

# 4.2V to 18V Input 1ch Buck Controller

## BD9305AFVM

#### **General Description**

BD9305AFVM is a 1-channel DC/DC converter controller. A Step-down DC/DC converter can be configured by BD9305AFVM. In addition, it has a built-in master-slave function which improves synchronization.

#### **Features**

- 1ch PWM Controlled DC/DC Converter Controller
- Built-in Soft Start Function
- Built-in Master / Slave Function
- Protection Circuits:
  - Under Voltage Lockout Protection Circuit
  - > Thermal Shutdown Circuit
  - > Short Protection Circuit of Timer Latch type

#### **Applications**

- TV, Power Supply for the TFT-LCD Panels used for LCD TVs, Back Lights
- DSC, DVC, Printer, DVD ,DVD Recorder, General Consumer Equipment, etc.

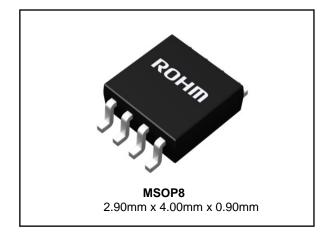
#### **Key Specifications**

Power Supply Voltage Range: 4.2V to 18V
 Error Amplifier Feed Back Voltage: 1.25±1.6%
 Oscillating Frequency Range: 100kHz to 800kHz
 Standby Current: 0µA(Typ)

■ Operating Temperature Range: -40°C to +85°C

#### Package

W(Typ) x D(Typ) x H(Max)



#### **Typical Application Circuit**

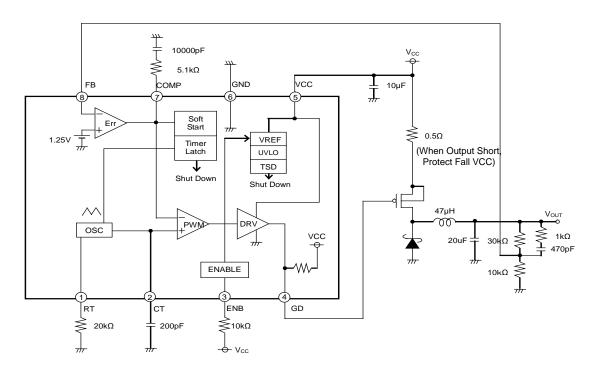
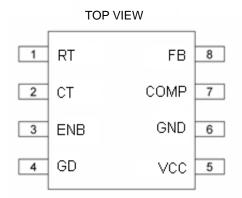


Figure 1. Typical Application Circuit

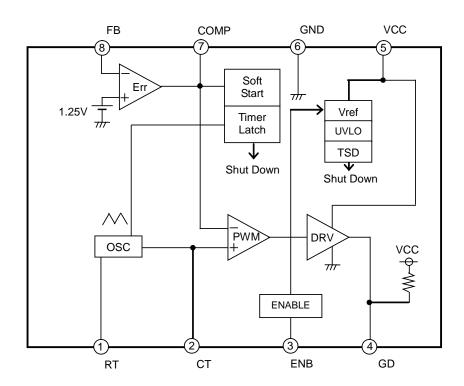
## **Pin Configuration**



**Pin Descriptions** 

Pin No	Pin Name	Function			
1	RT	External timing resistor pin			
2	СТ	External timing capacitor pin			
3	ENB	Control pin			
4	GD	Gate drive output pin			
5	VCC	Power supply pin			
6	GND	Ground pin			
7	COMP	Error amp output pin			
8	FB	Error amp inversion input pin			

## **Block Diagram**



#### **Block Operation**

#### 1. Error Amplifier (Err)

It compares a reference voltage of 1.25V (TYP) and the output feedback voltage.

This block produces the COMP terminal voltage that determines the duty cycle.

#### 2. Oscillator (OSC)

This block determines the switching frequency by RT and CT values. The triangular wave is determined by RT and CT.

#### 3. PWM

The duty cycle is determined by comparing the output of Error amplifier and the angular wave of Oscillator.

#### 4. DRV

This block drives the gate of the external power FET by the PWM switching Duty.

#### 5. VREF

This block outputs the internal reference voltage of 2.5V (TYP).

This circuit's reference voltage is controlled (ON / OFF) by the ENB terminal.

#### 6. Protection Circuits (UVLO / TSD)

UVLO (low-voltage Lock Out circuit) shuts down the circuits when the voltage is below 3.5V (MIN).

