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FAN6863WTY

Highly Integrated Green-Mode PWM Controller

Features

- Low Standby Power: Under 0.1 W
- Low Startup Current: 8 μ A
- Low Operating Current in Green Mode: 600 μ A
- Peak-Current Mode Operation with Cycle-by-Cycle Current Limiting
- PWM Frequency Continuously Decreasing with Burst Mode at Light Loads
- V_{DD} Over-Voltage Protection (OVP)
- Constant Output Power Limit (Full AC Input Range)
- Internal Latch Circuit for OVP, OTP
- SENSE Pin Short-Circuit Protection (SSCP)
- Fixed PWM Frequency (65 KHz) with Frequency Hopping
- Feedback Open-Loop Protection: 60 ms Delay
- GATE Output Maximum Voltage Clamp: 13.5 V
- Soft-Start Time: 5 ms
- Soft Driving for EMI Improvement
- Full Range Frequency Hopping
- Internal OTP Sensor with Hysteresis
- Gate Driving Capability: 400 mA

Applications

General-purpose switched-mode power supplies and flyback power converters, including:

- Power Adapters
- Open-Frame SMPS
- SMPS with Surge-Current Output, such as for Printers, Scanners, Motor Drivers

Description

A highly integrated PWM controller, FAN6863WTY provides several features to enhance the performance of flyback converters. To minimize standby power consumption, a proprietary Green Mode provides off-time modulation to continuously decrease the switching frequency under light-load conditions. Under zero-load conditions, the power supply enters Burst Mode, which completely shuts off PWM output. Output restarts just before the supply voltage drops below the UVLO lower limit. Green Mode enables power supplies to meet international power conservation requirements.

The FAN6863WTY is designed for SMPS and integrates a frequency-hopping function that helps to reduce EMI emission of a power supply with minimum line filters. To compensate the power limit variation over universal input range, a current limit (V_{LIMIT}) adaptively keeps the power limit substantially constant. The gate output is clamped at 13.5 V to protect the external MOSFET from over-voltage damage.

Other protection functions include SENSE pin Short-Circuit Protection (SSCP), V_{DD} Over-Voltage Protection (OVP), and Over-Temperature Protection (OTP). For OTP, an external NTC thermistor can be applied to sense the ambient temperature. When V_{DD} OVP or OTP is activated, an internal latch circuit latches off the controller. Protection types are shown in Table 1.

Table 1. Protection Type

| Part Number | OVP | OLP | OTP / OTP2 | SSCP |
|-------------|-------|-----|------------|------|
| FAN6863WTY | Latch | A/R | Latch | A/R |

Ordering Information

| Part Number | Operating Temperature Range | Package | Packing Method |
|-------------|-----------------------------|--|----------------|
| FAN6863WTY | -40 to +105°C | 6-Lead, SuperSOT™-6, JEDEC M0-193, 1.6 mm Wide | Tape & Reel |

