

Fast Settling, JFET Input Operational Amplifier

FEATURES

- 100% Tested Settling Time to 1mV at Sum Node, 10V Step
Tested with Fixed Feedback Capacitor 340ns Typ
540ns Max
- Slew Rate 60V/μs Min
- Gain-Bandwidth Product 14MHz
- Power Bandwidth (20V_{P-P}) 1.2 MHz
- Unity-Gain Stable; Phase Margin 60°
- Input Offset Voltage 600μV Max
- Input Bias Current 25°C 75pA Max
70°C 600pA Max
- Input Offset Current 25°C 40pA Max
70°C 150pA Max
- Low Distortion

APPLICATIONS

- Fast 12-Bit D/A Output Amplifiers
- High Speed Buffers
- Fast Sample-and-Hold Amplifiers
- High Speed Integrators
- Voltage to Frequency Converters
- Active Filters
- Log Amplifiers
- Peak Detectors

DESCRIPTION

The **LT®1122** JFET input operational amplifier combines high speed and precision performance.

A unique poly-gate JFET process minimizes gate series resistance and gate-to-drain capacitance, facilitating wide bandwidth performance, without degrading JFET transistor matching.

It slews at 80V/μs and settles in 340ns. The LT1122 is internally compensated to be unity-gain stable, yet it has a bandwidth of 14MHz at a supply current of only 7mA. Its speed makes the LT1122 an ideal choice for fast settling 12-bit data conversion and acquisition systems.

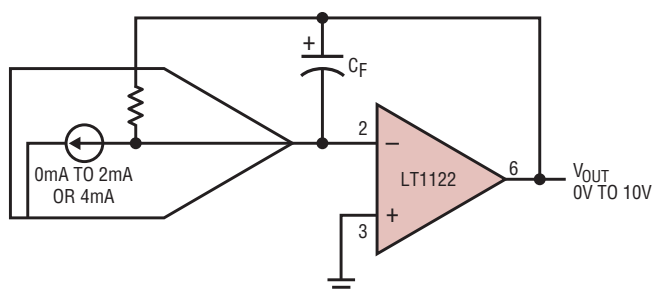
The LT1122 offset voltage of 120μV, and voltage gain of 500,000 also support the 12-bit accurate applications.

The input bias current of 10pA and offset current of 4pA combined with its speed allow the LT1122 to be used in such applications as high speed sample and hold amplifiers, peak detectors, and integrators.

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TYPICAL APPLICATION

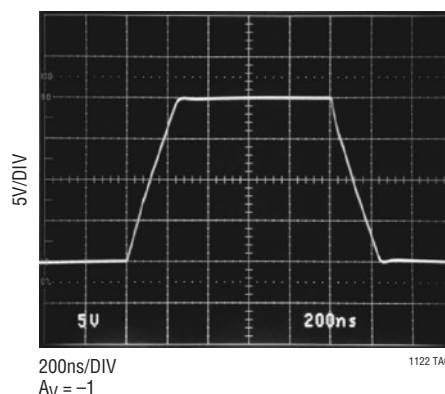
12-Bit Voltage Output D/A Converter



12-BIT CURRENT OUTPUT D/A CONVERTER
C_F = 5pF TO 17pF
(DEPENDENT ON D/A CONVERTER USED)

LT1122-TA01

Large-Scale Response



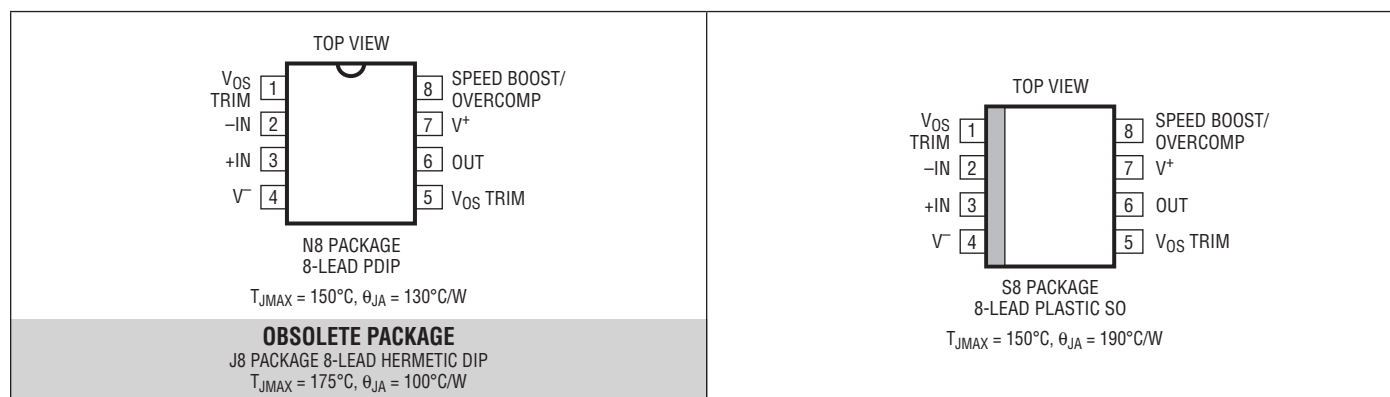
LT1122

ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage $\pm 20V$
Differential Input Voltage $\pm 40V$
Input Voltage $\pm 20V$
Output Short Circuit Duration Indefinite
Lead Temperature (Soldering, 10 sec.) $300^{\circ}C$

