



MAX4012/MAX4016/

MAX4018/MAX4020

Click here to ask an associate for production status of specific part numbers.

### Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **General Description**

The MAX4012 single, MAX4016 dual, MAX4018 triple, and MAX4020 quad op amps are unity-gain-stable devices that combine high-speed performance with Rail-to-Rail outputs. The MAX4018 has a disable feature that reduces power-supply current to 400 $\mu$ A and places its outputs into a high-impedance state. These devices operate from a 3.3V to 10V single supply or from  $\pm 1.65$ V to  $\pm 5$ V dual supplies. The common-mode input voltage range extends beyond the negative power-supply rail (ground in single-supply applications).

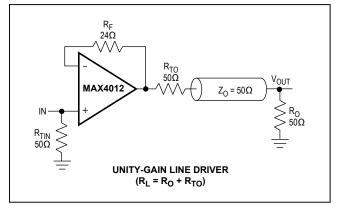
These devices require only 5.5mA of quiescent supply current while achieving a 200MHz -3dB bandwidth and a 600V/µs slew rate. These parts are an excellent solution in low-power/low-voltage systems that require wide bandwidth, such as video, communications, and instrumentation. In addition, when disabled, their high-output impedance makes them ideal for multiplexing applications.

The MAX4012 comes in a miniature 5-pin SOT23 and 8-pin SO package, while the MAX4016 comes in 8-pin  $\mu$ MAX<sup>®</sup> and SO packages. The MAX4018/MAX4020 are available in a space-saving 16-pin QSOP, as well as a 14-pin SO.

### **Applications**

- Set-Top Boxes
- Surveillance Video Systems
- Battery-Powered Instruments
- Video Line Driver
- Analog-to-Digital Converter Interface
- CCD Imaging Systems
- Video Routing and Switching Systems

### **Typical Operating Circuit**



µMAX is a registered trademark of Maxim Integrated Products, Inc.

### Features

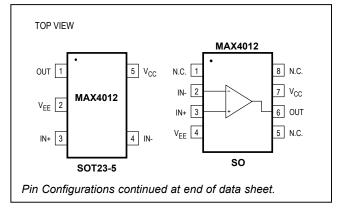
- Low-Cost
- High Speed:
  - 200MHz -3dB Bandwidth (MAX4012)
  - 150MHz -3dB Bandwidth
  - (MAX4016/MAX4018/MAX4020)
  - 30MHz 0.1dB Gain Flatness
  - · 600V/µs Slew Rate
- Single 3.3V/5.0V Operation
- Rail-to-Rail Outputs
- Input Common-Mode Range Extends Beyond V<sub>EE</sub>
- Low Differential Gain/Phase: 0.02%/0.02°
- Low Distortion at 5MHz:
  - -78dBc SFDR
  - -75dB Total Harmonic Distortion
- High-Output Drive: ±120mA
- 400µA Shutdown Capability (MAX4018)
- High-Output Impedance in Off State (MAX4018)
- Space-Saving SOT23, SO, µMAX, or QSOP Packages

### **Ordering Information**

PART	TEMP	PIN-	TOP	
PARI	RANGE	PACKAGE	MARK	
MAX4012EUK+T	-40°C to +85°C	5 SOT23-5	AMKD	
MAX4012ESA+T	-40°C to +85°C	8 SO	—	
MAX4016ESA+T	-40°C to +85°C	8 SO	—	
MAX4016EUA+T	-40°C to +85°C	8 µMAX	—	

Ordering Information continued at end of data sheet.

### **Pin Configurations**



#### 19-1246; Rev 4; 8/22

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# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Absolute Maximum Ratings**

Supply Voltage (V <sub>CC</sub> to V <sub>EE</sub> )12V	
IN, IN_+, OUT_, EN(V <sub>EE</sub> - 0.3V) to (V <sub>CC</sub> + 0.3V)	
Output Short-Circuit Duration to V <sub>CC</sub> or V <sub>EE</sub> Continuous	
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
5-Pin SOT23 (derate 7.1mW/°C above +70°C)571mW	
8-Pin SO (derate 5.9mW/°C above +70°C)471mW	

8-Pin µMAX (derate 4.1mW/°C above +70°C)	
14-Pin SO (derate 8.3mW/°C above +70°C)667mW	
16-Pin QSOP (derate 8.3mW/°C above +70°C)667mW	
Operating Temperature Range40°C to +85°C	
Storage Temperature Range65°C to +150°C	
Lead Temperature (soldering, 10s)+300°C	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### **DC Electrical Characteristics**

(V<sub>CC</sub> = 5V, V<sub>EE</sub> = 0, EN\_ = 5V, R<sub>L</sub> =  $\infty$  to V<sub>CC</sub>/2, V<sub>OUT</sub> = V<sub>CC</sub>/2, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, unless otherwise noted. Typical values are at T<sub>A</sub> = +25°C.) (Note 1)

