

CMOS LDO Regulator Series for Automotive

Ultra-Small Package FULL CMOS LDO Regulator

BUxxJA2MNVX-C series

General Description

BUxxJA2MNVX-C series are high-performance FULL CMOS regulators with 200mA output, which are mounted on versatile package SSON004R1010 (1.00mm x 1.00 mm x 0.60mm). These devices have excellent noise characteristics and load responsiveness characteristics despite its low circuit current consumption of 35µA. They are most appropriate for various applications such as power supplies for radar and camera of the automotive.

Features

- AEC-Q100 Qualified^(Note 1)
- High Accuracy Output
- Low Current Consumption
- Compatible With Small Ceramic Capacitor($C_{IN}=C_O=0.47\mu F$)
- With Built-in Output Discharge Circuit
- High Ripple Rejection
- ON/OFF Control of Output Voltage
- Built-in Over Current Protection Circuit
- Built-in Thermal Shutdown Circuit

(Note 1) Grade1

Key Specifications

- Input Voltage Range: 1.7V to 6.0V
- Output Voltage: 1.0V to 3.4V
- Output Voltage Accuracy: $\pm 2.0\%$ ($T_a=-40^\circ C$ to $125^\circ C$)
- Output Current: 200mA(Max)
- Standby Current: 35µA (Typ)
- Operating Temperature Range: $-40^\circ C$ to $+125^\circ C$

Package	W(Typ) x D(Typ) x H(Max)
SSON004R1010	1.00mm x 1.00mm x 0.60mm



SSON004R1010

Applications

- Radar and camera for automotive, etc.

Typical Application Circuit

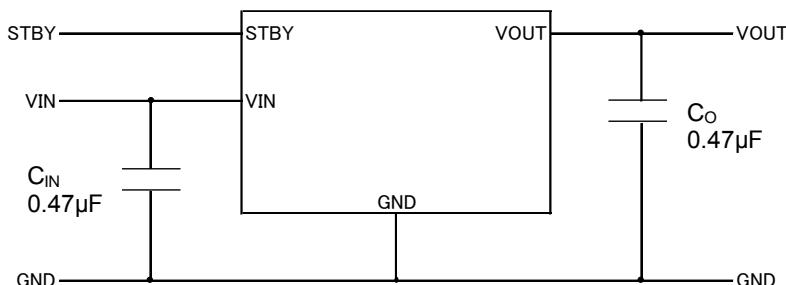


Figure 1. Application Circuit

Ordering Information

B	U	X	X	J	A	2	M	N	V	X	-	C TL
Part Number	Output Voltage 10 : 1.0V 11 : 1.1V 12 : 1.2V 1C : 1.25V 15 : 1.5V 18 : 1.8V 25 : 2.5V 28 : 2.8V 2J : 2.85V 29 : 2.9V 30 : 3.0V 33 : 3.3V 34 : 3.4V	Output Current 200mA	Category M:Automotive	Package NVX:SSON004R1010	Product Rank C:for Automotive Packaging and forming specification TL: Embossed tape and reel							

Block Diagram

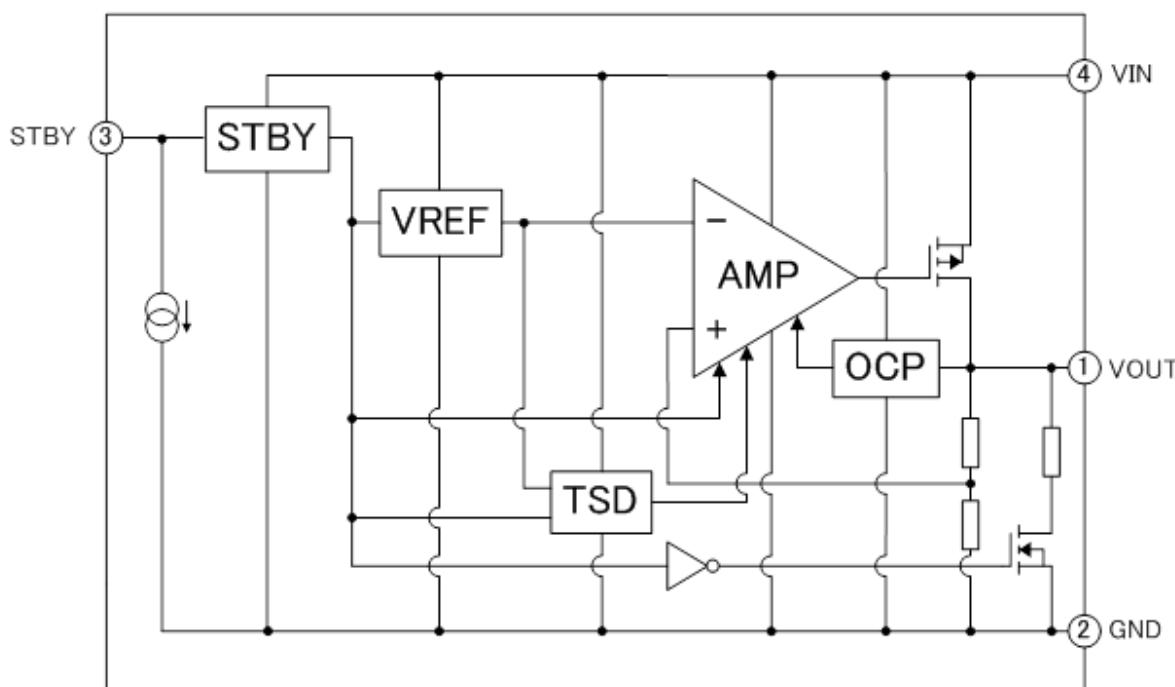
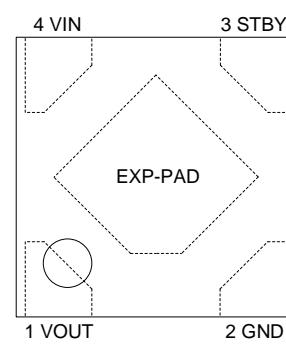


Figure 2. Block Diagram

Pin Descriptions

Pin Configurations

Pin No.	Pin name	Pin Function
1	VOUT	Output Voltage
2	GND	Ground
3	STBY	ON/OFF control of output voltage (High: ON, Low: OFF)
4	VIN	Power Supply Voltage
Back side	EXP-PAD	Connect to GND

SS004R1010 (TOP VIEW)^(Note1)^(Note1)The dashed line is the electrode position of the back side.

Absolute Maximum Ratings

Parameter	Symbol	Rating	Unit
Power Supply Voltage	V_{IN}	-0.3 to +6.5	V
STBY Voltage	V_{STBY}	-0.3 to +6.5	V
Operating Temperature Range	T_{opr}	-40 to +125	°C
Storage Temperature Range	T_{stg}	-55 to +150	°C
Maximum junction temperature	T_{jmax}	+150	°C

Recommended Operating Range

Parameter	Symbol	Min	Max	Unit
Power Supply Voltage	V_{IN}	1.7	6.0	V
STBY Voltage	V_{STBY}	0.0	6.0	V
Maximum Output Current	I_{OUT}	-	200	mA

Recommended Operating Conditions

Parameter	Symbol	Min	Typ	Max	Unit	Conditions
Input Capacitor	C_{IN}	0.22 (Note 1)	0.47	-	μF	Ceramic capacitor recommended
Output Capacitor	C_O	0.22 (Note 1)	0.47	-	μF	Ceramic capacitor recommended

(Note 1) Caution that the capacitance to be kept higher than this specified values under all conditions considering temperature, DC bias, etc.

Thermal Resistance ^(Note 1)

Parameter	Symbol	Thermal Resistance (Typ)		Unit
		1s ^(Note 3)	2s2p ^(Note 4)	
SSON004R1010				
Junction to Ambient	θ_{JA}	450.2	97.1	°C/W
Junction to Top Characterization Parameter ^(Note 2)	Ψ_{JT}	99	22	°C/W

(Note 1) Based on JESD51-2A(Still-Air).

(Note 2) The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

(Note 3) Using a PCB board based on JESD51-3.

(Note 4) Using a PCB board based on JESD51-5, 7.

