

## Integrated Protection IC on 12V Bus with an Integrated MOSFET, Lossless Current Sensing, and PMBus Interface

**MAX16545B/  
MAX16545C**

### General Description

The MAX16545B/C is a circuit-breaker protection IC with an integrated low-resistance MOSFET and lossless current-sense circuitry featuring PMBus™/SMBus control and reporting. The IC is designed to provide the optimum solution for distribution, control, monitoring and protection of the system's 12V power supply. An internal LDO provides the bias supply voltage for the protection IC.

If no fault is detected, the MAX16545B/C initiates the startup. The device has been designed to provide controlled, monotonic startup. Programmable soft-start ramp and delay is implemented to limit the inrush current during startup. The MAX16545B is set for 16A Startup OCP, and the MAX16545C is set for 24A Startup OCP.

The IC monitors the current and the voltage of the 12V system power rail and provides multiple levels of protection with fast turn off if a fault is detected. Three methods of over-current protection are provided. Programmable moderate OCP level allows surge currents for a limited time. User-selectable severe OCP level provides a fast disconnect if a current exceeding the severe OCP threshold is detected. An additional fixed high shutdown OCP level provides instantaneous disconnect to further protect the device.

The lossless current sense provides high-accuracy current sensing over load and temperature, improving overall system energy efficiency, and reducing dissipation. A signal proportional to the load current is reported through an analog output and extensive reporting is provided through the SMBus/PMBus.

Output voltage is monitored at all times. If at any time the output voltage falls below the programmable output undervoltage lockout threshold, the PWRGD signal is asserted low. If at any time the input voltage falls below the programmable input undervoltage lockout threshold, the PWRGD signal is asserted low. The MAX16545B/C can be programmed through PMBus to provide input overvoltage protection. Input overvoltage protection is disabled by default. When enabled through PMBus, if the input voltage exceeds a programmable overvoltage threshold, the MOSFET is latched off and a fault is communicated.

The IC is used to distribute and control power to a system from the +12V power supply. See [Basic Application Circuit](#).

### Benefits and Features

- High Density (6.5mm x 4mm for 60A): Less than 25% of the Board Area of Conventional Solutions
  - Monolithic Integration of Power, Control, and Monitoring (Integrated Power MOSFET with 0.95mΩ Total Resistance in 12V Power Path (R<sub>DS(on)</sub>), Including Package)
  - Integrated Lossless, Precise Current Sensing
  - Integrated LDO Provides V<sub>DD</sub> Supply (1.8V Bias Supply)
- Enables Advanced System Power Management
  - PMBus/SMBus Telemetry with Extensive Status Monitoring and Reporting
  - Load Current Indicator (ILOAD) Pin Provides Analog Output Current Reporting with High Accuracy
  - Programmable Soft-Start for Inrush Current Limiting
  - Increases Power-Supply Reliability with IC Self-Protection Features
    - Very Fast Fault Detection and Isolation
    - V<sub>IN</sub> to V<sub>OUT</sub> Short Protection During Startup
    - Overtemperature Protection
  - Three Levels of Overcurrent Protection
    - Programmable Moderate OCP
    - Programmable Severe OCP Provides Isolation < 5μs.
    - Fail-Safe Safe OCP Provides Isolation < 250ns.

### Applications

Servers, Networking, Storage, Communication Equipment and AC/DC Power Supplies

- Integrated Protection IC on 12V: Circuit Breaker, Hot Swap, Inrush Control

[Ordering Information](#) appears at end of data sheet.

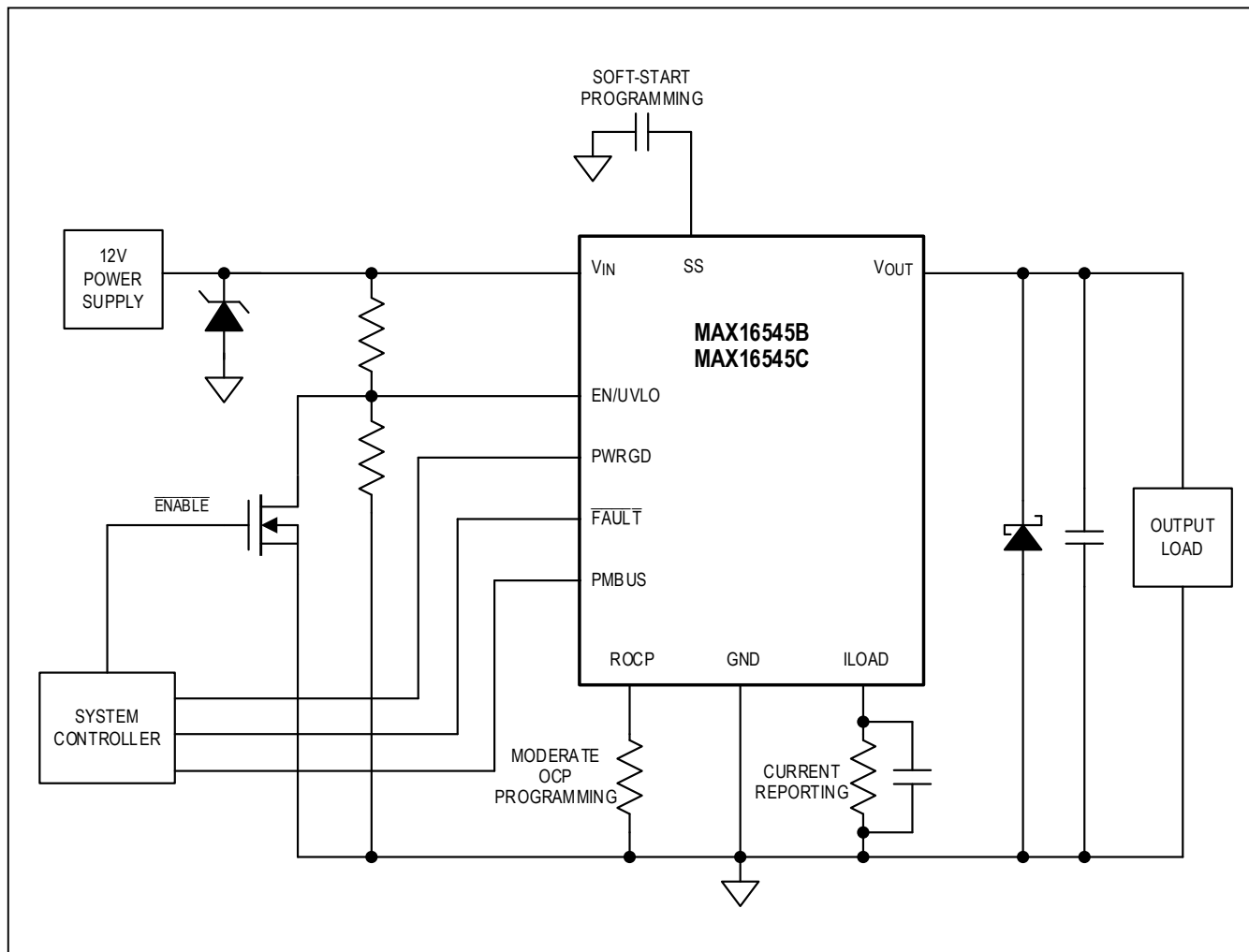
PMBus is a trademark of SMIF, Inc.

19-8779; Rev 6; 6/22

## MAX16545B MAX16545C

Integrated Protection IC on 12V Bus with an Integrated MOSFET, Lossless Current Sensing, and PMBus Interface

### Basic Application Circuit



# MAX16545B MAX16545C

## Integrated Protection IC on 12V Bus with an Integrated MOSFET, Lossless Current Sensing, and PMBus Interface

### Absolute Maximum Ratings

Supply Voltage ( $V_{IN}$ ) DC .....	-0.3V to +16V	ILOAD_IN, ILOAD .....	-0.3V to +2.5V
Supply Voltage ( $V_{IN}$ ) 150 $\mu$ s .....	+22V	Output Voltage ( $V_{OUT}$ ) DC .....	-0.3V to +16V
Bias Supply Voltage ( $V_{DD}$ ) .....	-0.3V to +2.5V	$V_{BST}$ (Relative to $V_{OUT}$ ) .....	-0.3V to +2.5V
PWRGD, FAULT .....	-0.3V to +5.5V	SS .....	-0.3V to +16V
EN/UVLO, ROCP .....	-0.3V to +2.5V	Junction Temperature ( $T_J$ ) .....	+150°C
SMBUS_CLK, SMBUS_DATA .....	-0.3V to +6V	Storage Temperature Range .....	-65°C to +165°C
SMBUS_ALERT .....	-0.3V to +6V	Peak Reflow Temperature .....	+260°C
SMBUS_ID .....	-0.3V to +2.5V		

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only; functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### Operating Ratings

Supply Voltage ( $V_{IN}$ ) .....	10.8V to 13.2V	Junction Temperature ( $T_J$ ) .....	0°C to +125°C
Bias Supply Voltage ( $V_{DD}$ ) .....	1.76V to 1.94V		

### Package Information

<b>PACKAGE TYPE: 22 FCQFN</b>	
Package Code	P224A6F+1
Outline Number	<a href="#">21-1006</a>
Land Pattern Number	<a href="#">90-0510</a>
<b>THERMAL RESISTANCE</b>	
Junction to Case ( $\theta_{JC}$ ) (max)	0.31°C/W

For the latest package outline information and land patterns (footprints), go to [www.maximintegrated.com/packages](http://www.maximintegrated.com/packages). Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

## Electrical Characteristics

( $V_{IN} = 12V \pm 10\%$ ,  $T_A = T_J = 0^\circ C$  to  $+125^\circ C$ , unless otherwise noted. Specifications are 100% production tested at  $T_A = T_J = +32^\circ C$ . Limits over operating temperature are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
12V SUPPLY (V <sub>IN</sub> )						
Supply Voltage Range	V <sub>IN</sub>		10.8	12	13.2	V
Supply Current	I <sub>IN</sub>	FET off	3.2		6.1	mA
		FET on: I <sub>OUT</sub> = 0 (Note 1)	3.7		6.5	
		FET on: I <sub>OUT</sub> = 60A (Note 1)			10	
INTEGRATED 1.8V LINEAR REGULATOR						
LDO Output Voltage Range	V <sub>DD</sub>		1.76	1.85	1.94	V
V <sub>DD</sub> UVLO, Falling			1.55	1.6	1.67	
V <sub>DD</sub> UVLO, Rising			1.62	1.67	1.71	
V <sub>DD</sub> UVLO Hysteresis		(Note 1)	30	60	80	mV
V <sub>DD</sub> UVLO Speed		(Note 1)		2		μs
UNDERVOLTAGE LOCKOUT: 12V SUPPLY (V <sub>IN</sub> )						
V <sub>IN_</sub> UVLO Rising Threshold	V <sub>IN_</sub> UVLO	At EN/UVLO Pin	0.95	1	1.05	V
V <sub>IN_</sub> UVLO Hysteresis				50		mV
Programmable 12V V <sub>IN</sub> Undervoltage Threshold for Rising Input	V <sub>IN_</sub> UVLO	Programmable through resistor-divider, measured at V <sub>IN</sub> with 1% resistors on the divider.	8			V
Response Time	t <sub>D</sub>	From EN/UVLO = 0V To FET Off (Note 1)		2		μs
EN/UVLO PIN INTERNAL PARAMETERS						
EN/UVLO Pin Leakage		EN/UVLO Pin = 1.8V	5	7.2	11	μA
EN/UVLO Internal-Pulldown Resistor			110	250	450	kΩ
BOOST VOLTAGE (V <sub>BST</sub> )						
V <sub>BST</sub> Charging Time to 1.6V above V <sub>OUT</sub>	t <sub>CHARGE</sub>	C <sub>BST</sub> = 100nF			275	μs
BST Voltage above V <sub>OUT</sub>	V <sub>BST</sub>			1.8		V
Rising UVLO Threshold above V <sub>OUT</sub>	BOOST UVLO		0.8	1.2	1.6	V
Falling UVLO Threshold above V <sub>OUT</sub>			0.7	1.1	1.5	
UVLO Hysteresis				60	130	400
Falling-Lockout Response Time	t <sub>D</sub>				2.3	μs
INTEGRATED MOSFET CHARACTERISTICS						
On-Resistance	R <sub>DS(ON)</sub>	(Note 1)		0.95		mΩ

### Electrical Characteristics (continued)

( $V_{IN} = 12V \pm 10\%$ ,  $T_A = T_J = 0^\circ C$  to  $+125^\circ C$ , unless otherwise noted. Specifications are 100% production tested at  $T_A = T_J = +32^\circ C$ . Limits over operating temperature are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>OVERCURRENT PROTECTION (OCP)</b>						
Gain from ROCP Pin Current to $I_{OUT}$	$G_{OCP}$	(Note 1)		8		A/ $\mu A$
Moderate OCP Voltage Threshold	$V_{OCPM}$	Referred to voltage over $R_{OCP}$ (Note 1)		0.8		V
Moderate OCP Threshold Accuracy		Referred to $I_{OUT}$ using Equation 1 (Note 1)		$\pm 12$		%
Moderate OCP Threshold Range	$I_{OCP}$ (MODERATE)	(Note 1)	30		70	A
Allowable Range for No Fault Detection	$R_{OCP}$		26		380	k $\Omega$
Moderate OCP Timeout	$t_{OCPM}$	$I_{OUT} >$ Moderate OCP, $I_{OUT} <$ Severe OCP PMBus Programmable, default 100 $\mu s$ (Note 1)		20		$\mu s$
				100		
				100		ms
				250		
Moderate OCP Fault Detect Delay Timeout Accuracy		(Note 1)		$\pm 25$		%
Severe OCP Threshold, % above Moderate OCP	$I_{OCP(SEVERE)}$	PMBus Programmable, default 130% (Note 1)		130		%
Severe OCP Threshold Accuracy		(Note 1)		150		
Severe OCP Delay	$t_{D(SEVERE)}$	From fault threshold exceeded to FET off		$\pm 15$		%
Safe OCP Threshold	$I_{OCP(SAFE)}$			10		$\mu s$
Safe OCP Threshold Accuracy				120		A
Safe OCP Delay	$t_{D(SAFE)}$	From fault threshold exceeded to FET off		$\pm 20$		%
Startup OCP Threshold	$I_{OCP(STARTUP)}$	PMBus programmable, MAX16545B default = 16A, MAX16545C default = 24A (Note 1)		250		ns
Startup OCP Threshold Accuracy		(Note 1)		16		A
Startup OCP Delay				24		
				8		
				$\pm 20$		%
	$t_{D(STARTUP)}$			10		$\mu s$
<b>SOFT-START AND <math>C_{OUT}</math></b>						
Soft-Start Discharge Resistance	SS			1	2	k $\Omega$
Soft-Start Charging Current	$I_{SS}$		24.8		38	$\mu A$
Soft-Start Time	$t_{SS}$	$C_{SS} = 0nF$ , $I_{OUT} = 0A$			1.5	ms

### Electrical Characteristics (continued)

( $V_{IN} = 12V \pm 10\%$ ,  $T_A = T_J = 0^\circ C$  to  $+125^\circ C$ , unless otherwise noted. Specifications are 100% production tested at  $T_A = T_J = +32^\circ C$ . Limits over operating temperature are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
Soft-Start Capacitor Discharge Threshold			0.10	0.21	0.29	V	
C <sub>SS</sub> Discharge Check Duration During Startup			2.2			s	
FET V <sub>GS</sub> UNDERVOLTAGE LOCKOUT							
V <sub>GS</sub> Rising UVLO Threshold	V <sub>GS_UVLO</sub>		0.87	1.3	1.61	V	
V <sub>GS</sub> Falling UVLO Threshold			0.74	1.17	1.47	V	
V <sub>GS</sub> UVLO Hysteresis			87	100	466	mV	
V <sub>GS</sub> UVLO Masking Time		V <sub>GS</sub> UVLO masking time at startup	90	100	110	ms	
PASS FET SHORT DETECTION DURING STARTUP							
Duration of V <sub>OUT</sub> Discharge Check	t <sub>DISCHARGE</sub>		2.2			s	
Pulldown Resistance on V <sub>OUT</sub> During Self-Check	R <sub>DISCHARGE</sub>		400	455	560	Ω	
Threshold for Self-Check Procedure	V <sub>OUT_SELF</sub> CHECK	PMBus Programmable, default 9V (Note 1)	5.8	6	6.2	V	
			6.8	7	7.2		
			7.8	8	8.2		
			9.0	9.2	9.4		
Delay from Self-Check Pass to Start V <sub>OUT</sub> Ramp	t <sub>D</sub>	PMBus Programmable, default 0.02ms (Note 1)	0.02			ms	
			10				
			20				
PWRGD PIN							
PWRGD Output-High Voltage	V <sub>OH</sub>	External pullup leakage below 1μA	5.5			V	
PWRGD Output-Low Voltage	V <sub>OL</sub>	Sinking 4mA	0.4			V	
Propagation Delay from V <sub>OUT_UVLO</sub> Detect to PWRGD Pin Asserted Low	t <sub>D</sub>		5			μs	
Rising Threshold for PWRGD Asserted High	V <sub>TH(PWRGD)</sub>	PMBus Programmable, default 11V (Note 1)	8			V	
			9				
			10				
			11				
Accuracy				-5	+5		%
PWRGD Assertion Delay After V <sub>OUT</sub> Settles			(Note 1)	5			ms
PWRGD Hysteresis			8V	0.36	0.41	0.46	V
PWRGD Hysteresis			9V	0.42	0.47	0.52	
PWRGD Hysteresis			10V	0.45	0.51	0.56	
PWRGD Hysteresis			11V	0.50	0.55	0.61	

