possible to the PCB's main ground plane in order to achieve good thermal performance.
- Solder mask openings for the ball landing pads must be arranged to prohibit solder balls flowing into vias.

For further PCB layout guidance, see PCB Layout Guidelines [3].



Appendix A Register Descriptions

This appendix describes the registers summarized in Section 7.

A.1 Register Page Control

Table 43: PAGE_CON

Register Address	Bit	Туре	Label	Description
0x00 PAGE_CON	7	R/W	REVERT	Resets REG_PAGE to 00 after read/write access has finished
	6	R/W	WRITE_MODE	2-WIRE multiple write mode 0: Page Write Mode 1: Repeated Write Mode
	5:3	R/W	Reserved	
	2:0	R/W	REG_PAGE	000: Selects Register 0x01 to 0x3F 001: Selects Register 0x81 to 0xCF 010: Selects Register 0x101 to 0x13F 011: Reserved for production and test

The PAGE_CON register is located at address 0x00 of each register page (0x00 and 0x80). The control interface provides an individual instance of the PAGE_CON register.

A.2 Register Page 0

A.2.1 Power Manager Control and Monitoring

The STATUS register reports the current value of the various signals at the time that it is read out. All the status bits have the same polarity as their corresponding signals.

Table 44: STATUS A

Register Address	Bit	Туре	Label	Description
0x01	7:4	R	Reserved	
STATUS_A	3	R	COMP1V2	Output state of 1.2 V comparator
	2	R	DVC	Asserted as long as at least one DVC supply performs voltage ramping
	1	R	WAKE	CHG_WAKE level
	0	R	nONKEY	Asserted as long nONKEY is pressed (low level)

Table 45: STATUS_B

Register Address	Bit	Туре	Label	Description
0x02	7	R	GPI7	GPI7 level
STATUS_B	6	R	GPI6	GPI6 level
	5	R	GPI5	GPI5 level
	4	R	GPI4	GPI4 level



Register Address	Bit	Туре	Label	Description
	3	R	GPI3	GPI3 level
	2	R	GPI2	GPI2 level or ADCIN3 threshold indicator (1 when overriding high limit)
	1	R	GPI1	GPI1 level or ADCIN2 threshold indicator (1 when overriding high limit)
	0	R	GPI0	GPI0 level or ADCIN1 threshold indicator (1 when overriding high limit)

Table 46: STATUS_C

Register Address	Bit	Туре	Label	Description
0x03	7	R	GPI15	GPI15 level
STATUS_C	6	R	GPI14	GPI14 level
	5	R	GPI13	GPI13/ EXT_WAKEUP/READY level
	4	R	GPI12	GPI12/nVDD_FAULT/VDD_MON level
	3	R	GPI11	GPI11 level
	2	R	GPI10	GPI10/PWR1_EN level
	1	R	GPI9	GPI9/PWR_EN level
	0	R	GPI8	GPI8/SYS_EN level

Table 47: STATUS_D

Register Address	Bit	Туре	Label	Description
0x04 STATUS_D	7	R/W	LDO11_LIM	Asserted as long LDO11 hits its over-current limit
	6	R/W	LDO8_LIM	Asserted as long LDO8 hits its over-current limit
	5	R/W	LDO7_LIM	Asserted as long LDO7 hits its over-current limit
	4	R/W	Reserved	
	3	R/W	LDO3_LIM	Asserted as long LDO3 hits its over-current limit
	2:0	R/W	Reserved	

Table 48: FAULT_LOG

Register Address	Bit	Type Note 1	Label	Description
0x05	7	R	WAIT_SHUT	Power down by time out of ID WAIT
FAULT_LOG	6	R	nSHUT_DOWN	Power down by assertion of port nOFF, nSHUTDOWN
	5	R	KEY_RESET	Power down from a long press of nONKEY or GPIO14/15



Register Address	Bit	Type Note 1	Label	Description
	4	R	TEMP_CRIT	Junction over-temperature detected
	3	R	VDD_START	Power down by V _{SYS} under-voltage detect before or within 16 seconds after entering ACTIVE mode
	2	R	VDD_FAULT	Power down by V _{SYS} under-voltage detect
	1	R	POR	DA9063L starts up from NO-POWER or DELIVERY mode
	0	R	TWD_ERROR	Watchdog time violated

Note 1 Cleared from the host by writing back the read value.

The EVENT registers hold information about events that have occurred in the DA9063L. Events are triggered by a change in the status registers that contains the status of monitored signals. When an EVENT bit is set in the event register the nIRQ signal is asserted (unless the nIRQ is to be masked by a bit in the IRQ mask register). The nIRQ is also masked during the power-up sequence and is not released until the event registers have been cleared. The IRQ triggering event register is cleared from the host by writing back its read value. The event registers may be read in page/repeated mode. New events that occur during clearing are delayed before they are passed to the event register, ensuring that the host controller does not miss them.

Table 49: EVENT_A

Register Address	Bit	Туре	Label	Description
0x06 EVENT_A	7	R	EVENTS_D	Asserted when register EVENT_B to EVENT_D have at least one event bit asserted
	6	R	EVENTS_C	Asserted when register EVENT_B to EVENT_C have at least one event bit asserted
	5	R	EVENTS_B	Asserted when register EVENT_B has at least one event bit asserted
	4	R Note 1	E_SEQ_RDY	Sequencer reached final position caused event
	3	R Note 1	E_ADC_RDY	ADC manual conversion result ready caused event
	2:1	R Note 1	Reserved	
	0	R Note 1	E_nONKEY	nONKEY caused event

Note 1 Cleared from the host by writing back the read value.

Table 50: EVENT B

Register Address	Bit	Type Note 1	Label	Description
0x07 EVENT_B	7	R	E_VDD_WARN	V _{SYS} dropped below VDD_FAULT_UPPER threshold
	6	R	E_VDD_MON	Vsys less or higher than VSYS_MON threshold caused event
	5	R	E_DVC_RDY	Finish of all DVC voltage ramping event



Register Address	Bit	Type Note 1	Label	Description
	4	R	E_REG_UVOV	Event triggered from a monitored regulator voltage being out of selected range or from new regulator voltage measurement being available (depends on settings of MON_MODE)
	3	R	E_LDO_LIM	LDO3, 7, 8 or 11 current limit exceeded for more than 10 ms
	2	R	E_COMP1V2	1.2 V comparator caused event
	1	R	E_TEMP	Junction high temp caused event
	0	R	E_WAKE	Detected rising edge on CHG_WAKE

Note 1 Cleared from the host by writing back the read value.

Table 51: EVENT_C

Register Address	Bit	Type Note 1	Label	Description
0x08	7	R	E_GPI7	GPI event according to active state setting
EVENT_C	6	R	E_GPI6	GPI event according to active state setting
	5	R	E_GPI5	GPI event according to active state setting
	4	R	E_GPI4	GPI event according to active state setting
	3	R	E_GPI3	GPI event according to active state setting
	2	R	E_GPI2	GPI event according to active state setting / ADCIN3 high / low threshold exceeded caused event
	1 R	R	E_GPI1	GPI event according to active state setting / ADCIN2 high / low threshold exceeded caused event
	0	R	E_GPI0	GPI event according to active state setting / ADCIN1 high / low threshold exceeded caused event

Note 1 Cleared from the host by writing back the read value.

Table 52: EVENT_D

Register Address	Bit	Type Note 1	Label	Description
0x09	7	R	E_GPI15	GPI event according to active state setting
EVENT_D	6	R	E_GPI14	GPI event according to active state setting/Event caused from host addressing HS 2-WIRE interface
	5	R	E_GPI13	GPI event according to active state setting
	4	R	E_GPI12	GPI event according to active state setting
	3	R	E_GPI11	GPI event according to active state setting
	2	R	E_GPI10	GPI/PWR1_EN event according to active state setting



Register Address	Bit	Type Note 1	Label	Description
	1	R	E_GPI9	GPI/PWR_EN event according to active state setting
	0	R	E_GPI8	GPI/SYS_EN event according to active state setting

Note 1 Cleared from the host by writing back the read value.

The nIRQ line is released only when all events have been cleared from the host processor by writing the read value into all registers with an asserted event bit.

Table 53: IRQ_MASK_A

Register Address	Bit	Туре	Label	Description
0x0A	7:5	R/W	Reserved	
IRQ_MASK_A	4	R/W	M_SEQ_RDY	Mask nIRQ from finishing power sequencing
	3	R/W	M_ADC_RDY	Mask ADC manual conversion result ready caused nIRQ
	2:1	R/W	Reserved	
	0	R/W	M_nONKEY	Mask nONKEY caused nIRQ

Table 54: IRQ_MASK_B

Register Address	Bit	Туре	Label	Description
0x0B IRQ_MASK_B	7	R/W	M_VDD_WARN	Mask VDDFAULT _UPPER comparator triggered event
	6	R/W	M_VDD_MON	Mask V _{SYS} caused nIRQ
	5	R/W	M_DVC_RDY	Mask DVC voltage ramping triggered event
	4	R/W	M_REG_UVOV	Mask events generated from regulator output voltage monitoring
	3	3 R/W	R/W	M_LDO_LIM
	2	R/W	M_COMP1V2	Mask 1.2 V comparator caused nIRQ
	1	R/W	M_TEMP	Mask junction over temp caused nIRQ
	0	R/W	M_WAKE	Mask companion charger caused event

Table 55: IRQ_MASK_E

Register Address	Bit	Туре	Label	Description
0x0C	7	R/W	M_GPI7	Mask GPI caused nIRQ
IRQ_MASK_E	6	R/W	M_GPI6	Mask GPI caused nIRQ
	5	R/W	M_GPI5	Mask GPI caused nIRQ
	4	R/W	M_GPI4	Mask GPI caused nIRQ



Register Address	Bit	Туре	Label	Description
	3	R/W	M_GPI3	Mask GPI caused nIRQ
	2	R/W	M_GPI2	Mask GPI caused / ADCIN3 high / low threshold exceeded caused nIRQ
	1	R/W	M_GPI1	Mask GPI caused / ADCIN2 high / low threshold exceeded caused nIRQ
	0	R/W	M_GPI0	Mask GPI caused / ADCIN1 high / low threshold exceeded caused nIRQ

Table 56: IRQ_MASK_F

Register Address	Bit	Туре	Label	Description
0x0D	7	R/W	M_GPI15	Mask GPI caused nIRQ
IRQ_MASK_F	6	R/W	M_GPI14	Mask GPI/HS 2-WIRE caused nIRQ
	5	R/W	M_GPI13	Mask GPI caused nIRQ
	4	R/W	M_GPI12	Mask GPI caused nIRQ
	3	R/W	M_GPI11	Mask GPI caused nIRQ
	2	R/W	M_GPI10	Mask GPI/PWR1_EN caused nIRQ
	1	R/W	M_GPI9	Mask GPI/PWR_EN caused nIRQ
	0	R/W	M_GPI8	Mask GPI/SYS_EN caused nIRQ

Table 57: CONTROL_A

Register Address	Bit	Туре	Label	Description
0xE	7	R/W	Reserved	
CONTROL_A	6	R/W	M_POWER1_EN	Mask the update of POWER1_EN when writing to CONTROL_A
	5	R/W	M_POWER_EN	Mask the update of POWER_EN when writing to CONTROL_A
	4	R/W	M_SYSTEM_EN	Mask the update of SYSTEM_EN when writing to CONTROL_A
	3	3 R/W STAND	STANDBY	Clearing SYSTEM_EN/releasing port SYS_EN press will
				0: completely power down to Slot 0 (Hibernate) 1: stop powering down at pointer PART_DOWN (Standby)
	2	R/W	POWER1_EN	Target status of power domain POWER1: controlled from OTP/PM interface and port PWR1_EN
	1	R/W	POWER_EN	Target status of power domain POWER: controlled from OTP/PM interface and port PWR_EN



Register Address	Bit	Туре	Label	Description
	0	R/W	SYSTEM_EN	Target status of power domain SYSTEM: controlled from OTP/PM interface and port SYS_EN

Table 58: CONTROL_B

Register Address	Bit	Туре	Label	Description
0xF CONTROL_B	7	R/W	BUCK_SLOWSTART	Enables soft-start for buck converters (recommended for application instant-on with discharged battery and weak external supply)
				Note 1
	6:5	R/W	Reserved	
	4	R/W	nONKEY_LOCK	0: Half-current POWERDOWN mode 1: Wakeup from POWERDOWN mode requires the nONKEY signal being low for longer than selected in KEY_DELAY (automatically cleared during power-up sequence)
	3	R/W	nRES_MODE	0: No assertion of nRESET for power down sequence 1: Assert nRESET before starting power down sequence (release after leaving POWERDOWN mode in case RESET_EVENT < '11')
	2	R/W	RES_BLINKING	Enables (time limited) VDD_START triggered GPO11/4/15 flashing in case of no connected external supply
	1	R/W	WATCHDOG_PD	Discontinue Watchdog timer during POWERDOWN mode Watchdog timer continues during POWERDOWN mode
	0	R/W	CHG_SEL	Port CHG_WAKE is connected to
				0: Dialog charger WAKE port 1: Charger SAFE_OUT

Note 1 Increases buck start-up time up to 3 ms.

Table 59: CONTROL_C

Register Address	Bit	Туре	Label	Description
0x10 CONTROL_C	7	R/W	DEF_SUPPLY	When asserted all supplies (except LDOCORE) are enabled/disabled from OTP default mode when entering sequencer Slot 0.
	6:5	R/W	SLEW_RATE	DVC slewing (bucks and LDOs) is executed at 00: 10 mV every 4.0 µs 01: 10 mV every 2.0 µs



Register Address	Bit	Туре	Label	Description
				10: 10 mV every 1.0 μs 11: 10 mV every 0.5 μs
	4	R/W	OTPREAD_EN	0: OTP read after POWERDOWN mode disabled 1: Power supplies are configured with OTP values when leaving POWERDOWN mode
	3	R/W	AUTO_BOOT	O: Start-up of power sequencer after progressing from RESET mode requires a valid wakeup event 1: PMIC automatically starts power sequencer after progressing from RESET mode
	2:0	R/W	DEBOUNCING	GPI, nONKEY and nSHUTDOWN debounce time 000: no debounce time 001: 0.1 ms 010: 1.0 ms 011: 10.2 ms 100: 51.2 ms 101: 256 ms 110: 512 ms 111: 1024 ms

Table 60: CONTROL_D

Register Address	Bit	Туре	Label	Description
0x11 CONTROL_D	7:6	R/W	BLINK_DUR	GPO10/GPO11 flashing on-time 00: 10 ms 01: 20 ms 10: 40 ms 11: 20 ms double stroke (180 ms period)
	5:3	R/W	BLINK_FRQ	GPO11/4/15 flashing frequency 000: no blinking (GPO11/14/15 state selected via GPIOxx_MODE) 001: every second 010: every two seconds 011: every four seconds 100: every 180 ms (flicker mode) 101: every two seconds enabled by VDD_START 110: every four seconds enabled by VDD_START 111: every 180 ms enabled by VDD_START Note 1
	2:0	R/W	TWDSCALE	000: Watchdog disabled 001: 1x scaling applied to two_MAX period 010: 2x 011: 4x 100: 8x 101: 16x 110: 32x 111: 64x



Note 1 Blinking from OTP settings 001 to 100 continues as long as an active charger is connected to port CHG_WAKE. In the absence of a battery charger a time limited blinking can be enabled via RES_BLINKING.

