

CMOS LDO Regulators for Automotive

1ch 200mA
CMOS LDO Regulators

BUxxJA2VG-C series

General Description

BUxxJA2VG-C series are high-performance CMOS LDO regulators with output current ability of up to 200mA. The SSOP5 package can contribute to the downsizing of the set. These devices have excellent noise and load response characteristics despite of its low circuit current consumption of 33 μ A. They are most appropriate for various applications such as power supplies for radar modules and camera modules.

Features

- AEC-Q100 qualified^(Note 1)
- High Output Voltage Accuracy: $\pm 2.0\%$
(In all recommended conditions)
- High Ripple Rejection: 68 dB (Typ, 1kHz)
- Compatible with small ceramic capacitor
($C_{in}=C_{out}=0.47\mu F$)
- Low Current Consumption: 33 μ A
- Output Voltage ON/OFF control
- Built-in Over Current Protection Circuit (OCP)
- Built-in Thermal Shutdown Circuit (TSD)
- Package SSOP5 is similar to SOT23-5(JEDEC)

(Note 1) Grade1

Applications

- Automotive Radar modules
- Automotive Camera modules

Typical Application Circuit

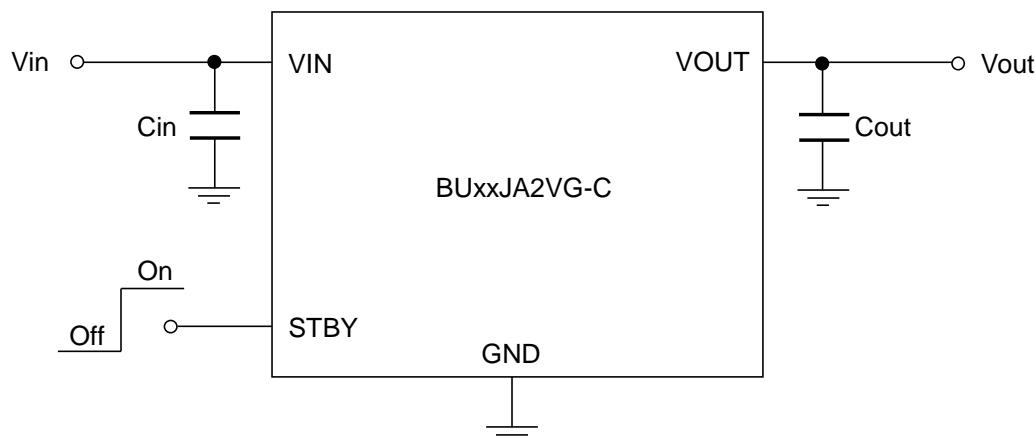


Figure 1. Typical Application Circuit

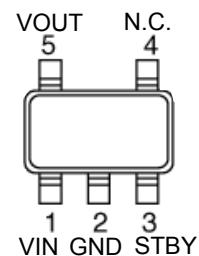
Ordering Information

B	U	X	X	J	A	2	V	G	-	C	G	Y	Y
Part Number	Output Voltage 10 : 1.0V 12 : 1.2V 1C : 1.25V 15 : 1.5V 18 : 1.8V 25 : 2.5V 28 : 2.8V 2d : 2.85V 30 : 3.0V 33 : 3.3V	Series name Maximum Output Current : 200mA Maximum Power Supply Voltage Range : 6.5V High-speed load response, Low noise, Shutdown SW	Package G : SSOP5	Product Rank C : for Automotive	Manufacturing Code	Packageing and forming specification Embossed tape and reel TR : The pin number 1 is the upper right TL (Note 1) : The pin number 1 is the lower left							

(Note 1) Only xx=18 and 33 models support TL version.

Pin Description^(Note 2)

Pin No.	Symbol	Function
1	VIN	Input Pin
2	GND	GND Pin
3	STBY	Output Control Pin (High:ON, Low:OFF)
4	N.C.	No Connect
5	VOUT	Output Pin



(Note 2) N.C. Pin can be open because it isn't connecting it inside of IC.

Block Diagram

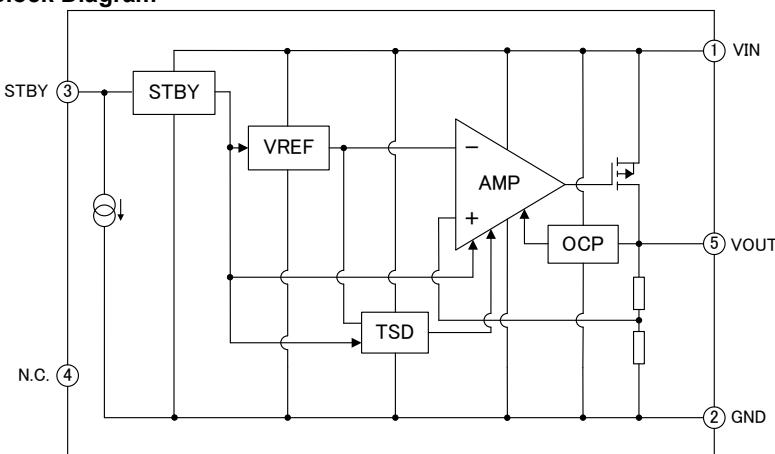


Figure 2. Block diagram

Block	Function	Description
STBY	Control Standby mode	STBY controls internal block active and standby state
VREF	Internal Reference Voltage	VREF generates reference voltage.
AMP	Error AMP	AMP amplifies electric signal and drives output power transistor.
OCP	Over Current Protection	When output current exceeds current ability, OCP restricts Output Current.
TSD	Thermal Shutdown	When Junction temperature rise and exceed Maximum junction temperature, TSD turns off Output power transistor.

Absolute Maximum Ratings

Parameter	Symbol	Rating	Unit
Power Supply Voltage Range	V _{IN}	-0.3 to +6.5 ^(Note1)	V
STBY Voltage	V _{STBY}	-0.3 to +6.5	V
Junction Temperature	T _{jmax}	+150	°C
Operating Temperature Range	T _{opr}	-40 to +125	°C
Storage Temperature Range	T _{stg}	-55 to +150	°C

(Note 1) Not to exceed T_{jmax}

Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Recommended Operating Ratings(Ta=-40°C to +125°C)

Parameter	Symbol	Limit			Unit
Power Supply Voltage Range	V _{IN}	1.7 to 6.0			V
STBY voltage	V _{STBY}	1.7 to 6.0			V
Maximum Output Current	I _{OUTMAX}	200			mA

Recommended Operating Conditions

Parameter	Symbol	Rating			Unit	Conditions
		Min	Typ	Max		
Input capacitor	C _{in}	0.47 ^(Note2)	1.0	—	μF	A ceramic capacitor is recommended.
Output capacitor	C _{out}	0.47 ^(Note2)	1.0	—	μF	A ceramic capacitor is recommended.

(Note 2) Set the value of the capacitor so that it does not fall below the minimum value.

Take into consideration the temperature characteristics, DC device characteristics and degradation with time.

Thermal Resistance (Note 3)

Parameter	Symbol	Thermal Resistance (Typ)		Unit
		1s ^(Note 5)	2s2p ^(Note 6)	
SSOP5				
Junction to Ambient	θ _{JA}	376.5	185.4	°C/W
Junction to Top Characterization Parameter ^(Note 4)	Ψ _{JT}	40	30	°C/W

(Note 3)Based on JESD51-2A(Still-Air).

(Note 4)The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

(Note 5)Using a PCB board based on JESD51-3.

Layer Number of Measurement Board	Material	Board Size	
Single	FR-4	114.3mm x 76.2mm x 1.57mm	
Top			
Copper Pattern			
Footprints and Traces			70μm

(Note 6)Using a PCB board based on JESD51-7.

Layer Number of Measurement Board	Material	Board Size	
4 Layers	FR-4	114.3mm x 76.2mm x 1.6mm	
Top		2 Internal Layers	
Copper Pattern		Copper Pattern	Thickness
Footprints and Traces		74.2mm x 74.2mm	35μm
Bottom		Copper Pattern	
Copper Pattern		74.2mm x 74.2mm	70μm
Footprints and Traces		74.2mm x 74.2mm	70μm

Electrical Characteristics

(Unless otherwise noted, $T_a = -40$ to 125°C , $V_{IN} = V_{OUT} + 1.0\text{V}$ ^(Note 1), $V_{STBY} = 1.5\text{V}$, $C_{in} = 1\mu\text{F}$, $C_{out} = 1\mu\text{F}$.The Typical value is defined at $T_a = 25^\circ\text{C}$)

Parameter	Symbol	Limit			Unit	Conditions
		MIN	TYP	MAX		
Output Voltage	V_{OUT}	$V_{OUT} \times 0.98$	V_{OUT}	$V_{OUT} \times 1.02$	V	$I_{OUT} = 0$ to 200mA $V_{OUT} > 2.5\text{V}$, $V_{IN} = V_{OUT} + 0.5$ to 6.0V $V_{OUT} \leq 2.5\text{V}$, $V_{IN} = 3.0$ to 6.0V
Line Regulation	V_{DLI}	-	4	15	mV	$I_{OUT} = 10\text{mA}$ $V_{OUT} \leq 2.5\text{V}$, $V_{IN} = 3.0$ to 6.0V
		-	6	20	mV	$I_{OUT} = 10\text{mA}$ $V_{OUT} > 2.5\text{V}$, $V_{IN} = V_{OUT} + 0.5$ to 6.0V
Load Regulation1	V_{DLO1}	-	0.5	5	mV	$I_{OUT} = 1$ to 100mA
Load Regulation2	V_{DLO2}	-	1	10	mV	$I_{OUT} = 1$ to 200mA
Dropout Voltage	V_{DROP}	-	160	315	mV	$V_{OUT} = 1.8\text{V}$, $I_{OUT} = 100\text{mA}$
		-	100	190	mV	$V_{OUT} = 2.5\text{V}$, $I_{OUT} = 100\text{mA}$
		-	85	155	mV	$V_{OUT} \geq 2.8\text{V}$, $I_{OUT} = 100\text{mA}$
Maximum Output Current	I_{OUTMAX}	200	-	-	mA	$V_{IN} = V_{OUT} + 1.0\text{V}$ ^(Note 1)
Limit Current	I_{LMAX}	250	400	-	mA	applied $V_{OUT} \times 0.98$ for V_{OUT} Pin, $T_a = 25^\circ\text{C}$
Short Current	I_{SHORT}	-	100	200	mA	$V_{OUT} = 0\text{V}$, $T_a = 25^\circ\text{C}$
Circuit Current	I_{GND}	-	33	80	μA	$I_{OUT} = 0\text{mA}$
Circuit Current (STBY)	I_{CCST}	-	-	2.0	μA	$V_{STBY} = 0\text{V}$
Ripple Rejection Ratio	R.R.	-	68	-	dB	$V_{RR} = -20\text{dBV}$, $f_{RR} = 1\text{kHz}$ $I_{OUT} = 10\text{mA}$, $T_a = 25^\circ\text{C}$
Load Transient Response	V_{LOT}	-	± 65	-	mV	$I_{OUT} = 1$ to 150mA , $T_{rise} = T_{fall} = 1\mu\text{s}$ $V_{IN} = V_{OUT} + 1.0\text{V}$, $T_a = 25^\circ\text{C}$
Line Transient Response	V_{LIT}	-	± 5	-	mV	$V_{IN} = V_{OUT} + 0.5$ to $V_{OUT} + 1.0\text{V}$ $T_{rise} = T_{fall} = 10\mu\text{s}$, $T_a = 25^\circ\text{C}$
Output Noise Voltage	V_{NOISE}	-	30	-	μVrms	Bandwidth 10 to 100kHz , $T_a = 25^\circ\text{C}$
Startup Time	T_{ST}	-	100	300	μs	Output Voltage settled within tolerances ^(Note 2) , $T_a = 25^\circ\text{C}$
STBY Control Voltage	ON	V_{STBH}	1.1	-	V_{IN}	$T_a = 25^\circ\text{C}$
	OFF	V_{STBL}	-0.2	-	0.5	
STBY Pin Current	I_{STBY}	-	-	4.0	μA	

(Note 1) $V_{IN} = 3.0\text{V}$ for $V_{OUT} < 2.5\text{V}$.(Note 2) Startup time = time from EN assertion to $V_{OUT} \times 0.98$

Reference data BU18JA2VG-C (Unless otherwise specified, $T_a=25^\circ\text{C}$)