### Electrical Characteristics (continued)

( $V_{IN} = 12V \pm 10\%$ ,  $T_A = T_J = 0^\circ C$  to  $+125^\circ C$ , unless otherwise noted. Specifications are 100% production tested at  $T_A = T_J = +32^\circ C$ . Limits over operating temperature are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>OVERTEMPERATURE PROTECTION (OTP)</b>						
Overtemperature Protection Threshold	$T_{OTP}$	PMBus programmable (Note 1)		135		$^\circ C$
<b>SMBus DATA (SMBUS_DATA) AND CLOCK (SMBUS_CLOCK) PINS</b>						
Leakage Current		For pullup voltage = 5.5V		6.2	10	$\mu A$
Input-Low Voltage	$V_{IL}$				0.8	V
Input-High Voltage	$V_{IH}$		2.1			V
Output-Low Voltage	$V_{OL}$	SMBUS_DATA, sinking 4mA			0.4	V
External-Pullup Voltage	$V_{OH}$ Max				5.5	V
<b>SMBUS_ALERT PIN</b>						
Output-Low Voltage	$V_{OL}$	Sinking 4mA			0.4	V
Leakage Current		For pullup = 5.5V			1	$\mu A$
External-Pullup Voltage	$V_{OH}$ Max				5.5	V
<b>FAULT ISOLATION DELAY</b>						
Fault-Detection Time	$t_D$	From fault detection to start of gate pulldown		60		ns
Pass FET Turn-Off Time	$t_F$	From start of gate pulldown to pass FET off		50		ns
<b>FAULT PIN</b>						
FAULT Input-High Voltage	$V_{IH}$		1			V
FAULT Input-Low Voltage	$V_{IL}$				0.66	V
FAULT Output-High Voltage	$V_{OH}$	External pullup, leakage below 1 $\mu A$			5.5	V
FAULT Output-Low Voltage	$V_{OL}$	Sinking 4mA			0.4	V
Propagation Delay from Fault Detect to FAULT Asserted Low	$t_d$				140	ns
<b><math>V_{IN}</math> OVERVOLTAGE DETECTION</b>						
Programmable Threshold for OV	$V_{IN\_OV}$	PMBus programmable, default 16V (Note 1)		14		V
				16		
				17		
				18		
Deglitching Time	$t_{FILTER\_OV}$			20		$\mu s$
Deglitching Time Accuracy				$\pm 5$		%

### Electrical Characteristics (continued)

( $V_{IN} = 12V \pm 10\%$ ,  $T_A = T_J = 0^\circ C$  to  $+125^\circ C$ , unless otherwise noted. Specifications are 100% production tested at  $T_A = T_J = +32^\circ C$ . Limits over operating temperature are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
CURRENT REPORTING (ILOAD) PIN						
Linear I <sub>OUT</sub> Reporting Range		Current reporting range from single MAX16545B/C device.	0		70	A
Linear-Voltage Range			0		1.35	V
Offset Current		I <sub>OUT</sub> = 0A		0.65		μA
Current-Reporting Gain	G <sub>ILOAD</sub>	ILOAD current divided by I <sub>OUT</sub> current (Note 1)		5		μA/A
Bandwidth		No capacitor in parallel with R <sub>LOAD</sub> resistor.	100			kHz
CURRENT REPORTING (ILOAD_IN) PIN						
Linear Current Range. Gain from ILOAD_IN Pin to ILOAD Pin = 1	I <sub>LOAD_IN</sub>	Only when used with VT503 (Note 2)	0		350	μA
ANALOG CURRENT REPORTING ACCURACY						
Analog Current-Reporting Accuracy	I <sub>OUT</sub>	I <sub>OUT</sub> = 60A (Note 2)	-1.3		+1.3	%
		I <sub>OUT</sub> = 8A (Note 2)	-1.8		+1.8	
SMBus/PMBus TELEMETRY ACCURACY						
Digital Current-Reporting Accuracy	READ_I <sub>OUT</sub> (8Ch), READ_I <sub>IN</sub> (89h)	I <sub>OUT</sub> = 60A (Note 2)	-1.6		+1.6	%
		I <sub>OUT</sub> = 8A (Note 2)	-2.3		+2.3	
Digital Power-Reporting Accuracy	READ_PIN (97h)	I <sub>OUT</sub> = 8A, V <sub>IN</sub> = 12V, R <sub>LOAD</sub> = 4.5kΩ (Note 2)	-3		+3	%
		I <sub>OUT</sub> = 60A, V <sub>IN</sub> = 12V, R <sub>LOAD</sub> = 4.5kΩ (Note 2)	-2		+2	
Digital Input-Voltage Reporting Accuracy	READ_V <sub>IN</sub> (88h)		-1		+1	%
Digital Output-Voltage Reporting Accuracy	READ_V <sub>OUT</sub> (8Bh)	10.8V < V <sub>OUT</sub> < 13.2V	-1		+1	%
		8V < V <sub>OUT</sub> < 10.8V	-3		+3	
Digital-Temperature Reporting Accuracy	READ_TEMPERATURE_1 (8Dh)	(Note 2)	-6		+6	°C
Digital-Energy Reporting Accuracy	READ_E <sub>IN</sub> (86h)	I <sub>OUT</sub> = 60A	-2.5		+2.5	%
		I <sub>OUT</sub> = 8A	-4.6		+4.6	

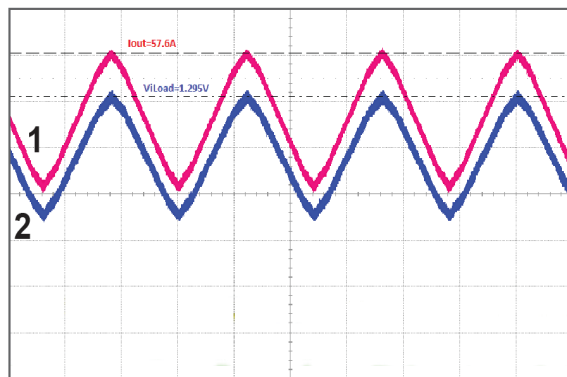
**Note 1:** Denotes specifications that apply for typical operating junction temperature ( $T_A = T_J = +32^\circ C$ ).

**Note 2:** Reporting accuracy presented includes 0.1% external resistor tolerance contribution.



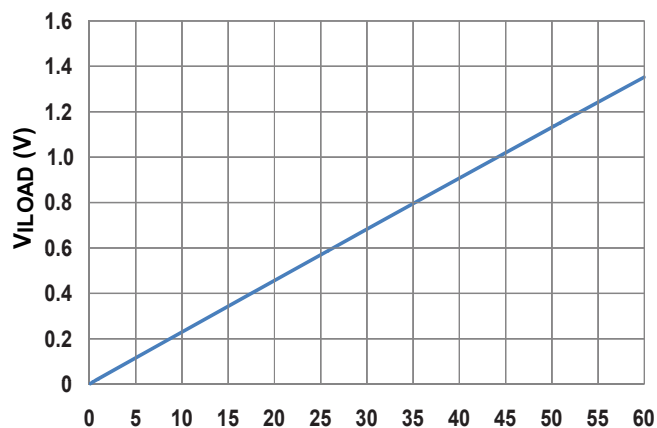
## Typical Operating Characteristics

**Output Current Monitoring and Reporting - Frequency Response**



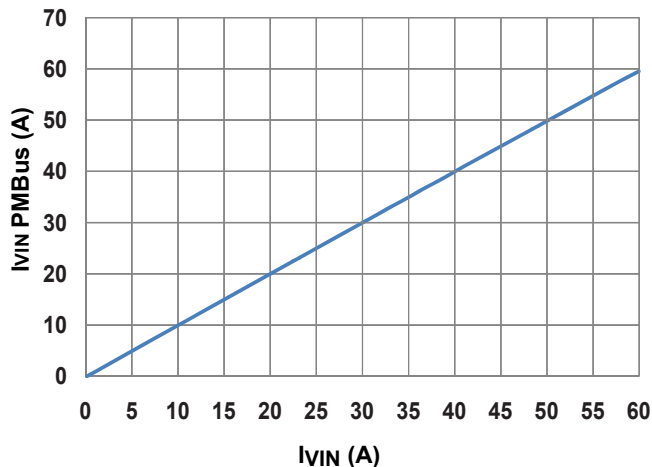
Time/div: 500µs  
Conditions:  $I_{OUT} = 0A$  to 60A  
1:  $I_{OUT}$  (20A/div)  
2:  $V_{ILOAD}$  (500mV/div)

**Output Current Monitoring and Reporting - Linearity**



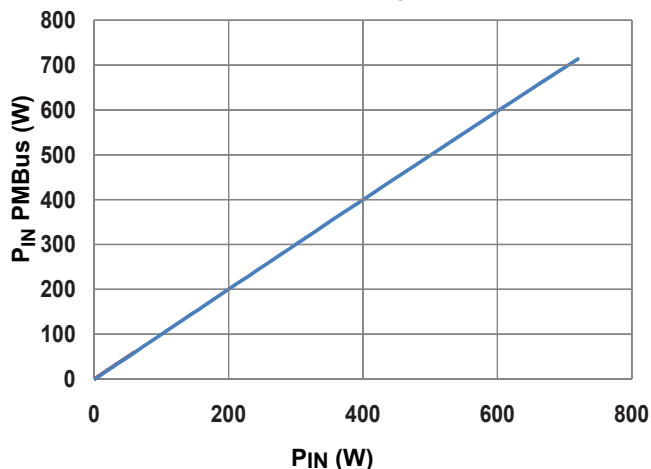
$I_{OUT}$  (A)  
Conditions:  $R_{LOAD} = 4.53k\Omega$  (0.1%)  
 $I_{OUT} = 0A$  to 60A

**Input Current Reporting - PMBus**



Conditions:  $I_{OUT} = 0A$  to 60A  
 $V_{IN} = 12V$   
Reporting of PMBus Register 89h

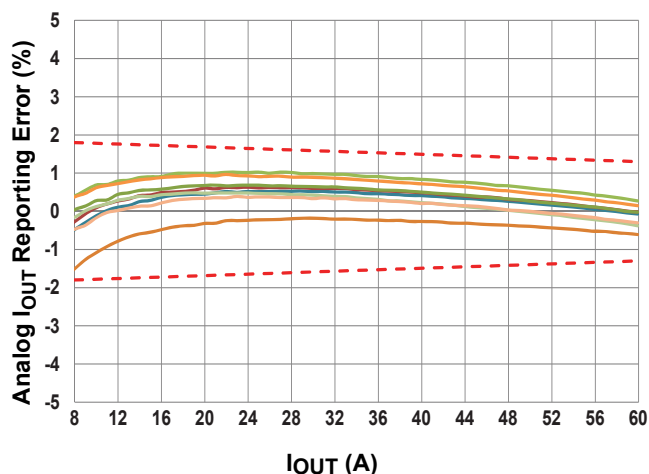
**Input Power Reporting - PMBus**



Conditions:  $I_{OUT} = 0A$  to 60A  
 $V_{IN} = 12V$   
Reporting of PMBus Register 97h

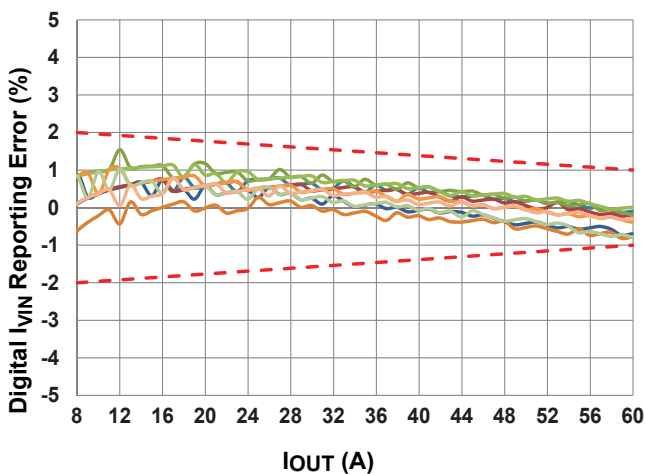
## Typical Operating Characteristics (continued)

**Analog  $I_{OUT}$  Reporting Accuracy**



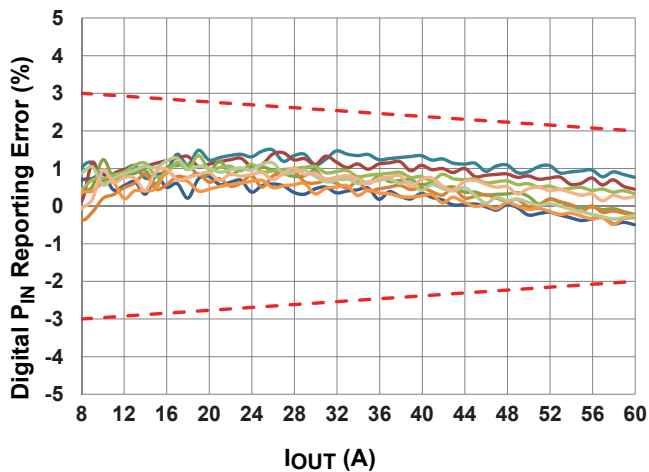
Conditions:  $I_{OUT}$  = 0-60A  
 $V_{IN}$  = 12V  
 $R_{LOAD}$  = 4.53k $\Omega$  (0.1%)  
Data on 8 Samples  
Limits as defined in EC table

**Digital  $I_{VIN}$  Reporting Accuracy**



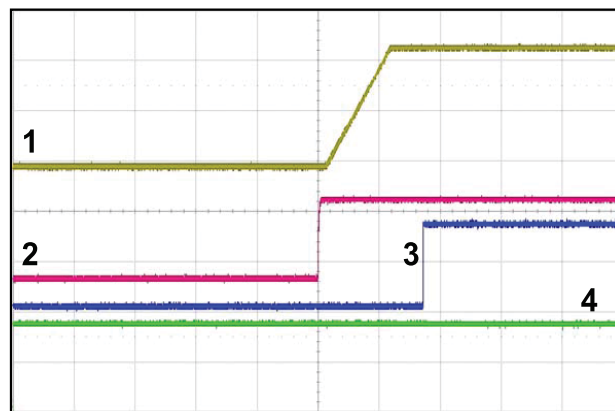
Conditions:  $I_{OUT}$  = 0A to 60A  
 $V_{IN}$  = 12V  
 $R_{LOAD}$  = 4.53k $\Omega$  (0.1%)  
Data on 8 Samples  
Limits as defined in EC table

**Digital Power Reporting Accuracy**



Conditions:  $I_{OUT}$  = 0-60A  
 $V_{IN}$  = 12V  
 $R_{LOAD}$  = 4.53k $\Omega$  (0.1%)  
Data on 8 Samples  
Limits as defined in EC table

**Startup Controlled with EN/UVLO - No Load**

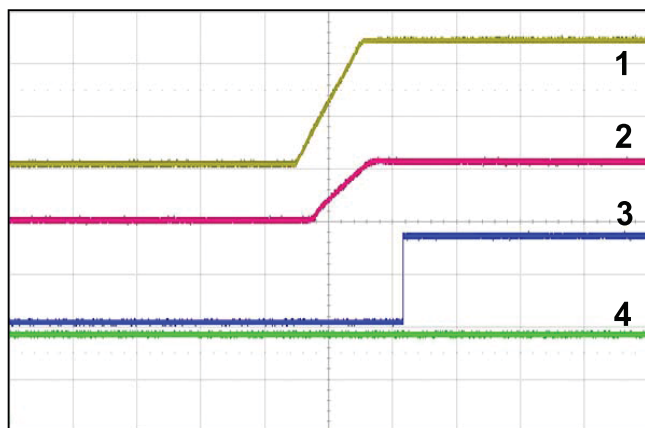


Time/Div: 20ms

Conditions:  $C_{SS}$  = 47nF  
 $C_{OUT}$  = 2 x 180 $\mu$ F + 6 x 10 $\mu$ F + 2 x 2.2 $\mu$ F  
1. VOUT (5V/div)  
2. EN/UVLO (1V/div)  
3. PWRGD (2V/div)  
4. FAULT (2V/div)

## Typical Operating Characteristics (continued)

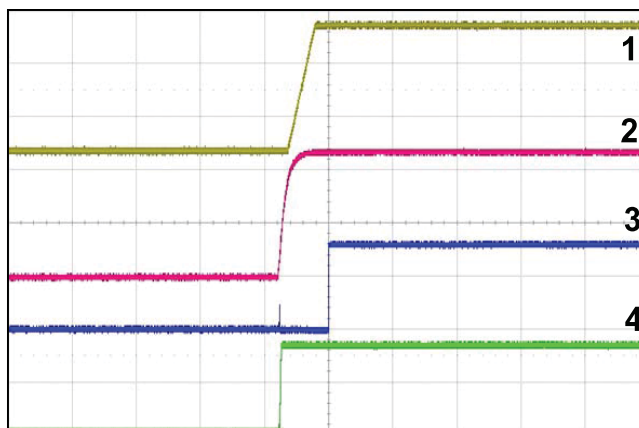
**Startup Into Load, Controlled with EN/UVLO -  
1Ω Resistive Load**



**Time/Div: 10ms**

Conditions:  $R_{LOAD} = 1\Omega$ , Present at Startup  
 $C_{SS} = 22nF$   
 $C_{OUT} = 2 \times 180\mu F + 6 \times 10\mu F + 2 \times 2.2\mu F$   
 1.  $V_{OUT}$  (5V/div)  
 2.  $I_{OUT}$  (10A/div)  
 3. PWRGD (2V/div)  
 4. FAULT (2V/div)