PARAMETER	SYMBOL	COND	ITIONS	MIN	TYP	MAX	UNITS	
Input Common-Mode Voltage Range	V <sub>CM</sub>	Guaranteed by CMRR test		V <sub>EE</sub> - 0.20		V <sub>CC</sub> - 2.25	V	
Input Offset Voltage (Note 2)	V <sub>OS</sub>				4	20	mV	
Input Offset Voltage Temperature Coefficient	TC <sub>VOS</sub>				8		µV/°C	
Input Offset Voltage Matching		Any channels for MAX MAX4020	(4016/MAX4018/		±1		mV	
Input Bias Current	Ι <sub>Β</sub>	(Note 2)			5.4	20	μA	
Input Offset Current	I <sub>OS</sub>	(Note 2)			0.1	20	μA	
Innut Desistance	Dur	Differential mode (-1V	′ ≤ V <sub>IN</sub> ≤ +1V)		70		kΩ	
Input Resistance	R <sub>IN</sub>	Common mode (-0.2V		3		MΩ		
Common-Mode Rejection Ratio	CMRR	$(V_{EE} - 0.2V) \le V_{CM} \le$	(V <sub>CC</sub> - 2.25V)	70	100		dB	
	A <sub>VOL</sub>	$0.25V \le V_{OUT} \le 4.75V$ , R <sub>L</sub> = 2k $\Omega$			61			
Open-Loop Gain (Note 2)		$0.5V \le V_{OUT} \le 4.5V$ , R <sub>L</sub> = 150 $\Omega$		52	59		dB	
		$1.0V \le V_{OUT} \le 4V, R_{L}$		57				
			V <sub>CC</sub> - V <sub>OH</sub>		0.06			
		$R_L = 2k\Omega$	V <sub>OL</sub> - V <sub>EE</sub>		0.06		- - - -	
		D = 1500	V <sub>CC</sub> - V <sub>OH</sub>		0.30			
Output Voltage Swing	M	R <sub>L</sub> = 150Ω	V <sub>OL</sub> - V <sub>EE</sub>		0.30			
(Note 2)	V <sub>OUT</sub>	D - 750	V <sub>CC</sub> - V <sub>OH</sub>		0.6	1.5		
		R <sub>L</sub> = 75Ω	V <sub>OL</sub> - V <sub>EE</sub>		0.6	1.5		
		R <sub>I</sub> = 75Ω	V <sub>CC</sub> - V <sub>OH</sub>		1.1	2.0		
		to ground	V <sub>OL</sub> - V <sub>EE</sub>		0.05	0.50		
Output Current	IOUT	$R_L = 20\Omega$ to $V_{CC}$ or	T <sub>A</sub> = +25°C	±70	±120		<u> </u>	
		V <sub>EE</sub>	$T_A = T_{MIN}$ to $T_{MAX}$	±60			mA	
Output Short-Circuit Current	I <sub>SC</sub>	Sinking or sourcing			±150		mA	
Open-Loop Output Resistance	R <sub>OUT</sub>				8		Ω	

# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **DC Electrical Characteristics (continued)**

(V<sub>CC</sub> = 5V, V<sub>EE</sub> = 0, EN\_ = 5V, R<sub>L</sub> =  $\infty$  to V<sub>CC</sub>/2, V<sub>OUT</sub> = V<sub>CC</sub>/2, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, unless otherwise noted. Typical values are at T<sub>A</sub> = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
		V <sub>CC</sub> = 5V, V <sub>EE</sub> = 0, V <sub>CM</sub> = 2.0V	46	57			
Power-Supply Rejection Ratio (Note 3)	PSRR	V <sub>CC</sub> = 5V, V <sub>EE</sub> = -5V, V <sub>CM</sub> = 0	54	66		dB	
		V <sub>CC</sub> = 3.3V, V <sub>EE</sub> = 0, V <sub>CM</sub> = 0.90V		45			
Operating Supply-Voltage Range	VS	$V_{CC}$ to $V_{EE}$	3.15		11.0	v	
Disabled Output Resistance	R <sub>OUT (OFF)</sub>	EN_ = 0, 0 ≤ V <sub>OUT</sub> ≤ 5V (Note 4)	28	35		kΩ	
EN_Logic-Low Threshold	V <sub>IL</sub>				V <sub>CC</sub> - 2.6	V	
EN_ Logic-High Threshold	V <sub>IH</sub>		V <sub>CC</sub> - 1.6			V	
ENL Logio Input Low Current	IIL	$(V_{EE} + 0.2V) \le EN_{\le} V_{CC}$		0.5		μΑ	
EN_ Logic Input Low Current		EN_ = 0		200	400		
EN_Logic Input High Current	IIH	EN_ = 5V		0.5	10	μA	
Quiescent Supply Current		Enabled		5.5	7.0	m۸	
(per Amplifier)	I <sub>S</sub>	MAX4018, disabled (EN_ = 0)		0.40	0.65	– mA	

### Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **AC Electrical Characteristics**