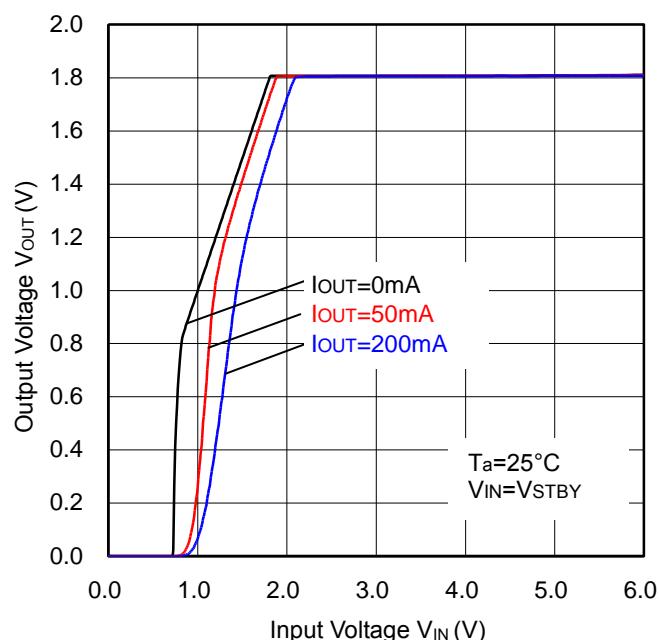


Figure 3. Output Voltage vs Input Voltage

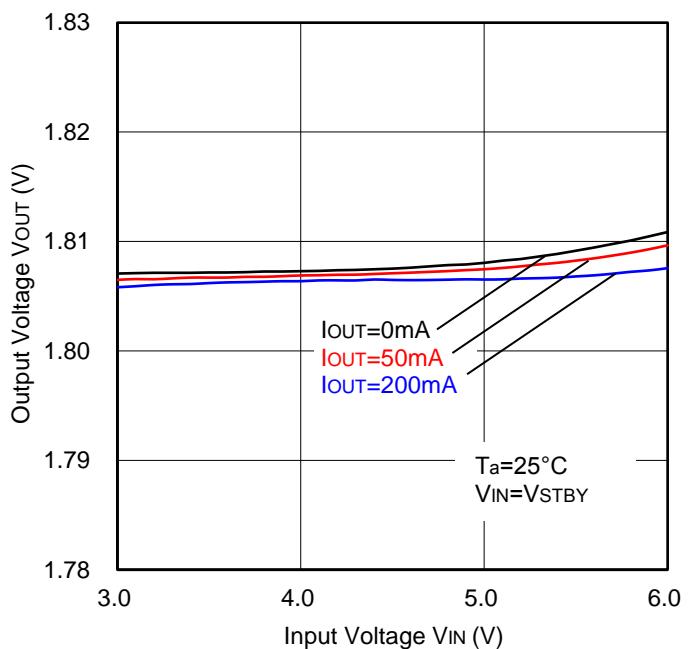


Figure 4. Line Regulation

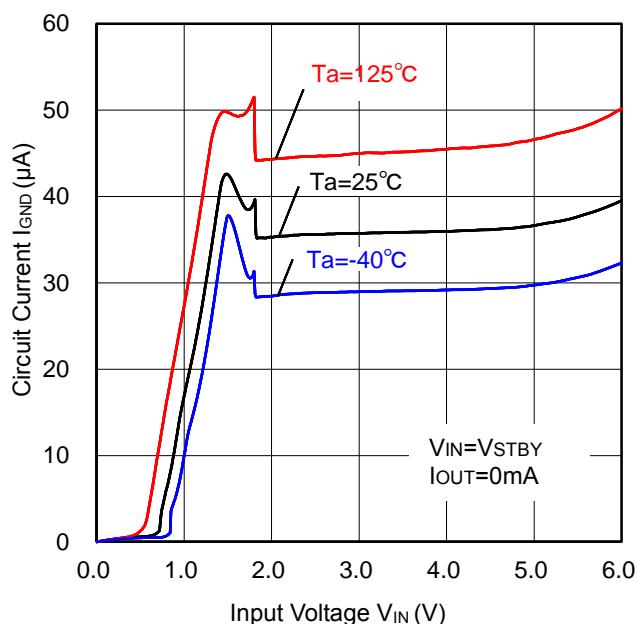


Figure 5. Circuit Current vs Input Voltage

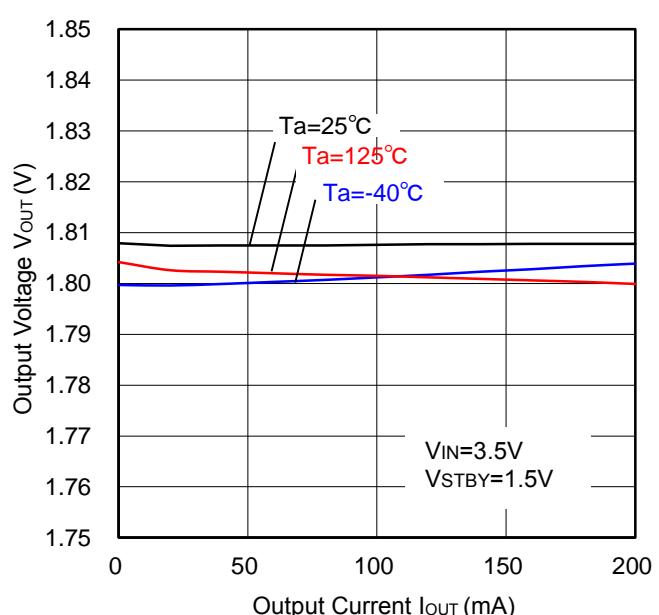


Figure 6. Load Regulation

Reference data BU18JA2VG-C (Unless otherwise specified, $T_a=25^\circ\text{C}$)

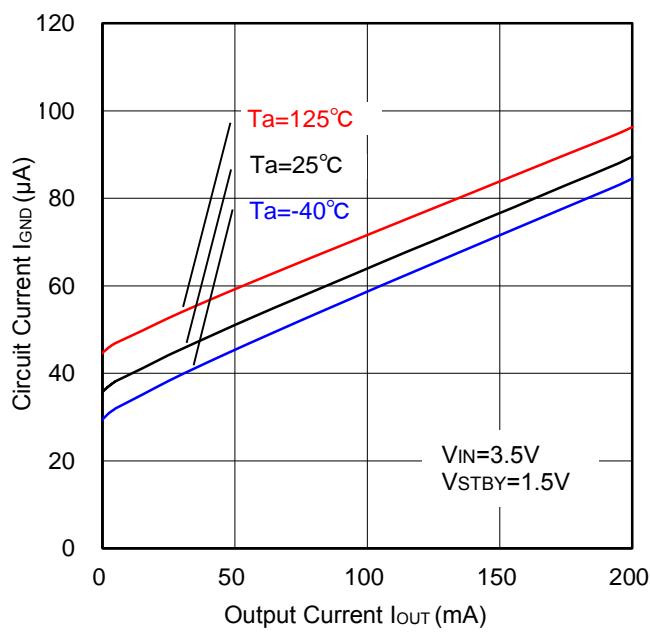


Figure 7. Circuit Current vs Output Current

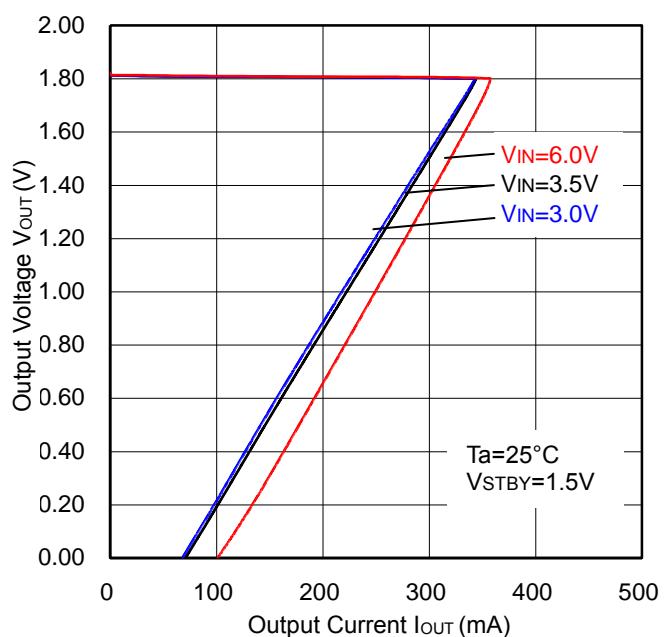


Figure 8. OCP Threshold

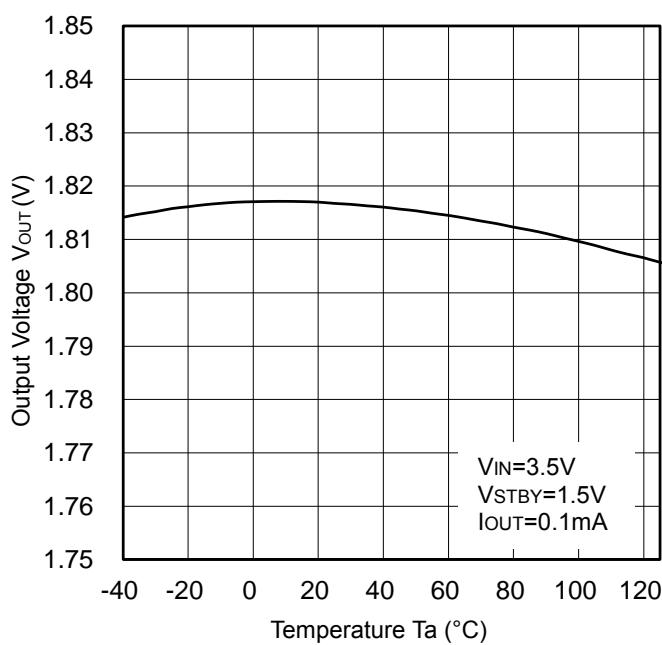


Figure 9. Output Voltage vs Temperature

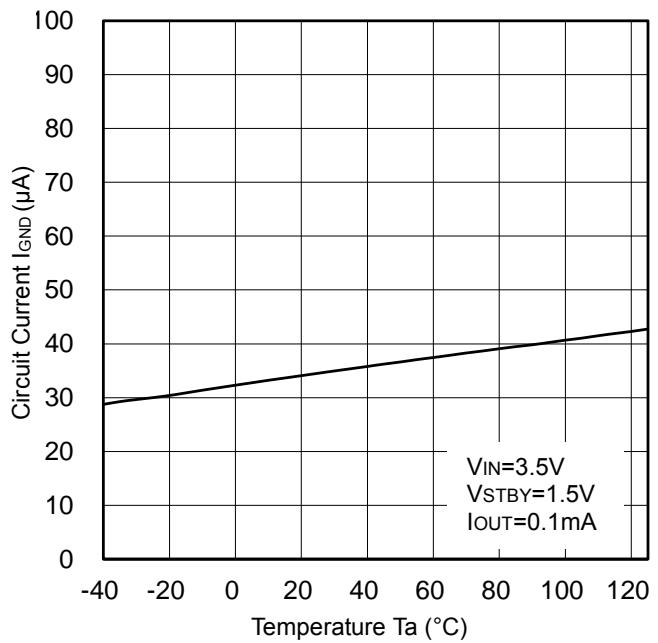


Figure 10. Circuit Current vs Temperature

Reference data BU18JA2VG-C (Unless otherwise specified, $T_a=25^\circ\text{C}$)