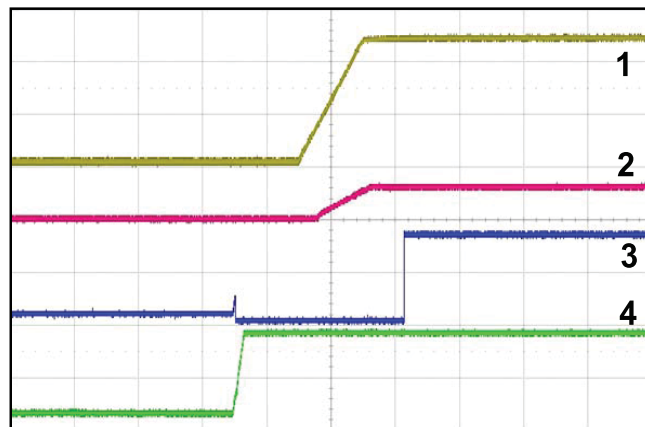
**Startup Controlled with  $V_{IN}$  - No Load**



**Time/Div: 50ms**

Conditions:  $C_{SS} = 47nF$   
 $C_{OUT} = 2 \times 180\mu F + 6 \times 10\mu F + 2 \times 2.2\mu F$   
 No Load at Startup  
 1.  $V_{OUT}$  (5V/div)  
 2.  $V_{IN}$  (5V/div)  
 3. PWRGD (2V/div)  
 4. FAULT (2V/div)

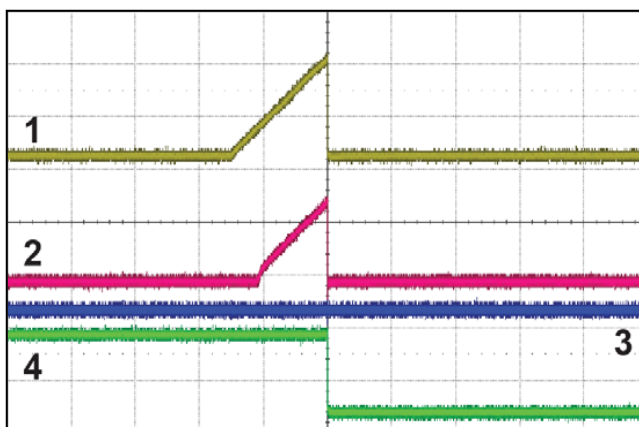
**Startup Into Load, Controlled with  $V_{IN}$  -  
1Ω Resistive Load**



**Time/Div: 10ms**

Conditions:  $C_{SS} = 22nF$   
 $C_{OUT} = 4 \times 330\mu F + 2 \times 180\mu F + 1 \times 10\mu F + 1 \times 2.2\mu F$   
 $R_{LOAD} = 1\Omega$  Present at Startup  
 1.  $V_{OUT}$  (5V/div)  
 2.  $I_{OUT}$  (20A/div)  
 3. PWRGD (2V/div)  
 4. FAULT (2V/div)

**Start Into Short Condition -  
 $V_{OUT}$  to GND Short Present at Startup**

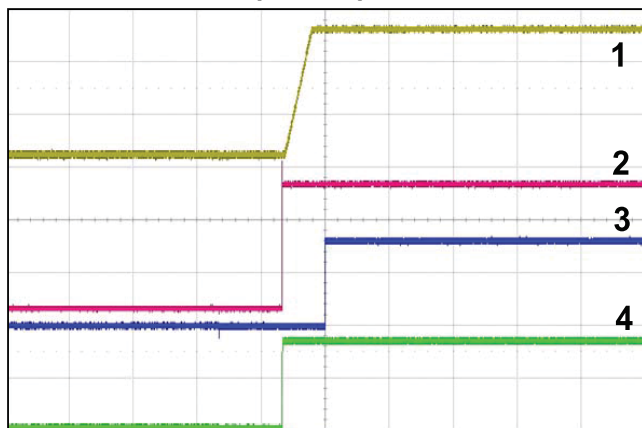


**Time/Div: 500μs**

Conditions:  $V_{OUT}$  Shorted to GND Prior to Startup  
 1.  $V_{OUT}$  (5V/div)  
 2.  $I_{IN}$  (10A/div)  
 3. PWRGD (2V/div)  
 4. FAULT (2V/div)

## Typical Operating Characteristics (continued)

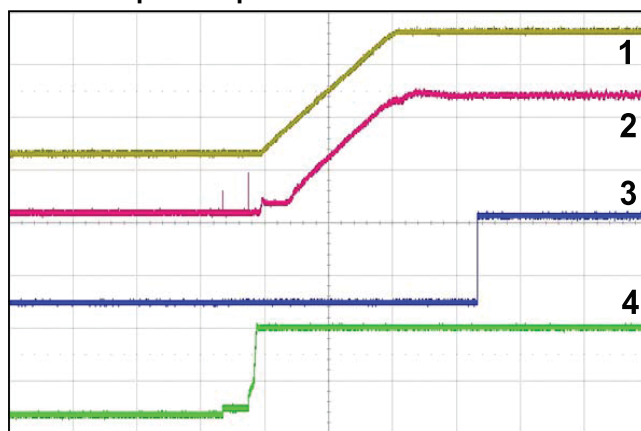
**Hot Swap Startup - No Load**



**Time/Div: 50ms**

Conditions:  $C_{SS} = 47\text{nF}$   
No Input Capacitors  
 $C_{OUT} = 2 \times 180\mu\text{F} + 6 \times 10\mu\text{F} + 2 \times 2.2\mu\text{F}$   
No Load Present at Startup  
1.  $V_{OUT}$  (5V/div)  
2.  $V_{IN}$  (5V/div)  
3. PWRGD (2V/div)  
4. FAULT (2V/div)

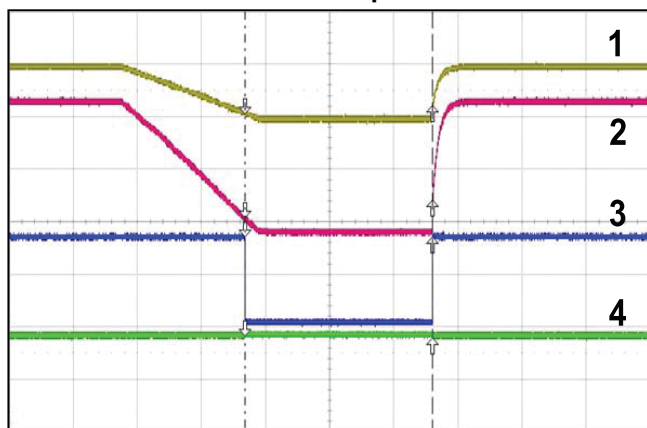
**Hot Swap Startup Into Load - 1Ω Resistive Load**



**Time/Div: 5ms**

Conditions:  $C_{SS} = 22\text{nF}$   
No Input Capacitors  
 $C_{OUT} = 4 \times 330\mu\text{F} + 2 \times 180\mu\text{F} + 1 \times 10\mu\text{F} + 1 \times 2.2\mu\text{F}$   
 $R_{LOAD} = 1\Omega$  Present at Startup  
1.  $V_{OUT}$  (5V/div)  
2.  $I_{OUT}$  (5A/div)  
3. PWRGD (2V/div)  
4. FAULT (2V/div)

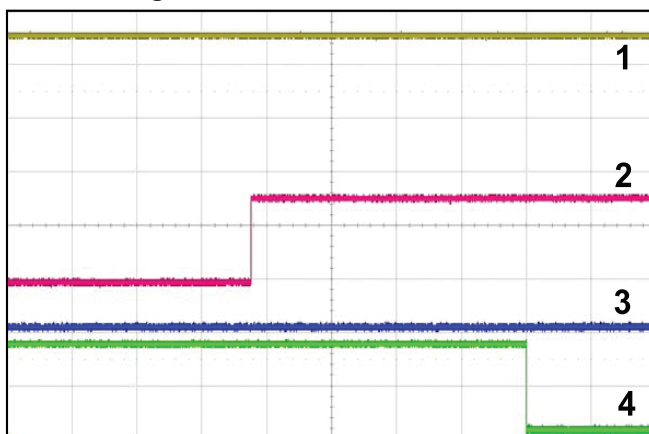
**PWRGD Response**



**Time/Div: 50ms**

Conditions:  $C_{SS} = 47\text{nF}$   
PWRGD Threshold = 8V  
 $V_{IN\_UVLO}$  Threshold = 6.2V  
No Load  
1.  $V_{OUT}$  (5V/div)  
2.  $V_{IN}$  (2V/div)  
3. PWRGD (2V/div)  
4. FAULT (2V/div)

**Integrated FET Self-Check Function**

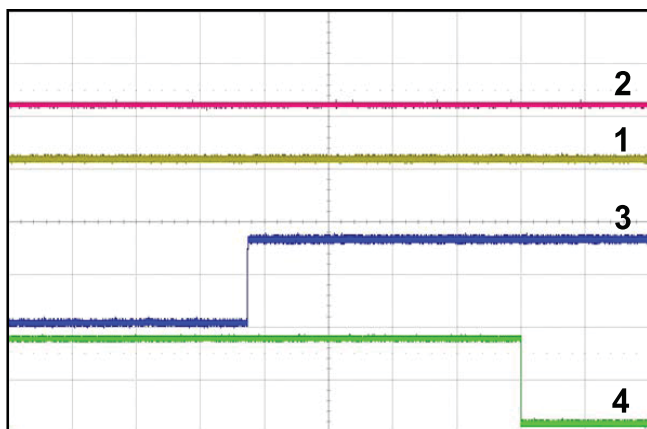


**Time/Div: 500ms**

Conditions:  $V_{OUT}$  Shorted to  $V_{IN}$  Prior to Startup  
1.  $V_{OUT}$  (5V/div)  
2. EN/UVLO (1V/div)  
3. PWRGD (2V/div)  
4. FAULT (2V/div)

## Typical Operating Characteristics (continued)

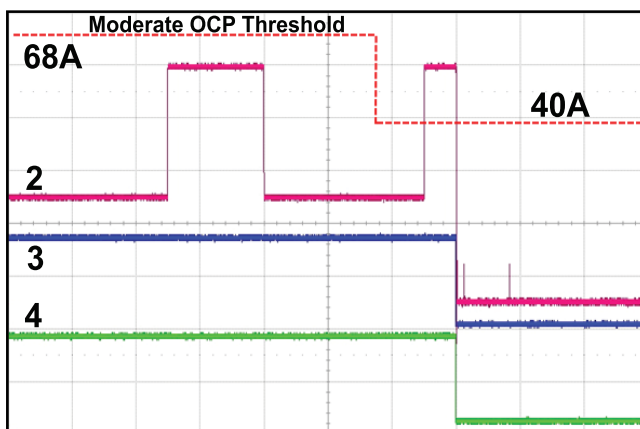
**C<sub>SS</sub> Discharge Fail Protection**



**Time/Div: 500ms**

Conditions: C<sub>SS</sub> = 47nF  
SS Shorted to V<sub>IN</sub> Prior to Startup  
1. V<sub>OUT</sub> (5V/div)  
2. SS (5V/div)  
3. EN/UVLO (1V/div)  
4. FAULT (2V/div)

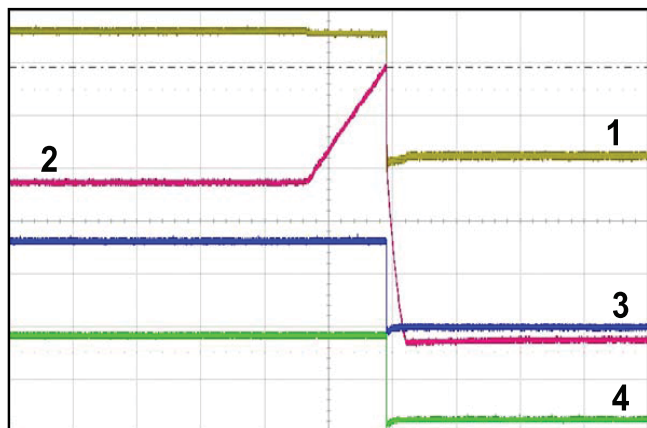
**R<sub>OCP</sub> Change During Operation**



**Time/Div: 200ms**

Conditions: OCPM<sub>High</sub> = 68A  
OCPM<sub>Low</sub> = 40A  
I<sub>OUT</sub> = 20A to 45A  
0.05A/μs  
t = 600ms  
Moderate OCP Timeout = 100ms  
2. I<sub>OUT</sub> (10A/div)  
3. PWRGD (2V/div)  
4. FAULT (2V/div)

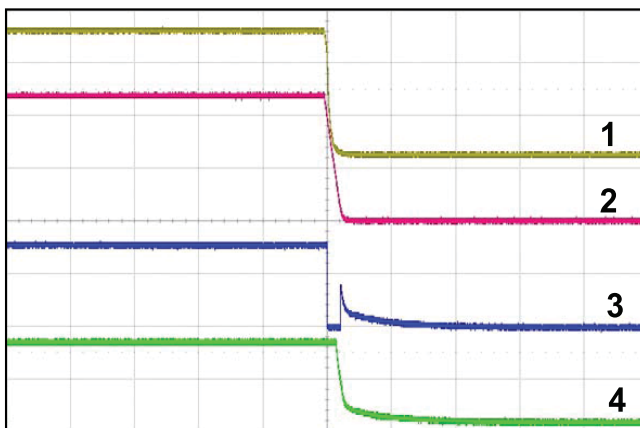
**Severe OCP**



**Time/Div: 200μs**

Conditions: I<sub>OUT</sub> = 60A to 120A  
OCPM = 68A  
Moderate OCP Timeout = 250ms  
Severe OCP Setting = 130%  
1. V<sub>OUT</sub> (5V/div)  
2. I<sub>OUT</sub> (20A/div)  
3. PWRGD (2V/div)  
4. FAULT (2V/div)

**Shutdown Controlled by V<sub>IN</sub> - No Load**

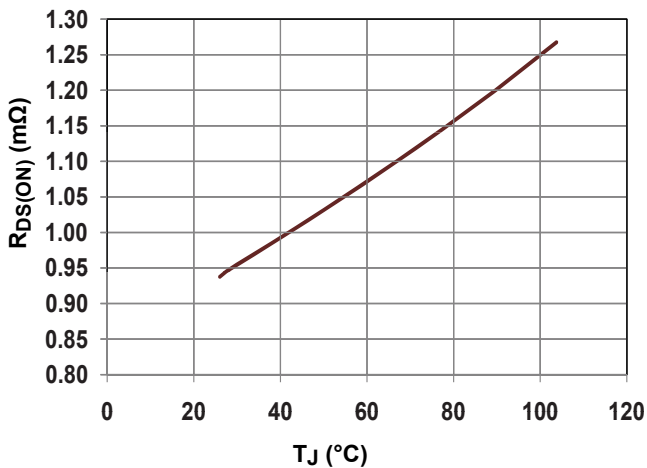


**Time/Div: 1s**

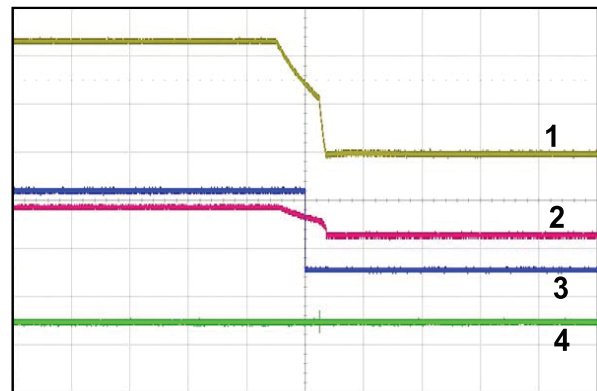
Conditions: I<sub>OUT</sub> = 0A  
FAULT and PWRGD Pullup Voltages Derived From 12V Supply  
1. V<sub>OUT</sub> (5V/div)  
2. V<sub>IN</sub> (5V/div)  
3. PWRGD (2V/div)  
4. FAULT (2V/div)

## Typical Operating Characteristics (continued)

**ON Resistance Including Package Resistance**

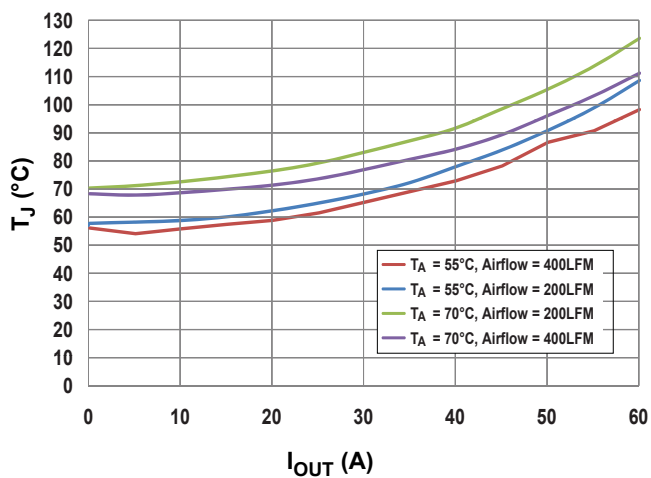


**Shutdown Controlled by  $V_{IN}$ , 1Ω Resistive Load**

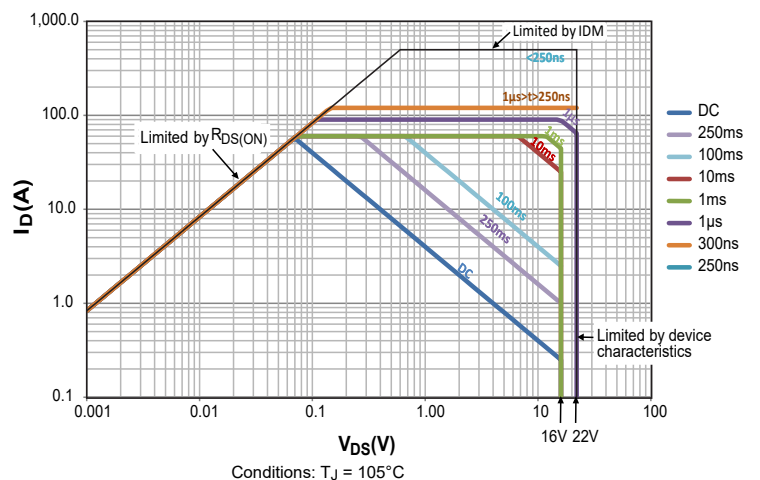


Conditions:  $R_{LOAD} = 1\Omega$   
1.  $V_{OUT}$  (5V/div)  
2.  $I_{OUT}$  (20A/div)  
3.  $PWRGD$  (2V/div)  
4.  $FAULT$  (2V/div)

