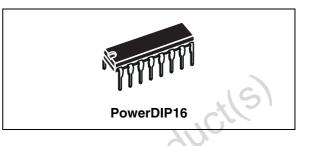
# TEA2025

## Stereo audio amplifier

### Datasheet – production data

### Features

- Dual or bridge connection modes
- Few external components
- Supply voltage 3 V to 15 V
- High channel separation
- Very low switch-on/off noise
- Max gain of 45 dB with adjustable external resistor
- Soft clipping
- Thermal protection
- $\blacksquare P_{O} = 2 \cdot 1 \text{ W}, \text{ V}_{S} = 6 \text{ V}, \text{ R}_{L} = 4 \Omega$
- $\blacksquare P_{O} = 2 \cdot 2.3 \text{ W}, \text{ V}_{S} = 9 \text{ V}, \text{ R}_{L} = 4 \Omega$
- $P_0 = 2 \cdot 0.1$  W,  $V_s = 3$  V,  $R_L = 4$  Ω



### Description

The TEA2025B is a monolithic integrated circuit housed in the 12+2+2 PowerDIP16 package, intended for use as a dual or bridge power audio amplifier in portable radio cassette players.

### Table 1. Device summary

Part number	Package
TEA2025B	PowerDIP16 (12+2+2)

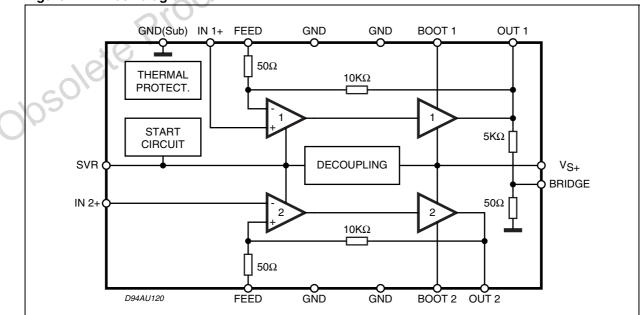


Figure 1. Block diagram

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This is information on a product in full production.

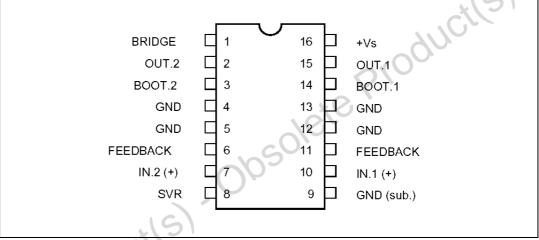
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## **Device overview and electrical specifications**

Table 2. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V <sub>S</sub>	Supply voltage	15	V
۱ <sub>۵</sub>	Ouput peak current	1.5	А
TJ	Junction temperature	150	°C
T <sub>stg</sub>	Storage temperature	150	°C

#### Figure 2. Pin connections PowerDIP16 (12+2+2)



#### Table 3. Thermal data

	Symbol	Description		PowerDIP16 (12+2+2) <sup>(1)</sup>	Unit
-	R <sub>th j-case</sub>	Thermal resistance junction-case	Max	15	°C/W
16	R <sub>th j-amb</sub>	Thermal resistance junction-ambient	Max	60	°C/W
2105 <sup>011</sup>	1. R <sub>th j-amb</sub> is n copper surfa	heasured on devices bonded on a 10 x 5 x 0.15 cm gl ce of 5 cm <sup>2</sup> .	ass-epox	y substrate with a 35 mm	thick



Symbol	Parameter	Test condition	Min.	Тур.	Max.	Uni		
VS	Supply voltage				3		12	V
Ι <sub>Q</sub>	Quiescent current					35	50	mA
Vo	Quiescent output voltage					4.5		V
^	Voltage goin	Stereo			43	45	47	dB
A <sub>V</sub>	Voltage gain	Bridge			49	51	53	dB
$\Delta A_V$	Voltage gain difference						±1	dB
Rj	Input Impedance					30		kΩ
			9 V	4 Ω	1.7	2.3	1 G	w
			9 V	8 Ω		1.3	11-	w
			6 V	4 Ω	0.7	1	<u> </u>	W
			6 V	8 Ω	-0	0.6		W
		Stereo 8 (per channel)	6 V	16 Ω		0.25		W
			6 V	32 Ω		0.13		W
Р	Output now or $(d = 10\%)$		3 V	4 Ω		0.1		W
Po	Output power (d = 10%)	50	3 V	32 Ω		0.02		W
		005	12 V	8Ω		2.4		W
			9 V	8Ω		4.7		W
		5	6 V	4 Ω		2.8		W
	Ċ	Bridge	6 V	8Ω		1.5		W
	or oduct		3 V	<b>16</b> Ω		0.18		W
	010		3 V	32 Ω		0.06		W
d	Distortion	$Vs = 9 V; R_L = 4 \Omega$	Ste			0.3	1.5	%
	xO		Brio	dge		0.5		
SVR	Supply voltage rejection	f = 100 Hz, V <sub>R</sub> = 0.5 V, R <sub>g</sub> =	: 0		40	46		dE
E <sub>N(IN)</sub>	Input noise voltage	R <sub>G</sub> = 0				1.5	3	m١
		R <sub>G</sub> = 10 4 Ω			3	6	m١	
СТ	Crosstalk	f = 1 kHz, R <sub>q</sub> = 10 kΩ			40	52		dE

