

LTC1257

Complete Single Supply 12-Bit Voltage Output DAC in SO-8

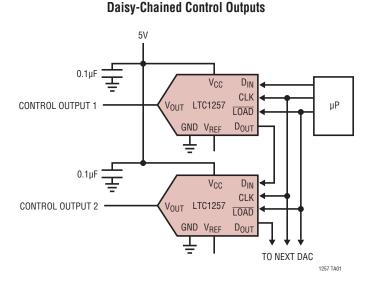
FEATURES

- 8-Pin SO Package
- Buffered Voltage Output
- Built-In 2.048V Reference
- 500µV/LSB with 2.048V Full Scale
- 1/2LSB Max DNL Error
- Guaranteed 12-Bit Monotonic
- 3-Wire Cascadable Serial Interface
- Wide Single Supply Range: V_{CC} = 4.75V to 15.75V
- Low Power: I_{CC} Typ = 350µA with 5V Supply

APPLICATIONS

- Digital Offset/Gain Adjustment
- Industrial Process Control
- Automatic Test Equipment

TYPICAL APPLICATION

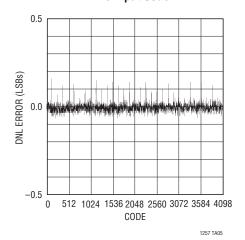


DESCRIPTION

The LTC®1257 is a complete single supply, 12-bit voltage output D/A converter (DAC) in an SO-8 package. The LTC1257 includes an output buffer amplifier, 2.048V voltage reference and an easy to use 3-wire cascadable serial interface. An external reference can be used to override the internal reference and extend the output voltage range to 12V. The power supply current is a low 350µA when operating from a 5V supply, making the LTC1257 ideal for battery-powered applications. The space-saving 8-pin SO package and operation with no external components provide the smallest 12-bit D/A system available.

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Differential Nonlinearity vs Input Code



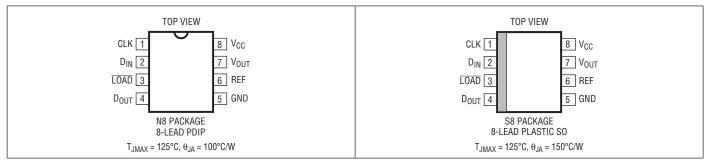


ABSOLUTE MAXIMUM RATINGS (Note 1)

V _{CC} to GND	0.5V to 16.5V
TTL Input Voltage	–0.5V to V _{CC} + 0.5V
V _{OUT}	$-0.5V$ to V _{CC} + 0.5V
REF	–0.5V to V _{CC} + 0.5V
Operating Temperature Range	
LTC1257C	0°C to 70°C
LTC12571	40°C to 85°C

Maximum Junction Temperature	
Plastic Package65°C to 125°C	;
Storage Temperature Range65°C to 150°C	;
Lead Temperature (Soldering, 10 sec) 300°C	;

PIN CONFIGURATION



ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LTC1257CN8#PBF	LTC1257CN8#TRPBF	LTC1257CN8	8-Lead PDIP	0°C to 70°C
LTC1257IN8#PBF	LTC1257IN8#TRPBF	LTC1257IN8	8-Lead PDIP	-40°C to 85°C
LTC1257CS8#PBF	LTC1257CS8#TRPBF	1257	8-Lead Plastic SO	0°C to 70°C
LTC1257IS8#PBF	LTC1257IS8#TRPBF	12571	8-Lead Plastic SO	-40°C to 85°C

Consult LTC Marketing for parts specified with wider operating temperature ranges.

Consult LTC Marketing for information on nonstandard lead based finish parts.

For more information on lead free part marking, go to: http://www.linear.com/leadfree/

For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/

ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = T_{MIN}$ to T_{MAX} . $V_{CC} = 4.75V$ to 15.75V, internal or external reference (2.475V $\leq V_{REF} \leq V_{CC} - 2.7V$), unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
DAC							
	Resolution		•	12			Bits
DNL	Differential Nonlinearity	Guaranteed Monotonic (Note 4)	•			±0.5	LSB
INL	Integral Nonlinearity	LTC1257C (Note 4)	•			±3.5	LSB
		LTC1257I (Note 4)	•			±4.0	LSB