(V<sub>CC</sub> = 5V, V<sub>EE</sub> = 0, V<sub>CM</sub> = 2.5V, EN\_ = 5V, R<sub>F</sub> = 24 $\Omega$ , R<sub>L</sub> = 100 $\Omega$  to V<sub>CC</sub>/2, V<sub>OUT</sub> = V<sub>CC</sub>/2, A<sub>VCL</sub> = 1, T<sub>A</sub> = +25°C, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS	
			MAX4012		200			
Small-Signal -3dB Bandwidth	BW <sub>SS</sub>	V <sub>OUT</sub> = 20mV <sub>P-P</sub>	MAX4016/MAX4018/ MAX4020	150			MHz	
Large-Signal -3dB Bandwidth	BW <sub>LS</sub>	V <sub>OUT</sub> = 2V <sub>P-P</sub>			140		MHz	
Bandwidth for 0.1dB Gain Flatness	BW0.1dB	V <sub>OUT</sub> = 20mV <sub>P-P</sub> (Note	5)	6	30		MHz	
Slew Rate	SR	V <sub>OUT</sub> = 2V step			600		V/µs	
Settling Time to 0.1%	t <sub>S</sub>	V <sub>OUT</sub> = 2V step			45		ns	
Rise/Fall Time	t <sub>R</sub> , t <sub>F</sub>	V <sub>OUT</sub> = 100mV <sub>P-P</sub>			1		ns	
Spurious-Free Dynamic Range	SFDR	f <sub>C</sub> = 5MHz, V <sub>OUT</sub> = 2V <sub>I</sub>	р_Р		-78		dBc	
			2nd harmonic		-78		dDa	
Harmonic Distortion	HD	f <sub>C</sub> = 5MHz,	3rd harmonic	-82		dBc		
		V <sub>OUT</sub> = 2V <sub>P-P</sub>	Total harmonic distortion		-75		dB	
Two-Tone, Third-Order Intermodulation Distortion	IP3	f <sub>1</sub> = 10.0MHz, f <sub>2</sub> = 10.1		35		dBc		
Input 1dB Compression Point		f <sub>C</sub> = 10MHz, A <sub>VCL</sub> = 2			11		dBm	
Differential Phase Error	DP	NTSC, R <sub>L</sub> = 150Ω			0.02		degrees	
Differential Gain Error	DG	NTSC, R <sub>L</sub> = 150Ω			0.02		%	
Input Noise-Voltage Density	e <sub>n</sub>	f = 10kHz	f = 10kHz		10		nV/√Hz	
Input Noise-Current Density	i <sub>n</sub>	f = 10kHz			1.3		nV/√Hz	
Input Capacitance	C <sub>IN</sub>				1		pF	
Disabled Output Capacitance	C <sub>OUT (OFF)</sub>	MAX4018, EN_ = 0			2		pF	
Output Impedance	Z <sub>OUT</sub>	f = 10MHz			6		Ω	
Amplifier Enable Time	t <sub>ON</sub>	MAX4018			100		ns	
Amplifier Disable Time	t <sub>OFF</sub>	MAX4018			1		μs	
Amplifier Gain Matching		MAX4016/MAX4018/MAX4020, f = 10MHz, V <sub>OUT</sub> = 20mV <sub>P-P</sub>			0.1		dB	
Amplifier Crosstalk	X <sub>TALK</sub>	MAX4016/MAX4018/M f = 10MHz, V <sub>OUT</sub> = 2V <sub>F</sub>		-95		dB		

**Note 1:** The MAX4012EUT is 100% production tested at  $T_A = +25^{\circ}C$ . Specifications over temperature limits are guaranteed by design.

Note 2: Tested with  $V_{CM}$  = 2.5V.

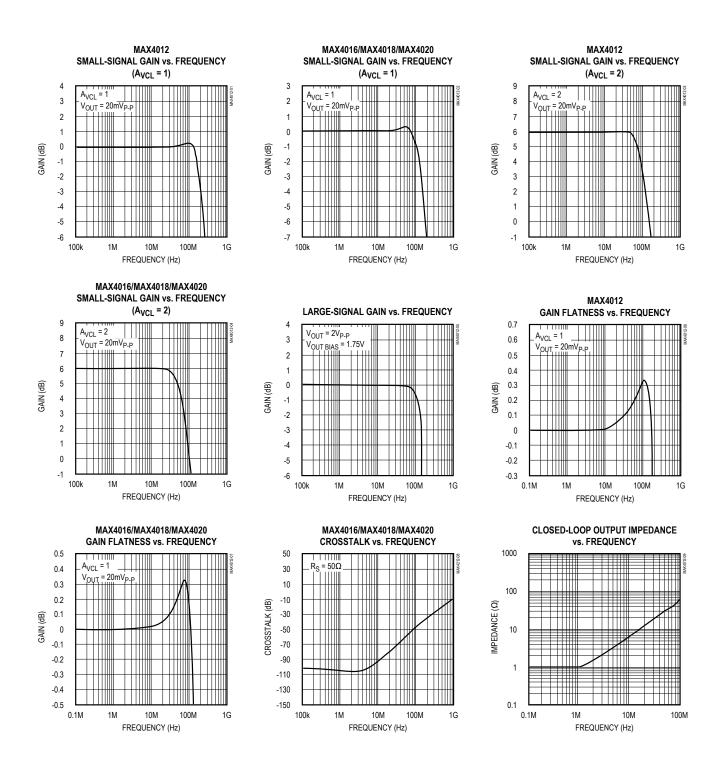
**Note 3:** PSR for single 5V supply tested with  $V_{EE} = 0$ ,  $V_{CC} = 4.5V$  to 5.5V; for dual ±5V supply with  $V_{EE} = -4.5V$  to -5.5V,  $V_{CC} = 4.5V$  to 5.5V; and for single 3.3V supply with  $V_{EE} = 0$ ,  $V_{CC} = 3.15V$  to 3.45V.

