

200mA Buck-Boost Synchronous DC/DC Converters

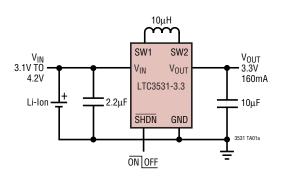
#### **FEATURES**

- Regulated Output with Input Above, Below or Equal to the Output
- **Single Inductor**
- Up to 90% Efficiency
- V<sub>IN</sub> Range: 1.8V to 5.5V
- 200mA at 3.3V<sub>DUT</sub> from 3.6V Input
- 125mA at  $3V_{OUT}$  from 2.5V Input
- Fixed VOLT Versions (TSOT, DFN): 3.3V, 3V
- Adjustable V<sub>OUT</sub> Version (DFN): 2V to 5V
- Burst Mode<sup>®</sup> Operation, No External Compensation
- Ultra Low Quiescent Current: 16µA, Shutdown Current <1uA
- Only 3 External Components Required (Fixed Voltage Versions)
- Short-Circuit Protection
- Output Disconnect in Shutdown
- Available in 6-Pin ThinSOT and 3mm × 3mm DFN Packages

### APPLICATIONS

- Handheld Instruments
- **MP3** Players
- Handheld computers
- PDA/GPS

### **TYPICAL APPLICATION**

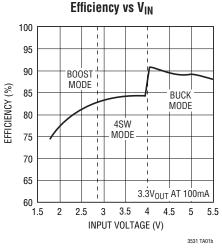


#### DESCRIPTION

The LTC®3531/LTC3531-3.3/LTC3531-3 are synchronous buck-boost DC/DC converters that operate from input voltages above, below or equal to the output voltage. The topology incorporated in the ICs provides a continuous transfer through all operating modes, making the product ideal for single cell Li-Ion and multicell alkaline or nickel applications. The converters operate in Burst Mode, minimizing solution footprint and component count as well as providing high conversion efficiency over a wide range of load currents.

The devices include two  $0.5\Omega$  N-channel MOSFET switches and two P-channel switches  $(0.5\Omega, 0.8\Omega)$ . Quiescent current is typically 16µA, making the parts ideal for battery power applications. Other features include a  $<1\mu$ A shutdown current, current limiting, thermal shutdown and output disconnect. The parts are offered in a 6-pin ThinSOT<sup>™</sup> package for fixed voltage versions or a 3mm × 3mm DFN package for fixed and adjustable versions.

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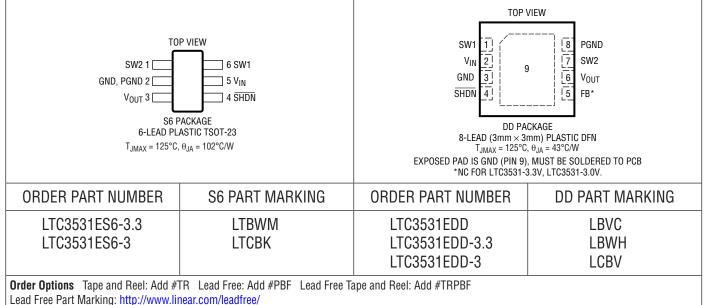


#### **ABSOLUTE MAXIMUM RATINGS**

(Note 1)

VIN, VOUT, SW1, SW2, SHDN Voltage .....-0.3 to 6V SW1, SW2 Voltage, <100ns Pulse .....-0.3 to 7V Operating Temperature Range (Notes 2,3)-40°C to 85°C 

## PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.

# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T<sub>A</sub> = 25°C. V<sub>IN</sub> = 3.6V V<sub>OUT</sub> = 3.3V unless otherwise noted.