Table 61: CONTROL_E

Register Address	Bit	Туре	Label	Description
0x12 CONTROL_E	7	R/W	V_LOCK	0: Allows host writes into registers 0x81 to 0x120 1: Disables register 0x81 to 0x120 reprogramming from host interfaces
	5	R/W	PM_FB3_PIN	0: nVIB_BRAKE disabled 1: nVIB_BRAKE enabled. Feedback pin is used as an input signal to stop and start the vibration motor (active low nVIB_BRAKE)
		R/W	PM_FB2_PIN	O: Feedback pin indicates the status of regulators being selected for voltage supervision (PWR_OK) 1: Feedback pin is used as KEEP_ACT signal for the Watchdog unit
		4 R/	R/W	PM_FB1_PIN
	3	R/W	ECO_MODE	When asserted DA9063L is armed for the pulsed mode when entering RESET
	2:0	R/W	Reserved	

Table 62: CONTROL_F

Register Address	Bit	Туре	Label	Description
0x13	7:3	R/W	Reserved	
CONTROL_F	2	R/W	WAKE_UP	If set to 1 PMU wakes up from POWERDOWN mode. The bit is cleared back to 0 automatically 16 sec after entering ACTIVE mode
	1	R/W	SHUTDOWN	If set to 1 the sequencer powers down to RESET mode. The bit is cleared back to 0 automatically when entering the RESET mode
	0	R/W	WATCHDOG	If set to 1 watchdog timer is reset. The bit is cleared back to 0 automatically.

Table 63: PD_DIS

Register Address	Bit	Type	Label	Description
0x14 PD_DIS	7	R/W	PMCONT_DIS	0: SYS_EN, PWR_EN, PWR1_EN enabled during power down 1: Auto-Disable of SYS_EN, PWR_EN and



Register Address	Bit	Туре	Label	Description
				PWR1_EN during POWERDOWN mode and force the detection hidden transition when re-enabling the control from ports
	6:4		Reserved	
	3	R/W	HS2WIRE_DIS	0: HS 2-WIRE not disabled during power down 1: Auto-disable of HS 2-WIRE interface during POWERDOWN mode
	2	R/W	Reserved	
	1	R/W	GPADC_PAUSE	0: ADC measurements continue during power down as configured 1: Auto-PAUSE auto measurements on A0, A1, A2 and A3 and manual measurement during POWERDOWN mode; if no autonomous automeasurements are required (Vsys from vibration motor driver) switch off the ADC completely
	0	R/W	GPI_DIS	O: GPIO extender enabled during power down 1: Auto-disable of features configured as GPI pins during POWERDOWN mode and force the detection hidden transition when re-enabling the pin

Note 1 When the related ID is configured to be 1 < PD_DIS_STEP ≤ MAX_COUNT the value of the above controls define whether functions are switched on when entering POWERDOWN mode from POR or wait until ID PD_DIS_STEP is processed.

A.2.2 **GPIO Control**

Table 64: GPIO0-1

Register Address	Bit	Туре	Label	Description
0x15 GPIO0-1	7	R/W	GPIO1_WEN	O: Passive to active transition triggers a wakeup 1: Wakeup suppressed
	6	R/W	GPIO1_TYPE	0: GPI: active low GPO: supplied from VDD_IO1 1: GPI: active high GPO: supplied from VDD_IO2
	5:4	R/W	GPIO1_PIN	PIN assigned to 00: ADCIN2/1.2 V comparator 01: GPI (optional regulator HW control) 10: GPO mode controlled (Open drain) 11: GPO mode controlled (Push-pull)
	3	R/W	GPIO0_WEN	O: Passive to active transition triggers a wakeup 1: Wakeup suppressed
	2	R/W	GPIO0_TYPE	0: GPI: active low GPO: supplied from VDD_IO1



Register Address	Bit	Туре	Label	Description
				1: GPI: active high GPO: supplied from VDD_IO2
	1:0	R/W	GPIO0_PIN	PIN assigned to
				00: ADCIN1
				01: GPI
				10: GPO mode controlled (Open drain)
				11: GPO mode controlled (Push-pull)

Table 65: GPIO2-3

Register Address	Bit	Туре	Label	Description
0x16 GPIO2-3	7	R/W	GPIO3_WEN	Passive to active transition triggers a wakeup Wakeup suppressed
	6	R/W	GPIO3_TYPE	0: GPI: active low GPO: supplied from VDD_IO1 1: GPI: active high GPO: supplied from VDD_IO2
	5:4	R/W	GPIO3_PIN	PIN assigned to 00: Reserved 01: GPI 10: GPO mode controlled (Open drain) 11: GPO mode controlled (Push-pull)
	3	R/W	GPIO2_WEN	Passive to active transition triggers a wakeup Wakeup suppressed
	2	R/W	GPIO2_TYPE	0: GPI: active low GPO: supplied from VDD_IO1 1: GPI: active high GPO: supplied from VDD_IO2
	1:0	R/W	GPIO2_PIN	PIN assigned to 00: ADCIN3 01: GPI (optional regulator HW control) 10: GPO Sequencer controlled (Push-pull) 11: GPO mode controlled (Push-pull)

Table 66: GPIO4-5

Register Address	Bit	Туре	Label	Description
0x17 GPIO4-5	7	R/W	GPIO5_WEN	O: Passive to active transition triggers a wakeup History 1: Wakeup suppressed
	6	R/W	GPIO5_TYPE	0: GPI: active low GPO: supplied from VDD_IO1 1: GPI: active high GPO: supplied from VDD_IO2
	5:4	R/W	GPIO5_PIN	PIN assigned to 00: Reserved 01: GPI 10: GPO mode controlled (Open drain) 11: GPO mode controlled (Push-pull)



Register Address	Bit	Туре	Label	Description
	3	R/W	GPIO4_WEN	O: Passive to active transition triggers a wakeup History 1: Wakeup suppressed
	2	R/W	GPIO4_TYPE	0: GPI: active low GPO: supplied from VDD_IO1 1: GPI: active high GPO: supplied from VDD_IO2
	1:0	R/W	GPIO4_PIN	PIN assigned to 00: CORE_SWS 01: GPI 10: GPO mode controlled (Open drain) 11: GPO mode controlled (Push-pull)

Table 67: GPIO6-7

Register Address	Bit	Туре	Label	Description
0x18 GPIO6-7	7	R/W	GPIO7_WEN	O: Passive to active transition triggers a wakeup 1: Wakeup suppressed
	6	R/W	GPIO7_TYPE	0: GPI: active low GPO: supplied from VDD_IO1 1: GPI: active high GPO: supplied from VDD_IO2
	5:4	R/W	GPIO7_PIN	PIN assigned to 00: Reserved 01: GPI 10: GPO Sequencer controlled (Pushpull) 11: GPO mode controlled (Push-pull)
	3	R/W	GPIO6_WEN	Passive to active transition triggers a wakeup Wakeup suppressed
	2	R/W	GPIO6_TYPE	0: GPI: active low GPO: supplied from VDD_IO1 1: GPI: active high GPO: supplied from VDD_IO2
	1:0	R/W	GPIO6_PIN	PIN assigned to 00: PERI_SWS 01: GPI 10: GPO mode controlled (Open drain) 11: GPO mode controlled (Push-pull)

Table 68: GPIO8-9

Register Address	Bit	Туре	Label	Description
0x19 GPIO8-9	7	R/W	GPIO9_WEN	O: Passive to active transition triggers a wakeup 1: Wakeup suppressed
	6	R/W	GPIO9_TYPE	0: GPI/PWR_EN: active low GPO: supplied from VDD_IO1



Register Address	Bit	Туре	Label	Description
				1: GPI/PWR_EN: active high GPO: supplied from VDD_IO2
	5:4	R/W	GPIO9_PIN	PIN and status register bit assigned to
				00: GPI with PWR_EN 01: GPI 10: GPO Sequencer controlled (Push-pull) 11: GPO mode controlled (Push-pull)
	3	R/W	GPIO8_WEN	O: Passive to active transition triggers a wakeup 1: Wakeup suppressed
	2	R/W	GPIO8_TYPE	0: GPI/SYS_EN: active low GPO: supplied from external/VDD_IO1
				1: GPI/SYS_EN: active high GPO: supplied from VDD_IO2
	1:0	R/W	GPIO8_PIN	PIN and status register bit assigned to
				00: GPI with SYS_EN
				10: GPO Sequencer controlled (Push-pull) 11: GPO mode controlled (Push-pull)

Table 69: GPIO10-11

Register Address	Bit	Туре	Label	Description
0x1A GPIO10-11	7	R/W	GPIO11_WEN	O: Passive to active transition triggers a wakeup History 1: Wakeup suppressed
	6	R/W	GPIO11_TYPE	0: GPI: active low GPO: supplied from external/VDD_IO1 1: GPI: active high GPO: supplied from VDD_IO2
	5:4	R/W	GPIO11_PIN	PIN assigned to 00: GPO (Open drain, with optional blinking) 01: GPI 10: GPO Sequencer controlled (Pushpull) 11: GPO mode controlled (Push-pull)
	3	R/W	GPIO10_WEN	O: Passive to active transition triggers a wakeup 1: Wakeup suppressed
	2	R/W	GPIO10_TYPE	0: GPI/PWR1_EN: active low GPO: supplied from external/VDD_IO1 1: GPI/PWR1_EN: active high GPO: supplied from VDD_IO2
	1:0	R/W	GPIO10_PIN	PIN and status register bit assigned to 00: GPI with PWR1_EN 01: GPI 10: GPO (Open drain) 11: GPO mode controlled (Push-pull)



Table 70: GPIO12-13

Register Address	Bit	Туре	Label	Description
0x1B GPIO12-13	7	R/W	GPIO13_WEN	O: Passive to active transition triggers a wakeup History is a wakeup suppressed
	6	R/W	GPIO13_TYPE	0: GPI: active low GPO/GP_FB1: supplied from external/VDD_IO1 1: GPI: active high GPO/GP_FB1: supplied from VDD_IO2
	5:4	R/W	GPIO13_PIN	PIN and status register bit assigned to 00: GPO controlled by state of GP_FB1 (EXT_WAKEUP/READY) (Push-pull) 01: GPI (optional regulator HW control) 10: GPO controlled by state of GP_FB1 (EXT_WAKEUP/READY) (Open drain) 11: GPO mode controlled (Push-pull)
	3	R/W	GPIO12_WEN	O: Passive to active transition triggers a wakeup 1: Wakeup suppressed
	2	R/W	GPIO12_TYPE	0: GPI: active low GPO/ nVDD_FAULT/Vsys monitor: supplied from VDD_IO1 1: GPI: active high GPO/DD_FAULT/Vsys monitor: supplied from VDD_IO2
	1:0	R/W	GPIO12_PIN	PIN assigned to 00: nVDD_FAULT (Push-pull) 01: GPI 10: GPO controlled by the state of V _{SYS} monitor (Push-pull) 11: GPO mode controlled (Push-pull)

Table 71: GPIO14-15

Register Address	Bit	Туре	Label	Description
0x1C GPIO14-15	7	R/W	GPIO15_WEN	O: Passive to active transition triggers a wakeup 1: Wakeup suppressed
	6	R/W	GPIO15_TYPE	0: GPI: active low GPO: supplied from external/VDD_IO1 1: GPI: active high GPO: supplied from VDD_IO2
	5:4	R/W	GPIO15_PIN	PIN assigned to 00: GPO (Open drain, with optional blinking) 01: GPI 10: CLK (configured via GPIO14_PIN) 11: GPO mode controlled (Open drain)
	3	R/W	GPIO14_WEN	0: Passive to active transition triggers a wakeup 1: Wakeup suppressed



Register Address	Bit	Туре	Label	Description
	2	R/W	GPIO14_TYPE	0: GPI: active low GPO: supplied from external/VDD_IO1 DATA/CLK supplied from VDD_IO1 (Note 1) 1: GPI: active high GPO: supplied from VDD_IO2 DATA/CLK supplied from VDD_IO2 (Note 1)
	1:0	R/W	GPIO14_PIN	PIN assigned to 00: GPO(Open drain, with optional blinking) 01: GPI 10: DATA (assigns GPIO15_PIN to CLK) 11: GPO mode controlled (Push-pull)

Note 1 When using as HS 2-WIRE interface input logic levels are derived from VDDCORE.

Table 72: GPIO_MODE0_7

Register Address	Bit	Туре	Label	Description
0x1D GPIO_MODE0_7	7	R/W	GPIO7_ MODE	O: GPI: debouncing off GPO: Sets output to low level (active low for sequencer control) 1: GPI: debouncing on GPO: Sets output to high level (active high for sequencer control)
	6	R/W	GPIO6_ MODE	O: GPI: debouncing off GPO: Sets output to low level (active low for sequencer control) 1: GPI: debouncing on GPO: Sets output to high level (active high for sequencer control)
	5	R/W	GPIO5_ MODE	O: GPI: debouncing off GPO: Sets output to low level 1: GPI: debouncing on GPO: Sets output to high level
	4	R/W	GPIO4_ MODE	O: GPI: debouncing off GPO: Sets output to low level(active low for sequencer control) 1: GPI: debouncing on GPO: Sets output to high level (active high for sequencer control)
	3	R/W	GPIO3_ MODE	O: GPI: debouncing off GPO: Sets output to low level (active low for sequencer control) 1: GPI: debouncing on GPO: Sets output to high level (active high for sequencer control)
	2	R/W	GPIO2_MODE	O: GPI: debouncing off GPO: Sets output to low level 1: GPI: debouncing on GPO: Sets output to high level



Register Address	Bit	Туре	Label	Description
	1	R/W	GPIO1_ MODE	O: GPI: debouncing off GPO: Sets output to low level 1: GPI: debouncing on GPO: Sets output to high level
	0	R/W	GPIO0_ MODE	O: GPI: debouncing off GPO: Sets output to low level 1: GPI: debouncing on GPO: Sets output to high level

Table 73: GPIO_MODE8_15

Register Address	Bit	Туре	Label	Description	
0x1E GPIO_MODE8_15	7	R/W	GPIO15_ MODE	O: GPI: debouncing off GPO: Sets output to low level (active high for blinking) 1: GPI: debouncing on GPO: Sets output to high level (active low for blinking)	
	6	R/W	GPIO14_ MODE	O: GPI: debouncing off GPO: Sets output to low level (active high for blinking) 1: GPI:debouncing on GPO: Sets output to high level (active low for blinking)	
	5	5	R/W	GPIO13_ MODE	0: GPI: debouncing off GPO: Sets output to low level (active low for GP_FB1) 1: GPI: debouncing on GPO: Sets output to high level (active high for GP_FB1)
	4	R/W	GPIO12_ MODE	0: GPI: debouncing off GPO: Sets output to low level (active low for nVDD_FAULT, V _{SYS} monitor state) 1: GPI: debouncing on GPO: Sets output to high level (active high for nVDD_FAULT, V _{SYS} monitor state)	
	3	3 R/W	R/W	GPIO11_ MODE	O: GPI: : debouncing off GPO: Sets output to low level (active high for blinking) 1: GPI: : debouncing on GPO: Sets output to high level (active low for blinking)
	2	R/W	GPIO10_ MODE	O: GPI/PWR1_EN: debouncing off GPO: Sets output to low level 1: GPI/PWR1_EN: debouncing on GPO: Sets output to high level	
	1	R/W	GPIO9_ MODE	O: GPI/PWR_EN: debouncing off GPO: Sets output to low level (active low for sequencer control) 1: GPI/PWR_EN debouncing on GPO: Sets output to high level (active high for sequencer control)	



Register Address	Bit	Туре	Label	Description
	0	R/W	GPIO8_ MODE	0: GPI/SYS_EN: debouncing off GPO: Sets output to low level 1: GPI/SYS_EN: debouncing on GPO: Sets output to high level

Table 74: Reserved

Register Address	Bit	Туре	Label	Description
0x1F	7:0		Reserved	

A.2.3 Regulator Control

Table 75: BCORE2_CONT

Register Address	Bit	Туре	Label	Description
0x20 BCORE2_CONT Note 1	7	R/W	Reserved	
	6:5	R/W	VBCORE2_GPI	GPIO select target voltage VBCORE2_A on passive to active transition, selects target voltage VBCORE2_B on active to passive transition (ramping)
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	4	R/W	Reserved	
	3	R/W	BCORE2_CONF	Sequencer target state of BCORE2_EN
	2:1	R/W	BCORE2_GPI	GPIO enables BUCKCORE2 on passive to active state transition, disables BUCKCORE2 on active to passive state transition
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	0	R/W	BCORE2_EN	0: BUCKCORE2 disabled 1: BUCKCORE2 enabled

Note 1 Disabled in BUCKCORE dual-phase mode.