**Thermal Evaluation**



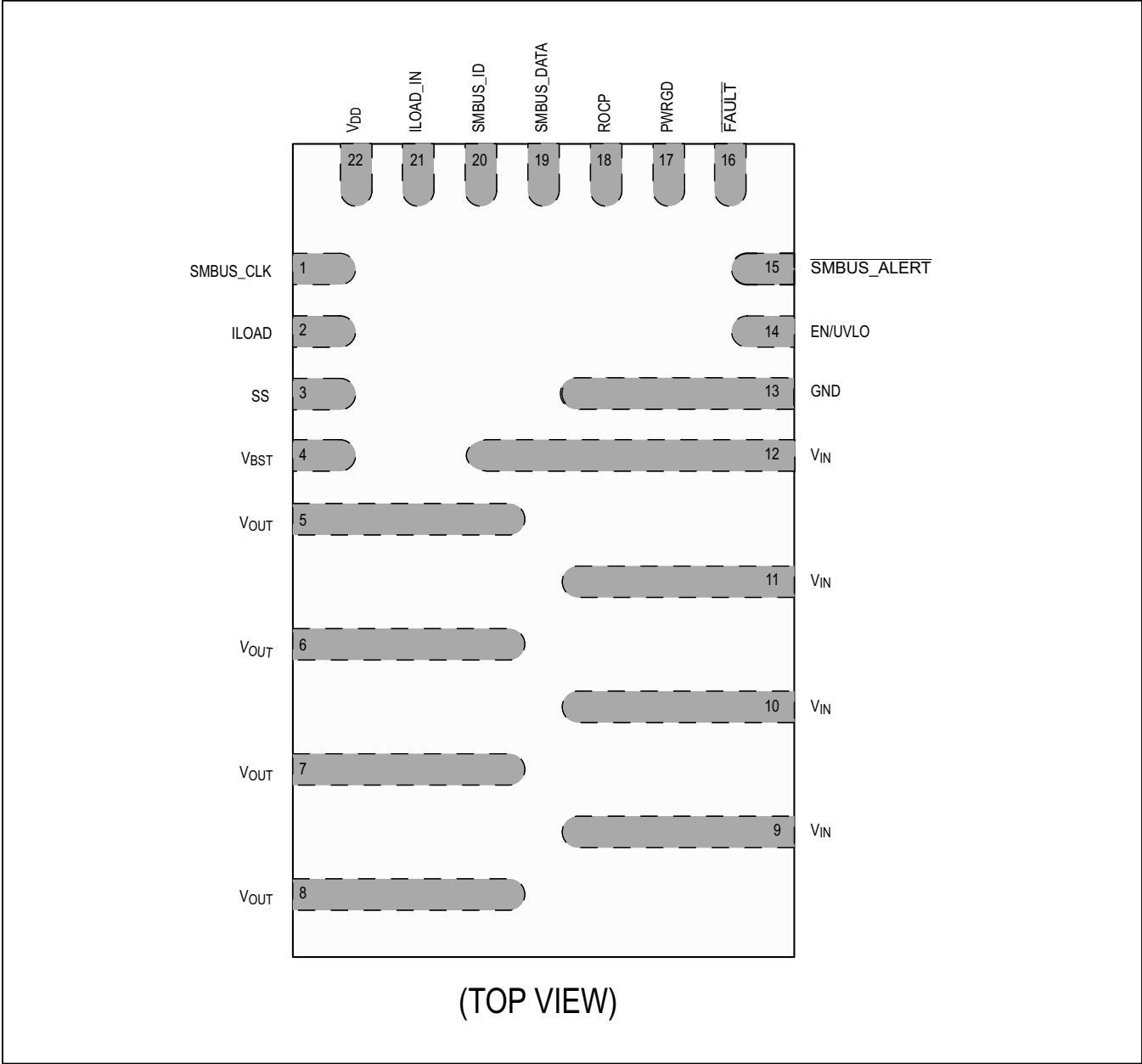
**Single-Pulse Operating Area (SOA)**



**MAX16545B**  
**MAX16545C**

Integrated Protection IC on 12V Bus with an Integrated MOSFET, Lossless Current Sensing, and PMBus Interface

**Pin Configuration**



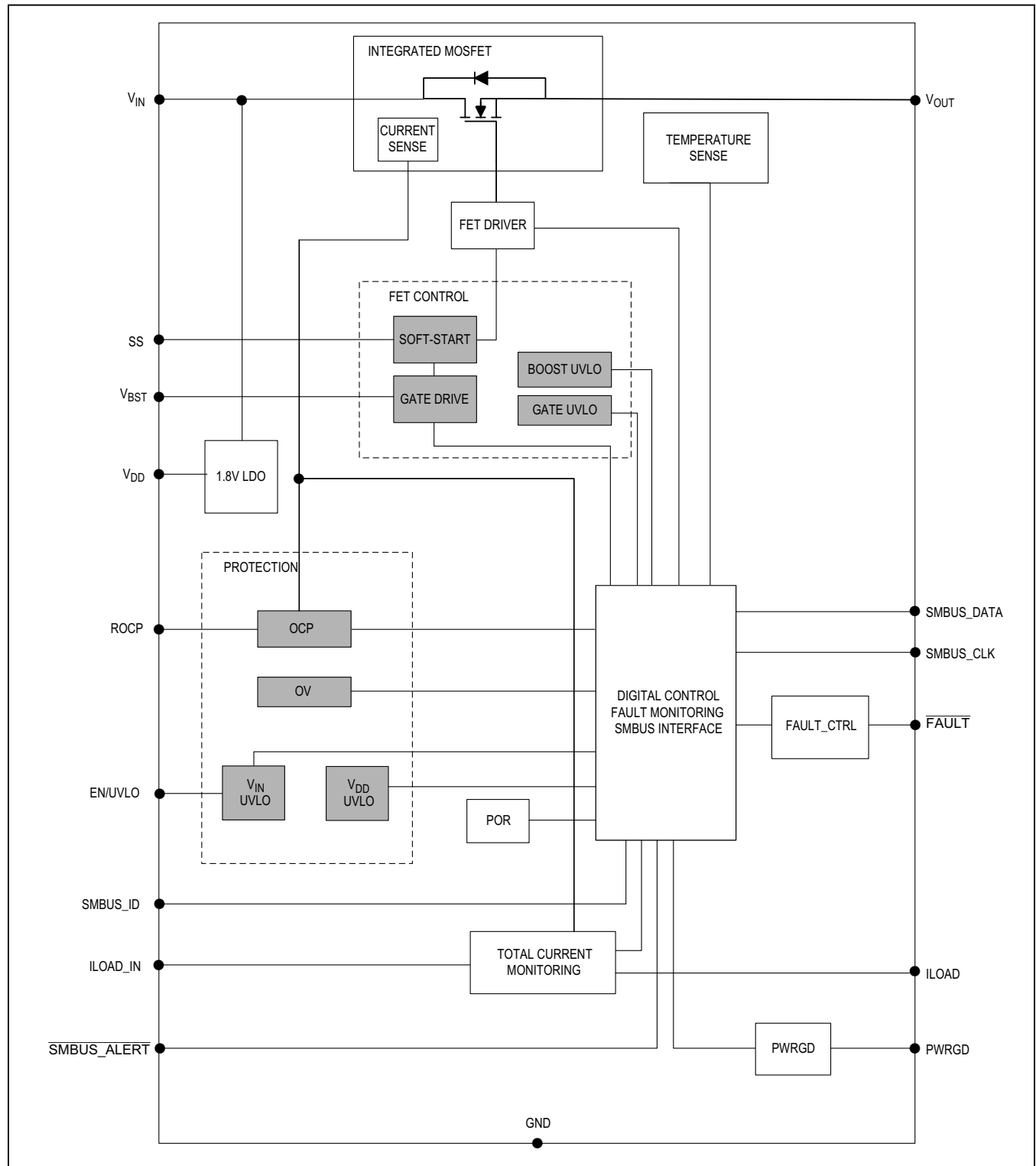


## Pin Description

PIN	NAME	FUNCTION
1	SMBUS_CLK	SMBus Clock Node.
2	ILOAD	Analog Current Representation of the Load Current. Connect this pin to ground through a properly sized resistor for proper voltage representation. Always keep this pin connected to ground through a resistor. A capacitor parallel to the $R_{ILOAD}$ resistor is required (see Figure 6 for the recommended value).
3	SS	Soft-Start Node. A capacitor is connected from SS to GND to program soft-start. The soft-start program capacitor should not exceed 75nF. If a fast-load transient ( $di/dt > 2.5A/ms$ ) resulting in fast and large transient output voltage deviation is possible in the application, an additional capacitor between SS and $V_{OUT}$ is recommended to keep the pass FET $V_{GS}$ above its UVLO threshold.
4	$V_{BST}$	Charge-Pump Supply for Pass FET Gate Drive. Connect this node to $V_{OUT}$ through a 220nF bypass capacitor. This supply is designed to be used by the MAX16545B/C and the MAX16543 only. No additional load or external components other than a bypass capacitor are allowed on the BST pin.
5-8	$V_{OUT}$	12V Output Power—Load Side.
9-12	$V_{IN}$	12V Input Power—Power Supply Side.
13	GND	Ground. Connect this node to GND plane through vias for proper operation.
14	EN/UVLO	12V Input Voltage UVLO pin. Normally connected to the center node of a resistor-divider connected from $V_{IN}$ to ground. A properly sized capacitor can be placed in parallel to the bottom resistor of the resistor-divider for additional filtering. In addition to 12V UVLO programming, this node can be used to enable/disable the MAX16545B/C.
15	$\overline{\text{SMBUS\_ALERT}}$	SMBus Alert. Open-drain, active-low pin.
16	$\overline{\text{FAULT}}$	Fault Communication (Bidirectional) Pin. This pin is used to indicate/receive detection of latching fault. See the <i>FAULT Reporting</i> section for more details. Connect this pin to the system bias supply rail through a 10k $\Omega$ resistor. 5V compliant, active-low pin.
17	PWRGD	Report $V_{IN}$ and $V_{OUT}$ status. See the <i>Power Good (PWRGD) Output</i> section for more details. Connect this pin to system bias supply rail through a 10k $\Omega$ resistor. 5V compliant.
18	ROCP	Moderate OCP Threshold Programming Input. Connect this pin to GND using an appropriate programming resistor. See the <i>Moderate OCP Threshold</i> section for additional details. No other components are allowed on this pin.
19	SMBUS_DATA	SMBus Data Node.
20	SMBUS_ID	SMBus Address Programming and Current Hysteresis Flag. Connect this pin to GND through a properly sized resistor to select desired address setting.
21	ILOAD_IN	System Current Reporting Input Pin. If current reporting from a single MAX16545B/C device is desired, connect this pin to GND through a 10k $\Omega$ resistor. Use this pin to sum currents from up to two MAX16543 devices if used in higher-current applications. In this case, the connection from the ILOAD pin of MAX16543 to ILOAD_IN of MAX16545B/C should be made through a 500 $\Omega$ resistor. Refer to the <i>MAX16543 Follower Device</i> section for details.
22	$V_{DD}$	Internal 1.8V Linear Regulator (LDO) Output. Connect this pin to GND through a 1 $\mu$ F (or higher) capacitor. See Figure 6 for the correct value. No additional loads or components other than up to two MAX16543 devices and external properly sized capacitor are allowed on the $V_{DD}$ pin.



## Block Diagram



## Theory of Operation

The MAX16545B/C integrated circuit-breaker IC is designed to provide a complete single-chip circuit-breaker protection solution for the 12V power bus, in applications where a power supply is either permanently connected (no removable assemblies), or is connected and disconnected at the input side only. It combines power monitoring and control functions with a low on-resistance pass FET device that acts as a disconnect switch to limit maximum power distributed to the load. The IC implements proprietary integrated lossless current sensing techniques to provide a highly accurate and compact circuit-breaker protection solution. The device integrates PMBus/SMBus interface for digital control and monitoring.

An integrated N-channel power MOSFET is driven by the FET control circuit that includes a boost circuit to provide a gate drive. An internal LDO provides  $V_{DD}$  bias supply, enabling effective use of the device in input hot-swap applications. See the [Block Diagram](#).

### Startup

The MAX16545B/C enables the integrated 1.8V  $V_{DD}$  LDO once the 12V supply voltage is high enough to guarantee LDO operation. Once  $V_{DD}$  is valid, the device reads the SMBus\_ID programming resistor value to set the SMBus address, and initiates itself. During this period, the gate drive supply capacitor becomes fully charged. Once these two functions are complete, the device can be controlled through the enable input.

The enable input (EN/UVLO) has a precise threshold and provides a  $V_{IN\_UVLO}$  function, where the enable voltage is derived from  $V_{IN}$  through a resistor-divider, with an optional control signal used in conjunction with the divider as shown in [Figure 1](#) and [Figure 5](#). The EN/UVLO pin must not be pulled high externally other than to pull the EN/UVLO pin high through a properly sized resistor-divider from the input supply. Though the enable signal can be pulled low to disable the part, this should only be allowed to occur when the input voltage supply is within operating range specified in [Electrical Characteristics](#) table.

### Safe Operating Area (SOA)

During soft-start, it is important to keep the FET within its safe operating area. The peak current allowed during startup is shown in [Figure 2](#). In case resistive short is possible in the application, it is required to use startup OCP feature to protect the device. The MAX16545B has a default startup OCP of 16A, and the MAX16545C has a default startup OCP of 24A. The startup OCP threshold must be programmed to a value lower than safe peak current. [Figure 2](#) assumes pure capacitive or RC load on the output.

During normal operation,  $V_{GS\_UVLO}$  ensures FET is operating in deep triode region while three levels of OCP limit the device operating current within safe boundaries. Refer to the SOA curves under [Typical Operating Characteristics](#) for more detail.

### Self-Check

Once  $V_{BST\_UVLO}$  has been cleared, the internal  $V_{DD\_LDO}$  is fully enabled and the EN/UVLO pin is above the enable threshold, the MAX16545B/C initiates a timed output discharge as a self-check procedure. By default, the device initiates this procedure immediately after  $V_{BST\_UVLO}$  has been cleared and the internal  $V_{DD\_LDO}$  is fully enabled. If the output does not fall below the programmed self-check threshold after a fixed period, the pass FET could be potentially shorted and the IC asserts both FAULT and PWRGD pins low. The latched fault condition remains until restart (EN/UVLO,  $V_{IN}$  toggling or restarting through the OPERATION command). The threshold for the self-check procedure can be programmed to one of four different values through the SMBus. See the Config Register (D0h) for more details. If the output voltage is below the programmed self-check threshold, the self-check is considered passed and the device proceeds with startup.

Because the device actively discharges the output when disabled, it should not be used in applications where the output can remain powered through another path.

In addition a soft-start capacitor discharge procedure is performed during every restart. The MAX16545B/C utilizes an integrated resistive element, with resistance approximately 1k $\Omega$ , to discharge the  $C_{SS}$  capacitor. The device checks if the voltage across the  $C_{SS}$  capacitor is below the soft-start threshold shown in the [Electrical Characteristics](#) table after a fixed period of time. If the voltage across the  $C_{SS}$  capacitor is not below the soft-start threshold, the device latches the pass FET OFF and asserts the  $\overline{\text{FAULT}}$  pin low.

The MAX16545B/C checks the  $R_{OCP}$  value at all times, including startup (after the bias supply is valid). This check ensures proper moderate OCP threshold selection. If the wrong  $R_{OCP}$  value is detected, the device reports the fault by asserting the  $\overline{\text{FAULT}}$  signal low. Each latched fault condition is kept until restart (EN/UVLO,  $V_{IN}$  toggling or restarting through the OPERATION command).

### Soft-Start

Once the self-check procedure is complete, the output voltage soft-start ramp is initiated. During startup, the  $C_{SS}$  capacitor is charged using a constant current source. Since the integrated FET is configured as a source follower, the output voltage is ramped monotonically at a rate determined by the external soft-start capacitor. The soft-start capacitor should not exceed 75nF.