PARAMETER		CONDITIONS		MIN	ТҮР	MAX	UNITS
V <sub>IN</sub>							
Minimum Startup Voltage			•		1.65	1.8	V
V <sub>OUT</sub> Regulation							
Output Voltage (3.3V Version)		No Load	•	3.25	3.32	3.39	V
Output Voltage (3V Version)		No Load	•	2.95	3.02	3.09	V
FB Voltage (Adj Version)		No Load	•	1.20	1.225	1.25	V
FB Input Current (Adj Version)		V <sub>FB</sub> = 1.225V			1	50	nA
Operating Current							
Quiescent Current in Sleep:	V <sub>IN</sub>	V <sub>IN</sub> = 5V, V <sub>OUT</sub> = 3.6V, FB = 1.3V			16	30	μA
	V <sub>OUT</sub>	V <sub>OUT</sub> = 3.6V			6	10	μA
Shutdown Current	V <sub>IN</sub>	$\overline{\text{SHDN}} = 0\text{V}, \text{ V}_{\text{OUT}} = 0\text{V}$				1	μA
Switch Performance							
NMOS Switch Leakage		Switches B and C			0.2	2	μA
PMOS Switch Leakage		Switches A and D			0.2	2	μA
NMOS B, C R <sub>DSON</sub>		$V_{IN} = 5V$			0.5		Ω
PMOS A R <sub>DSON</sub>		V <sub>IN</sub> = 5V			0.5		Ω
PMOS A R <sub>DSON</sub>		V <sub>IN</sub> = 5V			0.5		L



**ELECTRICAL CHARACTERISTICS** The  $\bullet$  denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T<sub>A</sub> = 25°C. V<sub>IN</sub> = 3.6V V<sub>OUT</sub> = 3.3V unless otherwise noted.

PARAMETER	CONDITIONS	MIN	ТҮР	MAX	UNITS
PMOS D R <sub>DSON</sub> (3.3V Version or Adjustable Version set to 3.3V)	V <sub>OUT</sub> = 3.1V		0.8		Ω
PMOS D R <sub>DSON</sub> (3V Version)	V <sub>OUT</sub> = 2.8V		0.9		Ω
Peak Current Limit	L = 10µH, V <sub>IN</sub> = 5V	295	365	460	mA
SHDN					
SHDN Input Threshold		0.4	1	1.4	V
SHDN Hysteresis			60		mV
SHDN Leakage Current	VSHDN		0.01	1	μA

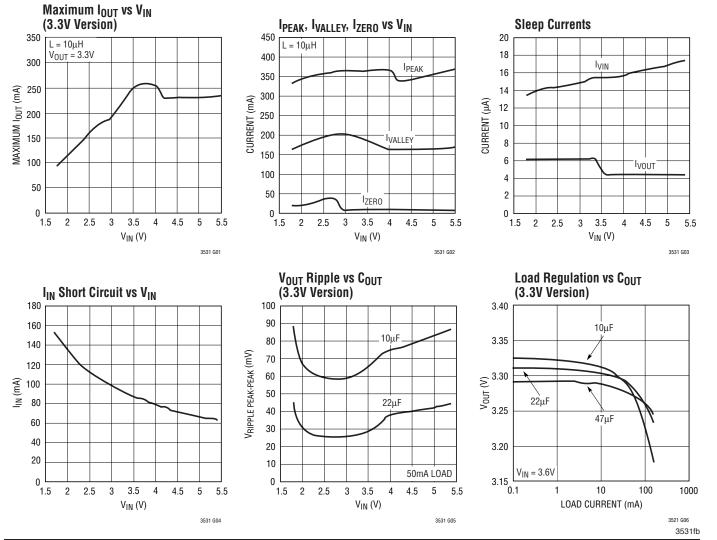
Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: This IC includes overtemperature protection that is intended to protect the device during momentary overload conditions. Junction temperature will exceed 125°C when overtemperature protection is active. Continuous operation above the specified maximum operating junction temperature may result in device degradation or failure.

Note 3: The LTC3531 is guaranteed to meet performance specifications from 0°C to 70°C. Specifications over the -40°C to 85°C operating temperature range are assured by design, characterization and correlation with statistical process controls.

#### **TYPICAL PERFORMANCE CHARACTERISTICS**

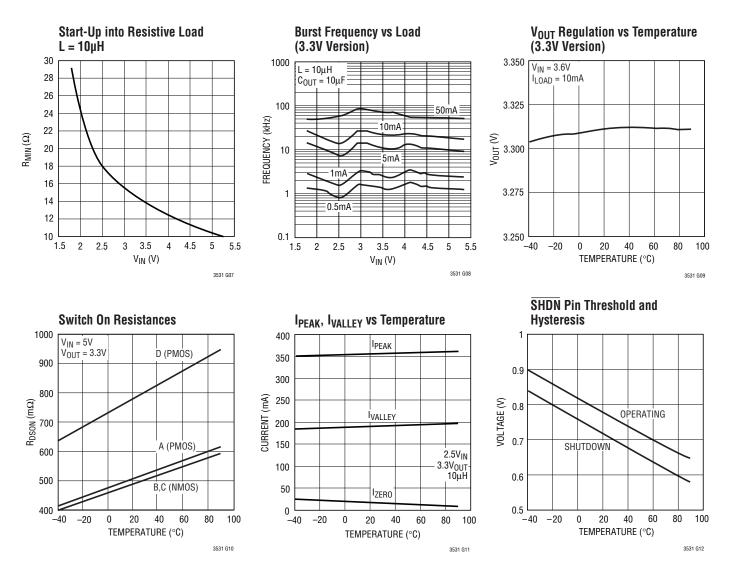
 $T_A = 25^{\circ}C$  unless otherwise specified.