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SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
OFF	Offset Error	When Using Internal Reference, LTC1257C				±8	LSB
		When Using Internal Reference, LTC12571	•			±10	LSB
		When Using External Reference, LTC1257C				±4	mV
		When Using External Reference, LTC12571				±5	mV
OFF _{TC}	Offset Error Tempco	When Using Internal Reference (Note 2) When Using External Reference (Note 2)			±0.02 ±15	±0.066 ±30	LSB/°C µV/°C
	Gain Error				0.5	±30 ±2	LSB
	Gain Error Tempco	(Note 2)			±0.01	±2 ±0.02	
Reference		(NOLE 2)	•		±0.01	±0.02	LSB/°C
Reference	Deference Output Voltege			0.000	0.040	0.000	V
	Reference Output Voltage	I _{REF} = 0, LTC1257C I _{RE} F = 0, LTC1257I		2.028 2.018	2.048	2.068 2.078	V
	Reference Output Tempco	$I_{\text{REF}} = 0$		2.010	±0.06	2.070	LSB/°C
	Reference Line Regulation	I _{RFF} = 0, LTC1257C			10.00	±0.4	LSB/V
	noisionee Ente nogulation	$I_{REF} = 0, LTC12571$				±0.7	LSB/V
	Reference Load Regulation	$0\mu A \le I_{REF} \le 100\mu A$	•			±1	LSB
	Reference Input Range	$V_{CC} > V_{BEF} + 2.7V$		2.475		12	V
	Reference Input Resistance			8	14	18	kΩ
	Reference Input Capacitance	(Note 2)			15		pF
	Short-Circuit Current	V _{BFF} Shorted to GND				90	mA
Power Sup							
V _{CC}	Positive Supply Voltage	For Specified Performance		4.75		15.75	V
	Supply Current	$4.75V \le V_{CC} \le 5.25V$			350	600	μA
00		$4.75V \le V_{CC} \le 15.75V$			800	1500	μΑ
Op Amp D(C Performance						
	Short-Circuit Current Low	V _{OUT} Shorted to GND				60	mA
	Short-Circuit Current High	V _{OUT} Shorted to V _{CC}				60	mA
	Output Impedance to GND	Input Code = 0			250	500	Ω
AC Perforn	nance						
	Voltage Output Slew Rate	5k Ω in Parallel with 100pF		1.0			V/µs
	Voltage Output Settling Time	To $\pm 1/2$ LSB, 5k Ω in Parallel with 100pF, V _{CC} = 4.75V			6		μs
	Digital Feedthrough	(Notes 2, 3)			50		nV/s
Digital I/O	,						
V _{IH}	Digital Input High Voltage			2.4			V
VIL	Digital Input Low Voltage					0.8	V
V _{OH}	Digital Output High Voltage	$I_{OUT} = -1mA, D_{OUT} Only$		V _{CC} – 1			V
V _{OL}	Digital Output Low Voltage	I _{OUT} = 1mA, D _{OUT} Only		0.4			V
I _{LEAK}	Digital Input Leakage	$V_{IN} = GND \text{ to } V_{CC}$				±10	μA
CIN	Digital Input Capacitance	(Note 2)				10	pF
Switching	(Note 2)						
t ₁	D _{IN} Valid to CLK Setup			100			ns
t ₂	D _{IN} Valid to CLK Hold			25			ns
t ₃	CLK High Time			350			ns
t ₄	CLK Low Time			350			ns



temperature range, otherwise specifications are at $T_A = T_{MIN}$ to T_{MAX} . $V_{CC} = 4.75V$ to 15.75V, internal or external reference (2.475V $\leq V_{REF} \leq V_{CC} - 2.7V$), unless otherwise noted.

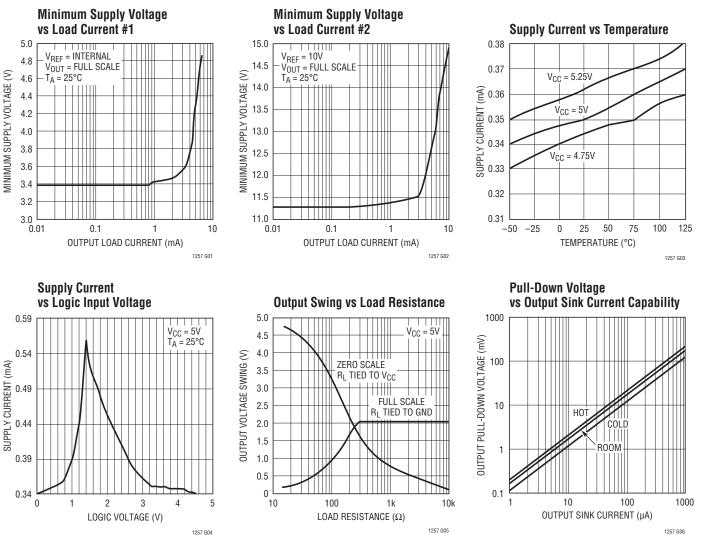
SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
t ₅	LOAD Pulse Width		•	150			ns
t ₆	LSB CLK to LOAD		•	0			ns
t ₇	LOAD High to CLK		•	0			ns
t ₈	D _{OUT} Output Delay	C _{LOAD} = 15pF	•	35		150	ns
f _{CLK}	Maximum Clock Frequency					1.4	MHz