TSD (temperature protection circuit) shuts down the IC when the temperature reaches 175°C (TYP).

#### 7. Soft Start Circuit

The Soft Start Circuit limits the current when the output voltage is slowly increasing during start-up.

Through this, the overshoot of output voltage and current sinking can be prevented.

#### 8 Timer Latch

It is an output short protection circuit that detects if the output of error amplifier (COMP voltage) is more than 1.7V (TYP). If the COMP voltage becomes more than 1.7V, the counter begins to operate. The LATCH is locked when the counter counts to 2200 and the GD output shuts down. The frequency of counter is determined by RT and CT. Once the LATCH was locked, the GD output will not operate until it is restarted by ENB or VCC. If the output short is removed while the Timer latch is counting, the counter will be reset.

Absolute Maximum Ratings (Ta = 25°C)

Parameter	Symbol	Limit	Unit
Power Supply Voltage (Note 2)	Vcc	20	V
Power Dissipation	Pd	0.58 <sup>(Note 1)</sup>	W
Operating Temperature Range	Topr	-40 to +85	°C
Storage Temperature Range	Tstg	-55 to +150	°C
Maximum Junction Temperature	Tjmax	150	°C

(Note 1) When mounted on a glass epoxy 4-layer board (70 mm x 70 mm x 1.6 mm). Derate by 4.7 mW/°C for Ta over 25°C. (Note 2) Must not exceed Pd.

Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Recommended Operating Conditions (Ta=-40°C to +85°C)

Parameter	Symbol		Unit		
Farameter		Min	Тур	Max	Offit
Power Supply Voltage	Vcc	4.2	12	18	V
Control Voltage	$V_{ENB}$	-	-	Vcc	V
Timing Capacitance	Сст	100	-	1000	pF
Timing Resistance	R <sub>RT</sub>	5	-	50	kΩ
Oscillating Frequency	fosc	100	-	800	kHz

Electrical Characteristics (Unless otherwise specified Ta=25°C, Vcc=12V, Cct=200pF, Rrt=20kΩ)

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Parameter	Symbol	Min	Тур	Max	Unit	Conditions
Triangular Waveform Oscillator Blo	ck】			1		
Oscillating Frequency	fosc	165	220	275	kHz	Vcc=5V
Charge Threshold Voltage	Vosc+	0.80	0.85	0.90	V	
Discharge Threshold Voltage	Vosc -	0.20	0.25	0.30	V	
Under-voltage Lockout Protection	Circuit ]			I.		
Threshold Voltage	$V_{UT}$	3.5	-	4.2	V	
[Error Amp Block]						I
Feed Back Voltage	$V_{FB}$	1.230	1.250	1.270	V	
Input Bias Current	I <sub>IB</sub>	-	0.05	1	μA	V <sub>FB</sub> =1.5V
COMP Sink Current	loı	35	50	65	μA	V <sub>FB</sub> =1.5V V <sub>COMP</sub> =1.25V
COMP Source Current	I <sub>00</sub>	35	50	65	μA	V <sub>FB</sub> =1.0V V <sub>COMP</sub> =1.25V
【Gate Drive Block】	•				•	
ON-Resistance	Ron	-	5	-	Ω	
Gate Drive Voltage L	V <sub>GDL</sub>	-	0	0.5	V	No Load
Gate Drive Voltage H	$V_{GDH}$	V <sub>CC</sub> -0.5	Vcc	-	V	No Load
MAX Duty	MDT	-	-	100	%	Vcc=5V
[Control Block]						
ON Voltage	V <sub>ON</sub>	2	-	-	V	
OFF Voltage	Voff	-	-	0.3	V	
ENB Sink Current	I <sub>ENB</sub>	40	60	90	μA	V <sub>ENB</sub> =5V
[Soft Start Block]						
Soft Start Time	ts	-	10	-	ms	
Timer Latch Protection Circuit						
Latch Detection COMP Voltage	V <sub>LC</sub>	1.5	1.7	1.9	V	
Latch Delay OSC Count Number	CNT	-	2200	-	COUNT	
Latch Delay Time	DLY	-	10	-	ms	
[Overall]						
Standby Current	Istby	-	0	10	μA	ENB=0FF
Average Consumption Current	Icc	1.0	1.5	2.5	mA	No Switching