**Note 4:** Does not include the external feedback network's impedance.

Note 5: Guaranteed by design.

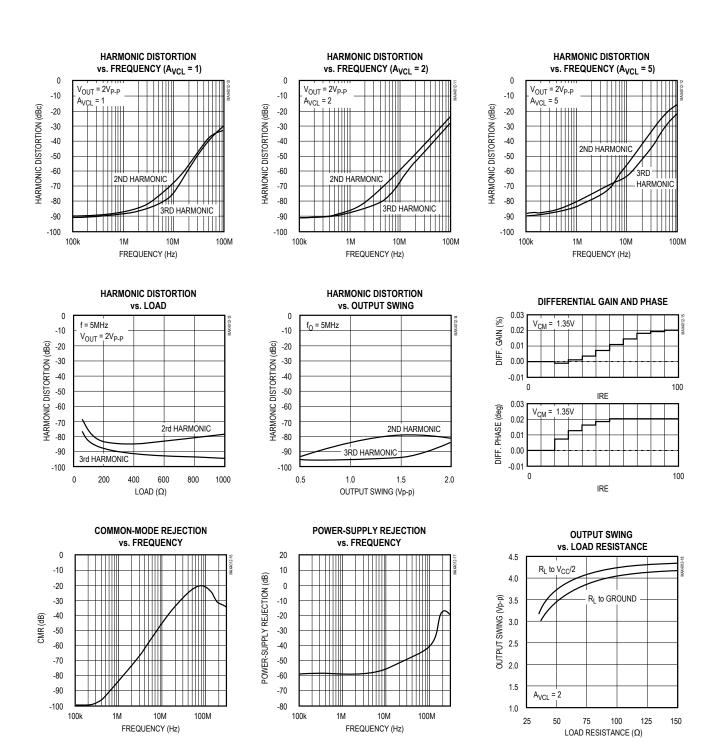
# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Typical Operating Characteristics**



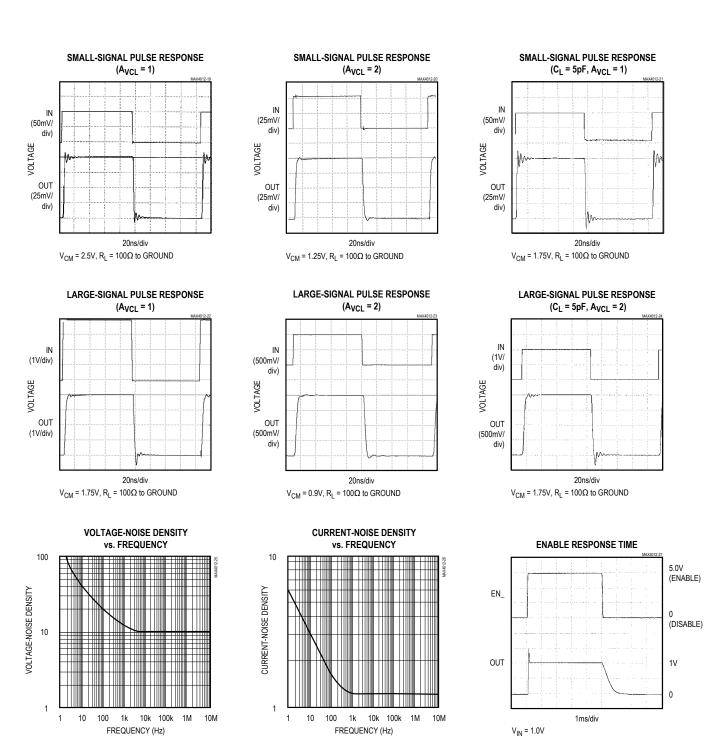
# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Typical Operating Characteristics (continued)**



