

# Sound Processor with Built-in 2-band Equalizer

### **BD37515FS**

#### **General Description**

BD37515FS is a sound processor with built-in 2-band equalizer for car audio. The functions are 4ch stereo input selector, input-gain control, main volume, loudness, and 5ch fader volume. Moreover, its "Advanced switch circuit", which is an original ROHM technology, can reduce various switching noise (ex. No-signal, low frequency like 20Hz & large signal inputs). "Advanced switch" makes control of microcomputer easier, supporting the construction of a high quality car audio system.

#### **Features**

- Reduce switching noise of input gain control, mute, main volume, fader volume, bass, treble, loudness by using advanced switch circuit.
- Built-in 1 differential input selector and 3 single-ended input selectors
- Built-in ground isolation amplifier inputs, ideal for external stereo input.
- Built-in input gain controller reduces switching noise for volume of a portable audio input.
- Decrease the number of external components due to built-in 2-band equalizer filter, LPF for subwoofer, and loudness filter. Also, it is possible to control Q, Gv, fo of 2-band equalizer, fc of LPF for subwoofer and Gv of loudness using I<sup>2</sup>C BUS control.
- It is possible to adjust the gain of the bass and treble up to ±20dB with 1 dB step gain adjustment.
- It is equipped with output terminals of Subwoofer.
- Energy-saving design resulting in low current consumption, by utilizing the Bi-CMOS process. It has the advantage in quality over scaling down the power heat control of the internal regulators.
- Input terminals and output terminals are organized and separately laid out to keep the signal flow in one direction which results in simpler and smaller PCB layout.
- It is possible to control the I<sup>2</sup>C BUS by 3.3V / 5V.

#### **Key Specifications**

Power Supply Voltage Range: 7.0V to 9.5V
 Circuit Current (No Signal): 38mA(Typ)
 Total harmonic Distortion 1:

(FRONT,REAR) 0.001%(Typ)

■ Total Harmonic Distortion 2:

(SUBWOOFER) 0.002% (Typ)
Maximum Input Voltage: 2.3Vrms (Typ)
Cross-talk Between Selectors: -100dB (Typ)

Volume Control Range: +15dB to -79dB

Output Noise Voltage1:

(FRONT,REAR) 3.8μVrms (Typ)

Output Noise Voltage2:

(SUBWOOFER) 4.8µVrms (Typ)

Residual Output Noise voltage: 1.8µVrms (Typ)Operating Temperature Range: -40°C to +85°C

#### **Package**

W(Typ) x D(Typ) x H(Max)

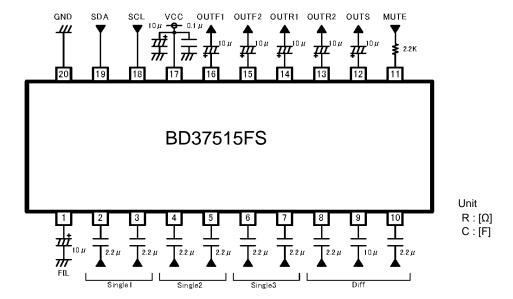


**SSOP-A20** 8.70mm x 7.80mm x 2.01mm

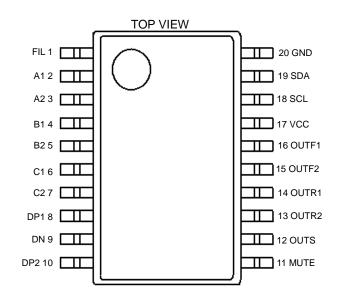
### **Applications**

It is optimal for use in car audio systems. It can also be used for audio equipment of mini Compo, micro Compo, TV. etc.

### **Typical Application Circuit**



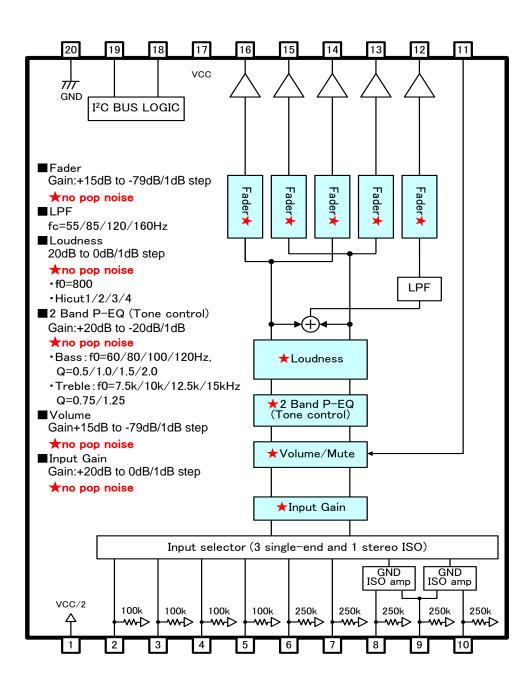
## **Pin Configuration**



**Pin Descriptions** 

2000. ipti	••				
Pin No.	Pin Name	Description	Pin No.	Pin Name	Description
1	FIL	VCC/2 terminal	11	MUTE	External compulsory mute terminal
2	A1	A input terminal of 1ch	12	OUTS	Subwoofer output terminal
3	A2	A input terminal of 2ch	13	OUTR2	Rear output terminal of 2ch
4	B1	B input terminal of 1ch	14	OUTR1	Rear output terminal of 1ch
5	B2	B input terminal of 2ch	15	OUTF2	Front output terminal of 2ch
6	C1	C input terminal of 1ch	16	OUTF1	Front output terminal of 1ch
7	C2	C input terminal of 2ch	17	VCC	Power supply terminal
8	DP1	D positive input terminal of 1ch	18	SCL	I <sup>2</sup> C Communication clock terminal
9	DN	D negative input terminal	19	SDA	I <sup>2</sup> C Communication data terminal
10	DP2	D positive input terminal of 2ch	20	GND	GND terminal

### **Block Diagram**



**Absolute Maximum Ratings** (Ta=25°C)

Parameter	Symbol	Rating	Unit
Power Supply Voltage	Vcc	10.0	V
Input Voltage	VIN	Vcc+0.3 to GND-0.3	V
Power Dissipation	Pd	0.94 <sup>(Note)</sup>	W
Storage Temperature	Tstg	-55 to +150	°C

(Note) This value derates by 7.5mW/°C for Ta=25°C or more when ROHM standard board is used. Thermal resistance θja = 133.3(°C/W)

ROHM Standard board Size: 70 x 70 x 1.6(mm<sup>3</sup>)

Material: A FR4 grass epoxy board(3% or less of copper foil area)

Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

**Recommended Operating Conditions** 

- <u></u>					
Parameter	Symbol	Min	Тур	Max	Unit
Power Supply Voltage	V <sub>CC</sub>	7.0	-	9.5	V
Temperature	Topr	-40	-	+85	V