Operating Temperature Range
LT1122AM/BM/CM/DM (**OBSOLETE**).. $-55^{\circ}C$ to $125^{\circ}C$
LT1122AC/BC/CC/DC/CS/DS $-40^{\circ}C$ to $85^{\circ}C$
Storage Temperature Range
All Devices $-65^{\circ}C$ to $150^{\circ}C$

PIN CONFIGURATION



ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LT1122ACN8#PBF	LT1122ACN8#TRPBF	LT1122ACN8	8-Lead Plastic DIP	$-40^{\circ}C$ to $85^{\circ}C$
LT1122BCN8#PBF	LT1122BCN8#TRPBF	LT1122BCN8	8-Lead Plastic DIP	$-40^{\circ}C$ to $85^{\circ}C$
LT1122CCN8#PBF	LT1122CCN8#TRPBF	LT1122CCN8	8-Lead Plastic DIP	$-40^{\circ}C$ to $85^{\circ}C$
LT1122DCN8#PBF	LT1122DCN8#TRPBF	LT1122DCN8	8-Lead Plastic DIP	$-40^{\circ}C$ to $85^{\circ}C$
LT1122CS8#PBF	LT1122CS8#TRPBF	1122C	8-Lead Plastic SO	$-40^{\circ}C$ to $85^{\circ}C$
LT1122DS8#PBF	LT1122DS8#TRPBF	1122D	8-Lead Plastic SO	$-40^{\circ}C$ to $85^{\circ}C$
OBSOLETE PACKAGE				
LT1122AMJ8#PBF	LT1122AMJ8#TRPBF	LT1122AMJ8	8-Lead Hermetic DIP	$-55^{\circ}C$ to $125^{\circ}C$
LT1122BMJ8#PBF	LT1122BMJ8#TRPBF	LT1122BMJ8	8-Lead Hermetic DIP	$-55^{\circ}C$ to $125^{\circ}C$
LT1122CMJ8#PBF	LT1122CMJ8#TRPBF	LT1122CMJ8	8-Lead Hermetic DIP	$-55^{\circ}C$ to $125^{\circ}C$
LT1122DMJ8#PBF	LT1122DMJ8#TRPBF	LT1122DMJ8	8-Lead Hermetic DIP	$-55^{\circ}C$ to $125^{\circ}C$
LT1122ACJ8#PBF	LT1122ACJ8#TRPBF	LT1122ACJ8	8-Lead Hermetic DIP	$-40^{\circ}C$ to $85^{\circ}C$
LT1122BCJ8#PBF	LT1122BCJ8#TRPBF	LT1122BCJ8	8-Lead Hermetic DIP	$-40^{\circ}C$ to $85^{\circ}C$
LT1122CCJ8#PBF	LT1122CCJ8#TRPBF	LT1122CCJ8	8-Lead Hermetic DIP	$-40^{\circ}C$ to $85^{\circ}C$
LT1122DCJ8#PBF	LT1122DCJ8#TRPBF	LT1122DCJ8	8-Lead Hermetic DIP	$-40^{\circ}C$ to $85^{\circ}C$

Consult LTC Marketing for parts specified with wider operating temperature ranges.

Consult LTC Marketing for information on nonstandard lead based finish parts.

For more information on lead free part markings, go to: <http://www.linear.com/leadfree/>

For more information on tape and reel specifications, go to: <http://www.linear.com/tapeand reel/>

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = \pm 15\text{V}$, $V_{CM} = 0\text{V}$ unless otherwise noted. (Note 2)

SYMBOL	PARAMETER	CONDITIONS	LT1122AM/BM LT1122AC/BC			LT1122CM/DM LT1122CC/DC LT1122CS/DS			UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	
V_{OS}	Input Offset Voltage			120	600		130	900	μV
I_{OS}	Input Offset Current			4	40		5	50	pA
I_B	Input Bias Current			10	75		12	100	pA
	Input Resistance								
	Differential			10^{12}			10^{12}		Ω
	Common Mode	$V_{CM} = -10\text{V to } 8\text{V}$ $V_{CM} = 8\text{V to } 11\text{V}$		10^{12} 10^{11}			10^{12} 10^{11}		Ω Ω
	Input Capacitance			4			4		pF
SR	Slew Rate	$A_V = -1$	60	80		50	75		$\text{V}/\mu\text{s}$
	Settling Time (Note 2)	10V to 0V, -10V to 0V 100% Tested: A- and C-Grades to 1mV at Sum Node B- and D-Grades to 1mV at Sum Node All Grades to 0.5mV at Sum Node		340 350 450	540		350 360 470	590	ns ns ns
GBW	Gain-Bandwidth Product Power Bandwidth	$V_{OUT} = 20V_{P-P}$		14 1.2			13 1.1		MHz MHz
A_{VOL}	Large-Signal Voltage Gain	$V_{OUT} = \pm 10\text{V}$, $R_L = 2\text{k}\Omega$ $V_{OUT} = \pm 10\text{V}$, $R_L = 600\Omega$	180 130	500 250		150 110	450 220		V/mV V/mV
CMRR	Common-Mode Rejection Ratio	$V_{CM} = \pm 10\text{V}$	83	99		80	98		dB
	Input Voltage Range	(Note 4)	± 10.5	± 11		± 10.5	± 11		V
PSRR	Power Supply Rejection Ratio	$V_S = \pm 10\text{V to } \pm 18\text{V}$	86	103		82	101		dB
	Input Noise Voltage	0.1Hz to 10Hz		3.0			3.3		μV_{P-P}
	Input Noise Voltage Density	$f_0 = 100\text{Hz}$ $f_0 = 10\text{kHz}$		25 14			27 15		$\text{nV}/\sqrt{\text{Hz}}$ $\text{nV}/\sqrt{\text{Hz}}$
	Input Noise Current Density	$f_0 = 100\text{Hz}$, $f_0 = 10\text{kHz}$		2			2		$\text{fA}/\sqrt{\text{Hz}}$
V_{OUT}	Output Voltage Swing	$R_L = 2\text{k}\Omega$ $R_L = 600\Omega$	± 12 ± 11.5	± 12.5 ± 12		± 12 ± 11.5	± 12.5 ± 12		V V
I_S	Supply Current			7.5	10		7.8	11	mA
	Minimum Supply Voltage	(Note 5)	± 5			± 5			V
	Offset Adjustment Range	$R_{POT} \geq 10\text{k}$, Wiper to V^+	± 4	± 10		± 4	± 10		mV