Typical Application

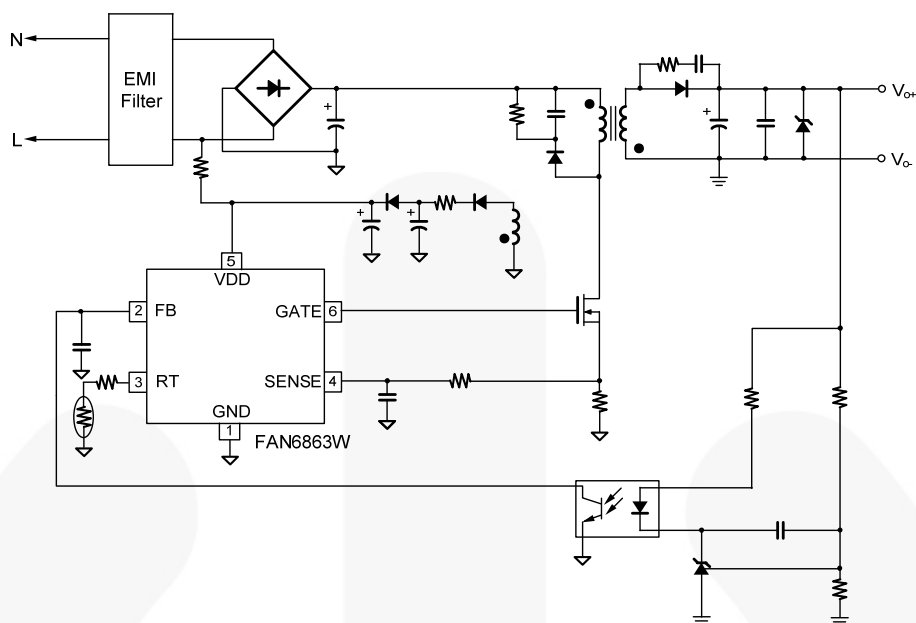


Figure 1. Typical Application

Block Diagram

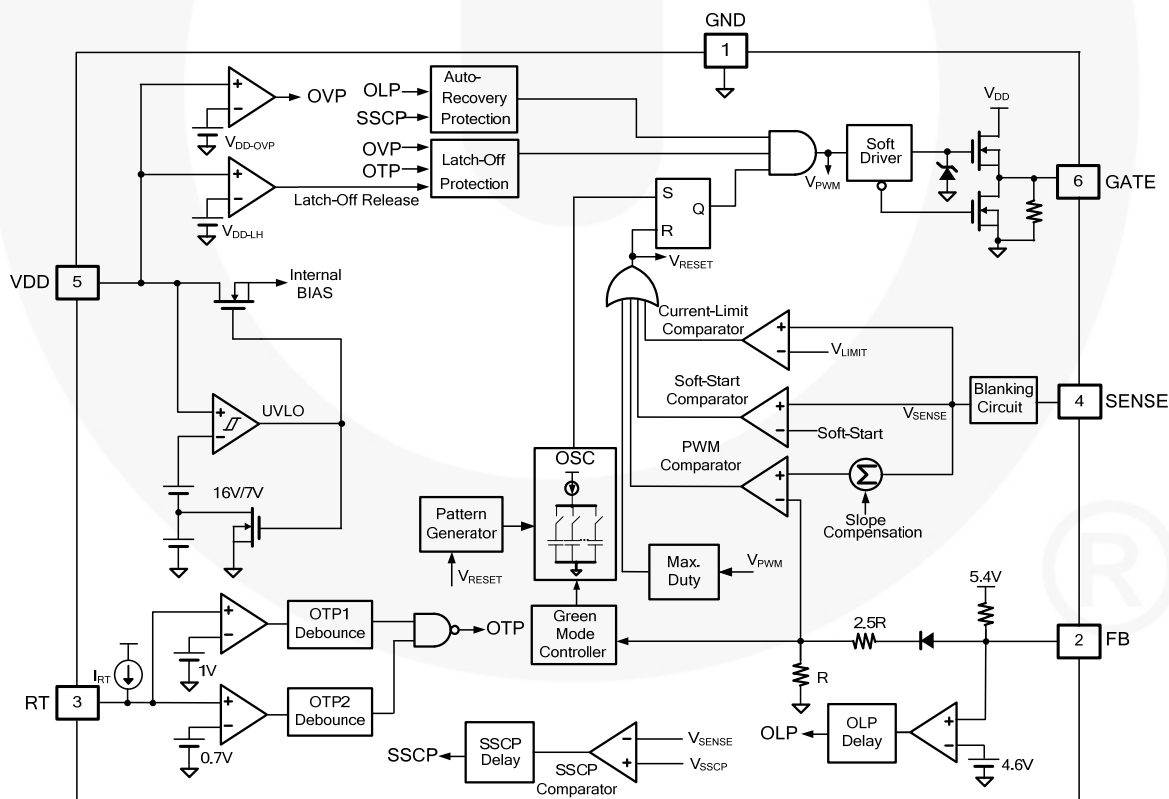


Figure 2. Block Diagram

Marking Information

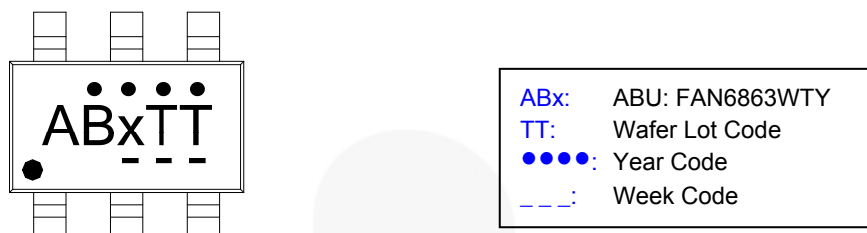


Figure 3. Top Mark

Pin Configuration

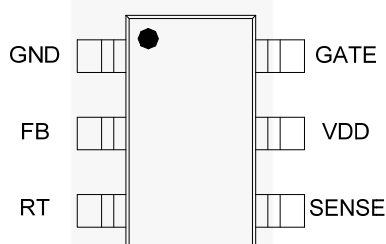


Figure 4. Pin Configuration

Pin Definitions

| Pin # | Name | Function | Description |
|-------|-------|-----------------------|--|
| 1 | GND | Ground | Ground |
| 2 | FB | Feedback | The FB pin provides the output voltage regulation signal. It provides feedback to the internal PWM comparator, so the PWM comparator can control the duty cycle. This pin also provides over-current protection. If V_{FB} is higher than the trigger level and persists at that level, the controller stops and restarts. |
| 3 | RT | Temperature Detection | An external NTC thermistor is connected from this pin to the GND pin. The impedance of the NTC decreases at high temperatures. If the voltage of the RT pin drops below the threshold, PWM output is disabled. |
| 4 | SENSE | Current Sense | This pin senses the voltage across a resistor. When the voltage reaches the internal threshold, PWM output is disabled and this activates over-current protection. This pin also provides current amplitude information for Current Mode control. |