#### **Electrical Characteristics**

(Unless specified otherwise, Ta=25°C,  $V_{CC}$ =8.5V, f=1kHz,  $V_{IN}$ =1Vrms, Rg=600 $\Omega$ , RL=10k $\Omega$ , A input, Input gain 0dB, Mute OFF, Volume 0dB, Tone control 0dB, Loudness 0dB, LPF OFF, Fader 0dB)

Ç	7 T, Volume odb, Tone control odb,		,	Limit	ader odb	,	
BLOCK	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
	Circuit Current	ΙQ	1	38	48	mA	No signal
	Voltage Gain	G∨	-1.5	0	+1.5	dB	Gv=20log(Vout/Vin)
	Channel Balance	СВ	-1.5	0	+1.5	dB	$CB = G_{V1}\text{-}G_{V2}$
	Total Harmonic Distortion 1 (FRONT,REAR)	THD+N1	1	0.001	0.05	%	V <sub>OUT</sub> =1Vrms BW=400Hz-30KHz
	Total Harmonic Distortion 2 (SUBWOOFER)	THD+N2	-	0.002	0.05	%	V <sub>OUT</sub> =1Vrms BW=400Hz-30KHz
RAL	Output Noise Voltage 1 (FRONT,REAR) *	V <sub>NO1</sub>	-	3.8	15	μVrms	$Rg = 0\Omega$ BW = IHF-A
GENERAL	Output Noise Voltage 2 (SUBWOOFER) *	V <sub>NO2</sub>	-	4.8	15	μVrms	$Rg = 0\Omega$ BW = IHF-A
0	Residual Output Noise Voltage *	V <sub>NOR</sub>	ı	1.8	10	μVrms	Fader = -∞dB Rg = 0Ω BW = IHF-A
	Cross-talk Between Channels *	СТС	-	-100	-90	dB	$Rg = 0\Omega$ $CTC=20log(V_{OUT}/V_{IN})$ $BW = IHF-A$
	Ripple Rejection	RR	-	-70	-40	dB	f=1KHz V <sub>RR</sub> =100mVrms RR=20log(Vcc IN/Vουτ)
	Input Impedance(A,B)	R <sub>IN_</sub> s	70	100	130	kΩ	
~	Input Impedance (C,D)	R <sub>IN_D</sub>	175	250	325	kΩ	
стоя	Maximum Input Voltage	VIM	2.1	2.3	-	Vrms	V <sub>IM</sub> at THD+N(V <sub>OUT</sub> )=1% BW=400Hz-30KHz
INPUT SELECTOR	Cross-talk Between Selectors *	CTS	-	-100	-90	dB	Rg = 0Ω CTS=20log(V <sub>OUT</sub> /V <sub>IN</sub> ) BW = IHF-A
INPL	Common Mode Rejection Ratio *	CMRR	50	65	-	dB	DP1 and DN input DP2 and DN input CMRR=20log(V <sub>IN</sub> /V <sub>OUT</sub> ) BW = IHF-A
SAIN	Minimum Input Gain	G <sub>IN_MIN</sub>	-2	0	+2	dB	Input gain 0dB V <sub>IN</sub> =100mVrms G <sub>IN</sub> =20log(V <sub>OUT</sub> /V <sub>IN</sub> )
INPUT GAIN	Maximum Input Gain	G <sub>IN_MAX</sub>	+18	+20	+22	dB	Input gain 20dB VIN=100mVrms GIN=20log(Vout/VIN)
	Gain Set Error	G <sub>IN_ERR</sub>	-2	0	+2	dB	GAIN=+1dB to +20dB

### **Electrical Characteristics - continued**

(Unless specified otherwise, Ta=25°C,  $V_{CC}$ =8.5V, f=1kHz,  $V_{IN}$ =1Vrms, Rg=600 $\Omega$ , R<sub>L</sub>=10k $\Omega$ , A input, Input gain 0dB, Mute OFF, Volume 0dB, Tone control 0dB, Loudness 0dB, LPF OFF, Fader 0dB)

	OFF, Volume 0dB, Tone control 0d	D, Loudiness	oub, Li	Limit	adei odi	) 	
BLOCK	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
MUTE	Mute Attenuation *	G <sub>мите</sub>	-	-105	-85	dB	Mute ON G <sub>MUTE</sub> =20log(V <sub>OUT</sub> /V <sub>IN</sub> ) BW = IHF-A
	Maximum Gain	G <sub>V_MAX</sub>	+13	+15	+17	dB	Volume = 15dB $V_{IN}$ =100mVrms $GV$ =20log( $V_{OUT}/V_{IN}$ )
VOLUME	Maximum Attenuation *	G <sub>V_MIN</sub>	-	-100	-85	dB	Volume = -∞dB GV=20log(V <sub>OUT</sub> /V <sub>IN</sub> ) BW = IHF-A
>	Attenuation Set Error 1	Gv_err1	-2	0	+2	dB	GAIN & ATT=+15dB to -15dB
	Attenuation Set Error 2	$G_{V\_ERR2}$	-3	0	+3	dB	ATT=-16dB to -47dB
	Attenuation Set Error 3	Gv_err3	-4	0	+4	dB	ATT=-48dB to -79dB
ပ္သ	Maximum Boost Gain	G <sub>B_BST</sub>	+18	+20	+22	dB	GAIN=+20dB f=100Hz V <sub>IN</sub> =100mVrms G <sub>B</sub> =20log (V <sub>OUT</sub> /V <sub>IN</sub> )
BASS	Maximum Cut Gain	<b>G</b> в_сит	-22	-20	-18	dB	Gain=-20dB f=100Hz V <sub>IN</sub> =2Vrms G <sub>B</sub> =20log (V <sub>OUT</sub> /V <sub>IN</sub> )
	Gain Set Error	$G_{B\_ERR}$	-2	0	+2	dB	Gain=+20dB to -20dB f=100Hz
Щ	Maximum Boost Gain	G <sub>T_BST</sub>	+18	+20	+22	dB	Gain=+20dB f=10kHz V <sub>IN</sub> =100mVrms G <sub>T</sub> =20log (V <sub>OUT</sub> /V <sub>IN</sub> )
TREBLE	Maximum Cut Gain	<b>G</b> т_сит	-23	-20	-17	dB	GAIN=-20dB f=10kHz V <sub>IN</sub> =2Vrms GT=20log (V <sub>OUT</sub> /V <sub>IN</sub> )
	Gain Set Error	G <sub>T_ERR</sub>	-2	0	+2	dB	Gain=+20dB to -20dB f=10kHz
~	Maximum Boost Gain	G <sub>F_BST</sub>	+13	+15	+17	dB	Fader=15dB V <sub>IN</sub> =100mVrms G <sub>F</sub> =20log(V <sub>OUT</sub> /V <sub>IN</sub> )
SUBWOOFER	Maximum Attenuation*	G <sub>F_MIN</sub>	-	-100	-90	dB	Fader = -∞dB G <sub>F</sub> =20log(V <sub>OUT</sub> /V <sub>IN</sub> ) BW = IHF-A
	Gain Set Error	G <sub>F_ERR</sub>	-2	0	+2	dB	Gain=+1dB to +15dB
_	Attenuation Set Error 1	G <sub>F_ERR1</sub>	-2	0	+2	dB	ATT=0dB to -15dB
ER.	Attenuation Set Error 2	G <sub>F_ERR2</sub>	-3	0	+3	dB	ATT=-16dB to -47dB
FADE	Attenuation Set Error 3	G <sub>F_ERR3</sub>	-4	0	+4	dB	ATT=-48dB to -79dB
<sub>7</sub>	Output Impedance	Rout	-	-	50	Ω	V <sub>IN</sub> =100mVrms
	Maximum Output Voltage	Vом	2	2.2	-	Vrms	THD+N=1% BW=400Hz-30KHz
LOUDNESS	Maximum Gain	G <sub>L_MAX</sub>	+17	+20	+23	dB	Gain 20dB V <sub>IN</sub> =100mVrms G <sub>L</sub> =20log(V <sub>OUT</sub> /V <sub>IN</sub> )
LOUE	Gain Set Error	G <sub>L_ERR</sub>	-2	0	+2	dB	GAIN=+20dB to +1dB

