# RENESAS

### Description

The 9DBV0441 is a member of Renesas' SOC-Friendly 1.8V Very-Low-Power (VLP) PCIe family. It has integrated output terminations providing Zo =  $100\Omega$  for direct connection to  $100\Omega$  transmission lines. The device has 4 output enables for clock management, and 3 selectable SMBus addresses.

### **Typical Applications**

 1.8V PCIe Gen1–5 Zero-Delay/Fan-out Buffer (ZDB/FOB)

### **Output Features**

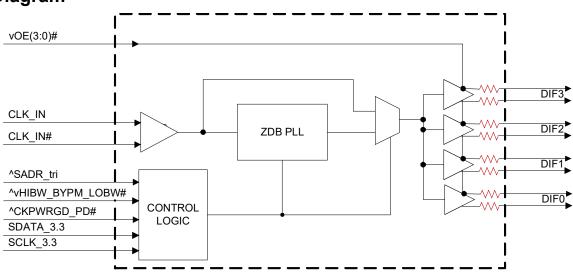
- Four 1–200MHz Low-Power (LP) HCSL DIF pairs with Zo = 100 $\Omega$ 

### **Key Specifications**

- DIF cycle-to-cycle jitter < 50ps
- DIF output-to-output skew < 50ps
- PCIe Gen5 CC additive phase jitter < 40fs RMS
- 12kHz–20MHz additive phase jitter = 156fs RMS at 156.25MHz (typical)

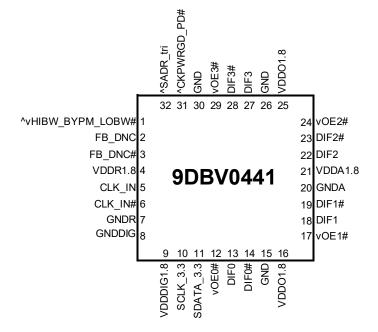
### Features/Benefits

- Direct connection to  $100\Omega$  transmission lines; saves 16 resistors compared to standard HCSL outputs
- 53mW typical power consumption in PLL mode; minimal power consumption
- Spread Spectrum (SS) compatible; allows use of SS for EMI reduction
- OE# pins; support DIF power management
- HCSL compatible differential input; can be driven by common clock sources
- Programmable Slew rate for each output; allows tuning for various line lengths
- Programmable output amplitude; allows tuning for various application environments
- Pin/software selectable PLL bandwidth and PLL Bypass; minimize phase jitter for each application
- Outputs blocked until PLL is locked; clean system start-up
- Software selectable 50MHz or 125MHz PLL operation; useful for Ethernet applications
- Configuration can be accomplished with strapping pins; SMBus interface not required for device control
- 3.3V tolerant SMBus interface works with legacy controllers
- Space saving 5 × 5mm 32-VFQFPN; minimal board space
- Selectable SMBus addresses; multiple devices can easily share an SMBus segment



### Block Diagram

### **Pin Configuration**



#### 32-pin VFQFPN, 5x5 mm, 0.5mm pitch

^ prefix indicates internal 120KOhm pull up resistor
^v prefix indicates internal 120KOhm pull up AND pull down resistor (biased to VDD/2)
v prefix indicates internal 120KOhm pull down resistor

#### **SMBus Address Selection Table**

	SADR	Address	+ Read/Write bit
State of SADR on first application of	0	1101011	X
CKPWRGD PD#	М	1101100	x
CKPVVRGD_PD#	1	1101101	х

#### **Power Management Table**

CKPWRGD PD#	CLK_IN	SMBus OEx# Pin		DIF	PLL	
		OEx bit		True O/P	Comp. O/P	FLL
0	Х	Х	Х	Low	Low	Off
1	Running	0	Х	Low	Low	On <sup>1</sup>
1	1 Running 1 0		Running	Running	On <sup>1</sup>	
1	Running	1	1	Low	Low	On <sup>1</sup>

1. If Bypass mode is selected, the PLL will be off, and outputs will be running.

#### **Power Connections**

Pin Numb	Pin Number				
VDD	GND	Description			
4	7	Input receiver analog			
9	8	Digital Power			
16, 25	15,20,26,30	DIF outputs			
21	20	PLL Analog			

#### **Frequency Select Table**

FSEL Byte3 [4:3]	CLK_IN (MHz)	DIFx (MHz)
00 (Default)	100.00	CLK_IN
01	50.00	CLK_IN
10	125.00	CLK_IN
11	Reserved	Reserved

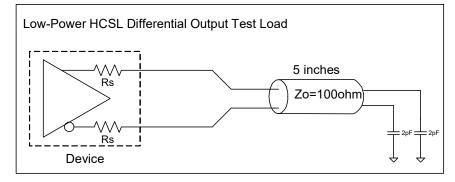
#### **PLL Operating Mode**

		Byte1 [7:6]	Byte1 [4:3]
HiBW_BypM_LoBW#	MODE	Readback	Control
0	PLL Lo BW	00	00
М	Bypass	01	01
1	PLL Hi BW	11	11