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$. $V_S = \pm 15\text{V}$, $V_{CM} = 0\text{V}$. (Note 2)

SYMBOL	PARAMETER	CONDITIONS		LT1122AC/BC			LT1122CC/DC LT1122CS/DS			UNITS
				MIN	TYP	MAX	MIN	TYP	MAX	
V_{OS}	Input Offset Voltage		●	350	1400		400	2000		μV
	Average Temperature Coefficient of Input Offset Voltage		●	5	18		6	25		$\mu\text{V}/^{\circ}\text{C}$
I_{OS}	Input Offset Current		●	12	150		15	200		pA
I_B	Input Bias Current		●	80	600		90	800		pA
A_{VOL}	Large-Signal Voltage Gain	$V_{OUT} = \pm 10\text{V}$, $R_L \geq 2\text{k}\Omega$	●	120	380		100	340		V/mV
CMRR	Common-Mode Rejection Ratio	$V_{CM} = \pm 10\text{V}$	●	82	98		78	96		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 10\text{V}$ to $\pm 17\text{V}$	●	84	101		80	99		dB
	Input Voltage Range		●	± 10	± 10.8		± 10	± 10.8		V
V_{OUT}	Output Voltage Swing	$R_L = 2\text{k}\Omega$	●	± 11.5	± 12.4		± 11.5	± 12.4		V
SR	Slew Rate	$A_V = -1$	●	50	70		40	65		$\text{V}/\mu\text{s}$

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $-55^{\circ}\text{C} \leq T_A \leq 125^{\circ}\text{C}$. $V_S = \pm 15\text{V}$, $V_{CM} = 0\text{V}$. (Note 2)

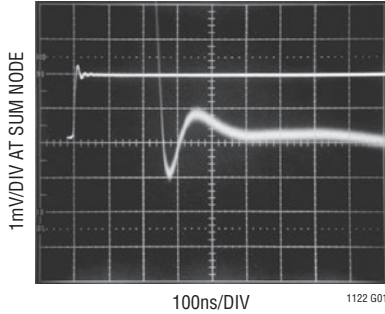
SYMBOL	PARAMETER	CONDITIONS		LT1122AM/BM			LT1122CS/DS			UNITS
				MIN	TYP	MAX	MIN	TYP	MAX	
V_{OS}	Input Offset Voltage		●	650	2400		800	3400		μV
	Average Temperature Coefficient of Input Offset Voltage		●	6	18		7	25		$\mu\text{V}/^{\circ}\text{C}$
I_{OS}	Input Offset Current		●	0.5	6		0.6	9		nA
I_B	Input Bias Current		●	6	25		7	35		nA
A_{VOL}	Large-Signal Voltage Gain	$V_{OUT} = \pm 10\text{V}$, $R_L \geq 2\text{k}\Omega$	●	70	230		60	200		V/mV
CMRR	Common-Mode Rejection Ratio	$V_{CM} = \pm 10\text{V}$	●	80	97		76	94		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 10\text{V}$ to $\pm 17\text{V}$	●	83	100		78	98		dB
	Input Voltage Range		●	± 10	± 10.5		± 10	± 10.5		V
V_{OUT}	Output Voltage Swing	$R_L = 2\text{k}\Omega$	●	± 11.3	± 12.1		± 11.3	± 12.1		V
SR	Slew Rate	$A_V = -1$	●	45	60		35	55		$\text{V}/\mu\text{s}$

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $-40^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$. $V_S = \pm 15\text{V}$, $V_{CM} = 0\text{V}$. (Note 6)

