TOSHIBA Bi-CMOS Integrated Circuit Silicon Monolithic

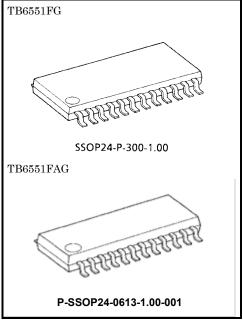
# **TB6551FG, TB6551FAG**

### 3-Phase Full-Wave Sine-Wave PWM Brushless Motor Controller

The TB6551FG/FAG is designed for motor fan applications for three-phase brushless DC (BLDC) motors.

#### **Features**

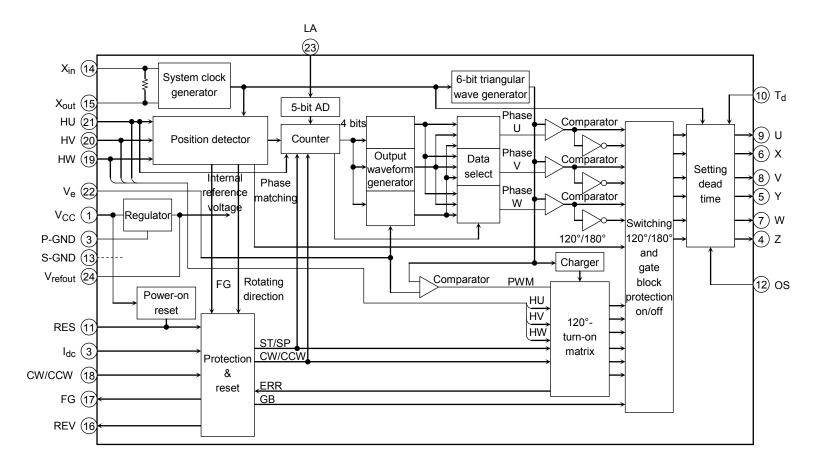
- Sine-wave PWM control
- Built-in triangular-wave generator (Carrier cycle = fOSC/252 (Hz))
- Built-in lead angle control function (0° to 58° in 32 steps)
- Built-in dead time function (setting 2.6 μs or 3.8 μs)
- Bootstrap circuit compliant
- Over-current protection signal input pin
- Built-in regulator ( $V_{ref} = 5 \text{ V (typ.)}, 30 \text{ mA (max)}$ )
- Operating supply voltage range:  $V_{CC} = 6 \text{ V}$  to 10 V



Weight:

SSOP24-P-300-1.00 : 0.33 g (typ.) P-SSOP24-0613-1.00-001: 0.28 g (typ.)

## **Block Diagram**





## **Pin Description**

Pin No.	Symbol	Description	Remarks
21	HU	Positional signal input pin U	
20	HV	Positional signal input pin V	When positional signal is HHH or LLL, gate block protection operates.
19	HW	Positional signal input pin W	With built-in pull-up resistor
18	CW/CCW	Rotation direction signal input pin	L: Forward H: Reverse
11	RES	Reset-signal-input pin	L: Reset (output is non-active)  Operation/Halt operation  Also used for gate block protection
22	V <sub>e</sub>	Inputs voltage instruction signal	With built-in pull-down resistor
23	LA	Lead angle setting signal input pin	Sets 0° to 58° in 32 steps
12	os	Inputs output logic select signal	L: Active low H: Active high
3	I <sub>dc</sub>	Inputs over-current- protection-signal	Inputs DC link current.  Reference voltage: 0.5 V  With built-in filter ( ≈ 1 µs)
14	X <sub>in</sub>	Inputs clock signal	
15	X <sub>out</sub>	Outputs clock signal	With built-in feedback resistor
24	V <sub>refout</sub>	Outputs reference voltage signal	5 V (typ.), 30 mA (max)
17	FG	FG signal output pin	Outputs 3PPR of positional signal
16	REV	Reverse rotation detection signal	Detects reverse rotation.
9	U	Outputs turn-on signal	
8	V	Outputs turn-on signal	
7	W	Outputs turn-on signal	Sologi pating high as active low units the southest lasis and a size
6	Х	Outputs turn-on signal	Select active high or active low using the output logic select pin.
5	Y	Outputs turn-on signal	
4	Z	Outputs turn-on signal	
1	V <sub>CC</sub>	Power supply voltage pin	V <sub>CC</sub> = 6 V to 10 V
10	T <sub>d</sub>	Inputs setting dead time	L: 3.8 µs, H or Open: 2.6 µs
2	P-GND	Ground for power supply	Ground pin
13	S-GND	Ground for signals	Ground pin