Table 76: BCORE1_CONT

Register Address	Bit	Туре	Label	Description
0x21	7	R/W	CORE_SW_CONF	Sequencer target state of CORE_SW_EN
BCORE1_CONT	6:5	R/W	VBCORE1_GPI	GPIO select target voltage VBCORE1_A on passive to active transition, selects target voltage VBCORE1_B on active to passive transition (ramping)
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled



Register Address	Bit	Туре	Label	Description
				10: GPIO2 controlled 11: GPIO13 controlled
	4	R/W	CORE_SW_EN	0: CORE_SW opened 1: CORE_SW closed
	3	R/W	BCORE1_CONF	Sequencer target state of BCORE1_EN
	2:1	R/W	BCORE1_GPI	GPIO enables BUCKCORE1 on passive to active state transition, disables BUCKCORE1 on active to passive state transition
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	0	R/W	BCORE1_EN	0: BUCKCORE1 disabled 1: BUCKCORE1 enabled

Table 77: BPRO_CONT

Register Address	Bit	Туре	Label	Description
0x22	7	R/W	Reserved	
BPRO_CONT	6:5	R/W	VBPRO_GPI	GPIO select target voltage VBPRO_A on passive to active transition, selects target voltage VBPRO_B on active to passive transition (ramping)
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	4	R/W	Reserved	
	3	R/W	BPRO_CONF	Sequencer target state of BPRO_EN
	2:1	R/W	BPRO_GPI	GPIO enables BUCKPRO on passive to active state transition, disables BUCKPRO on active to passive state transition 00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled
				11: GPIO13 controlled
	0	R/W	BPRO_EN	0: BUCKPRO disabled 1: BUCKPRO enabled

Table 78: BMEM_CONT

Register Address	Bit	Туре	Label	Description
0x23	7	R/W	Reserved	
BMEM_CONT	6:5	R/W	VBMEM_GPI	GPIO select target voltage VBMEM_A on passive to active transition, selects target voltage VBMEM_B on active to passive transition (ramping)



Register Address	Bit	Туре	Label	Description
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	4	R/W	Reserved	
	3	R/W	BMEM_CONF	Sequencer target state of BMEM_EN in case of being a default supply)
	2:1	R/W	BMEM_GPI	GPIO enables BUCKMEM on passive to active state transition, disables BUCKMEM on active to passive state transition 00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	0	R/W	BMEM_EN	0: BUCKMEM disabled 1: BUCKMEM enabled

Table 79: BIO_CONT

Register Address	Bit	Туре	Label	Description
0x24	7	R/W	Reserved	
BIO_CONT	6:5	R/W	VBIO_GPI	GPIO select target voltage VBIO_A on passive to active transition, selects target voltage VBIO_B on active to passive transition (ramping)
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	4	R/W	Reserved	
	3	R/W	BIO_CONF	Sequencer target state of BIO_EN
	2:1	R/W	BIO_GPI	GPIO enables BUCKIO on passive to active state transition, disables BUCKIO on active to passive state transition 00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	0	R/W	BIO_EN	0: BUCKIO disabled 1: BUCKIO enabled

Table 80: BPERI_CONT

Register Address	Bit	Туре	Label	Description
0x25	7	R/W	PERI_SW_CONF	Sequencer target state of PERI_SW_EN
BPERI_CONT	6:5	R/W	VBPERI_GPI	GPIO select target voltage VBPERI_A on passive to active transition, selects target



Register Address	Bit	Туре	Label	Description
				voltage VBPERI_B on active to passive transition (ramping)
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	4	R/W	PERI_SW_EN	0: PERI_SW opened 1: PERI_SW closed
	3	R/W	BPERI_CONF	Sequencer target state of BPERI_EN
	2:1	R/W	BPERI_GPI	GPIO enables BUCKPERI on passive to active state transition, disables BUCKPERI on active to passive state transition
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	0	R/W	BPERI_EN	0: BUCKPERI disabled 1: BUCKPERI enabled

Table 81: Reserved

Register Address	Bit	Туре	Label	Description
0x26, 0x27	7:0	R/W	Reserved	

Table 82: LDO3_CONT

Register Address	Bit	Туре	Label	Description
0x28	7	R/W	LDO3_CONF	Sequencer target state of LDO3_EN
LDO3_CONT	6:5	R/W	VLDO3_GPI	GPIO select target voltage VLDO3_A on passive to active transition, selects target voltage VLDO3_B on active to passive transition (ramping)
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	4	R/W	Reserved	
	3	R/W	LDO3_PD_DIS	C: Enable pull-down resistor No pull-down resistor in disabled mode
	2:1	R/W	LDO3_GPI	GPIO enables LDO3 on passive to active state transition, disables LDO3 on active to passive state transition
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled



Register Address	Bit	Туре	Label	Description
	0	R/W	LDO3_EN	0: LDO3 disabled 1: LDO3 enabled

Table 83: Reserved

Register Address	Bit	Туре	Label	Description
0x29 to 0x2B	7:0	R/W	Reserved	

Table 84: LDO7_CONT

Register Address	Bit	Туре	Label	Description
0x2C	7	R/W	LDO7_CONF	Sequencer target state of LDO7_EN
LDO7_CONT	6:5	R/W	VLDO7_GPI	GPIO select target voltage VLDO7_A on passive to active transition, selects target voltage VLDO7_B on active to passive transition (immediate voltage change) 00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	4	R/W	VLDO7_SEL	LDO7 voltage is selected from (immediate change): 0: VLDO7_A 1: VLDO7_B
	3	R/W	LDO7_PD_DIS	0: Enable pull-down resistor 1: No pull-down resistor in disabled mode
	2:1	R/W	LDO7_GPI	GPIO enables LDO7 on passive to active state transition, disables LDO7 on active to passive state transition 00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	0	R/W	LDO7_EN	0: LDO7 disabled 1: LDO7 enabled

Table 85: LDO8_CONT

Register Address	Bit	Туре	Label	Description
0x2D	7	R/W	LDO8_CONF	Sequencer target state of LDO8_EN
LDO8_CONT	6:5	R/W	VLDO8_GPI	GPIO select target voltage VLDO8_A on passive to active transition, selects target voltage VLDO8_B on active to passive transition (immediate voltage change) 00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled



Register Address	Bit	Type	Label	Description
	4	R/W	VLDO8_SEL	LDO8 voltage is selected from (immediate change):
				0: VLDO8_A 1: VLDO8_B
	3	R/W	LDO8_PD_DIS	C: Enable pull-down resistor No pull-down resistor in disabled mode
	2:1	R/W	LDO8_GPI	GPIO enables LDO8 on passive to active state transition, disables LDO8 on active to passive state transition
				00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	0	R/W	LDO8_EN	0: LDO8 disabled 1: LDO8 enabled

Table 86: LDO9_CONT

Register Address	Bit	Туре	Label	Description
0x2E	7	R/W	LDO9_CONF	Sequencer target state of LDO9_EN
LDO9_CONT	6:5	R/W	VLDO9_GPI	GPIO select target voltage VLDO9_A on passive to active transition, selects target voltage VLDO9_B on active to passive transition (immediate voltage change) 00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	4	R/W	VLDO9_SEL	LDO9 voltage is selected from (immediate change): 0: VLDO9_A 1: VLDO9_B
	3	R/W	LDO9_PD_DIS	0: Enable pull-down resistor 1: No pull-down resistor in disabled mode
	2:1	R/W	LDO9_GPI	GPIO enables LDO9 on passive to active state transition, disables LDO9 on active to passive state transition O0: Not controlled by GPIO (sequencer control) O1: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	0	R/W	LDO9_EN	0: LDO9 disabled 1: LDO9 enabled

Table 87: Reserved

Register Address	Bit	Туре	Label	Description
0x2F	7:0	R/W	Reserved	



Table 88: LDO11_CONT

Register Address	Bit	Туре	Label	Description
0x30	7	R/W	LDO11_CONF	Sequencer target state of LDO11_EN
LDO11_CONT	6:5	R/W	VLDO11_GPI	GPIO select target voltage VLDO11_A on passive to active transition, selects target voltage VLDO11_B on active to passive transition (immediate voltage change) 00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	4 R/W	R/W	VLDO11_SEL	LDO11 voltage is selected from (immediate change): 0: VLDO11_A 1: VLDO11_B
		R/W	LDO11_PD_DIS	0: Enable pull-down resistor 1: No pull-down resistor in disabled mode
	2:1	R/W	LDO11_GPI	GPIO enables LDO11 on passive to active state transition, disables LDO11 on active to passive state transition 00: Not controlled by GPIO (sequencer control) 01: GPIO1 controlled 10: GPIO2 controlled 11: GPIO13 controlled
	0	R/W	LDO11_EN	0: LDO11 disabled 1: LDO11 enabled

Table 89: VIB

Register Address	Bit	Туре	Label	Description
0x31	7:6	R/W	Reserved	
VIB	5:0	R/W	VIB_SET	000000: OFF-BREAK, NMOS on, PMOS off 000001: 47.55 mV 000010: 95.1 mV Average output level set in a range of 0 to 3 V in steps of 3 V/63 111111: 3.0 V

Table 90: DVC_1

Register Address	Bit	Туре	Label	Description
0x32 DVC_1	7	R/W	VLDO3_SEL	LDO3 voltage is selected from (ramping): 0: VLDO3_A 1: VLDO3_B
	6:5	R/W	Reserved	



Register Address	Bit	Туре	Label	Description
	4	R/W	VBPERI_SEL	BUCKPERI voltage is selected from (ramping): 0: VBPERI_A 1: VBPERI_B
	3	R/W	VBMEM_SEL	BUCKMEM voltage is selected from (ramping): 0: VBMEM_A 1: VBMEM_B
	2	R/W	VBPRO_SEL	BUCKPRO voltage is selected from (ramping): 0: VBPRO_A 1: VBPRO_B
	1	R/W	VBCORE2_SEL	BUCKCORE2 voltage is selected from (ramping): 0: VBCORE2_A 1: VBCORE2_B
	0	R/W	VBCORE1_SEL	BUCKCORE1 voltage is selected from (ramping): 0: VBCORE1_A 1: VBCORE1_B

Table 91: DVC_2

Register Address	Bit	Туре	Label	Description
0x33	7:1	R/W	Reserved	
DVC_2	0	R/W	VBIO_SEL	BUCKIO voltage is selected from (ramping): 0: VBIO_A 1: VBIO_B



A.2.4 GPADC

Table 92: ADC_MAN

Register Address	Bit	Туре	Label	Description
0x34	7:6	R/W	Reserved	
ADC_MAN	5	R/W	ADC_MODE	0: Measurement sequence interval 10 ms (economy mode) 1: Measurement sequence interval 1 ms
	4	R/W	ADC_MAN	Perform manual conversion. Bit is reset to 0 when conversion is complete.
	3:0	R/W	ADC_MUX	Manual measurement selects: 0000: VSYS port 0001: ADCIN1 0010: ADCIN2 0011: ADCIN3 0100: internal T-Sense 0101: reserved0110: reserved 0111: reserved 1000: Group 1 regulators voltage 1001: Group 2 regulators voltage 1010: Group 3 regulators voltage > 1010: reserved

Table 93: ADC_CONT

Register Address	Bit	Туре	Label	Description
0x35 ADC_CONT	7	R/W	COMP1V2_EN	0: Disable 1.2 V comparator at ADCIN2 1: Enable 1.2 V comparator
	6	R/W	AD3_ISRC_EN	0: Disable ADCIN3 current source 1: Enable ADCIN3 current source
	5	R/W	AD2_ISRC_EN	0: Disable ADCIN2 current source 1: Enable ADCIN2 current source
	4 3 2	R/W	AD1_ISRC_EN	0: Disable ADCIN1 current source 1: Enable ADCIN1 current source
		R/W	AUTO_AD3_EN	0: ADCIN3 auto-measurements disabled 1: ADCIN3 auto-measurements enabled
		R/W	AUTO_AD2_EN	0: ADCIN2 auto-measurements disabled 1: ADCIN2 auto-measurements enabled
		1 R/W	R/W	AUTO_AD1_EN
	0	R/W	AUTO_VSYS_EN	0: VSYS auto-measurements disabled when charger/vibration motor driver is off 1: VSYS auto-measurements enabled

Table 94: VSYS_MON

Register Address	Bit	Туре	Label	Description
0x36 VSYS_MON	7:0	R/W	VSYS_MON	VSYS_MON threshold setting (8-bit).



	00000000 corresponds to 2.5 V
	11111111 corresponds to 5.5 V

A.2.5 ADC Results

Table 95: ADC_RES_L

Register Address	Bit	Туре	Label	Description
0x37	7:6	R	ADC_RES_LSB	10-bit manual conversion result (2 LSBs)
ADC_RES_L	5:0	R	Reserved	

Table 96: ADC_RES_H

Register Address	Bit	Туре	Label	Description
0x38 ADC_RES_H	7:0	R	ADC_RES_MSB	10-bit manual conversion result (8 MSBs)

Table 97: VSYS_RES

Register Address	Bit	Туре	Label	Description
0x39 VSYS_RES	7:0	R	VSYS_RES	0x00 – 0xFF: Auto VSYS conversion result (A0) 0x00 corresponds to 2.5 V 0xFF corresponds to 5.5 V

Table 98: ADCIN1_RES

Register Address	Bit	Туре	Label	Description
0x3A ADCIN1_RES	7:0	R	ADCIN1_RES	0x00 – 0xFF: Auto ADC ADCIN1 conversion result

Table 99: ADCIN2_RES

Register Address	Bit	Туре	Label	Description
0x3B ADCIN2_RES	7:0	R	ADCIN2_RES	0x00 – 0xFF: Auto ADC ADCIN2 conversion result

Table 100: ADCIN3_RES

Register Address	Bit	Туре	Label	Description
0x3C ADCIN3_RES	7:0	R	ADCIN3_RES	0x00 – 0xFF: Auto ADC ADCIN3 conversion result



Table 101: MON_A8_RES

Register Address	Bit	Туре	Label	Description
0x3D MON_A8_RES	7:0	R	MON_A8_RES	0x00 – 0xFF: Regulator output voltage monitor 1 (A8) conversion result
				0x00 corresponds to 0.0 V 0xFF corresponds to 5.0 V

Table 102: MON_A9_RES

Register Address	Bit	Туре	Label	Description
0x3E MON_A9_RES	7:0	R	MON_A9_RES	0x00 – 0xFF: Regulator output voltage monitor 2 (A9) conversion result 0x00 corresponds to 0.0 V 0xFF corresponds to 5.0 V

Table 103: MON_A10_RES

Register Address	Bit	Туре	Label	Description
0x3F MON_A10_RES	7:0	R	MON_A10_RES	0x00 – 0xFF: Regulator output voltage monitor 3 (A10) conversion result
				0x00 corresponds to 0.0 V 0xFF corresponds to 5.0 V

Table 104: Reserved

Register Address	Bit	Туре	Label	Description
0x40 to 0x4F	7:0		Reserved	

Table 105: Copmic_S to Copmic_E

Register Address	Bit	Туре	Label	Description
0x50 CoPMIC_S	7:0	R	Reserved	Reserved for Co-PMIC
0x67 CoPMIC_E	7:0	R	Reserved	Reserved for Co-PMIC

Table 106: CHG_Co_S to CHG_Co_E

Register Address	Bit	Туре	Label	Description
0x68 CHG_Co_S	7:0	R	Reserved	Reserved for companion charger
0x7F CHG_Co_E	7:0	R	Reserved	Reserved for companion charger



A.3 Register Page 1

Table 107: PAGE_CON

Register Address	Bit	Туре	Label	Description
0x80	7	RW	REVERT	See register 0x00, Table 43
PAGE_CON	6	RW	WRITE_MODE	
	5:3	RW	Reserved	
	2:0	RW	REG_PAGE	

A.3.1 Power Sequencer

Table 108: SEQ

Register Address	Bit	Туре	Label	Description
0x81 SEQ	7:4	R/W	NXT_SEQ_START	Start time slot for first sequencing after being modified via register write
	3:0	R	SEQ_POINTER	Actual pointer position (time slot) of power sequencer