Layer Number of Measurement Board	Material	Board Size
Single	FR-4	114.3 mm x 76.2 mm x 1.57 mm

Top	
Copper Pattern	Thickness
Footprints and Traces	70 μ m

Layer Number of Measurement Board	Material	Board Size	Thermal Via ^(Note 5)	
			Pitch	Diameter
4 Layers	FR-4	114.3 mm x 76.2 mm x 1.6 mm	1.20 mm	Φ 0.30 mm

Top		2 Internal Layers		Bottom	
Copper Pattern	Thickness	Copper Pattern	Thickness	Copper Pattern	Thickness
Footprints and Traces	70 μ m	74.2 mm x 74.2 mm	35 μ m	74.2 mm x 74.2 mm	70 μ m

(Note 5) This thermal via connects with the copper pattern of all layers.

Electrical Characteristics(Ta= -40°C to 125°C, VIN=VOUT+1.0V ^(Note 1), STBY=VIN, C_{IN}=0.47μF, C_O=0.47μF, unless otherwise specified.)

Parameter	Symbol	Limit			Unit	Conditions
		Min	Typ	Max		
[Regulator Block]						
Output Voltage 1	VOUT1	VOUT ×0.98	-	VOUT ×1.02	V	IOUT=0.01mA, VOUT≥1.8V
		VOUT -36mV	-	VOUT +36mV		IOUT=0.01mA, VOUT<1.8V
Output Voltage 2	VOUT2	VOUT ×0.97	-	VOUT ×1.03	V	IOUT=0.01mA to 200mA VOUT≥1.8V
		VOUT -54mV	-	VOUT +54mV		IOUT=0.01mA to 200mA VOUT<1.8V
Circuit Current	IIN	-	35	90	μA	IOUT=0mA
Circuit Current (STBY)	ISTBY	-	-	2.0	μA	STBY=0V
Ripple Rejection Ratio	RR	45	70	-	dB	VRR=-20dBV, fRR=1kHz IOUT=10mA, Ta=25°C
Dropout Voltage	VSAT	-	800	1100	mV	1.0V ≤ VOUT < 1.2V(IOUT=200mA)
		-	600	900	mV	1.2V ≤ VOUT < 1.5V(IOUT=200mA)
		-	440	830	mV	1.5V ≤ VOUT < 1.8V(IOUT=200mA)
		-	380	710	mV	1.8V ≤ VOUT < 2.5(IOUT=200mA)
		-	280	620	mV	2.5V ≤ VOUT ≤ 2.6(IOUT=200mA)
		-	260	580	mV	2.7V ≤ VOUT ≤ 2.85(IOUT=200mA)
		-	240	530	mV	2.9V ≤ VOUT ≤ 3.1V(IOUT=200mA)
		-	220	490	mV	3.2V ≤ VOUT ≤ 3.4V(IOUT=200mA)
Line Regulation	VDL	-	2	20	mV	VIN=VOUT+1.0V to 5.5V ^(Note 2) IOUT=0.01mA
Load Regulation	VDLO	-	10	80	mV	IOUT=0.01mA to 100mA
[Over Current Protection (OCP) Block]						
Limit Current	ILMAX	220	400	700	mA	ILMAX@VOUT×0.95, Ta=25°C
Short Current	ISHORT	20	70	150	mA	VOUT=0V, Ta=25°C
[Standby Block]						
Discharge Resistor	RDSC	20	50	80	Ω	VIN=4.0V, STBY=0V VOUT=4.0V, Ta=25°C
STBY Pin Pull-down Current	ISTB	0.1	0.6	2.0	μA	STBY=1.5V
STBY Control Voltage	ON	VSTBH	1.2	-	V	
	OFF	VSTBL	0	-	V	

(Note 1) VIN=2.5V for VOUT≤1.5V

(Note 2) VIN=2.5V to 3.6V for VOUT≤1.5V

Reference data BU33JA2MNvx-C (Ta=25°C unless otherwise specified.)

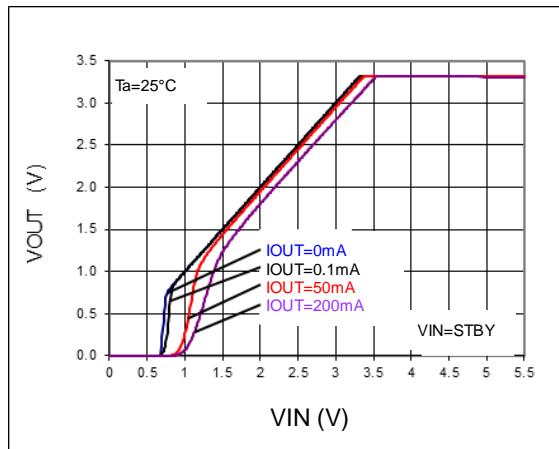


Figure 3. Output Voltage

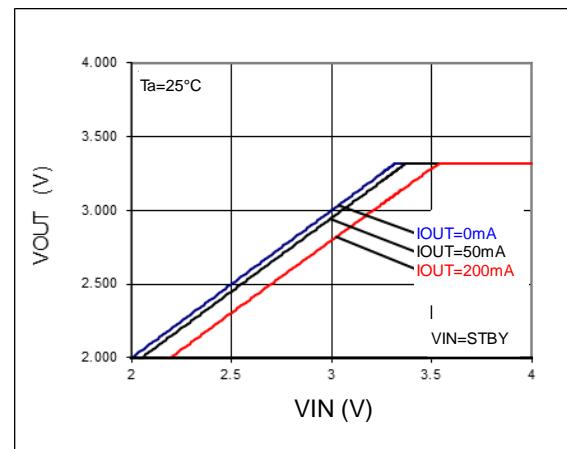


Figure 4. Output Voltage

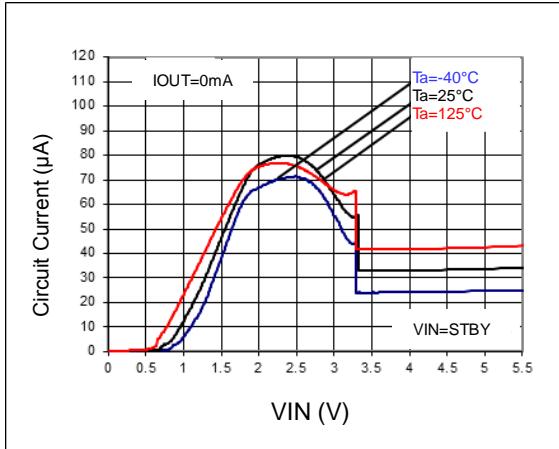


Figure 5. Circuit Current

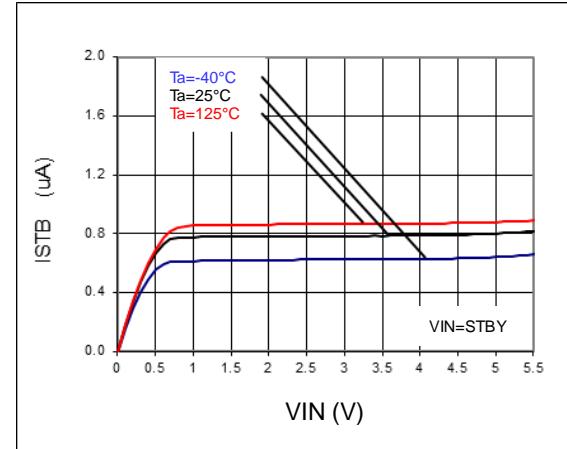


Figure 6. VSTBY - ISTB

Reference data BU33JA2MNvx-C (Ta=25°C unless otherwise specified.)

