

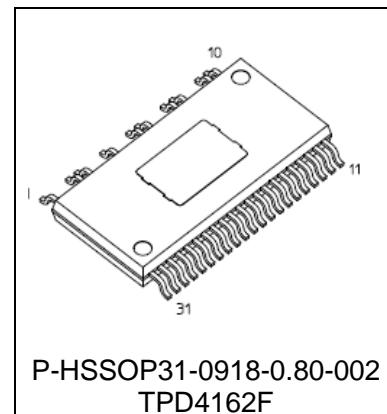
Intelligent Power Device

# TPD4162F

High Voltage Monolithic Silicon Power IC

## 1. Description

The TPD4162F is a DC brushless motor driver using high-voltage PWM control. It is fabricated using a high-voltage SOI process. The device contains PWM circuit, 3-phase decode circuit, level shift high-side driver, low-side driver, IGBT outputs, FRDs, over-current, and current limit and under-voltage protection circuits, and a thermal shutdown circuit. It is easy to control a DC brushless motor by applying a signal from a motor controller and a Hall element / Hall IC to the TPD4162F.



## 2. Applications

DC brushless motor driver IC

## 3. Features

- High voltage power side and low voltage signal side terminal are separated.
- Bootstrap circuits give simple high-side supply.
- Bootstrap diodes are built in.
- PWM and 3-phase decode circuit are built in.
- Pulses-per-revolution output:
  - FGC = High: 3 pulse/electrical angle: 360°
  - FGC = Low : 1 pulse/electrical angle: 360°
- Included over-current and current limit and under-voltage protection, and thermal shutdown.