## **Typical Performance Curves**

(Unless otherwise specified, V<sub>CC</sub>=12V, Ta=25°C)

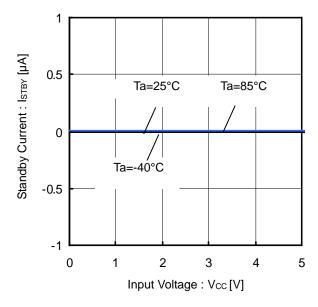


Figure 2. Standby Current vs Input Voltage

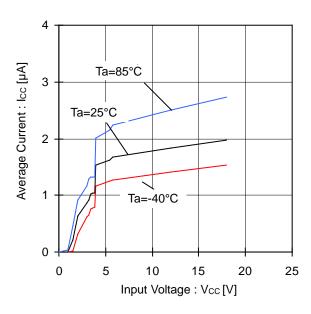


Figure 3. Average Consumption Current vs Input Voltage

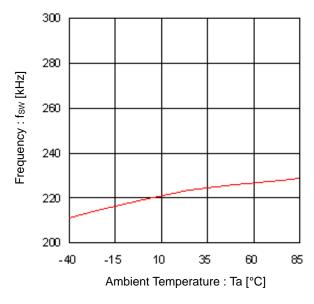


Figure 4. Frequency vs Temperature

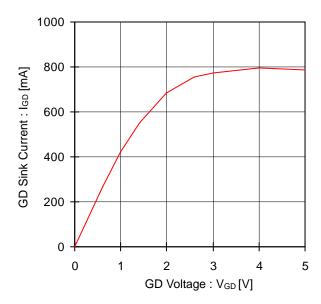


Figure 5. GD Sink Current vs GD Voltage

## **Typical Performance Curves – continued**

(Unless otherwise specified, V<sub>CC</sub>=12V, Ta=25°C)

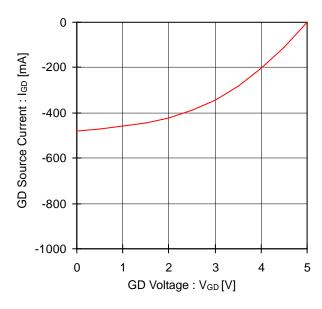


Figure 6. GD Source Current vs GD Voltage

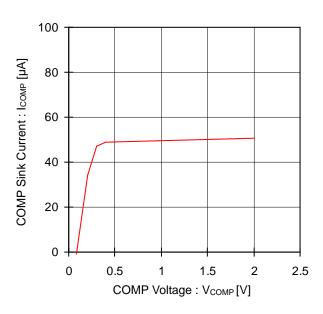


Figure 7. COMP Sink Current vs COMP Voltage

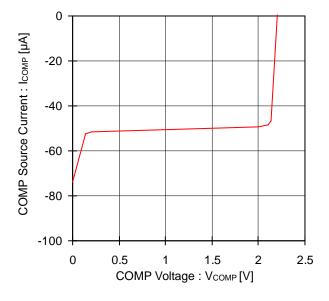


Figure 8. COMP Source Current vs COMP Voltage

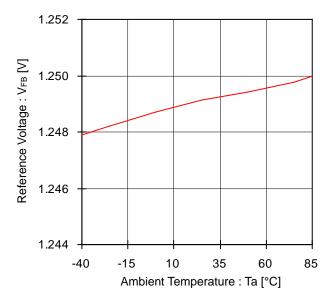


Figure 9. Reference Voltage vs Ambient Temperature

## **Typical Performance Curves – continued**

(Unless otherwise specified, V<sub>CC</sub>=12V, Ta=25°C)