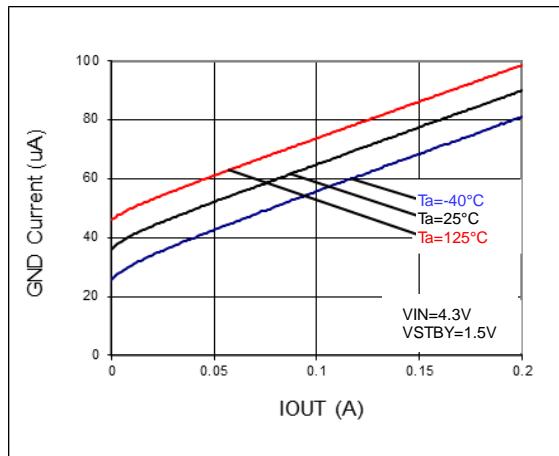


Figure 7. IOUT - IGND

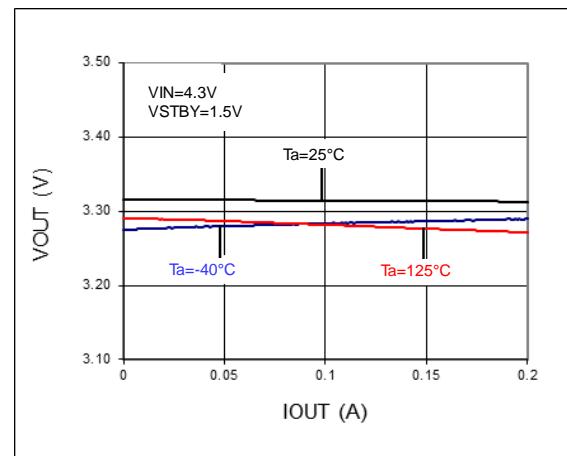


Figure 8. Load Regulation

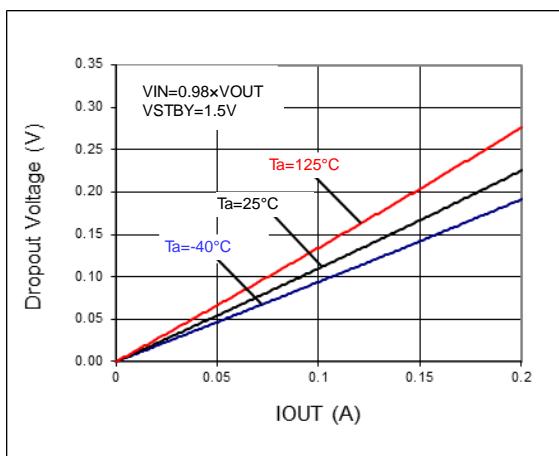


Figure 9. Dropout Voltage

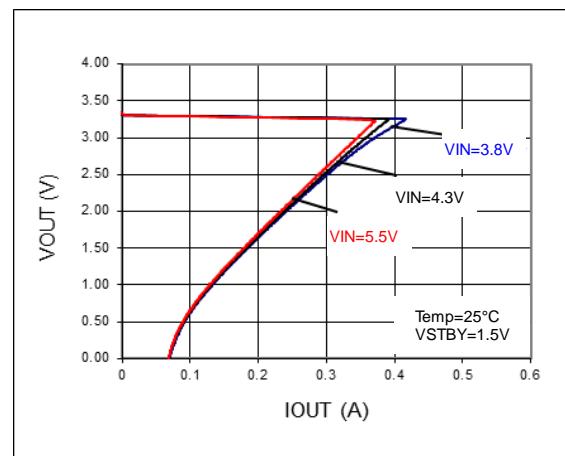


Figure 10. OCP Threshold

Reference data BU33JA2MNvx-C (Ta=25°C unless otherwise specified.)

