#### TELECOMMUNICATION SYSTEM SECONDARY PROTECTION

#### Ion-Implanted Breakdown Region Precise and Stable Voltage Low Voltage Overshoot under Surge

DEVICE	V <sub>DRM</sub>	V <sub>(BO)</sub>		
DEVICE	٧	٧		
'4072F3	58	72		
'4082F3	66	82		

# Planar Passivated Junctions Low Off-State Current < 10 μA</li>

#### • Rated for International Surge Wave Shapes

WAVE SHAPE	STANDARD	I <sub>TSP</sub> A
2/10 µs	FCC Part 68	80
8/20 µs	ANSI C62.41	70
10/160 µs	FCC Part 68	60
10/560 µs	FCC Part 68	45
0.5/700 μs	RLM 88	38
	FTZ R12	50
10/700 μs	VDE 0433	50
	CCITT IX K17/K20	50
10/1000 μs	REA PE-60	35

#### Surface Mount and Through-Hole Options

PACKAGE	PART # SUFFIX
Small-outline	D
Small-outline taped	DR
and reeled	DIC
Single-in-line	SL

#### • UL Recognized, E132482

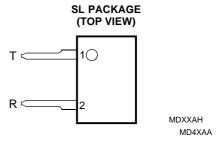
#### description

These low voltage symmetrical transient voltage suppressor devices are designed to protect two wire telecommunication applications against transients caused by lightning strikes and a.c. power lines. Offered in two voltage variants to meet battery and protection requirements they are guaranteed to suppress and withstand the listed international lightning surges in both polarities.

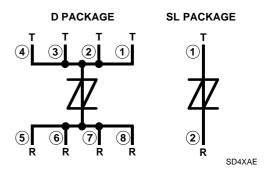
Transients are initially clipped by breakdown clamping until the voltage rises to the breakover level, which causes the device to crowbar. The high crowbar holding current prevents d.c. latchup as the current subsides.

#### 

Specified ratings require the connection of pins 1, 2, 3 and 4 for the T terminal.



#### device symbol



Terminals T and R correspond to the alternative line designators of A and B

These monolithic protection devices are fabricated in ion-implanted planar structures to ensure precise and matched breakover control and are virtually transparent to the system in normal operation

The small-outline 8-pin assignment has been carefully chosen for the TISP series to maximise the inter-pin clearance and creepage distances which are used by standards (e.g. IEC950) to establish voltage withstand ratings.



## TISP4072F3, TISP4082F3 SYMMETRICAL TRANSIENT VOLTAGE SUPPRESSORS

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#### absolute maximum ratings

RATING	SYMBOL	VALUE	UNIT		
Repetitive peak off-state voltage (0°C < T <sub>1</sub> < 70°C)	'4072F3	V	± 58	V	
Repetitive peak off-state voltage (0 C < 1) < 70 C)	'4082F3	$V_{DRM}$	± 66	V	
Non-repetitive peak on-state pulse current (see Notes 1, 2 and 3)	•				
1/2 μs (Gas tube differential transient, open-circuit voltage wave shape	1/2 µs)		120		
2/10 μs (FCC Part 68, open-circuit voltage wave shape 2/10 μs)			80		
8/20 μs (ANSI C62.41, open-circuit voltage wave shape 1.2/50 μs)			70		
10/160 μs (FCC Part 68, open-circuit voltage wave shape 10/160 μs)			60		
5/200 μs (VDE 0433, open-circuit voltage wave shape 2 kV, 10/700 μs)			50	Α	
0.2/310 μs (RLM 88, open-circuit voltage wave shape 1.5 kV, 0.5/700 μs)			38		
5/310 μs (CCITT IX K17/K20, open-circuit voltage wave shape 2 kV, 10/700 μs)			50		
5/310 µs (FTZ R12, open-circuit voltage wave shape 2 kV, 10/700 µs)			50		
10/560 µs (FCC Part 68, open-circuit voltage wave shape 10/560 µs)			45		
10/1000 μs (REA PE-60, open-circuit voltage wave shape 10/1000 μs)		35			
Non-repetitive peak on-state current (see Notes 2 and 3)  D Package			4	Λ κων ο	
50 Hz, 1 s	I <sub>TSM</sub>	6	A rms		
Initial rate of rise of on-state current, Linear current ramp, Maximum ramp value < 38 A			250	A/µs	
Junction temperature	TJ	-40 to +150	°C		
Storage temperature range	T <sub>stg</sub>	-40 to +150	°C		