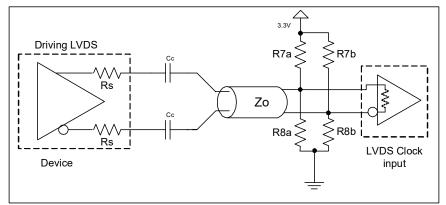
### **Pin Descriptions**

Pin#	Pin Name	Туре	Pin Description
1	^vHIBW BYPM LOBW#	LATCHED IN	Trilevel input to select High BW, Bypass or Low BW mode.
l.		LATCHED IN	See PLL Operating Mode Table for Details.
2	FB DNC	DNC	True clock of differential feedback. The feedback output and feedback input are
2		DINC	connected internally on this pin. Do not connect anything to this pin.
3	FB DNC#	DNC	Complement clock of differential feedback. The feedback output and feedback
0		BNO	input are connected internally on this pin. Do not connect anything to this pin.
4	VDDR1.8	PWR	1.8V power for differential input clock (receiver). This VDD should be treated as
			an Analog power rail and filtered appropriately.
5	CLK_IN	IN	True Input for differential reference clock.
6	CLK_IN#	IN	Complementary Input for differential reference clock.
7	GNDR	GND	Analog Ground pin for the differential input (receiver)
8	GNDDIG	GND	Ground pin for digital circuitry
9	VDDDIG1.8	PWR	1.8V digital power (dirty power)
10	SCLK_3.3	IN	Clock pin of SMBus circuitry, 3.3V tolerant.
11	SDATA_3.3	I/O	Data pin for SMBus circuitry, 3.3V tolerant.
12	vOE0#	IN	Active low input for enabling DIF pair 0. This pin has an internal pull-down.
10			1 =disable outputs, 0 = enable outputs
	DIF0	OUT	Differential true clock output
	DIF0#	OUT	Differential Complementary clock output
15		GND	Ground pin.
16	VDDO1.8	PWR	Power supply for outputs, nominally 1.8V. Active low input for enabling DIF pair 1. This pin has an internal pull-down.
17	vOE1#	IN	
18	DIF1	OUT	1 =disable outputs, 0 = enable outputs Differential true clock output
	DIF1#	OUT	Differential Complementary clock output
20	GNDA	GND	Ground pin for the PLL core.
	VDDA1.8	PWR	1.8V power for the PLL core.
	DIF2	OUT	Differential true clock output
22	DIF2#	OUT	Differential Complementary clock output
20		001	Active low input for enabling DIF pair 2. This pin has an internal pull-down.
24	vOE2#	IN	1 =disable outputs, 0 = enable outputs
25	VDDO1.8	PWR	Power supply for outputs, nominally 1.8V.
26	GND	GND	Ground pin.
	DIF3	OUT	Differential true clock output
	DIF3#	OUT	Differential Complementary clock output
			Active low input for enabling DIF pair 3. This pin has an internal pull-down.
29	vOE3#	IN	1 =disable outputs, 0 = enable outputs
30	GND	GND	Ground pin.
			Input notifies device to sample latched inputs and start up on first high assertion.
31	^CKPWRGD_PD#	IN	Low enters Power Down Mode, subsequent high assertions exit Power Down
	-		Mode. This pin has internal pull-up resistor.
22			
32	^SADR_tri	LATCHED IN	Tri-level latch to select SMBus Address. See SMBus Address Selection Table.

### **Test Loads**



## **Driving LVDS**



#### Driving LVDS inputs

	`		
	Receiver has Receiver does not		
Component	termination	have termination	Note
R7a, R7b	10K ohm	140 ohm	
R8a, R8b	5.6K ohm	75 ohm	
Сс	0.1 uF	0.1 uF	
Vcm	1.2 volts	1.2 volts	

### **Absolute Maximum Ratings**

Stresses above the ratings listed below can cause permanent damage to the 9DBV0441. These ratings, which are standard values for Renesas commercially rated parts, are stress ratings only. Functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods can affect product reliability. Electrical parameters are guaranteed only over the recommended operating temperature range.

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	NOTES
Supply Voltage	VDDx	Applies to all VDD pins	-0.5		2.5	V	1,2
Input Voltage	V <sub>IN</sub>		-0.5		V <sub>DD</sub> +0.5V	V	1, 3
Input High Voltage, SMBus	VIHSMB	SMBus clock and data pins			3.6V	V	1
Storage Temperature	Ts		-65		150	°C	1
Junction Temperature	Tj				125	°C	1
Input ESD protection	ESD prot	Human Body Model	2000			V	1

<sup>1</sup>Guaranteed by design and characterization, not 100% tested in production.

<sup>2</sup> Operation under these conditions is neither implied nor guaranteed.

<sup>3</sup> Not to exceed 2.5V.

### **Electrical Characteristics–Clock Input Parameters**

TA =  $T_{COM}$  or  $T_{IND}$ ; Supply Voltage per VDD of normal operation conditions, See Test Loads for Loading Conditions

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	NOTES
Input High Voltage - DIF_IN	VIHDIF	Differential inputs (single-ended measurement)	600	800	1150	mV	1
Input Low Voltage - DIF_IN	V <sub>ILDIF</sub>	Differential inputs (single-ended measurement)	V <sub>SS</sub> - 300	0	300	mV	1,3
Input Common Mode Voltage - DIF_IN	V <sub>COM</sub>	Common Mode Input Voltage	300		725	mV	1
Input Amplitude - DIF_IN	V <sub>SWING</sub>	Peak to Peak value (V <sub>IHDIF</sub> - V <sub>ILDIF</sub> )	300		1450	mV	1
Input Slew Rate - DIF_IN	dv/dt	Measured differentially	0.4			V/ns	1,2
Input Leakage Current	I <sub>IN</sub>	$V_{IN} = V_{DD}, V_{IN} = GND$	-5		5	uA	1
Input Duty Cycle	d <sub>tin</sub>	Measurement from differential waveform	45		55	%	1
Input Jitter - Cycle to Cycle	J <sub>DIFIn</sub>	Differential Measurement	0		150	ps	1

<sup>1</sup> Guaranteed by design and characterization, not 100% tested in production.

