# QUICK START GUIDE FOR DEMONSTRATION CIRCUIT 733A HIGH LINEARITY UPCONVERTING MIXER

LT5519

### DESCRIPTION

Demonstration circuit 733A is an upconverting mixer featuring the LT®5519. The LT®5519 is a 0.7GHz to 1.4GHz High Linearity Upconverting Mixer optimized for wireless and cable infrastructure applications. A high-speed, internally matched LO amplifier drives a double-balanced mixer core, allowing the use of a low power, single-ended LO source.

An RF output transformer is integrated, thus eliminating the need for external matching components at the RF output, while reducing system cost, component count, board area, and system-level variations.

The IF port can be easily matched to a broad range of frequencies for use in many different applications.

Demonstration circuit 733A is designed for an RF output frequency range from 0.7GHz to 1.4GHz and is optimized for a 140MHz IF input frequency.

Design files for this circuit board are available. Call the LTC factory.

Table 1. Typical Performance Summary  $(T_A = 25^{\circ}C)$ 

PARAMETER	CONDITION ( $f_{ie} = 140MHz$ , $f_{io} = 1140MHz$ )	VALUE
Supply Voltage		4.5V to 5.25V
Supply Current	V <sub>CC</sub> = 5V, EN = High	60mA
Maximum Shutdown Current	V <sub>CC</sub> = 5V, EN = Low	100μΑ
Frequency Range		0.7GHz to 1.4GHz
IF Input Return Loss	$Z_0 = 50$ , with external matching	20dB
LO Input Return Loss	$Z_{0} = 50$	17dB
RF Input Return Loss	$Z_{0} = 50$	20dB
LO Input Power		-10dBm to 0dBm
Conversion Gain	$P_{IF} = -10 dBm, P_{LO} = -5 dBm$	-0.6dB
SSB Noise Figure	$P_{LO} = -5dBm$	13.6dB
Input 3 <sup>rd</sup> Order Intercept	2-Tone, -10dBm/Tone, $\Delta f = 1$ MHz, $P_{LO} = -5$ dBm	17.1dBm
Input 2 <sup>nd</sup> Order Intercept	1-Tone, -10dBm, P <sub>LO</sub> = -5dBm	48dBm
Input 1dB Compression	$P_{LO} = -5dBm$	5.5dBm
LO to RF leakage	$P_{LO} = -5dBm$	-44dBm
LO to IF leakage	$P_{LO} = -5dBm$	-40dBm
IF Common Mode Voltage	Internally biased	1.77V



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### **APPLICATION NOTE**

#### **FREQUENCY RANGE**

Demonstration circuit 733A is optimized for an IF input frequency of 140MHz. This frequency is set by the input IF matching components on the PCB. Other values may be used to maintain best performance for IF frequencies ranging from 1MHz to 400MHz.

#### **CURRENT CONSUMPTION**

If lower power consumption is required, the LT5519's supply current can be reduced by increasing the value of

the DC return resistors, R1 & R2. Operation at a lower supply current will, however, degrade linearity.

#### LO TO RF LEAKAGE

Minimum LO to RF leakage is realized when R1 & R2 are closely matched; 0.1% tolerance resistors are recommended for this reason. Resistors with a greater tolerance (ie; 1%) may be used with some degradation of LO to RF leakage.

## **QUICK START PROCEDURE**

Demonstration circuit 733A is easy to set up to evaluate the performance of the LT5519. Refer to 0 for proper measurement equipment setup and follow the procedure below:

#### NOTE:

- a. Use high performance signal generators with low harmonic output for 2<sup>nd</sup> & 3<sup>rd</sup> order distortion measurements. Otherwise, low-pass filters at the signal generator outputs should be used to suppress harmonics, particularly the 2<sup>nd</sup> harmonic.
- b. High quality combiners that provide a 50 termination on all ports and have good port-to-port isolation should be used. Attenuators on the outputs of the signal generators are recommended to further improve source isolation and to reduce reflection into the sources.
- 1. Connect all test equipment as shown in Figure 1.
- 2. Set the DC power supply's current limit to 70mA, and adjust output voltage to 5V.
- 3. Connect Vcc to the 5V DC supply, and then connect EN to 5V; the mixer is enabled (on).
- 4. Set Signal Generator #1 to provide an 1140MHz, -5dBm, CW signal to the demo board LO input port.