- NOTES: 1. Further details on surge wave shapes are contained in the Applications Information section.
  - 2. Initially the TISP must be in thermal equilibrium with 0°C < T<sub>J</sub> <70°C. The surge may be repeated after the TISP returns to its initial conditions.
  - 3. Above 70°C, derate linearly to zero at 150°C lead temperature.

## electrical characteristics for the T and R terminals, $T_J = 25$ °C

				TISP4072F3		TISP4082F3				
	PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	MIN	TYP	MAX	UNIT
I <sub>DRM</sub>	Repetitive peak off- state current	$V_D = \pm V_{DRM}, 0^{\circ}C < T_J < 70^{\circ}C$				±10			±10	μА
V <sub>(BO)</sub>	Breakover voltage	$dv/dt = \pm 250 \text{ V/ms},  R_{SOURCE} = 300 \Omega$				±72			±82	V
V <sub>(BO)</sub>	Impulse breakover voltage	$dv/dt = \pm 1000 \text{ V/}\mu\text{s},  R_{SOURCE}$ $di/dt < 20 \text{ A/}\mu\text{s}$	= 50 Ω,		±86			±96		V
I <sub>(BO)</sub>	Breakover current	$dv/dt = \pm 250 \text{ V/ms},  R_{SOURCE} =$	= 300 Ω	±0.15		±0.6	±0.15		±0.6	Α
V <sub>T</sub>	On-state voltage	$I_T = \pm 5 \text{ A},  t_W = 100  \mu\text{s}$				±3			±3	V
I <sub>H</sub>	Holding current	di/dt = +/-30 mA/ms		±0.15			±0.15			Α
dv/dt	Critical rate of rise of off-state voltage	Linear voltage ramp  Maximum ramp value < 0.85V <sub>(BR)MIN</sub>		±5			±5			kV/μs
I <sub>D</sub>	Off-state current	$V_D = \pm 50 \text{ V}$				±10			±10	μΑ
		f = 100 kHz, V <sub>d</sub> = 100 mV	$V_{D} = 0,$		82	140		82	140	pF
C <sub>off</sub>	Off-state capacitance	(see Note 4)	$V_D = -5 V$		49	85		49	85	pF
		(222.1212.1)	$V_{D} = -50 \text{ V}$		25	40		25	40	pF

NOTE 4: Further details on capacitance are given in the Applications Information section.

#### thermal characteristics

	PARAMETER	TEST CONDITION	MIN	TYP	MAX	UNIT	
R	Junction to free air thermal resistance	$P_{tot} = 0.8 \text{ W}, T_A = 25^{\circ}\text{C}$	D Package			160	°C/W
R <sub>θ</sub> JA	difficient to free all thermal resistance	5 cm <sup>2</sup> , FR4 PCB	SL Package			105	0, 11

#### PARAMETER MEASUREMENT INFORMATION

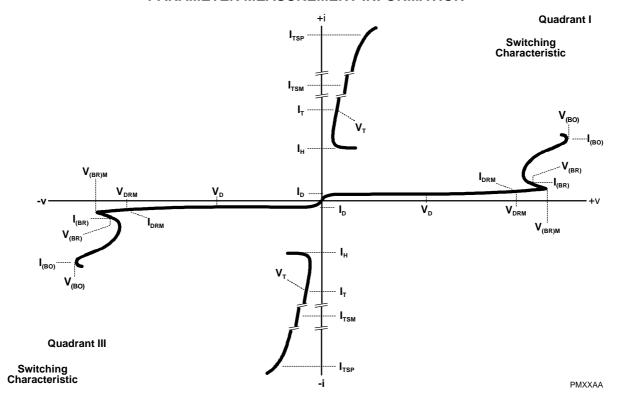
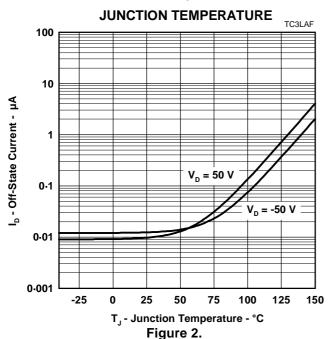