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: Guaranteed by design; not subject to test.

Note 3: DAC switched from all 1s to all 0s, and all 0s to all 1s code. **Note 4:** Guaranteed with internal V_{RFF} or with external V_{RFF} range of 2.475V to 12V. Tested at 10V.

TYPICAL PERFORMANCE CHARACTERISTICS

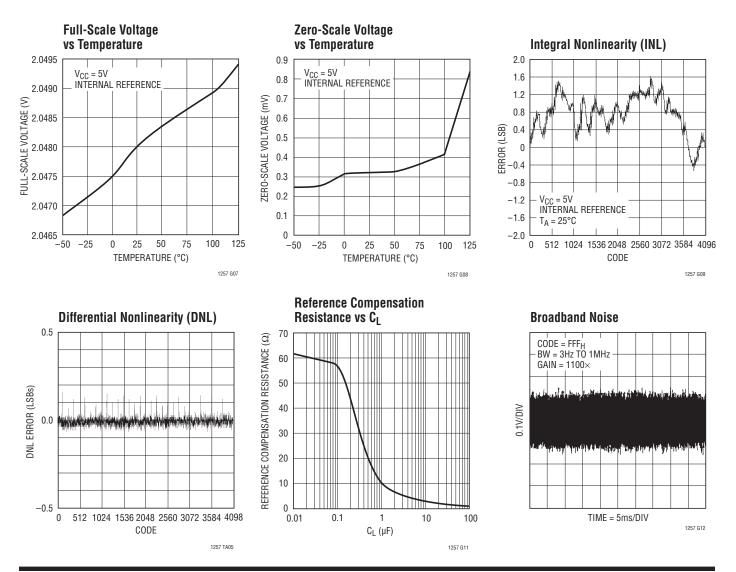


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TYPICAL PERFORMANCE CHARACTERISTICS



PIN FUNCTIONS

CLK (Pin 1): The TTL level input for the serial interface clock.

 D_{IN} (Pin 2): The TTL level input for the serial interface data. Data on the D_{IN} pin is latched into the shift register on the rising edge of the serial clock.

LOAD (Pin 3): The TTL level input for the serial interface load control. Data is loaded from the shift register into the DAC register, thus updating the DAC output when \overline{LOAD} is pulled low. The DAC register is transparent as long as \overline{LOAD} is held low.

 $D_{OUT}(Pin 4)$: The output of the shift register which becomes valid on the rising edge of the serial clock. The D_{OUT} pin is driven from GND to V_{CC} by an internal CMOS inverter. Multiple LTC1257s may be cascaded by connecting the D_{OUT} pin to the D_{IN} pin of the next chip.

GND (Pin 5): Ground.

REF (Pin 6): The output of the 2.048V reference and the input to the DAC resistor ladder. An external reference with voltage from 2.475V to V_{CC} – 2.7V may be used to override the internal reference.



PIN FUNCTIONS

 V_{OUT} (Pin 7): The buffered DAC output is capable of sourcing 2mA over temperature while pulling within 2.7V of V_{CC}. The output will pull to ground through an internal 250 Ω equivalent resistance.

 V_{CC} (Pin 8): The positive supply input. 4.75V \leq V_{CC} \leq 15.75V. Requires a bypass capacitor to ground.

DEFINITIONS

LSB: The least significant bit or the ideal voltage difference between two successive codes.

 $LSB = (V_{FS} - V_{OS})/2^{n} - 1$

- n = The number of digital input bits
- V_{OS} = The zero code error or offset of the DAC
- V_{FS} = The full-scale output voltage of the DAC measured when all bits are set to 1

Resolution: The resolution is the number of DAC output states (2^n) that divide the full-scale range. The resolution does not imply linearity.