**MAX16545B**  
**MAX16545C**

Integrated Protection IC on 12V Bus with an Integrated  
MOSFET, Lossless Current Sensing, and PMBus Interface

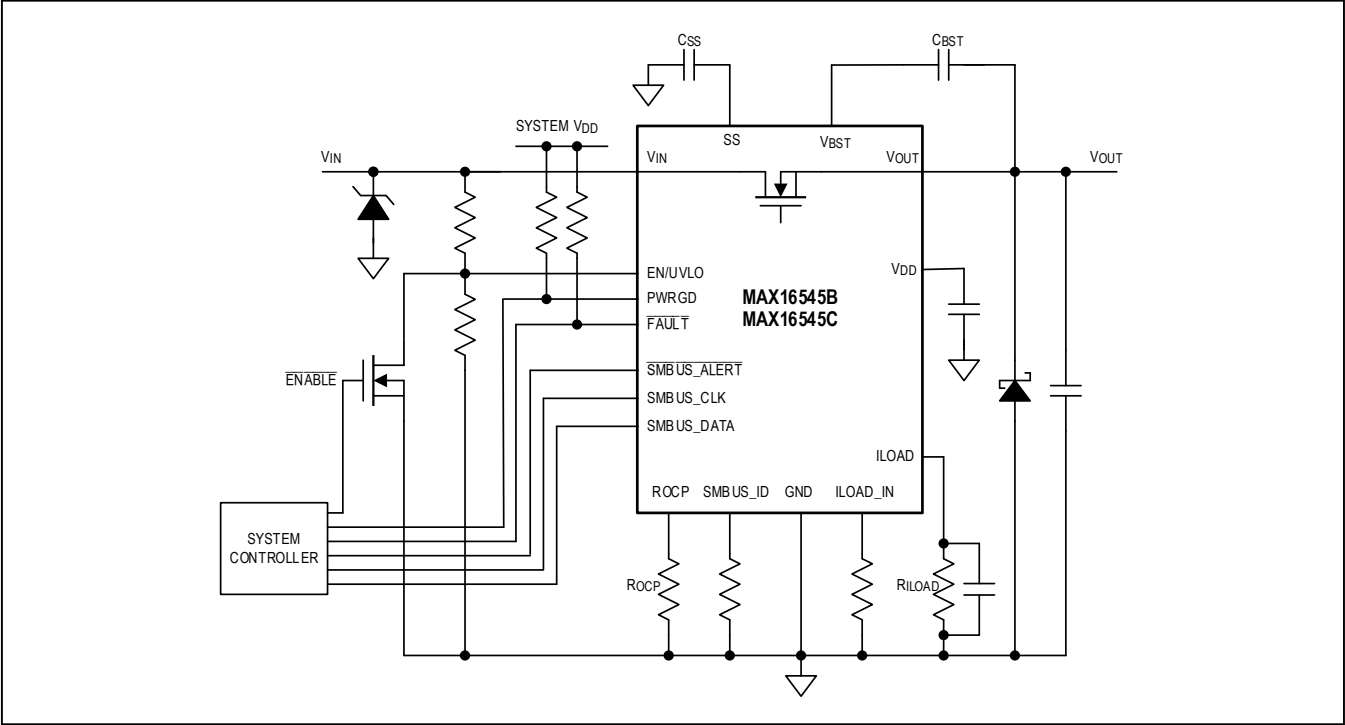


Figure 1. Configuration Example

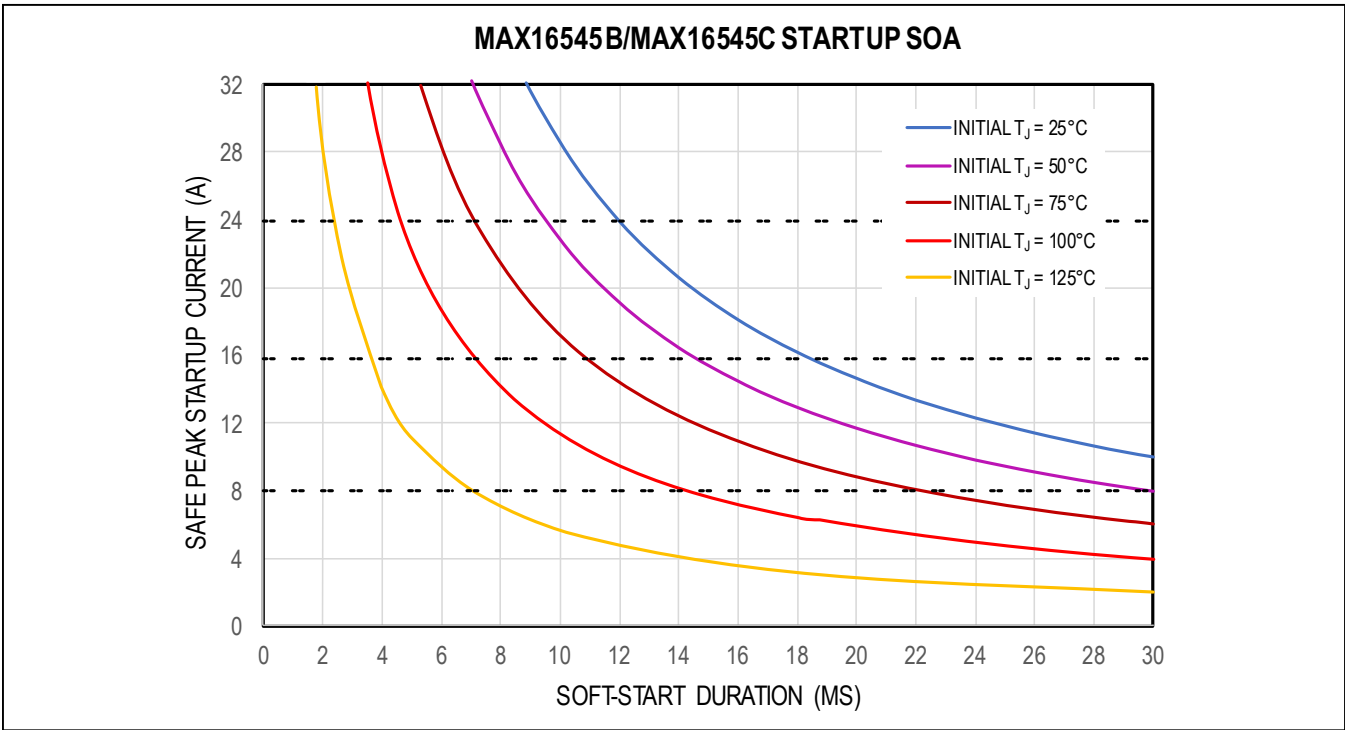


Figure 2. Startup SOA

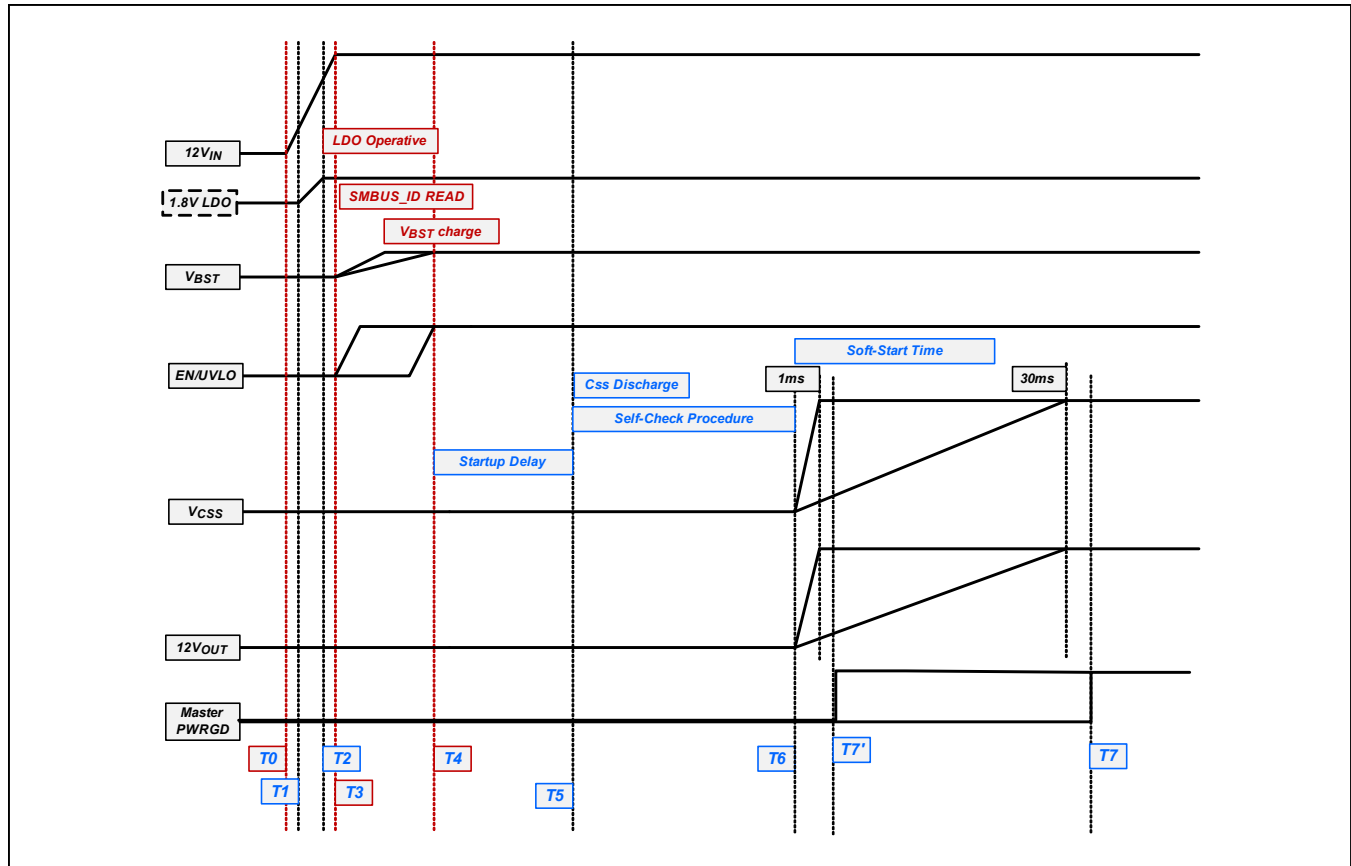


Figure 3. Startup Timing Diagram

### V<sub>DD</sub> Undervoltage Lockout

The MAX16545B/C implements V<sub>DD</sub>\_UVLO fault monitor and protection. V<sub>DD</sub> is monitored at all times. Startup procedures are not initiated until V<sub>DD</sub>\_UVLO is cleared. If V<sub>DD</sub> falls below V<sub>DD</sub>\_UVLO during normal operation, the part shuts down. Once V<sub>DD</sub> returns to acceptable values, the part performs a turn-on procedure from soft-start delay phase.

### Current Hysteresis

SMBUS\_ID pin is used as current hysteresis flag after R<sub>SMBUS\_ID</sub> decoding is done. If output current is greater than MFR\_SPECIFIC\_HYSTH level, the current hysteresis flag is set to high. If output current is less than MFR\_SPECIFIC\_HYSTL level, then the current hysteresis flag is set to low.

### SMBus\_ID Resistor Out of Range

After V<sub>DD</sub>\_UVLO is cleared, the SMBus\_ID programming resistor is read. If it is out of range, a latching fault is tripped. This fault can be cleared by EN/UVLO, V<sub>IN</sub>

toggling, or restarting through the OPERATIONS command. The EN/UVLO toggling may mean pulling EN/UVLO above the threshold and then low after the fault is detected to reset the fault latch. The value is read again when EN/UVLO goes high for the second time.

### V<sub>BST</sub> and V<sub>GS</sub> UVLO

V<sub>BST</sub>\_UVLO is checked when EN/UVLO is above the enable threshold and remains active while the device is enabled. The V<sub>BST</sub>\_UVLO can also be tripped before the FET is turned on, the latching fault can still be reset by EN/UVLO or V<sub>IN</sub> toggling or restarting through the OPERATION command. The gate drive (V<sub>GS</sub>) UVLO protection has 10μs internal deglitching filter to prevent false tripping due to output voltage overshoot if a large, fast-unloading transient is present. If additional filtering is desired, an external capacitor can be placed between V<sub>OUT</sub> and V<sub>SS</sub> pins. V<sub>BST</sub> and V<sub>GS</sub> UVLO are latching faults, which result in the pass FET being latched off and FAULT asserted low. V<sub>GS</sub>\_UVLO is masked for 100ms during startup to avoid tripping a latching fault.

## V<sub>IN</sub>\_UVLO

The V<sub>IN</sub> is sensed at the EN/UVLO pin through a resistor-divider that is used to program the desired V<sub>IN</sub>\_UVLO threshold. The V<sub>IN</sub>\_UVLO circuit has hysteresis, and if the EN/UVLO pin voltage falls below the programmed value during operation, the device turns the integrated FET off and PWRGD is deasserted. The integrated FET turns on again if the positive threshold is exceeded and performs self-test and soft-start. The MAX16545B/C supports V<sub>IN</sub>\_UVLO warning and fault reporting through PMBus.

## Overvoltage Detection

The input voltage is constantly monitored, and if at any time it rises above PMBus programmed threshold, the Power-Good output is de-asserted.

## Overcurrent Detection

The IC actively monitors load current on a 12V power bus at all times, including startup. Startup overcurrent protection (OCP) is active during startup. The startup OCP threshold can be changed to one of three programmable options using PMBus. Note that changing the startup OCP to a higher value can violate the startup SOA, in such cases, t<sub>SS</sub> needs to be lowered to remain in SOA (Figure 2). If, at any time during startup, the load current exceeds the programmed startup OCP threshold, the integrated pass FET is turned off within 10μs and the FAULT pin is asserted low. This is a latching fault. The device provides three levels of overcurrent protection during normal operation (after the startup procedure is complete and the PWRGD flag is set high). The moderate OCP threshold is set using an external resistor and is therefore programmable over a wide continuous range. Moderate OCP threshold can be exceeded for a limited, programmable, timeout period without resulting in a fault condition. If the load current on the 12V power bus exceeds the moderate OCP threshold for the entire timeout period, but its magnitude is less than a programmable severe overcurrent threshold (Table 3), the integrated pass FET is latched off at the end of the timeout period and the FAULT pin will be asserted low to indicate a moderate OCP fault condition. Restarting the system requires EN/UVLO or V<sub>IN</sub> toggling or restarting through the OPERATION command. Moderate OCP timeout is user programmable through PMBus with four different values.

Severe OCP is the second level of OCP. Severe OCP threshold is programmed relative to moderate OCP threshold. Fixed 10μs timeout is supported for the severe OCP. If at any time the load current exceeds the severe OCP threshold and the timeout is disabled, the device turns the integrated pass FET OFF and asserts the FAULT low. If timeout is enabled, the MAX16545B/C tolerates current exceeding the severe OCP threshold for the duration of the timeout. If the current is still higher than the severe OCP threshold by the end of the timeout, the device turns off and asserts FAULT low. Severe overcurrent protection is a latching fault.

The device features a third level of protection against severe overload faults, called safe OCP with internally-fixed threshold. If at any time the load current on a 12V power bus exceeds the safe OCP threshold, the device turns the pass FET OFF within 250ns and asserts the FAULT low. The severe OCP threshold should be set to a value less than the safe OCP threshold.

**Table 1. Timing Diagram Example Shown in Figure 6**

POINT/PERIOD	DESCRIPTION
t0	12V V <sub>IN</sub> applied.
t1	The LDO has enough headroom to start up.
t2	The internal 1.8V LDO is ready and MAX16545B/C logic is fully operational. External bypass capacitors connected to the V <sub>DD</sub> pin is fully charged.
t3-t14	SMBus ID resistor reading and V <sub>BST</sub> charging. EN/UVLO signal status is ignored from T3 to T4.
t4	EN/UVLO is active. Startup check-up procedures can be started after the EN/UVLO is cleared.
t4-t5	User-programmable startup delay.
t5-t6	Self-test procedure. C <sub>SS</sub> discharge.
t6-t7	V <sub>OUT</sub> ramp
t7	Power good (PWRGD) output asserted.
t7+	PWRGD is high, normal operation starts.

The IC supports on-the-fly moderate OCP threshold changes. The concept is shown in [Figure 4](#). If the resistance seen by the ROCP pin is changed during operation, the device adjusts the moderate OCP threshold matching the value selected by external resistive network.

### Wrong ROCP Protection

The MAX16545B/C is protected against out of range ROCP values. Protection is enabled at any time, including startup. If the ROCP resistor value is detected out of the permitted range, the device latches the integrated FET OFF, and it asserts the FAULT pin low. A EN/UVLO or VIN toggling or a restart through the OPERATION command must be performed in order to clear the fault and restart the device. If ROCP is changing during operation by the system ([Figure 4](#)), the design must be such that ROCP is within range at all times.

**Note:** If the switch that changes OCP threshold is slower to transition, the effective resistance on the ROCP pin can cause severe OCP to trigger during on-the-fly transitions.

### Overtemperature Protection (OTP)

The IC includes protection against overtemperature conditions (OTP). If the junction temperature exceeds the programmable fault threshold, the IC latches the integrated FET OFF and asserts the FAULT output low. To reenale the IC, the following options are available:

- Toggle EN/UVLO or VIN
  - Restart through the OPERATION command
- Overtemperature fault and warning thresholds are user-programmable through the PMBus. Exceeding the overtemperature warning threshold does not latch the FET off.