VP-9690A(Average value detection, effective value display) filter by Matsushita Communication is used for \* measurement. Phase between input / output is same.

### **Typical Performance Curves**

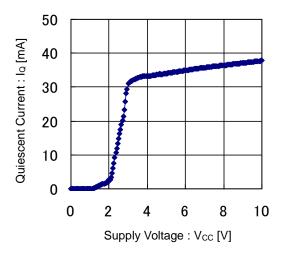


Figure 1. Quiescent Current vs Supply Voltage

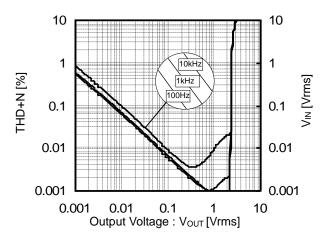


Figure 2. Total Harmonic Distortion vs Output Voltage

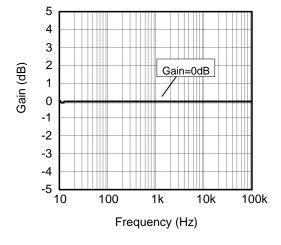
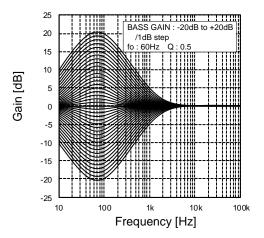


Figure 3. Gain vs Frequency



### **Typical Performance Curves - continued**

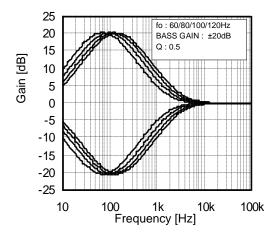


Figure 5. Bass fo vs Frequency (Bass fo is Changeable)

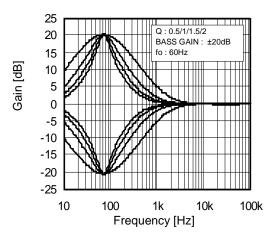


Figure 6. B Bass Q vs Frequency (Bass Q is Changeable)

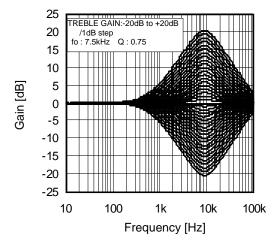


Figure 7. Treble Gain vs Frequency

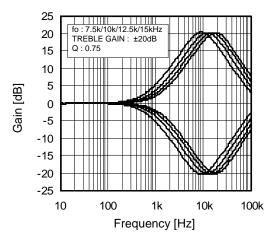


Figure 8. Treble fo vs Frequency (Treble fo is Changeable)

### **Typical Performance Curves - continued**

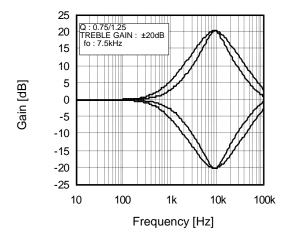


Figure 9. Treble Q vs Frequency (Treble Q is Changeable)

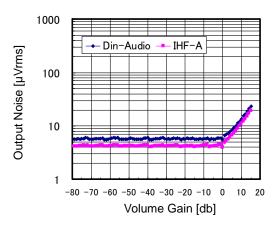


Figure 10. Output Noise vs Volume Gain

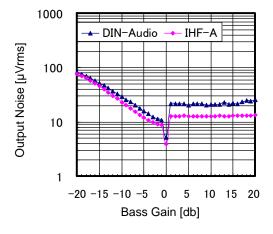


Figure 11. Output Noise vs Bass Gain

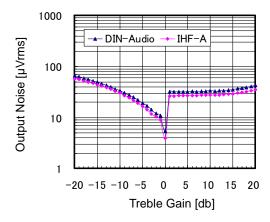


Figure 12. Output Noise vs Treble Gain

### **Typical Performance Curves - continued**

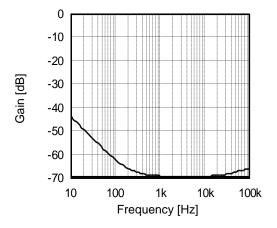


Figure 13. CMRR vs Frequency

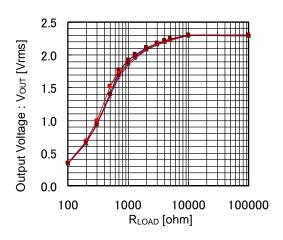


Figure 14. Output Voltage vs R<sub>LOAD</sub>

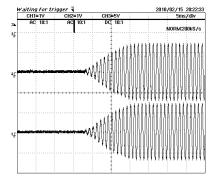


Figure 15. Advanced Switch 1

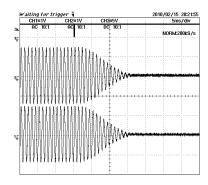


Figure 16. Advanced Switch 2

### **Timing Chart**

### **Control Signal Specification**

### (1) Electrical Specifications and Timing for Bus Lines and I/O Stage

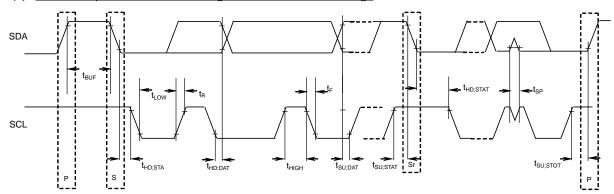


Figure 17. I<sup>2</sup>C-bus Signal Timing Diagram

Table 1 Characteristics of the SDA and SCL bus lines for  $I^2C$ -bus devices (Unless specified otherwise, Ta=25°C, Vcc=8.5V)

	Parameter	Symbol	Fast-mod	e I <sup>2</sup> C-bus	Unit
	Farameter	Symbol	Min	Max	Offic
1	SCL clock frequency	fscL	0	400	kHz
2	Bus free time between a STOP and START condition	<b>t</b> BUF	1.3	-	μS
3	Hold time (repeated) START condition. After this period, the first clock	tup oza	0.6		μS
3	pulse is generated	thd;sta	0.0		μΟ
4	LOW period of the SCL clock	tLOW	1.3	-	μS
5	HIGH period of the SCL clock	thigh	0.6	-	μS
6	Set-up time for a repeated START condition	tsu;sta	0.6	-	μS
7	Data hold time:	thd;dat	0.06 (Note)	-	μS
8	Data set-up time	tsu;dat	120	1	ns
9	Set-up time for STOP condition	tsu;sto	0.6	-	μS

All values referred to VIH Min and VIL Max Levels (see Table 2).