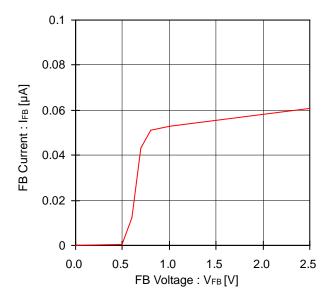


Figure 10. FB Input Bias Current vs FB Voltage

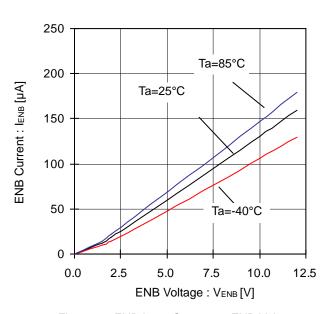


Figure 11. ENB Input Current vs ENB Voltage

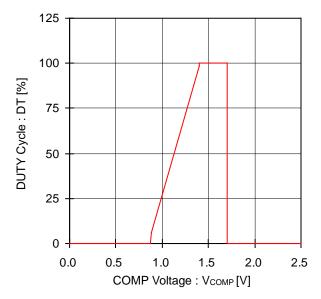


Figure 12. DUTY Cycle vs COMP Voltage

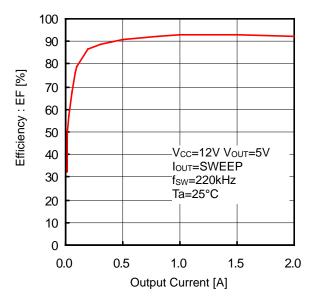


Figure 13. Efficiency vs Output Current

## **Typical Waveforms**

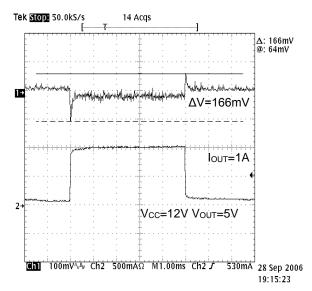


Figure 14. Load Response

#### **Application Information**

#### 1. Selecting Application Components

(1) Setting the Output L Constant (Step Down DC/DC) The inductance L used for output was decided by the rated current I<sub>LR</sub> and input current maximum value I<sub>OMAX</sub> of the inductance.

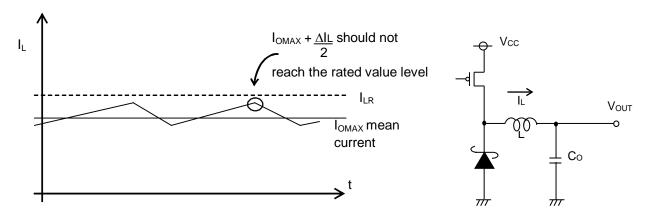


Figure 15. Coil Current Waveform (Step Down DC/DC)

Figure 16. Output Application Circuit (Step Down DC/DC)

Adjust so that  $I_{OMAX} + \Delta I_L/2$  does not reach the rated current value  $I_{LR}$ . At this time,  $\Delta I_L$  can be obtained by the following equation.

$$\Delta I_L = \frac{1}{L} \times \left( V_{CC} - V_{OUT} \right) \times \frac{V_{OUT}}{V_{CC}} \times \frac{1}{f} \quad [A]$$

Set a sufficient margin because the inductance L value may have ± 30% dispersion.

If the coil current exceeds the rating current ILR of the coil, it can cause damage to the IC internal elements.

(2) Setting the Output L Constant (Step Up DC/DC)

The inductance L to use for output is decided by the rated current I<sub>LR</sub> and input current maximum value I<sub>INMAX</sub> of the inductance.

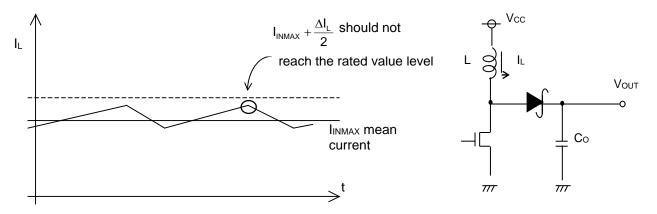


Figure 17. Coil Current Waveform (Step Up DC/DC)

Figure 18. Output Application Circuit (Step Up DC/DC)

Adjust so that  $I_{\text{INMAX}} + \Delta I_{\text{L}}/2$  does not reach the rated current value  $I_{\text{LR}}$ . At this time,  $\Delta I_{\text{L}}$  can be obtained by the following equation.

$$\Delta I_L = \frac{1}{L} V_{CC} \times \frac{V_{OUT} - V_{CC}}{V_{OUT}} \times \frac{1}{f} \ \left[A\right]$$

where:

f is the switching frequency

Set a sufficient margin because the inductance L value may have ± 30% dispersion.

If the coil current exceeds the rating current ILR of the coil, it can cause damage to the IC internal elements.

#### (3) Setting the Output Capacitor

For the output capacitor C, select a capacitor which has a larger value at the ripple voltage VPP allowance value and the drop voltage allowance value when there's a sudden load change.