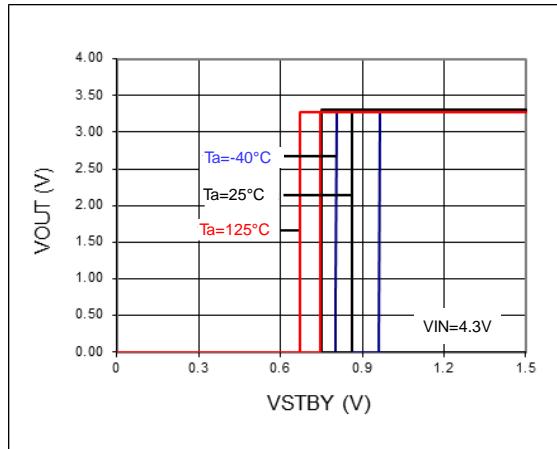


Figure 11. STBY Threshold

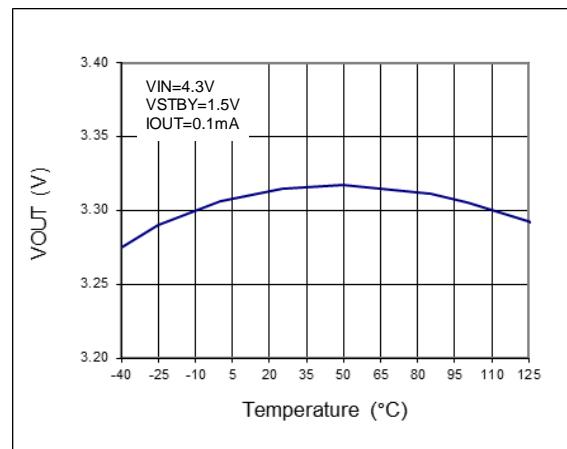


Figure 12. VOUT - Temp

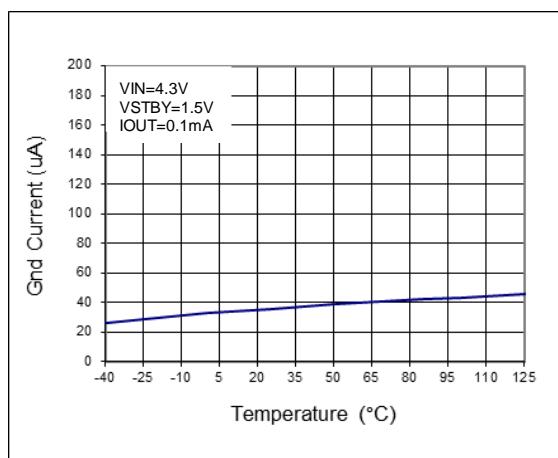


Figure 13. IGND - Temp

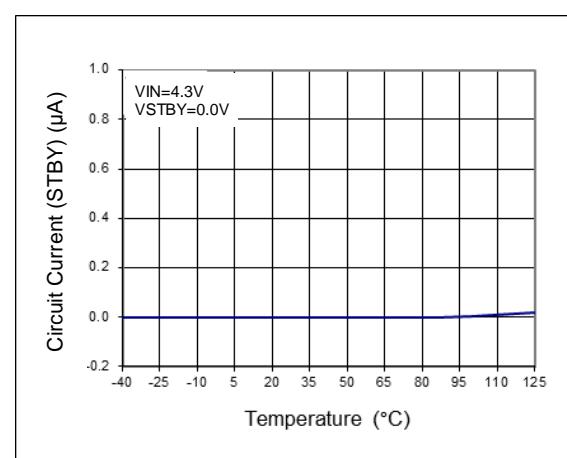
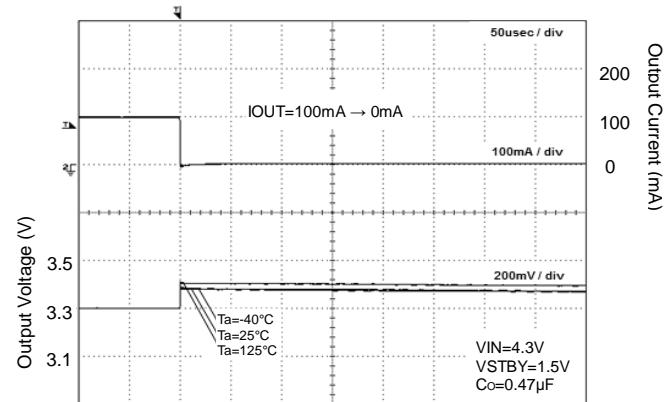
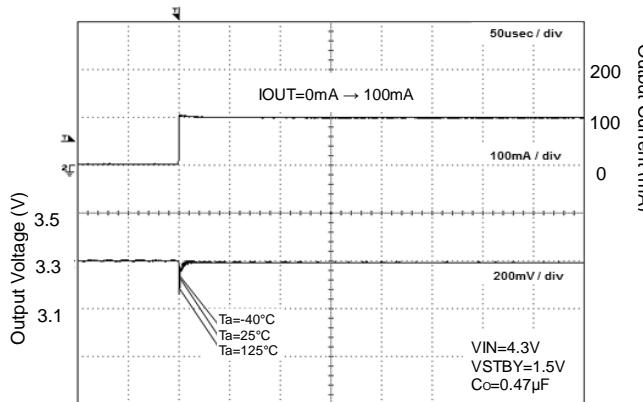
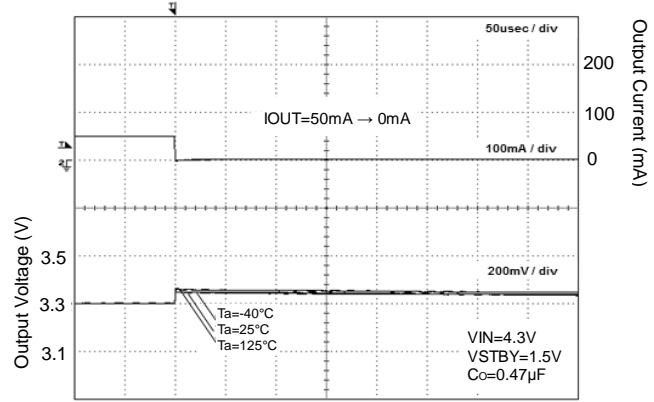
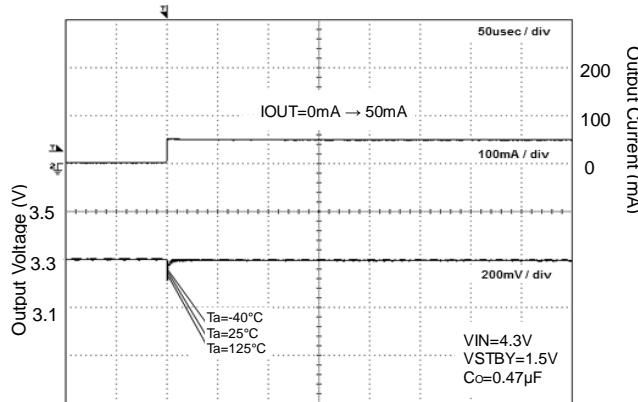
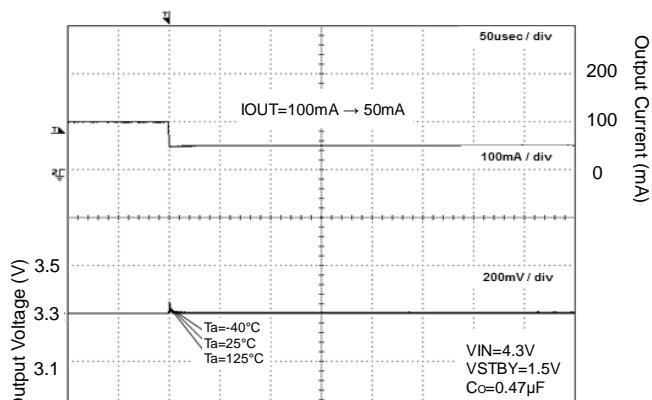
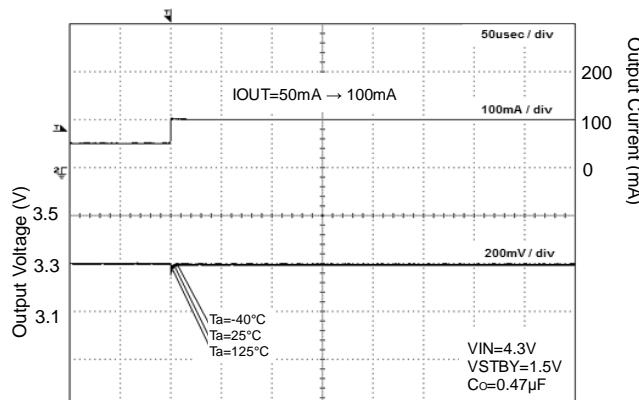
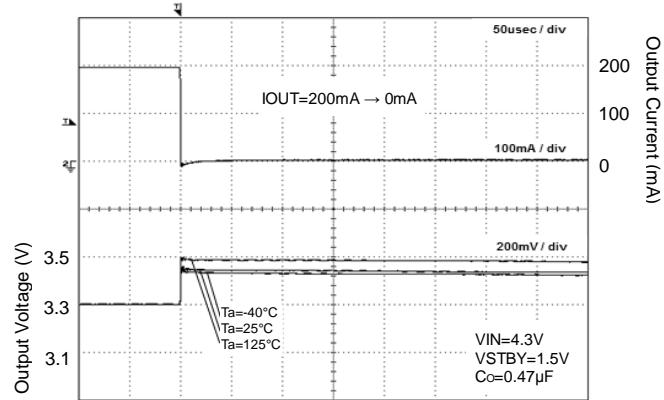
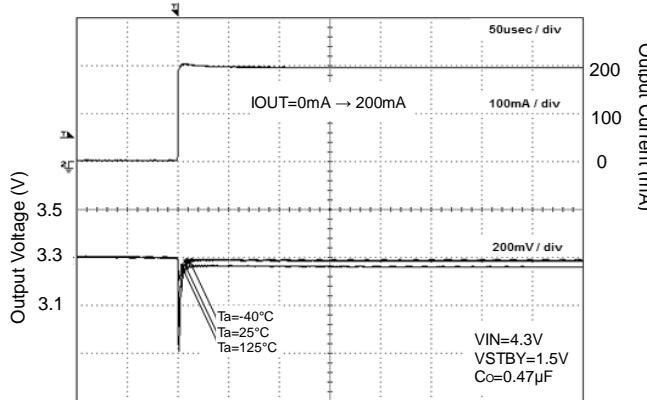


Figure 14. Circuit Current (STBY) - Temp

Reference data BU33JA2MNvx-C (Ta=25°C unless otherwise specified.)



Reference data BU33JA2MNvx-C (Ta=25°C unless otherwise specified.)



Reference data BU33JA2MNvx-C (Ta=25°C unless otherwise specified.)

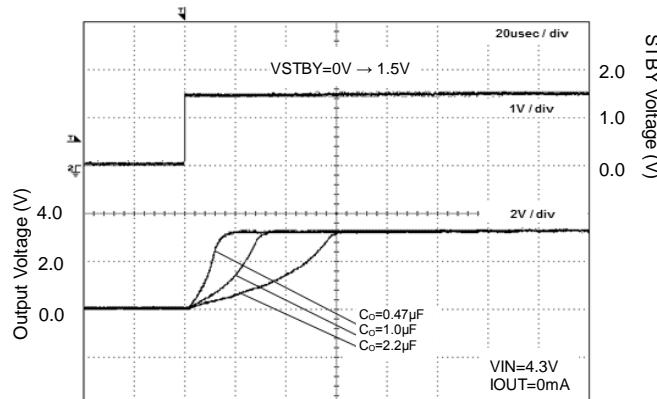


Figure 23. Start Up Time
IOUT=0mA

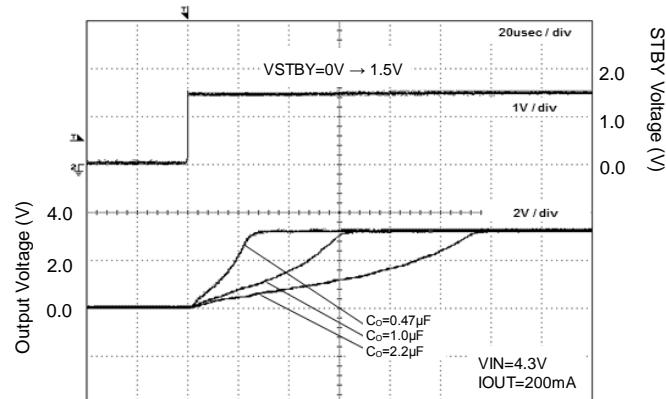


Figure 24. Start Up Time
IOUT=200mA

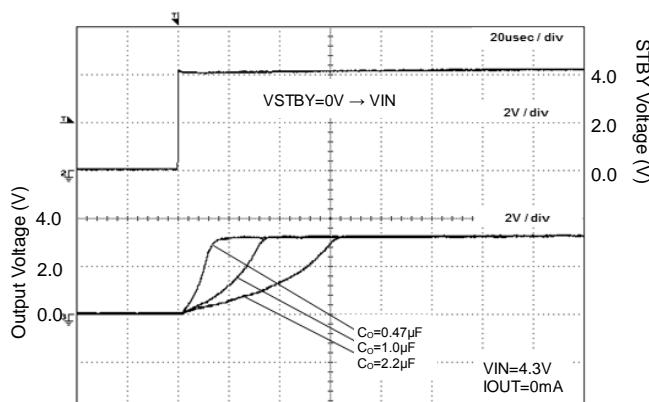


Figure 25. Start Up Time
(VIN=STBY) IOUT=0mA

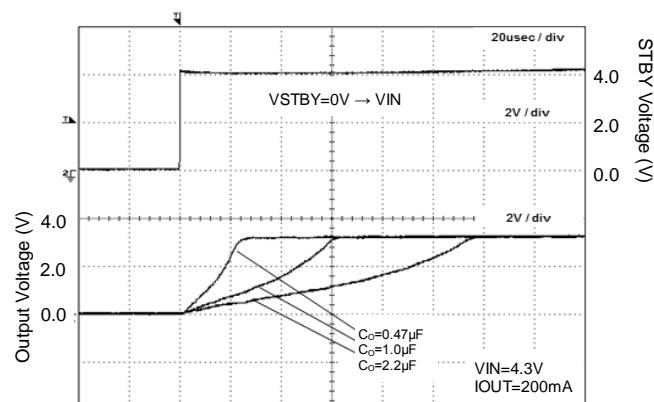


Figure 26. Start Up Time
(VIN=STBY) IOUT=200mA

Reference data BU33JA2MNvx-C (Ta=25°C unless otherwise specified.)