| 5 | VDD | Power Supply | Power supply |
| 6 | GATE | Driver Output | The totem-pole output driver for driving the power MOSFET |

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only. All voltage values, except differential voltages, are given with respect to GND pin.

| Symbol | Parameter | Min. | Max. | Unit |
|---------------|--|------|------|------|
| V_{DD} | Supply Voltage | | 30 | V |
| V_L | Input Voltage to FB, SENSE, and RT Pins | -0.3 | 7.0 | V |
| Θ_{JA} | Thermal Resistance (Junction-to-Ambient) | | 244 | °C/W |
| T_J | Operating Junction Temperature | -40 | +125 | °C |
| T_{STG} | Storage Temperature Range | -55 | +150 | °C |
| T_L | Lead Temperature, Wave Soldering, 10 Seconds | | +260 | °C |
| ESD | Human Body Model, JESD22-A114 | | 5.5 | kV |
| | Charge Device Model, JESD22-C101 | | 2.0 | |

Recommended Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

| Symbol | Parameter | Min. | Max. | Unit |
|--------|-------------------------------|------|------|------|
| T_A | Operating Ambient Temperature | -40 | +105 | °C |

Electrical Characteristics

$V_{DD} = 15\text{ V}$ and $T_A = 25^\circ\text{C}$, unless otherwise noted.