Output ripple voltage is determined by the following equation.

$$\Delta Vpp = \Delta I_L \times R_{ESR} + \frac{\Delta I_L}{2C_O} \times \frac{V_{OUT}}{V_{CC}} \times \frac{1}{f} [V]$$
 (Step Down DC/DC)

$$\Delta Vpp = I_{LMAX} \times R_{ESR} + \frac{1}{fC_O} \times \frac{V_{CC}}{V_{OUT}} \times \left(I_{LMAX} - \frac{\Delta I_L}{2}\right) \quad \text{[V]} \tag{Step Up DC/DC)}$$

Apply the setting so that the voltage is within the allowable ripple voltage range.

For the drop voltage during the sudden load change (VDR), perform a rough calculation by the following equation.

$$\Delta V_{DR} = \frac{\Delta I_L}{C_O} \times 10 \,\mu \sec \quad [V]$$

However, 10 µs is the rough calculation value of the DC/DC response speed. Set the capacitance while considering a sufficient margin so that these two values are within the standard value range.

#### (4) Setting of Feedback Resistance Constant

Refer to the following formula for setting of feedback resistance.

$$V_{OUT} = \frac{R_1 + R_2}{R_2} \times 1.25 \text{ [V]}$$

It is recommend to use  $10k\Omega$  to  $330k\Omega$  setting range. If a resistance below  $10k\Omega$  was set, voltage efficiency will be dropped. If a resistance of more than  $330k\Omega$  was set, the offset voltage becomes large because of the internal error amplifier's input bias current of  $0.05\mu A(Typ)$ .

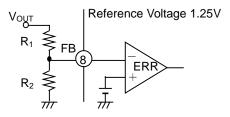


Figure 19. Feedback Resistance Setting

#### (5) Setting of Oscillating Frequency

The angular wave oscillation frequency can be set by connecting a resistor and a capacitor to RT (Pin 1) and CT (Pin 2) respectively. The charge and discharge currents at the capacitor of CT will be determined by the RT resistor. Refer to the configuration below for setting the RT's resistor and the CT's capacitor.

 $R_{RT}$ :  $5k\Omega$  to  $50k\Omega$ ,  $C_{CT}$ : 100pF to 1000pF. The frequency range of 100kHz to 800kHz are recommended.

Remember that the switching will stop if your setting is off this range.

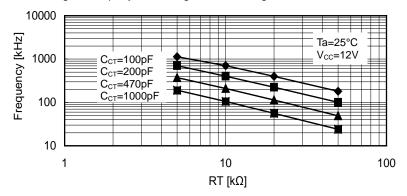


Figure 20. Frequency Setting

#### Selection of Input Capacitor

For DC/DC converter, the capacitor at the input side is also necessary because maximum current will be flowing between input and output. Therefore, it is recommended that an input capacitor with over 10µF and low ESR below 100mΩ. If a selected capacitor is outside this range, excessive large ripple voltage will overlap with the input voltage which may cause IC malfunction. However, this condition varies with negative overcurrent, input voltage, output voltage, inductor's value, and switching frequency so make sure to have a margin check with actual devices.

#### Selection of Output Rectifier Diode

Schottky barrier diode is recommended as the diode for rectification at the output stage of DC/DC converter. Refer below for choosing the maximum inductor current, the maximum output voltage, and the power supply voltage. <step-down DC/DC>

 $I_{OMAX} + \frac{\Delta I_L}{2}$  < Diode's rated current Maximum inductor current

Power supply voltage Vcc < Diode's rated voltage

<step-up DC/DC>

 $I_{INMAX} + \frac{\Delta I_L}{2}$ < Diode's rated current Maximum inductor current

Maximum output voltage  $V_{\text{OMAX}}$ < Diode's rated voltage

Furthermore, each parameter has a deviation of 30% to 40%, so design in such a way that you have provided enough margin for the deviation in your design.

