

# 1700 V SCALE-iDriver

Up to 8 A Single Channel IGBT/MOSFET Gate Driver  
Providing Basic Galvanic Isolation for 1700 V IGBT  
and MOSFET

## Product Highlights

### Highly Integrated, Compact Footprint

- Split outputs providing up to 8 A peak drive current
- Integrated FluxLink™ technology providing galvanic isolation between primary-side and secondary-side
- Rail-to-rail stabilized output voltage
- Unipolar supply voltage for secondary-side
- Suitable for 1700 V IGBT and MOSFET switches
- Up to 75 kHz switching frequency
- Low propagation delay time 260 ns
- Propagation delay jitter  $\pm 5$  ns
- $-40$  °C to  $125$  °C operating ambient temperature
- High common-mode transient immunity
- eSOP package with 9.5 mm creepage and clearance distances

### Advanced Protection / Safety Features

- Undervoltage lock-out protection for primary and secondary-side (UVLO) and fault feedback
- Short-circuit protection using  $V_{CE\ SAT}$  monitoring and fault feedback
- Advanced Soft Shut Down (ASSD)

### Full Safety and Regulatory Compliance

- 100% production partial discharge test
- 100% production HIPOT compliance testing at 6 kV RMS 1 s
- Basic insulation meets VDE 0884-10

### Green Package

- Halogen free and RoHS compliant

### Applications

- General purpose and servo drives
- UPS, solar, welding inverters and power supplies

## Description

The SID1183K is a single channel IGBT and MOSFET driver in an eSOP package. Galvanic isolation is provided by Power Integrations' innovative solid insulator FluxLink technology. The up to 8 A peak output drive current enables the product to drive devices up to 600 A without requiring any additional active components. For gate drive requirements that exceed the stand-alone capability of the SID1183K, an external booster may be added. Stable positive and negative voltages for gate control are provided by one unipolar isolated voltage source.

Additional features are short-circuit protection (DESAT) with Advanced Soft Shut Down (ASSD), undervoltage lock-out (UVLO) for primary-side and secondary-side and rail-to-rail output with temperature and process compensated output impedance guarantee safe operation even in harsh conditions.

Controller (PWM and fault) signals are compatible with 5 V CMOS logic, which may also be adjusted to 15 V levels by using external resistor divider.

## Product Portfolio

| Product <sup>1</sup> | Peak Output Drive Current |
|----------------------|---------------------------|
| <b>SID1183K</b>      | 8 A                       |

Table 1. SCALE-iDriver 1700 V Portfolio.

Notes:

1. Package: eSOP-R16B.



Figure 2. eSOP-R16B Package.

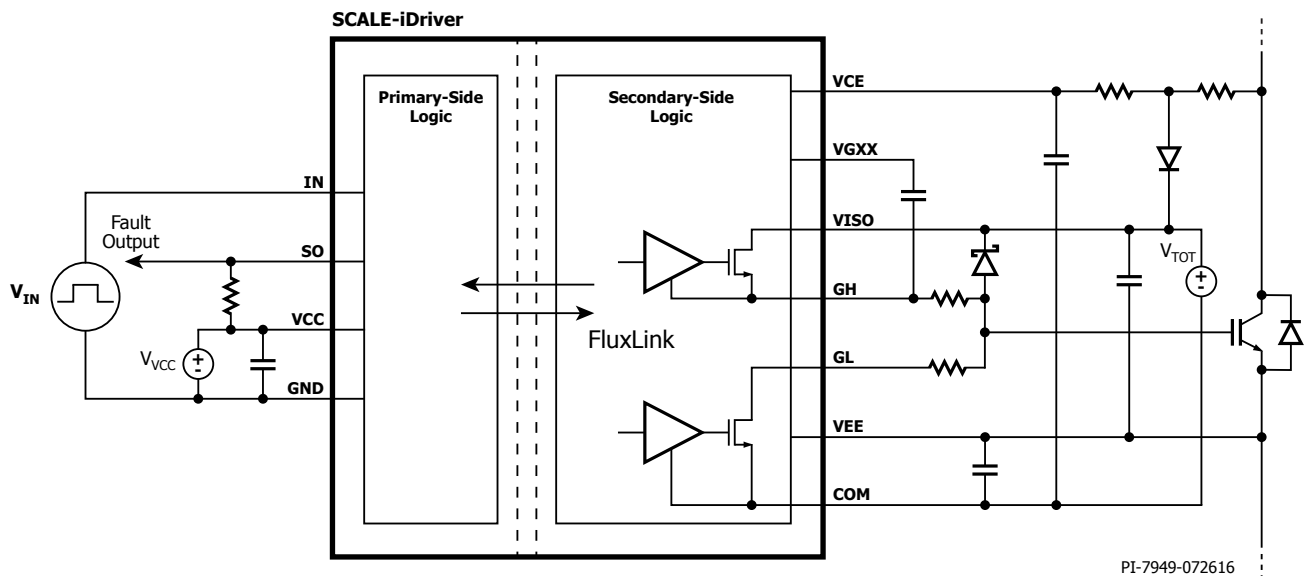
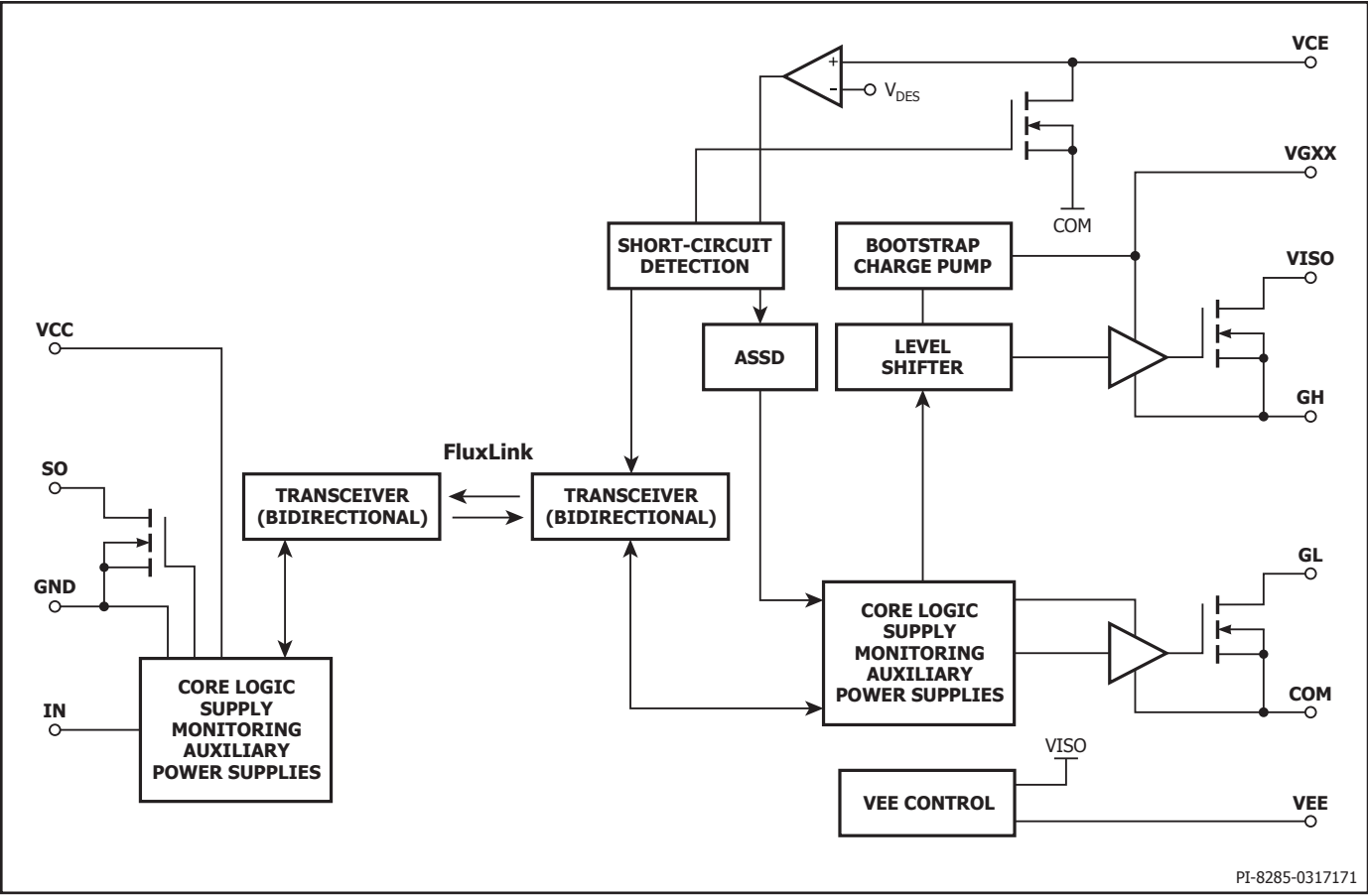


Figure 1. Typical Application Schematic.

PI-7949-072616



**Figure 3. Functional Block Diagram.**

## Pin Functional Description

### VCC Pin (Pin 1):

This pin is the primary-side supply voltage connection.

### GND Pin (Pin 3-6):

This pin is the connection for the primary-side ground potential. All primary-side voltages refer to this pin.

### IN Pin (Pin 7):

This pin is the input for the logic command signal.

### SO Pin (Pin 8):

This pin is the output for the logic fault signal (open drain).

### NC Pin (Pin 9):

This pin must be un-connected. Minimum PCB pad size for soldering is required.

### VEE Pin (Pin 10):

Common (IGBT emitter/MOSFET source) output supply voltage.

### VCE Pin (Pin 11):

This pin is the desaturation monitoring voltage input connection.

### VGXX Pin (Pin 12):

This pin is the bootstrap and charge pump supply voltage source.

### **GH Pin (Pin 13):**

This pin is the driver output – sourcing current (turn-on) connection.

### VISO Pin (Pin 14):

This pin is the input for the secondary-side positive supply voltage.

### COM Pin (Pin 15):

This pin provides the secondary-side reference potential.

### GL Pin (Pin 16):

This pin is the driver output – sinking current (turn-off).

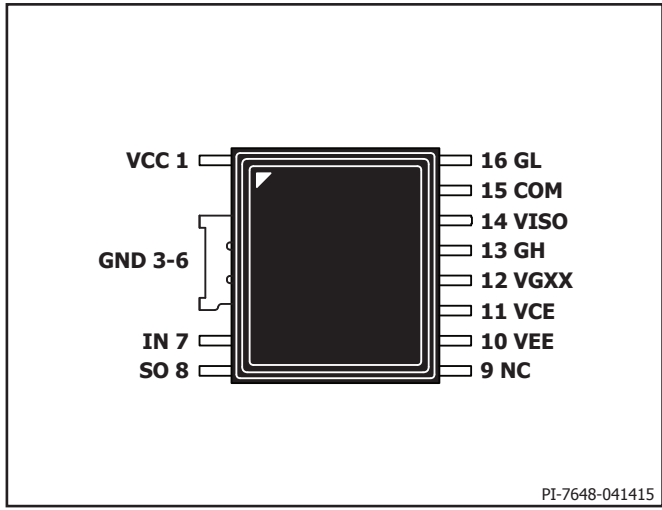


Figure 4. Pin Configuration.

## SCALE-iDriver Functional Description

The single channel SCALE-iDriver™ family drives IGBTs and MOSFETs or other semiconductor power switches with a blocking voltage of up to 1700 V and provides basic isolation between micro-controller and the power semiconductor switch. The status of the power semiconductor switch and SCALE-iDriver is monitored via the SO pin.

Command signals are transferred from the primary (IN) to secondary-side via FluxLink isolation technology. The GH pin supplies a positive gate voltage and charges the semiconductor gate during the turn-on process. The GL pin supplies the negative voltage and discharges the gate during the turn-off process.

Short-circuit protection is implemented using a desaturation detection technique monitored via the VCE pin. After the SCALE-iDriver detects a short-circuit, the semiconductor turn-off process is implemented using an Advanced Soft Shut Down (ASSD) technique.

### Power Supplies

The SID1183K requires two power supplies. One is the primary-side ( $V_{VCC}$ ) which powers the primary-side logic and communication with the secondary (insulated) side. One supply voltage is required for the secondary-side,  $V_{TOT}$  is applied between the VISO pin and the COM pin.  $V_{TOT}$  should be insulated from the primary-side and should provide at least the same insulation capabilities as the SCALE-iDriver.  $V_{TOT}$  should have a low capacitive coupling to the primary or any other secondary-side. The positive gate-emitter voltage is provided by  $V_{VISO}$  which is internally generated and stabilized to 15 V (typically) with respect to VEE. The negative gate-emitter voltage is provided by  $V_{VEE}$  with respect to COM. Due to the limited current sourcing capabilities of the VEE pin, any additional load needs to be applied between the VISO and COM pins. No additional load between VISO and VEE pins or between VEE and COM pins is allowed.

### Input and Fault Logic (Primary-Side)

The input (IN) and output (SO) logic is designed to work directly with micro-controllers using 5 V CMOS logic. If the physical distance between the controller and the SCALE-iDriver is large or if a different logic level is required, the resistive divider in Figure 5 is recommended. This solution adjusts the logic level as necessary and will also improve the driver's noise immunity.

Gate driver commands are transferred from the IN pin to the GH and GL pins with a propagation delay  $t_{P(LH)}$  and  $t_{P(HL)}$ .