(Note) To avoid sending right after the fall-edge of SCL (VIH min of the SCL signal), the transmitting device should set a hold time of 300ns or more for the SDA signal.

For 7(thd;DAT), 8(tsu;DAT), make the setup in which the margin is fully in.

Table 2 Characteristics of the SDA and SCL I/O stages for I<sup>2</sup>C-bus devices

	Dorometer	Cumbal	Fast-mod	Unit	
	Parameter	Symbol	Min	Max	Unit
10	LOW level input voltage:	VIL	-0.3	+1	V
11	HIGH level input voltage:	V <sub>IH</sub>	2.3	5	V
12	Pulse width of spikes which must be suppressed by the input filter.	tsp	0	50	ns
13	LOW level output voltage: at 3mA sink current	V <sub>OL1</sub>	0	0.4	V
14	Input current of each I/O pin with an input voltage between 0.4V and 4.5V.	Iı	-10	+10	μA

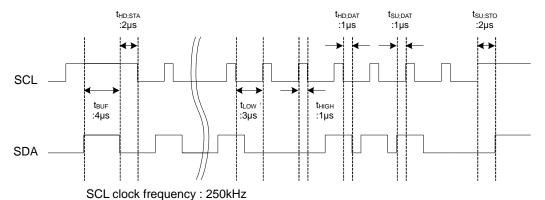


Figure 18. I<sup>2</sup>C Data Transmission Command Timing Diagram

### (2) <u>I<sup>2</sup>C BUS FORMAT</u>

	MSB LSB		MSB	LSB	MSB	LSB					
S	Slave Address	Α	Select Addr	ess A		Data	Α	Р			
1bit	: 8bit	1bit	8bit	1bit		8bit	1bit	1bit			
	S	= Start condition (Recognition of start bit)									
	Slave Address										
		The least significant bit is "L" which corresponds to write mode.									
	Α	= AC	KNOWLEDGE	bit (Recogni	tion of ac	knowledgement	:)				
	Select Address	dress = Select address corresponding to volume, bass or treble.									
	Data	Data = Data on every volume and tone.									
	Р	= Sto	op condition (Re	cognition of	stop bit)						

### (3) I<sup>2</sup>C BUS Interface Protocol

(a) Basic format

S	Slave Addre	SS	Α	Select Add	dress	Α	Da	ta	Α	Ρ
	MSB	LSB		MSB	LSB	Ν	1SB	LSE	3	

(b) Automatic Increment (Select Address increases (+1) according to the number of data.)

S	Slave Address	Α	Select Address	Α	Data1	Α	Data2	Α		DataN	Α	Р
1	MSB LSB	N	MSB LSE		MSB L	.SB	MSB L	SB	MS	BB LS	В	

(Example) 1 Data1 shall be set as data of address specified by Select Address.

- ② Data2 shall be set as data of address specified by Select Address +1.
- ③ DataN shall be set as data of address specified by Select Address +N-1.

(c) Configuration Unavailable for Transmission (In this case, only Select Address1 is set.)

S	Slave A	ddress	Α	Select A	ddress1	Α	Da	ata	Α	Select	Address	s 2	Α	Data	Α	Р
	MSB	LSE	3	MSB	LSB	N	ISB	LSE	3	MSB		LSB	MS	SB I	LSB	
	(Note	e) If any	data	a is transm	itted as Se	elec	t Ad	dress	s 2 r	next to da	ata, it is	reco	gnize	ed		
		as	data	, not as Se	elect Addre	SS	2.									

#### (4) Slave Address

MSB LSB												
	A6	A5	A4	A3	A2	A1	A0	R/W				
	1	0	0	0	0	0	0	0	80H			

#### (5) Select Address & Data

	Select	MSB			Da	ta			LSB
Items	Address (hex)	D7	D6	D5	D4	D3	D2	D1	D0
Initial setup 1	01	Advance d switch ON/OFF	0	of Input G					ced switch of Mute
Initial setup 2	02	LPF Phase 0°/180°	0	0	0	0	0	Subwoo	ofer LPF fc
Initial setup 3	03	0	0	0	1	0	0	0	1
Input Selector	05	0	0	0		In	out select	or	
Input gain	06	Mute ON/OFF	0	0		ı	nput Gair	1	
Volume gain	20			V	olume Gain /	Attenuatio	n		
Fader 1ch Front	28			F	ader Gain /	Attenuation	1		
Fader 2ch Front	29			F	ader Gain / A	Attenuation	1		
Fader 1ch Rear	2A			F	ader Gain /	Attenuation	1		
Fader 2ch Rear	2B			F	ader Gain / A	Attenuation	1		
Fader Subwoofer	2C			F	ader Gain /	Attenuation	1		
Bass setup	41	0	0	Ва	ss fo	0	0	Ва	ass Q
Test mode 1	44	0	0	0	0	0	0	0	0
Treble setup	47	0	0	Tre	ble fo	0	0	0	Treble Q
Bass gain	51	Bass Boost/ Cut	0	0	0 Bass Gain				
Test mode 2	54	1	0	0	0 0 0 0				0
Treble gain	57	Treble Boost/ Cut	0	0	Treble Gain				
Loudness Gain	75	0	Loudne	ness Hicut Loudness Gain					
System Reset	FE	1	0	0	0	0	0	0	1

Advanced switch

### Note

- 1. The Advanced Switch works in the latch part while changing from one function to another.
- 2. Upon continuous data transfer, the Select Address rolls over because of the automatic increment function, as shown below.

- 3. For the function of Input Selector etc, Advanced Switch is not used. Therefore, please apply mute on the set side when changing these settings.
- 4. When using mute function of this IC at the time of changing input selector, please switch mute ON/OFF while waiting for advanced-mute time.