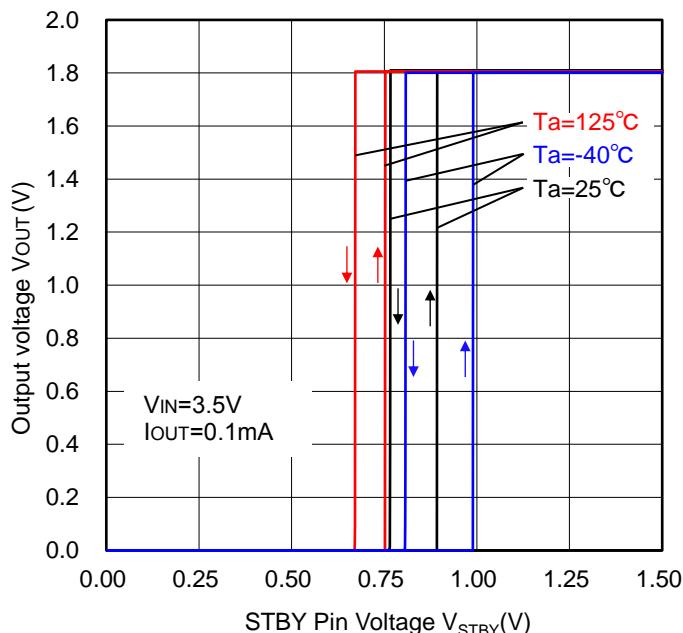


Figure 11. STBY Threshold

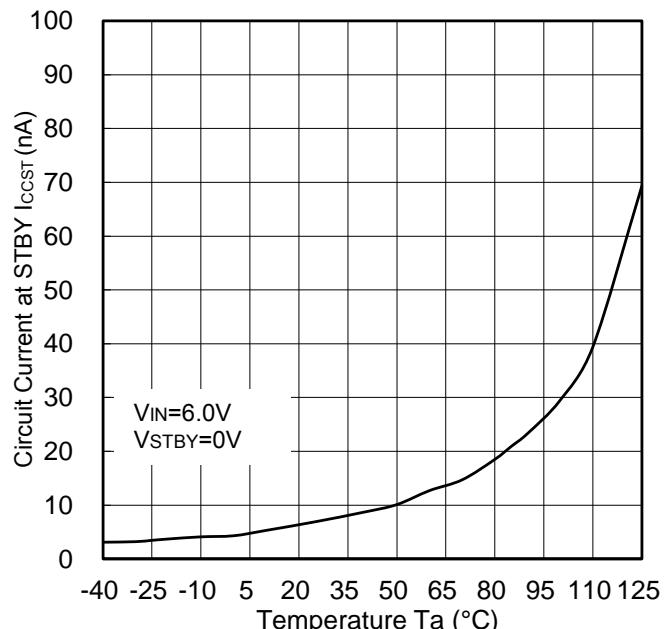


Figure 12. Circuit Current at STBY vs Temperature

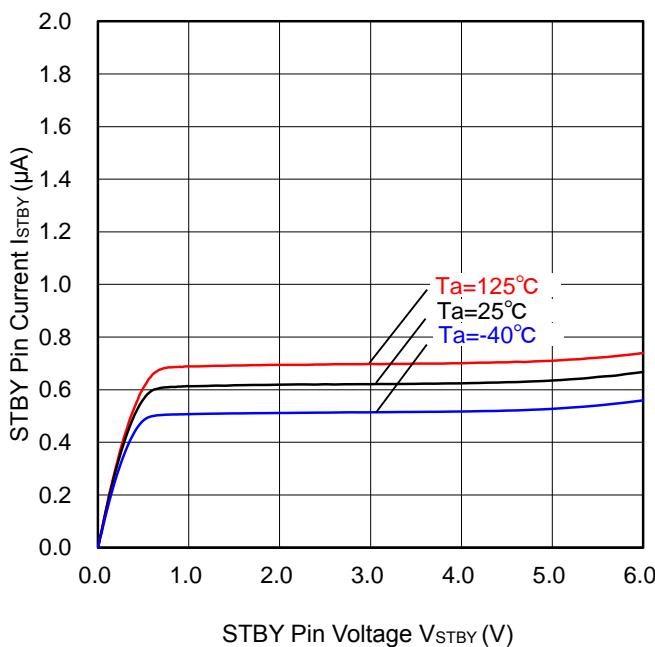


Figure 13. STBY Pin Current vs STBY Pin Voltage

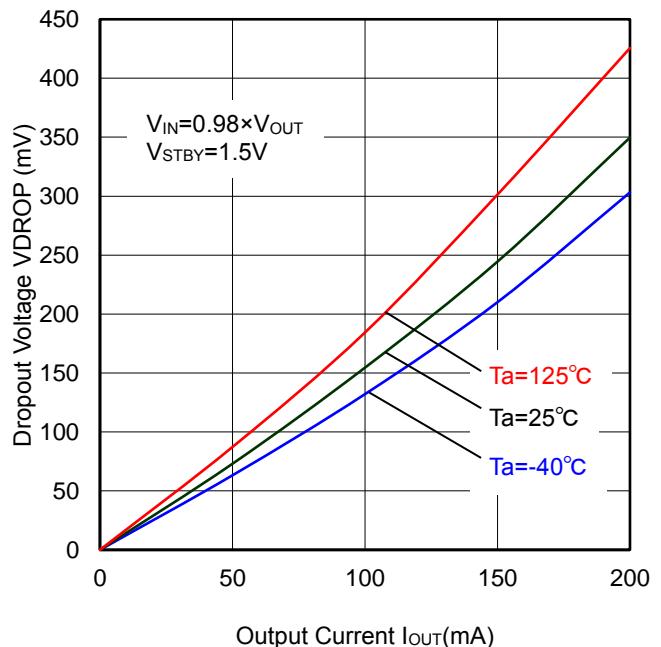


Figure 14. Dropout Voltage vs Output Current

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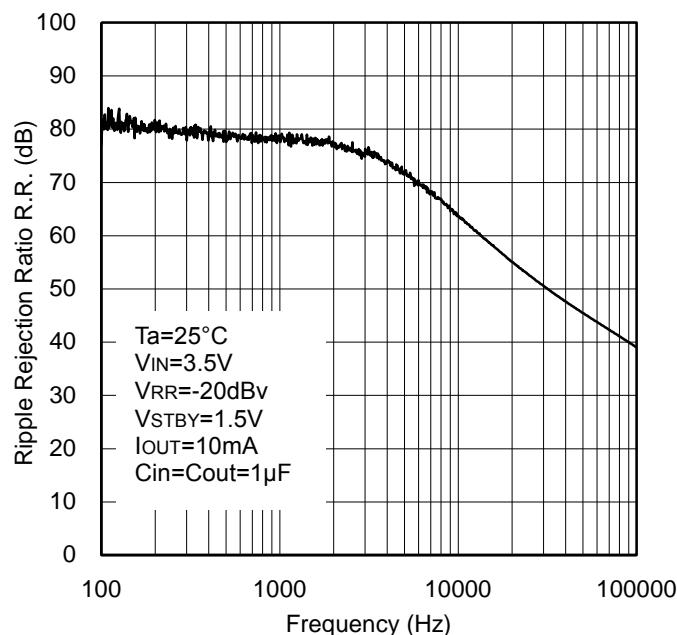


Figure 15. Ripple Rejection Ratio vs Frequency

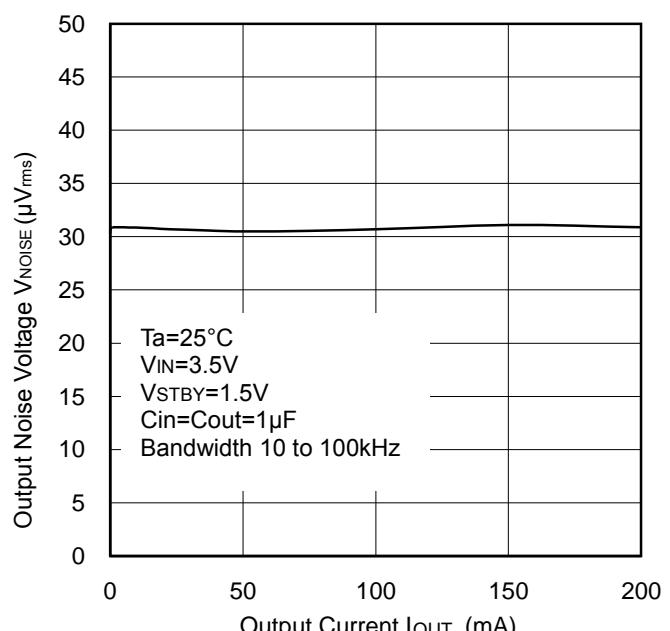


Figure 16. Output Noise Voltage vs Output Current

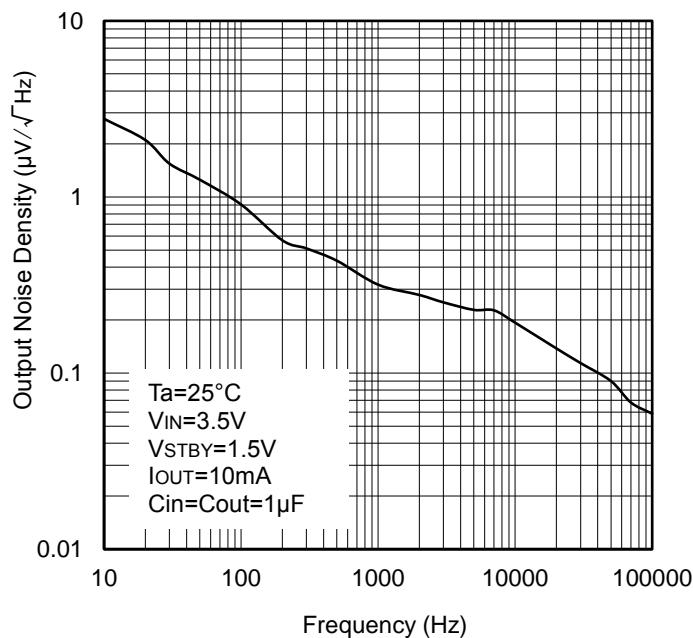
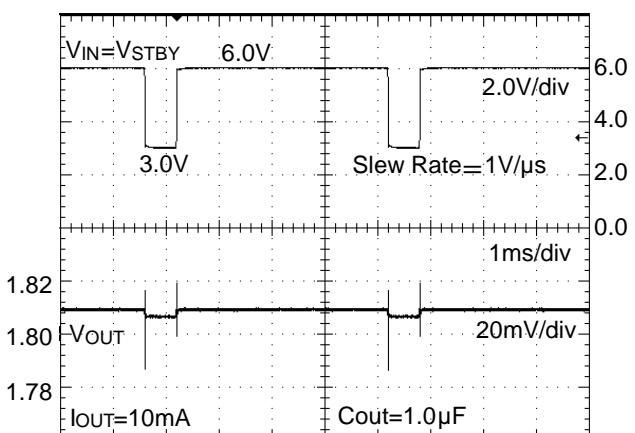
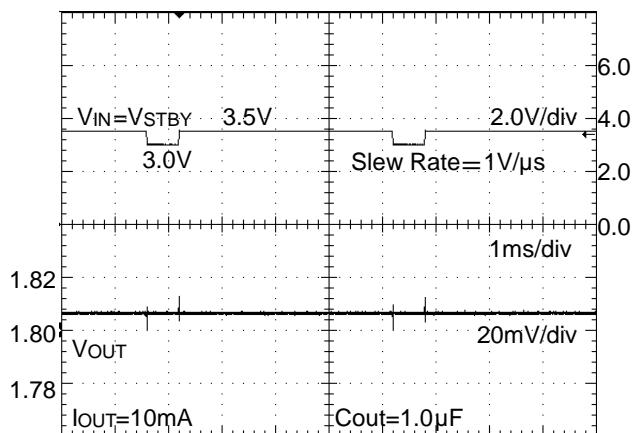
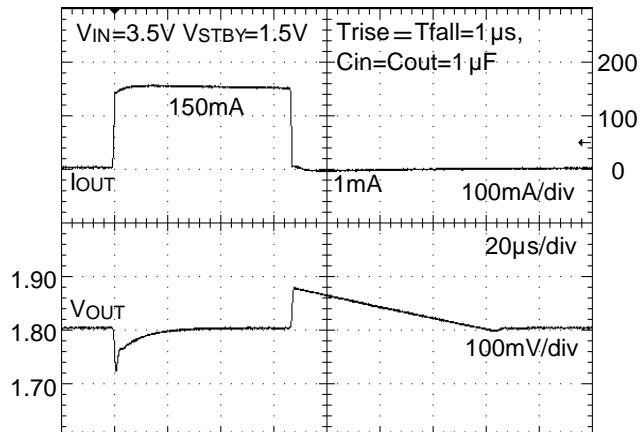
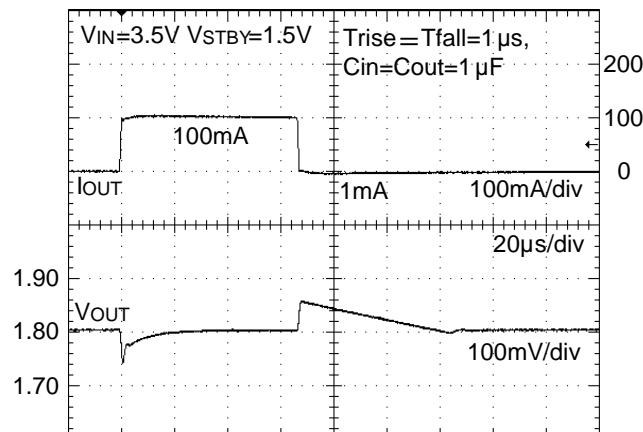


Figure 17. Output Noise Density vs Frequency

Reference data BU18JA2VG-C (Unless otherwise specified, $T_a=25^\circ\text{C}$)



Reference data BU18JA2VG-C (Unless otherwise specified, $T_a=25^\circ\text{C}$)