## Input/Output Equivalent Circuits

Pin Description	Symbol	Input/Output Signal	Input/Output Internal Circuit
Positional signal input pin U	ни	Digital  With Schmitt trigger	V <sub>refout</sub> V <sub>refout</sub>
Positional signal input pin V	HV	Hysteresis 300 mV (typ.)	240 KD
Positional signal input pin W	HW	L: 0.8 V (max) H: V <sub>refout</sub> - 1 V (min)	2.4 kΩ
Forward/reverse switching input pin  L: Forward (CW)  H: Reverse (CCW)	CW/CCW	Digital  With Schmitt trigger  Hysteresis 300 mV (typ.)  L: 0.8 V (max)  H: V <sub>refout</sub> - 1 V (min)	V <sub>refout</sub> V <sub>refout</sub> Q Q Q Q Q Q Q Q Q Q Q Q Q Q Q Q Q Q
Reset input L: Stops operation (reset). H: Operates.	RES	Digital  With Schmitt trigger  Hysteresis 300 mV (typ.)  L: 0.8 V (max)  H: V <sub>refout</sub> - 1 V (min)	V <sub>refout</sub> 2.4 kΩ  GY  ON  ON  ON  ON  ON  ON  ON  ON  ON  O
Voltage instruction signal input pin  Turn on the lower transistor at 0.2 V or less.  (X, Y, Z pins: On duty of 8%)	Ve	Analog  Input range 0 V to 5.0 V  Input voltage of Vrefout or higher is clipped to Vrefout.	0
Lead angle setting signal input pin  0 V: 0°  5 V: 58° (5-bit AD)	LA	Analog  Input range 0 V to 5.0 V  Input voltage of V <sub>refout</sub> or higher is clipped to V <sub>refout</sub> .	240 KD



Pin Description	Symbol	Input/Output Signal	Input/Output Internal Circuit
Setting dead time input pin L: 3.8 µs H or Open: 2.6 µs	T <sub>d</sub>	Digital L: 0.8 V (max) H: V <sub>refout</sub> - 1 V (min)	V <sub>refout</sub> V <sub>refout</sub> 1 G 2 0 2 1.2 kΩ
Output logic select signal input pin  L: Active low  H: Active high	os	Digital L: 0.8 V (max) H: V <sub>refout</sub> - 1 V (min)	V <sub>refout</sub> V <sub>refout</sub> G  Z  Z  Z  Z  Z  Z  Z  Z  Z  Z  Z  Z
Over-current protection signal input pin	I <sub>dc</sub>	Analog  Gate block protected at 0.5 V or higher (released at carrier cycle)	V <sub>CC</sub> 240 kΩ Comparator  y <sub>1</sub> y <sub>2</sub> y <sub>3</sub> y <sub>4</sub> y <sub>5</sub> y <sub>6</sub> y <sub>7</sub>
Clock signal input pin	X <sub>in</sub>	Operating range	V <sub>refout</sub> V <sub>refout</sub> X <sub>in</sub> O X <sub>out</sub>
Clock signal output pin	X <sub>out</sub>	2 MHz to 8 MHz (ceramic oscillation)	360 kΩ
Reference voltage signal output pin	Vrefout	5 ± 0.5 V (max 30 mA)	Vcc Vcc Vcc T



Pin Description	Symbol	Input/Output Signal	Input/Output Internal Circuit
Reverse-rotation-detection signal output pin		Digital  Push-pull output: ± 1 mA (max)	V <sub>refout</sub> V <sub>refout</sub> 120 Ω
FG signal output pin	FG	Digital Push-pull output: ± 1 mA (max)	V <sub>refout</sub> V <sub>refout</sub>
Turn-on signal output pin U Turn-on signal output pin V Turn-on signal output pin W Turn-on signal output pin X Turn-on signal output pin Y Turn-on signal output pin Z	U V W X Y	Analog  Push-pull output: ± 2 mA (max)  L: 0.78 V (max)  H: V <sub>refout</sub> = 0.78 V (min)	V <sub>refout</sub> 120 Ω

## Absolute Maximum Ratings (Ta = 25°C)

Characteristics Symbol Rating				Unit	
Supply voltage	V <sub>CC</sub>	1	2	V	
Input voltage	V <sub>in (1)</sub>	-0.3 to V <sub>CC</sub> (Note 1)		V	
input voitage	V <sub>in (2)</sub>	-0.3	v		
Turn-on signal output current	lout	2		mA	
Power Dissipation	D-	FG	0.9 (Note 3)	W	
	P <sub>D</sub>	FAG	1.0 (Note 3)		
Operating temperature	T <sub>opr</sub>	-30 to 115 (Note 4)		°C	
Storage temperature	T <sub>stg</sub>	-50 to 150		°C	