Start of commercial production  
2019-10

## 4. Block Diagram

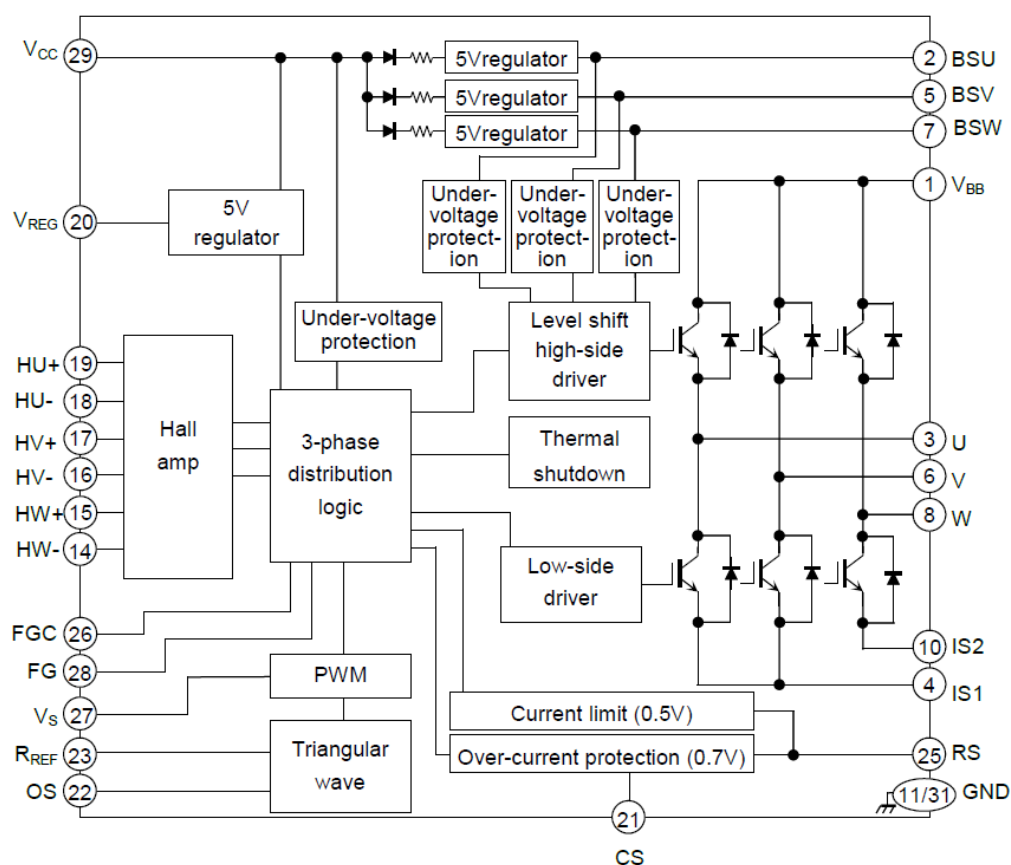


Figure 4.1 Block Diagram

## 5. Pin Assignments

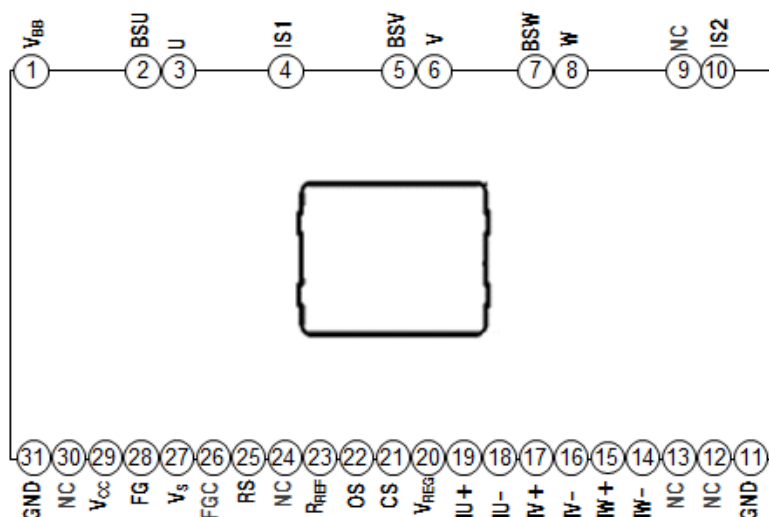


Figure 5.1 Pin Assignments

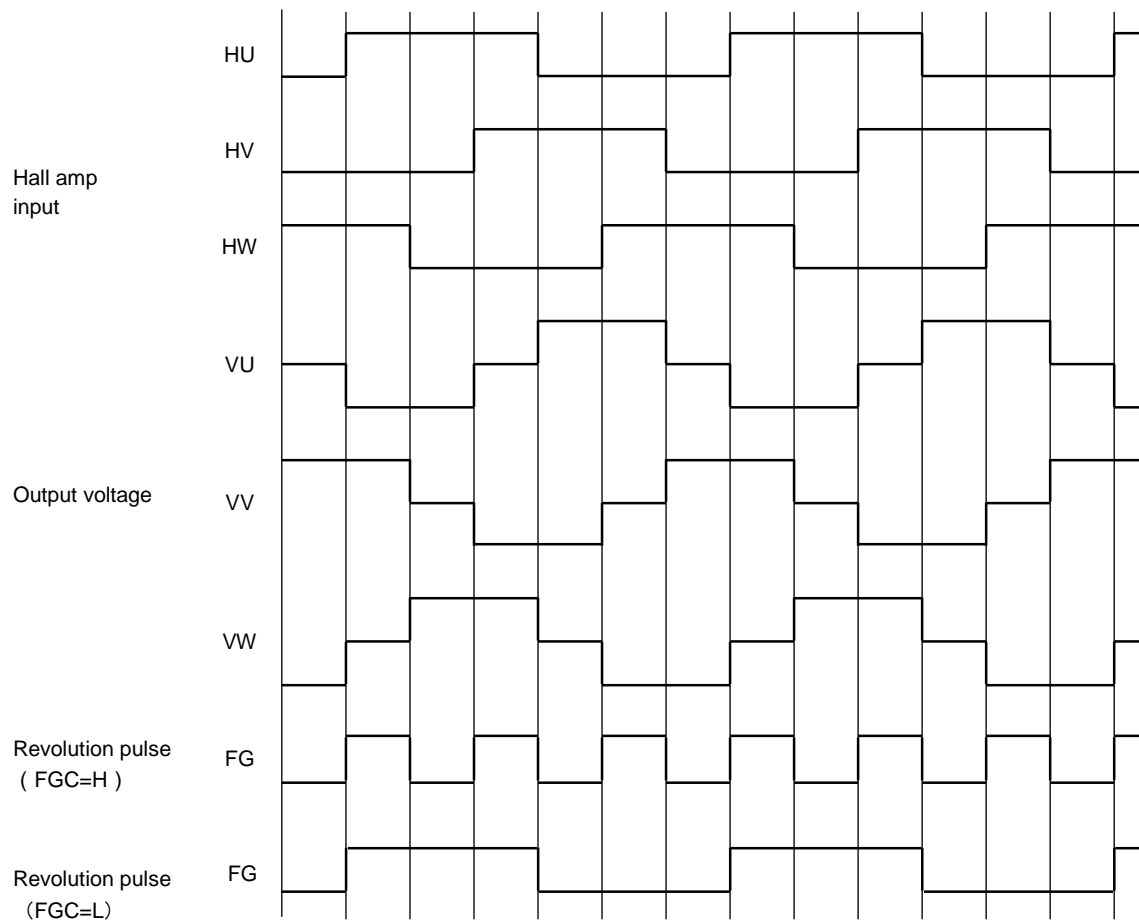
## 6. Pin Description

Table 6.1 Pin Description

Pin No.	Symbol	Pin Description
1	V <sub>BB</sub>	High-voltage power supply input pin.
2	BSU	U-phase bootstrap capacitor connecting pin.
3	U	U-phase output pin.
4	IS1	IGBT emitter/FRD anode pin.
5	BSV	V-phase bootstrap capacitor connecting pin.
6	V	V-phase output pin.
7	BSW	W-phase bootstrap capacitor connecting pin.
8	W	W-phase output pin.
9	NC	Unused pin, which is not connected to the chip internally.
10	IS2	IGBT emitter/FRD anode pin.
11	GND	Ground pin.
12	NC	Unused pin, which is not connected to the chip internally.
13	NC	Unused pin, which is not connected to the chip internally.
14	HW-	W-phase Hall element signal input pin. (Hall IC can be used.)
15	HW+	W-phase Hall element signal input pin. (Hall IC can be used.)
16	HV-	V-phase Hall element signal input pin. (Hall IC can be used.)
17	HV+	V-phase Hall element signal input pin. (Hall IC can be used.)
18	HU-	U-phase Hall element signal input pin. (Hall IC can be used.)
19	HU+	U-phase Hall element signal input pin. (Hall IC can be used.)
20	V <sub>REG</sub>	5 V regulator output pin.
21	CS	Over-current protection detection pin
22	OS	PWM triangular wave oscillation frequency setup pin. (Connect a capacitor to this pin.)
23	R <sub>REF</sub>	PWM triangular wave oscillation frequency setup pin. (Connect a resistor to this pin.)
24	NC	Unused pin, which is not connected to the chip internally.
25	RS	Input for current limit and over-current protection detection pin
26	FGC	FG pulse count select (High or open = 3 ppr; Low = 1 ppr).
27	V <sub>s</sub>	Speed control signal input pin. (PWM reference voltage input pin.)
28	FG	Revolution pulse output pin.
29	V <sub>CC</sub>	Control power supply pin.
30	NC	Unused pin, which is not connected to the chip internally.
31	GND	Ground pin.