INL: End-point integral nonlinearity is the maximum deviation from a straight line passing through the end-points of the DAC transfer curve. Because the part operates from a single supply and the output cannot go below ground, the linearity is measured between full-scale and the first code that guarantees a positive output. The INL error at a given input code is calculated as follows:

INL = $(V_{OUT} - V_{IDEAL})/LSB$

 $V_{IDEAL} = (Code)(LSB) + V_{OS}$

 V_{OUT} = The output voltage of the DAC measured at the given input code

DNL: Differential nonlinearity is the difference between the measured change and the ideal 1LSB change between any two adjacent codes. The DNL error between any two codes is calculated as follows:

DNL = $(\Delta V_{OUT} - LSB)/LSB$

 ΔV_{OUT} = The measured voltage difference between two adjacent codes

Offset Error: The theoretical voltage at the output when the DAC is loaded with all zeros. The output amplifier can have a true negative offset, but because the part is operated from a single supply, the output cannot go below ground. If the offset is negative, the output will remain near OV resulting in the transfer curve shown in Figure 1.

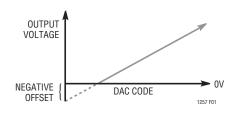


Figure 1. Effect of Negative Offset

The offset of the part is measured at the first code that produces an output voltage 0.5LSB greater than the previous code:

 $V_{OS} = V_{OUT} - [(Code)(V_{FS})/(2^n - 1)]$

Full-Scale Error: Full-scale error is the difference between the ideal and measured DAC output voltages with all bits set to one (Code = 4095). The full-scale error includes the offset error and is calculated as follows:

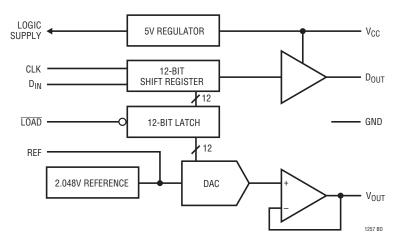
$$\begin{array}{ll} FSE &= (V_{OUT} - V_{IDEAL})/LSB \\ V_{IDEAL} &= (V_{REF})(1 - 2^{-n}) - V_{OS} \\ V_{BFF} &= The reference voltage, either internal or external \end{array}$$

Gain Error: Gain error is the difference between the ideal and measured slope of the DAC transfer characteristic. Gain error is equal to full-scale error minus offset error.

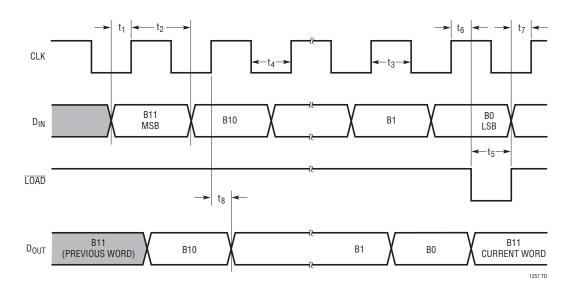
Digital Feedthrough: The glitch that appears at the analog output caused by AC coupling from the digital inputs when they change state. The area of the glitch is specified in (nV)(sec).



BLOCK DIAGRAM



TIMING DIAGRAM





OPERATION

Serial Interface

The data on the D_{IN} input is loaded into the shift register on the rising edge of the clock. The MSB is loaded first and the LSB last. The DAC register loads the data from the shift register when \overline{LOAD} is pulled low, and remains transparent until \overline{LOAD} is pulled high and the data is latched.

An internal 5V regulator provides the supply for the digital logic. By limiting the internal digital signal swings to 5V, digital noise is reduced. The buffered output of the 12-bit shift register is available on the D_{OUT} pin which will swing from GND to V_{CC} .

Multiple LTC1257s may be daisy chained together by connecting the D_{OUT} pin to the D_{IN} pin of the next chip, while the clock and load signals remain common to all chips in the daisy chain. The serial data is clocked to all of the chips, then the LOAD signal is pulled low to update all of them simultaneously. The maximum clocking rate is 1.4MHz.