During normal operation, when there is no fault detected, the SO pin stays at high impedance (open). Any fault is reported by connecting the SO pin to GND. The SO pin stays low as long as the  $V_{VCC}$  voltage (primary-side) stays below  $UVLO_{VCC}$  and the propagation delay is negligible. If desaturation is detected (there is a short-circuit), or the supply voltages  $V_{VISO}$ ,  $V_{VEE}$  (secondary-side) drop below  $UVLO_{VISO}$ ,  $UVLO_{VEE}$  the SO status changes with a delay time  $t_{FAULT}$  and keeps status low for a time defined as  $t_{SO}$ . In case of a fault condition the driver applies the off-state (the GL pin is connected to COM). During the  $t_{SO}$  period, command signal transitions from the IN pin are ignored. A new turn-on command transition is required before the driver will enter the on-state.

The SO pin current is defined as  $I_{SO}$ ; voltage during low status is defined as  $V_{SO(FAULT)}$ .

### Output (Secondary-Side)

The gate of the power semiconductor switch should be connected to the SCALE-iDriver output via pins GH and GL, using suitable turn-on and turn-off gate resistors. Turn-on gate resistor  $R_{GON}$  needs to be connected to the GH pin and turn-off gate resistor  $R_{GOFF}$  to the GL pin. If both gate resistors have the same value, the GL and GH pins may be

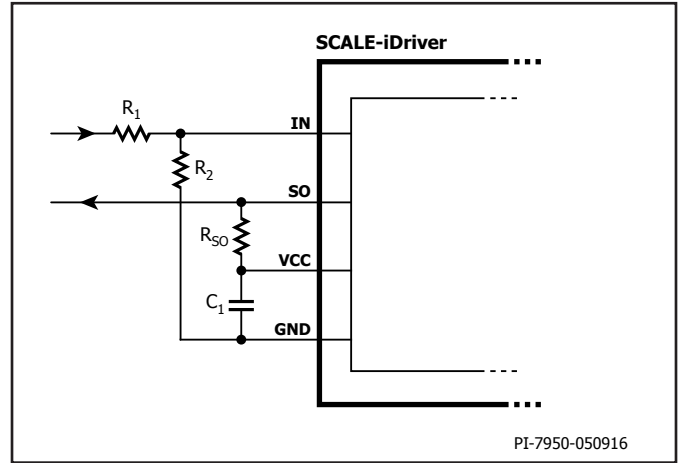


Figure 5. Increased Threshold Voltages  $V_{IN+LT}$  and  $V_{IN+HT}$  For  $R_1 = 3.3 \text{ k}\Omega$  and  $R_2 = 1 \text{ k}\Omega$  the IN Logic Level is 15 V.

connected together. Note: The SCALE-iDriver SID1183K data sheet defines the  $R_{GH}$  and  $R_{GL}$  values as total resistances connected to the respective pins GH and GL. Note that most power semiconductor data sheets specify an internal gate resistor  $R_{GINT}$  which is already integrated into the power semiconductor switch. In addition to  $R_{GINT}$  external resistor devices  $R_{GON}$  and  $R_{GOFF}$  are specified to setup the gate current levels to the application requirements. Consequently,  $R_{GH}$  is the sum of  $R_{GON}$  and  $R_{GINT}$  as shown in Figures 9 and 10. Careful consideration should be given to the power dissipation and peak current associated with the external gate resistors.

The GH pin output current source ( $I_{GH}$ ) of SID1183K is capable of handling (typically) up to 7.3 A during turn-on, and the GL pin output current source ( $I_{GL}$ ) is able to sink up to 8.0 A during turn-off at 25 °C. The SCALE-iDriver's internal resistances are described as  $R_{GHI}$  and  $R_{GLI}$  respectively. If the gate resistors for SCALE-iDriver family attempt to draw a higher peak current, the peak current will be internally limited to a safe value, see Figures 6 and 7. Figure 8 shows the peak current that can be achieved for a given supply voltage for same gate resistor values, load capacitance and layout design.

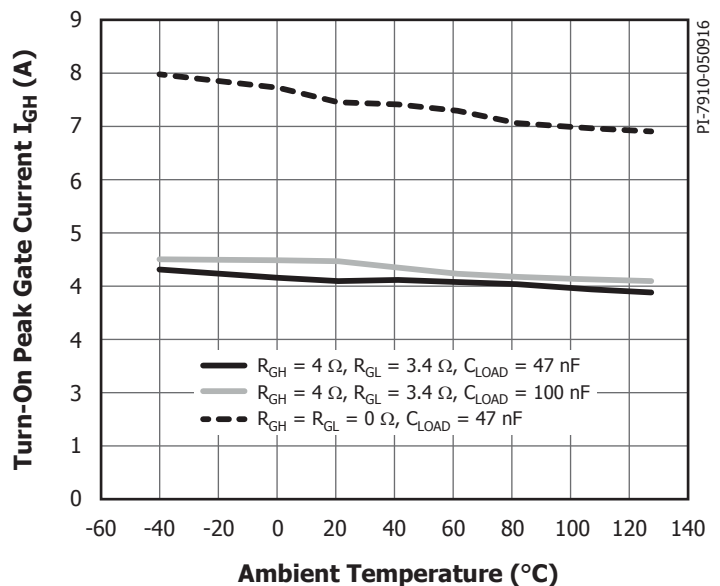


Figure 6. Turn-On Peak Output Current (Source) vs. Ambient Temperature. Conditions:  $V_{CC} = 5 \text{ V}$ ,  $V_{TOT} = 25 \text{ V}$ ,  $f_s = 20 \text{ kHz}$ , Duty Cycle = 50%.

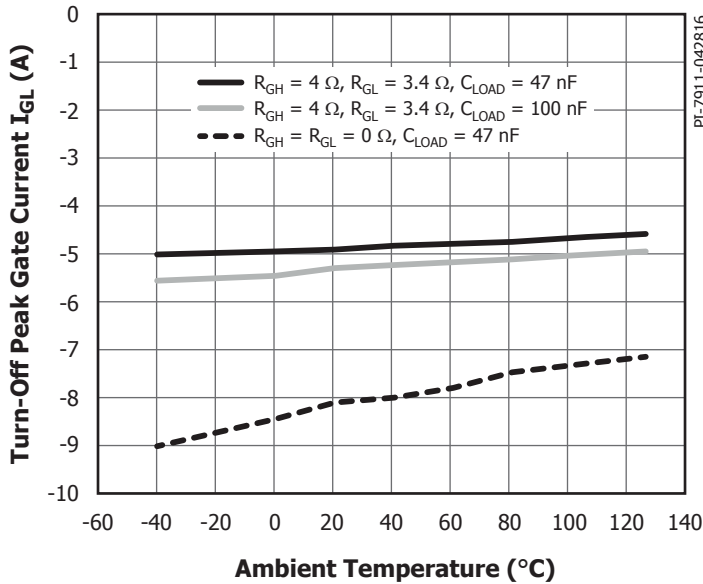


Figure 7. Turn-Off Peak Output Current (Sink) vs. Ambient Temperature. Conditions:  $V_{VCC} = 5\text{ V}$ ,  $V_{TOT} = 25\text{ V}$ ,  $f_s = 20\text{ kHz}$ , Duty Cycle = 50%.

### Short-Circuit Protection

The SCALE-iDriver uses the semiconductor desaturation effect to detect short-circuits and protects the device against damage by employing an Advanced Soft Shut Down (ASSD) technique. Desaturation can be detected using two different circuits, either with diode sense circuitry  $D_{VCE}$  (Figure 9) or with resistors  $R_{VCEX}$  (Figure 10). With the help of a well stabilized  $V_{VISO}$  and a Schottky diode ( $D_{STO}$ ) connected between semiconductor gate and VISO pin the short-circuit current value can be limited to a safe value.

During the off-state, the VCE pin is internally connected to the COM pin and  $C_{RES}$  is discharged (red curve in Figure 11 represents the potential of the VCE pin). When the power semiconductor switch receives a turn-on command, the collector-emitter voltage ( $V_{CE}$ ) decreases from the off-state level same as the DC-link voltage to a normally much lower on-state level (see blue curve in Figure 11) and  $C_{RES}$  begins to be charged up to the  $V_{CE}$  saturation level ( $V_{CE\text{SAT}}$ ). The  $V_{CE}$  voltage during on-state is continuously observed and compared with a reference voltage  $V_{DES}$ . As soon as  $V_{CE} > V_{DES}$  the driver turns off the power semiconductor switch with a controlled collector current slope, limiting the  $V_{CE}$  overvoltage excursions to below the maximum collector-emitter voltage ( $V_{CES}$ ). Turn-on commands during this time and during  $t_{SO}$  are ignored, and the SO pin is connected to GND.

The response time  $t_{RES}$  is the  $C_{RES}$  charging time and describes the delay between  $V_{CE}$  asserting and the voltage on the VCE pin rising (see Figure 11). Response time should be long enough to avoid false tripping during semiconductor turn-on and is adjustable via  $R_{RES}$  and  $C_{RES}$  (Figure 9) or  $R_{VCE}$  and  $C_{RES}$  (Figure 10) values. It should not be longer than the period allowed by the semiconductor manufacturer.

Note: The response time for short-circuit protection using a resistor network depends also on the actual DC-link voltage.

The implementation according to Figure 10 is preferred as it avoids unintended tripping of the gate driver during zero-crossing of the load current, which depending on actual application conditions may lead to a transient  $V_{CESAT}$  increase. The circuitry according to Figure 10 provides an inherent filter, which prevents a false short-circuit detection. The implementation according to Figure 9, however, does not provide this filtering and may lead to unintended tripping events.

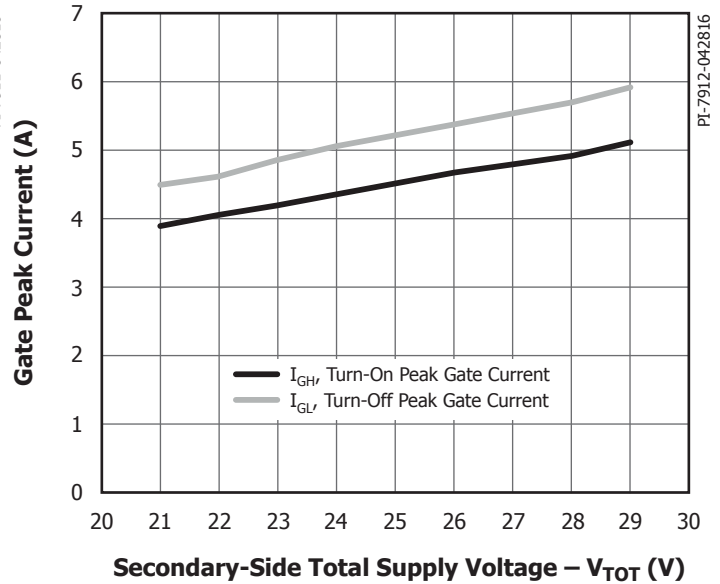


Figure 8. Turn-On and Turn-Off Peak Output Current vs. Secondary-Side Total Supply Voltage ( $V_{TOT}$ ). Conditions:  $V_{VCC} = 5\text{ V}$ ,  $T_J = 25\text{ °C}$ ,  $R_{GH} = 4\text{ Ω}$ ,  $R_{GL} = 3.4\text{ Ω}$ ,  $C_{LOAD} = 100\text{ nF}$ ,  $f_s = 1\text{ kHz}$ , Duty Cycle = 50%.

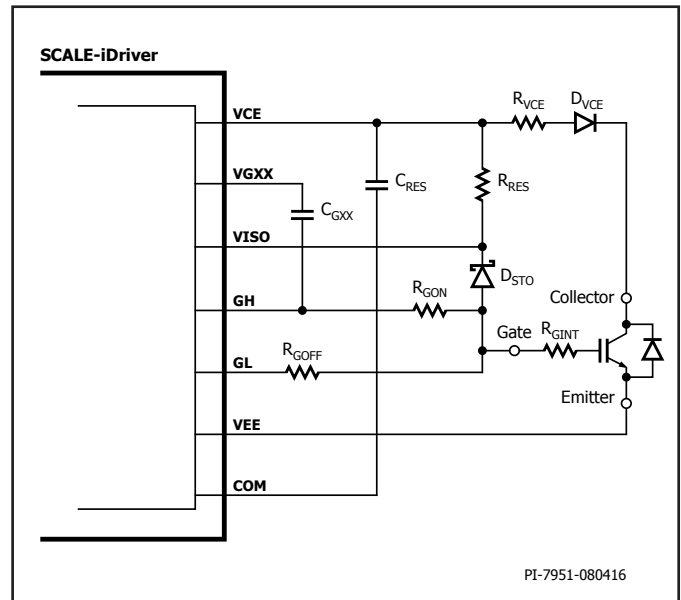


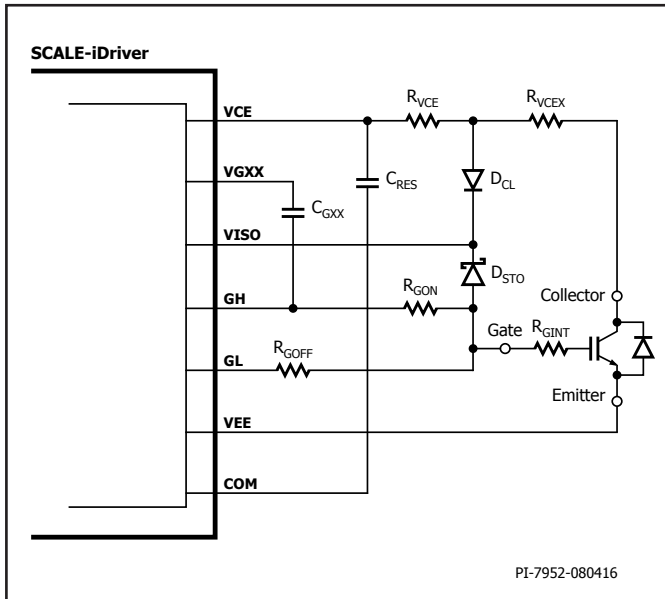
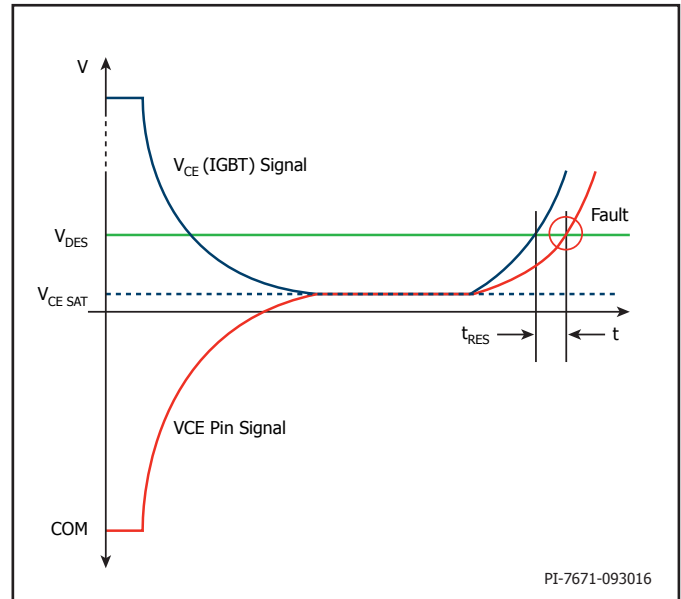
Figure 9. Short-Circuit Protection Using Diode  $D_{VCE}$ .