Figure 1. VOLTAGE-CURRENT CHARACTERISTIC FOR T AND R TERMINALS ALL MEASUREMENTS ARE REFERENCED TO THE R TERMINAL

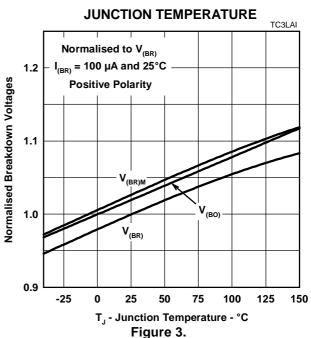


#### **TYPICAL CHARACTERISTICS** R and T terminals

#### **OFF-STATE CURRENT**

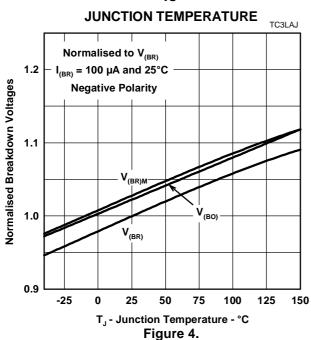


## NORMALISED BREAKDOWN VOLTAGES



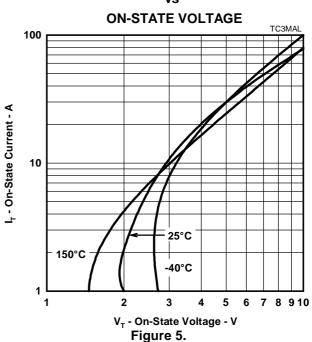
## NORMALISED BREAKDOWN VOLTAGES

vs



#### **ON-STATE CURRENT**

vs

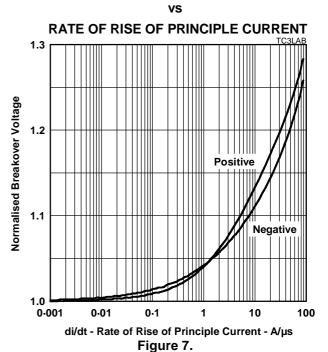


# TYPICAL CHARACTERISTICS R and T terminals

#### **HOLDING CURRENT & BREAKOVER CURRENT**

vs **JUNCTION TEMPERATURE** TC3LAH 1.0 I<sub>H</sub>, I<sub>(BO)</sub> - Holding Current, Breakover Current - A 0.9 0.8 0.7 0.6 0.5 I<sub>(BO)</sub> 0.4 0.3 IH 0.2 0.1 -25 25 50 75 100 125 150 T<sub>J</sub> - Junction Temperature - °C Figure 6.

## NORMALISED BREAKOVER VOLTAGE



## OFF-STATE CAPACITANCE

TERMINAL VOLTAGE

TC3LAE

Positive Bias

Negative Bias

Negative Bias

Terminal Voltage - V

Figure 8.

## OFF-STATE CAPACITANCE

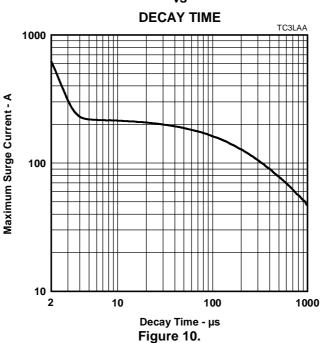
vs **JUNCTION TEMPERATURE** TC3LAD 500 Off-State Capacitance - pF 100 Terminal Bias = 0 Terminal Bias = 50 V Terminal Bias = -50 V 10 -25 100 125 150 T<sub>J</sub> - Junction Temperature - °C Figure 9.