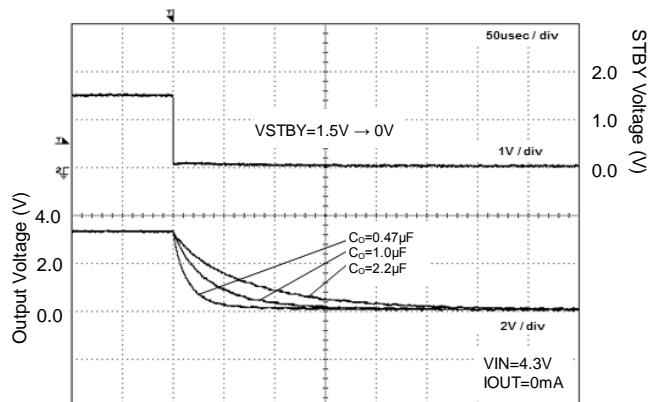


Figure 27. Discharge Time

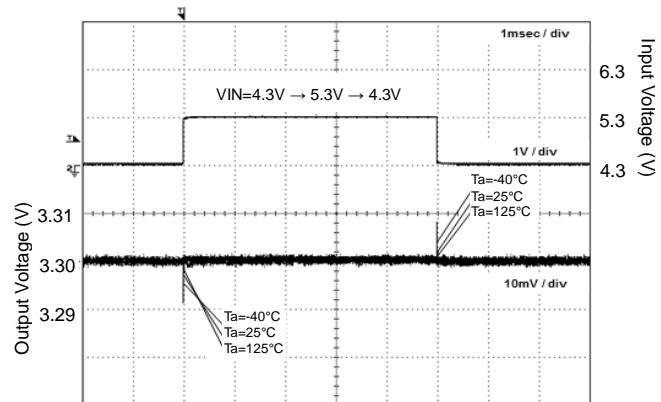


Figure 28. VIN Response

Power Dissipation

SSON004R1010

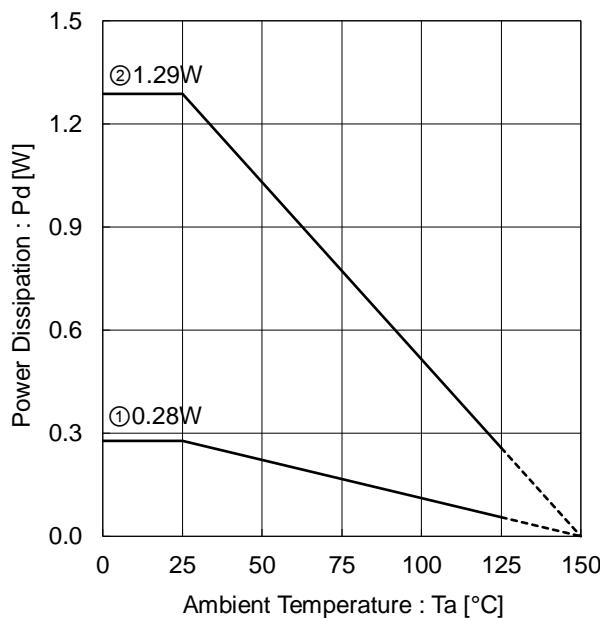


Figure 29. SSON004R1010 Package Data
(Reference Data)

IC mounted on ROHM standard board based on JEDEC.

① : 1-layer PCB
(Copper foil area on the reverse side of PCB: 0 mm × 0 mm)
Board material: FR4
Board size: 114.3 mm × 76.2 mm × 1.57 mm
Mount condition: PCB and exposed pad are soldered.
Top copper foil: ROHM recommended
footprint + wiring to measure, 2 oz. copper.

② : 4-layer PCB
(2 inner layers copper foil area of PCB, copper foil area on the
reverse side of PCB: 74.2 mm × 74.2 mm)
Board material: FR4
Board size: 114.3 mm × 76.2 mm × 1.6 mm
Mount condition: PCB and exposed pad are soldered.
Top copper foil: ROHM recommended
footprint + wiring to measure, 2 oz. copper.
2 inner layers copper foil area of PCB
: 74.2 mm × 74.2 mm, 1 oz. copper.
Copper foil area on the reverse side of PCB
: 74.2 mm × 74.2 mm, 2 oz. copper.

Condition① : $\theta_{JA} = 450.2 \text{ }^{\circ}\text{C/W}$, Ψ_{JT} (top center) = 99 $\text{ }^{\circ}\text{C/W}$

Condition② : $\theta_{JA} = 97.1 \text{ }^{\circ}\text{C/W}$, Ψ_{JT} (top center) = 22 $\text{ }^{\circ}\text{C/W}$

Thermal Design

Within this IC, the power consumption is decided by the dropout voltage condition, the load current and the circuit current. Refer to power dissipation curves illustrated in Figure 29 when using the IC in an environment of $T_a \geq 25^\circ\text{C}$. Even if the ambient temperature T_a is at 25°C , depending on the input voltage and the load current, chip junction temperature can be very high. Consider the design to be $T_j \leq T_{jmax} = 150^\circ\text{C}$ in all possible operating temperature range.

Should by any condition the maximum junction temperature $T_{jmax} = 150^\circ\text{C}$ rating be exceeded by the temperature increase of the chip, it may result in deterioration of the properties of the chip. The thermal impedance in this specification is based on recommended PCB and measurement condition by JEDEC standard. Verify the application and allow sufficient margins in the thermal design by the following method is used to calculate the junction temperature T_j .

T_j can be calculated by either of the two following methods.

1. The following method is used to calculate the junction temperature T_j .

$$T_j = T_a + P_c \times \theta_{JA}$$

Where:

- T_j : Junction Temperature
- T_a : Ambient Temperature
- P_c : Power Consumption
- θ_{JA} : Thermal Impedance
(Junction to Ambient)

2. The following method is also used to calculate the junction temperature T_j .

$$T_j = T_T + P_c \times \Psi_{JT}$$

Where:

- T_j : Junction Temperature
- T_T : Top Center of Case's (mold) Temperature
- P_c : Power consumption
- Ψ_{JT} : Thermal Impedance
(Junction to Top Center of Case)

The following method is used to calculate the power consumption P_c (W).

$$P_c = (V_{IN} - V_{OUT}) \times I_{OUT} + V_{IN} \times I_{GND}$$

Where:

- P_c : Power Consumption
- V_{IN} : Input Voltage
- V_{OUT} : Output Voltage
- I_{OUT} : Load Current
- I_{GND} : Circuit Current

- Calculation Example (SSON004R1010)

If $V_{IN} = 3.0$ V, $V_{OUT} = 1.8$ V, $I_{OUT} = 50$ mA, $I_{GND} = 35$ μ A, the power consumption P_c can be calculated as follows:

$$\begin{aligned} P_c &= (V_{IN} - V_{OUT}) \times I_{OUT} + V_{IN} \times I_{GND} \\ &= (3.0 \text{ V} - 1.8 \text{ V}) \times 50 \text{ mA} + 3.0 \text{ V} \times 35 \mu\text{A} \\ &= 0.06 \text{ W} \end{aligned}$$

At the ambient temperature $T_{max} = 125^\circ\text{C}$, the thermal Impedance (Junction to Ambient) $\theta_{JA} = 97.1$ $^\circ\text{C} / \text{W}$ (4-layer PCB),

$$\begin{aligned} T_j &= T_{max} + P_c \times \theta_{JA} \\ &= 125^\circ\text{C} + 0.06 \text{ W} \times 97.1^\circ\text{C} / \text{W} \\ &= 130.8^\circ\text{C} \end{aligned}$$

When operating the IC, the top center of case's (mold) temperature $T_T = 100$ $^\circ\text{C}$, $\Psi_{JT} = 22$ $^\circ\text{C} / \text{W}$ (4-layer PCB),

$$\begin{aligned} T_j &= T_T + P_c \times \Psi_{JT} \\ &= 100^\circ\text{C} + 0.06 \text{ W} \times 22^\circ\text{C} / \text{W} \\ &= 101.3^\circ\text{C} \end{aligned}$$

For optimum thermal performance, it is recommended to expand the copper foil area of the board, increasing the layer and thermal via between thermal land pad.