#### TYPICAL PERFORMANCE CHARACTERISTICS T<sub>A</sub> = 25°C uni

 $T_A = 25^{\circ}C$  unless otherwise specified.

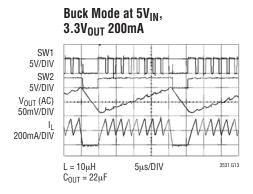




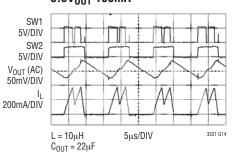


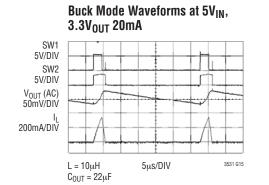
#### TYPICAL PERFORMANCE CHARACTERISTICS

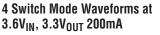
 $T_A = 25^{\circ}C$  unless otherwise specified.

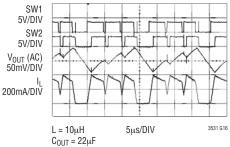


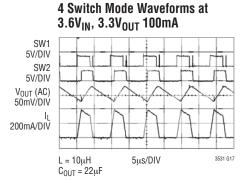
Buck Mode at 5V<sub>IN</sub>, 3.3V<sub>OUT</sub> 100mA



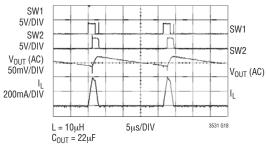






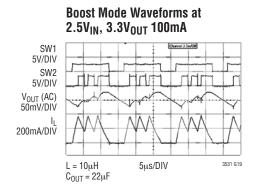








#### **TYPICAL PERFORMANCE CHARACTERISTICS** $T_A = 25^{\circ}C$ unless otherwise specified.

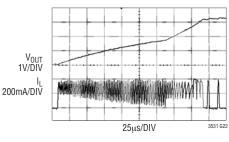


Boost Mode Waveforms at 2.5V<sub>IN</sub>, 3.3V<sub>OUT</sub> 20mA

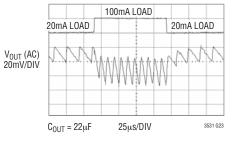
 $C_{OUT} = 22 \mu F$ 

Shorted Output SW1 SV/DIV SW2 SV/DIV V<sub>OUT</sub> (AC) 50mV/DIV L 200mA/DIV L L = 10 $\mu$ H 1 $\mu$ s/DIV 3531 621 COUT = 22 $\mu$ F V<sub>IN</sub> 5

#### Start-Up into 50mA Load at 3.3 V<sub>OUT</sub> (Shows Start, Buck, Then 4sw Modes)

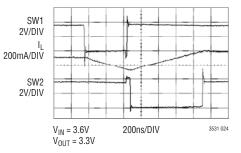






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## SW1 and SW2 Close-Up in Four Switch Mode





#### PIN FUNCTIONS ThinSOT/DFN Packages

**SW2 (Pin 1/Pin 7):** Buck-Boost Switch Pin Where Internal Switches C and D are Connected. An optional Schottky diode can be connected from SW2 to V<sub>OUT</sub> for a moderate efficiency improvement. Minimize trace length to keep EMI down.

GND (Pin 2/Pin 3): Signal Ground for the IC.

**PGND (Pin 2/Pin 8):** Power Ground for the IC. (Shared on ThinSOT version)

 $V_{OUT}$  (Pin 3/Pin 6): Output of the Buck-Boost Synchronous Rectifier. A filter capacitor is placed from V<sub>OUT</sub> to GND. A ceramic bypass capacitor is recommended as close to the V<sub>OUT</sub> and GND pins as possible.

**SHDN** (Pin 4/Pin 4): External Shutdown Pin. An applied voltage of < 0.4V shuts down the converter. A voltage above >1.4V will enable the converter.