# TYPICAL CHARACTERISTICS R and T terminals

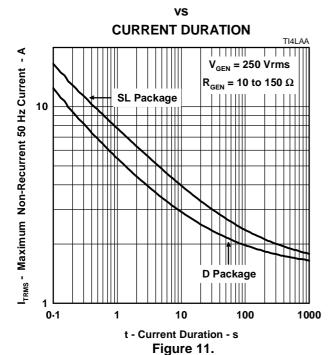
#### **SURGE CURRENT**

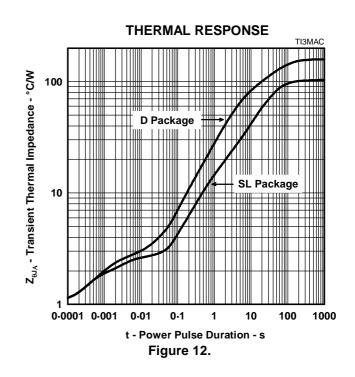
vs



#### THERMAL INFORMATION

#### **MAXIMUM NON-RECURRING 50 Hz CURRENT**





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#### APPLICATIONS INFORMATION

#### electrical characteristics

The electrical characteristics of a TISP are strongly dependent on junction temperature,  $T_J$ . Hence a characteristic value will depend on the junction temperature at the instant of measurement. The values given in this data sheet were measured on commercial testers, which generally minimise the temperature rise caused by testing. Application values may be calculated from the parameters' temperature curves, the power dissipated and the thermal response curve ( $Z_\theta$ ).

#### lightning surge

#### wave shape notation

Most lightning tests, used for equipment verification, specify a unidirectional sawtooth waveform which has an exponential rise and an exponential decay. Wave shapes are classified in terms of peak amplitude (voltage or current), rise time and a decay time to 50% of the maximum amplitude. The notation used for the wave shape is *amplitude*, *rise time/decay time*. A 50A,  $5/310 \, \mu s$  wave shape would have a peak current value of 50 A, a rise time of 5  $\mu s$  and a decay time of 310  $\mu s$ . The TISP surge current graph comprehends the wave shapes of commonly used surges.

#### generators

There are three categories of surge generator type, single wave shape, combination wave shape and circuit defined. Single wave shape generators have essentially the same wave shape for the open circuit voltage and short circuit current (e.g. 10/1000 µs open circuit voltage and short circuit current). Combination generators have two wave shapes, one for the open circuit voltage and the other for the short circuit current (e.g. 1.2/50 µs open circuit voltage and 8/20 µs short circuit current) Circuit specified generators usually equate to a combination generator, although typically only the open circuit voltage waveshape is referenced (e.g. a 10/700 µs open circuit voltage generator typically produces a 5/310 µs short circuit current). If the combination or circuit defined generators operate into a finite resistance the wave shape produced is intermediate between the open circuit and short circuit values.

#### current rating

When the TISP switches into the on-state it has a very low impedance. As a result, although the surge wave shape may be defined in terms of open circuit voltage, it is the current wave shape that must be used to assess the required TISP surge capability. As an example, the CCITT IX K17 1.5 kV,  $10/700~\mu s$  surge is changed to a 38 A,  $5/310~\mu s$  waveshape when driving into a short circuit. Thus the TISP surge current capability, when directly connected to the generator, will be found for the CCITT IX K17 waveform at 310  $\mu s$  on the surge graph and not 700  $\mu s$ . Some common short circuit equivalents are tabulated below:

STANDARD	OPEN CIRCUIT	SHORT CIRCUIT
	VOLTAGE	CURRENT
COLTT IV IVAZ	4.5.14/. 40/700	20 1 5/240
CCITT IX K17	1.5 kV, 10/700 µs	38 A, 5/310 μs
CCITT IX K20	1 kV, 10/700 μs	25 A, 5/310 µs
RLM88	1.5 kV, 0.5/700 μs	38 A, 0.2/310 μs
VDE 0433	2.0 kV, 10/700 µs	50 A, 5/200 μs
FTZ R12	2.0 kV, 10/700 µs	50 A, 5/310 μs

Any series resistance in the protected equipment will reduce the peak circuit current to less than the generators' short circuit value. A 2 kV open circuit voltage, 50 A short circuit current generator has an effective output impedance of 40  $\Omega$  (2000/50). If the equipment has a series resistance of 25  $\Omega$  then the surge current requirement of the TISP becomes 31 A (2000/65) and not 50 A.