Linear Regulators Surge Voltage Protection

The following provides instructions on surge voltage overs absolute maximum ratings polarity protection for ICs.

1. Applying positive surge to the input

If the possibility exists that surges higher than absolute maximum ratings 6.5 V will be applied to the input, a Zener Diode should be placed to protect the device in between the V_{IN} and the GND as shown in the figure 30.

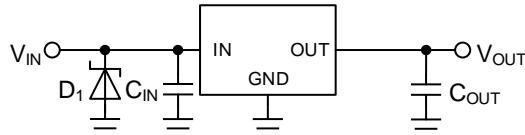


Figure 30. Surges Higher than 6.5 V will be Applied to the Input

2. Applying negative surge to the input

If the possibility exists that surges lower than absolute maximum ratings -0.3 V will be applied to the input, a Schottky Diode should be placed to protect the device in between the V_{IN} and the GND as shown in the figure 31.

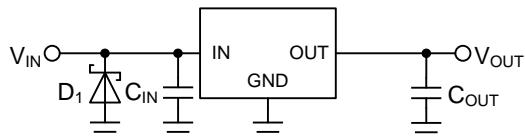


Figure 31. Surges Lower than -0.3 V will be Applied to the Input

Linear Regulators Reverse Voltage Protection

A linear regulator integrated circuit (IC) requires that the input voltage is always higher than the regulated voltage. Output voltage, however, may become higher than the input voltage under specific situations or circuit configurations, and that reverse voltage and current may cause damage to the IC. A reverse polarity connection or certain inductor components can also cause a polarity reversal between the input and output pins. The following provides instructions on reversed voltage polarity protection for ICs.

1. about Input /Output Voltage Reversal

In an MOS linear regulator, a parasitic element exists as a body diode in the drain-source junction portion of its power MOSFET. Reverse input/output voltage triggers the current flow from the output to the input through the body diode. The inverted current may damage or destroy the semiconductor elements of the regulator since the effect of the parasitic body diode is usually disregarded for the regulator behavior (Figure 32).

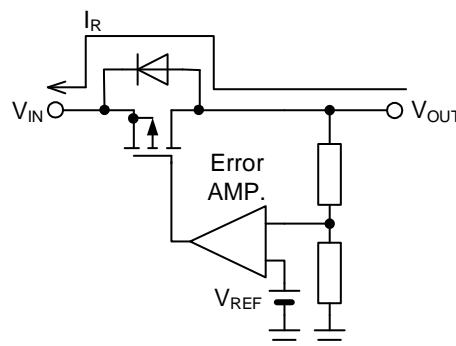


Figure 32. Reverse Current Path in an MOS Linear Regulator

An effective solution to this is an external bypass diode connected in-between the input and output to prevent the reverse current flow inside the IC (see Figure 33). Note that the bypass diode must be turned on before the internal circuit of the IC. Bypass diodes in the internal circuits of MOS linear regulators must have low forward voltage V_F . Some ICs are configured with current-limit thresholds to shut down high reverse current even when the output is off, allowing large leakage current from the diode to flow from the input to the output; therefore, it is necessary to choose one that has a small reverse current. Specifically, select a diode with a rated peak inverse voltage greater than the input to output voltage differential and rated forward current greater than the reverse current during use.

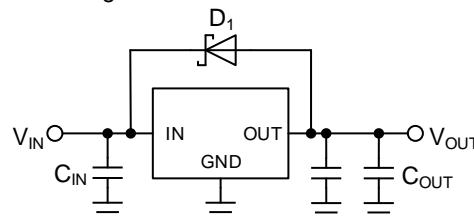


Figure 33. Bypass Diode for Reverse Current Diversion

The lower forward voltage (V_F) of Schottky barrier diodes cater to requirements of MOS linear regulators, however the main drawback is found in the level of their reverse current (I_R), which is relatively high. So, one with a low reverse current is recommended when choosing a Schottky diode. The V_R - I_R characteristics versus temperatures show increases at higher temperatures.

If V_{IN} is open in a circuit as shown in the following Figure 34 with its input/output voltage being reversed, the only current that flows in the reverse current path is the bias current of the IC. Because the amperage is too low to damage or destroy the parasitic element, a reverse current bypass diode is not required for this type of circuit.

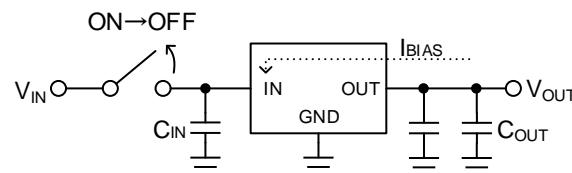


Figure 34. Open V_{IN}

2. Protection against Input Reverse Voltage

Accidental reverse polarity at the input connection flows a large current to the diode for electrostatic breakdown protection between the input pin of the IC and the GND pin, which may destroy the IC (see Figure 35).

A Schottky barrier diode or rectifier diode connected in series with the power supply as shown in Figure 36 is the simplest solution to prevent this from happening. The solution, however, is unsuitable for a circuit powered by batteries because there is a power loss calculated as $V_F \times I_{OUT}$, as the forward voltage V_F of the diode drops in a correct connection. The lower V_F of a Schottky barrier diode than that of a rectifier diode gives a slightly smaller power loss. Because diodes generate heat, care must be taken to select a diode that has enough allowance in power dissipation. A reverse connection allows a negligible reverse current to flow in the diode.

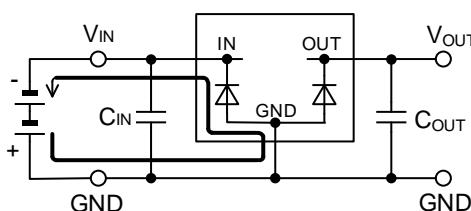


Figure 35. Current Path in Reverse Input Connection

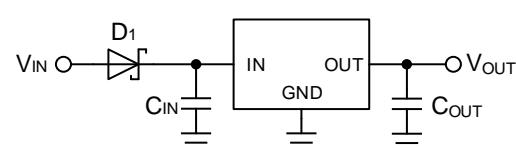


Figure 36. Protection against Reverse Polarity 1

Figure 37 shows a circuit in which a P-channel MOSFET is connected in series with the power. The diode located in the drain-source junction portion of the MOSFET is a body diode (parasitic element). The voltage drop in a correct connection is calculated by multiplying the resistance of the MOSFET being turned on by the output current I_{OUT} , therefore it is smaller than the voltage drop by the diode (see Figure 36) and results in less of a power loss. No current flows in a reverse connection where the MOSFET remains off.

If the voltage taking account of derating is greater than the voltage rating of MOSFET gate-source junction, lower the gate-source junction voltage by connecting voltage dividing resistors as shown in Figure 38.

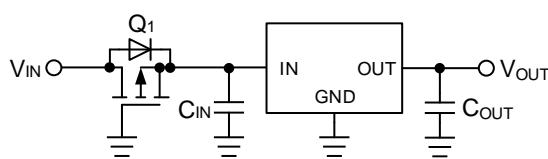


Figure 37. Protection against Reverse Polarity 2

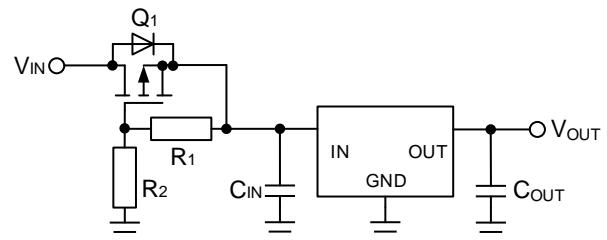


Figure 38. Protection against Reverse Polarity 3

3. Protection against Output Reverse Voltage when Output Connect to an Inductor

If the output load is inductive, electrical energy accumulated in the inductive load is released to the ground upon the output voltage turning off. In-between the IC output and ground pins is a diode for preventing electrostatic breakdown, in which a large current flows that could destroy the IC. To prevent this from happening, connect a Schottky barrier diode in parallel with the diode (see Figure 39).