- 5. Set the Signal Generators #2 and #3 to provide two 10dBm CW signals to the demo board RF input port—one at 140MHz, and the other at 141MHz.
- **6.** To measure 3<sup>rd</sup> order distortion and conversion gain, set the Spectrum Analyzer start and stop frequencies to 997MHz and 1002MHz, respectively. Sufficient spectrum analyzer input attenuation should be used to avoid saturating the instrument.
- 7. The  $3^{rd}$  order intercept point is equal to  $(P_1 P_3) / 2 + P_{in}$ , where  $P_1$  is the power level of the two fundamental output tones at 999MHz and 1000MHz,  $P_3$  is the  $3^{rd}$  order product at 998MHz and 1001MHz, and  $P_{in}$  is the input power (in this case, -10dBm). All units are in dBm.
- 8. To measure input 2<sup>nd</sup> order distortion, set the Spectrum Analyzer start and stop frequencies to 859MHz and 861MHz, respectively. Sufficient spectrum analyzer input attenuation should be used to avoid saturating the instrument.
- 9. The  $2^{nd}$  order intercept point is equal to  $P_1 P_2 + Pin$ , where  $P_1$  is the power level of the fundamental output tone at 1000MHz,  $P_2$  is the  $2^{nd}$  order product at 860MHz, and Pin is the input power (in this case, -10dBm). All units are in dBm



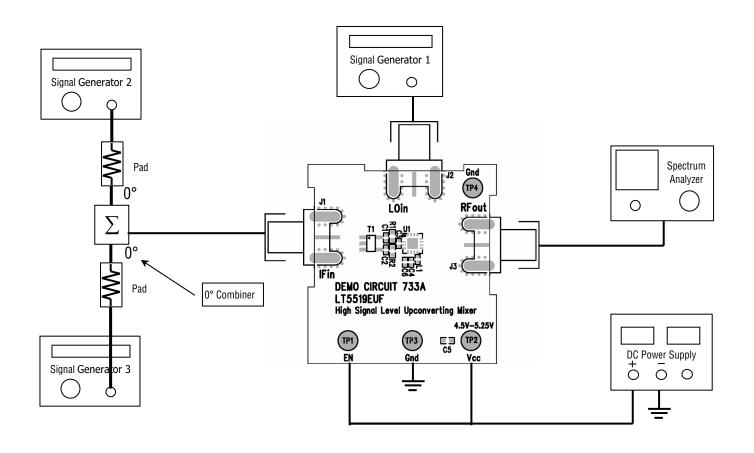
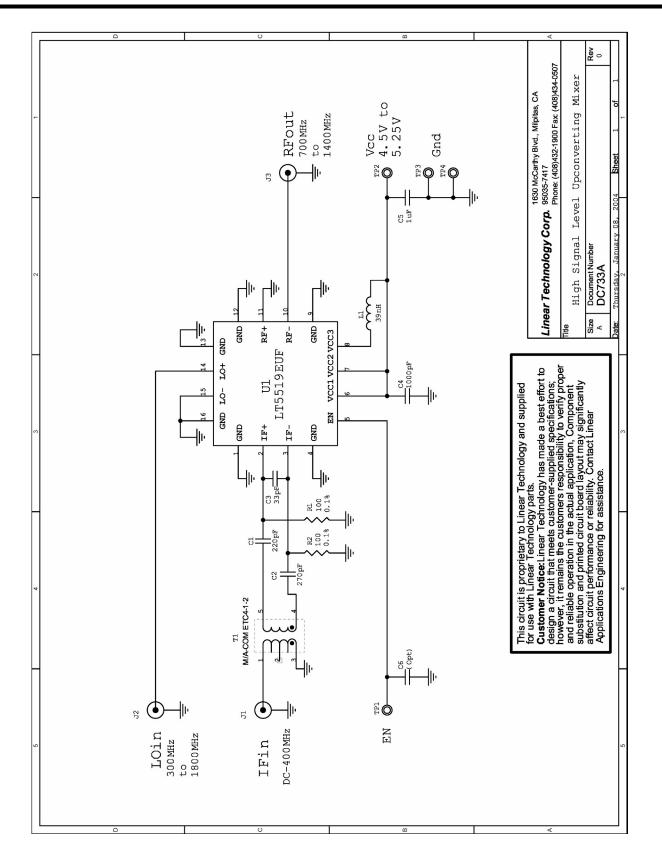


Figure 1. Proper Measurement Equipment Setup



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