Select address 01 (hex)

Time	MSB	MSB Advanced switch time of Mute									
Time	D7	D6	D5	D4	D3	D2	D1	D0			
0.6msec	A di (0 0 0 0 d		A al. (2 (2 a a a a a	avvitala tima a			0	0			
1.0msec	Advanced	0		switch time	0	0	0	1			
1.4msec	Switch ON/OFF	U		ain/Volume r/Loudness	U	U	1	0			
3.2msec	ON/OFF		Tone/Fade	i/Loudiless			1	1			

Time	MSB	Advanced switch time of Input gain/Volume/Tone/Fader/Loudness								
	D7	D6	D5	D4	D3	D2	D1	D0		
4.7 msec	A di 10 to o o o d		0	0						
7.1 msec	Advanced	witch 0	0	1	0	0	Advance	ed switch		
11.2 msec	ON/OFF		1	0			Time o	of Mute		
14.4 msec	OIN/OFF		1	1						

Mode	MSB	Advanced switch ON/OFF							
Mode	D7	D6	D5	D4	D3	D2	D1	D0	
OFF	0	0	Advanced switch time				Advance	Advanced switch	
ON	1	0	of Input gain/Volume Tone/Fader/Loudness		U	J	Time o	of Mute	

Select address 02(hex)

fo.	MSB			Subwoo	fer LPF fc			LSB
tc	D7	D6	D5	D4	D3	D2	D1	D0
OFF						0	0	0
55Hz						0	0	1
85Hz	LPF	0	0	_	0	0	1	0
120Hz	Phase	U	0	0	U	0	1	1
160Hz						1	0	0
Prohibition							Other setting	l

Phase	MSB		LSB					
Filase	D7	D6	D5	D4	D3	D2	D1	D0
0°	0	0	0	0	0	Subwoofer LPF fc		
180°	1	U	0	0	U			

Select address 05(hex)

Mode	MSB		I	nput S	electo	r		LSB
iviode	D7	D6	D5	D4	D3	D2	D1	D0
Initial					0	0	0	0
Α					0	0	0	1
В					0	0	1	0
С	0	0	0	0	0	0	1	1
D diff					0	1	1	1
Input SHORT					1	0	0	1
Prohibition						Other	setting	

Input SHORT : The input impedance of each input terminal is lowered from  $100k\Omega(TYP)$  to  $6~k\Omega(TYP)$ . (For quick charge of coupling capacitor)

:	Initial	condition

Select address 06 (hex)

Gain	MSB			Inp	ut Gain			LSB
Gain	D7	D6	D5	D4	D3	D2	D1	D0
0dB				0	0	0	0	0
1dB				0	0	0	0	1
2dB				0	0	0	1	0
3dB				0	0	0	1	1
4dB				0	0	1	0	0
5dB				0	0	1	0	1
6dB				0	0	1	1	0
7dB				0	0	1	1	1
8dB				0	1	0	0	0
9dB				0	1	0	0	1
10dB				0	1	0	1	0
11dB	Mute			0	1	0	1	1
12dB	ON/OFF	0	0	0	1	1	0	0
13dB				0	1	1	0	1
14dB				0	1	1	1	0
15dB				0	1	1	1	1
16dB				1	0	0	0	0
17dB				1	0	0	0	1
18dB				1	0	0	1	0
19dB	]			1	0	0	1	1
20dB	]			1	0	1	0	0
	]			1	1	0	1	1
Prohibition				:	:	:	:	:
				1	1	1	1	1

Mode	MSB	Mute ON/OFF							
Mode	D7	D6	D5	D4	D3	D2	D1	D0	
OFF	0	0	0			Innut Cain			
ON	1	U	U			Input Gain			

Select address 20, 28, 29, 2A, 2B, 2C (hex)

Gain & ATT	MSB		V	ol. Fader Ga	in / Attenuat	tion		LSB
Gain & ATT	D7	D6	D5	D4	D3	D2	D1	D0
	0	0	0	0	0	0	0	0
	0	0	0	0	0	0	0	1
Prohibition	:	:	:	:	:	:	:	:
	0	1	1	1	0	0	0	0
15dB	0	1	1	1	0	0	0	1
14dB	0	1	1	1	0	0	1	0
13dB	0	1	1	1	0	0	1	1
:	:	:	:	:	:	:	:	:
-77dB	1	1	0	0	1	1	0	1
-78dB	1	1	0	0	1	1	1	0
-79dB	1	1	0	0	1	1	1	1
	1	1	0	1	0	0	0	0
Prohibition	:	:	:	:	:	:	:	:
	1	1	1	1	1	1	1	0
-∞dB	1	1	1	1	1	1	1	1

: Initial condition

Select address 41(hex)

Ofastar	MSB	MSB Bass Q Factor									
Q factor	D7	D6	D5	D4	D3	D2	D1	D0			
0.5			Bass fo				0	0			
1.0						0	0	1			
1.5		0	Das	SS 10	0	U	1	0			
2.0	]						1	1			

40	MSB			Bas	s fo			LSB
to	D7	D6	D5	D4	D3	D2	D1	D0
60Hz			0	0				
80Hz	0		0	1		_	Bas	SS
100Hz	U	0	1	0		U	Bas Q fac	ctor
120Hz			1	1				

Select address 47 (hex)

0 (	MSB	MSB Treble Q Factor									
Q factor	D7	D6	D5	D4	D3	D2	D1	D0			
0.75	0	0	Trok	olo fo	0	0	0	0			
1.25	] 0	U	ITEL	ole fo	0	0	0	1			

fo	MSB			Treb	le fo			LSB
to	D7	D6	D5	D4	D3	D2	D1	D0
7.5kHz			0	0				
10kHz	]	_	0	1	_	_	0	Treble
12.5kHz	] "	0	1	0	] 0	0	0	Q factor
15kHz			1	1				

: Initial condition

Select address 51, 57 (hex)

Coin	MSB			Bass/ Tre	eble Gain			LSB																				
Gain	D7	D6	D5	D4	D3	D2	D1	D0																				
0dB				0	0	0	0	0																				
1dB				0	0	0	0	1																				
2dB				0	0	0	1	0																				
3dB				0	0	0	1	1																				
4dB				0	0	1	0	0																				
5dB							0	0	1	0	1																	
6dB				0	0	1	1	0																				
7dB						0	0	1	1	1																		
8dB								0	1	0	0	0																
9dB							0	1	0	0	1																	
10dB								0	1	0	1	0																
11dB	Bass/			0	1	0	1	1																				
12dB	Treble	0	0	0	1	1	0	0																				
13dB	Boost			0	1	1	0	1																				
14dB	/cut			0	1	1	1	0																				
15dB				0	1	1	1	1																				
16dB					1	0	0	0	0																			
17dB				1	0	0	0	1																				
18dB				1	0	0	1	0																				
19dB				1	0	0	1	1																				
20dB				1	0	1	0	0																				
				1	0	1	0	1																				
Prohibition																									:	÷	÷	:
1 1011101011				1	1	1	1	0																				
				1	1	1	1	1																				

Mode	MSB		Bass/ Treble Boost/Cut								
Wode	D7	D6	D5	D4	D3	D2	D1	D0			
Boost	0	0	0		De	ss/Treble G	nin				
Cut	1	U	U		Da	iss/ Heble G	ali i				

: Initial condition

Select address 75 (hex)