Further, if a long wire is in use for the connection between the output pin of the IC and the load, observe the waveform on an oscilloscope, since it is possible that the load becomes inductive. An additional diode is needed for a motor load that is affected by its counter electromotive force, as it produces an electrical current in a similar way.

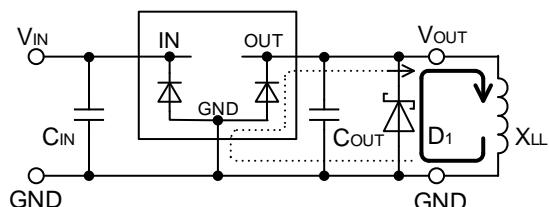


Figure 39. Current Path in Inductive Load (Output: Off)

Operation Notes

1. Absolute maximum ratings

Use of the IC exceeding the absolute maximum ratings (such as the input voltage or operating temperature range) may result in damage to the IC. Damage mode of the IC in such case cannot be assumed (e.g. short mode or open mode). If operational values are expected to exceed the maximum ratings for the device, consider adding protective circuitry (such as fuses) to eliminate the risk of damaging the IC.

2. GND potential

The potential of the GND pin must be the minimum potential in the system in all operating conditions. Never connect a potential lower than GND to any pin, even if only transiently.

3. Thermal design

Use a thermal design which ensure sufficient margin to the power dissipation rating (P_d) under actual operating conditions.

4. Inter-pin shorts and mounting errors

Caution on the orientation and positioning of the IC for mounting on printed circuit boards. Improper mounting or shorts between pins may result in damage to the IC.

5. Common impedance

Wiring traces should be as short and wide as possible to minimize common impedance. Bypass capacitors should be use to keep ripple to a minimum.

6. Voltage of STBY pin

To enable standby mode for all channels, set the STBY pin to 0.3 V or less, and for normal operation, to 1.2 V or more. Setting STBY to a voltage between 0.3 and 1.2 V may cause malfunction and should be avoided. Keep transition time between high and low (or vice versa) to a minimum.

Additionally, if STBY is shorted to VIN, the IC will switch to standby mode and disable the output discharge circuit, causing a temporary voltage to remain on the output pin. If the IC is switched on again while this voltage is present, overshoot may occur on the output. Therefore, in applications where these pins are shorted, the output should always be completely discharged before turning the IC on.

7. Over-current protection circuit (OCP)

This IC features an integrated over-current and short-protection circuitry on the output to prevent destruction of the IC when the output is shorted. The OCP circuitry is designed only to protect the IC from irregular conditions (such as motor output shorts) and is not designed to be used as an active security device for the application. Therefore, applications should not be designed under the assumption that this circuitry will engage.

8. Thermal shutdown circuit (TSD)

This IC also features a thermal shutdown circuit that is designed to turn the output off when the junction temperature of the IC exceeds approximately 150°C. This feature is intended to protect the IC only in the event of thermal overload and is not designed to guarantee operation or act as an active security device for the application. Therefore, applications should not be designed under the assumption that this circuitry will engage.

9. Input/output capacitor

Capacitors must be connected between the input/output pins and GND for stable operation, and should be physically mounted as close to the IC pins as possible. The input capacitor helps to counteract increases in power supply impedance, and increases stability in applications with long or winding power supply traces. The output capacitance value is directly related to the overall stability and transient response of the regulator, and should be set to the largest possible value for the application to increase these characteristics. During design, keep in mind that in general, ceramic capacitors have a wide range of tolerances, temperature coefficients and DC bias characteristics, and that their capacitance values tend to decrease over time. Confirm these details before choosing appropriate capacitors for your application. (Refer to the technical note of the intended ceramic capacitors.)

10. About the equivalent series resistance (ESR) of a ceramic capacitor

Capacitors generally have ESR (equivalent series resistance) and it operates stably in the ESR-I_{OUT} area shown on the below. Since ceramic capacitors, tantalum capacitors, electrolytic capacitors, etc. generally have different ESR, please check the ESR of the capacitor to be used and use it within the stability area range shown in the right graph for evaluation of the actual application.

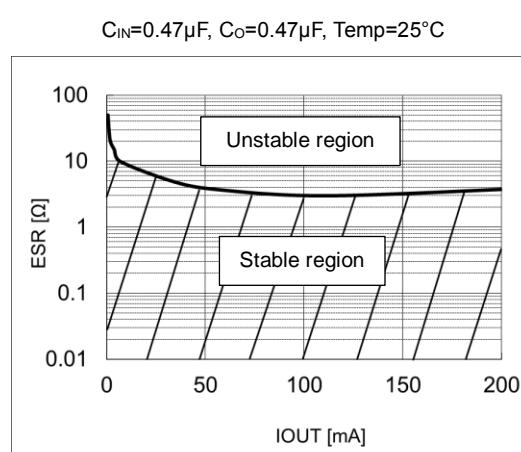


Figure 40. Stable region

Input/Output Capacitor

It is recommended that an input capacitor is placed near pins between the VIN pin and GND as well as an output capacitor between the VOUT pin and GND. The input is valid when the power supply impedance is high or when the PCB trace has significant length. For the output capacitor, the greater the capacitance, the more stable the output will be depending on the load and line voltage variations. However, please check the actual functionality of this capacitor by mounting it on a board for the actual application. Ceramic capacitors usually have different, thermal and equivalent series resistance characteristics, and may degrade gradually over continued use. For additional details, please check with the manufacturer, and select the best ceramic capacitor for your application.

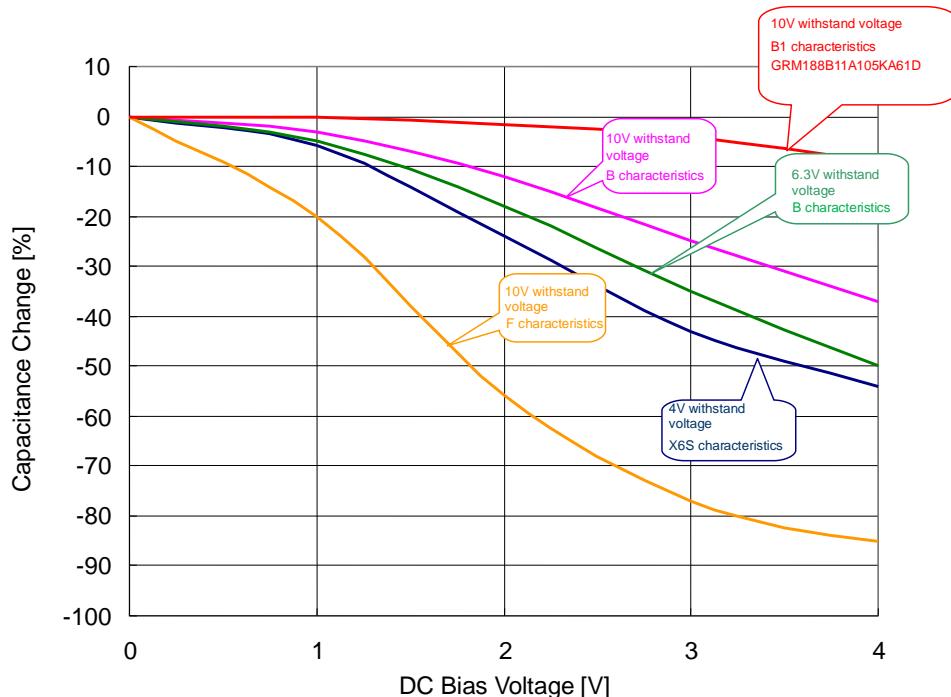


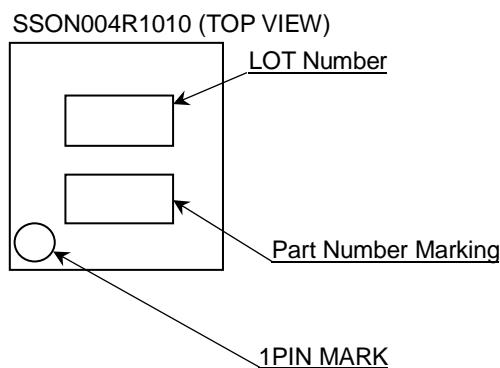
Figure 41. Capacity-bias characteristics
(Characteristics Example)