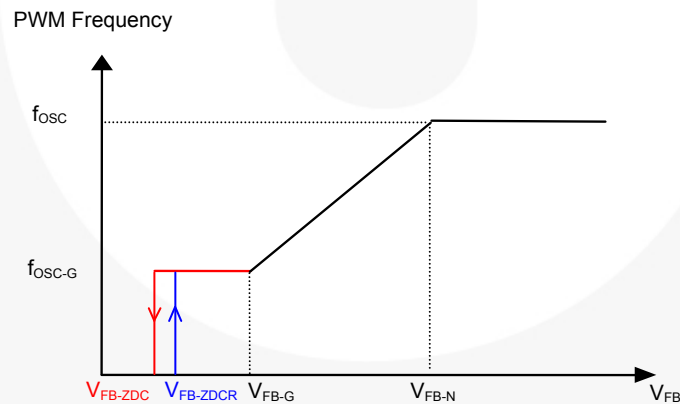
| Symbol | Parameter | Condition | Min. | Typ. | Max. | Unit |
|-------------------------------|---|--|-------|-------|-------|---------------|
| V_{DD} Section | | | | | | |
| V_{DD-OP} | Continuously Operating Voltage | | | | 20 | V |
| V_{DD-ON} | Turn-On Threshold Voltage | | 15 | 16 | 17 | V |
| V_{DD-OFF} | Turn-Off Voltage | | 6.5 | 7.0 | 7.5 | V |
| V_{DD-LH} | Threshold Voltage for Latch-Off Release | | | 4 | | V |
| I_{DD-ST} | Startup Current | $V_{DD-ON} - 0.16\text{ V}$ | | 8 | 15 | μA |
| I_{DD-OP1} | Operating Supply Current in PWM Operation | $V_{DD} = 20\text{ V}$, $V_{FB} = 3\text{ V}$ Gate Open | | | 2 | mA |
| I_{DD-OP2} | Operating Supply Current when $V_{FB} < V_{FB-ZDC}$ | $V_{DD} = 15\text{ V}$, $V_{FB} < V_{FB-ZDC}$ | | 600 | | μA |
| V_{DD-OVP} | V_{DD} Over-Voltage Protection | | 21.0 | 22.2 | 23.5 | V |
| $t_{D-VDDOVP}$ | V_{DD} OVP Debounce Time | | | 50 | | μs |
| I_{DD-LH} | Latch-Off Holding Current | $V_{DD} = 5\text{ V}$ | | 70 | 80 | μA |
| Feedback Input Section | | | | | | |
| A_V | Input-Voltage to Current-Sense Attenuation | | 1/4.0 | 1/3.5 | 1/3.0 | V/V |
| Z_{FB} | Input Impedance | | | 17 | | k Ω |
| $V_{FB-OPEN}$ | FB Pin Open Voltage | | 5.2 | 5.4 | 5.6 | V |
| V_{FB-OLP} | Threshold Voltage for Open-Loop Protection | | 4.3 | 4.6 | 4.9 | V |
| t_{D-OLP} | Open-Loop Protection Delay | $V_{FB} > V_{FB-OLP}$, $t_{ON} > 2.5\text{ }\mu\text{s}$, $T_A = -40\text{ to }+105^\circ\text{C}$ | 54 | 60 | 66 | ms |
| t_{D-SCP} | Secondary Short-Circuit Protection Delay | $FB > V_{FB-OLP}$, $t_{ON} < 2.5\text{ }\mu\text{s}$, $T_A = -40\text{ to }+105^\circ\text{C}$ | 6 | 7 | 8 | ms |
| t_{ON-SCP} | Short-Circuit Protection On-Time Detection | $V_{FB} > V_{FB-OLP}$, $T_A = -40\text{ to }+105^\circ\text{C}$ | | 2.5 | | μs |
| Current Sense Section | | | | | | |
| t_{PD} | Delay to Output | | | 100 | 250 | ns |
| t_{LEB} | Leading-Edge Blanking Time | | 200 | 250 | | ns |
| $V_{LIMIT-H}$ | HIGH Threshold Voltage for Current Limit | Duty > 55% | 0.57 | 0.60 | 0.63 | V |
| $V_{LIMIT-L}$ | LOW Threshold Voltage for Current Limit | Duty = 0% | 0.36 | 0.39 | 0.42 | V |
| $t_{SOFT-START}$ | Period During Startup Time | Startup Time | 4.75 | 5.00 | 10.00 | ms |
| V_{SSCP} | Threshold Voltage for SENSE Short-Circuit Protection | $t_{ON} > 4.5\text{ }\mu\text{s}$, $FB < V_{FB-OLP}$, $T_A = -40\text{ to }+105^\circ\text{C}$ | | 110 | | mV |
| $t_{ON-SSCP}$ | Detect SENSE On Time for SENSE Short-Circuit Protection | | | 5 | | μs |
| t_{SSCP} | Debounce Time for SENSE Short-Circuit Protection | $t_{ON} > 4.5\text{ }\mu\text{s}$, $V_{FB} < V_{FB-OLP}$, $T_A = -40\text{ to }+105^\circ\text{C}$ | | 100 | | μs |

Continued on the following page...

Electrical Characteristics (Continued) $V_{DD} = 15$ and $T_A = 25^\circ\text{C}$, unless otherwise noted.

| Symbol | Parameter | Condition | Min. | Typ. | Max. | Unit |
|----------------------------|--|--|---------------------------|------|-----------|------|
| Oscillator Section | | | | | | |
| f_{OSC} | Normal PWM Frequency | Center Frequency | $V_{FB} > V_{FB-N}$ | 60 | 65 | 68 |
| | | Hopping Range | $V_{FB} \geq V_{FB-N}$ | | ± 4.0 | kHz |
| | | | $V_{FB} = V_{FB-G}^{(1)}$ | | ± 2.9 | |
| t_{hop-1} | Hopping Period 1 ⁽¹⁾ | $V_{FB} \geq V_{FB-N}$ | | 4.4 | | ms |
| t_{hop-3} | Hopping Period 3 ⁽¹⁾ | $V_{FB} = V_{FB-G}$ | | 11.5 | | ms |
| f_{OSC-G} | Green Mode Minimum Frequency | | 18 | 22 | 26 | kHz |
| V_{FB-N} | FB Threshold Voltage for Frequency Reduction | | 2.35 | 2.50 | 2.65 | V |
| V_{FB-G} | FB Voltage at f_{OSC-G} | | 2.05 | 2.20 | 2.30 | V |
| V_{FB-ZDC} | FB Threshold Voltage for Zero-Duty | | | 1.6 | | V |
| $V_{FB-ZDCR} - V_{FB-ZDC}$ | ZDC Hysteresis | | | 0.15 | | V |
| f_{DV} | Frequency Variation vs. V_{DD} Deviation | $V_{DD} = 7.5\text{ V to }21\text{ V}$ | | 0.5 | 2.0 | % |
| f_{DT} | Frequency Variation vs. Temperature Deviation ⁽¹⁾ | $T_A = -40\text{ to }+105^\circ\text{C}$ | | | 2 | % |

Continued on following page...