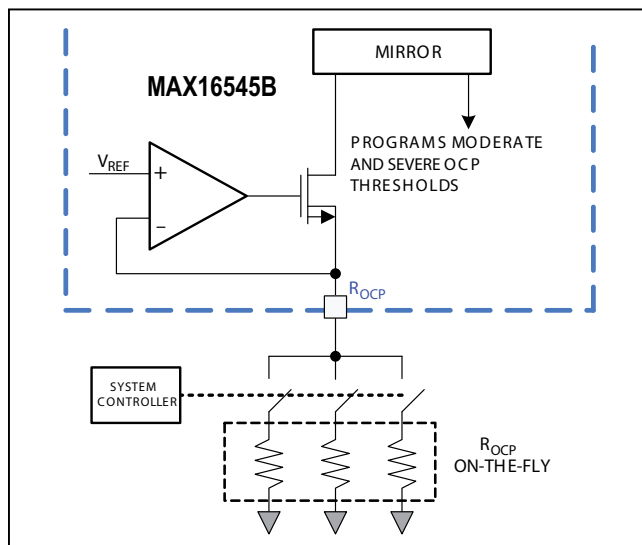


Figure 4. On-the-Fly Analog Programmable Moderate OCP Threshold

### FAULT Reporting

The device provides a dedicated pin for fault reporting, FAULT. If at any time a latching fault is detected, the FAULT pin is immediately latched low. EN/UVLO or VIN toggling, or a restart through the OPERATION command is required to reset the device after latching fault detection.

### Fault Input and FAULT Pullup

The FAULT pin is a bidirectional open-drain pin that can be used for fault communication, from an external circuitry to the MAX16545B/C. To communicate a fault to the device, the external circuitry must pull the FAULT pin low. If FAULT is externally pulled low, the device treats it as a latching fault. Therefore, the pullup voltage source must be considered to ensure the rail is operational and pulled high before the device startup cycle.

These are the options for the pullup rail:

- VDD: 1.8V internal LDO rail. This rail is limited to 1.8V, thus the external system has to be compliant with it.
- System 3.3V or 5V rail. FAULT is 5V compliant, and therefore, an external higher voltage rail can be used for pullup. This rail has to be stable when the device is initiating its startup procedure. If it is not, a false fault communication can occur.

### Power Good (PWRGD) Output

The MAX16545B/C provides a dedicated pin for power good reporting (PWRGD). PWRGD is asserted high after startup, when the output voltage exceeds the programmed PWRGD threshold and the FET is fully enhanced and operating in its resistive region. In all the other conditions, the PWRGD pin is deasserted low.

PWRGD is an open-drain pin, thus an external pull-up resistor connected to pullup supply rail is needed. Different options are available for the pullup rail:

- VDD: 1.8V internal LDO rail. This rail is limited to 1.8V, thus the external system has to be compliant with it.
- System 3.3V or 5V rail. PWRGD is 5V compliant, and therefore an external higher voltage rail can be used for pullup.

### Analog Load Current Signal Output

The IC includes an output pin, ILOAD to allow the user to monitor the load current device. The current sourced by the ILOAD pin is proportional to the current through the device with the ratio shown in the [Electrical Characteristics](#) table. A properly sized resistor between the ILOAD pin and GND should be added. A capacitor parallel to this resistor is required for proper operation. The MAX16545B/C reports zero current on the ILOAD pin



**Table 2. Faults Detected and Actions**

PARAMETER	DESCRIPTION	FAULT ASSERTED	LATCHING
V <sub>DD_UVLO</sub>	Internal V <sub>DD</sub> LDO UVLO	No	No
V <sub>BST_UVLO</sub>	UVLO for V <sub>BST</sub>	Yes (Note 1)	Yes
V <sub>GS_UVLO</sub>	UVLO for V <sub>GS</sub>	Yes	Yes
V <sub>IN_UVLO</sub>	EN/UVLO pin below UVLO threshold	N/A (Note 2)	N/A (Note 2)
V <sub>IN_UV</sub>	V <sub>IN</sub> below PMBus programmed threshold	Yes	Yes
ROCP_FAULT	R <sub>OCP</sub> detected out of valid range	Yes	Yes
R_SMBus_ID FAULT	SMBus_ID resistor detected out of range	Yes	Yes
CSS Discharge FAULT	Soft-start capacitor discharge failed	Yes	Yes
FET Short	Pass FET short detected during startup self-check	Yes	Yes
Startup OCP	Startup overcurrent fault detected	Yes	Yes
Moderate OCP	Moderate overcurrent fault detected	Yes	Yes
Severe OCP	Severe overcurrent fault detected	Yes	Yes
Safe OCP	Safe overcurrent fault detected	Yes	Yes
OTP	Overtemperature fault threshold exceeded	Yes	Yes
V <sub>IN_OV</sub>	V <sub>IN</sub> overvoltage detected	No	No

**Note 1:** If V<sub>BST\_UVLO</sub> fault occurs before startup, FAULT is not reported and status registers are not updated.

**Note 2:** V<sub>IN\_UVLO</sub> clears FAULT and the latching event.

during soft-start, it starts reporting load current 5ms after the pass FET V<sub>GS</sub> is above its UVLO threshold.

If another device reports the current to the MAX16545B/C through the ILOAD\_IN pin, the MAX16545B/C reports the total current on the ILOAD pin. When choosing a load resistor to provide a current reporting voltage, the total system current must be considered.

### MAX16543 Follower Device

Up to two MAX16543 protection ICs can be used in parallel with the MAX16545B/C to increase total current capability. Specifically,

- MAX16545B/C + 1 × MAX16543, 90A current capability
- MAX16545B/C + 2 × MAX16543, 120A current capability

In these configurations, connect the ILOAD pins of MAX16543 to the ILOAD\_IN pin of MAX16545B/C through 500Ω resistors for total current reporting. Refer to the MAX16543 datasheet for more details.

### PMBus/SMBus Reporting and Warning

The MAX16545B/C provides PMBus-compliant digital telemetry through the SMBus as shown in [Table 4](#). The IC supports input power and input energy reporting. For average values the sample size is programmable. Peak values for input/output voltage, output current, input

**Table 3. Requirements for PWRGD Assertion**

PARAMETER	CONDITION
V <sub>OUT</sub>	V <sub>OUT</sub> > PWRGD threshold
V <sub>IN</sub>	V <sub>IN</sub> > V <sub>IN_UVLO</sub>
Self-Check	MOSFET (V <sub>IN</sub> to V <sub>OUT</sub> ) short not detected during startup
V <sub>GS</sub>	FET is on and in triode region. There's a minimum delay of 5ms before PWRGD assertion after this condition is met.

power and temperature are stored in dedicated manufacturer-specific registers. The device also provides warning functions based on programmable warning thresholds, as per the PMBus specification. The sample size for warning flags is also programmable.

The device supports fault and status reporting except for V<sub>BST\_UVLO</sub> fault at startup. If the V<sub>BST\_UVLO</sub> occurs at startup the device latches off, FAULT is not reported, and status registers are not updated.

## Configuration

The MAX16545B/C is configured using both analog programming resistors and PMBus. See [Table 5](#) and [Table 6](#) for programmable parameters.

### Moderate OCP Threshold

The moderate overcurrent protection (OCP) threshold is externally programmable through a resistor connected to the ROCP pin. The moderate OCP threshold is programmed as shown in Equation 1 and Equation 2. See the [Electrical Characteristics](#) table for V<sub>OCPM</sub> and GOCP.

#### Equation 1:

$$I_{OCP} = \frac{V_{OCPM}}{R_{OCP}} \times GOCP$$

where:

I<sub>OCP</sub> = moderate overcurrent protection threshold (A)

V<sub>OCPM</sub> = overcurrent protection reference voltage shown in [Electrical Characteristics](#) table (V)

GOCP = overcurrent protection gain shown in [Electrical Characteristics](#) table (A/A)

R<sub>OCP</sub> = Value of overcurrent protection programming resistor (Ω)

#### Equation 2:

$$R_{OCP} = \frac{V_{OCPM}}{I_{OCP}} \times GOCP$$

where:

I<sub>OCP</sub> = moderate overcurrent protection threshold (A)

V<sub>OCPM</sub> = overcurrent protection reference voltage shown in [Electrical Characteristics](#) table (V)

GOCP = overcurrent protection gain shown in [Electrical Characteristics](#) table (A/A)

R<sub>OCP</sub> = Value of overcurrent protection programming resistor (Ω)

### Design Example

To set moderate OCP to 68.7A nominal, using Equation 2:

#### Equation 3:

$$R_{OCP} = \frac{0.8V}{68.7A} \times (8 \times 10^6) \\ = 93.1k\Omega$$

## Table 4. PMBus/SMBus Reporting

PARAMETER	DESCRIPTION
V <sub>IN</sub>	Input voltage (PMBus compliant)
V <sub>OUT</sub>	Output voltage (PMBus compliant)
I <sub>OUT</sub>	Output current (PMBus compliant)
P <sub>IN</sub>	Input power (PMBus compliant)
E <sub>IN</sub>	Input energy (PMBus compliant)
Temperature	MAX16545B/C chip temperature (PMBus compliant)
Peak V <sub>IN</sub>	Peak value of input voltage (direct reading)
Peak V <sub>OUT</sub>	Peak value of output voltage (direct reading)
Peak I <sub>OUT</sub>	Peak value of output current (direct reading)
Peak P <sub>IN</sub>	Peak value of input power (direct reading)
Peak Temperature	Peak value of IC temperature (direct reading)



### Soft-Start Capacitor $C_{SS}$

During startup, the pass FET device is operated as a source follower. The soft-start capacitor,  $C_{SS}$  is connected between the MOSFET's gate and ground and is charged from a fixed current source. The external  $C_{SS}$  capacitor therefore charges linearly and this produces a linear monotonic ramp for  $V_{OUT}$ . The ramp rate is programmable by selecting the appropriate value for  $C_{SS}$ .

The ramp rate for the voltage across  $C_{SS}$  and, hence,  $V_{OUT}$  is given by Equation 4.

**Equation 4:**

$$\frac{dV}{dt} = \frac{I_{SS}}{C_{SS}}$$

where:

$dV/dt$  = voltage ramp rate of  $V_{OUT}$  (V/ $\mu$ s)

$I_{SS}$  = MAX16545B/C soft-start current source (mA)

$C_{SS}$  = external  $C_{SS}$  capacitor value (nF)

Assuming  $V_{IN} = 12V$ , Equation 4 can be used to derive the soft-start time (Equation 5).

**Equation 5:**

$$t_{SS} = \frac{12V \times C_{SS}}{I_{SS}}$$

where:

$t_{SS}$  = ramp duration ( $\mu$ s)

$I_{SS}$  = MAX16545B/C soft-start current (mA)

$C_{SS}$  = external soft-start capacitor value (nF)

**Table 5. MAX16545B/C Programmability**

PARAMETER	PROGRAMMABILITY TYPE	COMPONENT	I/O
Moderate OCP Threshold	Analog	ROCP	ROCP
Softstart Ramp Rate	Analog	CSS	SS
Input UVLO Threshold	Analog	Divider	EN/UVLO
SMBus Address	Programming Resistor	RSMBus_ID	SMBus_ID
Moderate OCP Timeout	Digital	SMBus	SMB_DATA
Startup OCP Threshold	Digital	SMBus	SMB_DATA
Severe OCP Threshold	Digital	SMBus	SMB_DATA
Startup Delay	Digital	SMBus	SMB_DATA
Input OV Threshold	Digital	SMBus	SMB_DATA
Output PWRGD Threshold (Sets $V_{OUT\_UVLO}$ and Self-Test Thresholds)	Digital	SMBus	SMB_DATA
Self-Check Threshold	Digital	SMBus	SMB_DATA
Overtemperature Warning and Fault Thresholds	Digital	SMBus	SMB_DATA
Input Overpower Warning Threshold	Digital	SMBus	SMB_DATA
Reporting and Warning Averaging Sample Size	Digital	SMBus	SMB_DATA
$V_{IN}$ Undervoltage Warning	Digital	SMBus	SMB_DATA
$V_{OUT}$ Undervoltage Warning	Digital	SMBus	SMB_DATA
$V_{IN}$ Undervoltage Fault	Digital	SMBus	SMB_DATA
Overcurrent Warning	Digital	SMBus	SMB_DATA
Overcurrent Hysteresis	Digital	SMBus	SMB_DATA

**Table 6. MAX16545B/C Reporting External Components**

PARAMETER	PROGRAMMABILITY TYPE	COMPONENT	I/O
ILOAD Voltage Gain	Analog	R <sub>ILOAD</sub>	ILOAD

Assuming the load inrush current is due to the output capacitance only, the load current is given by Equation 6 and Equation 7.

**Equation 6:**

$$I_{VIN} = I_{OUT} = C_{OUT} \times \frac{dV}{dt}$$

where:

I<sub>OUT</sub> = load inrush current(A)

C<sub>OUT</sub> = load capacitance (μF)

**Equation 7:**

$$I_{VIN} = I_{OUT} = \frac{C_{OUT} \times I_{SS}}{C_{SS}}$$

where:

I<sub>OUT</sub> = load inrush current(A)

C<sub>OUT</sub> = load capacitance (μF)

C<sub>SS</sub> = external soft-start capacitor value (nF)

I<sub>SS</sub> = MAX16545B/C soft-start current (mA)

Therefore, the soft-start capacitor can be selected based on the design value of inrush current using Equation 8.

**Equation 8:**

$$C_{SS} = \frac{C_{OUT} \times I_{SS}}{I_{INRUSH}}$$

where:

I<sub>INRUSH</sub> = desired maximum inrush current due to C<sub>LOAD</sub> (A)

C<sub>OUT</sub> = load capacitance (mF)

C<sub>SS</sub> = external soft-start capacitor value (nF)

I<sub>SS</sub> = MAX16545B/C soft-start current (μA)

The max value of t<sub>SS</sub> is 30ms to guarantee linear startup as specified in the [Electrical Characteristics](#) table, select C<sub>SS</sub> to meet this requirement.

Note that unloading transients with di/dt > 2.5A/ms can be large enough to cause output voltage transients > ~100mV. Under these conditions, an additional capacitor between SS and V<sub>OUT</sub> is recommended to ensure the pass FET V<sub>GS</sub> remains above its UVLO threshold. This capacitor has no noticeable effect on soft-start ramp time as the differential voltage from V<sub>OUT</sub> to SS remains approximately constant during soft-start.

#### Design Example

Assume a maximum design value for inrush current of 10A, and a load capacitance of 2mF.

$$C_{SS} = \frac{2mF \times 30\mu A}{10A}$$

$$= 6nF \text{ minimum (to meet inrush maximum)}$$

Use Equation 5 to show that corresponding t<sub>SS</sub> in this example is 2.4ms.

## Input UVLO

The input UVLO is set using a resistor-divider, as shown in [Figure 5](#). The enable threshold,  $V_{IN\_UVLO}$  is given in the [Electrical Characteristics](#) table. The corresponding value for the  $V_{IN}$  rail is given by Equation 9.

**Equation 9:**

$$V_{IN} = \frac{V_{IN\_UVLO}}{K}$$

or

**Equation 10:**

$$K = \frac{V_{IN\_UVLO}}{V_{IN}}$$

where:

$V_{IN}$  = 12V rail input voltage to enable device (V)

$V_{IN\_UVLO}$  = EN/UVLO threshold (V)

$K$  = resistor-divider ratio,  $R_2/(R_1 + R_2)$

## Design Example

To set input UVLO to 10.8V. Using Equation 10.

**Equation 11:**

$$\begin{aligned} K &= \frac{1.0V}{10.8V} \\ &= 0.0926 \end{aligned}$$

where:

$R_2$  is set to 2.26k $\Omega$

$R_1$  = 20k $\Omega$  to guarantee startup at min supply voltage

## SMBus Address Programming

The SMBus address is programmed to one of sixteen values using the external resistor as shown in [Table 7](#).

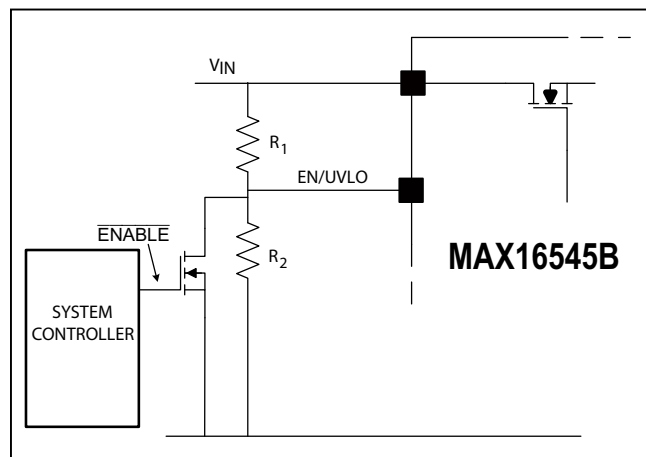


Figure 5. Programming Resistors for Input UVLO

## ILOAD Reporting

The current reporting voltage is set using an external resistor connected from ILOAD to ground. The maximum voltage is shown in the [Electrical Characteristics](#) table. The reporting voltage gain is given by Equation 12. Assuming a maximum signal voltage of 1.35V, Equation 13 can be used to select a value for  $R_{ILOAD}$  based on a desired full-scale current.

**Equation 12:**

$$V_{ILOAD} = R_{ILOAD} \times I_{LOAD} \times G_{ILOAD}$$

**Equation 13:**

$$R_{ILOAD} = \frac{1.35V}{I_{LOAD\_FSD} \times G_{ILOAD}}$$

where:

$V_{ILOAD}$  = current reporting voltage (V)

$R_{ILOAD}$  = external current reporting resistor ( $\Omega$ )

$I_{LOAD}$  = load current  $I_{OUT} = I_{VIN}$  (A)

$I_{LOAD\_FSD}$  = desired full scale of current reporting (A)

$G_{ILOAD}$  = current reporting gain from the [Electrical Characteristics](#) table

**Table 7. SMBus Address Programming Resistor Values**

	$R_{SMBus\_ID}$ (k $\Omega$ )	SMBus Address
0	1.78	40h
1	2.37	41h
2	3.16	42h
3	4.22	43h
4	5.62	44h
5	7.5	45h
6	9.76	46h
7	13	47h
8	17.4	10h
9	23.2	11h
10	30.9	12h
11	41.2	13h
12	54.9	50h
13	73.2	51h
14	97.6	52h
15	127	53h

**Note:**  $R_{SMBus\_ID}$  resistor tolerance should be 1% or better.

For example, to set the full-scale current reported to 67.5A.

$$R_{ILOAD} = \frac{1.35V}{67.5A \times (5 \times 10^{-6})}$$

$$= 4K\Omega$$

**Note:**  $R_{ILOAD}$  tolerance should be 0.1% or better.

### PMBus Reporting and Warning Settings

The MAX16545B/C provides single sample, averaged and peak values for voltage, current and power reporting, and single reading or averaged readings for warnings. The limits for these parameters and warnings should be set through PMBus prior to operation if a value other than the default is required.