 $V_{IN}$  (Pin 5/Pin 2): Input Supply Pin for the Buck-Boost Converter. A minimum 2.2µF Ceramic Capacitor should be placed between  $V_{IN}$  and GND.

**FB (NA/Pin 5):** Feedback Pin for the Adjustable Version. Connect the resistor divider tap here. The output voltage can be adjusted from 2V to 5V.

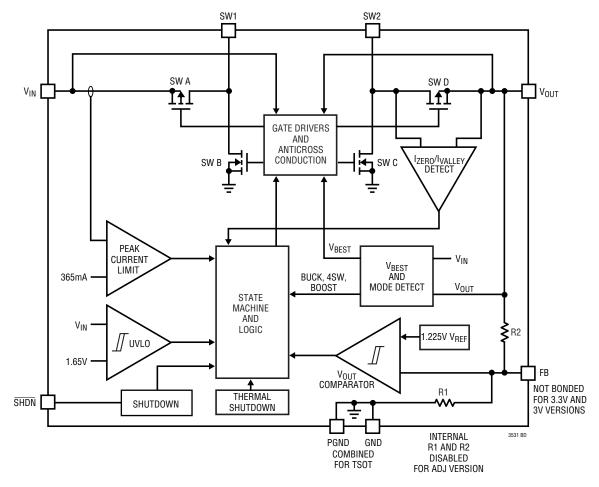
$$V_{OUT} = 1.225 \left(1 + \frac{R2}{R1}\right)$$

**SW1 (Pin 6/Pin 1):** Buck-Boost Switch Pin Where Internal Switches A and B are Connected. Connect the inductor from SW1 to SW2.

**Exposed Pad (Pin 9, DFN):** Solder to PCB ground for optimal thermal performance.



### **BLOCK DIAGRAM**





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### OPERATION

The LTC3531, LTC3531-3.3 and LTC3531-3 synchronous buck-boost converters utilize a Burst Mode control technique to achieve high efficiency over a wide dynamic range of load currents. A 2% accurate comparator is used to monitor the output voltage. If  $V_{OUT}$  is above its programmed reference threshold no switching occurs and only quiescent current is drawn from the power source (sleep mode). When  $V_{OUT}$  drops below the reference threshold the IC "wakes up", switching commences, and the output capacitor is charged. The value of the output capacitor, the load current, and the comparator hysteresis (~1%) determines the number of current pulses required to pump-up the output capacitor before the part returns to sleep.

In order to determine the best operating mode for the converter, the LTC3531 contains a second comparator that monitors the relative voltage difference between  $V_{IN}$  and  $V_{OUT}$ . Input and output voltages in the various modes as well as typical inductor currents are shown in Figure 1. Regions of the current waveforms where switches A and D are on provide the highest efficiency since energy is transferred directly from the input source to the output.

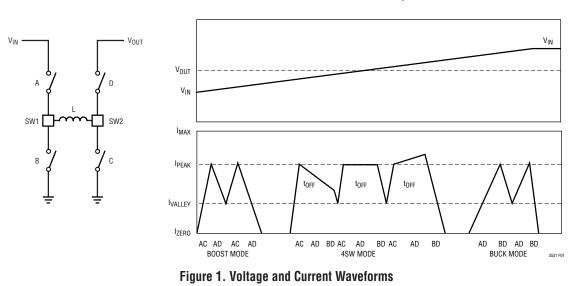
#### Boost Mode

If  $V_{IN}$  is ~400mV below  $V_{OUT}$ , the LTC3531 operates in boost or step-up mode. Referring to Figure 1 (left side) when  $V_{OUT}$  falls below its regulation voltage, switches A and C are turned on ( $V_{IN}$  is applied across the inductor) and current is ramped until  $I_{PEAK}$  is detected. When this

occurs, C is turned off, D is turned on and current is delivered to the output capacitor ( $V_{IN} - V_{OUT}$  is applied across the inductor). Inductor current falls when D is on, until an  $I_{VALLEY}$  is detected. Terminating at  $I_{VALLEY}$ , rather than  $I_{ZERO}$ , results in an increased load current capability for a given peak current. This AC then AD switch sequence is repeated until the output is pumped above its regulation voltage, a final  $I_{ZERO}$  is detected, and the part returns to sleep mode ( $I_{VALLEY}$  is ignored and  $I_{ZERO}$  is used in all modes once  $V_{OUT}$  is above its programmed value).