## 7. Functional Description

### 7.1.1. Timing Chart



**Figure7.1.1 Timing Chart**

Note: Hall amp input logic high (H) refers to  $H^+ > H^-$ . (\*: U/V/W)

## 7.1.2. Truth Table

Table 7.1.2 Truth Table

FGC	Hall amp Input			U Phase		V Phase		W Phase		FG
	HU	HV	HW	High side	Low side	High side	Low side	High side	Low side	
H	H	L	H	OFF	ON	ON	OFF	OFF	OFF	H
H	H	L	L	OFF	ON	OFF	OFF	ON	OFF	L
H	H	H	L	OFF	OFF	OFF	ON	ON	OFF	H
H	L	H	L	ON	OFF	OFF	ON	OFF	OFF	L
H	L	H	H	ON	OFF	OFF	OFF	OFF	ON	H
H	L	L	H	OFF	OFF	ON	OFF	OFF	ON	L
H	L	L	L	OFF	OFF	OFF	OFF	OFF	OFF	L
H	H	H	H	OFF	OFF	OFF	OFF	OFF	OFF	L
L	H	L	H	OFF	ON	ON	OFF	OFF	OFF	H
L	H	L	L	OFF	ON	OFF	OFF	ON	OFF	H
L	H	H	L	OFF	OFF	OFF	ON	ON	OFF	H
L	L	H	L	ON	OFF	OFF	ON	OFF	OFF	L
L	L	H	H	ON	OFF	OFF	OFF	OFF	ON	L
L	L	L	H	OFF	OFF	ON	OFF	OFF	ON	L
L	L	L	L	OFF	OFF	OFF	OFF	OFF	OFF	L
L	H	H	H	OFF	OFF	OFF	OFF	OFF	OFF	H

Note: Hall amp input logic high (H) refers to  $H^+ > H^-$ . (\*: U/V/W)

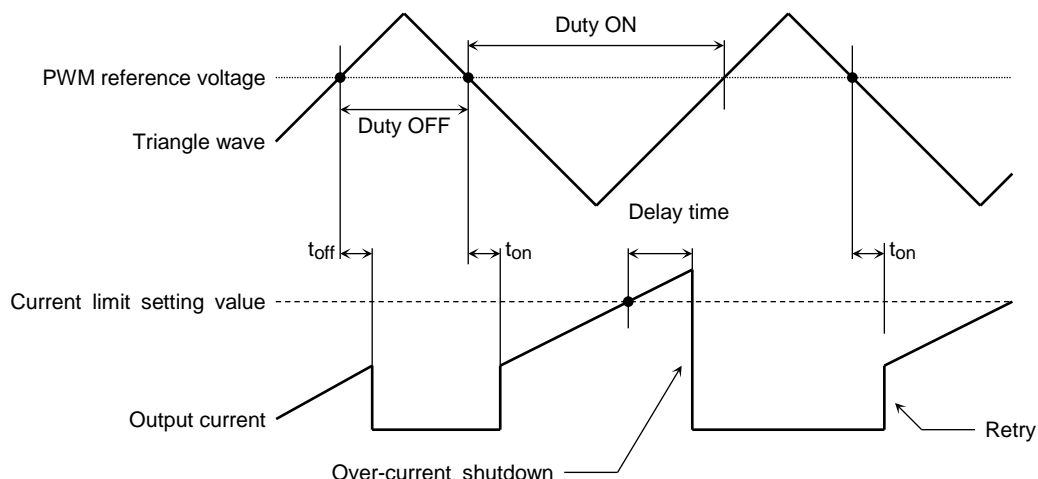
## 7.2. Handling precautions

- (1) When switching the power supply to the circuit on/off, ensure that  $V_S < V_{VS\text{OFF}}$  (all IGBT outputs off). At that time, either the  $V_{CC}$  or the  $V_{BB}$  can be turned on/off first. Note that if the power supply is switched off as described above, the IC may be destroyed if the current regeneration route to the  $V_{BB}$  power supply is blocked when the  $V_{BB}$  line is disconnected by a relay or similar while the motor is still running.
- (2) The triangular wave oscillator circuit, with externally connected  $C_5$  and  $R_3$ , charges and discharges minute amounts of current. Therefore, subjecting the IC to noise when mounting it on the board may distort the triangular wave or cause malfunction. To avoid this, attach external parts to the base of the IC leads or isolate them from any tracks or wiring which carries large current.
- (3) The PWM of this IC is controlled by the on/off state of the high-side IGBT.
- (4) If a motor is locked where  $V_{BB}$  voltage is low and duty is 100 %, it may not be possible to reboot after the load is released as a result of the high side being ON immediately prior to the motor being locked. This is because, over time, the bootstrap voltage falls, the high-side voltage decrease protection operates and the high-side output becomes OFF. In this case, since the level shift pulse necessary to turn the high side ON cannot be generated, reboot is not possible. A level shift pulse is generated by either the edge of a Hall sensor output or the edge of an internal PWM signal, but neither edge is available due to the motor lock and duty 100 % command. In order to reboot after a lock, the high-side power voltage must return to a level 0.5 V (typ.) higher than the voltage decrease protection level, and a high-side input signal must be introduced. As a high-side input signal is created by the aforementioned level shift pulse, it is possible to reboot by reducing PWM duty to less than 100 % or by forcing the motor to turn externally and creating an edge at a Hall sensor output. In order to ensure reboot after a system lock, the maximum duty of motor operation must be less than 100 %.

## 7.3. Description of Protection Function

### (1) Current limit protection

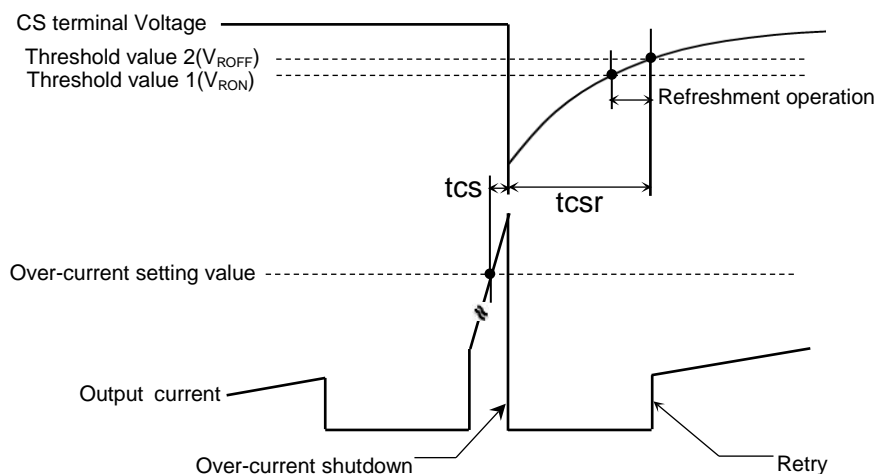
The IC incorporates a current limit protection circuit to protect itself against over current at startup or when a motor is locked. This protection function detects voltage generated in the current-detection resistor connected to the RS pin. When this voltage exceeds  $V_R = 0.5 \text{ V}$  (typ.), the high-side IGBT output, which is on, temporarily shuts down after a delay time, preventing any additional current from flowing to the IC. The next PWM ON signal releases the shutdown state.