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#### APPLICATIONS INFORMATION

#### protection voltage

The protection voltage,  $(V_{(BO)})$ , increases under lightning surge conditions due to thyristor regeneration. This increase is dependent on the rate of current rise, di/dt, when the TISP is clamping the voltage in its breakdown region. The  $V_{(BO)}$  value under surge conditions can be estimated by multiplying the 50 Hz rate  $V_{(BO)}$  (250 V/ms) value by the normalised increase at the surge's di/dt (Figure 7.) . An estimate of the di/dt can be made from the surge generator voltage rate of rise, dv/dt, and the circuit resistance.

As an example, the CCITT IX K17 1.5 kV,  $10/700~\mu s$  surge has an average dv/dt of 150 V/ $\mu s$ , but, as the rise is exponential, the initial dv/dt is higher, being in the region of 450 V/ $\mu s$ . The instantaneous generator output resistance is 25  $\Omega$ . If the equipment has an additional series resistance of 20  $\Omega$ , the total series resistance becomes 45  $\Omega$ . The maximum di/dt then can be estimated as  $450/45 = 10~A/\mu s$ . In practice the measured di/dt and protection voltage increase will be lower due to inductive effects and the finite slope resistance of the TISP breakdown region.

#### capacitance

#### off-state capacitance

The off-state capacitance of a TISP is sensitive to junction temperature,  $T_J$ , and the bias voltage, comprising of the dc voltage,  $V_D$ , and the ac voltage,  $V_d$ . All the capacitance values in this data sheet are measured with an ac voltage of 100 mV. The typical 25°C variation of capacitance value with ac bias is shown in Figure 13 When  $V_D >> V_d$  the capacitance value is independent on the value of  $V_d$ . The capacitance is essentially constant over the range of normal telecommunication frequencies.

### NORMALISED CAPACITANCE

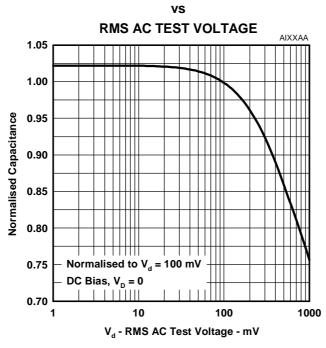


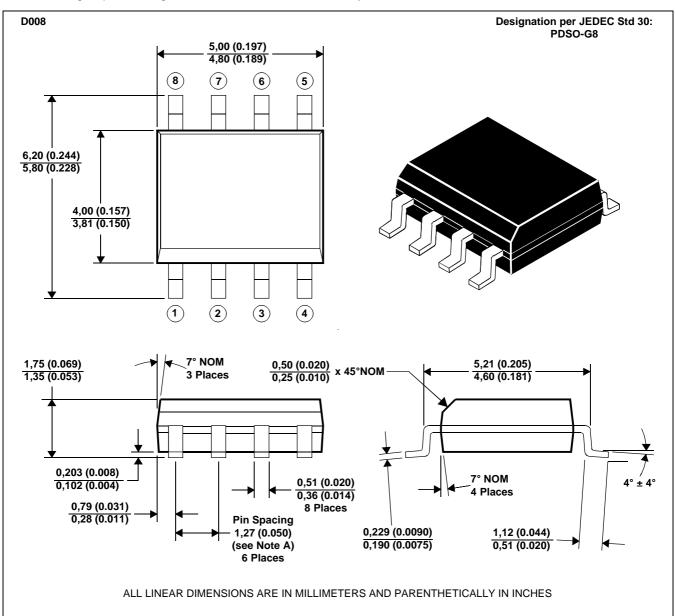
Figure 13.

#### **MECHANICAL DATA**

#### **D008**

#### plastic small-outline package

This small-outline package consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.