Note 1: V<sub>in (1)</sub> pin: V<sub>e</sub>, LA

Note 2: Vin (2) pin: HU, HV, HW, CW/CCW, RES, OS, Idc, Td

Note 3: When mounted on a PCB (universal 50 mm  $\times$  50 mm  $\times$  1.6 mm, Cu 30%)

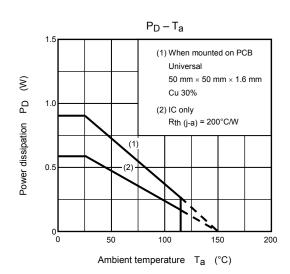
Note 4: Operating temperature range is determined by the  $P_D$  –  $T_a$  characteristic.

Note 5: Using continuously under heavy loads (e.g. the application of high temperature/current/voltage and the significant change in temperature, etc.) may cause this product to decrease in the reliability significantly even if the operating conditions (i.e. operating temperature/current/voltage, etc.) are within the absolute maximum ratings.

Please design the appropriate reliability upon reviewing the Toshiba Semiconductor Reliability Handbook ("Handling Precautions"/"Derating Concept and Methods") and individual reliability data (i.e. reliability test report and estimated failure rate, etc).

## Operating Conditions ( $T_a = 25$ °C)

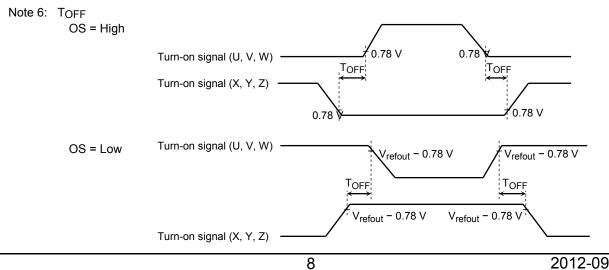
Characteristics	Symbol	Min	Тур.	Max	Unit
Supply voltage	V <sub>CC</sub>	6	7	10	V
Ceramic oscillation frequency	X <sub>in</sub>	2	4	8	MHz





## Electrical Characteristics ( $T_a = 25$ °C, $V_{CC} = 7$ V)

Characteristics	s Symbol		Test Circuit	Test Condition		Тур.	Max	Unit
Supply current	ly current I <sub>CC</sub>		_	V <sub>refout</sub> = open	_	3	6	mA
	l <sub>in (</sub>	1)		V <sub>in</sub> = 5 V V <sub>e</sub> , LA	_	20	40	
Input current	I <sub>in (2)</sub> -1			V <sub>in</sub> = 0 V HU, HV, HW	-40	-20	_	μΑ
input current			_	V <sub>in</sub> = 0 V CW/CCW, OS, T <sub>d</sub>	-80	-40		
	I <sub>in (2</sub>	)-3		V <sub>in</sub> = 5 V RES	_	40	80	
Input voltage	V <sub>in</sub>	High		HU, HV, HW, CW/CCW, RES, OS, T <sub>d</sub>	V <sub>refout</sub>		V <sub>refout</sub>	٧
		Low			_	_	0.8	
Input hysteresis voltage	V <sub>H</sub>	1	l	HU, HV, HW, CW/CCW, RES	_	0.3	ı	V
	V <sub>OUT</sub>	(H)-1		I <sub>OUT</sub> = 2 mA U, V, W, X, Y, Z	V <sub>refout</sub> - 0.78	V <sub>refout</sub> - 0.4	ı	-
	V <sub>OUT</sub>	(L)-1		$I_{OUT} = -2 \text{ mA}$ U, V, W, X, Y, Z	_	0.4	0.78	
	VREV (H) VREV (L) VFG(H) VFG(L) Vrefout		_	I <sub>OUT</sub> = 1 mA REV	V <sub>refout</sub> - 1.0	V <sub>refout</sub> - 0.5	1	V
Output voltage				I <sub>OUT</sub> = -1 mA REV	_	0.5	1.0	
				I <sub>OUT</sub> = 1 mA FG	V <sub>refout</sub>	V <sub>refout</sub> - 0.5	-	
				I <sub>OUT</sub> = −1 mA FG	_	0.5	1.0	
				I <sub>OUT</sub> = 30 mA V <sub>refout</sub>	4.5	5.0	5.5	
Output leakage	I <sub>L (H)</sub>			V <sub>OUT</sub> = 0 V U, V, W, X, Y, Z	_	0	10	
current			_	V <sub>OUT</sub> = 3.5 V U, V, W, X, Y, Z	_	0	10	μA
Output off-time by upper/lower transistor	nsistor		-	$T_d$ = High or OPEN, $X_{in}$ = 4.19 MHz, $I_{OUT}$ = $\pm$ 2 mA, OS = High/Low	2.2	2.6	ı	
(Note 6)			_	T <sub>d</sub> = Low, X <sub>in</sub> = 4.19 MHz, I <sub>OUT</sub> = ± 2 mA, OS = High/Low	3.0	3.8	_	μs
Over-current detection	V <sub>dc</sub>		_	I <sub>dc</sub>	0.46	0.5	0.54	V
	$T_{LA}$	(0)	_	L <sub>A</sub> = 0 V or Open, Hall IN = 100 Hz	_	0		
Lead angle correction	T <sub>LA (2.5)</sub>		-	L <sub>A</sub> = 2.5 V, Hall IN = 100 Hz	27.5	32	34.5	0
			_	L <sub>A</sub> = 5 V, Hall IN = 100 Hz	53.5	59	62.5	
	V <sub>CC</sub>	(H)	_	Output start operation point	4.2	4.5	4.8	
V <sub>CC</sub> monitor	V <sub>CC (L)</sub>		_	No output operation point	3.7	4.0	4.3	V
	V <sub>F</sub>	1	_	Input hysteresis width	_	0.5	_	