#### **4-Switch Mode**

If  $(V_{OUT} - 400 \text{mV}) < -V_{IN} < (V_{OUT} + 800 \text{mV})$ , the LTC3531 operates in 4-switch step-up/down mode. Returning to Figure 1 (center) when V<sub>OUT</sub> falls below its regulation voltage, switches A and C are turned on and current is ramped until IPFAK is detected. As with Boost Mode operation, C is then turned off, D is turned on and current is delivered to the output. When A and D are on, the inductor current slope is dependant on the relationship between  $V_{IN}$ ,  $V_{OUT}$ , and the  $R_{DSON}$  of the switches. In 4-switch mode, a  $t_{OFF}$ timer (approximately 3µs) is used to terminate the AD pulse. Once the  $t_{OFF}$  timer expires, switch A is turned off, B is turned on and inductor current is ramped down (V<sub>OUT</sub> is applied across the inductor) until IVALLEY is detected. This sequence is repeated until the output is regulated, BD switches are turned on, and a final I<sub>ZEBO</sub> is detected. Anticross conduction circuitry in all modes ensures the P-channel MOSFET and N-channel MOSFET switch pairs (A and B or D and C) are never turned on simultaneously.





### OPERATION

#### **Buck Mode**

If V<sub>IN</sub> is ~800mV above V<sub>OUT</sub>, the LTC3531 operates in buck or step-down mode. The higher offset between V<sub>IN</sub> and V<sub>OUT</sub> (800mV) is required to ensure sufficient magnetizing voltage across the inductor when the R<sub>DSONS</sub> are taken into account. At the beginning of a buck mode cycle (Figure 1 right side) switches A and D are turned on (V<sub>IN</sub> - V<sub>OUT</sub> is applied across the inductor), current is delivered to the output and ramped up until I<sub>PEAK</sub> is detected. When this occurs, A is turned off, B is turned on and inductor current falls (-V<sub>OUT</sub> across the inductor) until an I<sub>VALLEY</sub> is detected. This AD then BD switch sequence is repeated until the output is pumped above its regulation voltage, a final I<sub>ZERO</sub> is detected, and the part returns to sleep mode.

#### Start-Up Mode

Before V<sub>OUT</sub> reaches approximately 1.6V, the D switch is disabled and its body diode is used to transfer current to the output capacitor. In start-up mode, the I<sub>VALLEY</sub>/I<sub>ZERO</sub> sense circuit is disabled and an alternate algorithm is used to control inductor current. When the LTC3531 is brought out of shutdown (assuming V<sub>OUT</sub> is discharged) switches A and C are turned on until the inductor current reaches I<sub>PEAK</sub>. The AC switches are then turned off and inductor current flows to the output through the B switch and D body diode. The period for the B switch/D body diode is controlled by the t<sub>OFF</sub> timer to ~800nS. This sequence of AC switch-on to I<sub>PEAK</sub> then B switch and D body diode for ~800ns is repeated until V<sub>OUT</sub> reaches ~1.6V. Once this threshold is reached, the LTC3531 will transfer through the required modes until  $V_{\text{OUT}}$  is brought into regulation.

Due to propagation delays in the sense circuitry, the magnitudes of the  $I_{PEAK},\,I_{VALLEY}$ , and  $I_{ZERO}$  currents may shift depending on  $V_{IN},\,V_{OUT}$  and operating mode.

#### **OTHER LTC3531 FEATURES**

**Shutdown:** The part is shut down by pulling SHDN below 0.4V, and made active by pulling the pin up to  $V_{IN}$  or  $V_{OUT}$ . Note that SHDN an be driven above  $V_{IN}$  or  $V_{OUT}$ , as long as it is limited to less than 6V.

**Output Disconnect and Inrush Limiting:** The LTC3531 is designed to allow true output disconnect by opening both P-channel MOSFET rectifiers. This allows  $V_{OUT}$  to go to zero volts during shutdown, drawing no current from the input source. It also provides inrush current limiting at turn-on, minimizing surge currents seen by the input supply.

**Thermal Shutdown:** If the die temperature reaches approximately 150°C, the part will go into thermal shutdown and all switches will be turned off. The part will be enabled again when the die temperature has dropped by 10°C (nominal). To deliver the power that the LTC3531 is capable of, it is imperative that a good thermal path be provided to dissipate the heat generated within the package. It is recommended that multiple vias in the printed circuit board be used to conduct heat away from the IC and into a copper plane with as much area as possible. Soldering the Exposed Pad to the GND plane (DFN version) is recommended to improve thermal performance.