Averaging for voltage, power, and current uses an accumulate and dump technique, whereas temperature uses a shifting window/rolling average. Power is calculated for every voltage and current sample (as opposed to using averaged voltage and current), and the result is accumulated in a dedicated register for averaging.

Two independently programmable averaging sample sizes are used for reporting and warning/status register use for voltage, current, and power.

- Reporting:  $2^N$  samples, with  $N = 0$  to 16 (1 sample to 32K samples)  
Sample size is set using Reg\_DDh (CONFIG\_2, manufacturer-specific register)
- Warning:  $2^M$  samples, with  $M = 0$  to 3 (1 sample to 8 samples)  
Sample size is set using Reg D0h (CONFIG, manufacturer-specific register)

### Setting SMBus Programmable Parameters

See [Table 8](#) for parameters that are programmed through the SMBus. If a setting other than the default shown below is required, it must be programmed through SMBus. Note that some parameters have an enable bit as well as value bits.

**Table 8. Parameters Programmed Through SMBus and Default Values**

PARAMETER	DEFAULT VALUE
Overtemperature Protection Threshold	135°C
V <sub>OUT</sub> PWRGD/Self Check Thresholds	11V/9V
Moderate OCP Timeout	100μs
Startup Delay	0μs
Severe OCP Threshold	130%
Input OV Threshold	16V
Overtemperature Warning Threshold	220°C (disabled by default)
Input Power Warning Threshold	(Disabled by default)
Overcurrent Warning	(Disabled by default)
V <sub>IN</sub> Undervoltage Warning	(Disabled by default)
V <sub>OUT</sub> Undervoltage Warning	(Disabled by default)
Current Hysteresis	(Disabled by default)
Startup OCP	MAX16545B = 16A MAX16545C = 24A

**Table 9. PMBus/SMBus Registers (Note 1)**

ADDRESS (HEX)	NAME	TYPE	BYTES	DEFAULT (HEX)
1	<b>OPERATION</b>	RW	1	80
3	<b>CLEAR_FAULTS</b>	WO	0	0
19	<b>CAPABILITY</b>	RO	1	B0
1B	<b>SMBALERT_MASK</b>	(Note 2)	2	0000
43	<b>VOUT_UV_WARN_LIMIT</b>	RW	2	0000
4A	<b>IOUT_OC_WARN_LIMIT</b>	RW	2	03FF
4F	<b>OT_FAULT_LIMIT</b>	RW	2	0358
51	<b>OT_WARN_LIMIT</b>	RW	2	03FF
58	<b>VIN_UV_WARN_LIMIT</b>	RW	2	0000
59	<b>VIN_UV_FAULT_LIMIT</b>	RW	2	0000
6B	<b>PIN_OP_WARN_LIMIT</b>	RW	2	7FFF
78	<b>STATUS_BYTE</b>	RO	1	0
79	<b>STATUS_WORD</b>	RO	2	0
7A	<b>STATUS_VOUT</b>	RO	1	0
7B	<b>STATUS_IOUT</b>	RO	1	0
7C	<b>STATUS_INPUT</b>	RO	1	0
7D	<b>STATUS_TEMPERATURE</b>	RO	1	0
7E	<b>STATUS_CML</b>	RO	1	0
80	<b>STATUS_MFR_SPECIFIC</b>	RO	1	0

ADDRESS (HEX)	NAME	TYPE	BYTES	DEFAULT (HEX)
86	<b>READ_EIN</b>	RO	2	0
88	<b>READ_VIN</b>	RO	2	0
89	<b>READ_IIN</b>	RO	2	0
8B	<b>READ_VOUT</b>	RO	2	0
8C	<b>READ_IOUT</b>	RO	2	0
8D	<b>READ_TEMPERATURE</b>	RO	2	0
97	<b>READ_PIN</b>	RO	2	0
98	<b>PMBUS_REVISION</b>	RO	1	22
99	<b>MFR_ID</b>	BLK	5	VLTR
9A	<b>MFR_MODEL</b>	BLK	6	VT505
9B	<b>MFR_REVISION</b>	BLK	2	3
D0	<b>CONFIG</b>	RW	2	0
D1	<b>PEAK_VIN</b>	RO	2	0
D2	<b>PEAK_IOUT</b>	RO	2	0
D3	<b>PEAK_PIN</b>	RO	2	0
D4	<b>PEAK_TEMP</b>	RO	2	0
D5	<b>CLEAR_PEAKE</b>	WO	0	0
DD	<b>CONFIG2</b>	RW	1	0
F2	<b>MFR_SPEC_HYSTL</b>	RW	2	03FF
F3	<b>MFR_SPEC_HYSTH</b>	RW	2	03FF
F4	<b>MFR_SPEC_HYST_STAT</b>	RO	1	0

**Note 1:** Registers shown in bold (1h–9Dh) comply with PMBus Power Management Management Protocol Specifications. Refer to the PMBus specification and [Table 15](#) for more details.

**Note 2:** SMBALERT\_MASK is written to a write\_word command and read with a read from a write\_block\_read\_process\_call with a num\_bytes field of 1.

### SMBUS\_ALERT Behavior

Any of the STATUS bits (with some exception) asserts the alert line (SMBUS\_ALERT) low.

Exceptions are:

- Bit off of STATUS\_BYTE/STATUS\_WORD
- Bit POWERGOOD# of STATUS\_WORD

Alert line can be configured to mask any of the STATUS bits using the SMBALERT\_MASK register.

The only ways to release the alert line are:

- CLEAR\_FAULTS command
- ARA (see SMBUS spec v2.0)

**Table 10. Register 80h (STATUS\_MFR\_SPECIFIC) Interpretation**

BIT #	BIT NAME	DESCRIPTION	READING	INDICATION
[7]	SELFCHK_FAULT	Pass FET and soft-start self-check	0	Pass
			1	Fail
[6]	ROCP_FAULT	R <sub>OCP</sub> check	0	Pass
			1	Fail
[5]	R_SMBUSID_FAULT	SMBus_ID programming resistor check	0	Pass
			1	Fail
[4]	FOLLOWER_FAULT	Fault input	0	No Fault Input
			1	Fault Input
[3]	GATE_UVLO_FAULT	V <sub>GS_UVLO</sub> fault	0	No Fault
			1	Fault
[2]	BST_UVLO_FAULT	V <sub>BST_UVLO</sub> fault	0	No Fault
			1	Fault
[1]	—	Not used	0	—
			1	—
[0]	—	Always read as 0	0	—

**Table 11. MAX16545B/C PMBus Register Map**

ADDRESS		DEC	NAME	TYPE	WIDTH	DEFAULT	LOW BYTE							HIGH BYTE								
							b7	b6	b5	b4	b3	b2	b1	b0	b15	b14	b13	b12	b11	b10	b9	b8
01	1		OPERATION	RW	1	80	enable	0	0	0	0	0	0	0								
03	3		CLEAR_FAULTS	WO	0	0																
19	25		CAPABILITY	CN	1	B0	1	0	1	0	0	0	0	0								
1B	27		SMBALERT_MASK	**	2	0000																
43	67		VOUT_UV_WARN_LIMIT	RW	2	0	vout_uv_wam_th[7]	vout_uv_wam_th[6]	vout_uv_wam_th[5]	vout_uv_wam_th[4]	vout_uv_wam_th[3]	vout_uv_wam_th[2]	vout_uv_wam_th[1]	vout_uv_wam_th[0]	0	0	0	0	0	0	vout_uv_wam_th[9]	vout_uv_wam_th[8]
4A	74		IOUT_OC_WARN_LIMIT	RW	2	03FF	ocp_warn_th[7]	ocp_warn_th[6]	ocp_warn_th[5]	ocp_warn_th[4]	ocp_warn_th[3]	ocp_warn_th[2]	ocp_warn_th[1]	ocp_warn_th[0]	0	0	0	0	0	0	ocp_warn_th[9]	ocp_warn_th[8]
4F	79		OT_FAULT_LIMIT	RW	2	0358	ovt_fault_th[7]	ovt_fault_th[6]	ovt_fault_th[5]	ovt_fault_th[4]	ovt_fault_th[3]	ovt_fault_th[2]	ovt_fault_th[1]	ovt_fault_th[0]	0	0	0	0	0	0	ovt_fault_th[9]	ovt_fault_th[8]
51	81		OT_WARN_LIMIT	RW	2	03FF	ovt_warn_th[7]	ovt_warn_th[6]	ovt_warn_th[5]	ovt_warn_th[4]	ovt_warn_th[3]	ovt_warn_th[2]	ovt_warn_th[1]	ovt_warn_th[0]	0	0	0	0	0	0	ovt_warn_th[9]	ovt_warn_th[8]
47	87		VIN_OV_WARN_LIMIT	RW	2	03FF	ov_warn_th[7]	ov_warn_th[6]	ov_warn_th[5]	ov_warn_th[4]	ov_warn_th[3]	ov_warn_th[2]	ov_warn_th[1]	ov_warn_th[0]	0	0	0	0	0	0	ov_warn_th[9]	ov_warn_th[8]
58	88		VIN_UV_WARN_LIMIT	RW	2	0	uv_warn_th[7]	uv_warn_th[6]	uv_warn_th[5]	uv_warn_th[4]	uv_warn_th[3]	uv_warn_th[2]	uv_warn_th[1]	uv_warn_th[0]	0	0	0	0	0	0	uv_warn_th[9]	uv_warn_th[8]
59	89		VIN_UV_FAULT_LIMIT	RW	2	0	vin_uv_fault_th[7]	vin_uv_fault_th[6]	vin_uv_fault_th[5]	vin_uv_fault_th[4]	vin_uv_fault_th[3]	vin_uv_fault_th[2]	vin_uv_fault_th[1]	vin_uv_fault_th[0]	0	0	0	0	0	0	vin_uv_fault_th[9]	vin_uv_fault_th[8]
6B	107		PIN_OP_WARN_LIMIT	RW	2	FFFF	op_warn_th[7]	op_warn_th[6]	op_warn_th[5]	op_warn_th[4]	op_warn_th[3]	op_warn_th[2]	op_warn_th[1]	op_warn_th[0]	op_warn_th[15]	op_warn_th[14]	op_warn_th[13]	op_warn_th[12]	op_warn_th[11]	op_warn_th[10]	op_warn_th[9]	op_warn_th[8]
78	120		STATUS_BYTE	RO	1	0	0	OFF	0	IOUT_OC_FAULT	VIN_UV_FAULT	TEMPERATURE	CML	NOA								
79	121		STATUS_WORD	RO	2	0	0	OFF	0	IOUT_OC_FAULT	VIN_UV_FAULT	TEMPERATURE	CML	NOA	VOUT	IOUT/POUT	INPUT	MFR_SPECIFIC	POWER_GOOD#	0	0	BAD_FET_HEALTH
7A	122		STATUS_VOUT	RO	1	26	0	0	VOUT_UV_WARNING	0	0	0	0	0								
7B	123		STATUS_IOUT	RO	1	0	IOUT_OC_FAULT	0	IOUT_OC_WARNING	0	0	0	0	0								
7C	124		STATUS_INPUT	RO	1	0	VIN_OV_FAULT	0	VIN_UV_WARNING	VIN_UV_FAULT	Unit off for low in V	IIN_OC_FAULT	IIN_OC_WARNING	PIN_OP_WARNING								
7D	125		STATUS_TEMPERATURE	RO	1	0	OT_FAULT	OT_WARNING	0	0	0	0	0	0								
7E	126		STATUS_CML	RO	1	0	invalid/unsupported cmd	invalid/unsupported data	PEC Failed	0	0	0	Other communication fault	0								
80	128		STATUS_MFR_SPECIFIC	RO	1	0	SELFCHK_FAULT	ROCP_FAULT	R_SMBUS_ID_FAULT	FOLLOWER_FAULT	GATE_UV_FAULT	BST_UV_O_FAULT	0	0								
86	134		READ_EIN	BLK	7	0	81h06	enrgy_cnt[7:0]	enrgy_cnt[15:8]	roll_cnt[7:0]	samp_cnt[7:0]	samp_cnt[15:8]	samp_cnt[23:16]									
88	136		READ_VIN	RO	2	0	vadc_ave[7]	vadc_ave[6]	vadc_ave[5]	vadc_ave[4]	vadc_ave[3]	vadc_ave[2]	vadc_ave[1]	vadc_ave[0]	0	0	0	0	0	0	vadc_ave[9]	vadc_ave[8]
89	137		READ_IIN	RO	2	0	iadc_ave[7]	iadc_ave[6]	iadc_ave[5]	iadc_ave[4]	iadc_ave[3]	iadc_ave[2]	iadc_ave[1]	iadc_ave[0]	0	0	0	0	0	0	iadc_ave[9]	iadc_ave[8]

**Table 11: MAX16545B/C PMBus Register Map (continued)**

ADDRESS		DEC	NAME	TYPE	WIDTH	DEFAULT	LOW BYTE							HIGH BYTE								
							b7	b6	b5	b4	b3	b2	b1	b0	b15	b14	b13	b12	b11	b10	b9	b8
8B	139		READ_VOUT	RO	2	0	vadc_ave[7]	vadc_ave[6]	vadc_ave[5]	vadc_ave[4]	vadc_ave[3]	vadc_ave[2]	vadc_ave[1]	vadc_ave[0]	0	0	0	0	0	0	vadc_ave[9]	vadc_ave[8]
8C	140		READ_IOUT	RO	2	0	iadc_ave[7]	iadc_ave[6]	iadc_ave[5]	iadc_ave[4]	iadc_ave[3]	iadc_ave[2]	iadc_ave[1]	iadc_ave[0]	0	0	0	0	0	0	iadc_ave[9]	iadc_ave[8]
8D	141		READ_TEMPERATURE_1	RO	2	0	tadc_ave[7]	tadc_ave[6]	tadc_ave[5]	tadc_ave[4]	tadc_ave[3]	tadc_ave[2]	tadc_ave[1]	tadc_ave[0]	0	0	0	0	0	0	tadc_ave[9]	tadc_ave[8]
97	151		READ_PIN	RO	2	0	pwr_calc[7]	pwr_calc[6]	pwr_calc[5]	pwr_calc[4]	pwr_calc[3]	pwr_calc[2]	pwr_calc[1]	pwr_calc[0]	0	pwr_calc[14]	pwr_calc[13]	pwr_calc[12]	pwr_calc[11]	pwr_calc[10]	pwr_calc[9]	pwr_calc[8]
98	152		PMBUS_REVISION	RO	1	03	0	0	1	0	0	0	1	0								
99	153		MFR_ID	BLK	5	"VLTR"	8h04	"V"	"L"	"T"	"R"											
9A	154		MFR_MODEL	BLK	6	"VT505"	8h05	"V"	"T"	"5"	"0"	"5"										
9B	155		MFR_REVISION	BLK	2	"1"	8h01	"1"														
D0	208		CONFIG	RW	2	0	ocp_cfg	mocp_cfg[1]	mocp_cfg[0]		ov_th[1]	ov_th[0]	num_aves_alert[1]	num_aves_alert[0]			disable_iadc	ss_cfg[1]	ss_cfg[0]	otp_cfg	pwr_cfg[1]	pwr_cfg[0]
D1	209		PEAK_VIN	RO	2	0	vadc_peak[7]	vadc_peak[6]	vadc_peak[5]	vadc_peak[4]	vadc_peak[3]	vadc_peak[2]	vadc_peak[1]	vadc_peak[0]	0	0	0	0	0	0	vadc_peak[9]	vadc_peak[8]
D2	210		PEAK_IOUT	RO	2	0	iadc_peak[7]	iadc_peak[6]	iadc_peak[5]	iadc_peak[4]	iadc_peak[3]	iadc_peak[2]	iadc_peak[1]	iadc_peak[0]	0	0	0	0	0	0	iadc_peak[9]	iadc_peak[8]
D3	211		PEAK_PIN	RO	2	0	pwr_peak[7]	pwr_peak[6]	pwr_peak[5]	pwr_peak[4]	pwr_peak[3]	pwr_peak[2]	pwr_peak[1]	pwr_peak[0]	pwr_peak[15]	pwr_peak[14]	pwr_peak[13]	pwr_peak[12]	pwr_peak[11]	pwr_peak[10]	pwr_peak[9]	pwr_peak[8]
D4	212		PEAK_TEMP	RO	2	0	tadc_peak[7]	tadc_peak[6]	tadc_peak[5]	tadc_peak[4]	tadc_peak[3]	tadc_peak[2]	tadc_peak[1]	tadc_peak[0]	0	0	0	0	0	0	tadc_peak[9]	tadc_peak[8]
D5	213		CLEAR_PEAKS	WO	0	0																
DD	221		CONFIG_2	RW	2	0	0	0	ocp_su_cfg[1]	ocp_su_cfg[0]	num_aves_reporting[3]	num_aves_reporting[2]	num_aves_reporting[1]	num_aves_reporting[0]								
F2	242		MFR_SPECIFIC_HYST_L	RW	2	03FF	hystl[7]	hystl[6]	hystl[5]	hystl[4]	hystl[3]	hystl[2]	hystl[1]	hystl[0]							hystl[9]	hystl[8]
F3	243		MFR_SPECIFIC_HYST_H	RW	2	03FF	hysth[7]	hysth[6]	hysth[5]	hysth[4]	hysth[3]	hysth[2]	hysth[1]	hysth[0]							hysth[9]	hysth[8]
F4	244		MFR_SPECIFIC_HYST_STATUS	RO	1	0	0	0	0	0	out > oc_warn_limit	HYST pin status	HYST > reg_F3	HYST < reg_F2								
FD	253		PEAK_VOUT	RO	2	0	vout_peak[7]	vout_peak[6]	vout_peak[5]	vout_peak[4]	vout_peak[3]	vout_peak[2]	vout_peak[1]	vout_peak[0]	0	0	0	0	0	0	vout_peak[9]	vout_peak[8]



### CONFIG Register (D0h)

This is a 2-byte register used to configure the MAX16545B/C. The default value is 0000h. The meaning of the bits of this register is shown in [Table 12](#).