## **Functional Description**

### **Basic operation**

On start-up, the motor is driven by the square-wave turn-on signal based on a positional signal. When the positional signal reaches number of rotations f = 5 Hz or higher, the rotor position is inferred from the positional signal and a modulation wave is generated. The modulation wave and the triangular wave are compared; the sine-wave PWM signal is then generated and the motor is driven.

From start to 5 Hz: When driven by square wave (120° turn-on)  $f = fOSC/(2^{12} \times 32 \times 6)$  5 Hz or higher: When driven by sine-wave PWM (180° turn-on) When fOSC = 4 MHz, approx. 5 Hz

#### Function to stabilize bootstrap voltage

(1) When voltage instruction is input at  $V_e \le 0.2 \text{ V}$ :

The lower transistor is turned on at the regular (carrier) cycle. (On duty is approx. 8%.)

(2) When voltage instruction is input at  $V_e > 0.2 \text{ V}$ :

During sine-wave drive, the drive signal is output as it is.

During square-wave drive, the lower transistor is forcibly turned on at the regular (carrier) cycle. (On duty is approx. 8%.)

Note: At startup, to charge the upper transistor gate power supply, turn the lower transistor on for a fixed time with  $V_e \le 0.2 \ V$ .

## Dead time function: upper/lower transistor output off-time

When the motor is driven by a sine-wave PWM, dead time is generated digitally in the IC to prevent any short circuit caused by the simultaneous turning on of upper and lower external power devices.

When a square wave is generated in full duty cycle mode, the dead time function is turned on to prevent a short circuit.

T <sub>d</sub> Pin	Internal Counter	T <sub>OFF</sub>	
High or Open	11/f <sub>OSC</sub>	2.6 µs	
Low	16/f <sub>OSC</sub>	3.8 µs	

Toff values above are obtained when fOSC = 4.19 MHz.

fosc = reference clock (ceramic oscillation)

#### Correcting lead angle

The lead angle can be corrected in the turn-on signal range from 0 to 58° in relation to the induced voltage.

Analog input from LA pin (0 V to 5 V divided by 32):

 $0 \text{ A} = 0 \circ$ 

 $5 \text{ V} = 58^{\circ}$  (when more than 5 V is input,  $58^{\circ}$ )

### **Setting carrier frequency**

This feature sets the triangular wave cycle (carrier cycle) necessary for generating the PWM signal. (The triangular wave is used for forcibly turning on the lower transistor when the motor is driven by square wave.)

9

Carrier cycle = fOSC/252 (Hz)

fosc = Reference clock (ceramic oscillation)

### Switching the output of turn-on signal

This function switches the output of the turn-on signal between high and low.

Pin OS:

High = active high

Low = active low



## Outputting reverse rotation detection signal

The direction of motor rotation is detected for every electrical angle of 360°. (The output is high immediately after reset.)

The REV terminal increases to a 180° turn-on mode at the time of low (Hall IN  $\geq$  5 Hz).