<sup>2</sup> Slew rate measured through +/-75mV window centered around differential zero.

<sup>3</sup> The device can be driven from a single ended clock by driving the true clock and biasing the complement clock input to the  $V_{BIAS}$ , where  $V_{BIAS}$  is ( $V_{IHHIGH} - V_{IHLOW}$ )/2.

# Electrical Characteristics–Input/Supply/Common Parameters–Normal Operating Conditions

TA =  $T_{COM}$  or  $T_{IND}$ ; Supply Voltage per VDD of normal operation conditions, See Test Loads for Loading Conditions

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	NOTES
1.8V Supply Voltage	VDD	Supply voltage for core, analog and LVCMOS outputs	1.7	1.8	1.9	V	1
Ambient Operating	T <sub>COM</sub>	Commercial range	0	25	70	°C	1
Temperature	T <sub>IND</sub>	Industrial range	-40	25	85	°C	1
Input High Voltage	V <sub>IH</sub>	Single-ended inputs, except SMBus	0.75 V <sub>DD</sub>		V <sub>DD</sub> + 0.3	V	1
Input Mid Voltage	VIM	Single-ended tri-level inputs ('_tri' suffix)	0.4 V <sub>DD</sub>		0.6 V <sub>DD</sub>	V	1
Input Low Voltage	V <sub>IL</sub>	Single-ended inputs, except SMBus	-0.3		$0.25 V_{DD}$	V	1
	I <sub>IN</sub>	Single-ended inputs, $V_{IN}$ = GND, $V_{IN}$ = VDD	-5		5	uA	1
Input Current	I <sub>INP</sub>	Single-ended inputs $V_{IN} = 0 V$ ; Inputs with internal pull-up resistors $V_{IN} = VDD$ ; Inputs with internal pull-down resistors	-200		200	uA	1
	F <sub>ibyp</sub>	Bypass mode	1		200	MHz	2
Input Fraguanay	F <sub>ipll100</sub>	100MHz PLL mode	50	100.00	140	MHz	2
Input Frequency	F <sub>ipll125</sub>	125MHz PLL mode	62.5	125.00	175	MHz	2
	F <sub>ipll62</sub>	50MHz PLL mode	25	50.00	65	MHz	2
Pin Inductance	L <sub>pin</sub>				7	nH	1
	C <sub>IN</sub>	Logic Inputs, except DIF_IN	1.5		5	pF	1
Capacitance	$C_{INDIF_IN}$	DIF_IN differential clock inputs	1.5		2.7	pF	1,6
	C <sub>OUT</sub>	Output pin capacitance			6	pF	1
Clk Stabilization	T <sub>STAB</sub>	From V <sub>DD</sub> Power-Up and after input clock stabilization or de-assertion of PD# to 1st clock		0.6	1	ms	1,2
Input SS Modulation Frequency	f <sub>MODIN</sub>	Allowable Frequency (Triangular Modulation)	30	31.500	33	kHz	1
OE# Latency	t <sub>LATOE#</sub>	DIF start after OE# assertion DIF stop after OE# deassertion	1		3	clocks	1,3
Tdrive_PD#	t <sub>DRVPD</sub>	DIF output enable after PD# de-assertion			300	us	1,3
Tfall	t <sub>F</sub>	Fall time of single-ended control inputs			5	ns	1,2
Trise	t <sub>R</sub>	Rise time of single-ended control inputs			5	ns	1,2
SMBus Input Low Voltage	V <sub>ILSMB</sub>	$V_{DDSMB}$ = 3.3V, see note 4 for $V_{DDSMB}$ < 3.3V			0.8	V	1, 4
SMBus Input High Voltage	VIHSMB	$V_{DDSMB}$ = 3.3V, see note 5 for $V_{DDSMB}$ < 3.3V	2.1		3.6	V	1, 5
SMBus Output Low Voltage	V <sub>OLSMB</sub>	@ I <sub>PULLUP</sub>			0.4	V	1
SMBus Sink Current	I <sub>PULLUP</sub>	@ V <sub>OL</sub>	4			mA	1
Nominal Bus Voltage	V <sub>DDSMB</sub>		1.7		3.6	V	1
SCLK/SDATA Rise Time	t <sub>RSMB</sub>	(Max VIL - 0.15) to (Min VIH + 0.15)			1000	ns	1
SCLK/SDATA Fall Time	t <sub>FSMB</sub>	(Min VIH + 0.15) to (Max VIL - 0.15)			300	ns	1
SMBus Operating Frequency	f <sub>MAXSMB</sub>	Maximum SMBus operating frequency			400	kHz	1,7

<sup>1</sup> Guaranteed by design and characterization, not 100% tested in production.