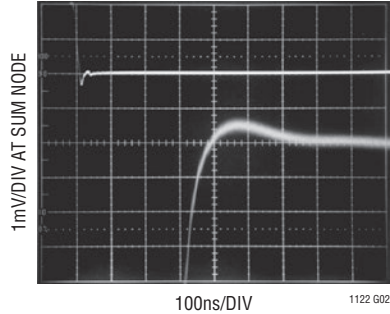
SYMBOL	PARAMETER	CONDITIONS		LT1122AM/BM			LT1122CS/DS			UNITS
				MIN	TYP	MAX	MIN	TYP	MAX	
V_{OS}	Input Offset Voltage		●	450	1900		500	2700		μV
	Average Temperature Coefficient of Input Offset Voltage		●	6	20		7	28		$\mu\text{V}/^{\circ}\text{C}$
I_{OS}	Input Offset Current		●	30	600		40	900		pA
I_B	Input Bias Current		●	230	2000		260	2700		pA
A_{VOL}	Large-Signal Voltage Gain	$V_{OUT} = \pm 10\text{V}$, $R_L \geq 2\text{k}\Omega$	●	95	340		80	300		V/mV
CMRR	Common-Mode Rejection Ratio	$V_{CM} = \pm 10\text{V}$	●	80	98		76	96		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 10\text{V}$ to $\pm 17\text{V}$	●	83	100		78	98		dB
	Input Voltage Range		●	± 10	± 10.6		± 10	± 10.6		V
V_{OUT}	Output Voltage Swing	$R_L = 2\text{k}\Omega$	●	± 11.3	± 12.2		± 11.3	± 12.2		V
SR	Slew Rate	$A_V = -1$	●	45	60		35	60		$\text{V}/\mu\text{s}$

TYPICAL PERFORMANCE CHARACTERISTICS

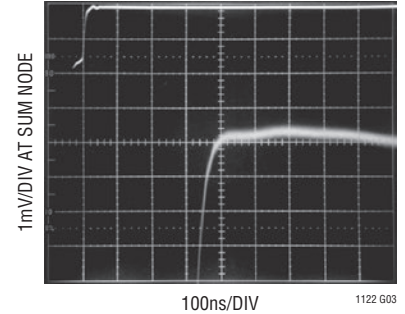
Settling Time
(Input from -10V to 0V)



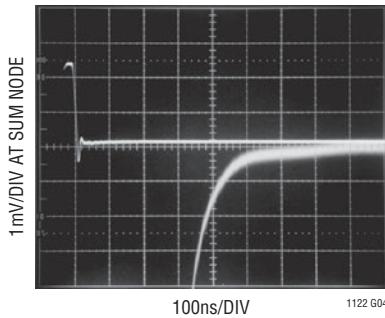
Settling Time
(Input from 10V to 0V)



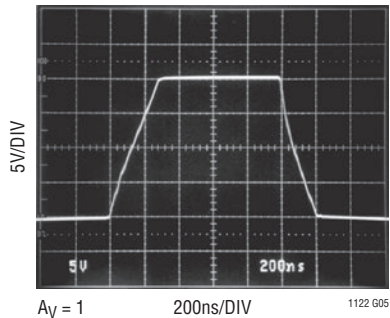
Settling Time
(Input from 0V to 10V)



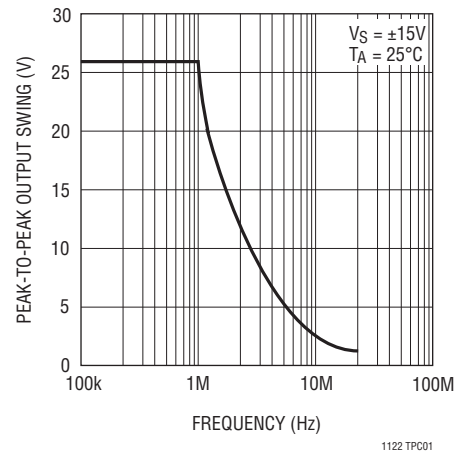
Settling Time
(Input from 0V to -10V)