CW/CCW Pin	Actual Motor Rotating Direction	REV Pin	
Low (CM)	CW (forward)	Low	
Low (CW)	CCW (reverse)	High	
High (CCM)	CW (forward)	High	
High (CCW)	CCW (reverse)	Low	

## **Protecting input pin**

1. Over-current protection (Pin Idc)

When the DC-link-current exceeds the internal reference voltage, gate block protection is performed. Over-current protection is released for each carrier frequency.

Reference voltage = 0.5 V (typ.)

2. Gate block protection (Pin RES)

When the input signal level is Low, the output is turned off; when the signal is High, the output is restarted.

Abnormalities are detected externally, and the signal is input to the pin RES.

RES Pin	OS Pin	Output Turn-on Signal (U, V, W, X, Y, Z)	
Low	Low	High	
Low	High	Low	

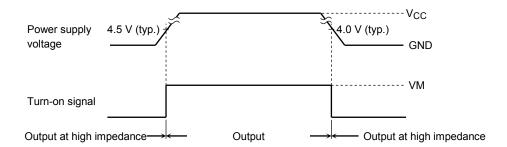
(When RES = Low, bootstrap capacitor charging stops.)

- 3. Internal protection
  - Positional signal abnormality protection

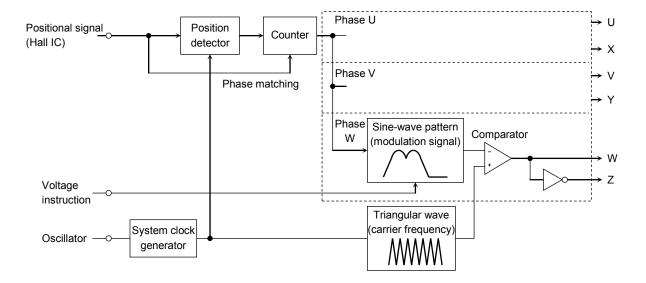
When the positional signal is HHH or LLL, the output is turned off; otherwise, the output is restarted.

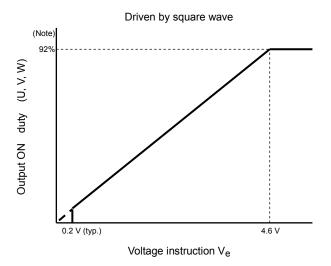
• Low power supply voltage protection (VCC monitor)

Outside the operating voltage range, the turn-on signal output is kept at high impedance to prevent damage caused by short-circuiting of power components when the power supply is turned on or off.

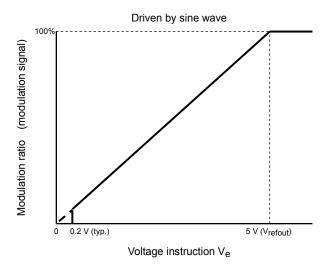


## **Operation Flow**





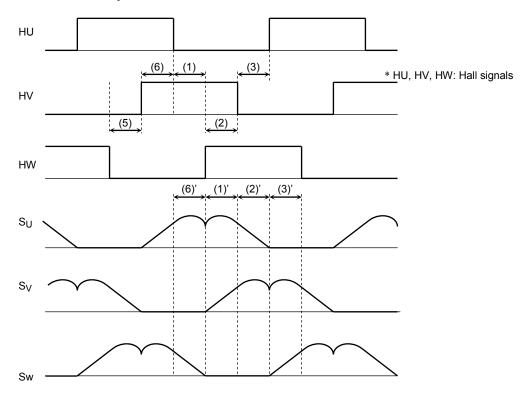
Note: Output ON time is decreased by the dead time (carrier frequency  $\times$  92% – T<sub>d</sub>  $\times$  2).



The modulation waveform is generated using Hall signals. The modulation waveform is then compared with the triangular wave and a sine-wave PWM signal is generated.

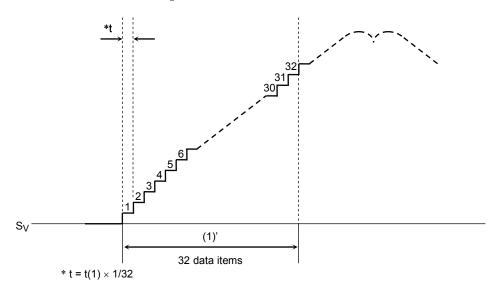
The time (electrical angle: 60°) from the rising (or falling) edges of the three Hall signals to the next falling (or rising) edges is counted. The counted time is used as the data for the next 60° phase of the modulation waveform.

There are 32 items of data for the 60° phase of the modulation waveform. The time width of one data item is 1/32 of the time width of the 60° phase of the previous modulation waveform. The modulation waveform moves forward by this width.