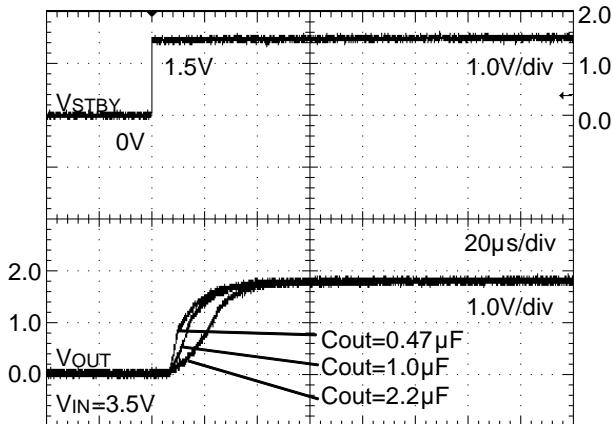


Figure 22. Startup Time
($\text{ROUT}=\text{open}$)

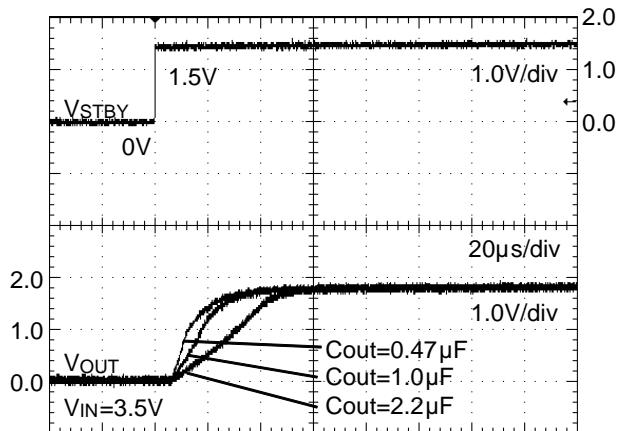


Figure 23. Startup Time
($\text{ROUT}=9\Omega$)

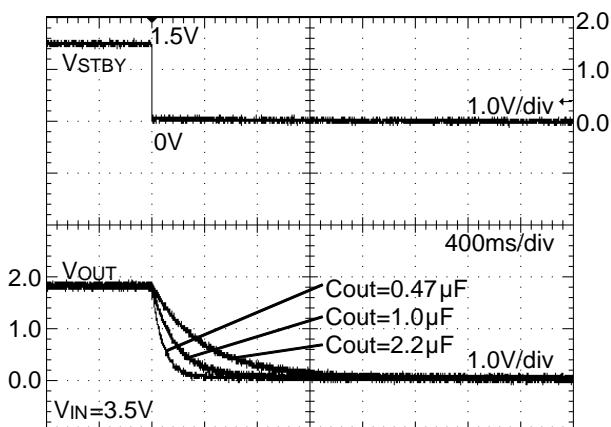


Figure 24. Discharge Time
($\text{ROUT}=\text{open}$)

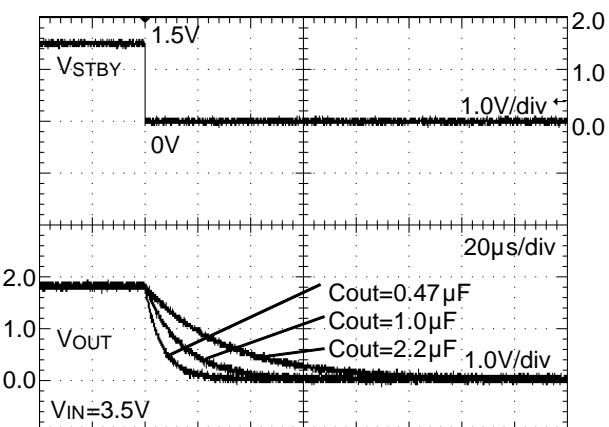
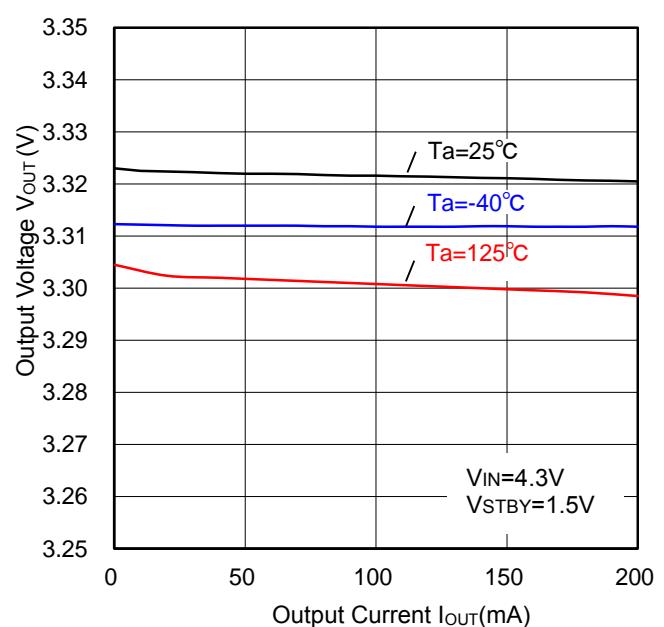
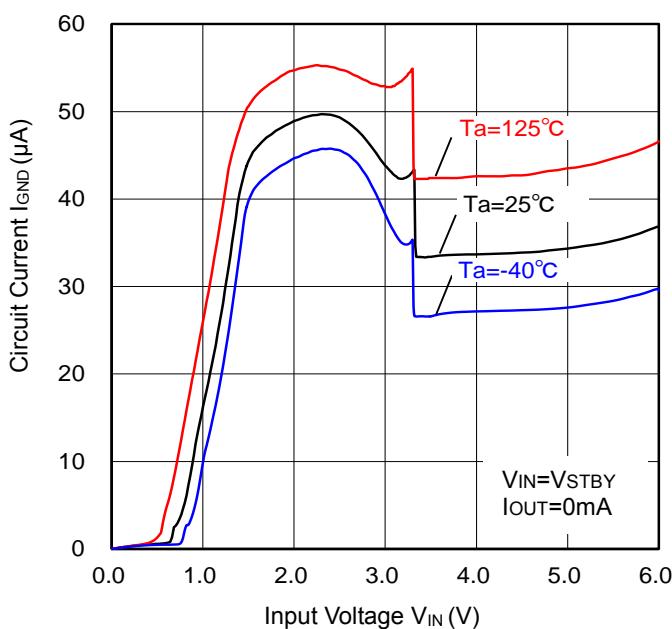
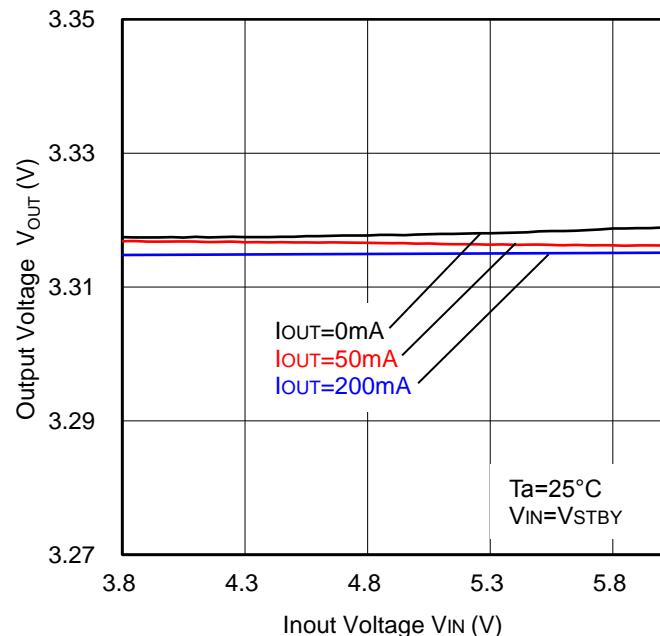
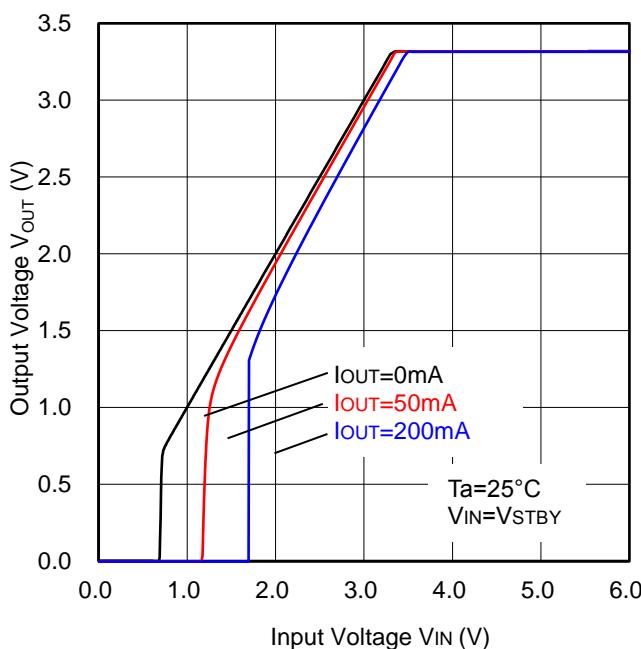


Figure 25. Discharge Time
($\text{ROUT}=9\Omega$)

Reference data BU33JA2VG-C (Unless otherwise specified, $T_a=25^\circ\text{C}$)



Reference data BU33JA2VG-C (Unless otherwise specified, $T_a=25^\circ\text{C}$)