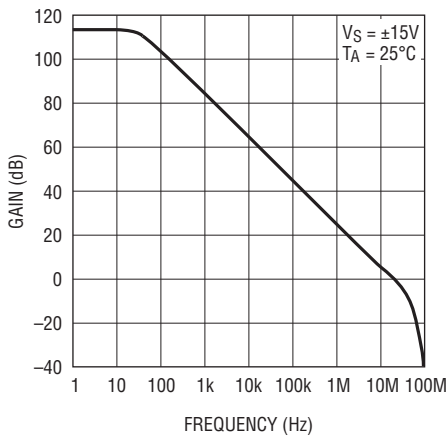
Large-Signal Response



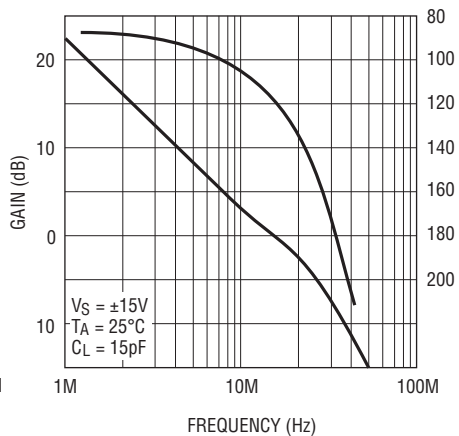
Undistorted Output Swing vs Frequency



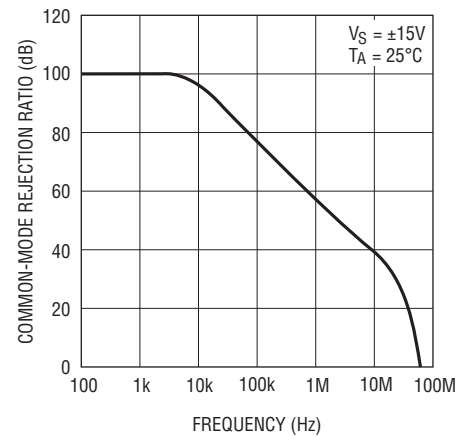
Voltage Gain vs Frequency



Gain, Phase vs Frequency

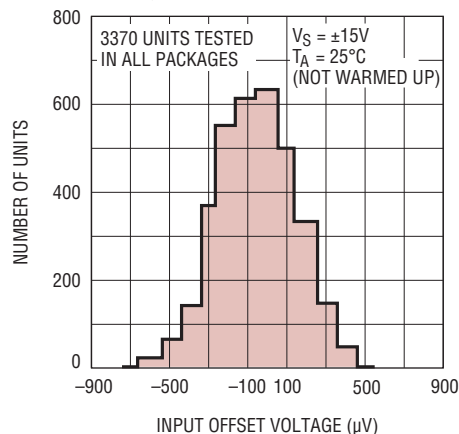


Common-Mode Rejection vs Frequency



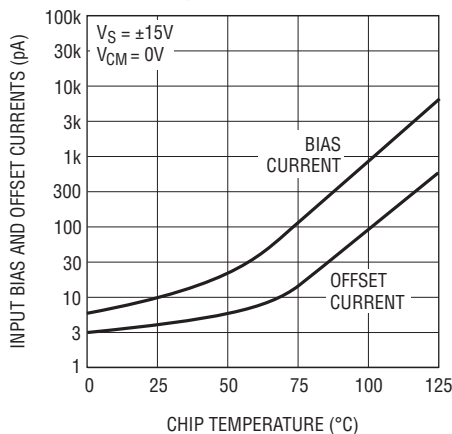
TYPICAL PERFORMANCE CHARACTERISTICS

Distribution of Input Offset Voltage



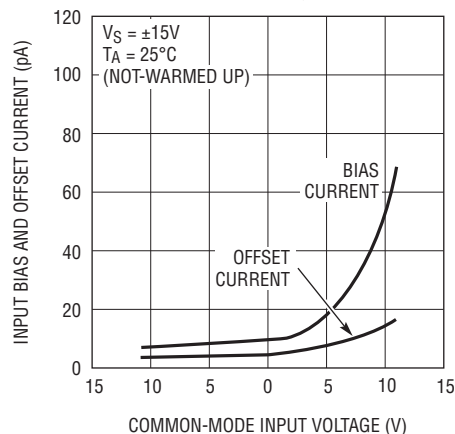
1122 TPC05

Input Bias and Offset Currents Over Temperature



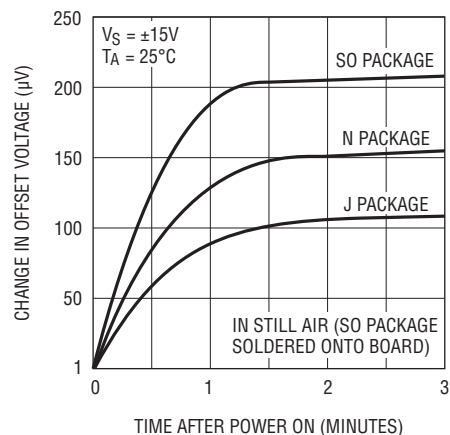
1122 TPC06

Bias and Offset Currents Over the Common-Mode Range



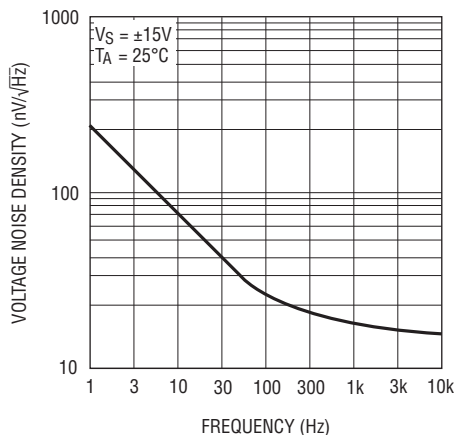
1122 TPC07

Warm-Up Drift



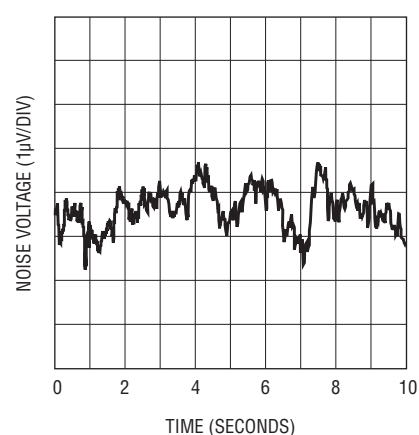
1122 TPC08

Noise Spectrum



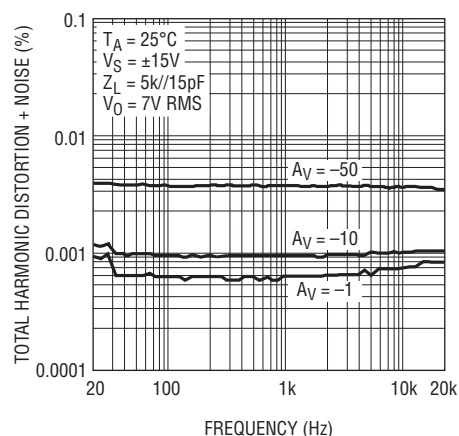
1122 TPC09

0.1Hz to 10Hz Noise



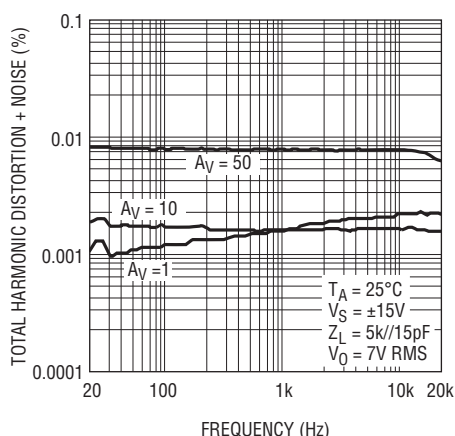
1122 TPC10

Total Harmonic Distortion + Noise vs Frequency Inverting Gain



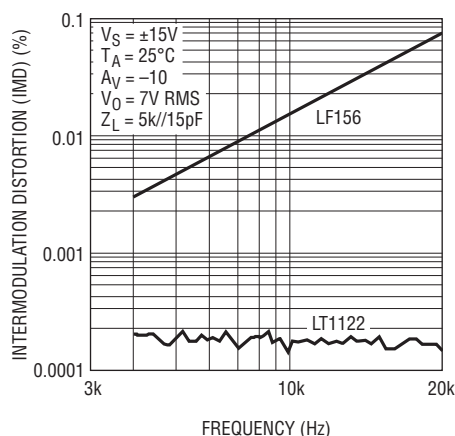
1122 TPC11

Total Harmonic Distortion + Noise vs Frequency Noninverting Gain



1122 TPC12

Intermodulation Distortion (CCIF Method) vs Frequency LT1122 and LF156*



*SEE LT1115 DATA SHEET FOR DEFINITION OF CCIF TESTING

1122 TPC13
1122fb

APPLICATIONS INFORMATION

Settling Time Measurements

Settling time test circuits shown on some competitive devices' data sheets require:

1. A "flat top" pulse generator. Unfortunately, flat top pulse generators are not commercially available.
2. A variable feedback capacitor around the device under test. This capacitor varies over a four-to-one range. Presumably, as each op amp is measured for settling time, the capacitor is fine tuned to optimize settling time for that particular device.
3. A small inductor load to optimize settling.

The LT1122's settling time is 100% tested in the test circuit shown. No "flat top" pulse generator is required. The test circuit can be readily constructed, using commercially available ICs. Of course, standard high frequency board construction techniques should be followed. All LT1122s are measured with a constant feedback capacitor. No fine tuning is required.