### **APPLICATIONS INFORMATION**

#### **Component Selection**

Only three power components are required to complete the design of the buck-boost converter,  $V_{OUT}$  programming resistors are needed for the adjustable version. The high operating frequency and low peak currents of the LTC3531 allow the use of low value, low profile inductors and tiny external ceramic capacitors.

#### **Inductor Selection**

For best efficiency, choose an inductor with high frequency core material, such as ferrite, to reduce core loses. The inductor should have low DCR (DC resistance) to reduce the  $I^2R$  losses, and must be able to handle the peak inductor current without saturating. A 10µH to 22µH in-

#### Table 2. Inductor Vendor Information

Supplier	Series	Phone	Website
COEV	DN4835	(800) 227-7040	www.coev.net
Coilcraft	MSS4020 LP03310 DS1608	(847) 639-6400	www.coilcraft.com
Murata	LQH43CN LQH32CN	USA: (814) 237-1431 (800) 831-9172	www.murata.com
Sumida	CDRH4D18 CDRH3D16/HP	USA: (847) 956-0666 Japan: 81-3-3607-5111	www.sumida.com
Toko	D312C D412C DB320C	(847) 297-0070	www.tokoam.com

ductor value with a >500mA current rating and <400m $\Omega$  DCR is recommended. For applications where radiated noise is a concern, a toroidal or shielded inductor can be used. Table 2 contains a list of inductor manufacturers.

#### **Capacitor Selection**

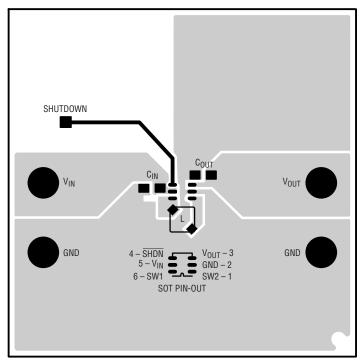
The buck-boost convertor requires two capacitors. Ceramic X5R types will minimize ESL and ESR while maintaining capacitance at rated voltage over temperature. The  $V_{IN}$  capacitor should be at least 2.2µF. The  $V_{OUT}$  capacitor should be between 4.7µF and 22µF. A larger output capacitor should be used if lower peak to peak output voltage ripple is desired. A larger output capacitor will also improve load regulation on  $V_{OUT}$ . See Table 3 for a list of capacitor manufacturers for input and output capacitor selection.

#### Table 3. Capacitor Vendor Information

Supplier	Series	Phone	Website
AVX	X5R	(803) 448-9411	www.avxcorp.com
Murata	X5R	USA: (814) 237-1431 (800) 831-9172	www.murata.com
Sanyo	POSCAP	(619) 661-6322	www.sanyovideo.com
Taiyo Yuden	X5R	(408) 573-4150	www.taiyo-yuden.com
TDK	X5R	(847) 803-6100	www.component.tdk.com



### **APPLICATIONS INFORMATION**

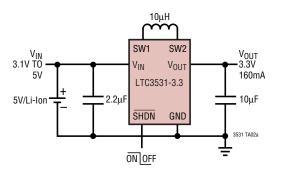


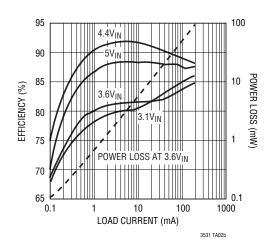
**Recommended Layout (SOT Versions)** 



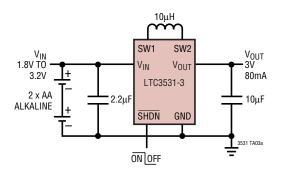
#### TYPICAL APPLICATION

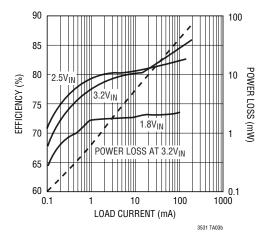
5V/Li-Ion to 3.3V with ThinSOT (3.3V Version)



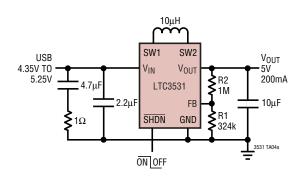


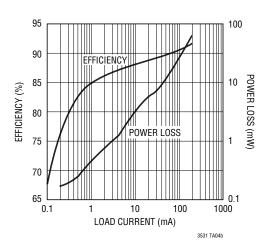
2 AA Alkaline to 3V with ThinSOT (3V Version)