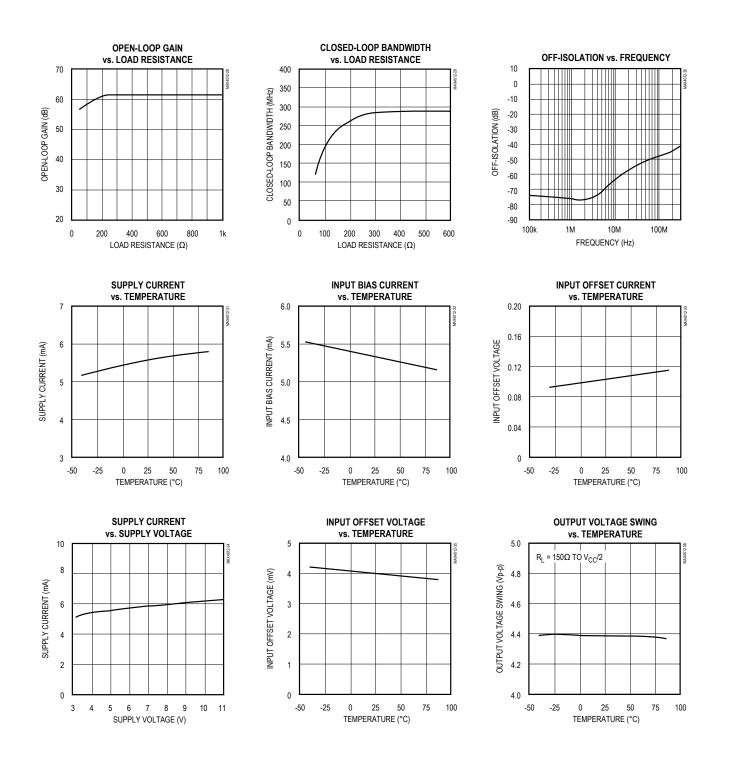
# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Typical Operating Characteristics (continued)**



# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Typical Operating Characteristics (continued)**



# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

# **Pin Description**

	PIN							
MAX4012	MAX4012	MAX4016	MAX	4018	MAX	4020	NAME	FUNCTION
SO-8	SOT23	SO/µMAX	SO	QSOP	SO	QSOP		
1, 5, 8	_	_	—	8, 9	—	8, 9	N.C.	No Connection. Not internally connected. Tie to ground or leave open.
6	1	—	_	—		_	OUT	Amplifier Output
4	2	4	11	13	11	13	$V_{EE}$	Negative Power Supply or Ground (in single- supply operation)
3	3	—		—		—	IN+	Noninverting Input
2	4	—	_	_	_	_	IN-	Inverting Input
7	5	8	4	4	4	4	V <sub>CC</sub>	Positive Power Supply
—	_	1	7	7	1	1	OUTA	Amplifier A Output
—	—	2	6	6	2	2	INA-	Amplifier A Inverting Input
—	_	3	5	5	3	3	INA+	Amplifier A Noninverting Input
—	_	7	8	10	7	7	OUTB	Amplifier B Output
—	_	6	9	11	6	6	INB-	Amplifier B Inverting Input
—	_	5	10	12	5	5	INB+	Amplifier B Noninverting Input
—	_	—	14	16	8	10	OUTC	Amplifier C Output
—	_	—	13	15	9	11	INC-	Amplifier C Inverting Input
—	_	—	12	14	10	12	INC+	Amplifier C Noninverting Input
_	_	—	_	_	14	16	OUTD	Amplifier D Output
_	_	—	_	_	13	15	IND-	Amplifier D Inverting Input
_	_	—	_	—	12	14	IND+	Amplifier D Noninverting Input
_	_	_				_	EN	Enable Amplifier
_	_	_	1	1			ENA	Enable Amplifier A
_	_	_	3	3		_	ENB	Enable Amplifier B
_	_	_	2	2			ENC	Enable Amplifier C

# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Detailed Description**

The MAX4012/MAX4016/MAX4018/MAX4020 are single-supply, rail-to-rail, voltage-feedback amplifiers that employ current-feedback techniques to achieve 600V/µs slew rates and 200MHz bandwidths. Excellent harmonic distortion and differential gain/phase performance make these amplifiers an ideal choice for a wide variety of video and RF signal-processing applications.

The output voltage swing comes to within 50mV of each supply rail. Local feedback around the output stage assures low open-loop output impedance to reduce gain sensitivity to load variations. This feedback also produces demand-driven current bias to the output transistors for  $\pm 120$ mA drive capability, while constraining total supply current to less than 7mA. The input stage permits common-mode voltages beyond the negative supply and to within 2.25V of the positive supply rail.

### **Applications Information**

#### **Choosing Resistor Values**

#### **Unity-Gain Configuration**

The MAX4012/MAX4016/MAX4018/MAX4020 are internally compensated for unity gain. When configured for unity gain, the devices require a  $24\Omega$  resistor (R<sub>F</sub>) in series with the feedback path. This resistor improves

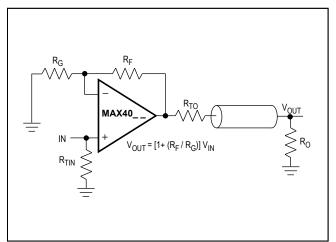


Figure 1a. Noninverting Gain Configuration

AC response by reducing the Q of the parallel LC circuit formed by the parasitic feedback capacitance and inductance.