Table 109: SEQ_TIMER

Register Address	Bit	Туре	Label	Description
0x82 SEQ_TIMER	7:4	R/W	SEQ_DUMMY	0000: 32 μs 0001: 64 μs 0010: 96 μs 0011: 128 μs 0100: 160 μs 0101: 192 μs 0110: 224 μs 0111: 256 μs 1000: 288 μs 1001: 384 μs 1010: 448 μs 1011: 512 μs 1100: 1.024 ms 1101: 2.048 ms 1111: 8.192 ms
	3:0	R/W	SEQ_TIME	0000: 32 μs 0001: 64 μs 0010: 96 μs 0010: 128 μs 0100: 160 μs 0101: 192 μs 0110: 224 μs 0111: 256 μs 1000: 288 μs 1001: 384 μs 1010: 448 μs 1011: 512 μs 1100: 1.024 ms 1101: 2.048 ms 1111: 8.192 ms

Table 110: Reserved

Register Address	Bit	Туре	Label	Description
0x83	7:0	R/W	Reserved	

Table 111: ID_4_3

Register Address	Bit	Туре	Label	Description
0x84	7:4	R/W	Reserved	
ID_4_3	3:0	R/W	LDO3_STEP	Power sequencer time slot for LDO3 control

Table 112: Reserved

Register Address	Bit	Туре	Label	Description
0x85	7:0		Reserved	

Datasheet Revision 2.4 11-Feb-2022



Table 113: ID_8_7

Register Address	Bit	Туре	Label	Description
0x86	7:4	R/W	LDO8_STEP	Power sequencer time slot for LDO8 control
ID_8_7	3:0	R/W	LDO7_STEP	Power sequencer time slot for LDO7 control

Table 114: ID_10_9

Register Address	Bit	Type	Label	Description
0x87	7:4	R/W	Reserved	
ID_10_9	3:0	R/W	LDO9_STEP	Power sequencer time slot for LDO9 control

Table 115: ID_12_11

Register Address	Bit	Туре	Label	Description
0x88 ID_12_11	7:4	R/W	PD_DIS_STEP	Power sequencer time slot for control of blocks to be disabled/paused during POWERDOWN mode
	3:0	R/W	LDO11_STEP	Power sequencer time slot for LDO11 control

Table 116: ID_14_13

Register Address	Bit	Туре	Label	Description
0x89 ID_14_13	7:4	R/W	BUCKCORE2_STEP	Power sequencer time slot for control of BUCKCORE2 (disabled in BUCKCORE dual phase mode)
	3:0	R/W	BUCKCORE1_STEP	Power sequencer time slot for control of BUCKCORE1

Table 117: ID_16_15

Register Address	Bit	Туре	Label	Description
0x8A ID_16_15	7:4	R/W	BUCK_IO_STEP	Power sequencer time slot for control of BUCKPRO
	3:0	R/W	BUCKPRO_STEP	Power sequencer time slot for control of BUCKPRO

Table 118: ID_18_17

Register Address	Bit	Туре	Label	Description
0x8B ID_18_17	7:4	R/W	BUCKPERI_STEP	Power sequencer time slot for control of BUCKPERI
	3:0	R/W	BUCKMEM_STEP	Power sequencer time slot for control of BUCKMEM



Table 119: Reserved

Register Address	Bit	Туре	Label	Description
0x8C	7:0	R/W	Reserved	

Table 120: ID_22_21

Register Address	Bit	Туре	Label	Description
0x8D ID_22_21	7:4	R/W	GP_FALL1_STEP	Power sequencer time slot for falling edge control of GPO2
	3:0	R/W	GP_RISE1_STEP	Power sequencer time slot for rising edge control of GPO2

Table 121: ID_24_23

Register Address	Bit	Туре	Label	Description
0x8E ID_24_23	7:4	R/W	GP_FALL2_STEP	Power sequencer time slot for falling edge control of GPO7
	3:0	R/W	GP_RISE2_STEP	Power sequencer time slot for rising edge control of GPO7

Table 122: ID_26_25

Register Address	Bit	Туре	Label	Description
0x8F ID_26_25	7:4	R/W	GP_FALL3_STEP	Power sequencer time slot for falling edge control of GPO8
	3:0	R/W	GP_RISE3_STEP	Power sequencer time slot for rising edge control of GPO8

Table 123: ID_28_27

Register Address	Bit	Туре	Label	Description
0x90 ID_28_27	7:4	R/W	GP_FALL4_STEP	Power sequencer time slot for falling edge control of GPO9
	3:0	R/W	GP_RISE4_STEP	Power sequencer time slot for rising edge control of GPO9

Table 124: ID_30_29

Register Address	Bit	Туре	Label	Description
0x91 ID_30_29	7:4	R/W	GP_FALL5_STEP	Power sequencer time slot for falling edge control of GPO11
	3:0	R/W	GP_RISE5_STEP	Power sequencer time slot for rising edge control of GPO11



Table 125: ID_32_31

Register Address	Bit	Туре	Label	Description
0x92	7:4	R/W	Reserved	
ID_32_31	3:0	R/W	WAIT_STEP	Power sequencer time slot that gates the progress with state of GPI10 (or used a dedicated delay timer)

Table 126: Reserved

Register Address	Bit	Туре	Label	Description
0x93, 0x94	7:0	R/W	Reserved	

Table 127: SEQ_A

Register Address	Bit	Туре	Label	Description
0x95 SEQ_A	7:4	R/W	POWER_END	OTP pointer to last supply of domain POWER
	3:0	R/W	SYSTEM_END	OTP pointer to last supply of domain SYSTEM

Table 128: SEQ_B

Register Address	Bit	Туре	Label	Description
0x96 SEQ_B	7:4	R/W	PART_DOWN	OTP pointer for partial POWERDOWN mode
	3:0	R/W	MAX_COUNT	OTP pointer to last supply of domain POWER1

Table 129: WAIT

Register Address	Bit	Туре	Label	Description
0x97 WAIT	7:6	R/W	WAIT_DIR	00: No wait during power sequencing 01: Wait during power-up sequence 10:: Wait during power-down sequence 11: Wait during power-up and power-down sequence
	5	R/W	TIME_OUT	0: No time limit 1: 500 ms time out for waiting GPIO10 to get active
	4	R/W	WAIT_MODE	0: Wait for GPIO10 to be active 1: Timer mode (start timer and wait for expire)



Register Address	Bit	Туре	Label	Description
	3:0	R/W	WAIT_TIME	0000: 0.0 μs 0001: 512 μs 0010: 1.0 ms 0011: 2.0 ms 0110: 4.1 ms 0101: 8.2 ms 0110: 16.4 ms 0111: 32.8 ms 1000: 65.5 ms 1001: 128 ms 1010: 256 ms 1011: 512 ms 1100: 1.0 s 1101: 2.1 s 1110: 4.2 s 1111: 8.4 s

Table 130: Reserved

Register Address	Bit	Туре	Label	Description
0x98	7:0	R/W	Reserved	

Table 131: RESET

Register Address	Bit	Туре	Label	Description
0x99 RESET	7:6	R/W	RESET_EVENT	RESET timer started by 00: EXT_WAKEUP 01: SYS_UP 10: PWR_UP 11: leaving PMIC RESET state (do not use in combination with nRES_MODE = 1)
	5:0	R/W	RESET_TIMER	000000: 0.000 ms 000001: 1.024 ms 000010: 2.048 ms 000011: 3.072 ms 000100: 4.096 ms 000101: 5.120 ms 011110: 30.720 ms 011111: 31.744 ms 100000: 32.768 ms 100001: 65.536 ms 100010: 98.304 ms 111101: 983.040 ms 111110: 1015.808 ms 111111: 1048.576 ms



A.3.2 Regulator Settings

Table 132: BUCK_ILIM_A

Register Address	Bit	Туре	Label	Description
0x9A BUCK_ILIM_A	7:4	R/W	BMEM_ILIM	BUCKMEM current limit (all limits x2 in MERGE mode)
				0000:1500 mA 0001:1600 mA 0010:1700 mA 0011:1800 mA 0100:1900 mA 0101:2000 mA 0110:2100 mA 0111:2200 mA 1000:2300 mA 1001:2400 mA 1011:2600 mA 1011:2600 mA 1111:2800 mA 1101:2900 mA
	3:0	R/W	BIO_ILIM	BUCKIO current limit 0000:1500 mA 0001:1600 mA 0010:1700 mA 0011:1800 mA 0100:1900 mA 0101:2000 mA 0110:2100 mA 0111:2200 mA 1000:2300 mA 1001:2400 mA 1011:2600 mA 1011:2600 mA 1111:2600 mA 1111:3000 mA

Table 133: BUCK_ILIM_B

Register Address	Bit	Туре	Label	Description
0x9B	7:4	R/W	BPERI_ILIM	BUCKPERI current limit
BUCK_ILIM_B				0000:1500 mA
				0001:1600 mA
				0010:1700 mA
				0011:1800 mA
				0100:1900 mA
				0101:2000 mA
				0110:2100 mA
				0111:2200 mA
				1000:2300 mA
				1001:2400 mA
				1010:2500 mA
				1011:2600 mA
				1100:2700 mA
				1101:2800 mA



Register Address	Bit	Туре	Label	Description
				1110:2900 mA 1111:3000 mA
	3:0	R/W	BPRO_ILIM	BUCKPRO current limit (all limits x2 in full-current mode) 0000:500 mA 0001:600 mA 0010:700 mA 0011:800 mA 0100:900 mA 0101:1000 mA 0110:1100 mA 0110:1100 mA 1010:1500 mA 1001:1400 mA 1011:1600 mA 1011:1600 mA 1101:1800 mA 1101:1800 mA 1101:1900 mA

Table 134: BUCK_ILIM_C

Register Address	Bit	Туре	Label	Description
0x9C BUCK_ILIM_C	7:4	R/W	BCORE2_ILIM	BUCKCORE2 current limit (all limits x2 in full-current mode)
				0000:500 mA 0001:600 mA 0010:700 mA 0011:800 mA 0100:900 mA 0101:1000 mA 0110:1100 mA 0111:1200 mA 1000:1300 mA 1001:1400 mA
				1010:1500 mA 1011:1600 mA
				1100:1700 mA 1101:1800 mA
				1110:1900 mA 1111:2000 mA



Register Address	Bit	Туре	Label	Description
	3:0	R/W	BCORE1_ILIM	BUCKCORE1 current limit (all limits x2 in full-current mode)
				0000:500 mA 0001:600 mA 0010:700 mA 0011:800 mA 0100:900 mA 0101:1000 mA 0110:1100 mA 0111:1200 mA 1000:1300 mA 1001:1400 mA 1010:1500 mA
				1100:1700 mA 1101:1800 mA 1110:1900 mA
				1111:2000 mA

Table 135: BCORE2_CONF

Register Address	Bit	Туре	Label	Description
0x9D BCORE2_CONF	7:6	R/W	BCORE2_MODE	00: Sleep/Synchronous mode controlled via voltage A and B registers 01: BUCKCORE2 always operates in Sleep mode 10: BUCKCORE2 always operates in Synchronous mode 11: BUCKCORE2 operates in Automatic mode
	5	R/W	BCORE2_PD_DIS	O: Enable pull-down resistor (automatically disabled in dual-phase mode) 1: No pull-down resistor in disabled mode
	4:3		Reserved	
	2:0	R/W	BCORE2_FB	BUCKCORE2 feedback signal is created out of: xx1: VBUCKCORE2 x1x: CORE_SWS 1xx: PERI_SWS Each switch connected to the output of the
				buck may be selected; setting 0b000 is invalid

Table 136: BCORE1_CONF

Register Address	Bit	Туре	Label	Description
0x9E BCORE1_CONF	7:6	R/W	BCORE1_MODE	00: Sleep/Synchronous mode controlled via voltage A and B registers 01: BUCKCORE1 always operates in Sleep mode 10: BUCKCORE1 always operates in Synchronous mode 11: BUCKCORE1 operates in Automatic mode



Register Address	Bit	Туре	Label	Description
	5	R/W	BCORE1_PD_DIS	0: Enable pull-down resistor 1: No pull-down resistor in disabled mode
	4:3		Reserved	
	2:0	R/W	BCORE1_FB	BUCKCORE feedback signal is created out of:
				000: BCORE_MERGE= 0: VBUCKCORE1 BCORE_MERGE= 1: Differential remote sensing via VBUCKCORE1 – VBUCKCORE2 and output capacitor voltage sense via port CORE_SWS or GP_FB_2
				xx1: VBUCKCORE1 x1x: CORE_SWS 1xx: PERI_SWS
				Each switch connected to the output of the buck may be selected; setting 0b000 disables sense voltage mixer for BUCKCORE

Table 137: BPRO_CONF

Register Address	Bit	Туре	Label	Description
0x9F BPRO_CONF	7:6	R/W	BPRO_MODE	00: Sleep/Synchronous mode controlled via voltage A and B registers 10: BUCKPRO always operates in Sleep mode 10: BUCKPRO always operates in Synchronous 11: BUCKPRO operates in Automatic mode
	5	R/W	BPRO_PD_DIS	C: Enable pull-down resistor No pull-down resistor in disabled mode
	4	R/W	BPRO_VTT_EN	0: Buck voltage mode 1: VTT mode, buck target voltage tracks 50% of VDDQ sense port (requires BPRO_VTTR_EN to be asserted as well)
	3	R/W	BPRO_VTTR_EN	0: VTTR port is assigned to E_CMP1V2, port VDDQ provides status of E_GPI2 1: VTTR port provides 50% of VDDQ voltage
	2:0	R/W	BPRO_FB	BUCKPRO feedback signal is created out of: xx1: VBUCKPRO x1x: CORE_SWS 1xx: PERI_SWS Each switch connected to the output of the buck may be selected; setting 0b000 is invalid



Table 138: BIO_CONF

Register Address	Bit	Туре	Label	Description
0xA0 BIO_CONF	7:6	R/W	BIO_MODE	00: Sleep/Synchronous mode controlled via voltage A and B registers 10: BUCKIO always operates in Sleep mode 10: BUCKIO always operates in Synchronous 11: BUCKIO operates in Automatic mode
	5	R/W	BIO_PD_DIS	C: Enable pull-down resistor No pull-down resistor in disabled mode
	4:3		Reserved	
	2:0	R/W	BIO_FB	BUCKIO feedback signal is created out of: xx1: VBUCKBIO x1x: CORE_SWS 1xx: PERI_SWS Each switch connected to the output of the buck may be selected; setting 0b000 is invalid

Table 139: BMEM_CONF

Register Address	Bit	Туре	Label	Description
0xA1 BMEM_CONF	7:6	R/W	BMEM_ MODE	00: Sleep/Synchronous mode controlled via voltage A and B registers 01: BUCKMEM always operates in Sleep mode 10: BUCKMEM always operates in Synchronous mode 11: BUCKMEM operates in Automatic mode
	5	R/W	BMEM_PD_DIS	C: Enable pull-down resistor No pull-down resistor in disabled mode
	4:3		Reserved	
	2:0	R/W	BMEM_FB	BUCKMEM feedback signal is created out of: xx1: VBUCKMEM x1x: CORE_SWS 1xx: PERI_SWS Each switch connected to the output of the buck may be selected; setting 0b000 is invalid



Table 140: BPERI_CONF

Register Address	Bit	Туре	Label	Description
0xA2 BPERI_CONF	7:6	R/W	BPERI_ MODE	00: Sleep/Synchronous mode controlled via voltage A and B registers 01: BUCKPERI always operates in Sleep mode 10: BUCKPERI always operates in Synchronous mode 11: BUCKPERI operates in Automatic mode
	5	R/W	BPERI_PD_DIS	0: Enable pull-down resistor 1: No pull-down resistor in disabled mode
	4:3		Reserved	
	2:0	R/W	BPERI_FB	BUCKPERI feedback signal is created out of: xx1: VBUCKPERI x1x: CORE_SWS 1xx: PERI_SWS Each switch connected to the output of the buck may be selected; setting 0b0000 is invalid