Reference

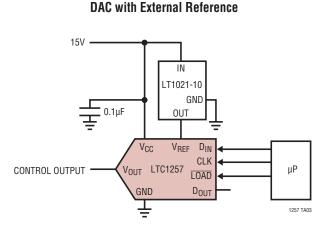
The LTC1257 includes an internal 2.048V reference, making 1LSB equal to 500μ V. The internal reference output is turned off when the pin is forced above the reference voltage, allowing an external reference to be connected to the reference pin. The external reference must be greater than 2.475V and less than V_{CC} – 2.7V, and be capable of driving the 10k minimum DAC resistor ladder.

If the reference output is driving a large capacitive load, a series resistor must be added to insure stability. For any capacitive load greater than $1\mu F$, a 10Ω series resistor will suffice.

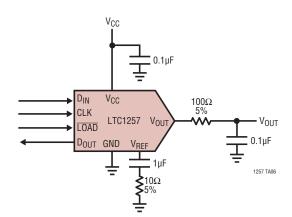
Voltage Output

The LTC1257 voltage output is able to pull within 2.7V of V_{CC} while sourcing 2mA. A internal NMOS transistor with a 200 Ω equivalent impedance pulls the output to ground. The output is protected against short circuits and is able to drive up to a 500pF capacitive load without oscillation. If digital noise on the output causes a problem, a simple 100 Ω , 0.1µF RC circuit can be used to filter the noise.

TYPICAL APPLICATIONS



Filtering V_{REF} and V_{OUT}



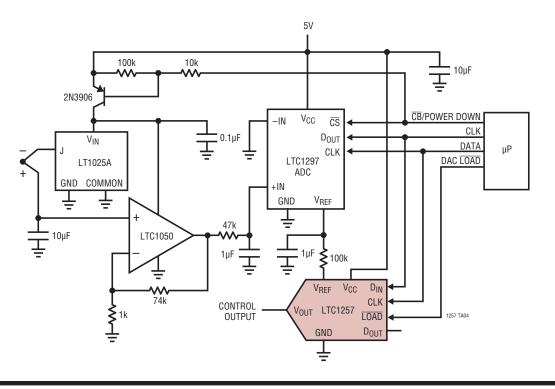
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TYPICAL APPLICATIONS

22µF 5V CH0 V_{CC} $\overline{\text{CS}}$ 8 ANALOG INPUT CHANNELS D_{OUT} : LTC1296 μP CLK D_{IN} CH7 COM REF⁺ REF⁻ SS0 Ŧ 50k 50k 5V 74HC04 0.1µF D_{IN} V_{CC} Ŧ 100Ω CLK V_{OUT} LTC1257 LOAD 0.1µF GND V_{REF} Ŧ D_{OUT} Т D_{IN} V_{CC} 100Ω CLK V_{OUT} LTC1257 LOAD 1257 TA02 GND V_{REF} D_{OUT} 0.1µF Ť Ŧ

Auto Ranging 8-Channel ADC with Shutdown

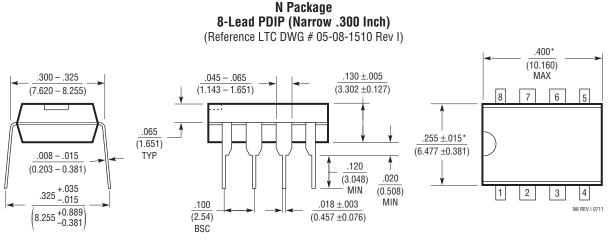






PACKAGE DESCRIPTION

Please refer to http://www.linear.com/designtools/packaging/ for the most recent package drawings.

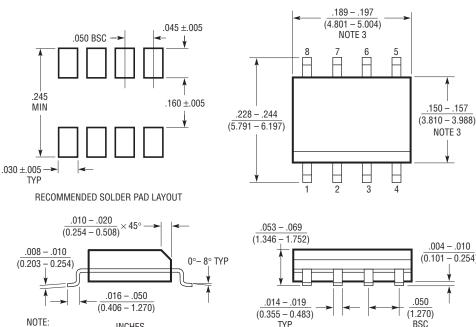


NOTE:

1. DIMENSIONS ARE MILLIMETERS INCHES

*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .010 INCH (0.254mm)

> S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610 Rev G)



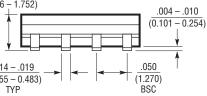
1. DIMENSIONS IN (MILLIMETERS)

2. DRAWING NOT TO SCALE

3. THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.

MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .006" (0.15mm)