**Figure 7.3.1 Description of Current limit protection**

### (2) Over-current protection

This protection function detects voltage generated in the current-detection resistor connected to the RS pin. When this voltage exceeds  $V_R = 0.7 \text{ V}$  (typ.), the all high-side and low-side IGBT outputs are shut down. Over-current protection recovery time can be adjusted with the capacitor and resistance connected to CS terminal. The CS terminal voltage is carried out by the damping time constant decided by this capacitor and resistance. The low side IGBT of the phase which a high side turns on by a timing chart will be made to turn on, a bootstrap capacitor will be charged, if the threshold value 1 ( $V_{RON}$ ) is exceeded (refreshment operation), and the threshold value 2 ( $V_{ROFF}$ ) is exceeded after that, IGBT turns on according to an input signal.



**Figure 7.3.2 Description of Over-current protection**

## (3) Under-voltage protection

The IC incorporates under-voltage protection circuits to prevent the IGBT from operating in unsaturated mode when the  $V_{CC}$  voltage or the  $V_{BS}$  voltage drops.

When the  $V_{CC}$  power supply falls to the IC internal setting  $V_{CCUVD} = 11\text{ V}$  (typ.), all IGBT outputs are shut down regardless of the input. This protection function has hysteresis. When the  $V_{CC}$  power supply reaches  $0.5\text{ V}$  higher than the shutdown voltage  $V_{CCUVR} = 11.5\text{ V}$  (typ.), the IC is automatically restored and the IGBT is turned on/off again by the input.

When the  $V_{BS}$  supply voltage drops  $V_{BSUVD} = 3.0\text{ V}$  (typ.), the high-side IGBT output shuts down. When the  $V_{BS}$  supply voltage reaches  $0.5\text{ V}$  higher than the shutdown voltage  $V_{BSUVR} = 3.5\text{ V}$  (typ.), the IGBT is turned on/off again by the input signal.

## (4) Thermal shutdown

The IC incorporates a thermal shutdown circuit to protect itself against excessive rise in temperature.

When the temperature of this chip rises to the internal setting TSD due to external causes or internal heat generation, all IGBT outputs are shut down regardless of the input. This protection function has hysteresis  $\Delta TSD = 50\text{ }^{\circ}\text{C}$  (typ.). When the chip temperature falls to  $TSD - \Delta TSD$ , the chip is automatically restored and the IGBT is turned on/off again by the input.

Because the chip contains just one temperature-detection location, when the chip heats up due to the IGBT for example, the distance between the detection location and the IGBT (the source of the heat) can cause differences in the time taken for shutdown to occur. Therefore, the temperature of the chip may rise higher than the thermal shutdown temperature.

## 7.4. Description of Bootstrap Capacitor Charging and Its Capacitance

The IC uses bootstrapping for the power supply for high-side drivers.

The bootstrap capacitor is charged by turning on the low-side IGBT of the same arm (approximately 1/5 of PWM cycle) while the high-side IGBT controlled by PWM is off. For example, to drive at  $20\text{ kHz}$ , it takes approximately  $10\text{ }\mu\text{s}$  per cycle to charge the capacitor. When the  $V_S$  voltage exceeds  $3.8\text{ V}$  (55 % duty), the low-side IGBT is continuously in the off state. This is because when the PWM on-duty becomes larger, the arm is short-circuited while the low-side IGBT is on. Even in this state, because PWM control is being performed on the high-side IGBT, the regenerative current of the diode flows to the low-side FRD of the same arm, and the bootstrap capacitor is charged. Note that when the on-duty is 100 %, diode regenerative current does not flow; thus, the bootstrap capacitor is not charged.

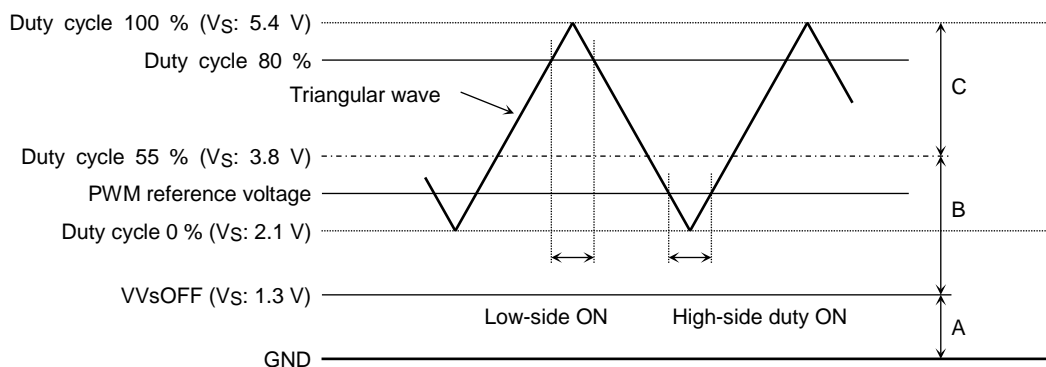
When driving a motor at 100 % duty cycle, take the voltage drop at 100 % duty (see the figure below) into consideration to determine the capacitance of the bootstrap capacitor.

Capacitance of the bootstrap capacitor = Current dissipation (max) of the high-side driver  $\times$  Maximum drive time  $/(V_{CC} - V_F(\text{BSD}) + V_F(\text{FRD}) - 13.5)\text{ [F]}$

$V_F(\text{BSD})$  : Bootstrap diode forward voltage

$V_F(\text{FRD})$  : Fast recovery diode forward voltage

Consideration must be made for aging and temperature change of the capacitor.