Speed Boost/Overcompensation Terminal

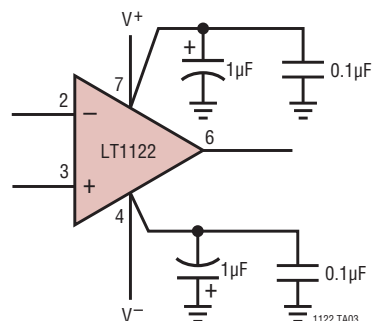
Pin 8 of the LT1122 can be used to change the input stage operating current of the device. Shorting Pin 8 to the positive supply (Pin 7) increases slew rate and bandwidth by about 25%, but at the expense of a reduction in phase margin by approximately 18 degrees. Unity-gain capacitive load handling decreases from typically 500pF to 100pF.

Conversely, connecting a 15k resistor from Pin 8 to ground pulls 1mA out of Pin 8 (with $V^+ = 15V$). This reduces slew rate and bandwidth by 25%. Phase margin and capacitive load handling improve; the latter typically increasing to 800pF.

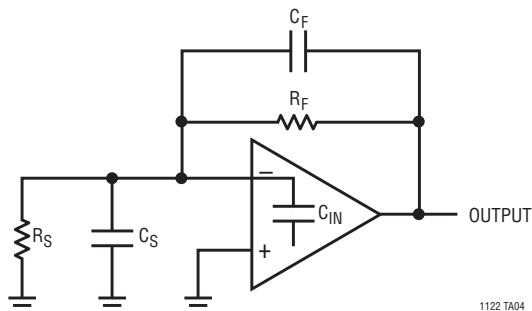
High Speed Operation

As with most high speed amplifiers, care should be taken with supply decoupling, lead dress and component placement.

The power supply connections to the LT1122 must maintain a low impedance to ground over a bandwidth of 20MHz. This is especially important when driving a significant resistive or capacitive load, since all current delivered to the load comes from the power supplies. Multiple high quality bypass capacitors are recommended for each power supply line in any critical application. A 0.1μF ceramic and a 1μF electrolytic capacitor, as shown, placed as close as possible to the amplifier (with short lead lengths to power supply common) will assure adequate high frequency bypassing, in most applications.



When the feedback around the op amp is resistive (R_F), a pole will be created with R_F , the source resistance and capacitance (R_S , C_S), and the amplifier input capacitance ($C_{IN} \approx 4pF$). In low closed-loop gain configurations and with R_S and R_F in the kilohm range, this pole can create excess phase shift and even oscillation. A small capacitor (C_F) in parallel with R_F eliminates this problem. With $R_S (C_S + C_{IN}) = R_F C_F$, the effect of the feedback pole is completely removed.