#### (8) Setting of Power FET

If step-down DC/DC was configured by BD9305AFVM, Pch FET is necessary. Consider the following conditions when you choose: <step-down DC/DC>

> $I_{OMAX} + \frac{\Delta I_L}{2}$  < FET's rated current Maximum inductor current

V<sub>CC</sub> V<sub>CC</sub> < FET's rated voltage
> FET's gate ON voltage
< 2000pF Power supply voltage Power supply voltage

Gate capacitance (Note 1)

<step-up DC/DC>

 $I_{INMAX} + \frac{\Delta I_L}{2}$ Maximum inductor current < FET's rated current

 $V_{OMAX}$ Maximum output voltage FET's rated voltage Vcc Power supply voltage FET's gate ON voltage

Gate capacitance (Note 1) CGATE 2000pF

Furthermore, each parameter has a deviation of 30% to 40%, so design in such a way that you have provided enough margin for the deviation in your design.

(Note 1) If the Gate capacity becomes large, the switching speed will be slower, which may cause heat generation and breakdown, so check thoroughly the actual devices

Phase Compensation

Phase Setting Method

The following conditions are required to ensure the stability of the negative feedback circuit.

Phase lag should be 150° or lower when gain is 1 (0 dB) (phase margin of 30° or higher).

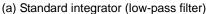
Because DC/DC converter applications are sampled using the switching frequency, the overall GBW should be set to 1/10 the switching frequency or lower. The target application characteristics can be summarized as follows:

- (a) Phase lag should be 150° or lower when gain is 1 (0 dB) (phase margin of 30° or higher).
- (b) The GBW at that time (i.e., the frequency of a 0-dB gain) is 1/10 of the switching frequency or below.

In other words, because the response is limited by the GBW, it is necessary to use higher switching frequencies to raise response.

One way to maintain stability through phase compensation involves cancellation of the secondary phase lag (-180°) caused by LC resonance with a secondary phase advance (by inserting 2 phase advances).

The GBW (i.e., the frequency with the gain set to 1) is determined by the phase compensation capacitor connected to the error amp. Increase the capacitance if a GBW reduction is required.





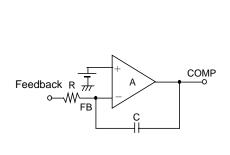
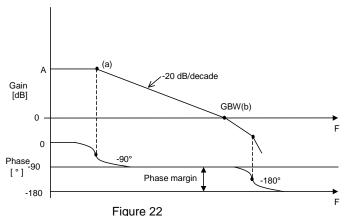


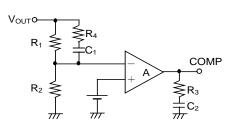
Figure 21



Point (a) 
$$fa = \frac{1}{2\pi RCA}$$
 [Hz]

Point (b) 
$$fb = GBW = \frac{1}{2\pi RC}$$
 [Hz]

The error amp performs the phase compensation at points (a) and (b) and it acts as a low-pass filter. For DC/DC converter applications, R refers to feedback resistors connected in parallel. From the LC resonance of output, the number of phase advances to be inserted is two.



LC resonant frequency

$$fp = \frac{1}{2\pi\sqrt{LC}} \quad [Hz]$$

Phase advance

$$fz1 = \frac{1}{2\pi C_1 R_1} \quad [Hz]$$

Phase advance

$$fz2 = \frac{1}{2\pi C_2 R_2} \quad [Hz]$$

Figure 23

Set a phase advance frequency close to the LC resonant frequency for the purpose of canceling the LC resonance.

(Note) If high-frequency noise is generated in the output, FB is affected through capacitor  $C_1$ . Therefore, insert the resistor  $R_4$ =1k $\Omega$  or so, which is in series with capacitor  $C_1$ .

#### 2.Example of Application

(Note) We strongly recommend the following application circuit examples but check thoroughly the characteristics before putting using them.

When you made changes at the external circuit, design a sufficient margin after considering the deviation, etc. of the external components and ROHM IC in terms of not only the static characteristic but also the transient characteristic. Moreover, understand that our company can not confirm fully with regard to the patent right.

#### <Master Slave Function>

The master slave function is built-in. Synchronous switching is possible by the multi-connection of BD9305AFVM/BD9306AFVM ICs. The following drawing shows an example of circuit connection in which BD9305AFVM is connected on the master side and BD9306AFVM is connected on the slave side.