Table 141: VBCORE2_A

Register Address	Bit	Туре	Label	Description
0xA3 VBCORE2_A	7	R/W	BCORE2_SL_A	0: Configures BUCKCORE2 to Synchronous mode, when selecting A voltage settings 1: Configures BUCKCORE2 to Sleep mode, when selecting A voltage settings
	6:0	R/W	VBCORE2_A	0000000: 0.30 V 0000001: 0.31 V 0000010: 0.32 V 0000011: 0.33 V 0000100: 0.34 V 0000101: 0.35 V 0100101: 0.67 V 0100111: 0.68 V 0100111: 0.69 V



Register Address	Bit	Туре	Label	Descr	iption
				0101000: 0.70 V 0101001: 0.71 V	
				 1010000: 1.10 V	
				1110011: 1.45 V 1110100: 1.46 V 1110101: 1.47 V 1110110: 1.48 V 1110111: 1.49 V 1111000: 1.50 V 1111001: 1.51 V 1111010: 1.52 V 1111011: 1.53 V 1111101: 1.55 V 1111101: 1.55 V 1111101: 1.56 V	PWM mode voltage range

Table 142: VBCORE1_A

Register Address	Bit	Туре	Label	Descr	ription
0xA4 VBCORE1_A	7	R/W	BCORE1_SL_A	0: Configures BUCK Synchronous mode, voltage settings 1: Configures BUCKC mode, when selecting	when selecting A ORE1 to Sleep
	6:0	R/W	VBCORE1_A	0000000: 0.30 V 0000001: 0.31 V 0000010: 0.32 V 0000011: 0.33 V 0000100: 0.34 V 0000101: 0.35 V 0100101: 0.67 V 0100110: 0.68 V 0100111: 0.69 V	
				0101000: 0.70 V 0101001: 0.71 V 1010000: 1.10 V 1110011: 1.45 V 1110100: 1.46 V 1110101: 1.47 V 1110101: 1.48 V 1110101: 1.50 V 1111001: 1.51 V 1111011: 1.53 V 1111101: 1.55 V 1111101: 1.56 V 1111111: 1.57 V	PWM mode voltage range



Table 143: VBPRO_A

Register Address	Bit	Туре	Label	Desci	ription
0xA5 VBPRO_A	7	R/W	BPRO_SL_A	0: Configures BUCK Synchronous mode, voltage settings 1: Configures BUCKP when selecting A voltage	when selecting A PRO to Sleep mode,
	6:0	R/W	VBPRO_A	0000000: 0.53 V 0000001: 0.54 V 0000010: 0.55 V 0000011: 0.56 V 0000100: 0.57 V 0000101: 0.58 V 	
				0010001: 0.70 V 0010010: 0.71 V 0010011: 0.72 V 0010100: 0.73 V 0010110: 0.75 V 1000011: 1.20 V 1110011: 1.68 V 1110100: 1.69 V 1110101: 1.70 V 1110111: 1.72 V	PWM mode voltage range
				1111000: 1.73 V 1111001: 1.74 V 1111010: 1.75 V 1111011: 1.76 V 1111100: 1.77 V 1111101: 1.78 V 1111111: 1.80 V	

Table 144: VBMEM_A

Register Address	Bit	Туре	Label	Description
0xA6 VBMEM_A	7	R/W	BMEM_SL_A	0: Configures BUCKMEM to Synchronous mode, when selecting A voltage settings 1: Configures BUCKMEM to Sleep mode, when selecting A voltage settings
	6:0	R/W	VBMEM_A	0000000: 0.80 V 0000001: 0.82 V 0000010: 0.84 V 0010100: 1.20 V 0111100: 2.00 V 0111101: 2.02 V 0111111: 2.04 V 0111111: 2.06 V



Register Address	Bit	Туре	Label	Description
				 1111111: 3.34 V

Table 145: VBIO_A

Register Address	Bit	Туре	Label	Description
0xA7 VBIO_A	7	R/W	BIO_SL_A	0: Configures BUCKIO to Synchronous mode, when selecting A voltage settings 1: Configures BUCKIO to Sleep mode, when selecting A voltage settings
	6:0	R/W	VBIO_A	0000000: 0.80 V 0000001: 0.82 V 0000010: 0.84 V 0010100: 1.20 V 0111100: 2.00 V 0111110: 2.02 V 0111111: 2.04 V 0111111: 3.34 V



Table 146: VBPERI_A

Register Address	Bit	Туре	Label	Description
0xA8 VBPERI_A	7	R/W	BPERI_SL_A	0: Configures BUCKPERI to Synchronous mode, when selecting A voltage settings 1: Configures BUCKPERI to Sleep mode, when selecting A voltage settings
	6:0	R/W	VBPERI_A	0000000: 0.80 V 0000001: 0.82 V 0000010: 0.84 V 0110010: 1.80 V 0111100: 2.00 V 0111101: 2.02 V 0111110: 2.04 V 0111111: 2.06 V 1111111: 3.34 V

Table 147: Reserved

Register Address	Bit	Туре	Label	Description
0xA9, 0xAA	7:0	R/W	Reserved	

Table 148: VLDO3_A

Register Address	Bit	Туре	Label	Description
0xAB VLDO3_A	7	R/W	LDO3_SL_A	O: Configures LDO to normal mode, when selecting A voltage settings 1: Configures LDO to Sleep mode, when selecting A voltage settings
	6:0	R/W	VLDO3_A	0000000: 0.90 V 0000001: 0.92 V 0000010: 0.94 V 0000011: 0.96 V 0000100: 0.98 V 0000101: 1.00 V 0000110: 1.02 V 0000111: 1.04 V 0001001: 1.08 V 0001001: 1.10 V 0001011: 1.10 V 0001101: 1.12 V 0001101: 1.16 V 0001101: 1.18 V 0001111: 1.20 V 0010001: 1.24 V 0010001: 1.26 V 0010011: 1.30 V 0010101: 1.30 V 0010101: 1.30 V 0010101: 1.32 V 1110000: 3.14 V 1110001: 3.16 V



Register Address	Bit	Туре	Label	Description
				1110010: 3.18 V
				1110011: 3.20 V
				1110100: 3.22 V
				1110101: 3.24 V
				1110110: 3.26 V
				1110111: 3.28 V
				1111000: 3.30 V
				1111001: 3.32 V
				1111010: 3.34 V
				1111011: 3.36 V
				1111100: 3.38 V
				1111101: 3.40 V
				1111110: 3.42 V
				1111111: 3.44 V

Table 149: Reserved

Register Address	Bit	Туре	Label	Description
0xAC to 0xAE	7:0	R/W	Reserved	

Table 150: VLDO7_A

Register Address	Bit	Type	Label	Description	
0xAF VLDO7_A	7	R/W	LDO7_SL_A	0: Configures LDO to normal mode, when selecting A voltage settings 1: Configures LDO to Sleep mode, when selecting A voltage settings	
	6	R/W	Reserved		
	5:0	R/W	VLDO7_A	000000: 0.90 V 00001: 0.90 V 000010: 0.90 V 000011: 0.95 V 000100: 1.00 V 000110: 1.10 V 000110: 1.10 V 000111: 1.15 V 001000: 1.20 V 001001: 1.30 V 001011: 1.35 V 001100: 1.40 V 001101: 1.45 V 001101: 1.55 V 001101: 1.55 V 001000: 1.60 V 001011: 1.75 V 010001: 1.77 V 010011: 1.75 V 010101: 1.85 V 100010: 2.50 V 10010: 2.60 V 100110: 2.70 V 100111: 2.75 V	



Register Address	Bit	Туре	Label	Description
				101000: 2.80 V
				101001: 2.85 V
				101010: 2.90 V
				101011: 2.95 V
				101100: 3.00 V
				101101: 3.05 V
				101110: 3.10 V
				101111: 3.15 V
				110000: 3.20 V
				110001: 3.25 V
				110010: 3.30 V
				110011: 3.35 V
				110100: 3.40 V
				110101: 3.45 V
				110110: 3.50 V
				110111: 3.55 V
				111000: 3.60 V
				>111000: 3.60 V

Table 151: VLDO8_A

Register Address	Bit	Type	Label	Description
0xB0 VLDO8_A	7	R/W	LDO8_SL_A 0: Configures LDO to normal mode, when selecting A voltage settings 1: Configures LDO to Sleep mode, wh selecting A voltage settings	
	6	R/W	Reserved	
	5:0	R/W	VLDO8_A	000000: 0.90 V 00001: 0.90 V 000010: 0.90 V 000011: 0.95 V 000100: 1.00 V 000101: 1.05 V 000110: 1.15 V 000100: 1.20 V 001001: 1.25 V 001010: 1.30 V 001011: 1.35 V 001100: 1.40 V 001111: 1.55 V 001100: 1.50 V 001111: 1.55 V 001000: 1.60 V 001011: 1.75 V 001001: 1.77 V 010011: 1.75 V 010001: 1.85 V 100010: 2.50 V 100101: 2.65 V 100100: 2.70 V 100111: 2.75 V 101000: 2.80 V 101000: 2.80 V 101001: 2.85 V



Register Address	Bit	Туре	Label	Description
				101010: 2.90 V
				101011: 2.95 V
				101100: 3.00 V
				101101: 3.05 V
				101110: 3.10 V
				101111: 3.15 V
				110000: 3.20 V
				110001: 3.25 V
				110010: 3.30 V
				110011: 3.35 V
				110100: 3.40 V
				110101: 3.45 V
				110110: 3.50 V
				110111: 3.55 V
				111000: 3.60 V
				>111000: 3.60 V

Table 152: VLDO9_A

Register Address	Bit	Type	Label	Description
0xB1 VLDO9_A	7	R/W	LDO9_SL_A 0: Configures LDO to normal mode, when selecting A voltage settings 1: Configures LDO to Sleep mode, when selecting A voltage settings	
	6	R/W	Reserved	
	5:0	R/W	VLDO9_A	000000: not used 000001: 0.95 V 000010: 0.95 V 000011: 0.95 V 000010: 1.00 V 00010: 1.05 V 000110: 1.10 V 000111: 1.15 V 001000: 1.20 V 001001: 1.33 V 001011: 1.35 V 001100: 1.40 V 001101: 1.45 V 001101: 1.55 V 001101: 1.55 V 001000: 1.60 V 001111: 1.75 V 010000: 1.60 V 010011: 1.75 V 010010: 1.80 V 010011: 1.85 V 100010: 2.50 V 100101: 2.55 V 100100: 2.60 V 100111: 2.75 V 101000: 2.80 V 101001: 2.85 V 101000: 2.80 V 101001: 2.90 V 101011: 2.95 V



Register Address	Bit	Туре	Label	Description	
Register Address	Bit	Туре	Label	Description 101100: 3.00 V 101101: 3.05 V 101110: 3.10 V 101111: 3.15 V 110000: 3.20 V 110001: 3.25 V 110010: 3.30 V 110011: 3.35 V 110100: 3.40 V	
				110100: 3.45 V 110110: 3.50 V 110111: 3.55 V 111000: 3.60 V >111000: 3.60 V	

Table 153: Reserved

Register Address	Bit	Туре	Label	Description
0xB2	7:0	R/W	Reserved	

Table 154: VLDO11_A

Register Address	Bit	Type	Label	Description	
0xB3 VLDO11_A	7:4	R/W	LDO11_SL_A	0: Configures LDO to normal mode, when selecting A voltage settings 1: Configures LDO to Sleep mode, when selecting A voltage settings	
	6	R/W	Reserved		
	5:0	R/W	VLDO11_A	000000: 0.90 V 000001: 0.90 V 000010: 0.90 V 000011: 0.95 V 000100: 1.00 V 000110: 1.05 V 000110: 1.10 V 000111: 1.15 V 001000: 1.20 V 001001: 1.25 V 001101: 1.35 V 001101: 1.45 V 001101: 1.45 V 001101: 1.55 V 001101: 1.55 V 010000: 1.60 V 010011: 1.75 V 010010: 1.70 V 010011: 1.85 V 010001: 1.85 V 010010: 2.50 V 100101: 2.55 V 100100: 2.60 V 100111: 2.75 V	



Register Address	Bit	Туре	Label	Description
				101000: 2.80 V
				101001: 2.85 V
				101010: 2.90 V
				101011: 2.95 V
				101100: 3.00 V
				101101: 3.05 V
				101110: 3.10 V
				101111: 3.15 V
				110000: 3.20 V
				110001: 3.25 V
				110010: 3.30 V
				110011: 3.35 V
				110100: 3.40 V
				110101: 3.45 V
				110110: 3.50 V
				110111: 3.55 V
				111000: 3.60 V
				>111000: 3.60 V



Table 155: VBCORE2_B

Register Address	Bit	Туре	Label	Desci	ription
0xB4 VBCORE2_B	7	R/W	BCORE2_SL_B	0: Configures BUCKCORE2 to Synchronous mode, when selecting B voltage settings 1: Configures BUCKCORE2 to Sleep mode, when selecting B voltage settings	
	6:0	R/W	VBCORE2_B	0000000: 0.30 V 0000001: 0.31 V 0000010: 0.32 V 0000011: 0.33 V 0000100: 0.34 V 0000101: 0.67 V 0100110: 0.68 V 0100111: 0.69 V 0101000: 0.70 V 0101001: 0.71 V 0111100: 0.90 V 1110011: 1.45 V 111010: 1.46 V 111010: 1.48 V 111011: 1.49 V 111000: 1.50 V 111101: 1.51 V 111101: 1.53 V 111110: 1.55 V 111110: 1.56 V 111111: 1.57 V	PWM mode voltage range

Table 156: VBCORE1_B

Register Address	Bit	Туре	Label	Descr	ription
0xB5 VBCORE1_B	7	R/W	BCORE1_SL_B	O: Configures BUCKCORE1 to Synchronous mode, when selecting B voltage settings 1: Configures BUCKCORE1 to Sleep mode, when selecting B voltage settings	
	6:0	R/W	VBCORE1_B	0000000: 0.30 V 0000001: 0.31 V 0000010: 0.32 V 0000011: 0.33 V 0000100: 0.34 V 0000101: 0.35 V 0100101: 0.67 V 0100110: 0.68 V 0100111: 0.69 V PWM mode voltage range	



Register Address	Bit	Туре	Label	Description	
				0111100: 0.90 V	
				1110011: 1.45 V 1110100: 1.46 V 1110101: 1.47 V 1110110: 1.48 V 1110111: 1.49 V 1111000: 1.50 V 1111001: 1.51 V 111101: 1.52 V 111101: 1.53 V 1111101: 1.55 V 111110: 1.56 V 111111: 1.57 V	

Table 157: VBPRO_B

Register Address	Bit	Туре	Label	Descr	ription
0xB6 VBPRO_B	7	R/W	BPRO_SL_B	0: Configures BUCKF mode, when selecting 1: Configures BUCK when selecting B vo	B voltage settings PRO to Sleep mode,
	6:0	R/W	VBPRO_B	0000000: 0.53 V 0000001: 0.54 V 0000010: 0.55 V 0000011: 0.56 V 0000100: 0.57 V 0000101: 0.58 V 0010000: 0.69 V	
				0010001: 0.70 V 0010010: 0.71 V 0010011: 0.72 V 0010100: 0.73 V 0010101: 0.74 V 0010110: 0.75 V 1000011: 1.20 V 1110011: 1.68 V 1110100: 1.69 V	PWM mode voltage
				1110101: 1.70 V 1110110: 1.71 V 1110111: 1.72 V 1111000: 1.73 V 1111001: 1.74 V 1111010: 1.75 V 1111011: 1.76 V 1111100: 1.77 V 1111101: 1.78 V 1111110: 1.79 V 1111111: 1.80 V	range



Table 158: VBMEM_B

Register Address	Bit	Туре	Label	Description
0xB7 VBMEM_B	7	R/W	BMEM_SL_B	Configures BUCKMEM to Synchronous mode, when selecting B voltage settings Configures BUCKMEM to Sleep mode, when selecting B voltage settings
	6:0	R/W	VBMEM_B	0000000: 0.80 V 0000001: 0.82 V 0000010: 0.84 V 0010100: 1.20 V 0111100: 2.00 V 0111110: 2.02 V 0111111: 2.04 V 0111111: 3.34 V