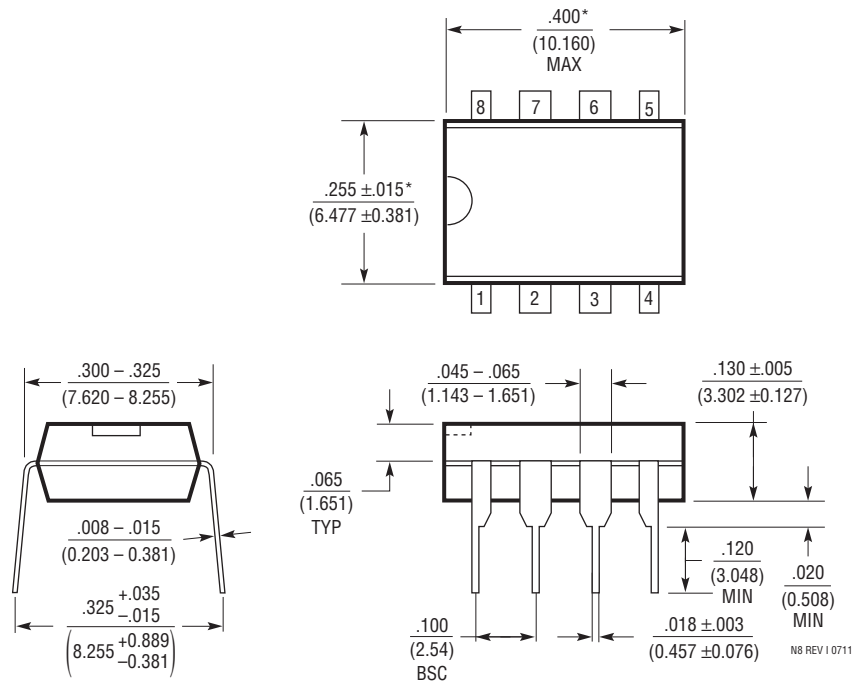


The circuit diagram shows a 100W audio amplifier. The input stage uses an LT1122 op-amp configured as a voltage follower, with a 4kHz J_{CUT} filter and a 47k resistor. The signal is then amplified by an LT1010 op-amp. A distortion trimmer (50k) is used to adjust the output. The output stage uses an LT1006 op-amp configured as a voltage follower, with a 15V supply and a 10μF capacitor. The output is connected to a 470W load. The circuit also includes a 15V supply, a 1M resistor, a 560k resistor, a 100k resistor, and a 2N3904 transistor (Q1). A ground crystal case is shown, containing a VACTEC VTL5C10 or CLAIREX CLM410 crystal. The output is labeled OUTPUT.

PACKAGE DESCRIPTION

Please refer to <http://www.linear.com/designtools/packaging/> for the most recent package drawings.

N Package 8-Lead PDIP (Narrow .300 Inch) (Reference LTC DWG # 05-08-1510 Rev I)



NOTE:

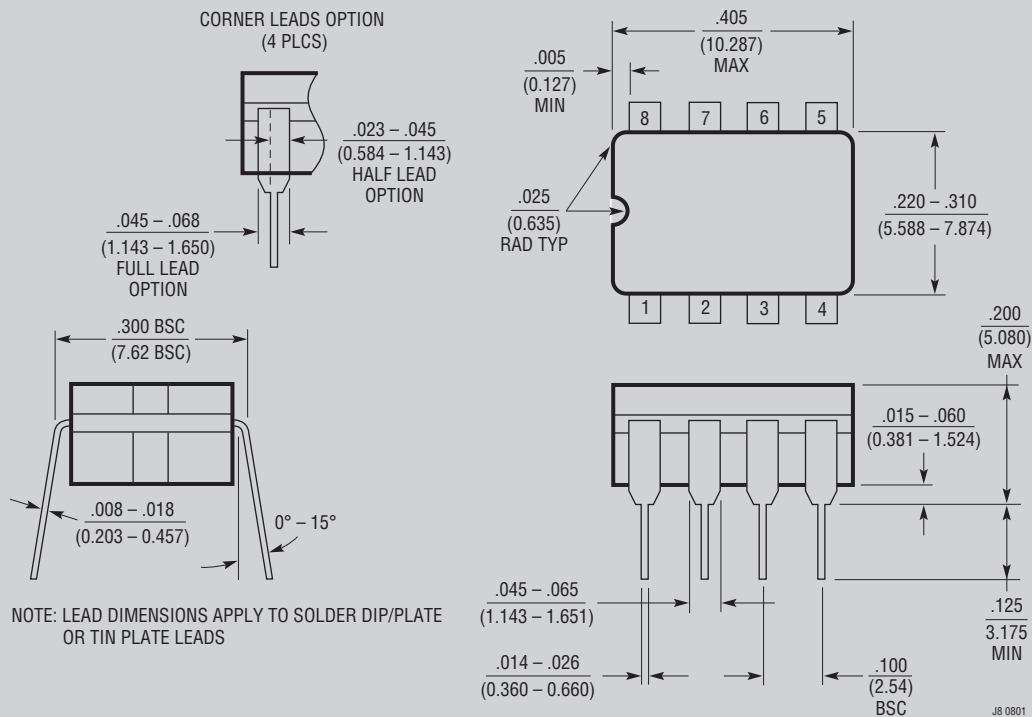
1. DIMENSIONS ARE $\frac{\text{INCHES}}{\text{MILLIMETERS}}$

*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .010 INCH (0.254mm)

PACKAGE DESCRIPTION

Please refer to <http://www.linear.com/designtools/packaging/> for the most recent package drawings.