 $^{2}$  Control input must be monotonic from 20% to 80% of input swing.

<sup>3</sup> Time from deassertion until outputs are > 200 mV.

 $^{4}$  For V<sub>DDSMB</sub> < 3.3V, V<sub>ILSMB</sub> < = 0.35V<sub>DDSMB</sub>.

 $^{5}$  For V<sub>DDSMB</sub> < 3.3V, V<sub>IHSMB</sub> > = 0.65V<sub>DDSMB</sub>.

<sup>6</sup> DIF\_IN input.

<sup>7</sup> The differential input clock must be running for the SMBus to be active.

### **Electrical Characteristics–DIF 0.7V Low Power HCSL Outputs**

TA = T<sub>COM</sub> or T<sub>IND</sub>; Supply Voltage per VDD of normal operation conditions, See Test Loads for Loading Conditions

		-					
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	NOTES
Slew rate	Trf	Scope averaging on 3.0V/ns setting	1.1	2	3	V/ns	1, 2, 3
Slew late	111	Scope averaging on 2.0V/ns setting	1.9	3	4	V/ns	1, 2, 3
Slew rate matching	∆Trf	Slew rate matching, Scope averaging on		7	20	%	1, 2, 4
Voltage High	V <sub>HIGH</sub>	Statistical measurement on single-ended signal using oscilloscope math function. (Scope	660	774	850	mV	1,7
Voltage Low	V <sub>LOW</sub>	averaging on)		18	150	v	1,7
Max Voltage	Vmax	Measurement on single ended signal using		821	1150	mV	1
Min Voltage	Vmin	absolute value. (Scope averaging off)	-300	-15		IIIV	1
Vswing	Vswing	Scope averaging off	300	1536		mV	1,2,7
Crossing Voltage (abs)	Vcross_abs	Scope averaging off	250	414	550	mV	1,5,7
Crossing Voltage (var)	∆-Vcross	Scope averaging off		13	140	mV	1, 6

<sup>1</sup>Guaranteed by design and characterization, not 100% tested in production.

<sup>2</sup> Measured from differential waveform

<sup>3</sup> Slew rate is measured through the Vswing voltage range centered around differential 0V. This results in a +/-150mV window around differential 0V.

<sup>4</sup> Matching applies to rising edge rate for Clock and falling edge rate for Clock#. It is measured using a +/-75mV window centered on the average cross point where Clock rising meets Clock# falling. The median cross point is used to calculate the voltage thresholds the oscilloscope is to use for the edge rate calculations.

<sup>5</sup> Vcross is defined as voltage where Clock = Clock# measured on a component test board and only applies to the differential rising edge (i.e. Clock rising and Clock# falling).

<sup>6</sup> The total variation of all Vcross measurements in any particular system. Note that this is a subset of Vcross\_min/max (Vcross absolute) allowed. The intent is to limit Vcross induced modulation by setting Δ-Vcross to be smaller than Vcross absolute.

<sup>7</sup> At default SMBus settings.

### **Electrical Characteristics–Current Consumption**

TA = T<sub>COM</sub> or T<sub>IND;</sub> Supply Voltage per VDD of normal operation conditions, See Test Loads for Loading Conditions

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	NOTES
Operating Supply Current	I <sub>DDAOP</sub>	VDDA+VDDR, PLL Mode, @100MHz		11	15	mA	1
	I <sub>DDOP</sub>	VDD1.8, All outputs active @100MHz		25	35	mA	1
Powerdown Current	I <sub>DDAPD</sub>	VDDA+VDDR, PLL Mode, @100MHz			1	mA	1,2
	I <sub>DDPD</sub>	VDD1.8, Outputs Low/Low			1.2	mA	1, 2

<sup>1</sup> Guaranteed by design and characterization, not 100% tested in production.

<sup>2</sup> Input clock stopped.

### Electrical Characteristics–Output Duty Cycle, Jitter, Skew and PLL Characteristics

TA = T<sub>COM</sub> or T<sub>IND</sub>; Supply Voltage per VDD of normal operation conditions, See Test Loads for Loading Conditions

	• •			-			
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	NOTES
PLL Bandwidth	BW	-3dB point in High BW Mode	2	2.7	4	MHz	1,5
FLE Balldwidth	-3dB point in Low BW Mode		1	1.4	2	MHz	1,5
PLL Jitter Peaking	t <sub>JPEAK</sub>	Peak Pass band Gain		1.2	2	dB	1
Duty Cycle	t <sub>DC</sub>	Measured differentially, PLL Mode	45	50.1	55	%	1
Duty Cycle Distortion	t <sub>DCD</sub>	Measured differentially, Bypass Mode @100MHz	-1	0	1	%	1,3
Skow Input to Output	t <sub>pdBYP</sub>	Bypass Mode, V <sub>T</sub> = 50%	3000	3600	4500	ps	1
Skew, Input to Output	t <sub>pdPLL</sub>	PLL Mode $V_T = 50\%$	0	92	200	ps	1,4
Skew, Output to Output	t <sub>sk3</sub>	V <sub>T</sub> = 50%		28	50	ps	1,4
Jitter, Cycle to cycle	t.	PLL mode		16	50	ps	1,2
JILLEI, CYCLE LO CYCLE	t <sub>jcyc-cyc</sub>	Additive Jitter in Bypass Mode		0.1	25	ps	1,2

<sup>1</sup> Guaranteed by design and characterization, not 100% tested in production.