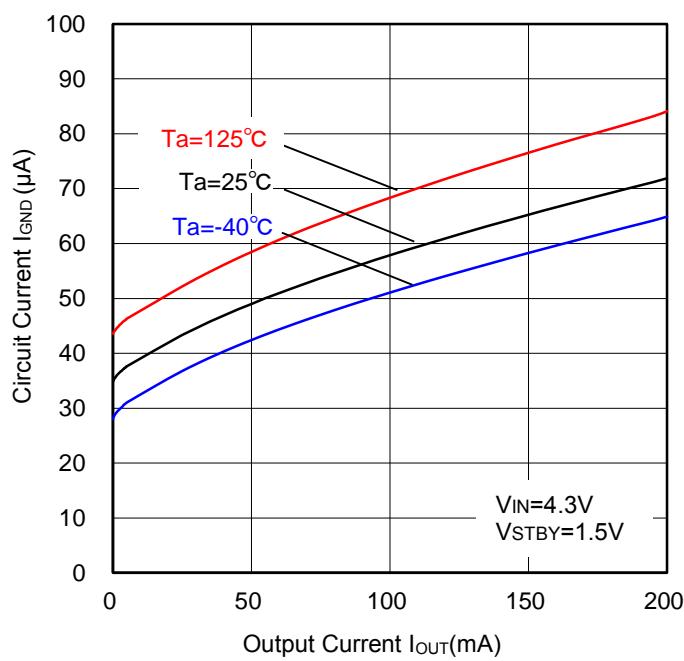


Figure 30. Circuit Current vs Output Current

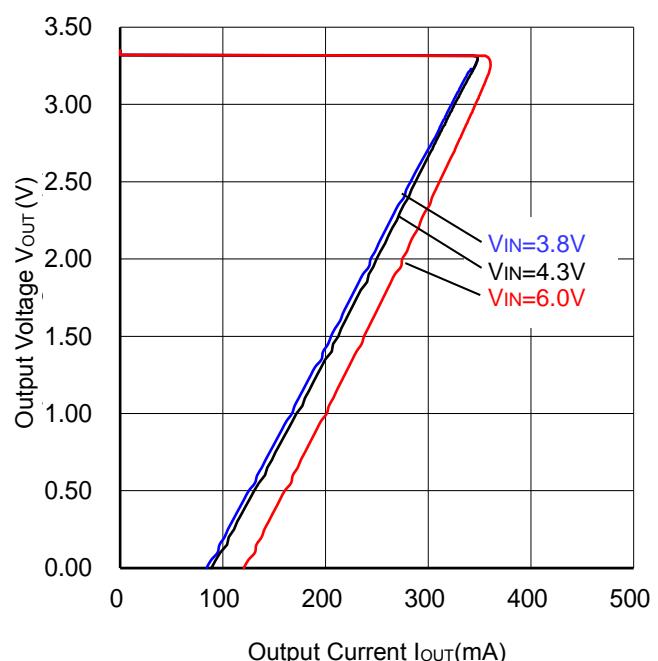


Figure 31. OCP Threshold

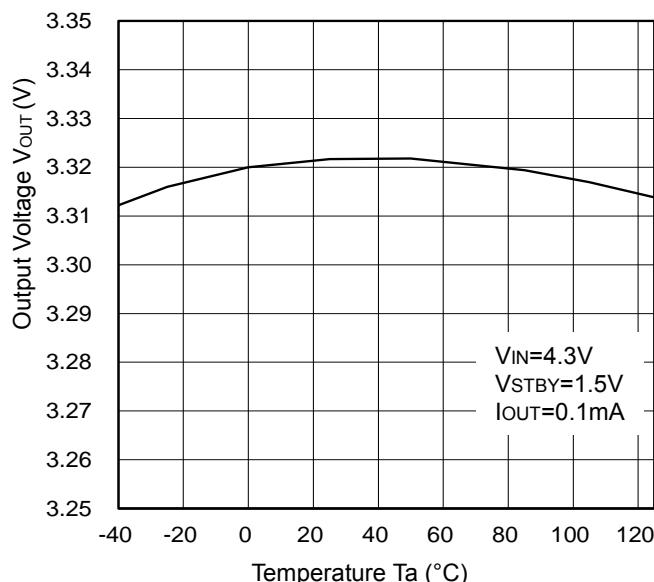


Figure 32. Output Voltage vs Temperature

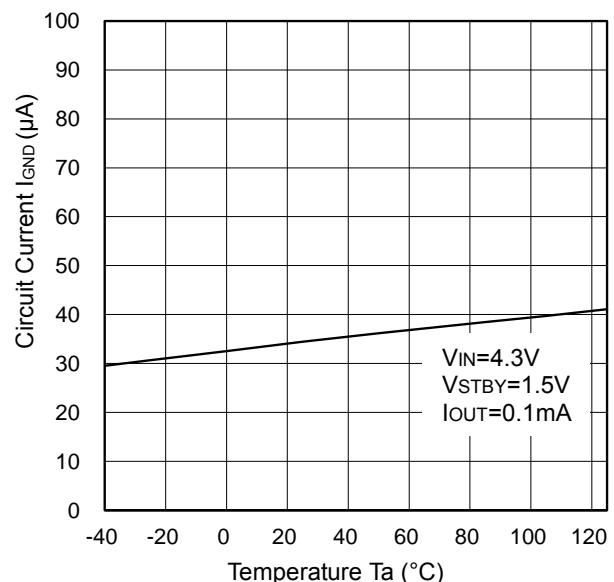


Figure 33. Circuit Current vs Temperature

Reference data BU33JA2VG-C (Unless otherwise specified, $T_a=25^\circ\text{C}$)

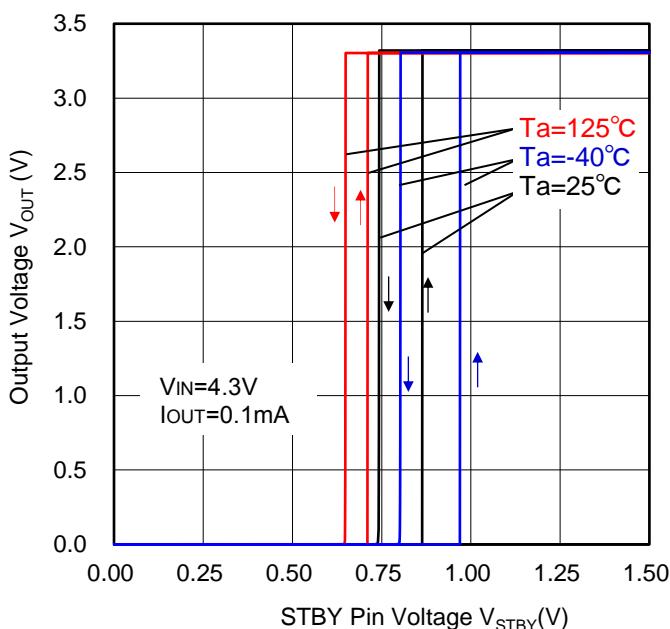


Figure 34. STBY Threshold

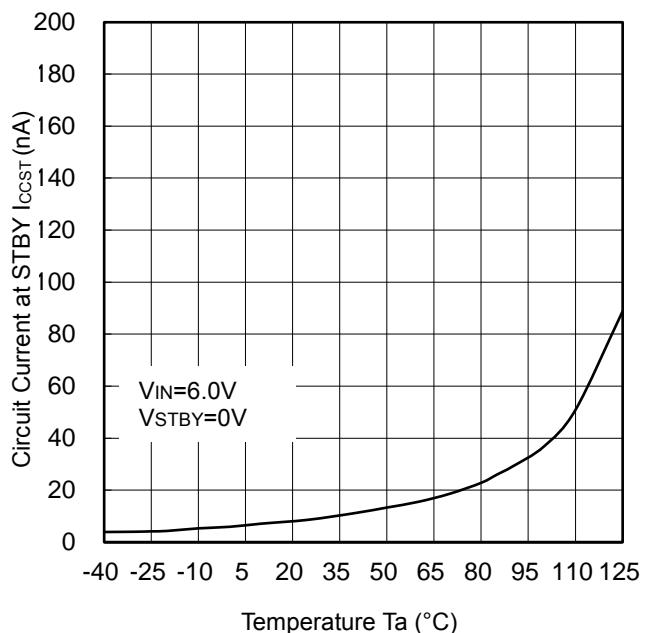


Figure 35. Circuit Current at STBY vs Temperature

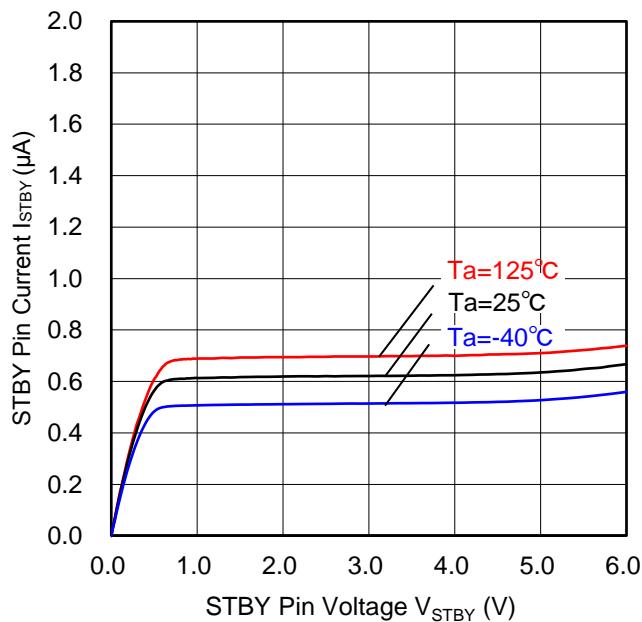


Figure 36. STBY Pin Current vs STBY Pin Voltage

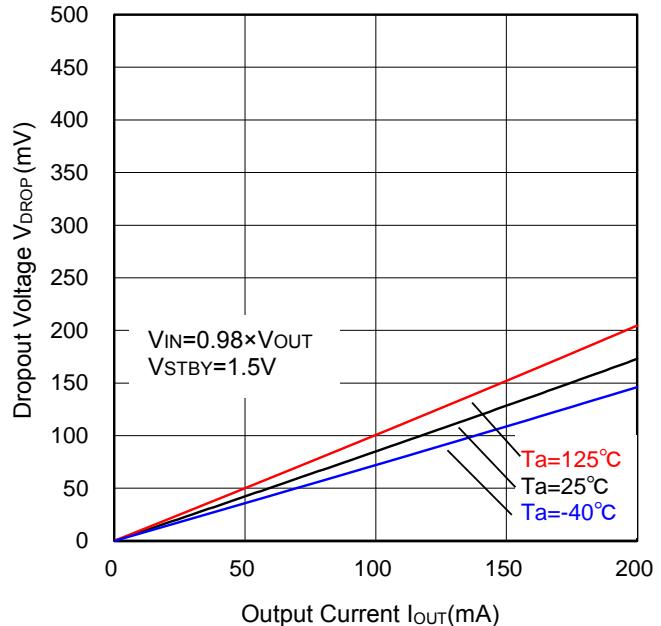


Figure 37. Dropout Voltage vs Output Current

Reference data BU33JA2VG-C (Unless otherwise specified, $T_a=25^{\circ}\text{C}$)