### CONFIG\_2 Register (DDh)

This single-byte register allows the selection of the number of samples to average for voltage, current and power reporting. It also sets the threshold for startup OCP. Only bits 5:0 are used; bits 7:6 have no effect and should be left as 00b. The meaning of the bits of this register is shown in [Table 13](#).

**Table 12. Register D0h**

BIT #	BIT NAME	SETTINGS (BINARY)	DESCRIPTION
[13]	I-ADC disable	0	I-ADC enabled (default value)
		1	I-ADC disabled
[12:11]	ss_cfg[1:0]	00	Soft-start delay = 0ms (default value)
		01	Soft-start delay = 10ms
		10	Soft-start delay = 20ms
		11	Not used
[10]	Not Used		
[9:8]	pwrgd_th[1:0]	00	Power good threshold = 11V, self-check threshold = 9V
		01	Power good threshold = 10V, self-check threshold = 8V
		10	Power good threshold = 9V, self-check threshold = 7V
		11	Power good threshold = 8V, self-check threshold = 6V
[7]	ocp_cfg	0	Severe OCP = 130% of Moderate OCP (default value)
		1	Severe OCP = 150% of moderate OCP

BIT #	BIT NAME	SETTINGS (BINARY)	DESCRIPTION
[6:5]	mocp_cfg[1:0]	00	Moderate OCP timeout = 100μs (default value)
		01	Moderate OCP timeout = 250ms
		10	Moderate OCP timeout = 100ms
		11	Moderate OCP timeout = 10μs
[3:2]	ov_th[1:0]	00	Set $V_{IN\_OV}$ threshold = 16V (default value)
		01	Set $V_{IN\_OV}$ threshold = 14V
		10	Set $V_{IN\_OV}$ threshold = 17V
		11	Set $V_{IN\_OV}$ threshold = 18V
[1:0]	num_aves_alert[1:0]	00	Sets averaging for PMBus warning levels to one sample (default value)
		01	Sets averaging for PMBus warning levels to two samples
		10	Sets averaging for PMBus warning levels to four samples
		11	Sets averaging for PMBus warning levels to eight samples

### Peak Voltage, Current, Power, and Temperature Reporting

The two-byte registers shown in [Table 14](#) provide readings of the peak values for input/output voltage, output current, input power and temperature in 10 bits direct format.

These registers can be reset with the CLEAR\_PEAKS command (send byte D5h) or power cycling the part.

**Table 13. Register DDh**

BITS #	BIT NAME	SETTING (BINARY)	STARTUP OCP SETTING/AVERAGING SAMPLE SIZE
[5:4]	su_ocrp_cfg	00	16A
		01	24A
		10	8A
		11	Not Used
[3:0]	num_aves_reporting	0	1 sample
		1	2 samples
		10	4 samples
		11	8 samples
		100	16 samples
		101	32 samples
		110	64 samples
		111	128 samples
		1000	256 samples
		1001	512 samples
		1010	1024 samples
		1011	2048 samples
		1100	4096 samples
		1101	8192 samples
		1110	16384 samples
		1111	32768 samples

**Note:** One sample for voltage/power/energy reporting is 2ms and 1ms for current reporting.

**Table 14. Manufacturer-Specific Direct Reporting Registers**

REGISTER ADDRESS	REGISTER NAME	PARAMETER
D1h	PEAK_VIN	Peak Input Voltage
D2h	PEAK_IOUT	Peak Output Current
D3h	PEAK_PIN	Peak Input Power
D4h	PEAK_TEMP	Peak Temperature
FDh	PEAK_VOUT	Peak Output Voltage

### PMBus Reporting Format

The MAX16545B/C uses the PMBus DIRECT number format; the actual readings in their respective units correlate with the numerical values read from registers, as shown in Equation 14.

- Current/Voltage/Temperature: 10-bit resolution
- Power: 16-bit two's complement representation

**Equation 14:**

$$X = \frac{1}{m} (Y \times 10^{-R} - b)$$

where:

X = the calculated, real-world value in the appropriate units (A, V, °C, etc.);

m = the slope coefficient, is a two byte, two's complement integer;

Y = is a two byte two's complement integer received from the PMBus device;

b = the offset is a two byte, two's complement integer

R = the exponent is a one byte, two's complement integer.

The values used in the MAX16545B/C for the parameters above are shown in [Table 15](#). Note that current and power readings depend on the value of R<sub>LOAD</sub>, the external current reporting resistor connected between ILOAD and GND.

**Table 15. PMBus Equation Parameters**

PMBUS REGISTER	FORMAT	DATA BYTES	m	b	R	UNITS
READ_IOUT (8Ch)*, READ_IIN (89h)*	Direct	2	3.824 x R <sub>LOAD</sub>	-4300	-3	A
READ_VOUT (8Bh), READ_VIN (88h)	Direct	2	7578	0	-2	V
READ_TEMPERATURE_1 (8Dh)	Direct	2	205	6545	-2	°C
READ_PIN (97h)	Direct	2	0.895 x R <sub>LOAD</sub>	-9100	-2	W
READ_EIN (86h)**	Direct	2	3.505 x R <sub>LOAD</sub>	0	-5	**

\*Digital current reporting is available even if the FET is OFF.

\*\*Refer to the PMBus specifications for more details on definition and calculation.

### Input Capacitance ( $C_{IN}$ ) Selection

Use of input capacitors is highly recommended to guarantee the input voltage is stable and noise free. For applications requiring no input capacitors before the MAX16545B/C, the input-voltage ripple should be less than 300mV peak-to-peak.

### Output Capacitance ( $C_{OUT}$ ) Selection

The maximum output capacitance can be calculated as shown in Equation 15.

**Equation 15:**

$$C_{OUT} = \frac{(I_{INRUSH} \times C_{SS})}{I_{SS}}$$

where:

$C_{SS}$  = Soft-start programming capacitance.

$I_{SS}$  = Soft-start current, 30μA (typ).

$C_{OUT}$  = Maximum load capacitance that can be used at soft-start with a purely capacitive load.

$I_{INRUSH}$  = Desired maximum inrush current during startup. Select  $I_{INRUSH}$  lower than startup OCP ( $I_{OCP(STARTUP)}$ ) and within startup SOA, refer to [Figure 2](#).

### Design Example

Assume a design value for maximum inrush current of 10A, and a soft-start capacitance of 25nF.

Assume 12V application and 30μA (typ) soft-start current; soft-start time in this case is 0.01ms. The maximum safe operating output current is 22A at 0.01ms (refer to the Startup SOA in [Figure 2](#)) and the default Startup OCP level is 16A (refer to [Table 8](#)).

The designed maximum inrush current of 10A is lower than 16A startup OCP and within startup SOA. The designed value is valid. Hence, the maximum load capacitance is calculated as shown in Equation 16.

**Equation 16:**

$$C_{OUT} = \frac{(10A \times 25nF)}{30\mu A} = 8.33mF$$

Additionally, the recommended output capacitance should be less than 10mF to prevent false triggering of self-check during startup into precharged output.

### Input TVS Diode Selection

The use of a transient voltage suppression (TVS) diode at input is necessary to clamp input-voltage transient within rating of  $V_{IN}$  pin of MAX16545B/C (see the [Absolute Maximum Ratings](#) section).

A general guide to select the proper TVS diode is listed below:

- Choose TVS diode reverse-standoff voltage ( $V_{RWM}$ )  $\geq$  operating voltage of MAX16545B/C, 12V (typ).
- Choose TVS diode peak-pulse current ( $I_{PPM}$ )  $\geq$  maximum transient peak-pulse current of the MAX16545B/C, 60A (typ).
- Choose TVS clamping voltage ( $V_C$ )  $\leq$  maximum voltage handling capability of the MAX16545B/C, 22V (typ) for 150μs.

Recommend SMCJ13A based on selection criteria above.

### Output Schottky diode selection

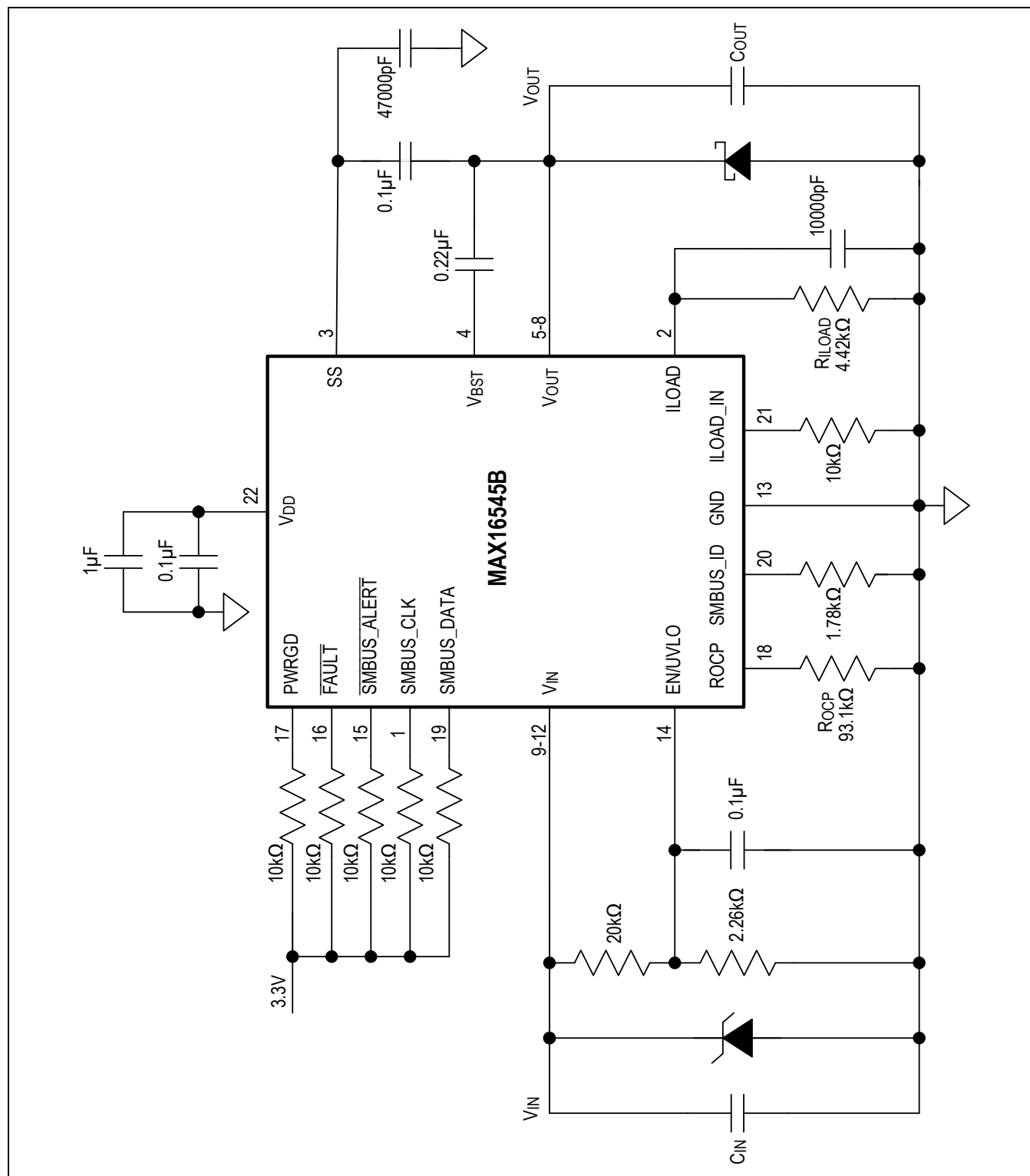
The use of a Schottky diode at output is necessary to clamp the negative output-voltage spike within the rating of  $V_{OUT}$  pin of the MAX16545B/C (see the [Absolute Maximum Ratings](#) section). Select the proper Schottky diode with low forward-voltage drop ( $V_F$ ) and peak forward-surge current ( $I_{FSM}$ ) higher than the expected inductive current.

### Reverse Current is Not Allowed During Output “Hot-Plug”

The MAX16545B/C devices must not be used in applications where a removable load assembly can be connected while there is stored energy on the input of the removable load. This is because stored energy in a load device can force current backwards through the pass-FET body diode, and the MAX16545B/C was not designed for this condition.

However, output load cards and devices (such as disk drives) may be removed and replaced while the system is powered, provided that the system design guarantees that the load device cannot be reconnected while there is still electrical charge on the input capacitance of the load device.

# Integrated Protection IC on 12V Bus with an Integrated MOSFET, Lossless Current Sensing, and PMBus Interface



### MAX16545B Layout Recommendations

Please request Application Note 6848, Optimal Application Circuit and Layout Guidelines for the MAX16545 and MAX16543, for detailed information on circuit board layout. Important layout details are summarized here.

#### $V_{IN}$ and $V_{OUT}$

- Minimize input and output trace inductance by using wide and multiple  $V_{IN}$  and  $V_{OUT}$  planes for optimal thermal performance.
- Use multiple vias to connect interlaying power planes.
- Place input capacitors (where applicable) as close to the IC as possible.
- Place output capacitors as close to the IC as possible.

#### Example

The EV kit layout in [Figure 7](#) shows use of large, wide power planes for  $V_{IN}$  and  $V_{OUT}$  on the top layer.  $C_{IN}$  and  $C_{OUT}$  are close to the IC. The other  $V_{IN}$  and  $V_{OUT}$  power layers are connected with multiple vias between the pins.

#### Ground

- Use a trace approximately 0.1mm wide by 0.7mm long from the ground pin (pin 13) on the top layer to the second-layer AGND island.

#### $V_{BST}$ and SS

Place the  $V_{BST}$  and SS capacitors on the top layer as close to the pins as possible.

#### $V_{DD}$

Add a  $V_{DD}$  plane on the top layer to decouple the  $V_{DD}$  caps close to the IC to form a tighter loop to ground.

#### ROCP and ILOAD

The ROCP and ILOAD resistors should be placed as close to the IC as possible.

#### ILOAD\_IN

Connect a 10k $\Omega$  resistor to ground.

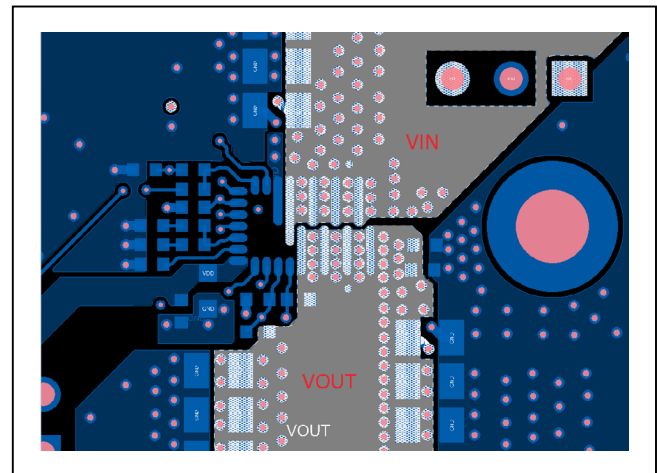


Figure 7. Top Layer (Power)

# MAX16545B MAX16545C

Integrated Protection IC on 12V Bus with an Integrated MOSFET, Lossless Current Sensing, and PMBus Interface

## Ordering Information

PART NUMBER	DESCRIPTION	PACKAGE	DRAWING NUMBER	SHIPPING METHOD	PACKAGE MARKING
MAX16545BGPF+	Startup OCP = 16A	22 FCQFN	ES AP-3236	2.5kμ Tape & Reel	MAX16545B
MAX16545BGPF+T					
MAX16545CGPF+	Startup OCP = 24A				
MAX16545CGPF+T					MAX16545C

+Denotes a lead(Pb)-free/RoHS-compliant part.

T = Tape and reel.

**Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	6/17	Initial release	—
1	12/18	Add new MAX16545C version, Improved Figure 2, Startup SOA	1, 5, 18, 19, 21, 28, 34, 35, 38
2	2/19	Revised application use-case and circuit layout guidelines. Clarified TOC titles and descriptions.	1, 9, 12-14, 18, 35, 37
3	4/19	Revised title of “Output Hot-Plug” warning in the <i>Detailed Description</i> section	35
4	6/19	Added “future product” note to the <i>Ordering Information</i> table.	38
5	10/19	Removed “future product” note from MAX16545C and corrected overvoltage protection to detection only	7, 17, 21, 23, 25, 28, 32, 33, 38
6	6/22	Updated Soft-Start Charging Current (min) in the <i>Electrical Characteristics</i> table	5



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