Table 159: VBIO_B

Register Address	Bit	Туре	Label	Description
0xB8 VBIO_B	7	R/W	BIO_SL_B	0: Configures BUCKIO to Synchronous mode, when selecting B voltage settings 1: Configures BUCKIO to Sleep mode, when selecting B voltage settings
	6:0	R/W	VBIO_B	0000000: 0.80 V 0000001: 0.82 V 0000010: 0.84 V 0010100: 1.20 V 0111100: 2.00 V 0111101: 2.02 V 0111110: 2.04 V 0111111: 2.06 V 1111111: 3.34 V

Table 160: VBPERI_B

Register Address	Bit	Туре	Label	Description
0xB9 VBPERI_B	7	R/W	BPERI_SL_B	O: Configures BUCKPERI to Synchronous mode, when selecting A voltage settings 1: Configures BUCKPERI to Sleep mode, when selecting B voltage settings
	6:0	R/W	VBPERI_B	0000000: 0.80 V 0000001: 0.82 V 0000010: 0.84 V
				0110010: 1.80 V
				 0111100: 2.00 V
				0111101: 2.02 V 0111110: 2.04 V
				0111111: 2.06 V



Register Address	Bit	Туре	Label	Description
				 1111111: 3.34 V

Table 161: Reserved

Register Address	Bit	Туре	Label	Description
0xBA, 0xBB	7:0	R/W	Reserved	

Table 162: VLDO3_B

Register Address	Bit	Туре	Label	Description
0xBC VLDO3_B	7	R/W	LDO3_SL_B	0: Configures LDO to normal mode, when selecting B voltage settings 1: Configures LDO to Sleep mode, when selecting B voltage settings
	6:0	R/W	VLDO3_B	0000000: 0.90 V 0000010: 0.92 V 0000011: 0.96 V 0000010: 0.98 V 0000101: 1.00 V 0000110: 1.02 V 0000111: 1.04 V 0000101: 1.08 V 0001001: 1.08 V 0001010: 1.10 V 0001010: 1.14 V 0001101: 1.16 V 0001101: 1.16 V 0001111: 1.20 V 0001101: 1.18 V 0001111: 1.20 V 0010001: 1.24 V 0010001: 1.26 V 001001: 1.28 V 001001: 1.30 V 001010: 1.30 V 001010: 3.16 V 111000: 3.14 V 111001: 3.20 V 111010: 3.22 V 111010: 3.24 V 111010: 3.32 V 111010: 3.33 V 111101: 3.34 V 111101: 3.34 V 111101: 3.34 V 111101: 3.40 V 111111: 3.44 V



Table 163: Reserved

Register Address	Bit	Туре	Label	Description
0xBD to 0xBF	7:0	R/W	Reserved	

Table 164: VLDO7_B

Register Address	Bit	Туре	Label	Description
0xC0 VLDO7_B	7	R/W	LDO7_SL_B	0: Configures LDO to normal mode, when selecting B voltage settings 1: Configures LDO to Sleep mode, when selecting B voltage settings
	6	R/W	Reserved	



Register Address	Bit	Туре	Label	Description
	5:0	R/W	VLDO7_B	000000: 0.90 V
			_	000001: 0.90 V
				000010: 0.90 V
				000011: 0.95 V
				000100: 1.00 V
				000101: 1.05 V
				000110: 1.10 V
				000111: 1.15 V
				001000: 1.20 V
				001001: 1.25 V
				001010: 1.30 V
				001011: 1.35 V
				001100: 1.40 V
				001101: 1.45 V
				001110: 1.50 V
				001111: 1.55 V
				010000: 1.60 V
				010001: 1.65 V
				010010: 1.70 V
				010011: 1.75 V
				010100: 1.80 V
				010101: 1.85 V
				 100010: 2.50 V
				100011: 2.55 V
				100100: 2.60 V
				100101: 2.65 V
				100110: 2.70 V
				100111: 2.75 V
				101000: 2.80 V
				101001: 2.85 V
				101010: 2.90 V
				101011: 2.95 V
				101100: 3.00 V
				101101: 3.05 V
				101110: 3.10 V
				101111: 3.15 V
				110000: 3.20 V
				110001: 3.25 V
				110010: 3.30 V
				110011: 3.35 V
				110100: 3.40 V
				110101: 3.45 V
				110110: 3.50 V
				110111: 3.55 V
				111000: 3.60 V
	1			>111000: 3.60 V



Table 165: VLDO8_B

Register Address	Bit	Туре	Label	Description
0xC1 VLDO8_B	7	R/W	LDO8_SL_B	O: Configures LDO to normal mode, when selecting B voltage settings 1: Configures LDO to Sleep mode, when selecting B voltage settings
	6	R/W	Reserved	
	5:0	R/W	VLDO8_B	000000: 0.90 V 000001: 0.90 V 000010: 0.90 V 000011: 0.95 V 000100: 1.00 V 000110: 1.10 V 000110: 1.10 V 000111: 1.15 V 001000: 1.20 V 001001: 1.25 V 001010: 1.30 V 001101: 1.35 V 001101: 1.45 V 001101: 1.45 V 001101: 1.55 V 011000: 1.60 V 010011: 1.75 V 010010: 1.80 V 010011: 1.85 V 100010: 2.50 V 100101: 2.55 V 100100: 2.60 V 100111: 2.75 V 101000: 2.80 V 101101: 2.90 V 101101: 2.95 V 101101: 2.95 V 101100: 3.00 V 101111: 3.15 V 111000: 3.20 V 110011: 3.35 V 111001: 3.35 V 110110: 3.35 V 110110: 3.45 V 110110: 3.55 V 111010: 3.50 V
				110110: 3.50 V 110111: 3.55 V



Table 166: VLDO9_B

Register Address	Bit	Туре	Label	Description
0xC2 VLDO9_B	7	R/W	LDO9_SL_B	O: Configures LDO to normal mode, when selecting B voltage settings 1: Configures LDO to Sleep mode, when selecting B voltage settings
	6	R/W	Reserved	
	5:0	R/W	VLDO9_B	000000: 0.95 V 000010: 0.95 V 000011: 0.95 V 0000110: 1.00 V 000101: 1.00 V 000110: 1.10 V 000111: 1.15 V 001000: 1.20 V 001001: 1.25 V 001001: 1.35 V 001101: 1.35 V 001101: 1.45 V 001101: 1.45 V 001101: 1.55 V 001101: 1.55 V 001010: 1.60 V 001111: 1.55 V 010000: 1.60 V 010001: 1.65 V 010010: 1.80 V 010011: 1.85 V 100010: 2.50 V 100101: 2.55 V 100100: 2.60 V 100110: 2.70 V 100111: 2.75 V 101100: 2.80 V 10111: 2.95 V 101101: 2.90 V 10111: 3.05 V 101101: 3.05 V 101101: 3.25 V 101101: 3.25 V 101101: 3.35 V 111000: 3.20 V 110011: 3.25 V 110010: 3.30 V 110111: 3.35 V 110101: 3.45 V 110101: 3.45 V 110101: 3.55 V 111010: 3.50 V
				>111000: 3.60 V



Table 167: Reserved

Register Address	Bit	Туре	Label	Description
0xC3	7:0	R/W	Reserved	

Table 168: VLDO11_B

Bit	Туре	Label	Description
7:4	R/W	LDO1_SL_B	O: Configures LDO to normal mode, when selecting B voltage settings 1: Configures LDO to Sleep mode, when selecting B voltage settings
6	R/W	Reserved	
5:0	R/W	VLDO11_B	000000: 0.90 V 000001: 0.90 V 000010: 0.90 V 000011: 0.95 V 000100: 1.00 V 000110: 1.10 V 000110: 1.10 V 000111: 1.15 V 001001: 1.30 V 00110: 1.30 V 00110: 1.40 V 00110: 1.45 V 001100: 1.40 V 001111: 1.55 V 001100: 1.60 V 001001: 1.65 V 010000: 1.60 V 010010: 1.70 V 010011: 1.75 V 010010: 1.80 V 010101: 1.85 V 100010: 2.50 V 100100: 2.60 V 100101: 2.65 V 100110: 2.70 V 100111: 2.75 V 101000: 2.80 V 101101: 2.90 V 101101: 3.35 V 101101: 3.05 V 101101: 3.05 V 101101: 3.25 V 110000: 3.20 V 110011: 3.35 V 110100: 3.30 V 11011: 3.35 V 110100: 3.40 V 11011: 3.55 V 111000: 3.60 V >111000: 3.60 V >111000: 3.60 V
	7:4	7:4 R/W 6 R/W	7:4 R/W LDO1_SL_B 6 R/W Reserved



A.3.3 High Power GPO PWM

Table 169: GPO11_LED

Register Address	Bit	Туре	Label	Description
0xC6 GPO11_LED	7	R/W	GPO11_DIM	0: PWM ratio changes instantly 1: GPO ramps between changes in PWM ratio with 32 ms per step
	6:0	R/W	GPO11_PWM	GPO11 LED on-time (low level at GPIO 11, period 21 kHz = 95 cycles of 0.5 µs)
				000000: off 0000001: 1% 0000010: 2% (1 μs bursts) 0000011: 3% 0000100: 4% 0000101: 5% 0000110: 6% 0000111: 7% 0001000: 8% 0001001: 9% 000101: 10% 000101: 11% 0001100: 12% 0001101: 13% 0001101: 13% 0001110: 14% 0001111: 15% 0001111: 15% 0011011: 15% 0011011: 100% >1011111: 100%

Table 170: GPO14_LED

Register Address	Bit	Туре	Label	Description
0xC7 GPO14_LED	7	R/W	GPO14_DIM	0: PWM ratio changes instantly 1: GPO ramps between changes in PWM ratio with 32 ms per step
	6:0	R/W	GPO14_PWM	GPO14 LED on-time (low level at GPIO 14, period 21 kHz = 95 cycles of 0.5 μs) 0000000: off 0000001: 1% 0000010: 2% (1 μs bursts) 0000011: 3% 0000100: 4% 0000110: 5% 0000110: 6% 0000111: 7% 0001000: 8% 0001001: 9% 0001011: 11% 0001101: 12% 0001101: 13% 0001111: 15% 0001000: 16%



Register Address	Bit	Туре	Label	Description
				1011111: 100% >1011111: 100%

Table 171: GPO15_LED

Register Address	Bit	Туре	Label	Description
0xC8 GPO15_LED	7	R/W	GPO15_DIM	0: PWM ratio changes instantly 1: GPO ramps between changes in PWM ratio with 32 ms per step
	6:0	R/W	GPO15_PWM	GPO15 LED on-time (low level at GPIO 15, period 21 kHz = 95 cycles of 0.5 µs)
				000000: off 0000001: 1% 0000010: 2% (1 μs bursts) 0000011: 3% 0000100: 4% 0000101: 5% 0000110: 6% 0000111: 7% 0001000: 8% 0001001: 9% 000101: 10% 000101: 11% 0001100: 12% 0001101: 13% 0001101: 13% 0001110: 14% 0001111: 15% 0001111: 15% 0011011: 15% 0111111: 100%

A.3.4 **GPADC** Thresholds

Table 172: ADC_CFG

Register Address	Bit	Туре	Label	Description
0xC9 ADC_CFG	7	R/W	ADCIN3_DEB	0: ADCIN3: debouncing off 1: ADCIN3: debouncing on
	6	R/W	ADCIN2_DEB	0: ADCIN2: debouncing off 1: ADCIN2: debouncing on
	5	R/W	ADCIN1_DEB	0: ADCIN1: debouncing off 1: ADCIN1: debouncing on
	4	R/W	ADCIN3_CUR	ADCIN3 current source:
				0: 10 μA 1: 40 μA
	3:2	R/W	ADCIN2_CUR	ADCIN2 current source:
				00: 1 μΑ
				01: 2.5 μA 10: 10 μA
				11: 40 µA
	1:0	R/W	ADCIN1_CUR	ADCIN1 current source:



Register Address	Bit	Туре	Label	Description
				00: 1 μA
				01: 2.5 μA
				10: 10 μΑ
				11: 40 µA

Table 173: AUTO1_HIGH

Register Address	Bit	Туре	Label	Description
0xCA AUTO1_HIGH	7:0	R/W	AUTO1_HIGH	00000000 – 11111111: ADCIN1 high level threshold

Table 174: AUTO1_LOW

Reg	ister Address	Bit	Туре	Label	Description
А	0xCB UTO1_LOW	7:0	R/W	AUTO1_LOW	00000000 – 11111111: ADCIN1 low level threshold

Table 175: AUTO2_HIGH

Register Address	Bit	Туре	Label	Description
0xCC AUTO2_HIGH	7:0	R/W	AUTO2_HIGH	00000000 – 11111111: ADCIN2 high level threshold

Table 176: AUTO2_LOW

Register Address	Bit	Туре	Label	Description
0xCD AUTO2_LOW	7:0	R/W	AUTO2_LOW	00000000 – 11111111: ADCIN2 low level threshold

Table 177: AUTO3 HIGH

Register Address	Bit	Туре	Label	Description
0xCE AUTO3_HIGH	7:0	R/W	AUTO3_HIGH	00000000 – 11111111: ADCIN3 high level threshold

Table 178: AUTO3_LOW

Register Address	Bit	Туре	Label	Description
0xCF AUTO3_LOW	7:0	R/W	AUTO3_LOW	00000000 – 11111111: ADCIN3 low level threshold

Table 179: Copmic_S to Copmic_E

Register Address	Bit	Туре	Label	Description
0xD0 CoPMIC_S	7:0	R	Reserved	Reserved for Co-PMIC
0xDF CoPMIC_E	7:0	R	Reserved	Reserved for Co-PMIC



Table 180: CHG_Co_S to CHG_Co_E

Register Address	Bit	Туре	Label	Description
0xE0 CHG_Co_S	7:0	R	Reserved	Reserved for companion charger
0xFF CHG_Co_E	7:0	R	Reserved	Reserved for companion charger



A.4 Register Page 2

Table 181: PAGE_CON

Register Address	Bit	Туре	Label	Description
0x100	7	RW	REVERT	See register 0x00, Table 43
PAGE_CON	6	RW	WRITE_MODE	
	5:3	RW	Reserved	
	2:0	RW	REG_PAGE	

A.4.1 OTP

Table 182: OTP_CONT

Register Address	Bit	Туре	Label	Description	
0x101 OTP_CONT	7	R	GP_WRITE_DIS	O: Enables write access to GP_ID registers 1: GP_ID_0to GP_ID_9 registers are 'read only'	
	6	R	OTP_CONF_LOC K	0: Registers 0x0A to 0x36 and 0x82 to 0xCF are not locked for OTP programming (only for evaluation samples) 1: OTP registers 0x0A to 0x36 and 0x82 to 0xCF are locked in OTP (set for all mass production parts, no further fusing possible)	
	3 R/	5 R		OTP_APPS_LOCK	0: Registers 0x104 to 0x117are not locked for OTP programming (only for evaluation samples) 1: OTP registers 0x104 to 0x117 are locked in OTP (set for all mass production parts, no further fusing possible)
		R	OTP_GP_LOCK	0: Registers 0x120 to 0x134 are not locked for OTP programming 1: Registers 0x120 to 0x134 are locked in OTP (no further fusing possible if once fused with 1)	
		R/W	PC_DONE	Asserted from PowerCommander SW after emulated OTP read has finished (control shared with Co-PMIC), automatically cleared when leaving emulated OTP read	
		R/W	OTP_APPS_RD	Reads on assertion application specific registers (0x104 to 0x117 and OTP_APPS_LOCK) from OTP	
	1	R/W	OTP_GP_RD	Reads on assertion device specific registers 0x120 to 0x134 (plus GP_WRITE_DIS and OTP_GP_LOCK) from OTP	
	0	R/W	OTP_TIM	OTP read timing 0: normal read 1: marginal read (for OTP fuse verification)	



Table 183: OTP_ADDR

Register Address	Bit	Туре	Label	Description
0x102 OTP_ADDR	7:0	R/W	OTP_ADDR	OTP Array address (shared with Companion ICs)

Table 184: OTP_DATA

Register Address	Bit	Туре	Label	Description
0x103 OTP_DATA	7:0	R/W	OTP_DATA	OTP read/write data (shared with Companion ICs) OTP_DATA written to OTP_ADDR selects the IC and accepts unlock sequence (1 + 3 bytes)

A.4.2 Customer Trim and Configuration

Table 185: T_OFFSET

Register Address	Bit	Туре	Label	Description
0x104 T_OFFSET	7:0	R	T_OFFSET	10000000 – 01111111: signed two's complement calibration offset for junction temperature measurement (loaded from the OTP memory, must be programmed during production)

Table 186: INTERFACE

Register Address	Bit	Type Note 1	Label	Description
0x105 INTERFACE	7:4	R	IF_BASE_ADDR	4 MSB of 2-WIRE control interfaces base address XXXX0000 10110100 = 0xB4 write address of HS 2-WIRE interface (page 0 and 1) 10110101 = 0xB5 read address of HS 2-WIRE interface (page 0 and 1) 10110110 = 0xB6 write address of HS 2-WIRE interface (page 2 and 3) 10110111 = 0xB7 read address of HS 2-WIRE interface (page 2 and 3) Code '0000' is reserved for unprogrammed OTP (triggers start-up with hardware default interface address)
	3:0	R	Reserved	

Note 1 The interface configuration can be written/modified only for unmarked samples which do not have the control OTP_APPS_LOCK asserted/fused.