<sup>2</sup> Measured from differential waveform

<sup>3</sup> Duty cycle distortion is the difference in duty cycle between the output and the input clock when the device is operated in bypass mode.

<sup>4</sup> All outputs at default slew rate

<sup>5</sup> The MIN/TYP/MAX values of each BW setting track each other, i.e., Low BW MAX will never occur with Hi BW MIN.

### Electrical Characteristics–Phase Jitter Parameters – 12kHz to 20MHz

T<sub>AMB</sub> = over the specified operating range. Supply Voltages per normal operation conditions. See Test Loads for loading conditions.

Parameter	Symbol	Conditions	Minimum	Typical	Maximum	Specification Limit	Units	Notes
12k-20M <i>Additive</i> Phase Jitter, Fan-out Buffer Mode	tjph12k-20MFOB	Fan-out Buffer Mode, SSC OFF, 156.25MHz		156		n/a	fs (rms)	1, 2, 3

Notes:

1. Applies to all differential outputs, guaranteed by design and characterization. See Test Loads for measurement setup details.

2. 12kHz to 20M Hz brick wall filter.

3. For RMS values additive jitter is calculated by solving for b where  $[b = sqrt(c^2 - a^2)]$ , a is rms input jitter and c is rms total jitter.

### Electrical Characteristics–Additive PCIe Phase Jitter for Fanout Buffer Mode<sup>[7]</sup>

T<sub>AMB</sub> = over the specified operating range. Supply Voltages per normal operation conditions. See Test Loads for loading conditions.

Parameter	Symbol	Conditions	Minimum	Typical	Maximum	Limit	Units	Notes
	tjphPCleG1-CC	PCle Gen 1 (2.5 GT/s)		1.7	3.0	86	рs (р-р)	1, 2
		PCIe Gen 2 Hi Band (5.0 GT/s)		0.033	0.049	3	ps (RMS)	1, 2
Additive PCIe Phase Jitter, Fan-out Buffer Mode (Common Clocked Architecture)	₿phPCleG2-CC	PCIe Gen 2 Lo Band (5.0 GT/s)		0.122	0.199	3.1	ps (RMS)	1, 2
	tjphPCleG3-CC	PCIe Gen 3 (8.0 GT/s)		0.059	0.098	1	ps (RMS)	1, 2
	tjphPCleG4-CC	PCIe Gen 4 (16.0 GT/s)		0.059	0.098	0.5	ps (RMS)	1, 2, 3, 4
	₿phPCleG5-CC	PCIe Gen 5 (32.0 GT/s)		0.023	0.038	0.15	ps (RMS)	1, 2, 3, 5
	tjphPCleG1-SRIS	PCIe Gen 1 (2.5 GT/s)		0.175	0.038	n/a	ps (RMS)	1, 2, 6
Additive PCIe Phase Jitter.	tjphPCleG2-SRIS	PCIe Gen 2 (5.0 GT/s)		0.156	0.275	n/a	ps (RMS)	1, 2, 6
Fan-out Buffer Mode	tjphPCleG3-SRIS	PCIe Gen 3 (8.0 GT/s)		0.041	0.247	n/a	ps (RMS)	1, 2, 6
(SRIS Architecture)	tjphPCleG4-SRIS	PCIe Gen 4 (16.0 GT/s)		0.043	0.064	n/a	ps (RMS)	1, 2, 6
	tjphPCleG5-SRIS	PCIe Gen 5 (32.0 GT/s)		0.036	0.066	n/a	ps (RMS)	1, 2, 6

Notes:

1. The Refclk jitter is measured after applying the filter functions found in PCI Express Base Specification 5.0, Revision 1.0. See the Test Loads section of the data sheet for the exact measurement setup. The total Ref Clk jitter limits for each data rate are listed for convenience. The worst case results for each data rate are summarized in this table. If oscilloscope data is used, equipment noise is removed from all results.

2. Jitter measurements shall be made with a capture of at least 100,000 clock cycles captured by a real-time oscilloscope (RTO) with a sample rate of 20 GS/s or greater. Broadband oscilloscope noise must be minimized in the measurement. The measured PP jitter is used (no extrapolation) for RTO measurements. Alternately - Jitter measurements may be used with a Phase Noise Analyzer (PNA) extending (flat) and integrating and folding the frequency content up to an offset from the carrier frequency of at least 200 MHz (at 300 MHz absolute frequency) below the Nyquist frequency. For PNA measurements for the 2.5 GT/s data rate, the RMS jitter is converted to peak to peak jitter using a multiplication factor of 8.83. In the case where real-time oscilloscope and PNA measurements have both been done and produce different results the RTO result must be used.

3. SSC spurs from the fundamental and harmonics are removed up to a cutoff frequency of 2 MHz taking care to minimize removal of any non-SSC content.