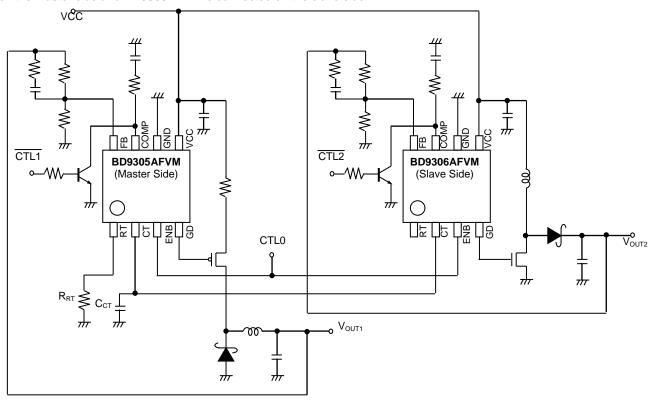


Figure 24. Master Slave Application Circuit

In the circuit above, BD9306AFVM is synchronized with the switching frequency which is determined by RT and CT of BD9305AFVM (master). In addition, the ON/OFF of output can be controlled by connecting the switch to the COMP terminal. (Refer to the following table)

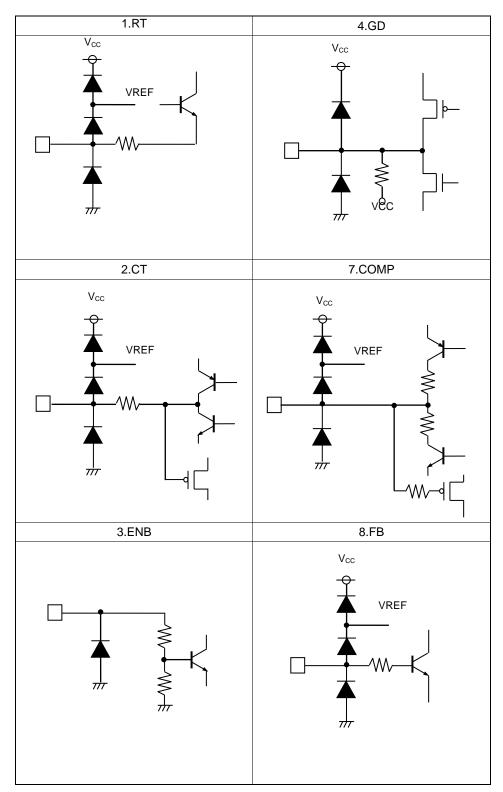
Control signal correspondence table

Output state		Control signal			
V <sub>OUT1</sub>	V <sub>OUT2</sub>	CTL0	CTL1	CTL2	
OFF	OFF	Low	(Note)	(Note)	
OFF	ON	High	High	Low	
ON	OFF	High	Low	High	
ON	ON	High	Low	Low	

(Note) The same in either case of High / Low.

Similarly in the case of connecting three or more than three, synchronization is still possible by connecting the CT terminal of Master and the CT terminal of Slave.

## I/O Equivalent Circuits



#### **Operational Notes**

#### 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

#### 2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

#### 3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

#### 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

#### 5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. The absolute maximum rating of the Pd stated in this specification is when the IC is mounted on a 70mm x 70mm x 1.6mm glass epoxy board. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

#### 6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

#### 7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

#### 8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

#### 9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

#### 10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

#### **Operational Notes - continued**

#### 11. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

#### 12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

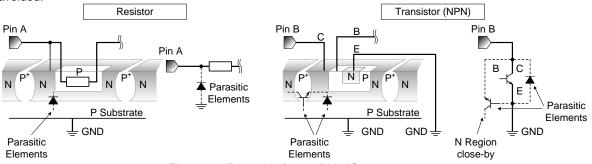
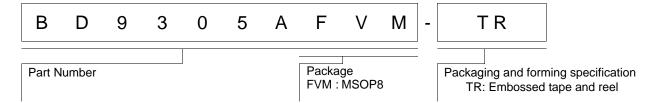


Figure 25. Example of monolithic IC structure

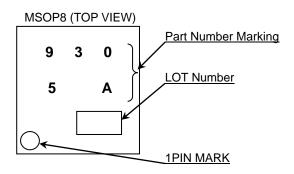
#### 13. Over Current Protection Circuit (OCP)