In the above diagram, the modulation waveform (1)' data moves forward by the 1/32 time width of the time (1) from HU:  $\downarrow$  to HW:  $\uparrow$  Similarly, data (2)' moves forward by the 1/32 time width of the time (2) from HW:  $\uparrow$  to HV: $\downarrow$ .

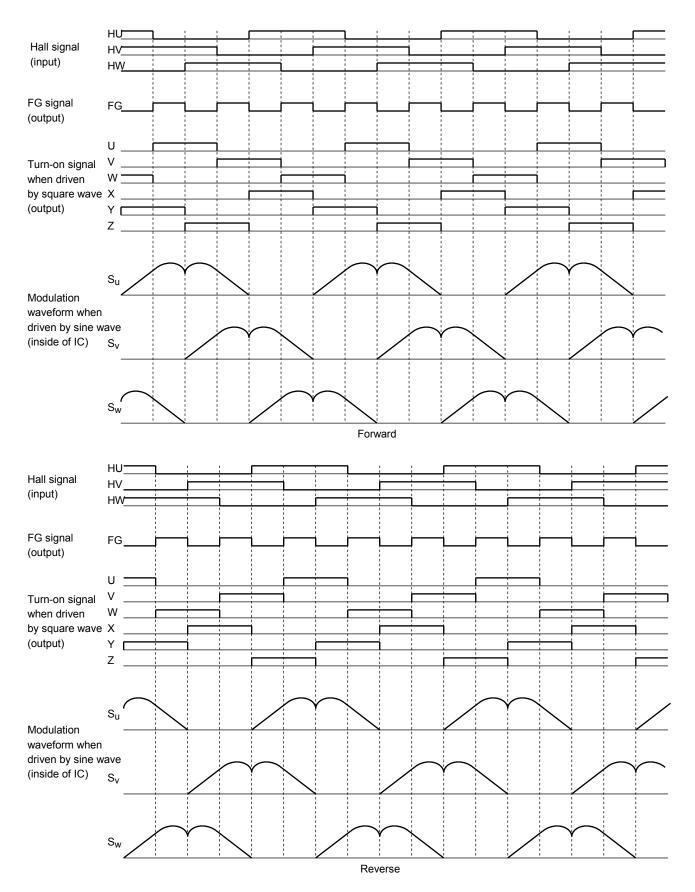
If the next edge does not occur after the 32 data items end, the next 32 data items move forward by the same time width until the next edge occurs.



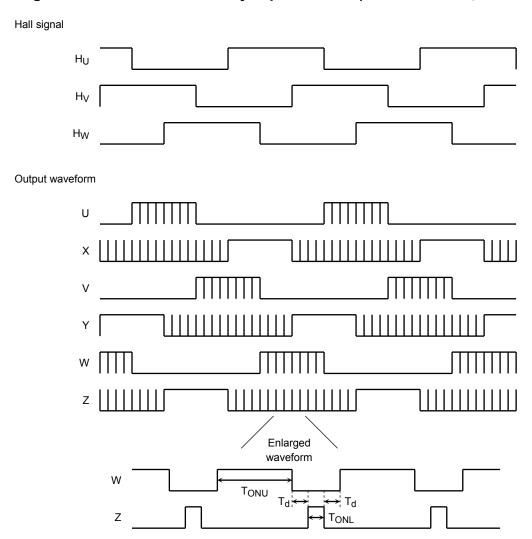
The modulation wave is brought into phase with every edge of the Hall signal.

The modulation wave is reset in synchronization with the rising and falling edges of the Hall signal at every electrical angle of 60°. Thus, when the Hall device is not placed in the correct position or during accelerating or decelerating, the modulation waveform is not continuous at every reset.

## **Timing Charts**



## Operating Waveform When Driven by Square Wave (CW/CCW = Low, OS = High)



To stabilize the bootstrap voltage, the lower outputs  $(X,\,Y,\,\text{and}\,Z)$  are always turned on at the carrier cycle even during off time. At that time, the upper outputs  $(U,\,V,\,\text{and}\,W)$  are assigned dead time and turned off at the timing when the lower outputs are turned on.  $(T_d\,\text{varies}\,\text{with}\,\text{input}\,V_e.)$ 