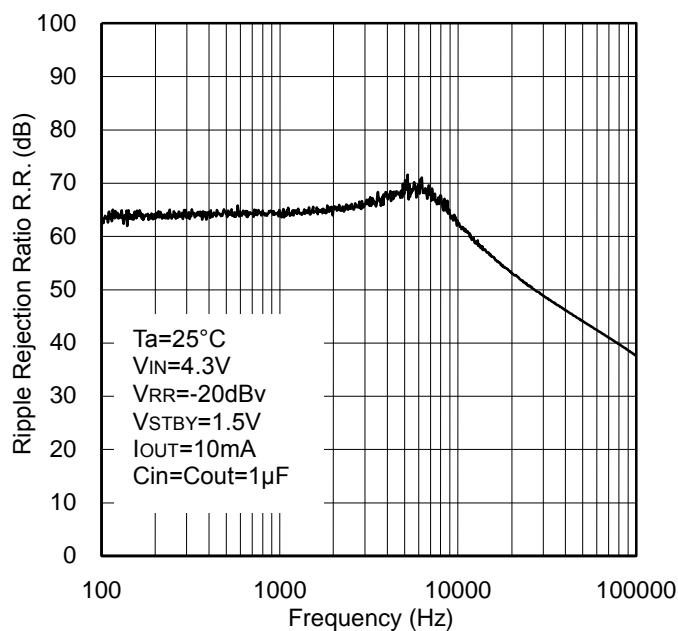


Figure 38. Ripple Rejection Ratio vs Frequency

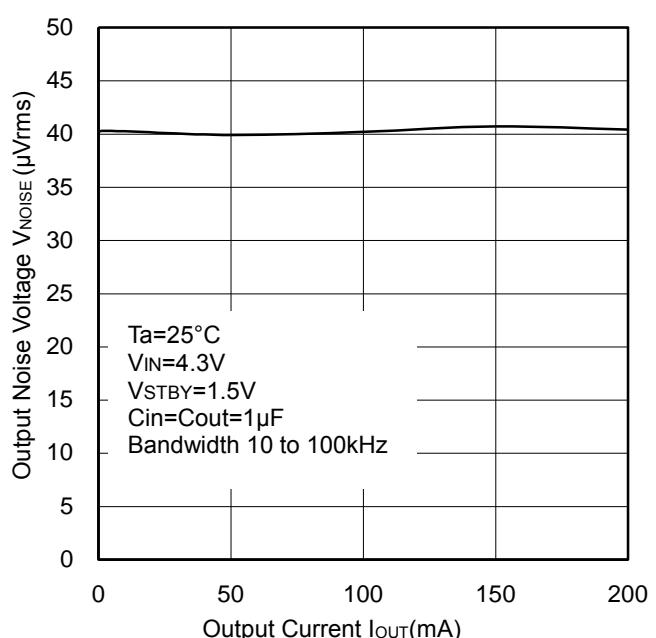


Figure 39. Output Noise Voltage vs Output Current

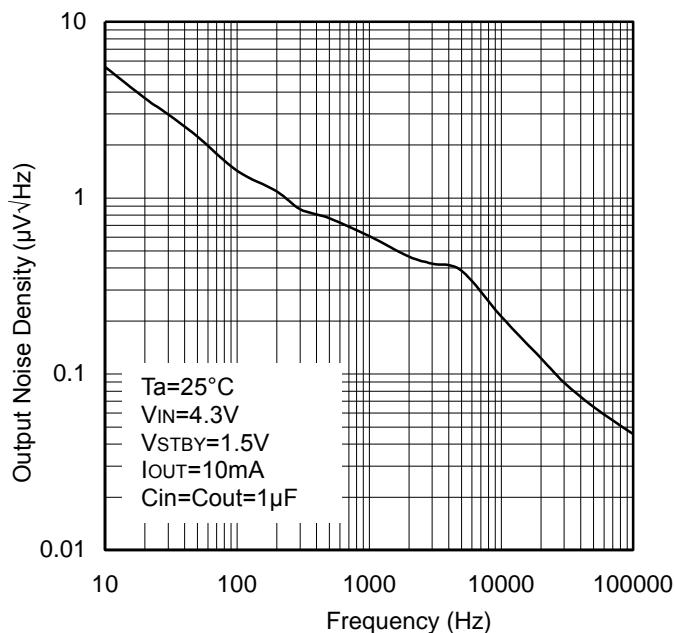
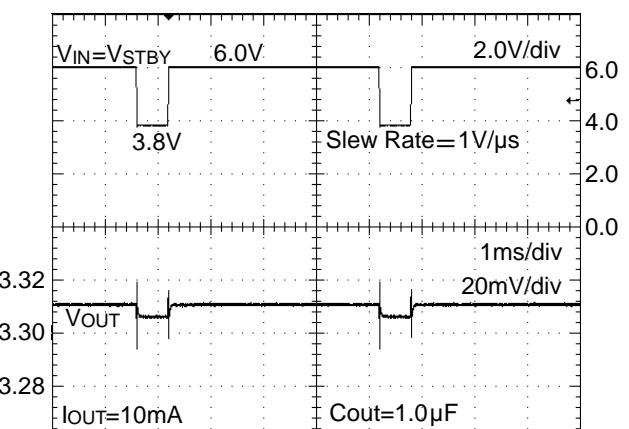
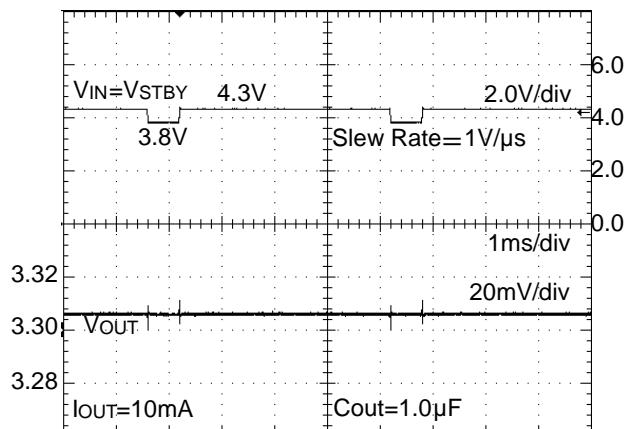
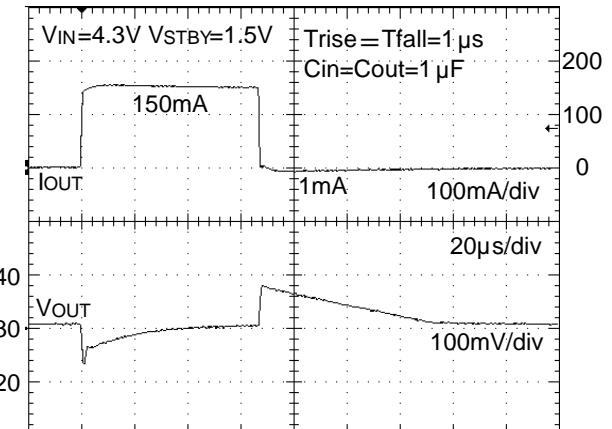
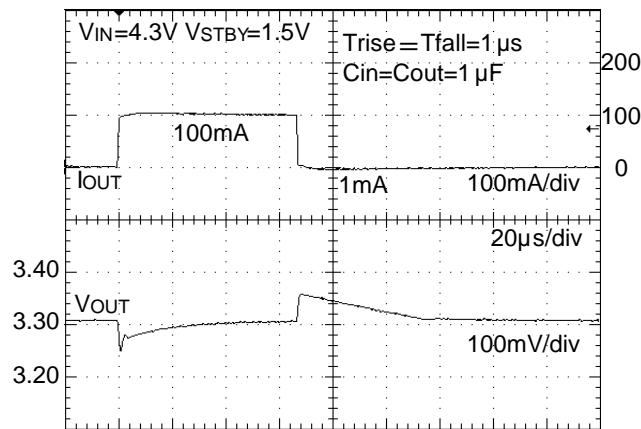


Figure 40. Output Noise Density vs Frequency

Reference data BU33JA2VG-C (Unless otherwise specified, $T_a=25^\circ\text{C}$)



Reference data BU33JA2VG-C (Unless otherwise specified, $T_a=25^\circ\text{C}$)

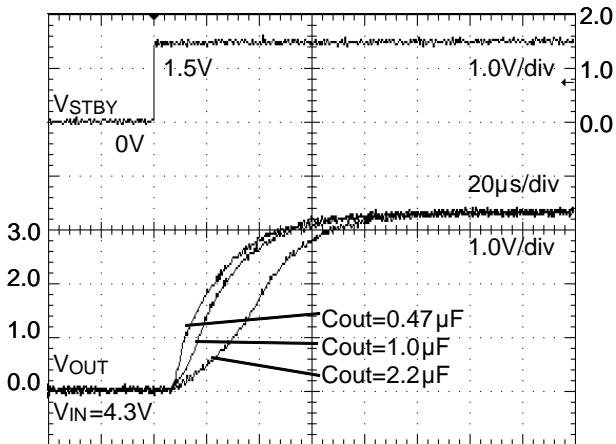


Figure 45. Startup Time (ROUT=open)

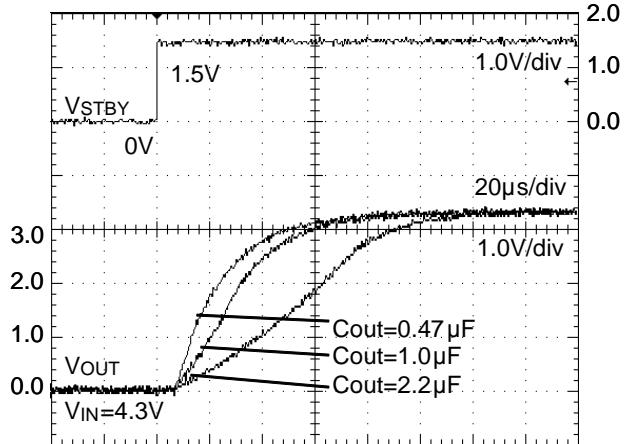


Figure 46. Startup Time
($R_{OUT}=16.5\Omega$)

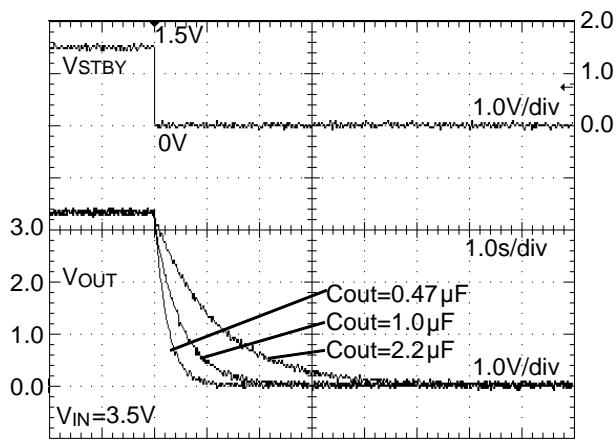


Figure 47. Discharge Time (ROUT=open)

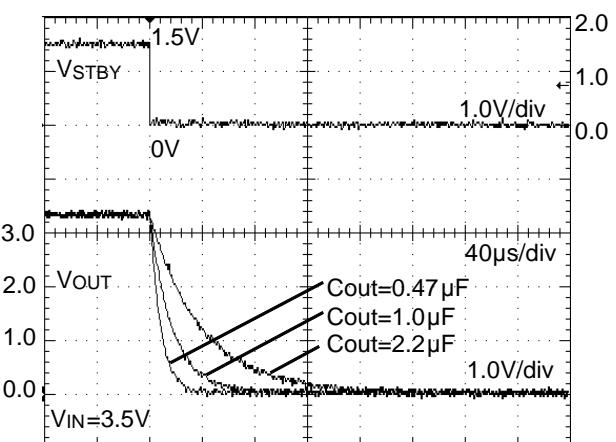


Figure 48. Discharge Time
($R_{OUT}=16.5\Omega$)

Input/Output Capacitor

It is recommended that a capacitor is placed close to pin between input pin and GND as well as output pin and GND. The input capacitor becomes more necessary when the power supply impedance is high or when the PCB trace has significant length. Moreover, the higher the capacitance of the output capacitor the more stable the output will be, even with load and line voltage variations. However, please check the actual functionality by mounting on a board for the actual application. Also, ceramic capacitors usually have different thermal and equivalent series resistance characteristics and may degrade gradually over continued use.

For additional details, please check with the manufacturer and select the best ceramic capacitor for your application.

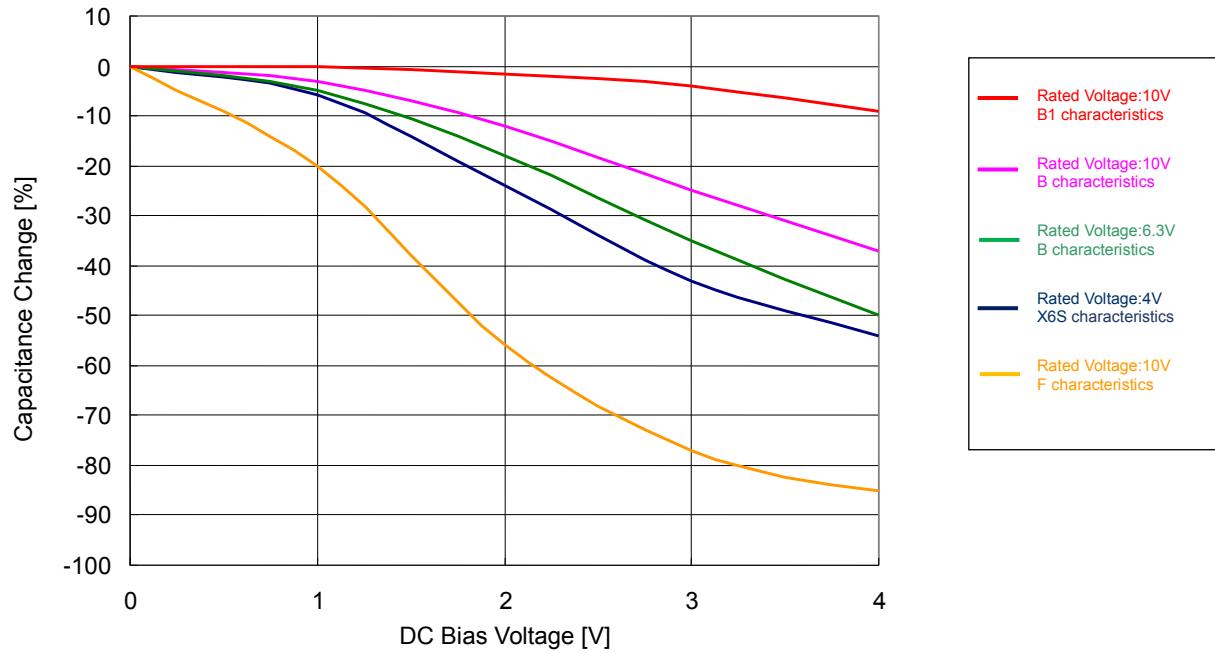


Figure 49. Ceramic Capacitor Capacitance Value vs DC Bias Characteristics
(Characteristics Example)

Equivalent Series Resistance (ESR) of a Ceramic Capacitor

To prevent oscillation, please attach a capacitor between VOUT and GND. Capacitors generally have ESR (equivalent series resistance) and it operates stably in the ESR-I_{OUT} area shown on the right. Since ceramic capacitors, tantalum capacitors, electrolytic capacitors, etc. generally have different ESR, please check the ESR of the capacitor to be used and use it within the stability area range shown in the right graph for evaluation of the actual application.

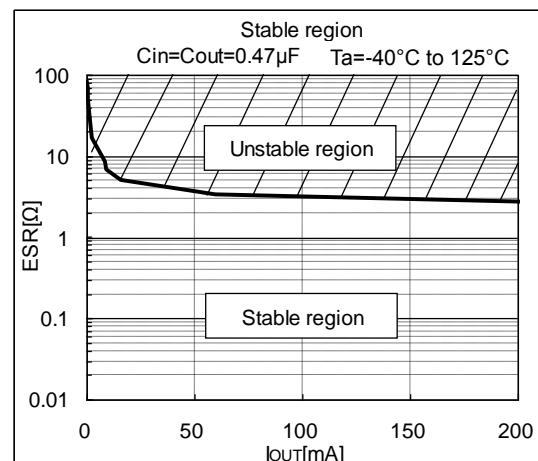


Figure 50. Stability area characteristics
(V_{IN}=1.7^(Note1) to 6.0V)
(Note1) Set V_{IN} voltage considering Dropout Voltage

Power Dissipation

■SSOP5

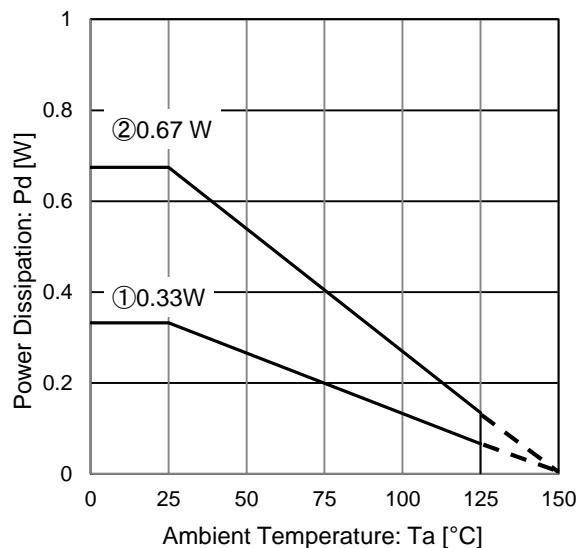


Figure 51. SSOP5 Package Data
(Reference Data)

IC mounted on ROHM standard board based on JEDEC.