Table 187: CONFIG_A

Register Address	Bit	Туре	Label	Description
0x106	7:4	R/W	Reserved	
CONFIG_A	3	R/W	IRQ_TYPE	nIRQ output is: 0: Active low 1: Active high (invert signal)
	2	R/W	PM_O_TYPE	nRESET, nIRQ output are: 0: Push-pull 1: Open drain (requires external pull-up resistor)
	1	R/W	PM_O_V	E_GPI_2, COMP1V2, nRESET, nIRQ are supplied from: 0: VDD_IO1 1: VDD_IO2
	0	R/W	PM_I_V	nONKEY, nOFF, nSHUTDOWN, SYS_EN, PWR_EN, PWR1_EN, KEEP_ACT, nVIB_BRAKE are supplied from: 0: VDDCORE 1: VDD_IO2

Table 188: CONFIG_B

Register Address	Bit	Туре	Label	Description
0x107	7	R/W	Reserved	
CONFIG_B	6:4	R/W	VDD_HYST_ADJ	Hysteresis adjust of VDD_FAULT comparator (VDD_FAULT_UPPER) in 50 mV steps 000: 100 mV 001: 150 mV 111: 450 mV
	3:0	R/W	VDD_FAULT_ADJ	Setting of VDD_FAULT_LOWER comparator in 50 mV steps 0000: 2.50 V 0001: 2.55 V 0110: 2.80 V 1110: 3.20 V 1111: 3.25 V

Table 189: CONFIG_C

Register Address	Bit	Туре	Label	Description
0x108 CONFIG_C	7	R/W	BPERI_CLK_INV	BUCKPERI clock polarity 0: Normal 1: Inverted
	6	R/W	BIO_CLK_INV	BUCKIO clock polarity



Register Address	Bit	Туре	Label	Description
				0: Normal 1: Inverted
	5	R/W	BMEM_CLK_INV	BUCKMEM clock polarity 0: Normal 1: Inverted
	4	R/W	BPRO_CLK_INV	BUCKPRO clock polarity (should be configured opposite to BUCKMEM clock polarity) 0: Normal 1: Inverted
	3	R/W	BCORE1_CLK_INV	BUCKCORE1 clock polarity (BUCKCORE2 always runs on opposite clock polarity) 0: Normal 1: Inverted
	2	R/W	BUCK_DISCHG	Enable active discharge of buck rails
	1:0	R/W	Reserved	

Table 190: CONFIG_D

Register Address	Bit	Туре	Label	Description
0x109 CONFIG_D	7	R/W	GP_FB3_TYPE	GP_FB3 output is: 0: Active low 1: Active high (invert signal)
	6	R/W	GP_FB2_TYPE	GP_FB2 output is: 0: Active low (invert signal, push-pull for PWR_OK) 1: Active high (open drain for PWR_OK)
	5	R/W	FORCE_RESET	Asserts port nRESET in case of being set
	4	R/W	HS_IF_HSM	Enables continuous High Speed mode on HS 2-WIRE IF (no master code required)
	3	R/W	HS_IF_FMP	Selects fast-mode+ timings for HS 2-WIRE IF if asserted
	2	R/W	SYSTEM_EN_RD	During second OTP read control SYSTEM_EN is 0: updated from OTP 1: not changed
	1	R/W	nIRQ_MODE	nIRQ will be asserted from events during POWERDOWN mode (and modes lower than ACTICE) 0: immediately 1: after powering up to ACTIVE mode
	0	R/W	GPI_V	GPIs (not configured as Power Manager control inputs) and HS 2-WIRE IF are supplied from: 0: VDDCORE 1: VDD_IO2



Table 191: CONFIG_E

Register Address	Bit	Туре	Label	Description
0x10A CONFIG_E	7	R/W	PERI_SW_AUTO	Selects PERI_SW (during powering up): 0: configured from PERI_SW_CONF 1: enabled
	6	R/W	CORE_SW_AUTO	Selects CORE_SW (during powering up): 0: configured from CORE_SW_CONF 1: enabled
	5	R/W	BPERI_AUTO	Selects BUCKPERI (during powering up): 0: configured from BPERI_CONF 1: enabled
	4	R/W	BIO_AUTO	Selects BUCKIO (during powering up): 0: configured from BIO_CONF 1: enabled
	3	R/W	BMEM_AUTO	Selects BUCKMEM (during powering up): 0: configured from BMEM_CONF 1: enabled
	2	R/W	BPRO_AUTO	Selects BUCKPRO (during powering up): 0: configured from BPRO_CONF 1: enabled
	1	R/W	BCORE2_AUTO	Selects BUCKCORE2 (during powering up): 0: configured from BCORE2_CONF 1: enabled
	0	R/W	BCORE1_AUTO	Selects BUCKCORE1 (during powering up): 0: configured from BCORE1_CONF 1: enabled

Table 192: CONFIG_F

Register Address	Bit	Туре	Label	Description
0x10B CONFIG_F	7	R/W	LDO11_BYP	0: LDO11 is configured for regulator mode 1: LDO11 bypass mode enabled
	6	R/W	LDO8_BYP	0: LDO8 is configured for regulator mode 1: LDO8 bypass mode enabled
	5	R/W	LDO7_BYP	0: LDO7 is configured for regulator mode 1: LDO7 bypass mode enabled
	4	R/W	Reserved	
	3	R/W	LDO3_BYP	0: LDO3 is configured for regulator mode 1: LDO3 bypass mode enabled
	2	R/W	LDO11_AUTO	Selects LDO11 (during powering up): 0: configured from LDO11_CONF 1: enabled
	1	R/W	Reserved	
	0	R/W	LDO9_AUTO	Selects LDO9 (during powering up):



Register Address	Bit	Туре	Label	Description
				0: configured from LDO9_CONF 1: enabled

Table 193: CONFIG_G

Register Address	Bit	Туре	Label	Description
0x10C CONFIG_G	7	R/W	LDO8_AUTO	Selects LDO8 (during powering up): 0: configured from LDO8_CONF 1: enabled
	6	R/W	LDO7_AUTO	Selects LDO7 (during powering up): 0: configured from LDO7_CONF 1: enabled
	5:3	R/W	Reserved	
	2	R/W	LDO3_AUTO	Selects LDO3 (during powering up): 0: configured from LDO3_CONF 1: enabled
	1:0	R/W	Reserved	

Table 194: CONFIG_H

Register Address	Bit	Туре	Label	Description
0x10D CONFIG_H	7	R/W	BUCK_MERGE	Has to be set if the outputs of BUCKMEM and BUCKIO are merged towards a single coil; the control from BUCKIO registers is disabled
	6	R/W	BCORE1_OD	If set, BUCKCORE1 changes to full-current mode (double pass device and current limit)
	5	R/W	BCORE2_OD	If set, BUCKCORE2 changes to full-current mode (double pass device and current limit)
	4	R/W	BPRO_OD	If set, BUCKPRO changes to full-current mode (double pass device and current limit)
	3	R/W	BCORE_MERGE	Has to be set if the outputs of BUCKCORE1 and BUCKCORE2 are merged towards a dual phase buck; the control from BUCKCORE2 registers is disabled
	2	R/W	MERGE_SENSE	In case BUCKCORE is merged and configured for remote sensing the output capacitor voltage rail is routed to port:
				0: GP_FB_2 (setting disables normal GP_FB_2 functionality) 1: CORE_SWS
				Note: In case MERGE_SENSE is asserted all Bxxx_FB control settings 0bx1x are invalid
	1	R/W	LDO8_MODE	0: LDO mode (external capacitor required)



Register Address	Bit	Туре	Label	Description
				Vibration motor driver (no external capacitor)
	0	R/W	PWM_CLK	0: 2.0 MHz (31.25 kHz repetition frequency) 1: 1.0 MHz (15.6 kHz repetition frequency)

Table 195: CONFIG_I

Register Address	Bit	Туре	Label	Description	
0x10E CONFIG_I	7	R/W	LDO_SD	If asserted LDO3, 7, 8 and 11 will shut down after current limit was hit for more than 200 ms	
	6	R/W	INT_SD_MODE	Shut down sequence from internal fault condition is: 0: Normal 1: Fast (skipping seq and dummy slot timers)	
	5	R/W	HOST_SD_MODE	Shut down sequence from SHUTDOWN (register bit or port nSHUTDOWN) is: 0: Normal 1: Fast (skipping seq and dummy slot timers)	
	4	R/W	KEY_SD_MODE	User triggered (nONKEY, GPIO14/15) shutdown sequence is:	
				0: Normal 1: PMU POR: triggers an instant disable of all regulators incl. LDOCORE (FAULTLOG registers remain unchanged). After leaving POR automatically the RESET_DURATION timer must expire before starting a power-up sequence.	
	3	R/W	GPI14_15_SD	0: Disables shutdown via parallel assertion of GPI14 and GPI15 1: Enables shutdown via GPI14 & GPI15	
	2	2 R/W	R/W	nONKEY_SD	nONKEY is configured 0: without shutdown via long press of nONKEY 1: with shutdown via long press of nONKEY
	1:0	R/W	nONKEY_PIN	nONKEY is configured to 00: Port mode 01: Key mode with key lock during SW triggered POWERDOWN mode 10: Key mode with key locked autonomous powering down (multi-functional key) 11: Key mode with autonomous powering down to partial or key locked full POWERDOWN mode (dedicated power key) Details: see Section 6.1.1	



Table 196: CONFIG_J

Register Address	Bit	Туре	Label	Description
0x10F CONFIG_J	7	R/W	IF_RESET	Enables automatic reset of all control interfaces when port nSHUTDOWN is asserted
	6	R/W	TWOWIRE_TO	Enables automatic reset of 2-WIRE-IF in case of clock ceases to toggle for >19 ms
	5:4	R/W	RESET_DURATION	Power controller stays in RESET mode for minimum duration of 00: 20 ms 01: 100 ms 10: 500 ms 11: 1000 ms
	3:2	R/W	SHUT_DELAY	Long press time threshold for shutdown feature from nONKEY and GPIO14/15: 00: KEY_DELAY + 0 s 01: KEY_DELAY + 4 s 10: KEY_DELAY + 5 s 11: KEY_DELAY + 6 s
	1:0	R/W	KEY_DELAY	Long press threshold for nONKEY lock: 00: nONKEY_LOCK after 1 s 01: nONKEY_LOCK after 1.5 s 10: nONKEY_LOCK after 2 s 11: nONKEY_LOCK after 7 s

Note 1 This setting may trigger glidges on regulator outputs and disable the automatic RESET/POR of slave PMUs).

Table 197: CONFIG_K

Register Address	Bit	Туре	Label	Description
0x110 CONFIG_K	7	R/W	GPIO7_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	6	R/W	GPIO6_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	5	R/W	GPIO5_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	4	R/W	GPIO4_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)



Register Address	Bit	Туре	Label	Description
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	3	R/W	GPIO3_PUPD	O: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor) 1: GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	2	R/W	GPIO2_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	1	R/W	GPIO1_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	0	R/W	GPIO0_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)

Table 198: CONFIG_L

Register Address	Bit	Туре	Label	Description
0x111 CONFIG_L	7	R/W	GPIO15_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	6	R/W	GPIO14_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	5	R/W	GPIO13_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)



Register Address	Bit	Туре	Label	Description
	4	R/W	GPIO12_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	3	R/W	GPIO11_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				1: GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	2	R/W	GPIO10_PUPD	O: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor) 1: GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	1	R/W	GPIO9_PUPD	0: GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)
	0	R/W	GPIO8_PUPD	GPI: pull-down resistor disabled GPO (open drain): pull-up resistor disabled (external pull-up resistor)
				1: GPI: pull-down resistor enabled GPO (open drain): pull-up resistor enabled (supply rail selected via GPIOx_TYPE)

Table 199: CONFIG_M

Register Address	Bit	Туре	Label	Description
0x112 CONFIG_M	7:4	R/W	OSC_FRQ	Offset for internal HF oscillator frequency 0x8: -10.67% 0xF: -1.33% 0x0: 0.00% 0x1: 1.33% 0x7: 9.33%
	3:0	R/W	Reserved	

Table 200: Reserved

Register Address	Bit	Туре	Label	Description
0x113	7:0	R/W	Reserved	



Table 201: MON_REG_1

Register Address	Bit	Туре	Label	Description
0x114 MON_REG_1	7:6	R/W	UVOV_DELAY	Range comparison is enabled after regulator enable: 00: immediately 01: with one measurement delay 10: with two measurements delay 11: with four measurements delay
	5:4	R/W	MON_MODE	Regulator monitor executes 00: Under-voltage/Over-voltage lockout with an E_REG_UVOV event (nIRQ assertion) and regulator shutdown from an output voltage being out of the selected range 01: Normal auto measurement with an E_REG_UVOV event (nIRQ assertion) from any finished auto measurement on A8, A9 and A10 10: Burst auto measurement with an E_REG_UVOV event (nIRQ assertion) generated after the time slot of A10 has been processed 11: reserved
	3	R/W	MON_DEB	0: Regulator monitor (A8, 9, 10): debouncing off 1: Regulator monitor (A8, 9, 10): debouncing on
	2	R/W	MON_RES	Control requires M_REG_UVOV = 1: 1: Enables assertion of nRESET from out-of-range detection Note: It is not recommended to assert this control inside OTP
	1:0	R/W	MON_THRES	Regulator Monitor Threshold 00: Approx = 25 % 01: Approx = 12.5 % 10: Approx = 6.25 % 11: Approx = 3.125 %

Table 202: MON_REG_2

Register Address	Bit	Туре	Label	Description
0x115	7	R/W	LDO8_MON_EN	Enable LDO8 regulator monitoring
MON_REG_2	6	R/W	LDO7_MON_EN	Enable LDO7 regulator monitoring
	5:3	R/W	Reserved	
	2	R/W	LDO3_MON_EN	Enable LDO3 regulator monitoring
	1:0	R/W	Reserved	



Table 203: MON_REG_3

Register Address	Bit	Туре	Label	Description
0x116	7:3	R/W	Reserved	
MON_REG_3	2	R/W	LDO11_MON_EN	Enable LDO11 regulator monitoring
	1	R/W	Reserved	
	0	R/W	LDO9_MON_EN	Enable LDO9 regulator monitoring

Table 204: MON_REG_4

Register Address	Bit	Туре	Label	Description
0x117	7	R/W	BPERI_MON_EN	Enable BUCKPERI regulator monitoring
MON_REG_4	6	R/W	BMEM_MON_EN	Enable BUCKMEM regulator monitoring
	5	R/W	BIO_MON_EN	Enable BUCKIO regulator monitoring
	4	R/W	BPRO_MON_EN	Enable BUCKPRO regulator monitoring
	3	R/W	BCORE2_MON_EN	Enable BUCKCORE2 regulator monitoring
	2	R/W	BCORE1_MON_EN	Enable BUCKCORE1 regulator monitoring
	1:0	R/W	Reserved	

Table 205: MON_REG_5

Register Address	Bit	Туре	Label	Description
0x11E	7	R	Reserved	
MON_REG_5	6:4	R	MON_A9_IDX	Latest measurement at channel A9 was: 000: none 001: BUCKIO 010: BUCKMEM 011: BUCKPERI > 011: reserved
	3	R	Reserved	
	2:0	R	MON_A8_IDX	Latest measurement at channel A8 was: 000: none 001: BUCKCORE1 010: BUCKCORE2 011: BUCKPRO 100: LDO3 101: Reserved 110: LDO11 > 110: reserved

Table 206: MON_REG_6

Register Address	Bit	Type	Label	Description
0x11F	7:3	R	Reserved	
MON_REG_6	2:0	R	MON_A10_IDX	Latest measurement at channel A10 was:



Register Address	Bit	Туре	Label	Description
				000: none 001: Reserved 010: LDO7 011: LDO8 100: LDO9 > 100: reserved

Table 207: Reserved

Register Address	Bit	Туре	Label	Description
0x120	7:0	R/W	Reserved	

Table 208: GP_ID_0

Register Address	Bit	Туре	Label	Description
0x121 GP_ID_0	7:0	R/W Note 1	GP_0	Data from fuse array (OTP)

Note 1 Write access can be disabled by OTP if required.