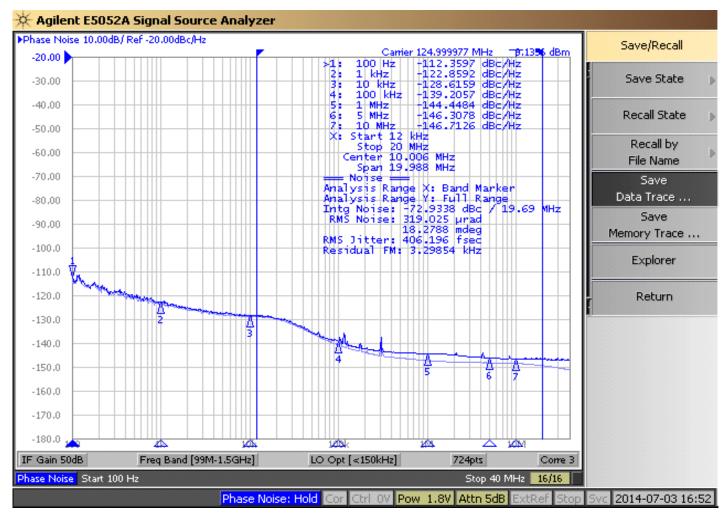
4. Note that 0.7 ps RMS is to be used in channel simulations to account for additional noise in a real system.

5. Note that 0.25 ps RMS is to be used in channel simulations to account for additional noise in a real system.

6. The PCI Express Base Specification 5.0, Revision 1.0 provides the filters necessary to calculate SRIS jitter values, however, it does not provide specification limits, hence the n/a in the Limit column. SRIS values are informative only. In general, a clock operating in an SRIS system must be twice as good as a clock operating in a Common Clock system. For RMS values, twice as good is equivalent to dividing the CC value by  $\sqrt{2}$ . And additional consideration is the value for which to divide by  $\sqrt{2}$ . The conservative approach is to divide the ref clock jitter limit, and the case can be made for dividing the channel simulation values by  $\sqrt{2}$ , if the ref clock is close to the Tx clock input. An example for Gen4 is as follows. A "rule-of-thumb" SRIS limit would be either 0.5ps RMS/ $\sqrt{2}$  = 0.35ps RMS if the clock chip is far from the clock input, or 0.7ps RMS/ $\sqrt{2}$  = 0.5ps RMS if the clock chip is near the clock input.

7. Additive jitter for RMS values is calculated by solving for b where  $b = \sqrt{(c^2 - a^2)}$ , and a is rms input jitter and c is rms output jitter.

### Additive Phase Jitter Plot: 125M (12kHz to 20MHz)



### **General SMBus Serial Interface Information**

#### How to Write

- Controller (host) sends a start bit
- Controller (host) sends the write address
- Renesas clock will acknowledge
- Controller (host) sends the beginning byte location = N
- Renesas clock will acknowledge
- Controller (host) sends the byte count = X
- Renesas clock will **acknowledge**
- Controller (host) starts sending Byte N through Byte N+X-1
- Renesas clock will **acknowledg**e each byte **one at a time**
- Controller (host) sends a Stop bit

#### **Index Block Write Operation** Controller (Host) Renesas (Slave/Receiver) Т starT bit Slave Address WR WRite ACK Beginning Byte = N ACK Data Byte Count = X ACK ×B Beginning Byte N ACK ₹ 0 0 Ο 0 0 0 Byte N + X - 1 ACK Р stoP bit

Note: SMBus Address is Latched on SADR pin.

#### How to Read

- Controller (host) will send a start bit
- Controller (host) sends the write address
- Renesas clock will acknowledge
- Controller (host) sends the beginning byte location = N
- Renesas clock will acknowledge
- · Controller (host) will send a separate start bit
- Controller (host) sends the read address
- Renesas clock will acknowledge
- Renesas clock will send the data byte count = X
- Renesas clock sends Byte N+X-1
- Renesas clock sends Byte 0 through Byte X (if X<sub>(H)</sub> was written to Byte 8)
- Controller (host) will need to acknowledge each byte
- Controller (host) will send a not acknowledge bit
- Controller (host) will send a stop bit

	Index Bloc	k Rea	d Operation
С	ontroller (Host)		Renesas (Slave/Receiver)
Т	starT bit		
5	Slave Address		
WR	WRite		
			ACK
Be	ginning Byte = N		
			ACK
RT	Repeat starT		
ę	Slave Address		
RD	ReaD		
			ACK
			Data Byte Count=X
	ACK		
			Beginning Byte N
	ACK		
			0
	0		0
	0	a	0
	0		
	i	X Byte	Byte N + X - 1
Ν	Not acknowledge		
Р	stoP bit		

#### SMBus Table: Output Enable Register <sup>1</sup>

Byte 0	Name	Control Function	Туре	0	1	Default	
Bit 7		Reserved				1	
Bit 6	DIF OE3	Output Enable	RW	Low/Low	Enabled	1	
Bit 5	DIF OE2	Output Enable	RW	Low/Low	Enabled	1	
Bit 4	Reserved						
Bit 3	DIF OE1	Output Enable	RW	Low/Low	Enabled	1	
Bit 2		Reserved				1	
Bit 1	DIF OE0	Output Enable	RW	Low/Low	Enabled	1	
Bit 0		Reserved				1	