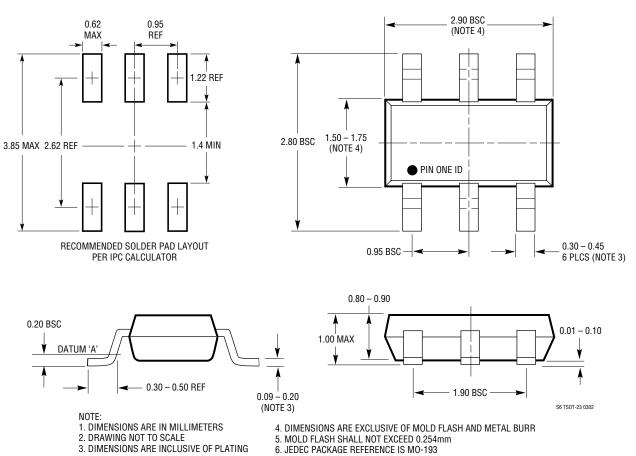








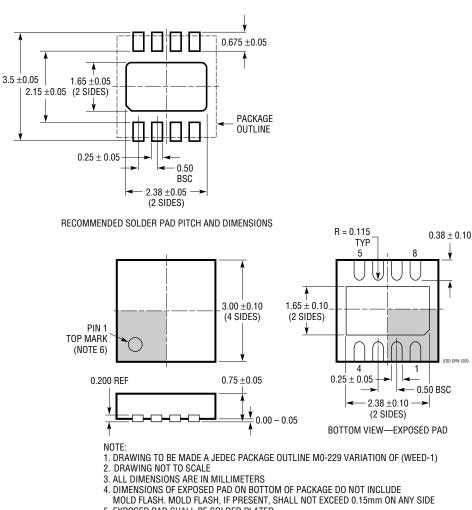
### PACKAGE DESCRIPTION



S6 Package 6-Lead Plastic TSOT-23 (Reference LTC DWG # 05-08-1636)