MDXXAA

- NOTES: A. Leads are within 0,25 (0.010) radius of true position at maximum material condition.
  - B. Body dimensions do not include mold flash or protrusion.
  - C. Mold flash or protrusion shall not exceed 0,15 (0.006).
  - D. Lead tips to be planar within  $\pm 0,051$  (0.002).

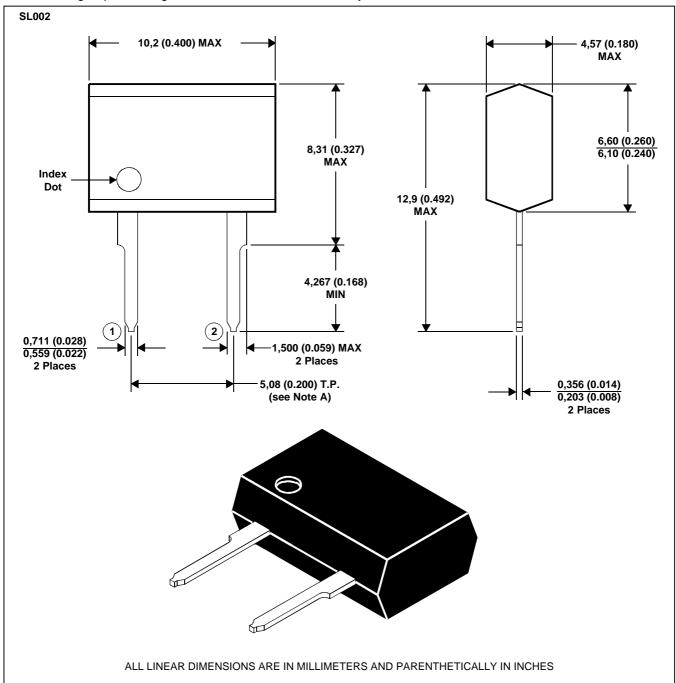


#### **MECHANICAL DATA**

#### **SL002**

#### 2-pin plastic single-in-line package

This single-in-line package consists of a circuit mounted on a lead frame and encapsulated within a plastic compound. The compound will withstand soldering temperature with no deformation, and circuit performance characteristics will remain stable when operated in high humidity conditions. Leads require no additional cleaning or processing when used in soldered assembly.



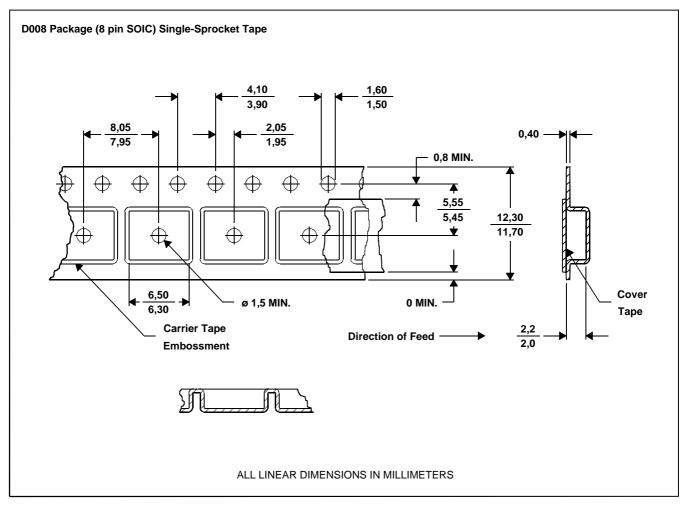
NOTES: A. Each pin centerline is located within 0,25 (0.010) of its true longitudinal position.

B. Body molding flash of up to 0,15 (0.006) may occur in the package lead plane.

MDXXAC

#### **MECHANICAL DATA**

# D008 tape dimensions



NOTES: A. Taped devices are supplied on a reel of the following dimensions:-

MDXXAT

Reel diameter: 330 +0.0/-4.0 mmReel hub diameter:  $100 \pm 2.0 \text{ mm}$ Reel axial hole:  $13.0 \pm 0.2 \text{ mm}$ 

B. 2500 devices are on a reel.



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