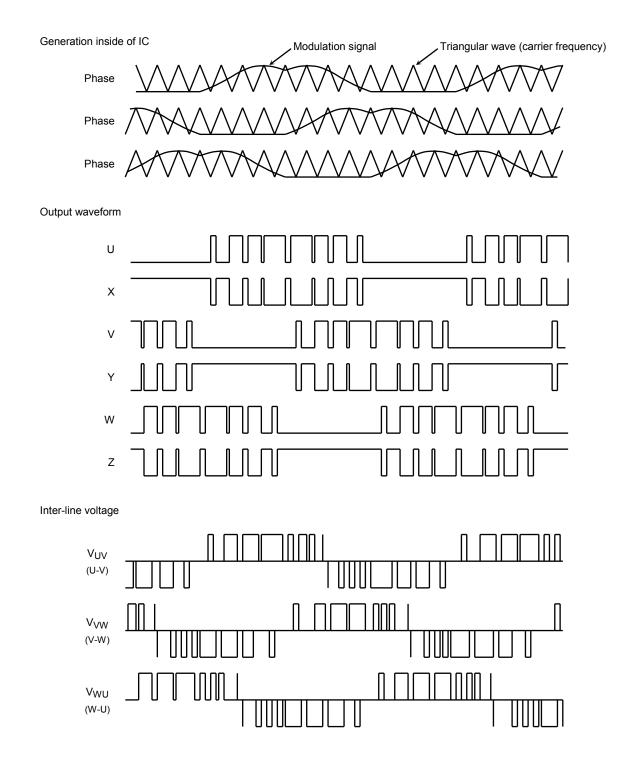
Carrier cycle = 
$$f_{OSC}/252$$
 (Hz) Dead time:  $T_d$  =  $16/f_{OSC}$  (s) (when  $V_e$  = 4.6 V or more)

 $T_{ONL}$  = carrier cycle  $\times$  8% (s) (uniform regardless of Ve input)

When the motor is driven by a square wave, acceleration or deceleration is determined by voltage  $V_e$ . The motor accelerates or decelerates according to the On duty of  $T_{\mbox{ONU}}$ . (See the diagram for output On duty on page 11.)

Note: The motor is driven by a square wave if REV = High, i.e., if the Hall signals at start-up are 5 Hz ( $f_{OSC}$  = 4 MHz) or lower and the motor is rotating in the reverse direction to that of the TB6551FG/FAG setting.

## Operating Waveform When Driven by Sine-Wave PWM (CW/CCW = Low, OS = High)

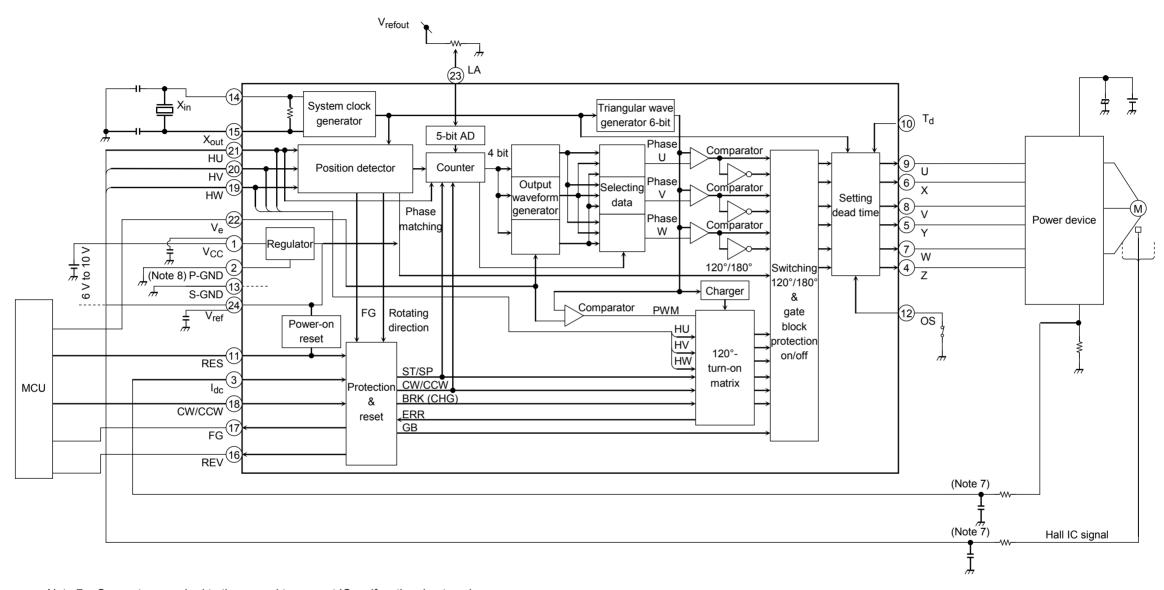


When the motor is driven by a sine wave, the motor is accelerated or decelerated according to the On duty of  $T_{ONU}$  when the amplitude of the modulation symbol changes by voltage  $V_e$  (see the diagram of output On duty on page 11):

Triangular wave frequency = carrier frequency =  $f_{OSC}/252$  (Hz).