001001 0001000 10 (110	_/										
Mode	MSB			Loudne	ss Hicut			LSB			
iviode	D7	D6	D5	D4	D3	D2	D1	D0			
Hicut1		0	0								
Hicut2		0	1			audaaa Cai	in				
Hicut3	] "	1	0	Loudness Gain							
Hicut4		1	1								

Gain	MSB			Loudne	Loudness Gain						
Gaiii	D7	D6	D5	D4	D3	D2	D1	D0			
0dB				0	0	0	0	0			
1dB				0	0	0	0	1			
2dB				0	0	0	1	0			
3dB				0	0	0	1	1			
4dB				0	0	1	0	0			
5dB				0	0	1	0	1			
6dB				0	0	1	1	0			
7dB				0	0	1	1	1			
8dB				0	1	0	0	0			
9dB				0	1	0	0	1			
10dB				0	1	0	1	0			
11dB				0	1	0	1	1			
12dB	0	Loudne	ss Hicut	0	1	1	0	0			
13dB				0	1	1	0	1			
14dB				0	1	1	1	0			
15dB				0	1	1	1	1			
16dB				1	0	0	0	0			
17dB				1	0	0	0	1			
18dB				1	0	0	1	0			
19dB				1	0	0	1	1			
20dB				1	0	1	0	0			
				1	0	1	0	1			
Prohibition				:	:	:	:	:			
				1	1	1	1	1			

: Initial condition

(6) <u>About Power ON Reset</u> Built-in IC initialization is made during power ON of the supply voltage. Please send initial data to all addresses at supply voltage on. And please turn ON mute at the set side until this initial data is sent.

Danamatan	0		Limit		1.114	O and distingtion
Parameter	Symbol	Min	Тур	Max	Unit	Conditions
Rise Time of VCC t <sub>RISE</sub>		33	-	-	µsec	V <sub>CC</sub> rise time from 0V to 5V
VCC Voltage of Release Power ON Reset	V <sub>POR</sub>	-	4.1	-	V	

#### (7) About External Compulsory Mute Terminal

It is possible to force mute externally by setting an input voltage to the MUTE terminal.

Mute Voltage Condition	Mode
GND to 1.0V	MUTE ON
2.3V to V <sub>CC</sub>	MUTE OFF

Establish the voltage of MUTE in the condition to be defined.

## **Application Information**

### 1. Function and Specifications

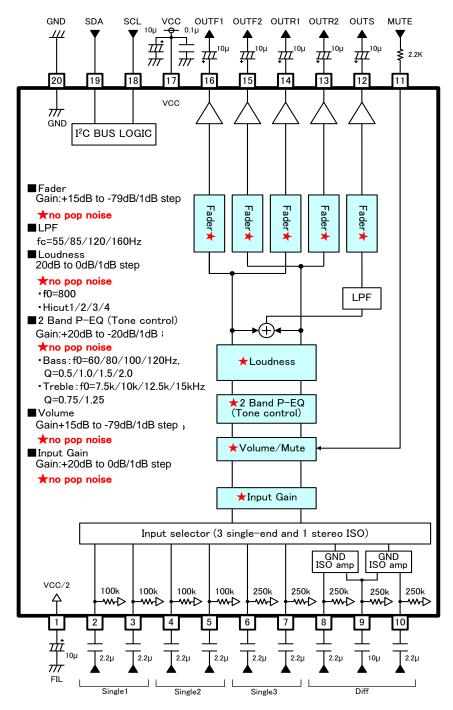
Function	Specification	s
Input selector	Stereo 3 input     Differential 1 input	
Input goin	• +20dB to 0dB (1dB step)	
Input gain	<ul> <li>Possible to use "Advanced switch" for prevention of sw</li> </ul>	itching noise.
Mute	<ul> <li>Possible to use "Advanced switch" for prevention of sw</li> </ul>	itching noise.
Volume	• +15dB to -79dB (1dB step), -∞dB	
volume	<ul> <li>Possible to use "Advanced switch" for prevention of sw</li> </ul>	itching noise.
Bass	<ul> <li>+20dB to -20dB (1dB step)</li> </ul>	• Q=0.5, 1, 1.5, 2
DdSS	<ul> <li>Possible to use "Advanced switch" at changing gain</li> </ul>	• fo=60, 80, 100, 120Hz
Treble	<ul> <li>+20dB to -20dB (1dB step)</li> </ul>	• Q=0.75, 1.25
Treble	<ul> <li>Possible to use "Advanced switch" at changing gain</li> </ul>	• fo=7.5k, 10k, 12.5k, 15kHz
Fader	<ul> <li>+15dB to -79dB(1dB step), -∞dB</li> </ul>	
rauei	<ul> <li>Possible to use "Advanced switch" for prevention of sw</li> </ul>	itching noise.
Loudness	• 20dB to 0dB(1dB step) • fo=800Hz	
Loudiless	<ul> <li>Possible to use "Advanced switch" for prevention of sw</li> </ul>	itching noise.
LPF	• fc=55/85/120/160Hz, pass • Phase shift (0°/180°	·)