### Safe Power-Up and Power-Down

During power-up and power-down the IN pin should stay at logic low. In order to avoid these effects, it is recommended that the IN pin is kept at logic low during power-up and power-down. Any supply voltage related to VCC, VISO, VEE and VGXX pins should be stabilized using ceramic capacitors  $C_1$ ,  $C_{S1X}$ ,  $C_{S2X}$ ,  $C_{GXX}$  respectively as shown in Figures 13 and 14. After supply voltages reach their nominal values, the driver will begin to function after a time delay  $t_{START}$ .

### Short-Pulse Operation

If command signals applied to the IN pin are shorter than the minimum specified by  $t_{GE(MIN)}$ , then SID1183K output signals, GH and GL pins, will extend to value  $t_{GE(MIN)}$ . The duration of pulses longer than  $t_{GE(MIN)}$  will not be changed.


Figure 10. Short-Circuit Protection Using a Resistor Network  $R_{VCEX}$ 

Figure 11. Short-Circuit Protection Using Resistor Network  $R_{VCEX}$ 

### Advanced Soft Shut Down (ASSD)

This function is activated after a short-circuit is detected. It protects the power semiconductor switch against destruction by ending the turn-on state and limiting the current slope in order to keep momentary  $V_{CE}$  overvoltages below  $V_{CES}$ . This function is particularly suited to IGBT applications. Figure 12 shows how the ASSD function operates. The  $V_{CE}$  desaturation is visible during time period P1 (yellow line). During this time, the gate-emitter voltage (green line) is kept very stable. Collector current (pink line) is also well stabilized and limited to a safe value. At the end of period P1,  $V_{GE}$  is reduced until the

beginning of  $t_{FSSD1}$ . Due to collector current decrease a small  $V_{CE}$  overvoltage is seen. During  $t_{FSSD1}$   $V_{GE}$  is further reduced and the gate of the power semiconductor switch is further discharged. During  $t_{FSSD2}$  additional small  $V_{CE}$  overvoltage events may occur. Once  $V_{GE}$  drops below the gate threshold of the IGBT, the collector current has decayed almost to zero and the remaining gate charge is removed – ending the short-circuit event. The whole short-circuit current detection and safe switch-off is shorter than 10  $\mu s$  (7.1  $\mu s$  from 10%-to-10% of the collector current in this example).

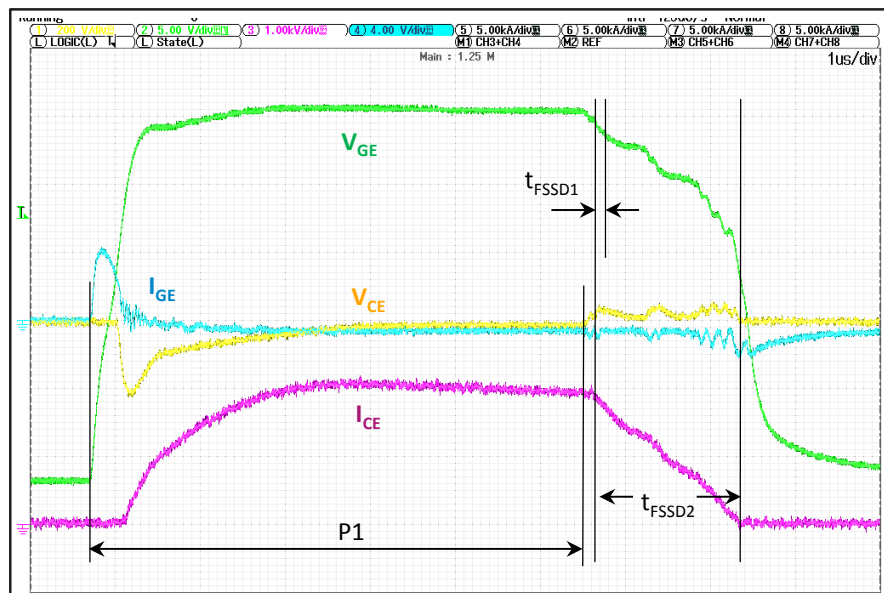


Figure 12. Advanced Soft Shut Down Function.

## Application Examples and Components Selection

Figures 13 and 14 show the schematic and typical components used for a SID1183K design. In both cases the primary-side supply voltage ( $V_{VCC}$ ) is connected between VCC and GND pins and supported through a supply bypass ceramic capacitor  $C_1$  (4.7  $\mu$ F typically). If the command signal voltage level is higher than the rated IN pin voltage (in this case 15 V) a resistive voltage divider should be used. Additional capacitor  $C_F$  and Schmitt trigger  $IC_1$  can be used to provide input signal filtering. The SO output has 5 V logic and the  $R_{SO}$  is selected so that it does not exceed absolute maximum rated  $I_{SO}$  current.

The secondary-side isolated power supply ( $V_{TOT}$ ) is connected between VISO and COM. The positive voltage rail ( $V_{VISO}$ ) is supported through 4.7  $\mu$ F ceramic capacitors  $C_{S21}$  and  $C_{S22}$  connected in parallel. The negative voltage rail ( $V_{VEE}$ ) is similarly supported through capacitors

$C_{S11}$  and  $C_{S12}$ . The gate charge will vary according to the type of power semiconductor switch that is being driven. Typically,  $C_{S11} + C_{S12}$  should be at least 3  $\mu$ F multiplied by the total gate charge of the power semiconductor switch ( $Q_{GATE}$ ) divided by 1  $\mu$ C. A 10 nF capacitor  $C_{GXX}$  is connected between the GH and VGXX pins.

The gate of the power semiconductor switch is connected through resistor  $R_{GON}$  to the GH pin and by  $R_{GOFF}$  to the GL pin. If the value of  $R_{GON}$  is the same as  $R_{GOFF}$  the GH pin can be connected to the GL pin and a common gate resistor can be connected to the gate. In each case, proper consideration needs to be given to the power dissipation and temperature performance of the gate resistors.

To ensure gate voltage stabilization and collector current limitation during a short-circuit, the gate is connected to the VISO pin through a Schottky diode  $D_{STO}$  (for example PMEG4010).

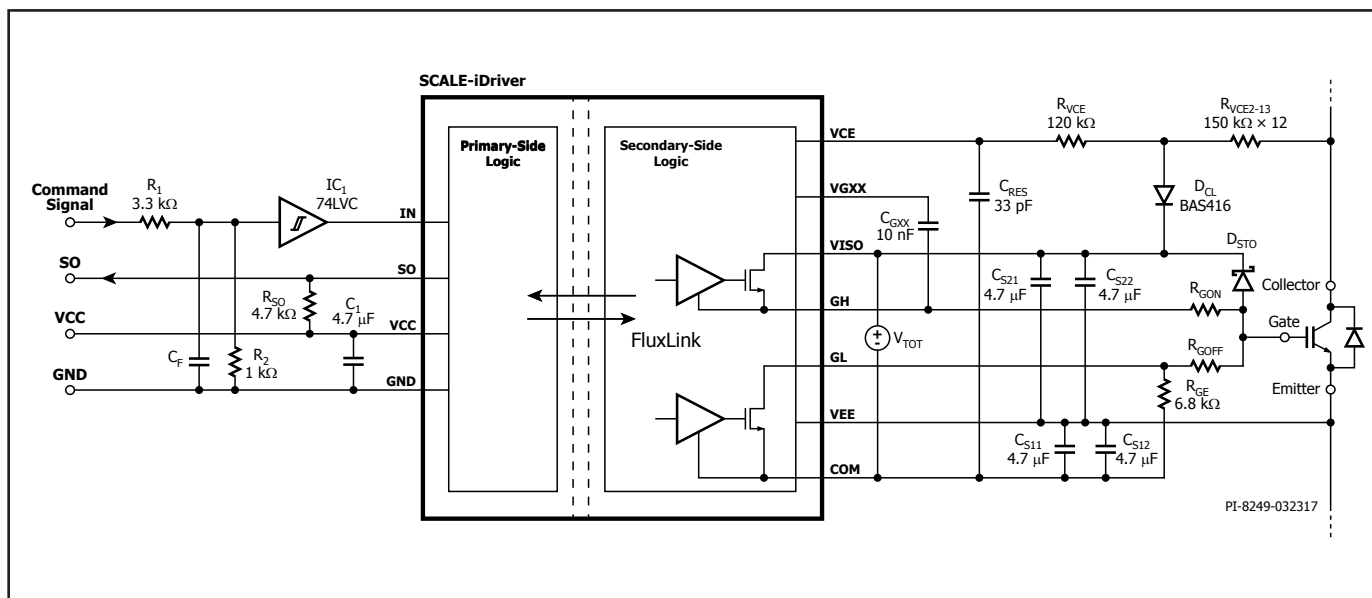


Figure 13. SCALE-iDriver Application Example Using a Resistor Network for Desaturation Detection.

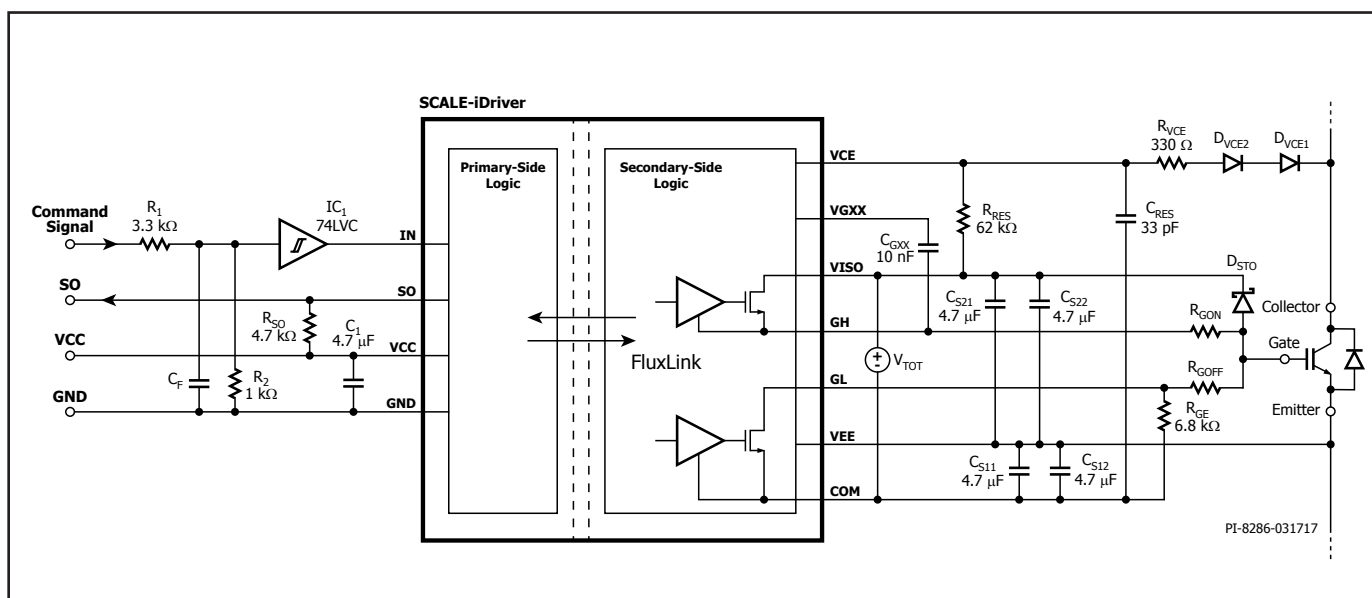


Figure 14. SCALE-iDriver Application Example Using Diodes for Desaturation Detection.



To avoid parasitic power-switch-conduction during system power-on, the gate is connected to COM through 6.8 kΩ resistor.

Figure 13 shows how switch desaturation can be measured using resistors  $R_{VCE2} - R_{VCE13}$ . In this example all the resistors have a value of 150 kΩ and 1206 size. The total resistance is 1.8 MΩ. The resistors should be chosen to limit current to between 0.6 mA to 0.8 mA at maximum DC-link voltage. The sum of  $R_{VCE2} - R_{VCE13}$  should be typically 1.8 MΩ for 1700 V semiconductors. In each case the resistor string must provide sufficient creepage and clearance distances between collector of the semiconductor and SCALE-iDriver. The low leakage diode  $D_{CL}$  keeps the short-circuit duration constant over a wide DC-link voltage range.