$V_S$ Range	IGBT Operation
A	Both high and low-side OFF.
B	Charging range. Low-side IGBT refreshing on the phase the high-side IGBT in PWM.
C	No charging range. High-side at PWM according to the timing chart. Low-side no refreshing.

## 8. Absolute Maximum Ratings

**Table 8.1 Absolute Maximum Ratings**

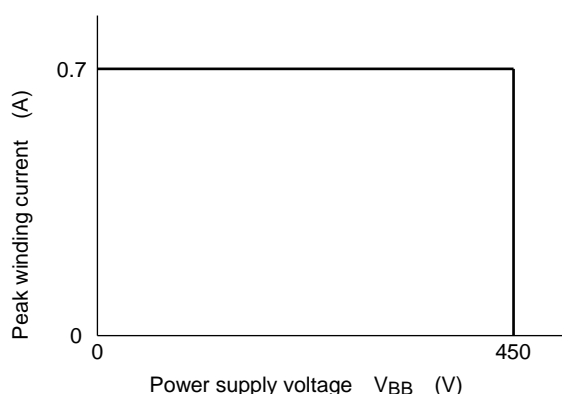
( $T_a = 25^\circ\text{C}$  unless otherwise specified)

Characteristics	Symbol	Rating	Unit
Power supply voltage	$V_{BB}$	600	V
	$V_{CC}$	20	V
Output current (DC)	$I_{out}$	0.7	A
Output current (pulse)	$I_{outp}$	1.2	A
Input voltage (except $V_S$ )	$V_{IN}$	-0.5 to $V_{REG} + 0.5$	V
Input voltage (only $V_S$ )	$V_{V_S}$	8.2	V
$V_{REG}$ current	$I_{REG}$	50	mA
FG voltage	$V_{FG}$	6	V
FG current	$I_{FG}$	20	mA
Power dissipation ( $T_c = 25^\circ\text{C}$ )	$P_C$	20	W
Operating junction temperature	$T_{jopr}$	-40 to 135	$^\circ\text{C}$
Junction temperature	$T_j$	150	$^\circ\text{C}$
Storage temperature	$T_{stg}$	-55 to 150	$^\circ\text{C}$

Note: Using continuously under heavy loads (e.g. the application of high temperature/current/voltage and the significant change in temperature, etc.) may cause this product to decrease in the reliability significantly even if the operating conditions (i.e. operating temperature/current/voltage, etc.) are within the absolute maximum ratings and the operating ranges.

Please design the appropriate reliability upon reviewing the Toshiba Semiconductor Reliability Handbook ("Handling Precautions"/"Derating Concept and Methods") and individual reliability data (i.e. reliability test report and estimated failure rate, etc).

### 8.1. Safe Operating Area



**Figure8.1 SOA at  $T_j = 135^\circ\text{C}$**



## 9. Operating Ranges

Table 9.1 Operating Ranges(Ta=25 °C)

Characteristics	Symbol	Test Condition	Min	Typ.	Max	Unit
Operating power supply voltage	V <sub>BB</sub>	—	50	280	450	V
	V <sub>CC</sub>	—	13.5	15	17.5	

## 10. Electrical Characteristics

Table 10.1 Electrical Characteristics(Ta=25 °C)

Characteristics	Symbol	Test Condition	Min	Typ.	Max	Unit
Current dissipation	I <sub>BB</sub>	V <sub>BB</sub> = 450 V, Duty cycle = 0 %	—	—	0.5	mA
	I <sub>CC</sub>	V <sub>CC</sub> = 15 V, Duty cycle = 0 %	—	0.9	1.5	
	I <sub>BS</sub> (ON)	V <sub>BS</sub> = 15 V, high side ON	—	85	150	μA
	I <sub>BS</sub> (OFF)	V <sub>BS</sub> = 15 V, high side OFF	—	75	130	
Hall amp input sensitivity	VHSENS(HA)	—	50	—	—	mV <sub>p-p</sub>
Hall amp input current	IHB(HA)	—	-2	0	2	μA
Hall amp common input voltage	CMV <sub>IN</sub> (HA)	—	0.7	—	V <sub>REG</sub> -1.3	V
Hall amp hysteresis width	ΔV <sub>IN</sub> (HA)	—	8	30	62	mV
Hall amp input voltage L→H	VLH(HA)	—	4	15	31	
Hall amp input voltage H→L	VHL(HA)	—	-31	-15	-4	
Output saturation voltage	V <sub>CEsat</sub>	V <sub>CC</sub> = 15 V, I <sub>outp</sub> = 0.5 A	—	2.0	3.0	V
FRD forward voltage	V <sub>F</sub>	I <sub>F</sub> = 0.5 A	—	1.5	3.0	V
PWM ON-duty cycle	PWMMIN	—	0	—	—	%
	PWMMAX	—	—	—	100	
PWM ON-duty cycle, 0 %	VV <sub>S</sub> 0 %	PWM = 0 %	1.7	2.1	2.5	V
PWM ON-duty cycle, 100 %	VV <sub>S</sub> 100 %	PWM = 100 %	4.9	5.4	6.1	V
PWM ON-duty voltage range	VV <sub>S</sub> W	VV <sub>S</sub> 100 % – VV <sub>S</sub> 0 %	2.8	3.3	3.8	V
Output all-OFF voltage	VV <sub>S</sub> OFF	Output all OFF	1.1	1.3	1.5	V
Regulator voltage	V <sub>REG</sub>	V <sub>CC</sub> = 15 V, I <sub>REG</sub> = 30 mA	4.5	5.0	5.5	V
Speed control voltage range	V <sub>S</sub>	—	0	—	6.5	V
FG output saturation voltage	V <sub>FGsat</sub>	V <sub>CC</sub> = 15 V, I <sub>FG</sub> = 5 mA	—	—	0.5	V
Current limit voltage	V <sub>R</sub>	—	0.46	0.5	0.54	V
Current limit delay time	DtR	—	—	3	4.5	μs
Over-current protection voltage	V <sub>CS</sub>	—	0.64	0.7	0.76	V
Over-current protection delay time	t <sub>cs</sub>	C <sub>4</sub> =470 pF, R <sub>2</sub> =2 MΩ	—	2.2	3.5	μs
Over-current protection recovery time	t <sub>csr</sub>		—	1.0	2.0	ms
Thermal shutdown temperature	TSD	—	135	—	185	°C
Thermal shutdown hysteresis	ΔTSD	—	—	50	—	°C
V <sub>CC</sub> under-voltage protection	V <sub>CC</sub> UVD	—	10	11	12	V
V <sub>CC</sub> under-voltage protection recovery	V <sub>CC</sub> UVR	—	10.5	11.5	12.5	V