J8 Package 3-Lead Cerdip (Narrow .300 Inch, Hermetic) (Reference LTC DWG # 05-08-1110)

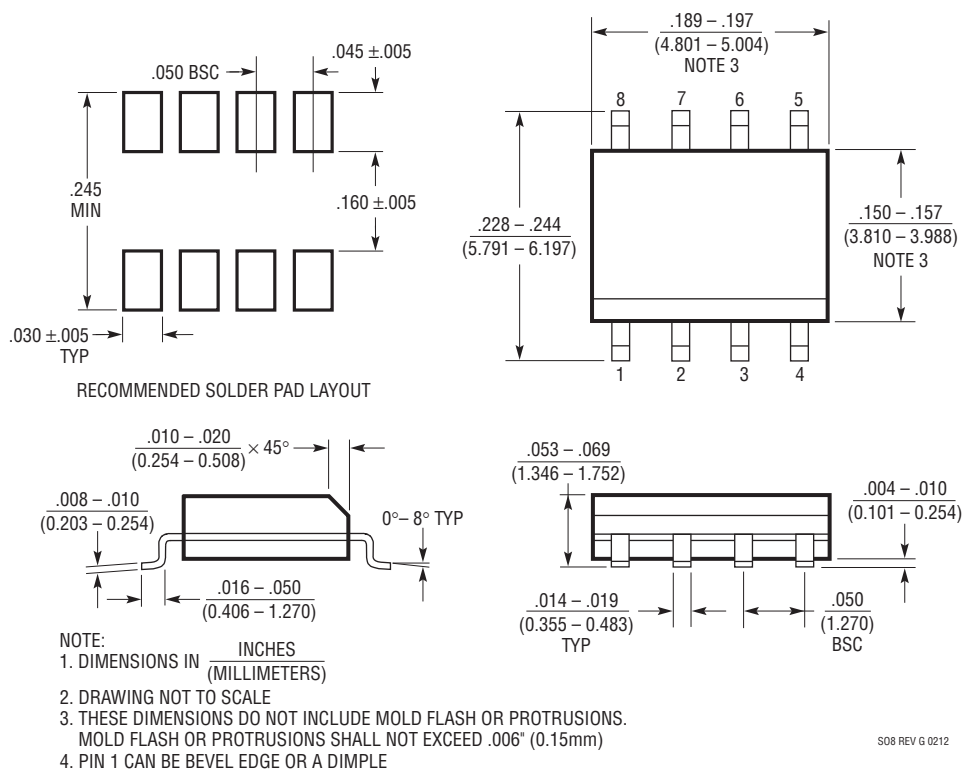


OBSOLETE PACKAGE

PACKAGE DESCRIPTION

Please refer to <http://www.linear.com/designtools/packaging/> for the most recent package drawings.

S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610 Rev G)

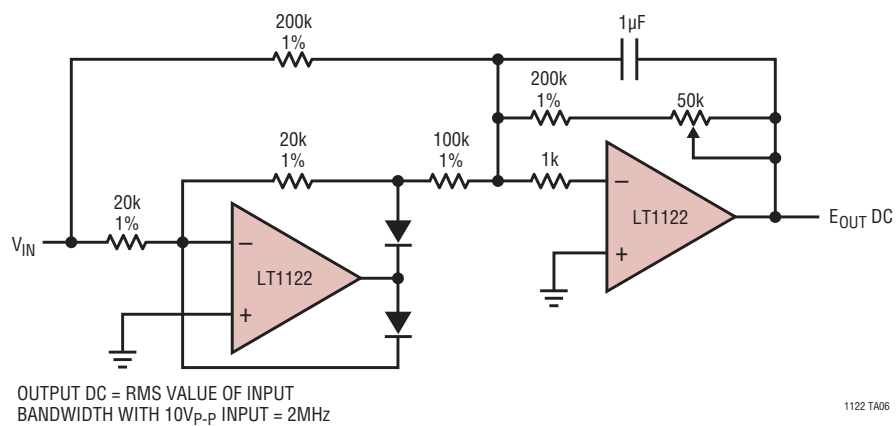


REVISION HISTORY (Revision history begins at Rev B)

REV	DATE	DESCRIPTION	PAGE NUMBER
B	02/14	Updated data sheet to current standards. New Order Information Table, Package Descriptions	2, 10-12

TYPICAL APPLICATION

Wide-Band, Filtered, Full Wave Rectifier



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1022	High Speed Precision JFET Op Amp	23V/µs Min Slew Rate, 250µV V _{OS}
LT1055/LT1056	Precision High Speed JFET Op Amps	16V/µs Slew Rate, 150µV V _{OS}
LT1464	1MHz C-Load™ Stable JFET Op Amp	Capacitive Loads Up to 10nF
LTC®6244	50MHz Low Noise CMOS Op Amp	1pA I _B , 100µV Max V _{OS} , 1.5µV _{P-P} , 0.1Hz to 10Hz Noise

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