Note: The motor is driven by a sine wave if REV = Low, i.e., if the Hall signals at start-up are 5 Hz (f<sub>OSC</sub> = 4 MHz) or higher and the motor is rotating in the same direction as that of the TB6551FG/FAG setting.

## **Example of Application Circuit**

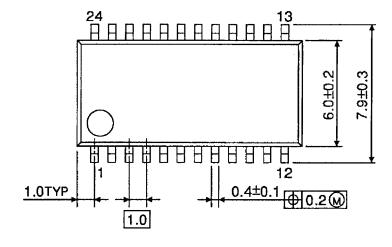


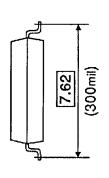
- Note 7: Connect as required to the ground to prevent IC malfunction due to noise.
- Note 8: Connect P-GND to signal ground on the application circuit.
- Note 9: Utmost care is necessary in the design of the output, V<sub>CC</sub>, VM, and GND lines since the IC may be destroyed by short-circuiting between outputs, air contamination faults, or faults due to improper grounding, or by short-circuiting between contiguous pins.

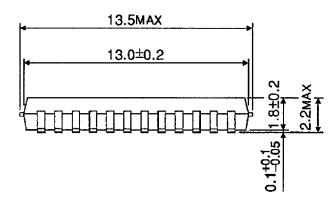
Unit: mm

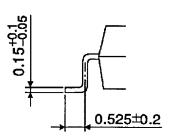
## **Package Dimensions**

SSOP24-P-300-1.00







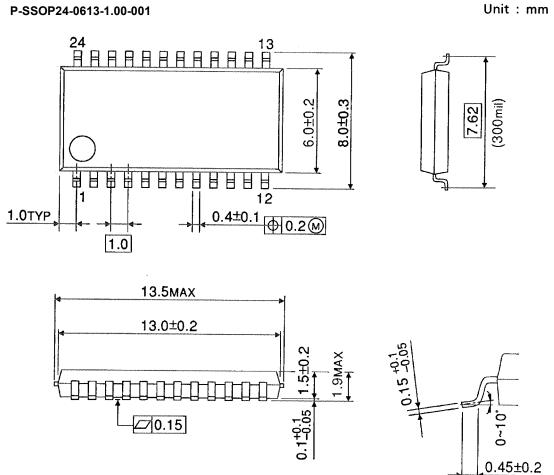


Weight: 0.33 g (typ.)

## **Package Dimensions**

**TOSHIBA** 

P-SSOP24-0613-1.00-001



Weight: 0.28 g (typ.)

#### **Notes on Contents**

#### 1. Block Diagrams

Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purposes.

#### 2. Equivalent Circuits

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

#### 3. Timing Charts

Timing charts may be simplified for explanatory purposes.

### 4. Application Circuits

The application circuits shown in this document are provided for reference purposes only. Thorough evaluation is required, especially at the mass production design stage.

To shib does not grant any license to any industrial property rights by providing these examples of application circuits.

#### 5. Test Circuits

Components in the test circuits are used only to obtain and confirm the device characteristics. These components and circuits are not guaranteed to prevent malfunction or failure from occurring in the application equipment.

## IC Usage Considerations Notes on handling of ICs

- [1] The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings.
  - Exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.
- [2] Do not insert devices in the wrong orientation or incorrectly.
  - Make sure that the positive and negative terminals of power supplies are connected properly.
  - Otherwise, the current or power consumption may exceed the absolute maximum rating, and exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.

In addition, do not use any device that is applied the current with inserting in the wrong orientation or incorrectly even just one time.

19

## Points to remember on handling of ICs

#### (1) Over current Protection Circuit

Over current protection circuits (referred to as current limiter circuits) do not necessarily protect ICs under all circumstances. If the over current protection circuits operate against the over current, clear the over current status immediately.

Depending on the method of use and usage conditions, such as exceeding absolute maximum ratings can cause the over current protection circuit to not operate properly or IC breakdown before operation. In addition, depending on the method of use and usage conditions, if over current continues to flow for a long time after operation, the IC may generate heat resulting in breakdown.

#### (2) Back-EMF

When a motor rotates in the reverse direction, stops or slows down abruptly, a current flow back to the motor's power supply due to the effect of back-EMF. If the current sink capability of the power supply is small, the device's motor power supply and output pins might be exposed to conditions beyond absolute maximum ratings. To avoid this problem, take the effect of back-EMF into consideration in system design

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