V <sub>BS</sub> under-voltage protection	V <sub>BS</sub> UVD	—	2.0	3.0	4.0	V
V <sub>BS</sub> under-voltage protection recovery	V <sub>BS</sub> UVR	—	2.5	3.5	4.5	V
Refresh operating ON voltage	V <sub>R</sub> ON	Refresh operation ON	1.1	1.3	1.5	V
Refresh operating OFF voltage	V <sub>R</sub> OFF	Refresh operation OFF	3.1	3.8	4.6	V
Triangular wave frequency	f <sub>c</sub>	R <sub>3</sub> = 27 kΩ, C <sub>5</sub> = 1000 pF	16.5	20	25	kHz
Output-on delay time	t <sub>on</sub>	V <sub>BB</sub> = 280 V, V <sub>CC</sub> = 15 V, I <sub>outp</sub> = 0.5 A	—	2.5	3.5	μs
Output-off delay time	t <sub>off</sub>	V <sub>BB</sub> = 280 V, V <sub>CC</sub> = 15 V, I <sub>outp</sub> = 0.5 A	—	2.0	3.0	μs
FRD reverse recovery time	t <sub>rr</sub>	V <sub>BB</sub> = 280 V, V <sub>CC</sub> = 15 V, I <sub>outp</sub> = 0.5 A	—	200	—	ns

## 11. Application Circuit Example

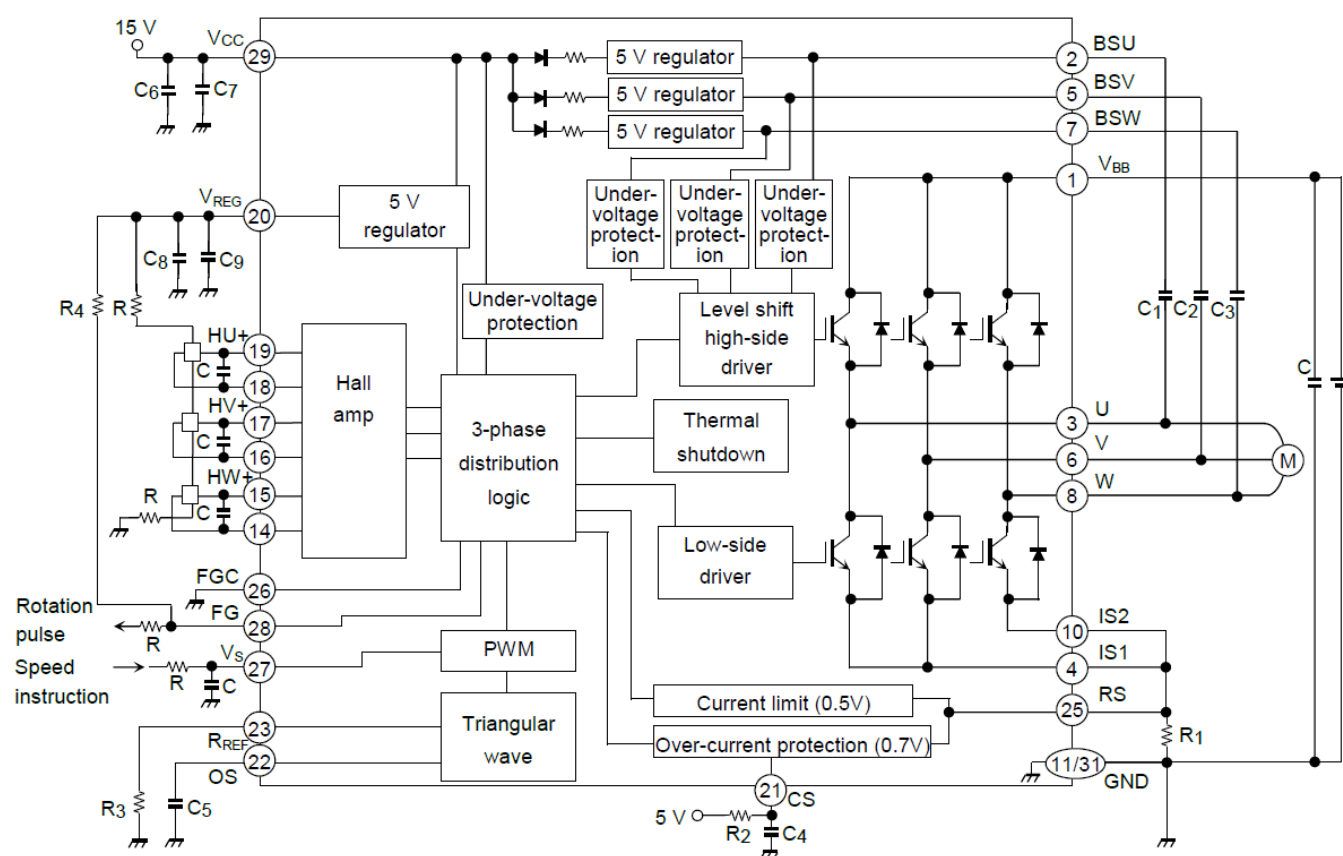


Figure 11.1 Application Circuit Example

Typical external parts are shown in the following table.