Response time is set up through  $R_{VCE}$  and  $C_{RES}$  (typically 120 kΩ and 33 pF respectively for 1700 V semiconductors). If short-circuit detection proves to be too sensitive, the  $C_{RES}$  value can be increased. The maximum short-circuit duration must be limited to the maximum value given in the semiconductor data sheet.

Figure 14 illustrates how diodes  $D_{VCE1}$  and  $D_{VCE2}$  may be used to measure switch desaturation. For insulation, two or more diodes in SMD packages are used (STTH212U for example depending on actual application conditions).  $R_{RES}$  connected to VISO guarantees current flow through the diodes when the semiconductor is in the on-state. When the switch desaturates,  $C_{RES}$  starts to be charged through  $R_{RES}$ . In this configuration the response time is controlled by  $R_{RES}$  and  $C_{RES}$ . In this application example  $C_{RES} = 33$  pF and  $R_{RES} = 62$  kΩ; if desaturation is too sensitive or the short-circuit duration too long, both  $C_{RES}$  and  $R_{RES}$  can be adjusted.

Figure 15 shows the recommended PCB layout and corresponds to the schematic in Figure 13. The PCB is a two layer design. It is important to ensure that PCB traces do not cover the area below the desaturation resistors or diodes  $D_{VCE1}$  and  $D_{VCE2}$ . This is a critical design requirement to avoid coupling capacitance with the SCALE-iDriver's VCE pin and isolation issues within the PCB.

Gate resistors are located physically close to the power semiconductor switch. As these components can get hot, it is recommended that they are placed away from the SCALE-iDriver.

## Power Dissipation and IC Junction Temperature Estimation

First calculation in designing the power semiconductor switch gate driver stage is to calculate the required gate power -  $P_{DRV}$ . The power is calculated based on equation 1:

$$P_{DRV} = Q_{GATE} \times f_s \times V_{TOT} \quad (1)$$

where,

$Q_{GATE}$  – Controlled power semiconductor switch gate charge (derived for the particular gate potential range defined by  $V_{TOT}$ ). See semiconductor manufacturer data sheet.

$f_s$  – Switching frequency which is same as applied to the IN pin of SCALE-iDriver.

$V_{TOT}$  – SCALE-iDriver secondary-side supply voltage.

In addition to  $P_{DRV}$ ,  $P_p$  (primary-side IC power dissipation) and  $P_{SNL}$  (secondary-side IC power dissipation without capacitive load) must be considered. Both are ambient temperature and switching frequency dependent (see typical performance characteristics).

$$P_p = V_{VCC} \times I_{VCC} \quad (2)$$

$$P_{SNL} = V_{TOT} \times I_{VISO} \quad (3)$$

During IC operation, the  $P_{DRV}$  power is shared between turn-on ( $R_{GH}$ ), turn-off ( $R_{GL}$ ) external gate resistors and internal driver resistances  $R_{GHI}$  and  $R_{GLI}$ . For junction temperature estimation purposes, the dissipated power under load ( $P_{OL}$ ) inside the IC can be calculated accordingly to equation 4:

$$P_{OL} = 0.5 \times Q_{GATE} \times f_s \times V_{TOT} \times \left( \frac{R_{GHI}}{R_{GHI} + R_{GH}} + \frac{R_{GLI}}{R_{GLI} + R_{GL}} \right) \quad (4)$$

$R_{GH}$  and  $R_{GL}$  represent sum of external ( $R_{GON}$ ,  $R_{GOFF}$ ) and power semiconductor internal gate resistance ( $R_{GINT}$ ):

$$R_{GH} = R_{GON} + R_{GINT}$$

$$R_{GL} = R_{GOFF} + R_{GINT}$$

Total IC power dissipation ( $P_{DIS}$ ) is estimated as sum of equations 2, 3 and 4:

$$P_{DIS} = P_p + P_{SNL} + P_{OL} \quad (5)$$

The operating junction temperature ( $T_j$ ) for given ambient temperature ( $T_A$ ) can be estimated according to equation 6:

$$T_j = \theta_{JA} \times P_{DIS} + T_A \quad (6)$$

## Example

An example is given below,

$f_s = 20$  kHz,  $T_A = 85$  °C,  $V_{TOT} = 25$  V,  $V_{VCC} = 5$  V.

$Q_{GATE} = 1.5$  μC (the gate charge value here should correspond to selected  $V_{TOT}$ ),  $R_{GINT} = 2.5$  Ω,  $R_{GON} = R_{GOFF} = 1.8$  Ω.

$P_{DRV} = 1.5$  μC  $\times$  20 kHz  $\times$  25 V = 0.75 W, according to equation 1.

$P_p = 5$  V  $\times$  13.5 mA = 68 mW, according to equation 2 (see Figure 18).

$P_{SNL} = 25$  V  $\times$  7.7 mA = 193 mW, according to equation 3 (see Figure 20).

The dissipated power under load is:

$$P_{OL} = 0.5 \times 1.5 \mu C \times 20 \text{ kHz} \times 25 \text{ V} \times \left( \frac{1.2 \Omega}{1.2 \Omega + 4.3 \Omega} + \frac{1.1 \Omega}{1.1 \Omega + 4.3 \Omega} \right) \cong 176 \text{ mW},$$

according to equation 4.

$R_{GHI} = 1.2$  Ω as maximum data sheet value.

$R_{GLI} = 1.1$  Ω as maximum data sheet value.

$R_{GH} = R_{GL} = 1.8$  Ω + 2.5 Ω = 4.3 Ω.

$P_{DIS} = 68$  mW + 193 mW + 176 mW = 437 mW according to equation 5.

$T_j = 67$  °C/W  $\times$  437 mW + 85 °C = 113 °C according to equation 6.

Estimated junction temperature for this design would be approximately 113 °C and is lower than the recommended maximum value. As the internal IC resistor values are maximum values, it is understood that the example represents worst-case conditions.

Table 2 describes the recommended capacitor and resistor characteristics and layout requirements to achieve optimum performances of SCALE-iDriver.

| Pin  | Return to Pin | Recommended Value | Symbol                                       | Notes   |
|------|---------------|-------------------|--|---|
| VCC  | GND           | 4.7 $\mu$ F       | $C_1$  | VCC blocking capacitor must be placed close to IC. Enlarged loop could result in inadequate VCC supply voltage during operation.  |
| VISO | VEE           | 4.7 $\mu$ F       | $C_{S21}/C_{S22}$                            | 25V X7R type is recommended. Example part number could be Murata 25 V part #GRM31CR71E475KA88. This capacitor needs to be close to IC pins.   |
| VEE  | COM           | 4.7 $\mu$ F       | $C_{S11}/C_{S12}$                            | 25 V X7R type is recommended. Example part number could be Murata 25 V part #GRM31CR71E-475KA88. This capacitor needs to be close to IC pins.   |
| VGXX | GH            | 10 nF             | $C_{GXX}$                                    | To avoid misoperation, this pin should not be connected to anything else. This capacitor needs to be as close to IC pins as possible. 25 V X7R type is recommended. Example part number could be Yageo 25 V part#CC0603KRX7R9BB103.   |
| VCE  | COM           | 33 pF             | $C_{RES}$                                    | Select $C_{RES}$ to achieve needed desaturation protection response time. 50 V COG/NPO is recommended. A value of 33 pF is initially recommended. Example part number could be KEMET 50 V part C0603C330J5GACTU. Any net and any other layer should provide sufficient distance to components $C_{RES}$ in order to avoid parasitic effects (capacitance) |
| VCE  |               |                   | $R_{VCE}, D_{VCE}, C_{RES}, R_{RES}, D_{CL}$ | Select $R_{VCE}$ or $R_{RES}$ for the proper operation of the short-circuit protection. Any net and any other layer should provide sufficient distance to components $R_{VCE}, D_{VCE}, R_{RES}$ and $D_{CL}$ in order to avoid parasitic effects.  |

Table 2. PCB Layout and Component Guidelines.



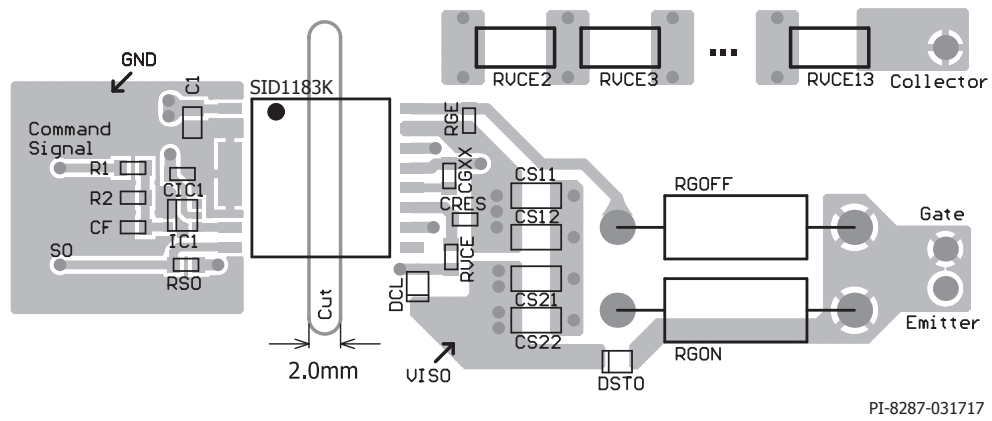


Figure 15a. Top View of Recommended PCB Layout. Corresponds to Schematic Shown in Figure 13.

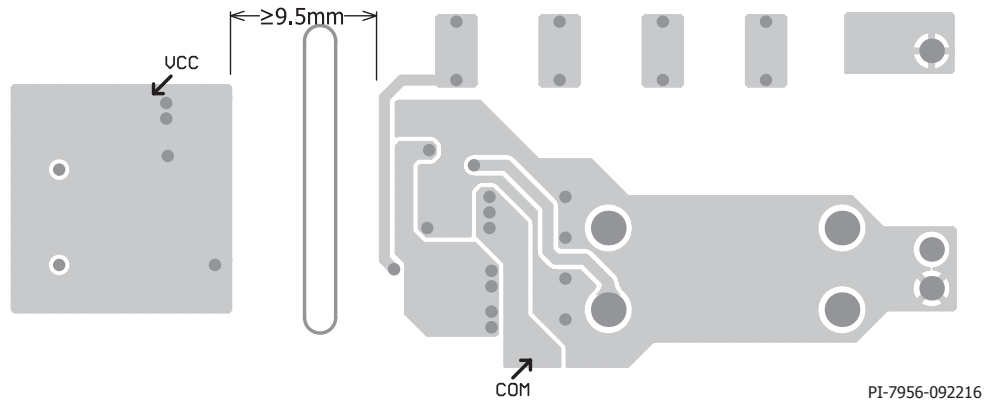


Figure 15b. Bottom View of Recommended PCB Layout. Corresponds to Schematic Shown in Figure 13.

| Parameter                                   | Symbol     | Conditions  | Min  | Max              | Units |
|---|------------|---|------|------------------|-------|
| <b>Absolute Maximum Ratings<sup>1</sup></b> |            |   |      |                  |       |
| Primary-Side Supply Voltage <sup>2</sup>    | $V_{VCC}$  | VCC to GND  | -0.5 | 6.5              | V     |
| Secondary-Side Total Supply Voltage         | $V_{TOT}$  | VISO to COM                                       | -0.5 | 30               | V     |
| Secondary-Side Positive Supply Voltage      | $V_{VISO}$ | VISO to VEE                                       | -0.5 | 17.5             | V     |
| Secondary-Side Negative Supply Voltage      | $V_{VEE}$  | VEE to COM  | -0.5 | 15               | V     |
| Logic Input Voltage (command signal)        | $V_{IN}$   | IN to GND   | -0.5 | $V_{VCC} + 0.5$  | V     |
| Logic Output Voltage (fault signal)         | $V_{SO}$   | SO to GND   | -0.5 | $V_{VCC} + 0.5$  | V     |
| Logic Output Current (fault signal)         | $I_{SO}$   | Positive Current Flowing into the Pin             |      | 10               | mA    |
| VCE Pin Voltage                             | $V_{VCE}$  | VCE - COM   | -0.5 | $V_{TOT} + 0.5$  | V     |
| Switching Frequency                         | $f_s$      |   |      | 75               | kHz   |
| Storage Temperature                         | $T_s$      |   | -65  | 150              | °C    |
| Operating Junction Temperature              | $T_j$      |   | -40  | 150 <sup>3</sup> | °C    |
| Operating Ambient Temperature               | $T_A$      |   | -40  | 125              | °C    |
| Operating Case Temperature                  | $T_C$      |   | -40  | 125              | °C    |
| Input Power Dissipation <sup>4</sup>        | $P_p$      | $V_{VCC} = 5\text{ V}$<br>$V_{TOT} = 28\text{ V}$ |      | 115              | mW    |
| Output Power Dissipation <sup>4</sup>       | $P_s$      |   |      | 1675             |       |
| Total IC Power Dissipation <sup>4</sup>     | $P_{DIS}$  |   |      | 1790             |       |

## NOTES:

1. Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device.
2. Defined as peak voltage measured directly on VCC pin.
3. Transmission of command signals could be affected by PCB layout parasitic inductances at junction temperatures higher than recommended.
4. Input Power Dissipation refers to equation 2. Output Power Dissipation is secondary-side IC power dissipation without capacitive load ( $P_{SNL}$ , equation 3) and dissipated power under load ( $P_{OL}$ , equation 4). Total IC power dissipation is sum of  $P_p$  and  $P_s$ .