① : 1-layer PCB
(Copper foil area on the reverse side of PCB: 0 mm × 0 mm)

Board material: FR4

Board size: 114.3 mm × 76.2 mm × 1.57 mm

Mount condition: PCB and exposed pad are soldered.

Top copper foil: ROHM recommended
footprint + wiring to measure, 2 oz. copper.

② : 4-layer PCB

(2 inner layers copper foil area of PCB, copper foil area on the
reverse side of PCB: 74.2 mm × 74.2 mm)

Board material: FR4

Board size: 114.3 mm × 76.2 mm × 1.6 mm

Mount condition: PCB and exposed pad are soldered.

Top copper foil: ROHM recommended
footprint + wiring to measure, 2 oz. copper.

2 inner layers copper foil area of PCB

: 74.2 mm × 74.2 mm, 1 oz. copper.

Copper foil area on the reverse side of PCB

: 74.2 mm × 74.2 mm, 2 oz. copper.

Condition①: $\theta_{JA} = 376.5 \text{ }^{\circ}\text{C/W}$, Ψ_{JT} (top center) = 40 $\text{ }^{\circ}\text{C/W}$

Condition②: $\theta_{JA} = 185.4 \text{ }^{\circ}\text{C/W}$, Ψ_{JT} (top center) = 30 $\text{ }^{\circ}\text{C/W}$

Thermal Design

Within this IC, the power consumption is decided by the dropout voltage condition, the load current and the circuit current. Refer to power dissipation curves illustrated in Figure 51 when using the IC in an environment of $T_a \geq 25^\circ\text{C}$. Even if the ambient temperature T_a is at 25°C , depending on the input voltage and the load current, chip junction temperature can be very high. Consider the design to be $T_j \leq T_{jmax} = 150^\circ\text{C}$ in all possible operating temperature range.

Should by any condition the maximum junction temperature $T_{jmax} = 150^\circ\text{C}$ rating be exceeded by the temperature increase of the chip, it may result in deterioration of the properties of the chip. The thermal impedance in this specification is based on recommended PCB and measurement condition by JEDEC standard. Verify the application and allow sufficient margins in the thermal design by the following method is used to calculate the junction temperature T_j .

T_j can be calculated by either of the two following methods.

1. The following method is used to calculate the junction temperature T_j .

$$T_j = T_a + P_c \times \theta_{JA}$$

Where:

T_j	: Junction Temperature
T_a	: Ambient Temperature
P_c	: Power Consumption
θ_{JA}	: Thermal Impedance (Junction to Ambient)

2. The following method is also used to calculate the junction temperature T_j .

$$T_j = T_r + P_c \times \Psi_{JT}$$

Where:

T_j	: Junction Temperature
T_r	: Top Center of Case's (mold) Temperature
P_c	: Power consumption
Ψ_{JT}	: Thermal Impedance (Junction to Top Center of Case)

The following method is used to calculate the power consumption P_c (W).

$$P_c = (V_{IN} - V_{OUT}) \times I_{OUT} + V_{IN} \times I_{GND}$$

Where:

P_c	: Power Consumption
V_{IN}	: Input Voltage
V_{OUT}	: Output Voltage
I_{OUT}	: Load Current
I_{GND}	: Circuit Current

- Calculation Example (SSOP5)

If $V_{IN} = 3.0$ V, $V_{OUT} = 1.8$ V, $I_{OUT} = 50$ mA, $I_{GND} = 33$ μ A, the power consumption P_C can be calculated as follows:

$$\begin{aligned}
 P_C &= (V_{IN} - V_{OUT}) \times I_{OUT} + V_{IN} \times I_{GND} \\
 &= (3.0 \text{ V} - 1.8 \text{ V}) \times 50 \text{ mA} + 3.0 \text{ V} \times 33 \mu\text{A} \\
 &= 0.06 \text{ W}
 \end{aligned}$$

At the ambient temperature $T_{max} = 125^\circ\text{C}$, the thermal Impedance (Junction to Ambient) $\theta_{JA} = 185.4$ $^\circ\text{C} / \text{W}$ (4-layer PCB),

$$\begin{aligned}
 T_j &= T_{max} + P_C \times \theta_{JA} \\
 &= 125^\circ\text{C} + 0.06 \text{ W} \times 185.4 \text{ }^\circ\text{C} / \text{W} \\
 &= 136.1^\circ\text{C}
 \end{aligned}$$

When operating the IC, the top center of case's (mold) temperature $T_T = 100$ $^\circ\text{C}$, $\Psi_{JT} = 40$ $^\circ\text{C} / \text{W}$ (1-layer PCB),

$$\begin{aligned}
 T_j &= T_T + P_C \times \Psi_{JT} \\
 &= 100^\circ\text{C} + 0.06 \text{ W} \times 40 \text{ }^\circ\text{C} / \text{W} \\
 &= 102.4^\circ\text{C}
 \end{aligned}$$

For optimum thermal performance, it is recommended to expand the copper foil area of the board, increasing the layer and thermal via between thermal land pad.

I/O Equivalence Circuits

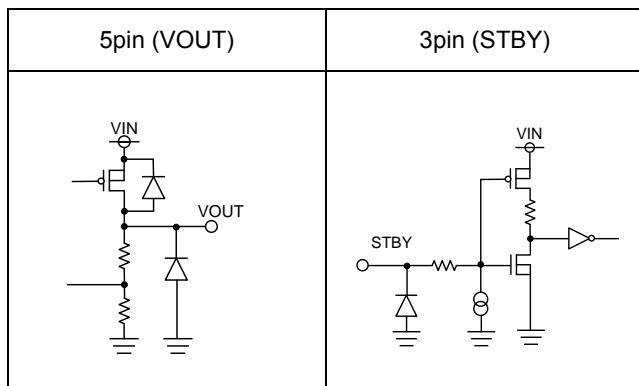


Figure 52. Input / Output equivalent circuit

Linear Regulators Surge Voltage Protection

The following provides instructions on surge voltage overs absolute maximum ratings polarity protection for ICs.

1. Applying positive surge to the input

If the possibility exists that surges higher than absolute maximum ratings 6.5 V will be applied to the input, a Zener Diode should be placed to protect the device in between the V_{IN} and the GND as shown in the figure 53.

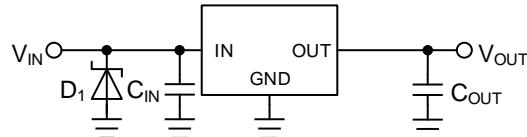


Figure 53. Surges Higher than 6.5 V will be Applied to the Input

2. Applying negative surge to the input

If the possibility exists that surges lower than absolute maximum ratings -0.3 V will be applied to the input, a Schottky Diode should be place to protect the device in between the V_{IN} and the GND as shown in the figure 54.

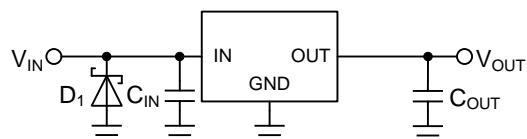


Figure 54. Surges Lower than -0.3 V will be Applied to the Input

Linear Regulators Reverse Voltage Protection

A linear regulator integrated circuit (IC) requires that the input voltage is always higher than the regulated voltage. Output voltage, however, may become higher than the input voltage under specific situations or circuit configurations, and that reverse voltage and current may cause damage to the IC. A reverse polarity connection or certain inductor components can also cause a polarity reversal between the input and output pins. The following provides instructions on reversed voltage polarity protection for ICs.

1. about Input /Output Voltage Reversal

In an MOS linear regulator, a parasitic element exists as a body diode in the drain-source junction portion of its power MOSFET. Reverse input/output voltage triggers the current flow from the output to the input through the body diode. The inverted current may damage or destroy the semiconductor elements of the regulator since the effect of the parasitic body diode is usually disregarded for the regulator behavior (Figure 55).

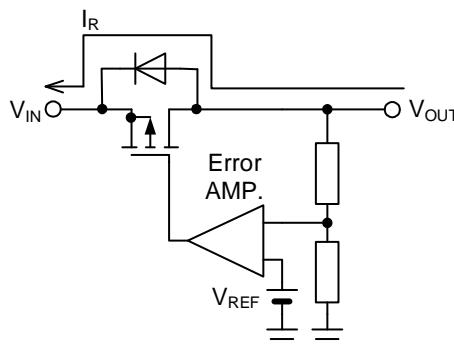


Figure 55. Reverse Current Path in an MOS Linear Regulator

An effective solution to this is an external bypass diode connected in-between the input and output to prevent the reverse current flow inside the IC (see Figure 56). Note that the bypass diode must be turned on before the internal circuit of the IC. Bypass diodes in the internal circuits of MOS linear regulators must have low forward voltage V_F . Some ICs are configured with current-limit thresholds to shut down high reverse current even when the output is off, allowing large leakage current from the diode to flow from the input to the output; therefore, it is necessary to choose one that has a small reverse current. Specifically, select a diode with a rated peak inverse voltage greater than the input to output voltage differential and rated forward current greater than the reverse current during use.

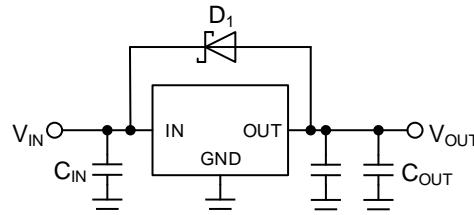


Figure 56. Bypass Diode for Reverse Current Diversion

The lower forward voltage (V_F) of Schottky barrier diodes cater to requirements of MOS linear regulators, however the main drawback is found in the level of their reverse current (I_R), which is relatively high. So, one with a low reverse current is recommended when choosing a Schottky diode. The V_R - I_R characteristics versus temperatures show increases at higher temperatures.

If V_{IN} is open in a circuit as shown in the following Figure 57 with its input/output voltage being reversed, the only current that flows in the reverse current path is the bias current of the IC. Because the amperage is too low to damage or destroy the parasitic element, a reverse current bypass diode is not required for this type of circuit.

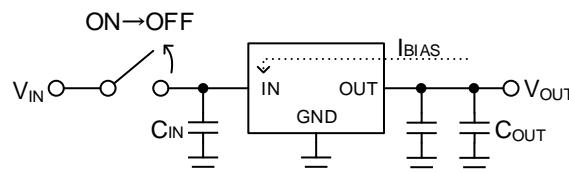


Figure 57. Open V_{IN}

2. Protection against Input Reverse Voltage

Accidental reverse polarity at the input connection flows a large current to the diode for electrostatic breakdown protection between the input pin of the IC and the GND pin, which may destroy the IC (see Figure 58).

A Schottky barrier diode or rectifier diode connected in series with the power supply as shown in Figure 59 is the simplest solution to prevent this from happening. The solution, however, is unsuitable for a circuit powered by batteries because there is a power loss calculated as $V_F \times I_{OUT}$, as the forward voltage V_F of the diode drops in a correct connection. The lower V_F of a Schottky barrier diode than that of a rectifier diode gives a slightly smaller power loss. Because diodes generate heat, care must be taken to select a diode that has enough allowance in power dissipation. A reverse connection allows a negligible reverse current to flow in the diode.

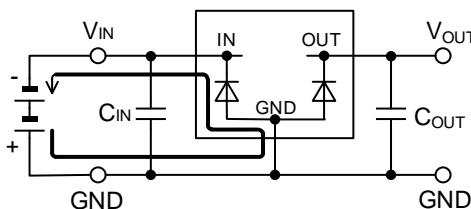


Figure 58. Current Path in Reverse Input Connection

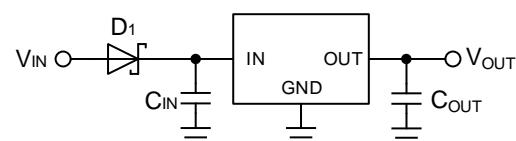


Figure 59. Protection against Reverse Polarity 1

Figure 60 shows a circuit in which a P-channel MOSFET is connected in series with the power. The diode located in the drain-source junction portion of the MOSFET is a body diode (parasitic element). The voltage drop in a correct connection is calculated by multiplying the resistance of the MOSFET being turned on by the output current I_{OUT} , therefore it is smaller than the voltage drop by the diode (see Figure 59) and results in less of a power loss. No current flows in a reverse connection where the MOSFET remains off.