**Table 11.1 Typical external parts**

Part	Typical	Purpose	Remarks
C <sub>1</sub> , C <sub>2</sub> , C <sub>3</sub>	25 V/2.2 $\mu$ F	Bootstrap capacitor	(Note 1)
R <sub>1</sub>	1 $\Omega \pm 1\%$ (1 W)	Current detection	(Note 2)
R <sub>2</sub>	2 M $\Omega \pm 5\%$	Over-current protection recovery setup	(Note 3)
C <sub>4</sub>	25 V / 470 pF	Over-current protection recovery setup	(Note 3)
C <sub>5</sub>	25 V/1000 pF $\pm 5\%$	PWM frequency setup	(Note 4)
R <sub>3</sub>	27 k $\Omega \pm 5\%$	PWM frequency setup	(Note 4)
C <sub>6</sub>	25 V/10 $\mu$ F	Control power supply stability	(Note 5)
C <sub>7</sub>	25 V/0.1 $\mu$ F		
C <sub>8</sub>	25 V/10 $\mu$ F	V <sub>REG</sub> power supply stability	(Note 5)
C <sub>9</sub>	25 V/0.1 $\mu$ F		
R <sub>4</sub>	5.1 k $\Omega \pm 5\%$	FG pin pull-up resistor	(Note 6)

Note 1: The required bootstrap capacitance value varies according to the motor drive conditions.

Although the IC can operate at above the V<sub>BS</sub> undervoltage level, it is however recommended that the capacitor voltage be greater than or equal to 3.5 V to keep the power dissipation small.

Note 2: The following formula shows the detection current:  $I_O = V_R \div R_1$  (at V<sub>R</sub> = 0.5 V (typ.))

Do not exceed a detection current of 0.7 A when using the IC.

Note 3: A setup of over-current protection recovery and the refreshment operating time at the time of a return are set up in the combination of C<sub>4</sub> and R<sub>2</sub> which were shown in the formula.

And recommends a setup of C<sub>4</sub> and R<sub>2</sub> from which refreshment operating time is set to 190  $\mu$ s or more. If the CS pin is not used, connect to the V<sub>REG</sub> pin.

Over-current protection recovery time =  $1.06 \times C_4 \times R_2$  [s]

Refreshment operating time =  $0.21 \times C_4 \times R_2$  [s]

Note 4: With the combination of C<sub>5</sub> and R<sub>3</sub> shown in the table, the PWM frequency is around 20 kHz. The IC intrinsic error factor is around 10 %.

The PWM frequency is broadly expressed by the following formula. (In this case, the stray capacitance of the printed circuit board needs to be considered.)

$$f_c = 0.65 \div \{ C_5 \times (R_3 + 4.25 \text{ k}\Omega) \} \text{ [Hz]}$$

R<sub>3</sub> creates the reference current of the PWM triangular wave charge/discharge circuit. If R<sub>3</sub> is set too small it exceeds the current capacity of the IC internal circuits and the triangular wave distorts. Set R<sub>3</sub> to at least 9 k $\Omega$ .

Note 5: When using the IC, adjustment is required in accordance with the use environment. When mounting, place as close to the base of the IC leads as possible to improve noise elimination.

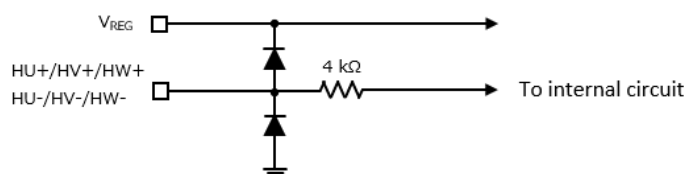
Note 6: The FG pin is open drain. If the FG pin is not used, connect to the GND.

Note 7: If noise is detected on the Input signal pin, add a capacitor between inputs.

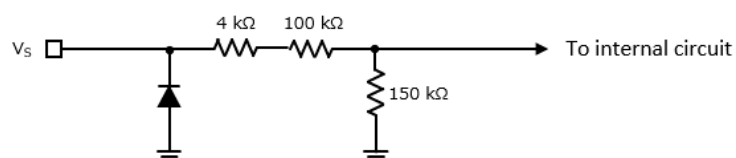
Note 8: A Hall device should be an indium antimony system. It is recommend that the peak Hall device voltage be set more than 300 mV

## 12. Internal circuit diagrams

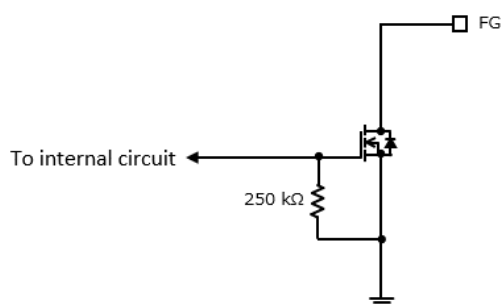
Internal circuit diagram of HU+, HU-, HV+, HV-, HW+, HW- input pins