#### Inverting and Noninverting Configurations

Select the gain-setting feedback ( $R_F$ ) and input ( $R_G$ ) resistor values to fit your application. Large resistor values increase voltage noise and interact with the amplifier's input and PC board capacitance. This can generate undesirable poles and zeros and decrease bandwidth or cause oscillations. For example, a noninverting gain-oftwo configuration ( $R_F = R_G$ ) using 1k $\Omega$  resistors, combined with 1pF of amplifier input capacitance and 1pF of PC board capacitance, causes a pole at 159MHz. Since this pole is within the amplifier bandwidth, it jeopardizes stability. Reducing the  $1k\Omega$  resistors to  $100\Omega$  extends the pole frequency to 1.59GHz, but could limit output swing by adding  $200\Omega$  in parallel with the amplifier's load resistor. Table 1 shows suggested feedback, gain resistors, and bandwidth for several gain values in the configurations shown in Figures 1a and 1b.

#### Layout and Power-Supply Bypassing

These amplifiers operate from a single 3.3V to 11V power supply or from dual supplies to  $\pm 5.5$ V. For single-supply operation, bypass V<sub>CC</sub> to ground with a 0.1µF capacitor as close to the pin as possible. If operating with dual supplies, bypass each supply with a 0.1µF capacitor.

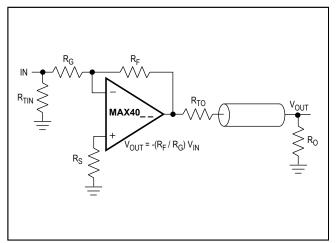


Figure 1b. Inverting Gain Configuration

### Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

Analog Devices recommends using microstrip and stripline techniques to obtain full bandwidth. To ensure that the PC board does not degrade the amplifier's performance, design it for a frequency greater than 1GHz. Pay careful attention to inputs and outputs to avoid large parasitic capacitance. Whether or not you use a constantimpedance board, observe the following guidelines when designing the board:

- Don't use wire-wrap boards because they are too inductive.
- Don't use IC sockets because they increase parasitic capacitance and inductance.
- Use surface-mount instead of through-hole components for better high-frequency performance.
- Use a PC board with at least two layers; it should be as free from voids as possible.
- Keep signal lines as short and as straight as possible. Do not make 90° turns; round all corners.

#### Rail-to-Rail Outputs, Ground-Sensing Input

The input common-mode range extends from (V<sub>EE</sub> - 200mV) to (V<sub>CC</sub> - 2.25V) with excellent common-mode rejection. Beyond this range, the amplifier output is a non-linear function of the input, but does not undergo phase reversal or latchup.

The output swings to within 60mV of either power-supply rail with a  $2k\Omega$  load. The input ground-sensing and the rail-to-rail output substantially increase the dynamic range. With a symmetric input in a single 5V application, the input can swing  $2.95V_{P-P}$ , and the output can swing  $4.9V_{P-P}$  with minimal distortion.

#### **Enable Input and Disabled Output**

The enable feature (EN\_) allows the amplifier to be placed in a low-power, high-output-impedance state. Typically, the EN\_ logic low input current (I<sub>IL</sub>) is small. However, as the EN voltage (V<sub>IL</sub>) approaches the negative supply rail, I<sub>IL</sub> increases (Figure 2). A single resistor connected as shown in Figure 3 prevents the rise in the logic-low input current. This resistor provides a feedback mechanism that increases V<sub>IL</sub> as the logic input is brought to V<sub>EE</sub>. Figure 4 shows the resulting input current (I<sub>IL</sub>).

When the MAX4018 is disabled, the amplifier's output impedance is  $35k\Omega$ . This high resistance and the low 2pF output capacitance make this part ideal in RF/video multiplexer or switch applications. For larger arrays, pay careful attention to capacitive loading. See the *Output Capacitive Loading and Stability* section for more information.

#### GAIN (V/V) COMPONENT +1 -1 +2 -2 +5 -5 +10 -10 +25 -25 500 $R_F(\Omega)$ 24 500 500 500 500 500 500 500 1200 $R_{G}(\Omega)$ ∞ 500 500 250 124 100 56 50 20 50 $R_{S}(\Omega)$ 0 \_ 0 0 0 0 62 100 $R_{TIN}(\Omega)$ 49.9 56 49.9 49.9 49.9 ∞ 49.9 ∞ $R_{TO}(\Omega)$ 49.9 49.9 49.9 49.9 49.9 49.9 49.9 49.9 49.9 49.9 Small-Signal -3dB Bandwidth (MHz) 200 90 105 60 25 33 11 25 6 10

### **Table 1. Recommended Component Values**

**Note:**  $R_L = R_O + R_{TO}$ ;  $R_{TIN}$  and  $R_{TO}$  are calculated for 50 $\Omega$  applications. For 75 $\Omega$  systems,  $R_{TO} = 75\Omega$ ; calculate RTIN from the following equation:

$$R_{TIN} = \frac{75}{1 - \frac{75}{R_G}} \Omega$$

# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

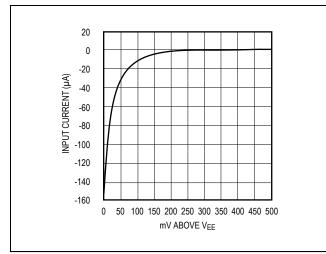


Figure 2. Enable Logic-Low Input Current vs. VIL

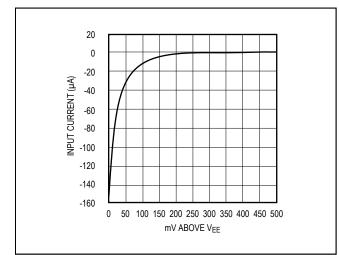


Figure 4. Enable Logic-Low Input Current vs.  $V_{I\!L}$  with  $10k\Omega$  Series Resistor