**Thermal Resistance**

Thermal Resistance: eSOP-R16B Package:

$\theta_{JA}$  ..... 67 °C/W<sup>1</sup>  
 $\theta_{JC}$  ..... 34 °C/W<sup>2</sup>

Notes:

1. 2 oz. (610 g/m<sup>2</sup>) copper clad. Measured with layout shown in Figure 15.
2. The case temperature is measured at the plastic surface at the top of the package.

| Parameter   | Symbol            | Conditions<br>$T_J = -40\text{ }^{\circ}\text{C to } +125\text{ }^{\circ}\text{C}$<br>See Note 1 (Unless Otherwise Specified) | Min  | Typ   | Max  | Units              |
|---|-------------------|---|------|-------|------|--------------------|
| <b>Recommended Operation Conditions</b>                                       |                   |   |      |       |      |                    |
| Primary-Side Supply Voltage   | $V_{VCC}$         | VCC - GND   | 4.75 |       | 5.25 | V                  |
| Secondary-Side Total Supply Voltage   | $V_{TOT}$         | VISO - COM  | 22   |       | 28   | V                  |
| Logic Low Input Voltage   | $V_{IL}$          |   |      |       | 0.5  | V                  |
| Logic High Input Voltage  | $V_{IH}$          |   | 3.3  |       |      | V                  |
| Switching Frequency   | $f_S$             |   | 0    |       | 75   | kHz                |
| Operating IC Junction Temperature   | $T_J$             |   | -40  |       | 125  | $^{\circ}\text{C}$ |
| <b>Electrical Characteristics</b>   |                   |   |      |       |      |                    |
| Logic Low Input Threshold Voltage   | $V_{IN+LT}$       | $f_S = 0\text{ Hz}$   | 0.6  | 1.25  | 1.8  | V                  |
| Logic High Input Threshold Voltage  | $V_{IN+HT}$       | $f_S = 0\text{ Hz}$   | 1.7  | 2.2   | 3.05 | V                  |
| Logic Input Voltage Hysteresis  | $V_{IN+HS}$       | $f_S = 0\text{ Hz}$   | 0.1  |       |      | V                  |
| Input Bias Current  | $I_{IN}$          | $V_{IN} = 5\text{ V}$   | 56   | 113   | 165  | $\mu\text{A}$      |
|   |                   | $V_{IN} > 3\text{ V}$<br>See Note 12  |      | 106   |      |                    |
| Supply Current (Primary-Side)   | $I_{VCC}$         | $V_{IN} = 0\text{ V}$   |      |       | 17   | mA                 |
|   |                   | $V_{IN} = 5\text{ V}$   |      |       | 23   |                    |
|   |                   | $f_S = 20\text{ kHz}$   |      |       | 20   |                    |
|   |                   | $f_S = 75\text{ kHz}$   |      |       | 23   |                    |
| Supply Current (Secondary-Side)   | $I_{VISO}$        | $V_{IN} = 0\text{ V}$   |      |       | 8    | mA                 |
|   |                   | $V_{IN} = 5\text{ V}$   |      |       | 9    |                    |
|   |                   | $f_S = 20\text{ kHz}$   |      |       | 10   |                    |
|   |                   | $f_S = 75\text{ kHz}$   |      |       | 14   |                    |
| Power Supply Monitoring Threshold (Primary-Side)                              | $UVLO_{VCC}$      | Clear Fault   |      | 4.28  | 4.65 | V                  |
|   |                   | Set Fault   | 3.85 | 4.12  |      |                    |
|   |                   | Hysteresis, See Notes 3, 4  | 0.02 |       |      |                    |
| Power Supply Monitoring Threshold (Secondary-Side, Positive Rail $V_{VISO}$ ) | $UVLO_{VISO}$     | Clear Fault   |      | 12.85 | 13.5 | V                  |
|   |                   | Set Fault, Note 3   | 11.7 | 12.35 |      |                    |
|   |                   | Hysteresis  | 0.3  |       |      |                    |
| Power Supply Monitoring Blanking Time, $V_{VISO}$                             | $UVLO_{VISO(BL)}$ | Voltage Drop 13.5 V to 11.5 V<br>See Note 12  | 0.5  |       |      | $\mu\text{s}$      |
| Power Supply Monitoring Threshold (Secondary-Side, Negative Rail $V_{VEE}$ )  | $UVLO_{VEE}$      | Clear Fault, $V_{TOT} = 20\text{ V}$  |      | 5.15  | 5.5  | V                  |
|   |                   | Set Fault, $V_{TOT} = 20\text{ V}$  | 4.67 | 4.93  |      |                    |
|   |                   | Hysteresis  | 0.1  |       |      |                    |

| Parameter  | Symbol           | Conditions<br>$T_J = -40\text{ }^{\circ}\text{C}$ to $+125\text{ }^{\circ}\text{C}$<br>See Note 1 (Unless Otherwise Specified) | Min  | Typ   | Max   | Units         |
|--|------------------|--|------|-------|-------|---------------|
| <b>Electrical Characteristics (cont.)</b>                          |                  |  |      |       |       |               |
| <b>Power Supply Monitoring Blanking Time, <math>V_{VEE}</math></b> | $UVLO_{VEE(BL)}$ | Voltage Drop 5.5 V to 4.5 V<br>See Note 12   | 0.5  |       |       | $\mu\text{s}$ |
| <b>Secondary-Side Positive Supply Voltage Regulation</b>           | $V_{VISO(HS)}$   | $21\text{ V} \leq V_{TOT} \leq 30\text{ V}$ ,<br>$ i(V_{VEE})  \leq 1.5\text{ mA}$   | 14.4 | 15.07 | 15.75 | V             |
| <b>VEE Source Capability</b>                                       | $I_{VEE(SO)}$    | $V_{TOT} = 15\text{ V}$ , $V_{VEE}$ set to 0 V   | 0.1  |       |       | mA            |
|  |                  | $V_{TOT} = 25\text{ V}$ , $V_{VEE}$ set to 7.5 V<br>See Note 13  | 1.85 | 3.3   | 4.5   |               |
| <b>VEE Sink Capability</b>   | $I_{VEE(SI)}$    | $V_{TOT} = 25\text{ V}$ , $V_{VEE}$ set to 12.5 V  | 1.74 | 3.1   | 4.5   | mA            |
| <b>DESAT Detection Level</b>                                       | $V_{DES}$        | VCE-VEE, $V_{IN} = 5\text{ V}$   | 7.2  | 7.8   | 8.3   | V             |
| <b>DESAT Sink Current</b>  | $I_{DES}$        | $V_{VCE} = 10\text{ V}$ , $V_{IN} = 0\text{ V}$  | 15   | 28    | 50    | mA            |
| <b>DESAT Bias Current</b>  | $I_{DES(BS)}$    | $V_{VCE} - V_{VEE} = 4.5\text{ V}$ , $V_{IN} = 5\text{ V}$   | -0.5 |       | 3     | $\mu\text{A}$ |
| <b>VCE Pin Capacitance</b>   | $C_{VCE}$        | Between VCE and COM pins<br>See Note 12  |      | 12.5  |       | pF            |
| <b>Turn-On Propagation Delay</b>                                   | $t_{P(LH)}$      | $T_J = 25\text{ }^{\circ}\text{C}$<br>See Note 5   | 180  | 253   | 340   | ns            |
|  |                  | $T_J = 125\text{ }^{\circ}\text{C}$<br>See Note 5  | 210  | 278   | 364   |               |
| <b>Turn-Off Propagation Delay</b>                                  | $t_{P(HL)}$      | $T_J = 25\text{ }^{\circ}\text{C}$<br>See Note 6   | 200  | 262   | 330   | ns            |
|  |                  | $T_J = 125\text{ }^{\circ}\text{C}$<br>See Note 6  | 211  | 287   | 359   |               |
| <b>Minimum Turn-On and Off Pulses</b>                              | $t_{GE(MIN)}$    | See Note 12  |      |       | 650   | ns            |
| <b>Output Rise Time</b>  | $t_R$            | No $C_G$<br>See Note 7   |      | 22    | 45    | ns            |
|  |                  | $C_G = 10\text{ nF}$ ,<br>See Note 7   | 55   | 90    | 150   |               |
|  |                  | $C_G = 47\text{ nF}$ ,<br>See Note 7   | 300  | 465   | 650   |               |
| <b>Output Fall Time</b>  | $t_F$            | No $C_G$<br>See Note 8   |      | 18    | 45    | ns            |
|  |                  | $C_G = 10\text{ nF}$<br>See Note 8   | 40   | 81    | 150   |               |
|  |                  | $C_G = 47\text{ nF}$<br>See Note 8   | 300  | 460   | 650   |               |

| Parameter                                 | Symbol              | Conditions<br>$T_J = -40\text{ }^{\circ}\text{C}$ to $+125\text{ }^{\circ}\text{C}$<br>See Note 1 (Unless Otherwise Specified) | Min                     | Typ     | Max  | Units         |
|---|---------------------|--|-------------------------|---------|------|---------------|
| <b>Electrical Characteristics (cont.)</b> |                     |  |                         |         |      |               |
| <b>ASSD Rate of Change</b>                | $t_{\text{FSSD1}}$  | VGE change from 14.5 V to 14 V<br>See Note 12  |                         | 60      |      | ns            |
|   | $t_{\text{FSSD2}}$  | VGE change from 14.5 V to 2.5 V<br>See Note 12   | 950                     | 1828    | 2800 |               |
| <b>Propagation Delay Jitter</b>           |                     | See Note 12  |                         | $\pm 5$ |      | ns            |
| <b>Fault Signalization Delay Time</b>     | $t_{\text{FAULT}}$  | See Note 10  |                         | 190     | 750  | ns            |
| <b>SO Fault Signalization time</b>        | $t_{\text{SO}}$     |  | 6.8                     | 10      | 13.4 | $\mu\text{s}$ |
| <b>Power-On Start-Up Time</b>             | $t_{\text{START}}$  | See Note 11  |                         |         | 10   | ms            |
| <b>Gate Sourcing Peak Current, GH Pin</b> | $I_{\text{GH}}$     | $V_{\text{GH}} \geq V_{\text{TOT}} - 8.8\text{ V}$<br>$C_{\text{G}} = 470\text{ nF}$<br>See Note 13                            | 3.6                     | 4.6     | 5.5  | A             |
|   |                     | $R_{\text{G}} = 0, C_{\text{G}} = 47\text{ nF}$<br>See Notes 2, 12, 13<br>$T_{\text{A}} = 25\text{ }^{\circ}\text{C}$          |                         | 7.3     |      |               |
| <b>Gate Sinking Peak Current GL Pin</b>   | $I_{\text{GL}}$     | $V_{\text{GL}} \leq 7.5\text{ V}$<br>$C_{\text{G}} = 470\text{ nF}$<br>$V_{\text{GL}}$ is referenced to COM                    | 4                       | 4.8     | 5.5  | A             |
|   |                     | $R_{\text{G}} = 0, C_{\text{G}} = 47\text{ nF}$<br>See Notes 2, 12<br>$T_{\text{A}} = 25\text{ }^{\circ}\text{C}$              |                         | 7.8     |      |               |
| <b>Turn-On Internal Gate Resistance</b>   | $R_{\text{GHI}}$    | $I(\text{GH}) = 250\text{ mA}$<br>$V_{\text{IN}} = 5\text{ V}$<br>See Note 13  |                         | 0.76    | 1.2  | $\Omega$      |
| <b>Turn-Off Internal Gate Resistance</b>  | $R_{\text{GLI}}$    | $I(\text{GL}) = 250\text{ mA}$<br>$V_{\text{IN}} = 0\text{ V}$<br>See Note 13  |                         | 0.68    | 1.1  | $\Omega$      |
| <b>Turn-On Gate Output Voltage</b>        | $V_{\text{GH(ON)}}$ | $I(\text{GH}) = 20\text{ mA}$<br>$V_{\text{IN}} = 5\text{ V}$<br>See Note 13   | $V_{\text{TOT}} - 0.04$ |         |      | V             |