2. V	olume .	/ Fader	Volume	Attenuation	Data
------	---------	---------	--------	-------------	------

2 <u>.                                    </u>	Volume /	rader	voiu	me A	ttenua	ation	Data											
	(dB)	D7	D6	D5	D4	D3	D2	D1	D0	(dB)	D7	D6	D5	D4	D3	D2	D1	D0
	+15	0	1	1	1	0	0	0	1	-33	1	0	1	0	0	0	0	1
	+14	0	1	1	1	0	0	1	0	-34	1	0	1	0	0	0	1	0
	+13	0	1	1	1	0	0	1	1	-35	1	0	1	0	0	0	1	1
	+12	0	1	1	1	0	1	0	0	-36	1	0	1	0	0	1	0	0
	+11	0	1	1	1	0	1	0	1	-37	1	0	1	0	0	1	0	1
	+10	0	1	1	1	0	1	1	0	-38	1	0	1	0	0	1	1	0
	+9	0	1	1	1	0	1	1	1	-39	1	0	1	0	0	1	1	1
	+8	0	1	1	1	1	0	0	0	-40	1	0	1	0	1	0	0	0
	+7	0	1	1	1	1	0	0	1	-41	1	0	1	0	1	0	0	1
	+6	0	1	1	1	1	0	1	0	-42	1	0	1	0	1	0	1	0
	+5	0	1	1	1	1	0	1	1	-43	1	0	1	0	1	0	1	1
	+4	0	1	1	1	1	1	0	0	-44	1	0	1	0	1	1	0	0
	+3	0	1	1	1	1	1	0	1	-45	1	0	1	0	1	1	0	1
	+2	0	1	1	1	1	1	1	0	-46	1	0	1	0	1	1	1	0
	+1	0	1	1	1	1	1	1	1	-47	1	0	1	0	1	1	1	1
	0	1	0	0	0	0	0	0	0	-48	1	0	1	1	0	0	0	0
	-1	1	0	0	0	0	0	0	1	-49	1	0	1	1	0	0	0	1
	-2	1	0	0	0	0	0	1	0	-50	1	0	1	1	0	0	1	0
	-3	1	0	0	0	0	0	1	1	-51	1	0	1	1	0	0	1	1
	-4	1	0	0	0	0	1	0	0	-52	1	0	1	1	0	1	0	0
	-5	1	0	0	0	0	1	0	1	-53	1	0	1	1	0	1	0	1
	-6	1	0	0	0	0	1	1	0	-54	1	0	1	1	0	1	1	0
	-7	1	0	0	0	0	1	1	1	-55	1	0	1	1	0	1	1	1
	-8	1	0	0	0	1	0	0	0	-56	1	0	1	1	1	0	0	0
	-9	1	0	0	0	1	0	0	1	-57	1	0	1	1	1	0	0	1
	-10	1	0	0	0	1	0	1	0	-58	1	0	1	1	1	0	1	0
	-11	1	0	0	0	1	0	1	1	-59	1	0	1	1	1	0	1	1
	-12	1	0	0	0	1	1	0	0	-60	1	0	1	1	1	1	0	0
	-13	1	0	0	0	1	1	0	1	-61	1	0	1	1	1	1	0	1
	-14	1	0	0	0	1	1	1	0	-62	1	0	1	1	1	1	1	0
	-15	1	0	0	0	1	1	1	1	-63	1	0	1	1	1	1	1	1
	-16	1	0	0	1	0	0	0	0	-64	1	1	0	0	0	0	0	0
	-17	1	0	0	1	0	0	0	1	-65	1	1	0	0	0	0	0	1
	-18	1	0	0	1	0	0	1	0	-66	1	1	0	0	0	0	1	0
	-19	1	0	0	1	0	0	1	1	-67	1	1	0	0	0	0	1	1
	-20	1	0	0	1	0	1	0	0	-68	1	1	0	0	0	1	0	0
	-21	1	0	0	1	0	1	0	1	-69	1	1	0	0	0	1	0	1
	-22	1	0	0	1	0	1	1	0	-70	1	1	0	0	0	1	1	0
	-23	1	0	0	1	0	1	1	1	-71	1	1	0	0	0	1	1	1
	-24	1	0	0	1	1	0	0	0	-72	1	1	0	0	1	0	0	0
	-25	1	0	0	1	1	0	0	1	-73	1	1	0	0	1	0	0	1
	-26	1	0	0	1	1	0	1	0	-74	1	1	0	0	1	0	1	0
	-27	1	0	0	1	1	0	1	1	-75	1	1	0	0	1	0	1	1
	-28	1	0	0	1	1	1	0	0	-76	1	1	0	0	1	1	0	0
	-29	1	0	0	1	1	1	0	1	-77	1	1	0	0	1	1	0	1
	-30	1	0	0	1	1	1	1	0	-78	1	1	0	0	1	1	1	0
	-31	1	0	0	1	1	1	1	1	-79	1	1	0	0	1	1	1	1
	-32	1	0	1	0	0	0	0	0	-∞	1	1	1	1	1	1	1	1

: Initial condition

#### 3. Application Circuit



Unit R : [Ω] C : [F]

### Notes on wiring

- ① Please connect the decoupling capacitor of the power supply in the shortest possible distance to GND.
- ② GND lines should be one-point connected.
- ③ Wiring pattern of Digital should be away from that of Analog unit and cross-talk should not be acceptable.
- § SCL and SDA lines of I<sup>2</sup>C BUS should not be parallel if possible.

  The lines should be shielded, if they are adjacent to each other.
- ⑤ Analog input lines should not be parallel if possible. The lines should be shielded, if they are adjacent to each other.

#### **Power Dissipation**

About the thermal design of the IC

Characteristics of an IC have a great deal to do with the temperature at which it is used, and exceeding absolute maximum ratings may degrade and destroy the device. Careful consideration must be given to the heat of the IC from the two standpoints of immediate damage and long-term reliability of operation.

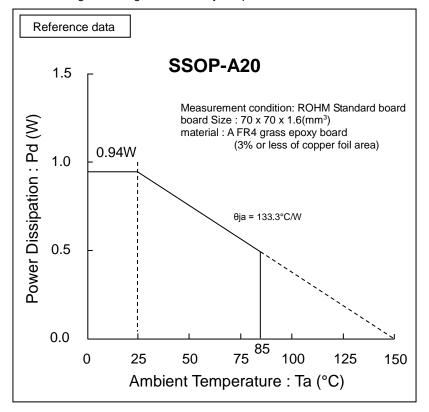


Figure 19. Temperature Derating Curve

(Note) Values are actual measurements and are not guaranteed.

Power dissipation values vary according to the board on which the IC is mounted

I/O Equivalent Circuits

Terminal	T		
	Terminal	Equivalent Circuit	Terminal Description
A1	4.25		A terminal for signal input. The input impedance is 100kΩ(typ).
A2			The input impedance is rooksz(typ).
B1		│	
B2		0	
0.4	4.05		A terminal for signal input.
	4.25	0 +	The input impedance is $250k\Omega(typ)$ .
C2		\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	
		250k	
DP1	4.25	vcc	Input terminal available to Single/Differential mode.
DN			The input impedance is $250k\Omega(typ)$ .
DP2			(3/2)
		GND	
MUTE	_	VCC By By By 1.65V	A terminal for external compulsory mute. If terminal voltage is High level, the mute is OFF. And if the terminal voltage is Low level, the mute is on.
		GND	
OUTS	4.25	VCC	A terminal for fader and Subwoofer output.
		<del>                                    </del>	
OUIFI		<del>-    </del>	
		GND BY	
	Name A1 A2 B1 B2  C1 C2  DP1 DN DP2  MUTE	Name         voltage           A1         4.25           A2         B1           B1         B2    C1 C2  4.25 C2  MUTE  OUTS OUTR2 OUTR1 OUTF2	Name voltage Equivalent Circuit  A1

Values in the pin explanation and input/output equivalent circuit are reference values only and are not guaranteed.

I/O Equivalent Circuits - continued

D Equivalent Circuits - continued								
Terminal No.	Terminal Name	Terminal voltage	Equivalent Circuit	Terminal Description				
17	VCC	8.5		Power supply terminal.				
18	SCL	-	VCC O 1.65V	A terminal for clock input of I <sup>2</sup> C BUS communication.				
19	SDA	-	VCC O J J J J J J J J J J J J J J J J J J	A terminal for data input of I <sup>2</sup> C BUS communication.				
20	GND	0		Ground terminal.				
1	FIL	4.25	VCC	1/2 VCC terminal.  Voltage for reference bias of analog signal system. The simple precharge circuit and simple discharge circuit for an external capacitor are built in.				

Values in the pin explanation and input/output equivalent circuit are reference values only and are not guaranteed.