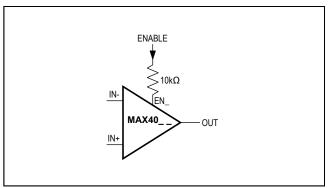


Figure 3. Circuit to Reduce Enable Logic-Low Input Current

To implement the mux function, the outputs of multiple amplifiers can be tied together, and only the amplifier with the selected input will be enabled. All of the other amplifiers will be placed in the low-power shutdown mode, with their high output impedance presenting very little load to the active amplifier output. For gains of +2 or greater, the feedback network impedance of all the amplifiers used in a mux application must be considered when calculating the total load on the active amplifier output

### **Output Capacitive Loading and Stability**

The MAX4012/MAX4016/MAX4018/MAX4020 are optimized for AC performance. They are not designed to drive highly reactive loads, which decreases phase margin and may produce excessive ringing and oscillation. Figure 5 shows a circuit that eliminates this problem. Figure 6 is a graph of the optimal isolation resistor ( $R_S$ ) vs. capacitive load. Figure 7 shows how a capacitive load causes excessive peaking of the amplifier's frequency response if the capacitor is not isolated from the amplifier by a resistor. A small isolation resistor (usually 20 $\Omega$  to 30 $\Omega$ ) placed before the reactive load prevents ringing and oscillation. At higher capacitive loads, AC performance is controlled by the interaction of the load capacitance and the isolation resistor. Figure 8 shows the effect of a 27 $\Omega$  isolation resistor on closed-loop response.

Coaxial cable and other transmission lines are easily driven when properly terminated at both ends with their characteristic impedance. Driving back-terminated transmission lines essentially eliminates the line's capacitance.

# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

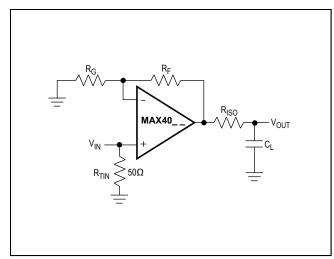


Figure 5. Driving a Capacitive Load through an Isolation Resistor

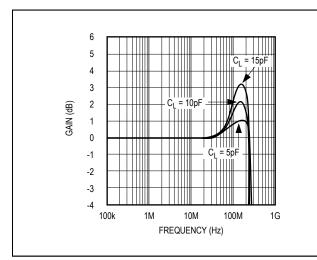


Figure 7. Small-Signal Gain vs. Frequency with Load Capacitance and No Isolation Resistor

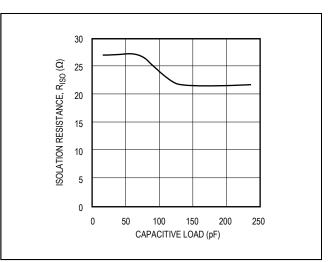


Figure 6. Capacitive Load vs. Isolation Resistance

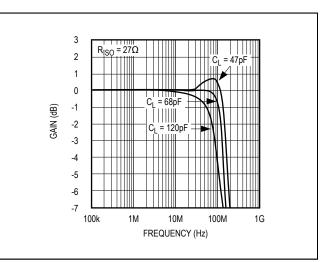
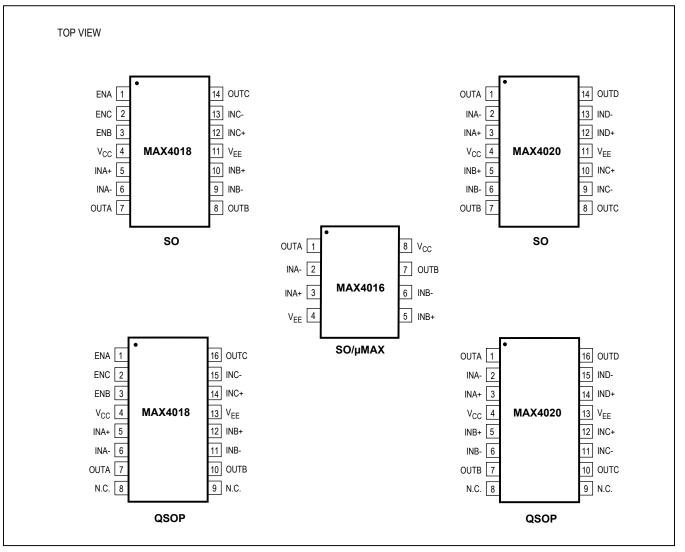