Table 4.	<b>Electrical characteristcs</b> ( $T_{amb} = 25 \text{ °C}$ , $V_S = 9 \text{ V}$ , stereo unless otherwise specified)

Term. N° (PowerDIP16)	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16
DC volt. (V)	0.04	4.5	8.9	0	0	0.6	0.04	8.5	0	0.04	0.6	0	0	8.9	4.5	9



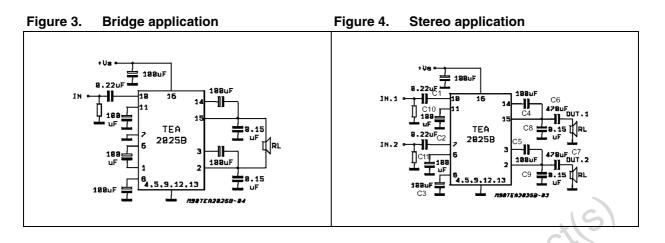
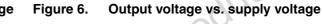


Figure 5. Supply current vs. supply voltage  $(R_L = 4 \Omega)$ 



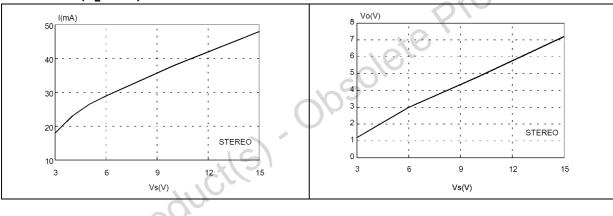
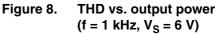
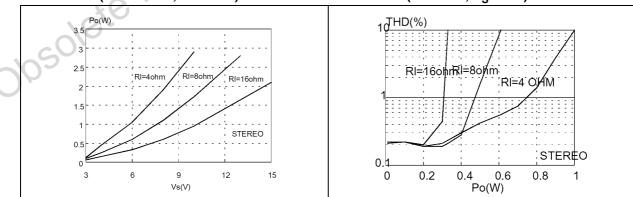


Figure 7. Output power vs. supply voltage (THD = 10%, f = 1 kHz)







#### **Application information** 2

#### 2.1 Input capacitor

The input capacitor is PNP type allowing the source to be referenced to ground. In this way no input coupling capacitor is required. However, a series capacitor (0.22 µF) to the input side can be useful in case of noise due to variable resistor contact.

#### 2.2 **Bootstrap**

The bootstrap connection allows increasing the output swing.

The recommended value for the bootstrap capacitors (100 µF) avoids a reduction of the output signal also at low frequencies and low supply voltages. Produ

#### 2.3 Voltage gain adjustment

#### 2.3.1 Stereo mode

The voltage gain is determined by on-chip resistors R1 and R2 together with the external RfC1 series connected between pin 6 (11) and ground. The frequency response is approximated by:

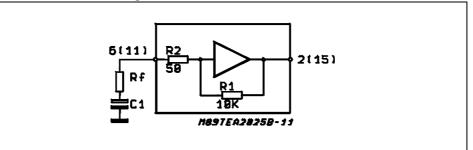
$$\frac{V_{OUT}}{V_{IN}} = \frac{R1}{Rf \div R2 + \frac{1}{JWC1}}$$

With Rf = 0, C1 = 100  $\mu$ F, the gain results in 46 dB with pole at f = 32 Hz.

The purpose of Rf is to reduce the gain. It is recommended to not reduce it under 36 dB.

#### Bridge mode 2.3.2

Figure 9. Internal voltage divider



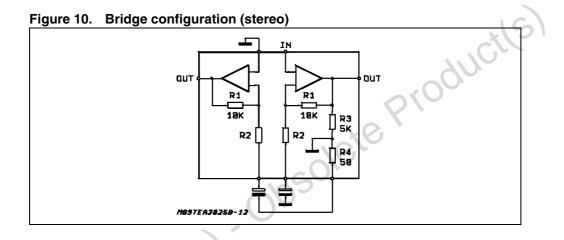
The bridge configuration is realized very easily thanks to an internal voltage divider which provides (at pin 1) the CH 1 output signal after reduction.