| Parameter   | Symbol          | Conditions<br>$T_J = -40\text{ }^{\circ}\text{C}$ to $+125\text{ }^{\circ}\text{C}$<br>See Note 1 (Unless Otherwise Specified)                                      | Min       | Typ | Max     | Units              |
|---|-----------------|---|-----------|-----|---------|--------------------|
| <b>Electrical Characteristics (cont.)</b>             |                 |   |           |     |         |                    |
| Turn-Off Gate Output Voltage<br>(Referred to COM Pin) | $V_{GL(OFF)}$   | $I_{GL} = 20\text{ mA}$<br>$V_{IN} = 0\text{ V}$  |           |     | 0.04    | V                  |
| SO Output Voltage                                     | $V_{SO(FAULT)}$ | Fault Condition, $I_{SO} = 3.4\text{ mA}$ , $V_{VCC} \geq 3.9\text{ V}$   |           | 210 | 450     | mV                 |
| <b>Package Characteristics (See Notes 12, 14)</b>     |                 |   |           |     |         |                    |
| Distance Through the Insulation                       | DTI             | Minimum Internal Gap (Internal Clearance)   | 0.4       |     |         | mm                 |
| Minimum Air Gap (Clearance)                           | L1 (IO1)        | Shortest Terminal-to-Terminal Distance Through Air  | 9.5       |     |         | mm                 |
| Minimum External Tracking (Creepage)                  | L2 (IO2)        | Shortest Terminal-to-Terminal Distance Across the Package Surface   | 9.5       |     |         | mm                 |
| Tracking Resistance (Comparative Tracking Index)      | CTI             | DIN EN 60112 (VDE 0303-11): 2010-05<br>EN / IEC 60112:2003 + A1:2009  | 600       |     |         |                    |
| Isolation Resistance, Input to Output<br>See Note 16  | $R_{IO}$        | $V_{IO} = 500\text{ V}$ , $T_J = 25\text{ }^{\circ}\text{C}$  | $10^{12}$ |     |         | $\Omega$           |
|   |                 | $V_{IO} = 500\text{ V}$ , $100\text{ }^{\circ}\text{C} \leq T_J \leq T_{C(MAX)}$  | $10^{11}$ |     |         |                    |
| Isolation Capacitance, Input to Output<br>See Note 16 | $C_{IO}$        |   |           | 1   |         | pF                 |
| <b>Package Insulation Characteristics</b>             |                 |   |           |     |         |                    |
| Maximum Working Isolation Voltage                     | $V_{IOWM}$      |   |           |     | 1202    | $V_{RMS}$          |
| Maximum Repetitive Peak Isolation Voltage             | $V_{IORM}$      |   |           |     | 1700    | $V_{PEAK}$         |
| Input to Output Test Voltage                          | $V_{PD}$        | Method A, After Environmental Tests<br>Subgroup 1, $V_{PR} = 1.3 \times V_{IORM}$ , $t = 10\text{ s}$<br>(qualification) Partial Discharge $< 5\text{ pC}$          |           |     | 2210    | $V_{PEAK}$         |
|   |                 | Method A, After Input/Output Safety Test<br>Subgroup 2/3, $V_{PR} = 1.2 \times V_{IORM}$ , $t = 10\text{ s}$ ,<br>(qualification) Partial Discharge $< 5\text{ pC}$ |           |     | 2040    |                    |
|   |                 | Method B1, 100% Production Test,<br>$V_{PR} = 1.5 \times V_{IORM}$ , $t = 1\text{ s}$<br>Partial Discharge $< 5\text{ pC}$  |           |     | 2550    |                    |
| Maximum Transient Isolation Voltage                   | $V_{IOTM}$      | $V_{TEST} = V_{IOTM}$ , $t = 60\text{ s}$ (qualification),<br>$t = 1\text{ s}$ (100% production)  |           |     | 8000    | $V_{PEAK}$         |
| Maximum Surge Isolation Voltage                       | $V_{IOSM}$      | Test Method Per IEC 60065, 1.2/50 $\mu\text{s}$<br>Waveform, $V_{TEST} = 1.3 \times V_{IOSM} = 10400\text{ V}$<br>(qualification)                                   |           |     | 8000    | $V_{PEAK}$         |
| Insulation Resistance                                 | $R_S$           | $V_{IO} = 500\text{ V}$ at $T_S$  |           |     | $>10^9$ | $\Omega$           |
| Maximum Case Temperature                              | $T_S$           |   |           |     | 150     | $^{\circ}\text{C}$ |

| Parameter   | Symbol    | Conditions<br>$T_j = -40\text{ °C to }+125\text{ °C}$<br>See Note 1 (Unless Otherwise Specified)  | Min | Typ       | Max  | Units     |
|---|-----------|---|-----|-----------|------|-----------|
| <b>Package Insulation Characteristics (cont.)</b> |           |   |     |           |      |           |
| <b>Safety Total Dissipated Power</b>              | $P_S$     | $T_j = 25\text{ °C}$  |     |           | 1.79 | W         |
| <b>Pollution Degree</b>                           |           |   |     | 2         |      |           |
| <b>Climatic Classification</b>                    |           |   |     | 50/105/21 |      |           |
| <b>Withstanding Isolation Voltage</b>             | $V_{ISO}$ | $V_{TEST} = V_{ISO}, t = 60\text{ s (qualification)}$ ,<br>$V_{TEST} = 1.2 \times V_{ISO} = 6000\text{ V}_{RMS}, t = 1\text{ s}$<br>(100% production) |     | 5000      |      | $V_{RMS}$ |



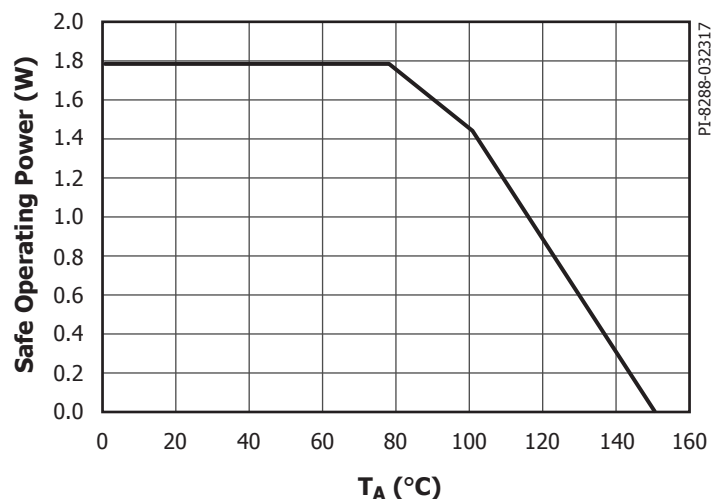


Figure 16. Thermal Derating Curve Showing Dependence of Limited Dissipated Power on Case Temperature (DIN V VDE V 0884-10).

Operation is allowed until reaching  $T_j$  and/or case temperature of 125 °C are reached. Thermal stress beyond those values but below thermal derating curve may lead to permanent functional product damage. Operating beyond thermal SR derating curve may affect product reliability.

## NOTES:

1.  $V_{VCC} = 5$  V,  $V_{TOT} = 25$  V; GH and GL pins are shorted together.  $R_G = 4$   $\Omega$ , No  $C_G$ ; VCC pin is connected to the SO pin through a 2 k $\Omega$  resistor. The VGXX pin is connected to the GH pin through a 10 nF capacitor. Typical values are defined at  $T_j = 25$  °C;  $f_s = 20$  kHz, Duty Cycle = 50%. Positive currents are assumed to be flowing into pins.
2. Pulse width  $\leq 10$   $\mu$ s, duty cycle  $\leq 1\%$ . The maximum value is controlled by the ASIC to a safe level. The internal peak power is safely controlled for  $R_G \geq 0$  and power semiconductor module input gate capacitance  $C_{IES} \leq 47$  nF.
3. During very slow  $V_{VCC}$  power-up and power-down related to  $V_{TOT}$ ,  $V_{VCC}$  and  $V_{VEE}$  respectively, several SO fault pulses may be generated.
4. SO pin connected to GND as long as  $V_{VCC}$  stays below minimum value. No signal is transferred from primary to secondary-side.
5.  $V_{IN}$  potential changes from 0 V to 5 V within 10 ns. Delay is measured from 50% voltage increase on IN pin to 10% voltage increase on GH pin.
6.  $V_{IN}$  potential changes from 5 V to 0 V within 10 ns. Delay is measured from 50% voltage decrease on IN pin to 10% voltage decrease on GL pin.
7. Measured from 10% to 90% of  $V_{GE}$  ( $C_G$  simulates semiconductor gate capacitance). The  $V_{GE}$  is measured across  $C_G$ .
8. Measured from 90% to 10% of  $V_{GE}$  ( $C_G$  simulates semiconductor gate capacitance). The  $V_{GE}$  is measured across  $C_G$ .
9. ASSD function limits G-E voltage of controlled semiconductor in specified time. Conditions:  $C_G = 10$  nF,  $V_{TOT} = V_{VISO} = 15$  V,  $V_{VEE} = 0$  V (VEE shorted to COM).
10. The amount of time needed to transfer fault event (UVLO or DESAT) from secondary-side to SO pin.
11. The amount of time after primary and secondary-side supply voltages ( $V_{VCC}$  and  $V_{TOT}$ ) reach minimal required level for driver proper operation. No signal is transferred from primary to secondary-side during that time, and no fault condition will be transferred from the secondary-side to the primary-side.
12. Guaranteed by design.
13. Positive current is flowing out of the pin.
14. Safety distances are application dependent and the creepage and clearance requirements should follow specific equipment isolation standards of an application. Board design should ensure that the soldering pads of an IC maintain required safety relevant distances.
15. Measured accordingly to IEC 61000-4-8 ( $f_s = 50$  Hz, and 60 Hz) and IEC 61000-4-9.
16. All pins on each side of the barrier tied together creating a two-terminal device.

# Typical Performance Characteristics

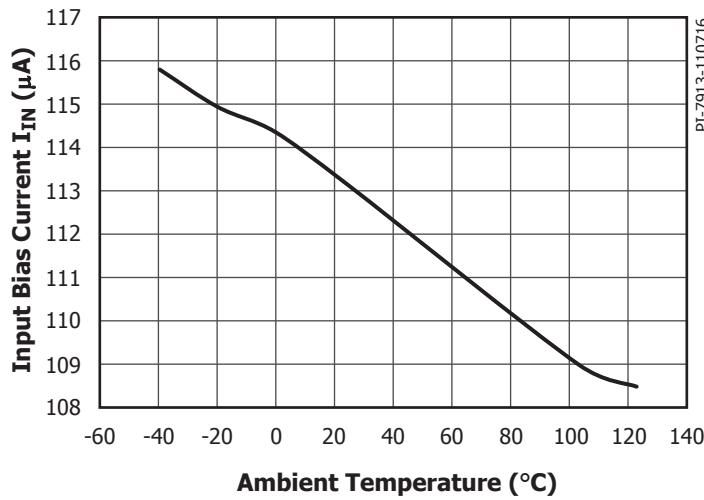


Figure 17. Input Bias Current vs. Ambient Temperature.  
Conditions:  $V_{VCC} = 5$  V,  $V_{IN} = 5$  V,  $V_{TOT} = 25$  V.

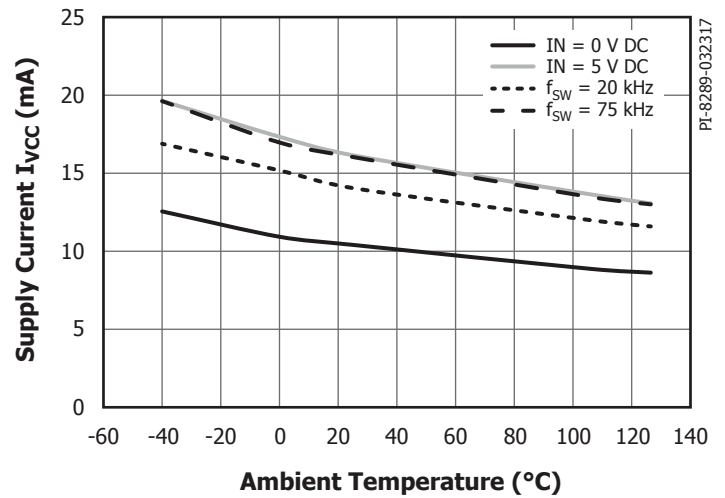


Figure 18. Supply Current Primary-Side  $I_{VCC}$  vs. Ambient Temperature.  
Conditions:  $V_{VCC} = 5$  V,  $V_{TOT} = 25$  V, No-Load.

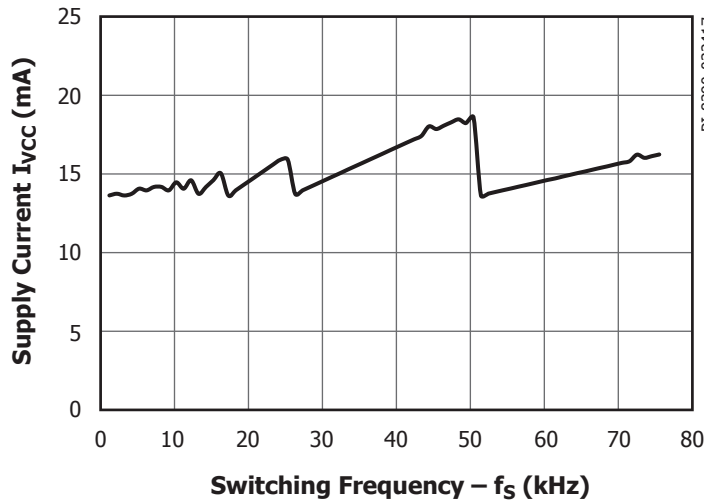


Figure 19. Supply Current Primary-Side  $I_{VCC}$  vs. Switching Frequency.  
Conditions:  $V_{VCC} = 5$  V,  $V_{TOT} = 25$  V,  $T_J = 25$  °C, No-Load.

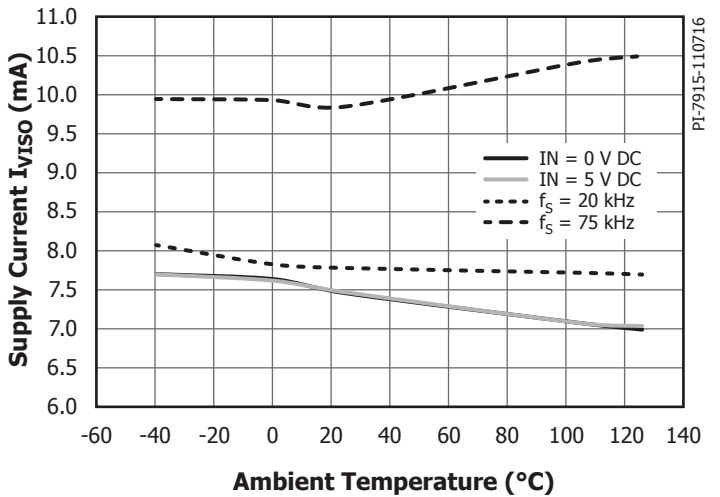


Figure 20. Supply Current Secondary-Side  $I_{VISO}$  vs. Ambient Temperature.  
Conditions:  $V_{VCC} = 5$  V,  $V_{TOT} = 25$  V, No-Load.