Figure 8. Small-Signal Gain vs. Frequency with Load Capacitance and  $27\Omega$  Isolation Resistor

# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Pin Configurations (continued)**



# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Ordering Information (continued)**

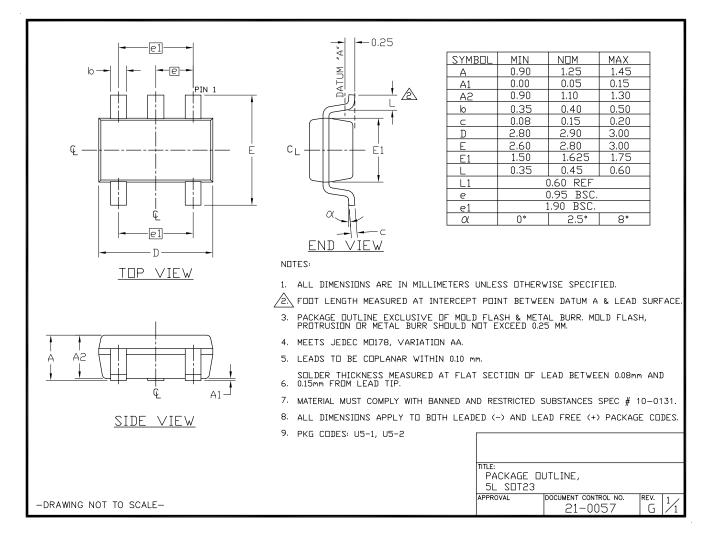
PART	TEMP	PIN-	TOP
PARI	RANGE	PACKAGE	MARK
MAX4018ESD+T	-40°C to +85°C	14 SO	—
MAX4018EEE+T	-40°C to +85°C	16 QSOP	—
MAX4020ESD+T	-40°C to +85°C	14 SO	—
MAX4020EEE+T	-40°C to +85°C	16 QSOP	_

+ Denotes a lead(Pb)-free/RoHS-compliant package.

T Denotes tape-and-reel.

### **Package Information**

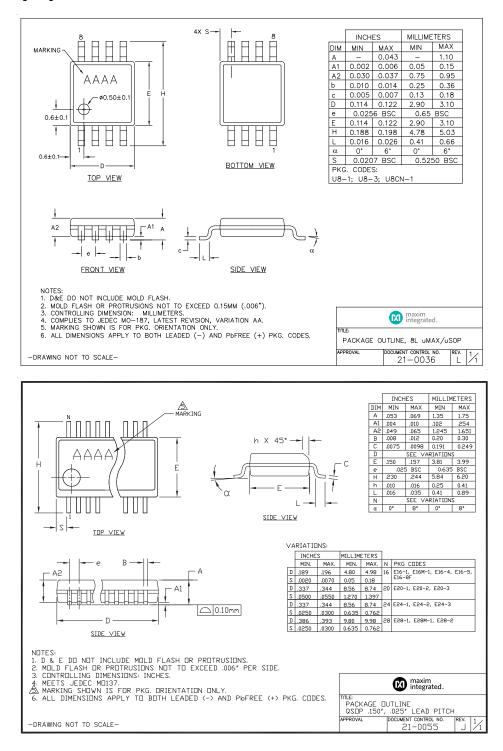
For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.



# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Package Information (continued)**

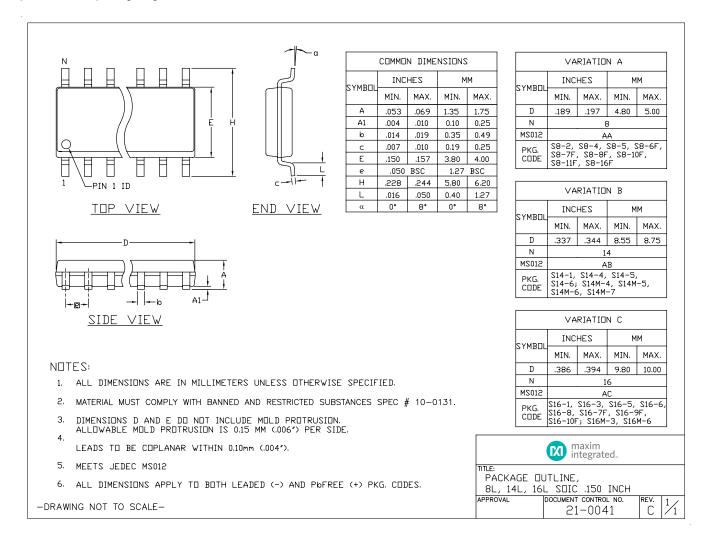
For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.



# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Package Information (continued)**

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.



# Low-Cost, High-Speed, Single-Supply Op Amps with Rail-to-Rail Outputs

### **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	6/97	Initial release	—
1	8/01	Updated Electrical Characteristics table	2
2	10/03	Updated Electrical Characteristics table	3
3	8/04	Added 8 SO package	All
4	8/22	Updated Ordering Information table, deleted Chip Information	1, 15



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