#### **Operational Notes**

#### 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

#### 2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

#### 3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

#### 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

#### 5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

#### 6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

#### 7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

#### 8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

#### 9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

#### 10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

#### 11. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

#### **Operational Notes - continued**

#### 12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

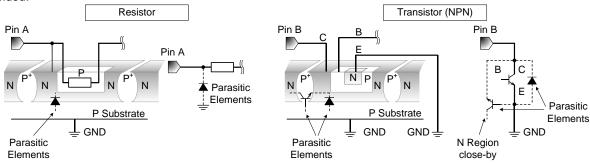
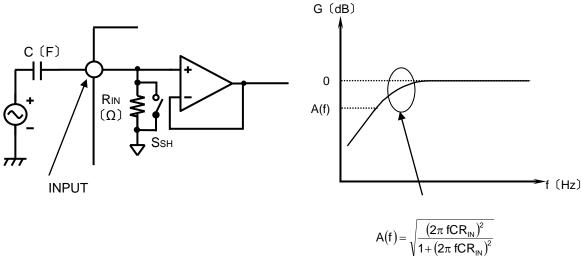


Figure 20. Example of monolithic IC structure

#### 13. About a Signal Input Part

#### (a) About Input Coupling Capacitor Constant Value

In the input signal terminal, please decide the constant value of the input coupling capacitor C(F) that would be sufficient to form an RC characterized HPF with input impedance  $R_{IN}(\Omega)$  inside the IC.



### (b) About the Input Selector SHORT

SHORT mode is the command which makes switch  $S_{SH}$  =ON of input selector part so that the input impedance  $R_{IN}$  of all terminals becomes small. Switch  $S_{SH}$  is OFF when SHORT command is not selected. The constant time brought about by the small resistance inside and the capacitor outside the LSI becomes

small when this command is used. The charge time of the capacitor becomes short. Since SHORT mode turns ON the switch of  $S_{SH}$  and makes it low impedance, please use it at no signal condition.

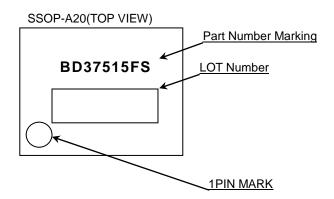
#### 14. About Mute Terminal(Pin 11) when power supply is OFF

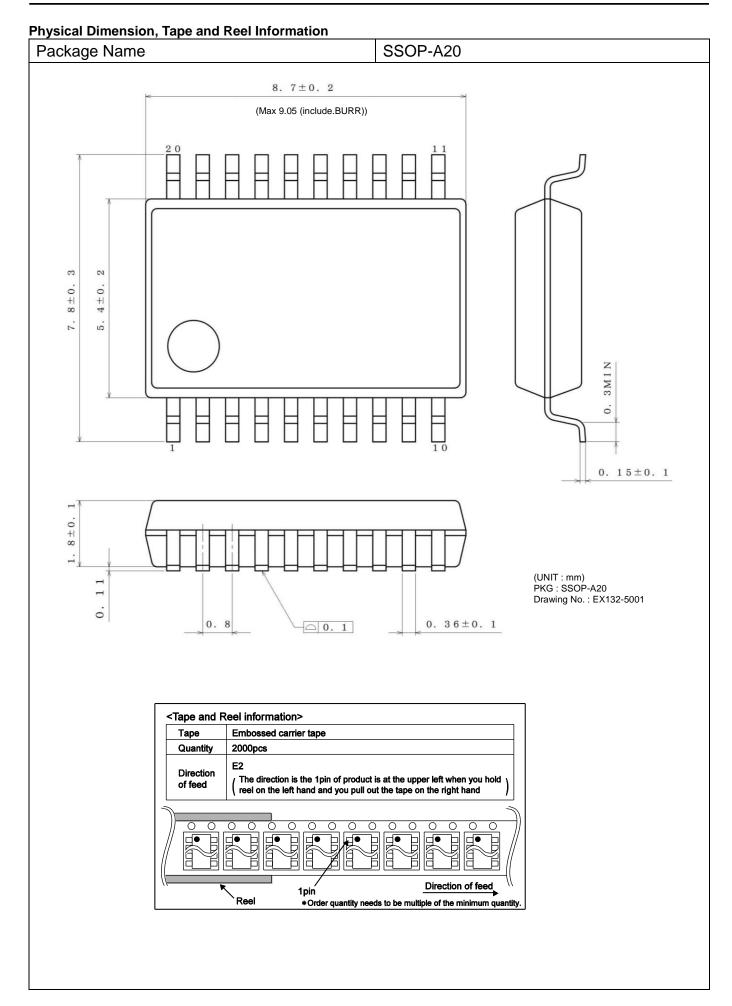
There should be no applied voltage across the Mute terminal (Pin 11) when power-supply is OFF. A resistor (about  $2.2k\Omega$ ) should be connected in series to Mute terminal in case a voltage is supplied to Mute terminal. (Please refer Application Circuit Diagram.)

### **Ordering Information**



### **Marking Diagram**





## **Revision History**

Date	Revision	Changes	
16.Dec.2015	001	New Release	

## **Notice**

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(Note1) Medical Equipment Classification of the Specific Applications

JAPAN	USA	EU	CHINA
CLASSⅢ	CL ACCIII	CLASSIIb	П 20
CLASSIV	CLASSⅢ	CLASSⅢ	- CLASSIII

- 2. ROHM designs and manufactures its Products subject to strict quality control system. However, semiconductor products can fail or malfunction at a certain rate. Please be sure to implement, at your own responsibilities, adequate safety measures including but not limited to fail-safe design against the physical injury, damage to any property, which a failure or malfunction of our Products may cause. The following are examples of safety measures:
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  - [b] Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
  - [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl<sub>2</sub>, H<sub>2</sub>S, NH<sub>3</sub>, SO<sub>2</sub>, and NO<sub>2</sub>
  - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
  - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
  - [f] Sealing or coating our Products with resin or other coating materials
  - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
  - [h] Use of the Products in places subject to dew condensation
- 4. The Products are not subject to radiation-proof design.
- 5. Please verify and confirm characteristics of the final or mounted products in using the Products.
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- 7. De-rate Power Dissipation depending on ambient temperature. When used in sealed area, confirm that it is the use in the range that does not exceed the maximum junction temperature.
- 8. Confirm that operation temperature is within the specified range described in the product specification.
- 9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

#### Precaution for Mounting / Circuit board design

- 1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- 2. In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

#### **Precautions Regarding Application Examples and External Circuits**

- 1. If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
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#### **Precaution for Electrostatic**

This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of lonizer, friction prevention and temperature / humidity control).

### **Precaution for Storage / Transportation**

- 1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
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  - [b] the temperature or humidity exceeds those recommended by ROHM
  - [c] the Products are exposed to direct sunshine or condensation
  - [d] the Products are exposed to high Electrostatic
- Even under ROHM recommended storage condition, solderability of products out of recommended storage time period
  may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is
  exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- 4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

#### **Precaution for Product Label**

QR code printed on ROHM Products label is for ROHM's internal use only.

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