1. A low on these bits will override the OE# pin and force the differential output Low/Low

#### SMBus Table: PLL Operating Mode and Output Amplitude Control Register

Byte 1	Name	Control Function	Туре	0	1	Default
Bit 7	PLLMODERB1	PLL Mode Readback Bit 1	R	See PLL Operat	ing Mode Table	Latch
Bit 6	PLLMODERB0	PLL Mode Readback Bit 0	R		ing mode rable	Latch
Bit 5	PLLMODE_SWCNTRL	Enable SW control of PLL Mode	RW	Values in B1[7:6] set PLL Mode	Values in B1[4:3] set PLL Mode	0
Bit 4	PLLMODE1	PLL Mode Control Bit 1	RW <sup>1</sup>	See PLL Operat	ing Mada Tabla	0
Bit 3	PLLMODE0	PLL Mode Control Bit 0	RW <sup>1</sup>	See PLL Operat	ing mode rable	0
Bit 2		Reserved				1
Bit 1	AMPLITUDE 1	Controls Output Amplitude	RW	00 = 0.6V	01 = 0.7V	1
Bit 0	AMPLITUDE 0		RW	10= 0.8V	11 = 0.9V	0

1. B1[5] must be set to a 1 for these bits to have any effect on the part.

#### SMBus Table: DIF Slew Rate Control Register

Byte 2	Name	Control Function	Туре	0	1	Default	
Bit 7		Reserved				1	
Bit 6	SLEWRATESEL DIF3	Slew Rate Selection	RW	2 V/ns	3 V/ns	1	
Bit 5	SLEWRATESEL DIF2	3 V/ns	1				
Bit 4	Reserved						
Bit 3	SLEWRATESEL DIF1	Slew Rate Selection	RW	2 V/ns	3 V/ns	1	
Bit 2		Reserved				1	
Bit 1	SLEWRATESEL DIF0	Slew Rate Selection	RW	2 V/ns	3 V/ns	1	
Bit 0		Reserved				1	

#### SMBus Table: Frequency Select Control Register

Byte 3	Name	Name Control Function		0	1	Default		
Bit 7		Reserved				1		
Bit 6		Reserved						
Bit 5	FREQ_SEL_EN	Enable SW selection of frequency	RW	SW frequency SW frequency change disabled change enabled		0		
Bit 4	FSEL1	Freq. Select Bit 1	RW <sup>1</sup>	See Frequency	0			
Bit 3	FSEL0	Freq. Select Bit 0	RW <sup>1</sup>	See Trequency		0		
Bit 2		Reserved				1		
Bit 1	Reserved							
Bit 0	SLEWRATESEL FB	Adjust Slew Rate of FB	RW	2V/ns	3V/ns	1		

1. B3[5] must be set to a 1 for these bits to have any effect on the part.

#### Byte 4 is Reserved and reads back 'hFF

#### SMBus Table: Revision and Vendor ID Register

Byte 5	Name	Control Function	Туре	0	1	Default
Bit 7	RID3		R			0
Bit 6	RID2	Revision ID	R	A rev :	0	
Bit 5	RID1	Revision ID	R	A 160 -	0	
Bit 4	RID0		R		0	
Bit 3	VID3		R			0
Bit 2	VID2	VENDOR ID	R	0001 = IDT		0
Bit 1	VID1	VENDOR ID	R			0
Bit 0	VID0		R			1

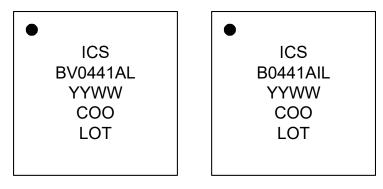
#### SMBus Table: Device Type/Device ID

Byte 6	Name	Control Function	Туре	0	1	Default
Bit 7	Device Type1	Device Type	R	00 = FGV, 01 = DBV,		0
Bit 6	Device Type0	Device Type	R	10 = DMV, 1	1	
Bit 5	Device ID5		R			0
Bit 4	Device ID4		R		0	
Bit 3	Device ID3	Device ID	R	000100 bina	0	
Bit 2	Device ID2		R		1	
Bit 1	Device ID1	]	R			0
Bit 0	Device ID0	]	R			0

#### SMBus Table: Byte Count Register

Byte 7	Name	Control Function	Туре	0	1	Default		
Bit 7	Reserved							
Bit 6	Reserved							
Bit 5	5 Reserved							
Bit 4	BC4		RW			0		
Bit 3	BC3		RW	Writing to this regist	er will configure how	1		
Bit 2	BC2	Byte Count Programming	RW	many bytes will be r	read back, default is	0		
Bit 1	BC1		RW	= 8 b	ytes.	0		
Bit 0	BC0		RW			0		

### **Marking Diagrams**



Notes:

- 1. "LOT" is the lot sequence number.
- 2. "COO" denotes country of origin.
- 3. YYWW is the last two digits of the year and week that the part was assembled.
- 4. Line 2: truncated part number
- 5. "L" denotes RoHS compliant package.
- 6. "I" denotes industrial temperature range device.

### **Thermal Characteristics**

PARAMETER	SYMBOL	CONDITIONS	PKG	TYP VALUE	UNITS	NOTES
	θ <sub>JC</sub>	Junction to Case		42	°C/W	1
	$\theta_{Jb}$	Junction to Base		2.4	°C/W	1
Thermal Resistance	$\theta_{JA0}$	Junction to Air, still air	NLG32	39	°C/W	1
merma Resistance	$\theta_{JA1}$	Junction to Air, 1 m/s air flow	INLG32	33	°C/W	1
	$\theta_{JA3}$	Junction to Air, 3 m/s air flow		28	°C/W	1
	$\theta_{JA5}$	Junction to Air, 5 m/s air flow		27	°C/W	1

<sup>1</sup>ePad soldered to board

### **Package Outline Drawings**

The package outline drawings are appended at the end of this document and are accessible from the link below. The package information is the most current data available.