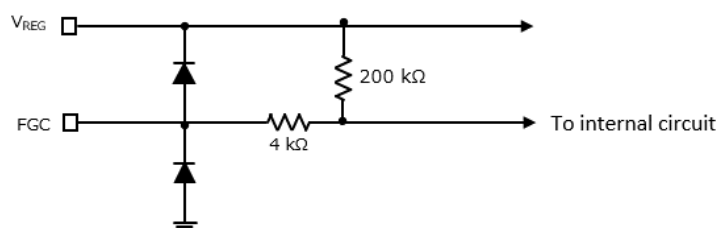
Internal circuit diagram of Vs pin



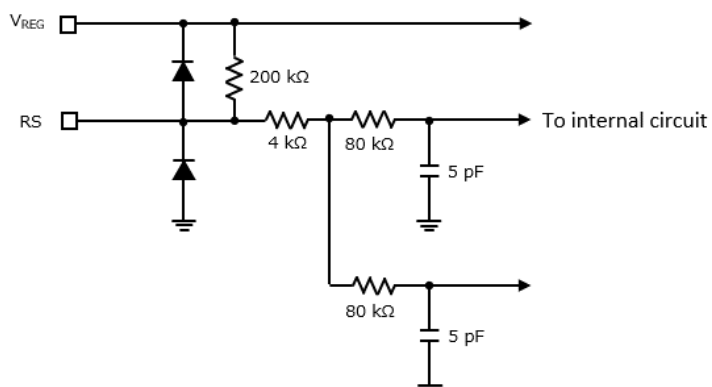
Internal circuit diagram of FG pin



Internal circuit diagram of FGC pin



Internal circuit diagram of RS pin

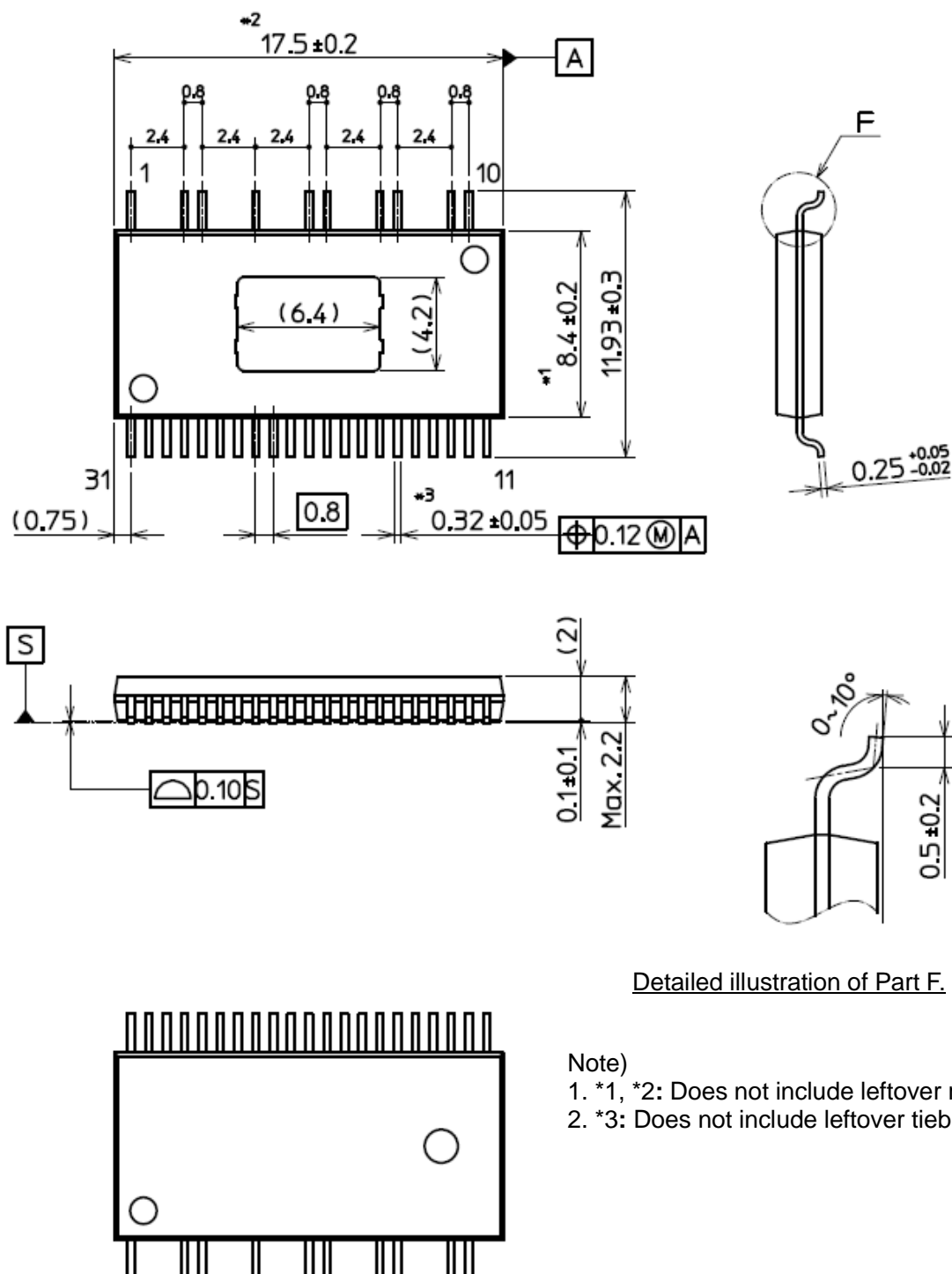


## 13. Package Information

### 13.1. Package Dimensions

P-HSSOP31-0918-0.80-002

Unit:mm



Weight: 0.7 g (typ.)

Figure 13.1 Package Dimensions

## 13.2. Marking

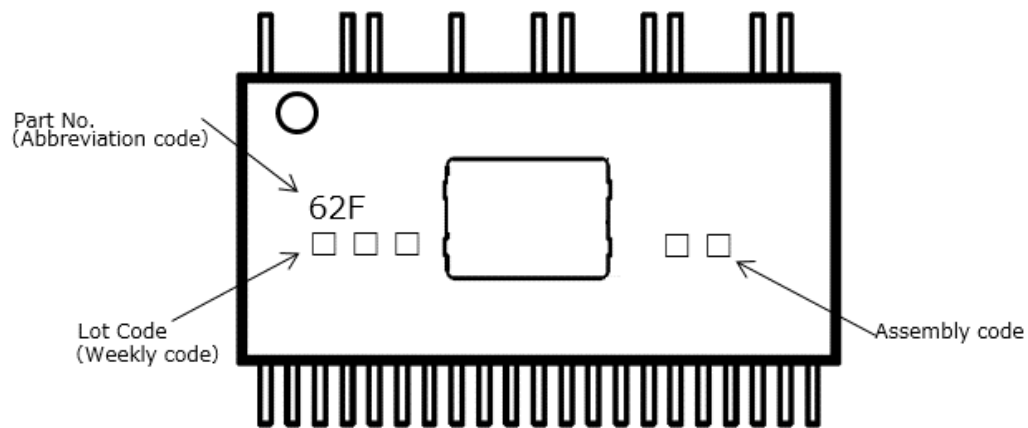


Figure 13.2 Marking

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