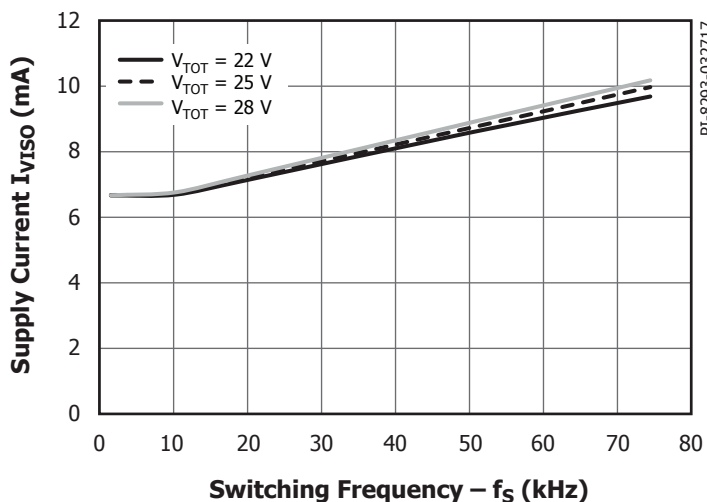


Figure 21. Supply Current Secondary-Side  $I_{VISO}$  vs. Switching Frequency.  
Conditions:  $V_{VCC} = 5$  V, No-Load.

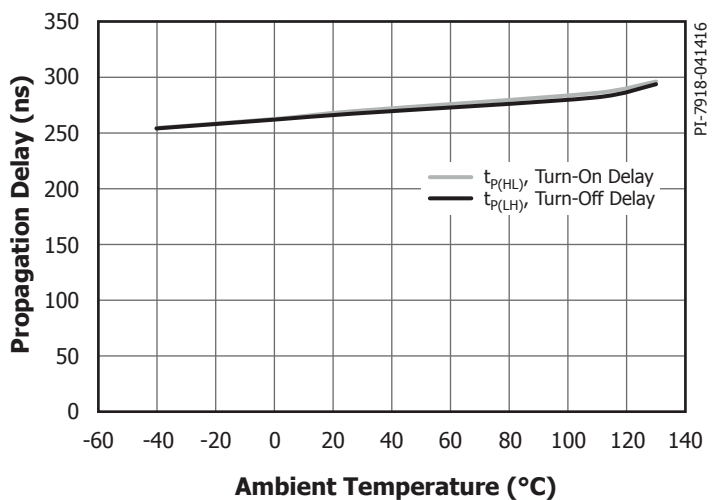
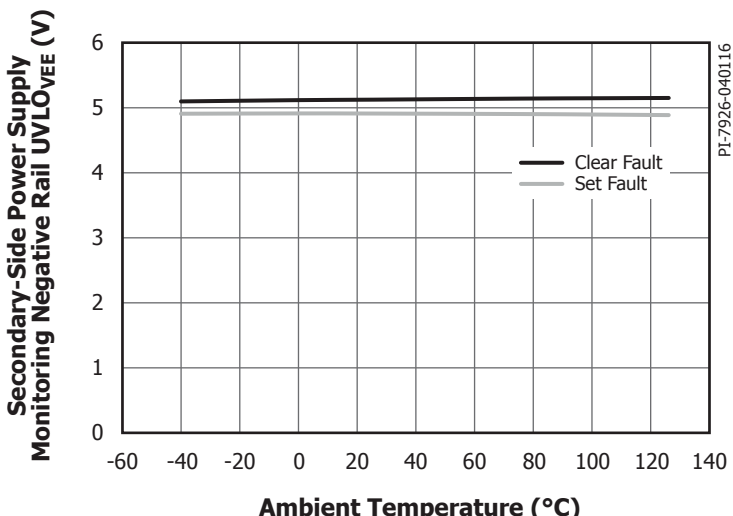
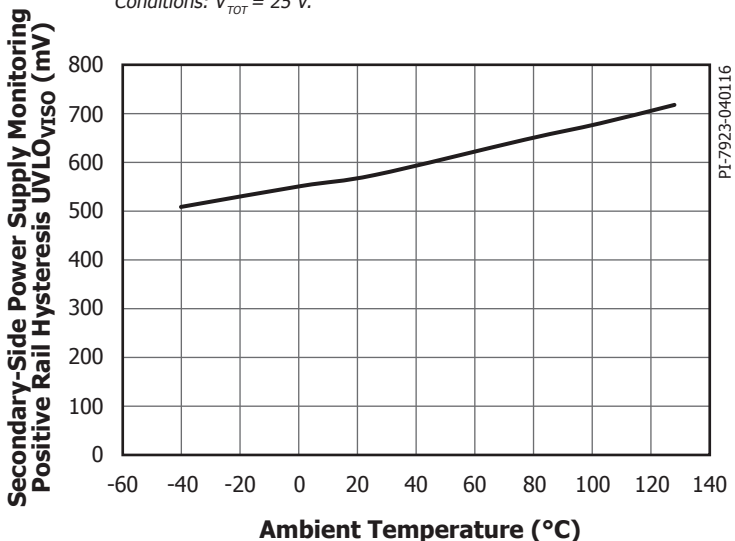
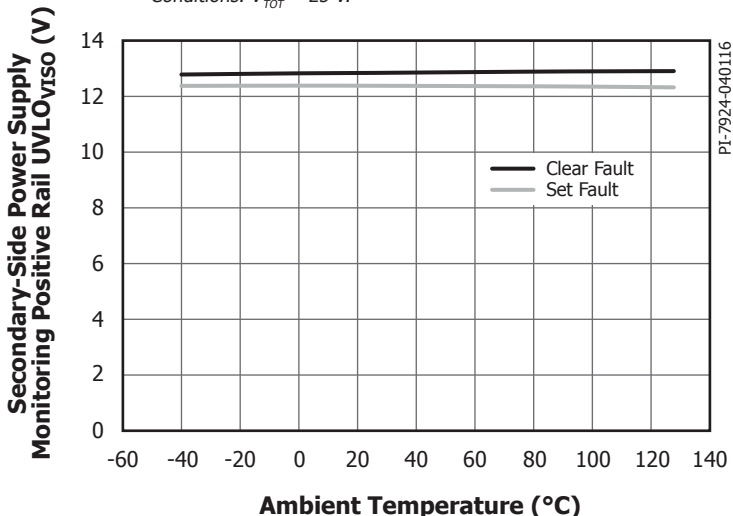
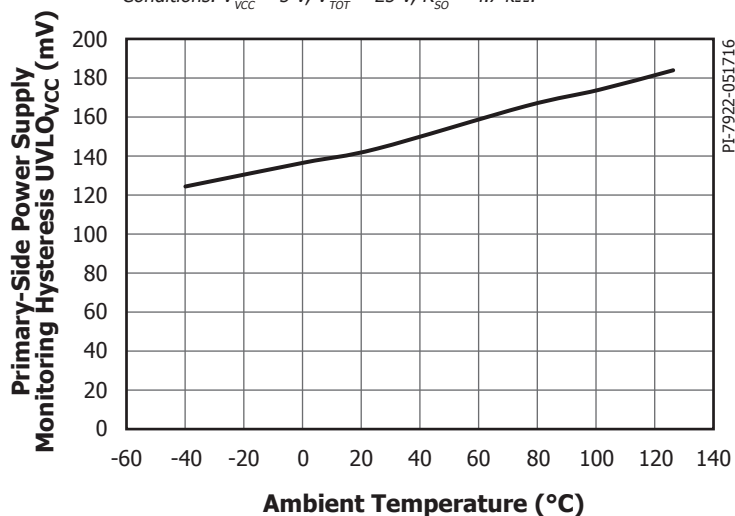
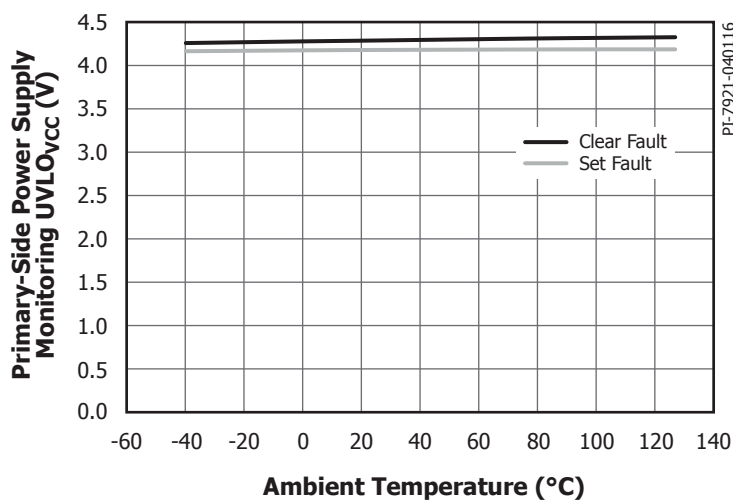
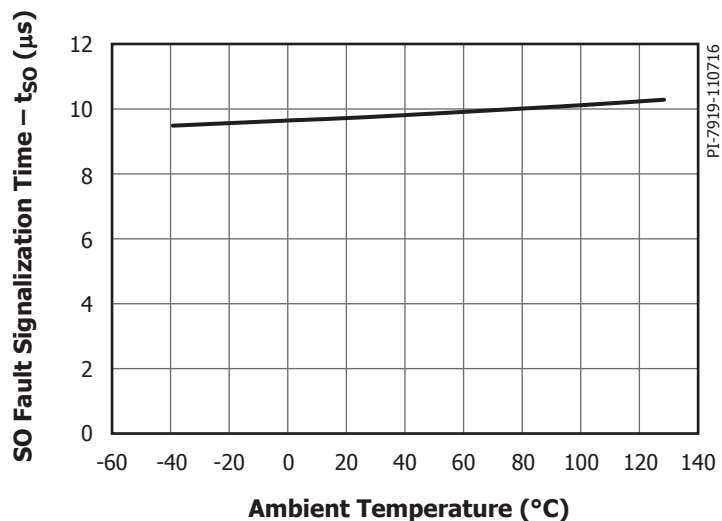


Figure 22. Propagation Delay Time vs. Ambient Temperature.  
Conditions:  $V_{VCC} = 5$  V,  $V_{TOT} = 25$  V,  $f_S = 20$  kHz,  $C_{LOAD} = 2.2$  nF.

# Typical Performance Characteristics



# Typical Performance Characteristics

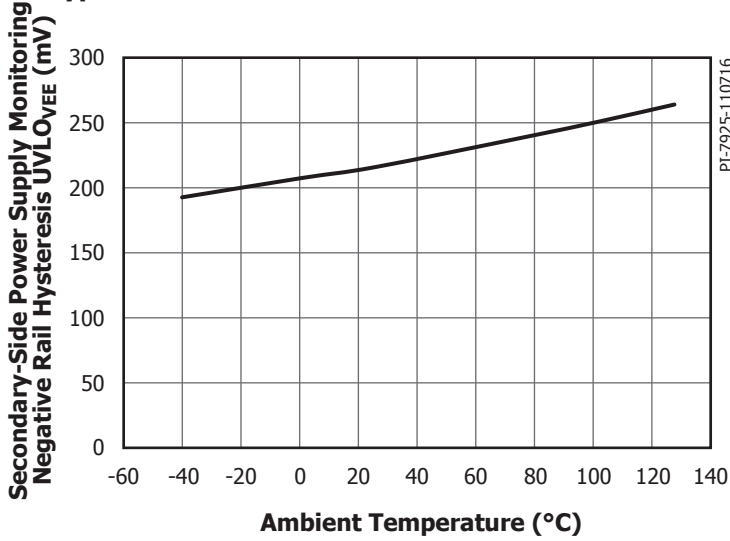


Figure 29. Power Supply Monitoring Negative Rail Hysteresis  $UVLO_{VEE}$  vs. Ambient Temperature. Conditions:  $V_{VCC} = 5 V$ .

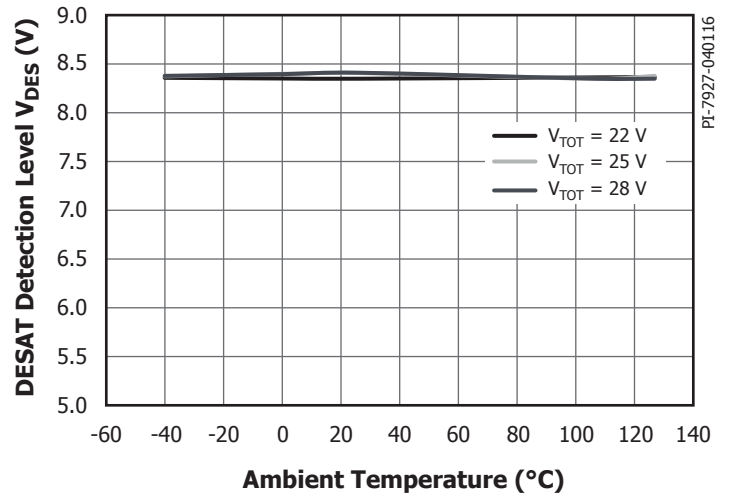


Figure 30. Desaturation Detection Level  $V_{DES}$  vs. Ambient Temperature. Conditions:  $V_{VCC} = 5 V$ .