4. PIN 1 CAN BE BEVEL EDGE OR A DIMPLE



S08 REV G 0212



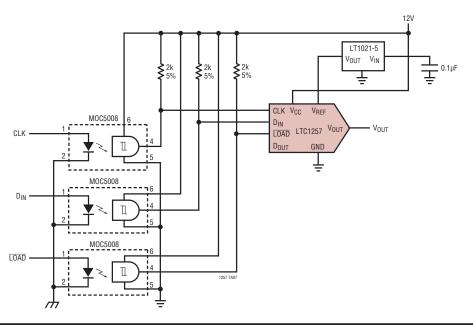
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REVISION HISTORY (Revision history begins at Rev C)

REV	DATE	DESCRIPTION	PAGE NUMBER
С	12/12	Removed MAX Voltage Output Settling Time value in Electrical Characteristics section	3



TYPICAL APPLICATION



Driving LTC1257 with Opto-Isolators

RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
12 Bit		
LTC1446/LTC1446L	Dual 12-Bit V _{OUT} DACs in SO-8 Package	LTC1446: V_{CC} = 4.5V to 5.5V, V_{OUT} = 0V to 4.095V LTC1446L: V_{CC} = 2.7V to 5.5V, V_{OUT} = 0V to 2.5V
LTC1448	Dual 12-Bit $V_{\mbox{OUT}}$ DAC in SO-8 Package, $V_{\mbox{CC}}$: 2.7V to 5.5V	Output Swings from GND to REF, REF Input Can Be Tied to $V_{\mbox{CC}}$
LTC1450/LTC1450L	Single 12-Bit V _{OUT} DACs with Parallel Interface	LTC1450: V_{CC} = 4.5V to 5.5V, V_{OUT} = 0V to 4.095V LTC1450L: V_{CC} = 2.7V to 5.5V, V_{OUT} = 0V to 2.5V
LTC1451	Single Rail-to-Rail 12-Bit V_{OUT} DAC, Full Scale: 4.095V, V_{CC} : 4.5V to 5.5V, Internal 2.048V Reference Brought Out to Pin	Low Power, Complete V _{OUT} DAC in SO-8 Package
LTC1452	Single Rail-to-Rail 12-Bit V_{OUT} Multiplying DAC, V_{CC} : 2.7V to 5.5V	Low Power, Multiplying V _{OUT} DAC with Rail-to-Rail Buffer Amplifier in SO-8 Package
LTC1453	Single Rail-to-Rail 12-Bit V _{OUT} DAC, Full Scale: 2.5V, V _{CC} : 2.7V to 5.5V	3V, Low Power, Complete V _{OUT} DAC in SO-8 Package
LTC1454/LTC1454L	Dual 12-Bit V_{OUT} DACs in SO-16 Package with Added Functionality	LTC1454: V_{CC} = 4.5V to 5.5V, V_{OUT} = 0V to 4.095V LTC1454L: V_{CC} = 2.7V to 5.5V, V_{OUT} = 0V to 2.5V
LTC1456	Single Rail-to-Rail Output 12-Bit DAC with Clear Pin, Full Scale: 4.095V, $V_{CC}{:}$ 4.5V to 5.5V	Low Power, Complete $V_{\mbox{OUT}}$ DAC in SO-8 Package with Clear Pin
LTC1458/LTC1458L	Quad 12 Bit Rail-to-Rail Output DACs with Added Functionality	LTC1458: V_{CC} = 4.5V to 5.5V, V_{OUT} = 0V to 4.095V LTC1458L: V_{CC} = 2.7V to 5.5V, V_{OUT} = 0V to 2.5V
LTC1659	Single Rail-to-Rail 12-Bit V_{OUT} DAC in MSOP-8 Package, V_{CC} = 2.7V to 5.5V	Output Swings from GND to REF, REF Input Can Be Tied to $V_{\mbox{CC}}$
14 Bit		
LTC1658	14-Bit Rail-to-Rail Micropower DAC in MSOP, V_{CC} = 2.7V to 5.5V	Output Swings from GND to REF, REF Input Can Be Tied to $V_{\mbox{CC}}$
LTC1654	Dual 14-Bit V _{OUT} DAC	Programmable Speed/Power, SO-8 Footprint
16 Bit		
LTC1655(L)	Single 16-Bit V _{OUT} DAC with Serial Interface in SO-8	V_{CC} = 5V (3V), Low Power, Deglitched, V_{OUT} = 0V to 4.096V (0V to 2.5V)



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