It is sufficient to connect pin 6 (inverting input of CH 2) with a capacitor to pin 1 and to connect pin 7 to ground. The total gain of the bridge is given by:

$$\frac{V_{OUT}}{V_{IN}} = \frac{R1}{Rf \div R2 + \frac{1}{JWC1}} \left(1 + \frac{R3}{R4} \frac{R1}{R2 + R4 + \frac{1}{JWC1}}\right)$$

and with the recommended values (C1 = C2 = 100  $\mu$ F, Rf= 0), then Gv = 52 dB with first pole at f = 32 Hz



#### **Output capacitors** 2.4

The low cutoff frequency due to the output capacitor depending on the load is given by:

$$\mathsf{F}_{\mathsf{L}} = \frac{1}{2\Pi\mathsf{C}_{\mathsf{OUT}}\cdot\mathsf{R}_{\mathsf{L}}}$$

· L =  $\frac{I}{2\Pi C_{OUT} \cdot R_L}$ with C<sub>OUT</sub> 470 µF and R<sub>L</sub> = 4 ohm, then F<sub>L</sub> = 80 Hz.

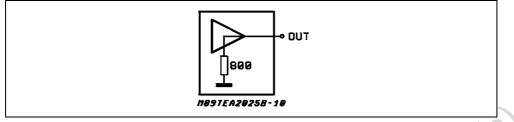


ndu

### 2.5 Pop noise

Most amplifiers similar to the TEA2025B need external resistors between the DC outputs and ground in order to minimize pop on/off noise and crossover distortion.

### Figure 11. Internal resistor



The TEA2025B requires less components as these resistors (800 ohm) are in the device.

### 2.6 Stability

Jbsolete P

A good layout is recommended in order to avoid oscillations.

Generally the designer must pay attention to the following points:

- Short wires of components and short connections.
- No ground loops
- Bypass of supply voltage with capacitors as close as possible to the supply IC pin. The low value (polyester) capacitors must have suitable temperature and frequency characteristics.
- No sockets

The heatsink can have a smaller factor of safety compared with that of a conventional circuit. There is no device damage in the case of excessive junction temperature:  $P_O$  (and therefore  $P_{tot}$ ) and  $I_d$  are simply reduced.



## 3 Application suggestions

The recommended values of the components are those shown in the stereo application circuit of *Figure 4*, although different values can be used (refer to the following table).

Component	Recommended value	Purpose	Larger than	Smaller than
C1, C2	0.22 µF	Input DC decoupling in case of slider contact noise of variable resistor		
C3	100 µF	Ripple rejecton		Degradation of SVR, increase of a low frequency and low voltage
C4, C5	100 µF	Bootstrap		200
C6, C7	470 μF	Output DC decoupling		Increase of low frequency cutoff
C8, C9	0.15 µF	Frequency stability		Danger of oscillations
C10, C11	100 µF	Inverting input DC decoupling		Increase of low frequency cutoff
	I	005	oler	
	Prof	Inverting input DC decoupling	0167	

Table 5. Recommended values for stereo applications

## 4 Package mechanical data

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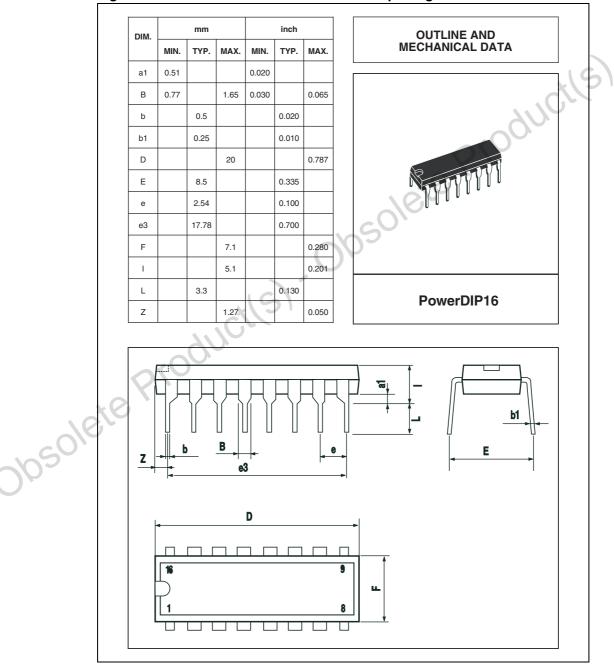


Figure 12. PowerDIP16 mechanical data and package dimensions



## 5 Revision history

### Table 6. Revision history

	Date	Revision	Changes
	September 2003	2	Updates not recorded
	30-Apr-2010	3	Updated title and added environmental compliance statement for package
	01-Oct-2012	4	Removed SO20 package option from datasheet Minor textual updates Revised document presentation
opsole	teprod	ucils	obsolete



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