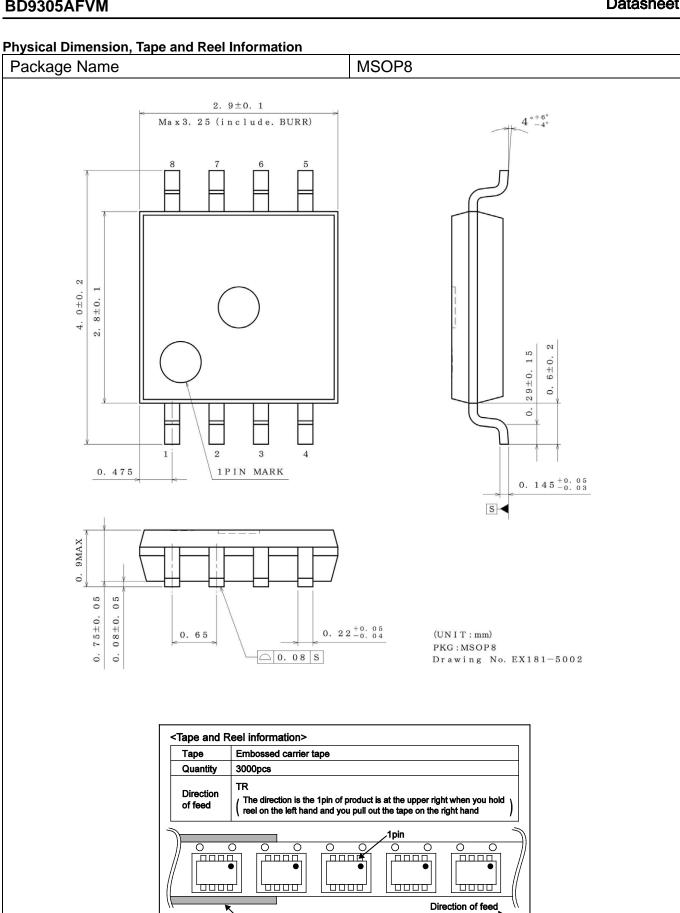
This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

## **Ordering Information**



## **Marking Diagram**





\*Order quantity needs to be multiple of the minimum quantity.

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## **Revision History**

Date	Revision	Changes
09.Sep.2014	001	New Release

## **Notice**

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Our Products are designed and manufactured for application in ordinary electronic equipments (such as AV equipment, OA equipment, telecommunication equipment, home electronic appliances, amusement equipment, etc.). If you intend to use our Products in devices requiring extremely high reliability (such as medical equipment (Note 1), transport equipment, traffic equipment, aircraft/spacecraft, nuclear power controllers, fuel controllers, car equipment including car accessories, safety devices, etc.) and whose malfunction or failure may cause loss of human life, bodily injury or serious damage to property ("Specific Applications"), please consult with the ROHM sales representative in advance. Unless otherwise agreed in writing by ROHM in advance, ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of any ROHM's Products for Specific Applications.

(Note1) Medical Equipment Classification of the Specific Applications

JÁPAN	USA	EU	CHINA	
CLASSⅢ	CL ACCIII	CLASS II b	CL A CC TT	
CLASSIV	CLASSⅢ	CLASSⅢ	CLASSⅢ	

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  - [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl<sub>2</sub>, H<sub>2</sub>S, NH<sub>3</sub>, SO<sub>2</sub>, and NO<sub>2</sub>
  - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
  - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
  - [f] Sealing or coating our Products with resin or other coating materials
  - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
  - [h] Use of the Products in places subject to dew condensation
- 4. The Products are not subject to radiation-proof design.
- 5. Please verify and confirm characteristics of the final or mounted products in using the Products.
- 6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- 7. De-rate Power Dissipation (Pd) depending on Ambient temperature (Ta). When used in sealed area, confirm the actual ambient temperature.
- 8. Confirm that operation temperature is within the specified range described in the product specification.
- 9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

#### Precaution for Mounting / Circuit board design

- 1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- 2. In principle, the reflow soldering method must be used; if flow soldering method is preferred, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

#### Precautions Regarding Application Examples and External Circuits

- If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
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#### **Precaution for Electrostatic**

This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of Ionizer, friction prevention and temperature / humidity control).

### **Precaution for Storage / Transportation**

- 1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
  - [a] the Products are exposed to sea winds or corrosive gases, including Cl2, H2S, NH3, SO2, and NO2
  - [b] the temperature or humidity exceeds those recommended by ROHM
  - [c] the Products are exposed to direct sunshine or condensation
  - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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QR code printed on ROHM Products label is for ROHM's internal use only.

#### **Precaution for Disposition**

When disposing Products please dispose them properly using an authorized industry waste company.

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