Table 209: GP_ID_1

	gister Idress	Bit	Туре	Label	Description
_	x122 ^_ID_1	7:0	R/W Note 1	GP_1	Data from fuse array (OTP)

Note 1 Write access can be disabled by OTP if required.

Table 210: GP_ID_2

Register Address	Bit	Туре	Label	Description
0x123 GP_ID_2	7:0	R/W Note 1	GP_2	Data from fuse array (OTP)

Note 1 Write access can be disabled by OTP if required.

Table 211: GP_ID_3

Register Address	Bit	Туре	Label	Description
0x124 GP_ID_3	7:0	R/W Note 1	GP_3	Data from fuse array (OTP)

Note 1 Write access can be disabled by OTP if required.

Table 212: GP_ID_4

Register Address	Bit	Туре	Label	Description
0x125 GP_ID_4	7:0	R/W	GP_4	Data from fuse array (OTP)

Datasheet Revision 2.4 11-Feb-2022



Register Address	Bit	Туре	Label	Description
		Note 1		

Note 1 Write access can be disabled by OTP if required.

Table 213: GP_ID_5

Register Address	Bit	Туре	Label	Description
0x126 GP_ID_5	7:0	R/W Note 1	GP_5	Data from fuse array (OTP)

Note 1 Write access can be disabled by OTP if required.

Table 214: GP_ID_6

Register Address	Bit	Туре	Label	Description
0x127 GP_ID_6	7:0	R/W Note 1	GP_6	Data from fuse array (OTP)

Note 1 Write access can be disabled by OTP if required.

Table 215: GP_ID_7

Register Address	Bit	Туре	Label	Description
0x128 GP_ID_7	7:0	R/W Note 1	GP_7	Data from fuse array (OTP)

Note 1 Write access can be disabled by OTP if required.

Table 216: GP_ID_8

Register Address	Bit	Туре	Label	Description
0x129 GP_ID_8	7:0	R/W Note 1	GP_8	Data from fuse array (OTP)

Note 1 Write access can be disabled by OTP if required.

Table 217: GP_ID_9

Register Address	Bit	Туре	Label	Description
0x12A GP_ID_9	7:0	R/W Note 1	GP_9	Data from fuse array (OTP)

Note 1 Write access can be disabled by OTP if required.

Table 218: GP_ID_10

Register Address	Bit	Туре	Label	Description
0x12B GP_ID_10	7:0	R/W	GP_10	Data from fuse array (OTP), no OTP reload after powering up from NO-POWER mode

Datasheet Revision 2.4 11-Feb-2022



Table 219: GP_ID_11

Register Address	Bit	Туре	Label	Description
0x12C GP_ID_11	7:0	R/W	GP_11	Data from fuse array (OTP)

Table 220: Reserved

Register Address	Bit	Type	Label	Description
0x12D - 0x134 Reserved	7:0	R	Reserved	Reserved

Table 221: CoPMIC_S to CoPMIC_E

Register Address	Bit	Туре	Label	Description
0x140 CoPMIC_S	7:0	R	Reserved	Reserved for Co-PMIC
0x14F CoPMIC_E	7:0	R	Reserved	Reserved for Co-PMIC

Table 222: CHG_Co_S to CHG_Co_E

Register Address	Bit	Type	Label	Description
0x150 CHG_Co_S	7:0	R	Reserved	Reserved for companion charger
0x17F CHG_Co_E	7:0	R	Reserved	Reserved for companion charger

A.5 Register Page 3

Table 223: PAGE_CON

Register Address	Bit	Туре	Label	Description
0x180	7	RW	REVERT	See register 0x00, Table 43
PAGE_CON	6	RW	WRITE_MODE	
	5:3	RW	Reserved	
	2:0	RW	REG_PAGE	

Table 224: DEVICE_ID

Register Address	Bit	Type	Label	Description
0x181 DEVICE_ID	7:0	R	DEVICE_ID	Read back of chip ID

Table 225: VARIANT_ID

Register Address	Bit	Туре	Label	Description
	7:4	R	MRC	Read back of mask revision code (MRC)



Register Address	Bit	Туре	Label	Description
0x182 VARIANT_ID	3:0	R	VRC	Read back of package variant code (VRC)

Table 226: CUSTOMER_ID

Register Address	Bit	Туре	Label	Description
0x183 CUSTOMER_ID	7:0	R	CUSTOMER_ID	ID for customer and target application platform, written during production of variant

Table 227: CONFIG_ID

Register Address	Bit	Туре	Label	Description
0x184 CONFIG_ID	7:0	R	CONFIG_REV	ID for revision of OTP settings, written during production of variant 00000000 – OTP unprogrammed (RESERVED) > 00000000 – OTP configuration revision xxx

Table 228: PMIC_STATUS

Register Address	Bit	Туре	Label	Description
0x1A8 PMIC_STATUS	7	R	PC_DONE	Power Commander download complete
	6:5	R/W	Reserved	
	4:0	R	STATUS	Decimal Decode: 03 = RESET (Shutdown) 28 = SYSTEM 25 = POWER 23 = POWER-DOWN 20 = POWER1 17 = ACTIVE



Status Definitions

Revision	Datasheet Status	Product Status	Definition
1. <n></n>	Target	Development	This datasheet contains the design specifications for product development. Specifications may be changed in any manner without notice.
2. <n></n>	Preliminary	Qualification	This datasheet contains the specifications and preliminary characterization data for products in pre-production. Specifications may be changed at any time without notice in order to improve the design.
3. <n></n>	Final	Production	This datasheet contains the final specifications for products in volume production. The specifications may be changed at any time in order to improve the design, manufacturing and supply. Major specification changes are communicated via Customer Product Notifications. Datasheet changes are communicated via www.dialog-semiconductor.com.
4. <n></n>	Obsolete	Archived	This datasheet contains the specifications for discontinued products. The information is provided for reference only.

RoHS Compliance

Dialog Semiconductor's suppliers certify that its products are in compliance with the requirements of Directive 2011/65/EU of the European Parliament on the restriction of the use of certain hazardous substances in electrical and electronic equipment. RoHS certificates from our suppliers are available on request.

IMPORTANT NOTICE AND DISCLAIMER

RENESAS ELECTRONICS CORPORATION AND ITS SUBSIDIARIES ("RENESAS") PROVIDES TECHNICAL SPECIFICATIONS AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS OR IMPLIED, INCLUDING, WITHOUT LIMITATION, ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE, OR NON-INFRINGEMENT OF THIRD-PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for developers who are designing with Renesas products. You are solely responsible for (1) selecting the appropriate products for your application, (2) designing, validating, and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, or other requirements. These resources are subject to change without notice. Renesas grants you permission to use these resources only to develop an application that uses Renesas products. Other reproduction or use of these resources is strictly prohibited. No license is granted to any other Renesas intellectual property or to any third-party intellectual property. Renesas disclaims responsibility for, and you will fully indemnify Renesas and its representatives against, any claims, damages, costs, losses, or liabilities arising from your use of these resources. Renesas' products are provided only subject to Renesas' Terms and Conditions of Sale or other applicable terms agreed to in writing. No use of any Renesas resources expands or otherwise alters any applicable warranties or warranty disclaimers for these products.

(Disclaimer Rev.1.01 Jan 2024)

Corporate Headquarters

TOYOSU FORESIA, 3-2-24 Toyosu, Koto-ku, Tokyo 135-0061, Japan www.renesas.com

Trademarks

Renesas and the Renesas logo are trademarks of Renesas Electronics Corporation. All trademarks and registered trademarks are the property of their respective owners.

Contact Information

For further information on a product, technology, the most up-to-date version of a document, or your nearest sales office, please visit www.renesas.com/contact-us/.

Notice

- 1. Descriptions of circuits, software and other related information in this document are provided only to illustrate the operation of semiconductor products and application examples. You are fully responsible for the incorporation or any other use of the circuits, software, and information in the design of your product or system. Renesas Electronics disclaims any and all liability for any losses and damages incurred by you or third parties arising from the use of these circuits, software, or information.
- Renesas Electronics hereby expressly disclaims any warranties against and liability for infringement or any other claims involving patents, copyrights, or other intellectual property rights of third parties, by or arising from the use of Renesas Electronics products or technical information described in this document, including but not limited to, the product data, drawings, charts, programs, algorithms, and application examples.
- No license, express, implied or otherwise, is granted hereby under any patents, copyrights or other intellectual property rights of Renesas Electronics or others.
- 4. You shall be responsible for determining what licenses are required from any third parties, and obtaining such licenses for the lawful import, export, manufacture, sales, utilization, distribution or other disposal of any products incorporating Renesas Electronics products, if required.
- 5. You shall not alter, modify, copy, or reverse engineer any Renesas Electronics product, whether in whole or in part. Renesas Electronics disclaims any and all liability for any losses or damages incurred by you or third parties arising from such alteration, modification, copying or reverse engineering.
- 6. Renesas Electronics products are classified according to the following two quality grades: "Standard" and "High Quality". The intended applications for each Renesas Electronics product depends on the product's quality grade, as indicated below.
 - "Standard": Computers; office equipment; communications equipment; test and measurement equipment; audio and visual equipment; home electronic appliances; machine tools; personal electronic equipment; industrial robots; etc.
 - "High Quality": Transportation equipment (automobiles, trains, ships, etc.); traffic control (traffic lights); large-scale communication equipment; key financial terminal systems; safety control equipment; etc.

Unless expressly designated as a high reliability product or a product for harsh environments in a Renesas Electronics data sheet or other Renesas Electronics document, Renesas Electronics products are not intended or authorized for use in products or systems that may pose a direct threat to human life or bodily injury (artificial life support devices or systems; surgical implantations; etc.), or may cause serious property damage (space system; undersea repeaters; nuclear power control systems; aircraft control systems; key plant systems; military equipment; etc.). Renesas Electronics disclaims any and all liability for any damages or losses incurred by you or any third parties arising from the use of any Renesas Electronics product that is inconsistent with any Renesas Electronics data sheet, user's manual or other Renesas Electronics document.

- 7. No semiconductor product is absolutely secure. Notwithstanding any security measures or features that may be implemented in Renesas Electronics hardware or software products, Renesas Electronics shall have absolutely no liability arising out of any vulnerability or security breach, including but not limited to any unauthorized access to or use of a Renesas Electronics product or a system that uses a Renesas Electronics product. RENESAS ELECTRONICS DOES NOT WARRANT OR GUARANTEE THAT RENESAS ELECTRONICS PRODUCTS, OR ANY SYSTEMS CREATED USING RENESAS ELECTRONICS PRODUCTS WILL BE INVULNERABLE OR FREE FROM CORRUPTION, ATTACK, VIRUSES, INTERFERENCE, HACKING, DATA LOSS OR THEFT, OR OTHER SECURITY INTRUSION ("Vulnerability Issues"). RENESAS ELECTRONICS DISCLAIMS ANY AND ALL RESPONSIBILITY OR LIABILITY ARISING FROM OR RELATED TO ANY VULNERABILITY ISSUES. FURTHERMORE, TO THE EXTENT PERMITTED BY APPLICABLE LAW, RENESAS ELECTRONICS DISCLAIMS ANY AND ALL WARRANTIES, EXPRESS OR IMPLIED, WITH RESPECT TO THIS DOCUMENT AND ANY RELATED OR ACCOMPANYING SOFTWARE OR HARDWARE, INCLUDING BUT NOT LIMITED TO THE IMPLIED WARRANTIES OF MERCHANTABILITY, OR FITNESS FOR A PARTICULAR PURPOSE.
- 8. When using Renesas Electronics products, refer to the latest product information (data sheets, user's manuals, application notes, "General Notes for Handling and Using Semiconductor Devices" in the reliability handbook, etc.), and ensure that usage conditions are within the ranges specified by Renesas Electronics with respect to maximum ratings, operating power supply voltage range, heat dissipation characteristics, installation, etc. Renesas Electronics disclaims any and all liability for any malfunctions, failure or accident arising out of the use of Renesas Electronics products outside of such specified ranges.
- 9. Although Renesas Electronics endeavors to improve the quality and reliability of Renesas Electronics products, semiconductor products have specific characteristics, such as the occurrence of failure at a certain rate and malfunctions under certain use conditions. Unless designated as a high reliability product or a product for harsh environments in a Renesas Electronics data sheet or other Renesas Electronics document, Renesas Electronics products are not subject to radiation resistance design. You are responsible for implementing safety measures to guard against the possibility of bodily injury, injury or damage caused by fire, and/or danger to the public in the event of a failure or malfunction of Renesas Electronics products, such as safety design for hardware and software, including but not limited to redundancy, fire control and malfunction prevention, appropriate treatment for aging degradation or any other appropriate measures. Because the evaluation of microcomputer software alone is very difficult and impractical, you are responsible for evaluating the safety of the final products or systems manufactured by you.
- 10. Please contact a Renesas Electronics sales office for details as to environmental matters such as the environmental compatibility of each Renesas Electronics product. You are responsible for carefully and sufficiently investigating applicable laws and regulations that regulate the inclusion or use of controlled substances, including without limitation, the EU RoHS Directive, and using Renesas Electronics products in compliance with all these applicable laws and regulations. Renesas Electronics disclaims any and all liability for damages or losses occurring as a result of your noncompliance with applicable laws and regulations.
- 11. Renesas Electronics products and technologies shall not be used for or incorporated into any products or systems whose manufacture, use, or sale is prohibited under any applicable domestic or foreign laws or regulations. You shall comply with any applicable export control laws and regulations promulgated and administered by the governments of any countries asserting jurisdiction over the parties or transactions.
- 12. It is the responsibility of the buyer or distributor of Renesas Electronics products, or any other party who distributes, disposes of, or otherwise sells or transfers the product to a third party, to notify such third party in advance of the contents and conditions set forth in this document.
- 13. This document shall not be reprinted, reproduced or duplicated in any form, in whole or in part, without prior written consent of Renesas Electronics.
- 14. Please contact a Renesas Electronics sales office if you have any questions regarding the information contained in this document or Renesas Electronics products.
- (Note1) "Renesas Electronics" as used in this document means Renesas Electronics Corporation and also includes its directly or indirectly controlled subsidiaries.
- (Note2) "Renesas Electronics product(s)" means any product developed or manufactured by or for Renesas Electronics.

(Disclaimer Rev.5.0-1 October 2020)

Corporate Headquarters

TOYOSU FORESIA, 3-2-24 Toyosu, Koto-ku, Tokyo 135-0061, Japan www.renesas.com

Trademarks

Renesas and the Renesas logo are trademarks of Renesas Electronics Corporation. All trademarks and registered trademarks are the property of their respective owners.

Contact Information

For further information on a product, technology, the most up-to-date version of a document, or your nearest sales office, please visit: www.renesas.com/contact/

Mouser Electronics

Authorized Distributor

Click to View Pricing, Inventory, Delivery & Lifecycle Information:

Renesas Electronics:

DA9063L-00HO1 DA9063L-XXHO1 DA9063L-XXHO2 DA9063L-00HO2