#### PACKAGE DESCRIPTION



**DD Package** 8-Lead Plastic DFN (3mm × 3mm) (Reference LTC DWG # 05-08-1698)

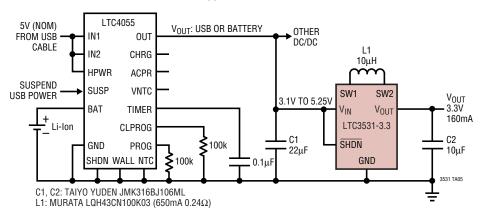
5. EXPOSED PAD SHALL BE SOLDER PLATED

6. SHADED AREA IS ONLY A REFERENCE FOR PIN 1 LOCATION





### TYPICAL APPLICATION



## Complete USB/Li-Ion Powered System with 3.3V<sub>OUT</sub> and Linear Charger

#### **RELATED PARTS**

PART NUMBER	DESCRIPTION	COMMENTS	
LT1930/LT1930A	1A (I <sub>SW</sub> ), 1.2MHz/2.2MHz, High Efficiency Step-Up DC/DC Converter	$V_{\text{IN}}$ : 2.6V to 16V, $V_{\text{OUT}(\text{MAX})}$ = 34V, $I_{\text{Q}}$ = 4.2mA/5.5mA, $I_{\text{SD}}$ < 1µA, ThinSOT Package	
LTC3400/LTC3400B	$600\text{mA}~(\text{I}_{\text{SW}}),1.2\text{MHz}$ Synchronous Step-Up DC/DC Converter	$V_{\text{IN}}$ : 0.85V to 5V, $V_{\text{OUT}(\text{MAX})}$ = 5V, $I_{\text{Q}}$ = 19µA/300µA, $I_{\text{SD}}$ < 1µA, ThinSOT Package	
LTC3401/LTC3402	1A/2A (I <sub>SW</sub> ), 3MHz Synchronous Step-Up DC/DC Converter	$V_{IN}$ : 0.5V to 5V, $V_{OUT(MAX)}$ = 6V, $I_Q$ = 38µA, $I_{SD}$ < 1µA, MS Package	
LTC3405/LTC3405A	300mA (I <sub>OUT</sub> ), 1.5MHz Synchronous Step-Down DC/DC Converter	$V_{IN}$ : 2.7V to 6V, $V_{OUT(MIN)}$ = 0.8V, $I_Q$ = 20µA, $I_{SD} \leq$ 1µA, MS10 Package	
LTC3406/LTC3406B	600mA (I <sub>OUT</sub> ), 1.5MHz Synchronous Step-Down DC/DC Converter	$V_{\text{IN}}$ : 2.5V to 5.5V, $V_{\text{OUT}(\text{MIN})}$ = 0.6V, $I_{\text{Q}}$ = 20µA, $I_{\text{SD}}$ $\leq$ 1µA, ThinSOT Package	
LTC3421	3A (I <sub>SW</sub> ), 3MHz Synchronous Step-Up DC/DC Converter	$V_{\text{IN}}$ : 0.5V to 4.5V, $V_{\text{OUT}(\text{MAX})}$ = 5.25V, $I_{\text{Q}}$ = 12µA, $I_{\text{SD}}$ < 1µA, QFN Package	
LTC3422	1.5A (I <sub>SW</sub> ), 3MHz Synchronous Step-Up DC/DC Converter	$V_{\text{IN}}$ : 0.5V to 4.5V, $V_{\text{OUT}(\text{MAX})}$ = 5.25V, $I_{\text{Q}}$ = 25µA, $I_{\text{SD}}$ < 1µA, 3mm × 3mm DFN Package	
LTC3426	2A (I <sub>SW</sub> ), 1.2MHz Step-Up DC/DC Converter in SOT-23	V <sub>IN</sub> : 1.6V to 5V, V <sub>OUT</sub> up to 5.5V	
LTC3428	4A (I <sub>SW</sub> ), 1.2MHz Step-Up DC/DC Converter	V <sub>IN</sub> : 1.6V to 5V, V <sub>OUT</sub> up to 5.5V	
LTC3429	600mA ( $I_{SW}$ ), 500kHz Synchronous Step-Up DC/DC Converter	$V_{IN}$ : 0.5V to 4.4V, $V_{OUT(MIN)}$ = 5V, $I_Q$ = 20µA, $I_{SD}$ < 1µA, QFN Package	
LTC3440	600mA (I <sub>OUT</sub> ), 2MHz Synchronous Buck-Boost DC/DC Converter	$V_{\text{IN}}$ : 2.5V to 5.5V, $V_{\text{OUT}(\text{MIN})}$ = 5.5V, $I_{\text{Q}}$ = 25µA, $I_{\text{SD}}$ < 1µA, MS, DFN Packages	
LTC3441	600mA (I <sub>OUT</sub> ), 2MHz Synchronous Buck-Boost DC/DC Converter	V <sub>IN</sub> : 2.5V to 5.5V, V <sub>OUT(MIN)</sub> = 5.5V, I <sub>Q</sub> = 25μA, I <sub>SD</sub> < 1μA, DFN Package	
LTC3442	2MHz Synchronous Buck-Boost with Auto-Burst	V <sub>IN</sub> : 2.4V to 5.5V, V <sub>OUT</sub> up to 5.25V	
LTC3443	1.2A (I <sub>OUT</sub> ), 600kHz Synchronous Buck-Boost DC/DC Converter	$V_{\text{IN}}$ : 2.4V to 5.5V, $V_{\text{OUT}(\text{MIN})}$ = 5.25V, $I_{\text{Q}}$ = 28µA, $I_{\text{SD}}$ < 1µA, MS Package	
LTC3458	1.4A, 1.5MHz Synchronous Step-Up DC/DC Converter	V <sub>IN</sub> : 1.5V to 6V, V <sub>OUT</sub> up to 7.5V	
LTC3458L	1.7A, 1.5MHz Synchronous Step-Up DC/DC Converter	V <sub>IN</sub> : 1.5V to 6V, V <sub>OUT</sub> up to 6V	
LTC3459	10V Micropower Synchronous Step-Up DC/DC Converter	V <sub>IN</sub> : 1.5V to 5.5V, V <sub>OUT</sub> up to 10V	
LTC3525/LTC3525-3.3/ LTC3525-5	400mA (I <sub>SW</sub> ), Synchronous Step-Up DC/DC Converter with Output Disconnect	V <sub>IN</sub> : 0.5V to 4.5V, I <sub>Q</sub> = 7µA, I <sub>SD</sub> < 1µA, 2mm × 2mm SC70 Package	



### **Mouser Electronics**

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