If the voltage taking account of derating is greater than the voltage rating of MOSFET gate-source junction, lower the gate-source junction voltage by connecting voltage dividing resistors as shown in Figure 61.

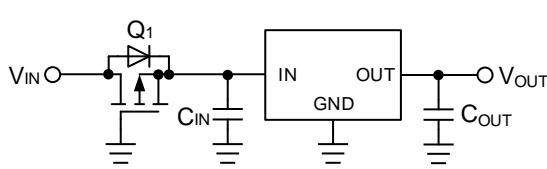


Figure 60. Protection against Reverse Polarity 2

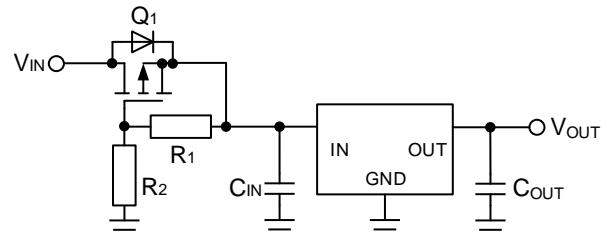


Figure 61. Protection against Reverse Polarity 3

3. Protection against Output Reverse Voltage when Output Connect to an Inductor

If the output load is inductive, electrical energy accumulated in the inductive load is released to the ground upon the output voltage turning off. In-between the IC output and ground pins is a diode for preventing electrostatic breakdown, in which a large current flows that could destroy the IC. To prevent this from happening, connect a Schottky barrier diode in parallel with the diode (see Figure 62).

Further, if a long wire is in use for the connection between the output pin of the IC and the load, observe the waveform on an oscilloscope, since it is possible that the load becomes inductive. An additional diode is needed for a motor load that is affected by its counter electromotive force, as it produces an electrical current in a similar way.

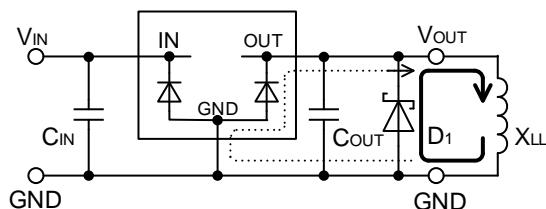


Figure 62. Current Path in Inductive Load (Output: Off)

Operational Notes**1) Absolute maximum ratings**

This product is produced with strict quality control, however it may be destroyed if operated beyond its absolute maximum ratings. In addition, it is impossible to predict all destructive situations such as short-circuit modes, open circuit modes, etc. Therefore, it is important to consider circuit protection measures, like adding a fuse, in case the IC is operated in a special mode exceeding the absolute maximum ratings.

2) GND Potential

GND potential must be the lowest potential of all pins of the IC at all operating conditions. Ensure that no pins are at a voltage below the ground pin at any time, even during transient condition.

3) Setting of Heat

Carry out the heat design that have adequate margin considering P_d of actual working states.

4) Pin Short and Mistake Fitting

When mounting the IC on the PCB, pay attention to the orientation of the IC. If there is mistake in the placement, the IC may be burned up.

5) Mutual Impedance

Use short and wide wiring tracks for the power supply and ground to keep the mutual impedance as small as possible.

Use a capacitor to keep ripple to a minimum.

6) STBY Pin Voltage

To enable standby mode for all channels, set the STBY pin to 0.5 V or less, and for normal operation, to 1.1 V or more. Setting STBY to a voltage between 0.5 and 1.1 V may cause malfunction and should be avoided. Keep transition time between high and low (or vice versa) to a minimum.

Additionally, if STBY is shorted to VIN, the IC will switch to standby mode and disable the output discharge circuit, causing a temporary voltage to remain on the output pin. If the IC is switched on again while this voltage is present, overshoot may occur on the output. Therefore, in applications where these pins are shorted, the output should always be completely discharged before turning the IC on.

7) Over Current Protection Circuit

Over current and short circuit protection is built-in at the output, and IC destruction is prevented at the time of load short circuit. These protection circuits are effective in the destructive prevention by sudden accidents, please avoid applications to where the over current protection circuit operates continuously.

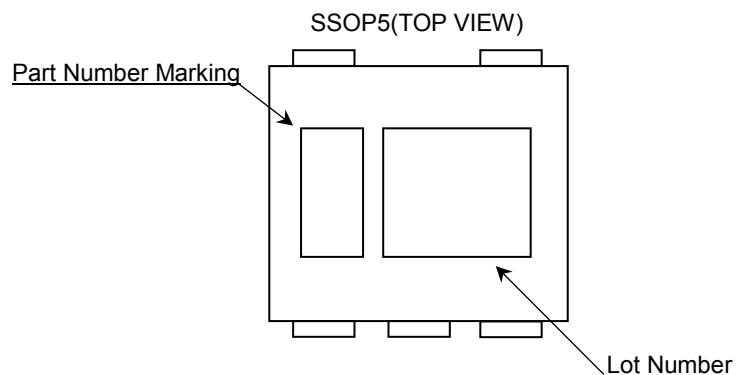
8) Thermal Shutdown

This IC has Thermal Shutdown Circuit (TSD Circuit). When the temperature of IC Chip is higher than 180°C(typ), the output is turned off by TSD Circuit. TSD Circuit is only designed for protecting IC from thermal over load. Therefore it is not recommended that you design application where TSD will work in normal condition.

9) Output capacitor

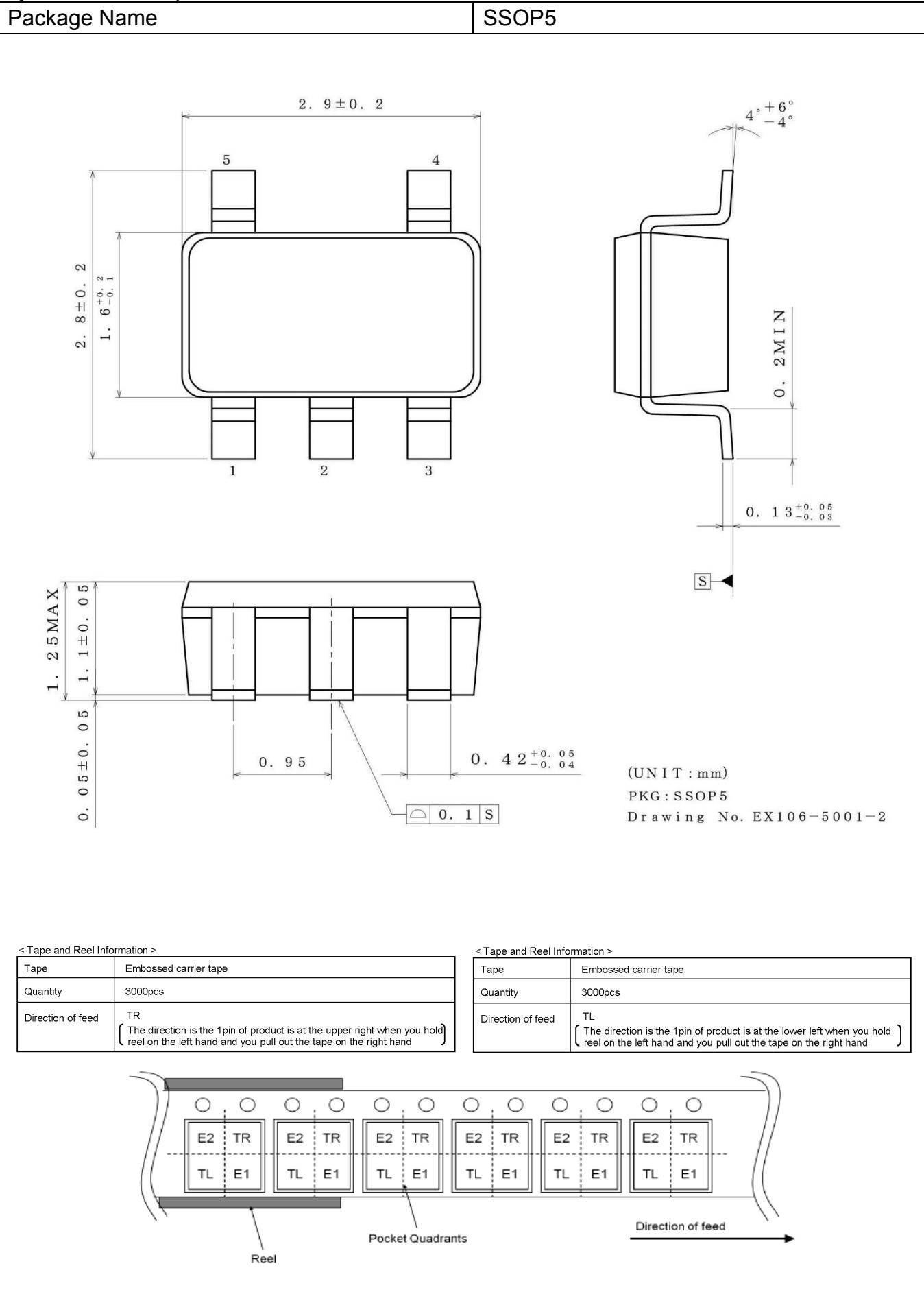
To prevent oscillation at output, it is recommended that the IC be operated at the stable region shown in Figure 50. It operates at the capacitance of more than 0.47 μ F. As capacitance is larger, stability becomes more stable and characteristic of output load fluctuation is also improved.

Marking Diagram



Part Number	Output Voltage [V]	Part Number Marking
BU10JA2VG-C	1.0	5T
BU12JA2VG-C	1.2	5U
BU1CJA2VG-C	1.25	5V
BU15JA2VG-C	1.5	5W
BU18JA2VG-C	1.8	XM
BU25JA2VG-C	2.5	5X
BU28JA2VG-C	2.8	Z6
BU2JJA2VG-C	2.85	5Y
BU30JA2VG-C	3.0	5Z
BU33JA2VG-C	3.3	XN

Physical Dimension Tape and Reel Information



Revision History

Date	Revision	Changes
10.Dec.2014	001	New Release
20.Mar.2015	002	Thermal Characteristics is changed.
24.Mar.2015	003	Correction of errors.
30.Aug.2017	004	Lineup is added P.2 TL version is added to the "Ordering Information". Block Diagram is updated P.3 The item of the STBY pin is added to "Absolute Maximum Ratings". P.7 "Figure 14. Dropout Voltage vs Output Current" is added P.21 to P.23 The item of "Linear Regulators Surge Voltage Protection" is added The item of "Linear Regulators Reverse Voltage Protection" is added P.25 An expression method of "Marking Diagram" is changed P.26 TL version is added to the "Physical Dimension Tape and Reel Information". Others, correction of errors.

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(Note1) Medical Equipment Classification of the Specific Applications

JAPAN	USA	EU	CHINA
CLASS III	CLASS III	CLASS II b	
CLASS IV		CLASS III	CLASS III

2. ROHM designs and manufactures its Products subject to strict quality control system. However, semiconductor products can fail or malfunction at a certain rate. Please be sure to implement, at your own responsibilities, adequate safety measures including but not limited to fail-safe design against the physical injury, damage to any property, which a failure or malfunction of our Products may cause. The following are examples of safety measures:
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 - [b] Installation of redundant circuits to reduce the impact of single or multiple circuit failure
3. Our Products are not designed under any special or extraordinary environments or conditions, as exemplified below. Accordingly, ROHM shall not be in any way responsible or liable for any damages, expenses or losses arising from the use of any ROHM's Products under any special or extraordinary environments or conditions. If you intend to use our Products under any special or extraordinary environments or conditions (as exemplified below), your independent verification and confirmation of product performance, reliability, etc, prior to use, must be necessary:
 - [a] Use of our Products in any types of liquid, including water, oils, chemicals, and organic solvents
 - [b] Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
 - [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
 - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
4. The Products are not subject to radiation-proof design.
5. Please verify and confirm characteristics of the final or mounted products in using the Products.
6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
7. De-rate Power Dissipation depending on ambient temperature. When used in sealed area, confirm that it is the use in the range that does not exceed the maximum junction temperature.
8. Confirm that operation temperature is within the specified range described in the product specification.
9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

Precaution for Mounting / Circuit board design

1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
2. In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

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1. If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
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Precaution for Electrostatic

This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of Ionizer, friction prevention and temperature / humidity control).

Precaution for Storage / Transportation

1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
 - [a] the Products are exposed to sea winds or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
 - [b] the temperature or humidity exceeds those recommended by ROHM
 - [c] the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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