32-VFQFPN (NLG32P1)

### **Ordering Information**

Part / Order Number	Shipping Packaging	Package	Temperature
9DBV0441AKLF	Trays	32-pin VFQFPN	0 to +70° C
9DBV0441AKLFT	Tape and Reel	32-pin VFQFPN	0 to +70° C
9DBV0441AKILF	Trays	32-pin VFQFPN	-40 to +85° C
9DBV0441AKILFT	Tape and Reel	32-pin VFQFPN	-40 to +85° C

"LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

"A" is the device revision designator (will not correlate with the datasheet revision).

### **Revision History**

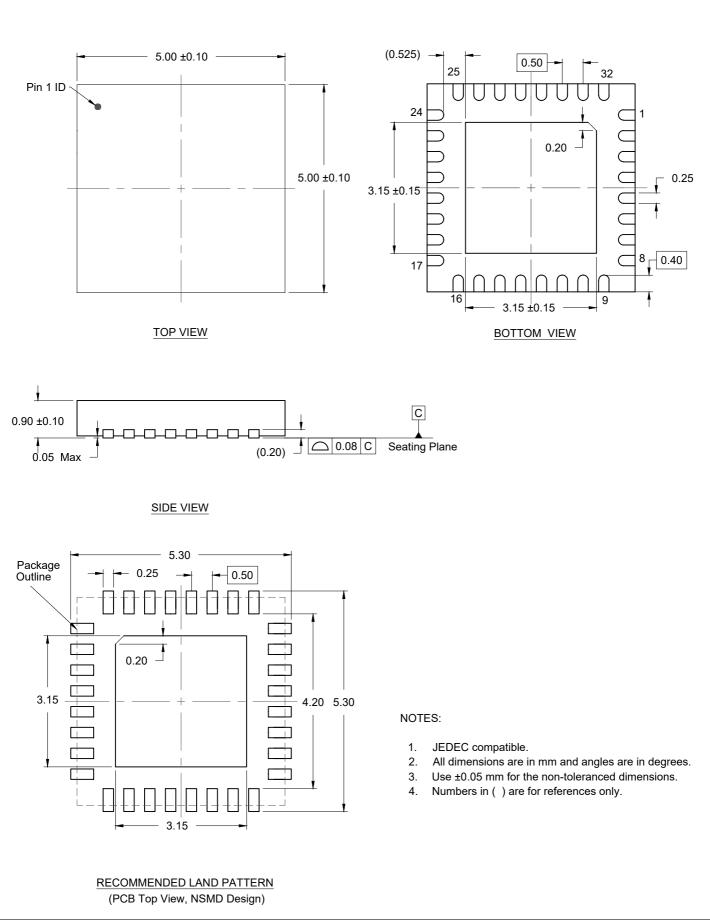
<b>Revision Date</b>	Description
August 13, 2012	1. Removed "Differential" from DS title and Recommended Application, corrected typo's in
	Description. Updated block diagram to indicate internal terminations.
	2. Corrected spelling error in pull-up/pulldown text under pinout
	3. Updated all electrical tables and added "Industry Limit" column to "Phase Jitter
	Parameters".
	4. Updated Byte3[0] to be consistent with Byte 2. Updated Byte6[7:6] definition.
	5. Added thermal data to page 12.
	6. Added NLG32 to "Package Outline and Package Dimensions" on page 13.
	7. Move to final.
February 25, 2013	1. Changed VIH min. from 0.65*VDD to 0.75*VDD
	2. Changed VIL max. from 0.35*VDD to 0.25*VDD
	3. Added missing mid-level input voltage spec (VIM) of 0.4*VDD to 0.6*VDD.
October 27, 2014	1. Updated front page text for consistency and updated block diagram resistor colors to
	highlight internal resistors.
	2. Updated max frequency of 100MHz PLL mode from 110MHz to 140MHz
	3. Updated max frequency of 125MHz PLL mode from 137.5MHz to 175MHz
	4. Updated max frequency of 50MHz PLL mode from 55MHz to 65MHz
November 26, 2014	1. Updated Key Specifications with additive phase jitter.
	2. Added additive phase jitter plot to specifications.
April 22, 2016	1. Updated max frequency of 100MHz PLL mode to 140MHz
	2. Updated max frequency of 125MHz PLL mode to 175MHz
	3. Updated max frequency of 50MHz PLL mode to 65MHz
August 27, 2019	Update to PCIe Gen4.
July 29, 2020	Corrected typo in Output Features on front page; changed 200Hz to 200MHz.
July 29, 2021	1. Updated document title.
	2. Updated Recommended Applications.
	3. Updated Key Specifications.
	4. Updated Phase Jitter tables.
February 6, 2023	Updated POD link.

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### Package Outline Drawing

Package Code:NLG32P1 32-VFQFPN 5.0 x 5.0 x 0.9 mm Body, 0.5mm Pitch PSC-4171-01, Revision: 04, Date Created: Aug 15, 2022



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