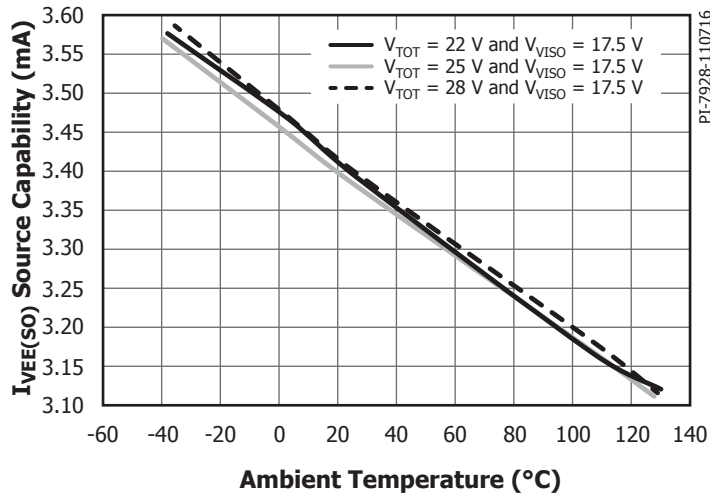


Figure 31. VEE Source Capability  $I_{VEE(SO)}$  vs. Ambient Temperature and  $V_{VISO}$ . Conditions:  $V_{VCC} = 5 V$ ,  $f_s = 20 kHz$ , Duty Cycle = 50%.

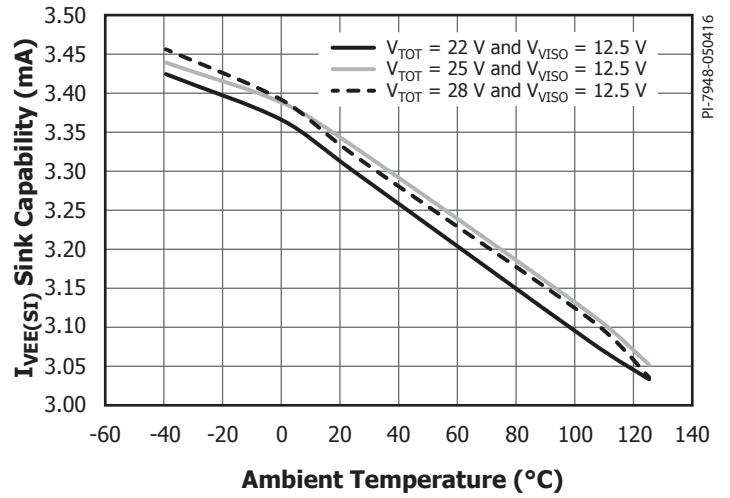
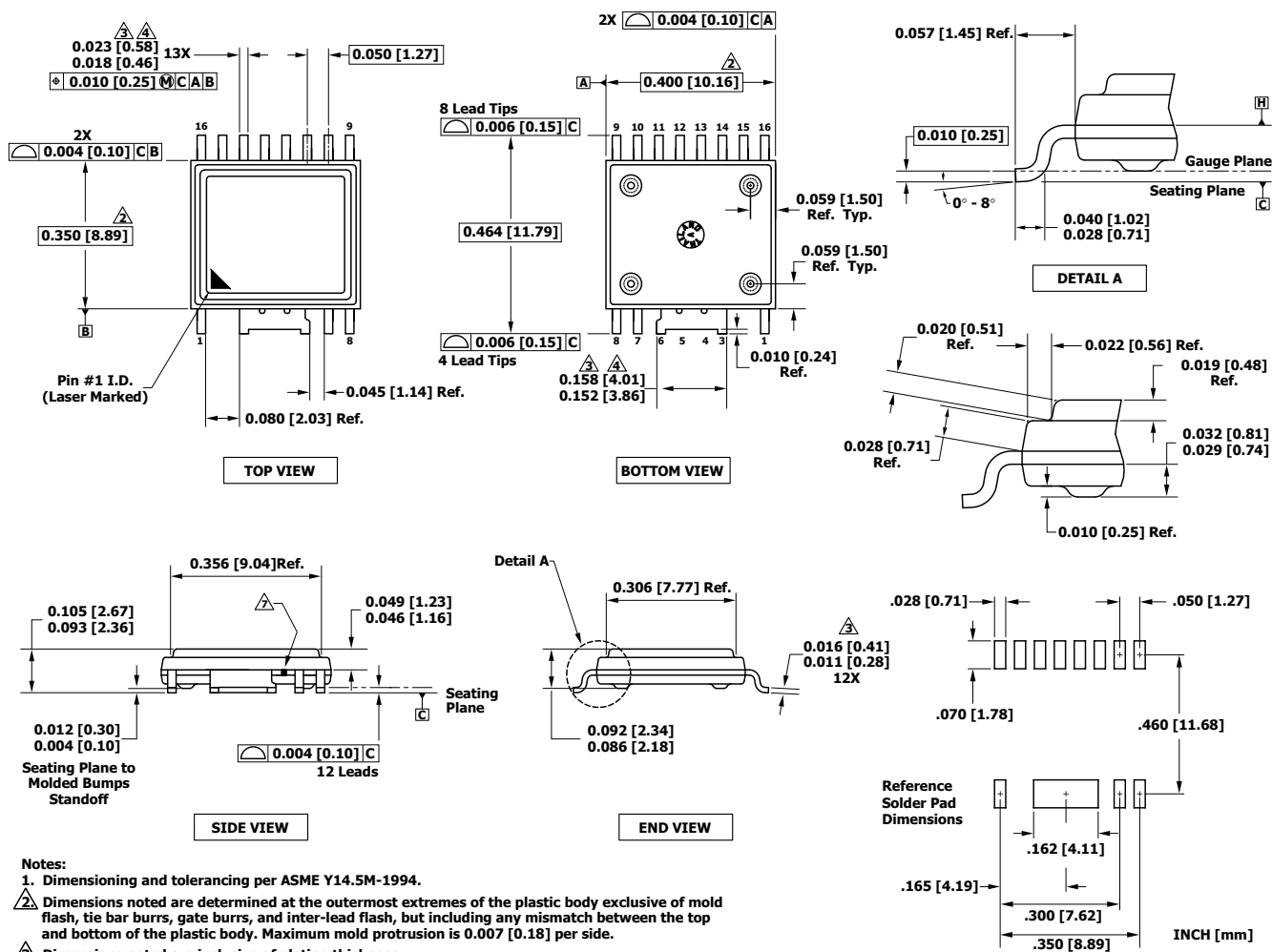


Figure 32. VEE Sink Capability  $I_{VEE(SI)}$  vs. Ambient Temperature and  $V_{VISO}$ . Conditions:  $V_{VCC} = 5 V$ ,  $f_s = 20 kHz$ , Duty Cycle = 50%.

## eSOP-R16B



## Notes:

1. Dimensioning and tolerancing per ASME Y14.5M-1994.

⚠ Dimensions noted are determined at the outermost extremes of the plastic body exclusive of mold flash, tie bar burrs, gate burrs, and inter-lead flash, but including any mismatch between the top and bottom of the plastic body. Maximum mold protrusion is 0.007 [0.18] per side.

⚠ Dimensions noted are inclusive of plating thickness.

⚠ Does not include inter-lead flash or protrusions.

5. Controlling dimensions in inches [mm].

6. Datums A and B to be determined in Datum H.

⚠ Exposed metal at the plastic package body outline/surface between leads 6 and 7, connected internally to wide lead 3/4/5/6.

PI-6995-051716  
POD-eSOP-R16B Rev B

## MSL Table

| Part Number | MSL Rating |
|-------------|------------|
| SID1183K    | 3          |

## ESD and Latch-Up Table

| Test                     | Conditions   | Results   |
|--------------------------|--------------|---|
| Latch-up at 125 °C       | JESD78D      | $> \pm 100 \text{ mA}$ or $> 1.5 \times V_{\text{MAX}}$ on all pins |
| Human Body Model ESD     | JESD22-A114F | $> \pm 2000 \text{ V}$ on all pins                                  |
| Charged Device Model ESD | JESD22-C101  | $> \pm 500 \text{ V}$ on all pins                                   |
| Machine Model ESD        | JESD22-A115C | $> \pm 200 \text{ V}$ on all pins                                   |

## IEC 60664-1 Rating Table

| Parameter                   | Conditions   | Specifications |
|-----------------------------|--|----------------|
| Basic Isolation Group       | Material Group   | I              |
| Installation Classification | Rated mains voltage $\leq 150 \text{ V}_{\text{RMS}}$  | I - IV         |
|                             | Rated mains voltage $\leq 300 \text{ V}_{\text{RMS}}$  | I - IV         |
|                             | Rated mains voltage $\leq 600 \text{ V}_{\text{RMS}}$  | I - IV         |
|                             | Rated mains voltage $\leq 1000 \text{ V}_{\text{RMS}}$ | I - III        |

## Electrical Characteristics (EMI) Table

| Parameter                                  | Symbol             | Conditions   | Min | Typ      | Max        | Units             |
|--|--------------------|--|-----|----------|------------|-------------------|
| Common-Mode Transient Immunity, Logic High | $CM_H$             | Typical values measured according to Figures 33, 34. Maximum values are design values assuming trapezoid waveforms |     | -35 / 50 | -100 / 100 | kV/ $\mu\text{s}$ |
| Common-Mode Transient Immunity, Logic Low  | $CM_L$             | Typical values measured according to Figures 33, 34. Maximum values are design values assuming trapezoid waveforms |     | -35 / 50 | -100 / 100 | kV/ $\mu\text{s}$ |
| Variable Magnetic Field Immunity           | $H_{\text{HPEAK}}$ | See Note 15  |     | 1000     |            | A/m               |
|  | $H_{\text{LPEAK}}$ | See Note 15  |     | 1000     |            |                   |

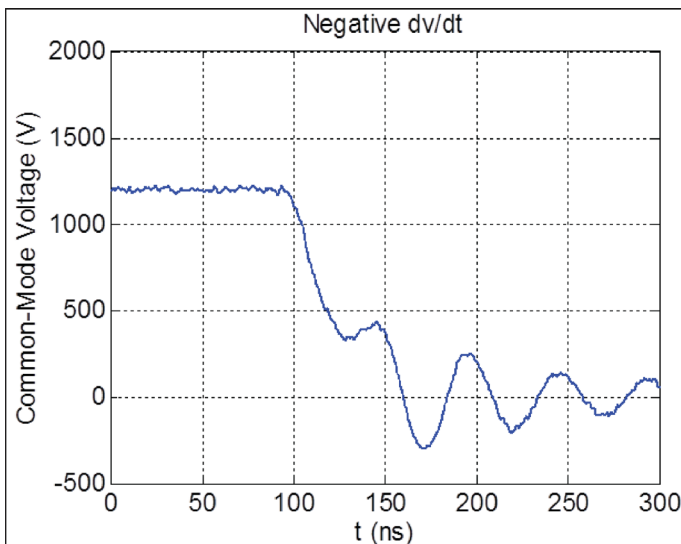


Figure 33. Applied Common Mode Pulses for Generating Negative dv/dt.

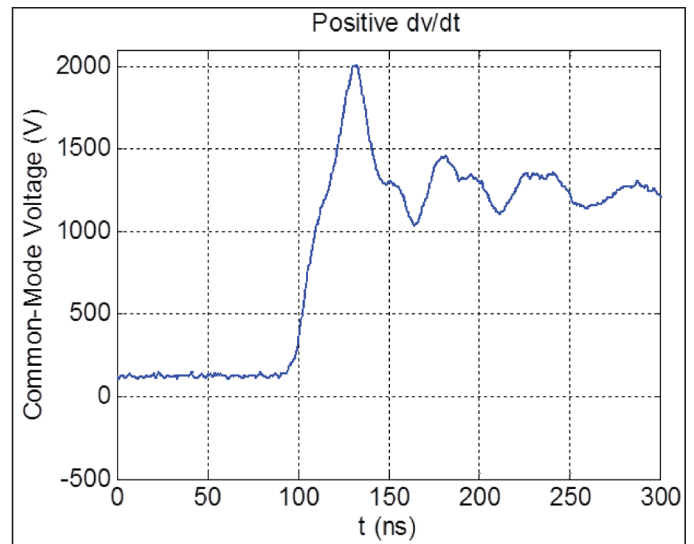
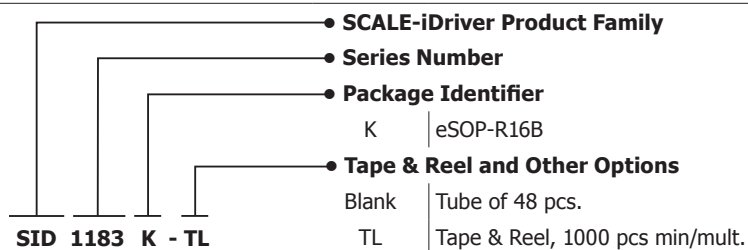


Figure 34. Applied Common Mode Pulses for Generating Positive dv/dt.

## Regulatory Information Table

| VDE  | UL  | CSA   |
|--|---|---|
| Certified to DIN V VDE V 0884-10 (VDE V 0884-10): 2006-12  | UR recognized under UL1577 Component Recognition Program              | UR recognized to CSA Component Acceptance Notice 5A                   |
| Basic insulation for Max. Transient Isolation voltage 8 kV <sub>PEAK</sub> , Max. Surge Isolation voltage 8 kV <sub>PEAK</sub> , Max. Repetitive Peak Isolation voltage 1700 V <sub>PEAK</sub> | Single protection, 5000 V <sub>RMS</sub> dielectric voltage withstand | Single protection, 5000 V <sub>RMS</sub> dielectric voltage withstand |
| File No. 5020828-4880-0002   | File No. Pending  | File No. Pending  |

## Part Ordering Information





## Notes

| Revision | Notes                   | Date  |
|----------|-------------------------|-------|
| A        | Code A Initial Release. | 05/17 |

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