I/O Equivalence Circuits

1pin (VOUT)				3pin (STBY)	4pin (VIN)																																																						
	<table border="1"> <thead> <tr> <th>xx</th> <th>Output Voltage [V] (Typ)</th> <th>R1 [kΩ] (Typ)</th> <th>R2 [kΩ] (Typ)</th> </tr> </thead> <tbody> <tr><td>10</td><td>1.0</td><td>173</td><td>185</td></tr> <tr><td>11</td><td>1.1</td><td>206</td><td>185</td></tr> <tr><td>12</td><td>1.2</td><td>241</td><td>185</td></tr> <tr><td>1C</td><td>1.25</td><td>260</td><td>185</td></tr> <tr><td>15</td><td>1.5</td><td>352</td><td>185</td></tr> <tr><td>18</td><td>1.8</td><td>463</td><td>185</td></tr> <tr><td>25</td><td>2.5</td><td>716</td><td>185</td></tr> <tr><td>28</td><td>2.8</td><td>821</td><td>185</td></tr> <tr><td>2J</td><td>2.85</td><td>829</td><td>185</td></tr> <tr><td>29</td><td>2.9</td><td>847</td><td>185</td></tr> <tr><td>30</td><td>3.0</td><td>889</td><td>185</td></tr> <tr><td>33</td><td>3.3</td><td>1001</td><td>185</td></tr> <tr><td>34</td><td>3.4</td><td>1025</td><td>185</td></tr> </tbody> </table>	xx	Output Voltage [V] (Typ)	R1 [kΩ] (Typ)	R2 [kΩ] (Typ)	10	1.0	173	185	11	1.1	206	185	12	1.2	241	185	1C	1.25	260	185	15	1.5	352	185	18	1.8	463	185	25	2.5	716	185	28	2.8	821	185	2J	2.85	829	185	29	2.9	847	185	30	3.0	889	185	33	3.3	1001	185	34	3.4	1025	185		
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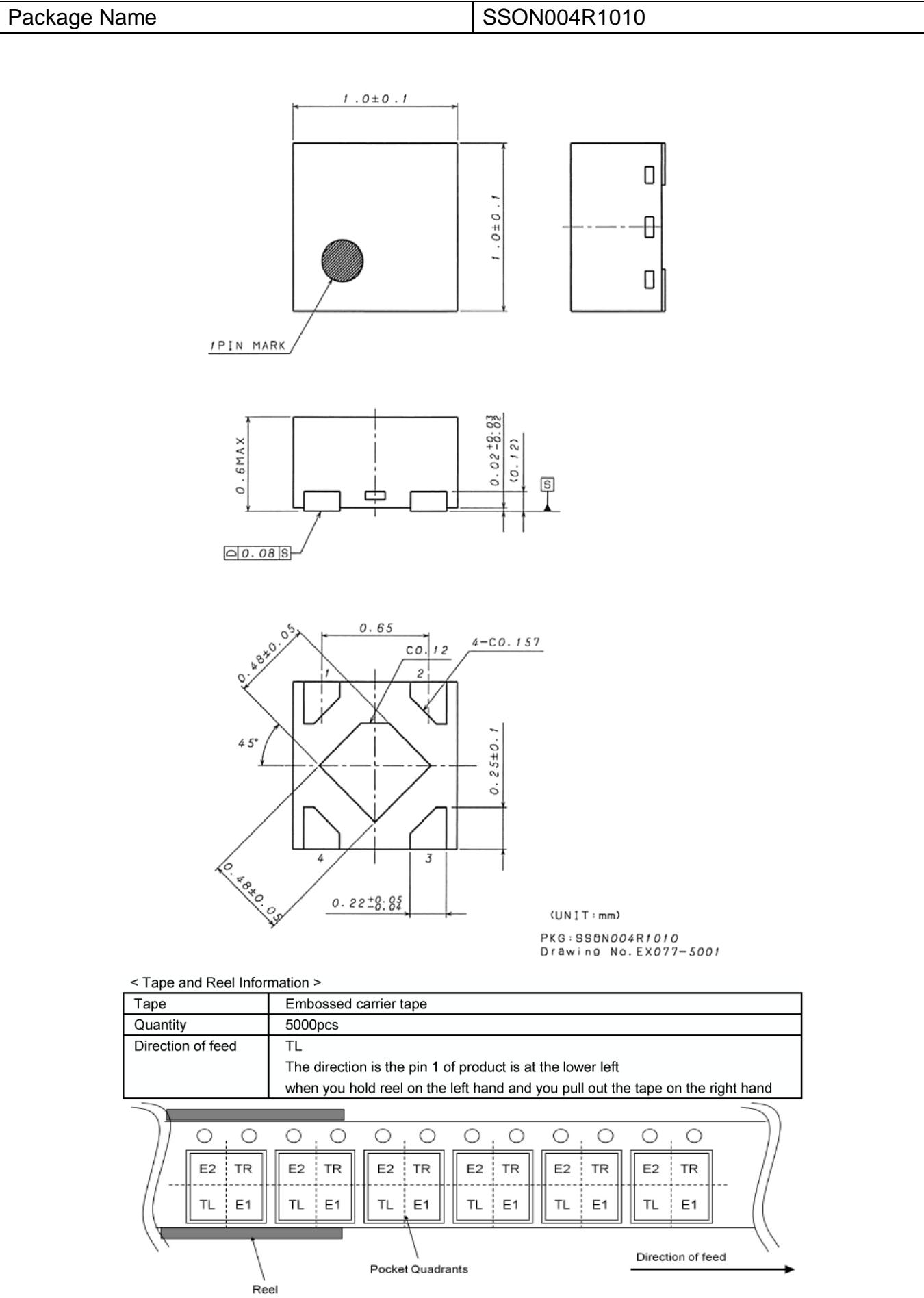
Figure 42. Input / Output equivalent circuit

Marking Diagram



Part Number	Output Voltage [V]	Part Number Marking
BU10JA2MNVX-C	1.0	5
BU11JA2MNVX-C	1.1	6
BU12JA2MNVX-C	1.2	4
BU1CJA2MNVX-C	1.25	3
BU15JA2MNVX-C	1.5	2
BU18JA2MNVX-C	1.8	Q
BU25JA2MNVX-C	2.5	1
BU28JA2MNVX-C	2.8	U
BU2JJA2MNVX-C	2.85	0
BU29JA2MNVX-C	2.9	Ui
BU30JA2MNVX-C	3.0	Y
BU33JA2MNVX-C	3.3	R
BU34JA2MNVX-C	3.4	Yi

Physical Dimension and Packing Information



Revision History

Date	Revision	Changes
26.Dec.2014	001	New Release.
27.Aug.2015	002	P2 Add Lineup.
11.Apr.2016	003	Applied the ROHM Standard Style and improved understandability. Add Equivalence Circuits.
21.Mar.2017	004	<p>p.1-20 Update of the footer. (Applied the rule.)</p> <p>p.2 The voltage lineup is added. (Output Voltage:1.1V)</p> <p>p.4 Changed the expression from "Power dissipation" to "Thermal Resistance". (Based on the JEDEC standard)</p> <p>p.5 Temperature condition of "Electrical Characteristics" is added. Changed the expression from "Operating Current" to "Circuit Current".</p> <p>p.8 Unified the item name of figure 14 for the parameter name of "Electrical Characteristics".</p> <p>p.13 Changed the expression from "About power dissipation(Pd)" to "Power Dissipation". (Based on the change of p.4)</p> <p>p.14 The item of "Thermal Design" is added. (Based on the change of p.4)</p> <p>p.15 The item of "Calculation Example(SSON004R1010)" is added. (Based on the change of p.4)</p> <p>p.17 Item of VIN is added in I/O Equivalence Circuits and resistance value is listed in I/O Equivalence Circuits.</p> <p>p.18 The lineup of "Marking Diagram" is added. (Output Voltage:1.1V)</p> <p>p.19 Update "Physical Dimension Tape and Reel Information" to the latest version.</p>
12. Mar. 2018	005	<p>p.2 Add Lineup. Added the electrode position of the back side to "Pin Configurations" in a dashed line.</p> <p>p.16-18 Added of the operation notes about the use of general linear regulator.</p> <p>p.20 Add Lineup</p> <p>p.21 Add Lineup</p>
9. Jul. 2018	006	<p>p.2 Add Lineup</p> <p>p.5 Correction of conditions errors in Ripple Rejection Ratio.</p> <p>p.20 Add Lineup</p> <p>p.21 Add Lineup</p> <p>Others, correction of errors.</p>

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(Note1) Medical Equipment Classification of the Specific Applications

JAPAN	USA	EU	CHINA
CLASS III	CLASS III	CLASS II b	
CLASS IV		CLASS III	CLASS III

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