**Figure 5. PWM Frequency**

Electrical Characteristics (Continued) $V_{DD} = 15V$ and $T_A = 25^\circ C$, unless otherwise noted.

| Symbol | Parameter | Condition | Min. | Typ. | Max. | Unit |
|--|---|----------------------------|------|--------------|------|------------|
| PWM Output Section | | | | | | |
| DCY_{MAX} | Maximum Duty Cycle | | 68 | 75 | 85 | % |
| V_{OL} | Output Voltage LOW | $V_{DD} = 15V, I_O = 50mA$ | | | 1.5 | V |
| V_{OH} | Output Voltage HIGH | $V_{DD} = 8V, I_O = 50mA$ | 6 | | | V |
| t_R | Rising Time (with Soft Driving) | $GATE = 1nF$ | | 150 | | ns |
| t_F | Falling Time | $GATE = 1nF$ | | 35 | | ns |
| V_{CLAMP} | Gate Output Clamping Voltage | $V_{DD} = 20V$ | 12.0 | 13.5 | 15.0 | V |
| $I_{O-SOURCE}$ | Gate Source Driving Capability ⁽¹⁾ | $V_{DD} = 15V$ | | 400 | | mA |
| I_{O-SINK} | Gate Sink Driving Capability ⁽¹⁾ | $V_{DD} = 15V$ | | 400 | | mA |
| Over-Temperature Protection (OTP) Section | | | | | | |
| R_{RT} | Maximum External Resistance of RT Pin to Trigger Protection | | 9 | 10 | 11 | k Ω |
| V_{OTP} | Threshold Voltage for Over-Temperature Protection | | 0.94 | 1.00 | 1.06 | V |
| I_{RT} | Output Current of RT Pin | | 92 | 100 | 108 | μA |
| t_{DOTP} | Over-Temperature Debounce Time | $V_{FB} = V_{FB-N}$ | 14 | 17 | 19 | ms |
| V_{OTP2} | Second Threshold Voltage for Over-Temperature Protection | | 0.65 | 0.70 | 0.75 | V |
| t_{DOTP2} | Second Over-Temperature Debounce Time | | 80 | 135 | 200 | μs |
| T_{OTP} | Protection Junction Temperature ^(1, 2) | | | +135 | | $^\circ C$ |
| $T_{Restart}$ | Restart Junction Temperature ^(1, 3) | | | T_{OTP-25} | | $^\circ C$ |

Notes:

1. Guarantee by design.
2. When activated, the output is disabled and the latch is turned off.
3. The threshold temperature for enabling the output again and resetting the latch after OTP has been activated.

Typical Performance Characteristics

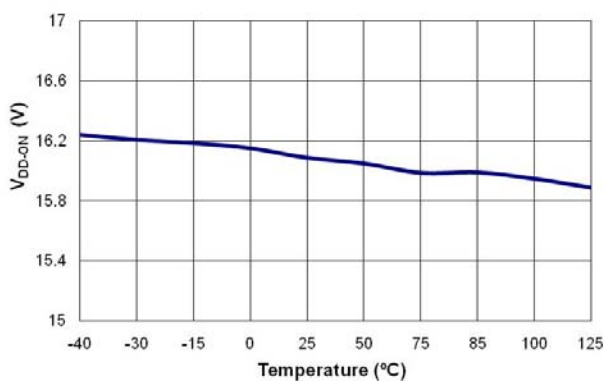


Figure 6. Turn-On Threshold Voltage (V_{DD-ON}) vs. Temperature

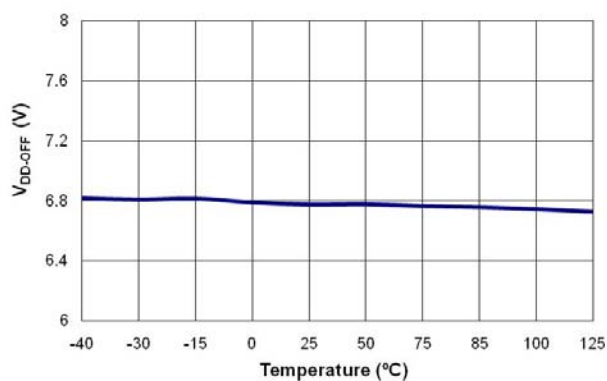


Figure 7. Turn-Off Threshold Voltage (V_{DD-OFF}) vs. Temperature

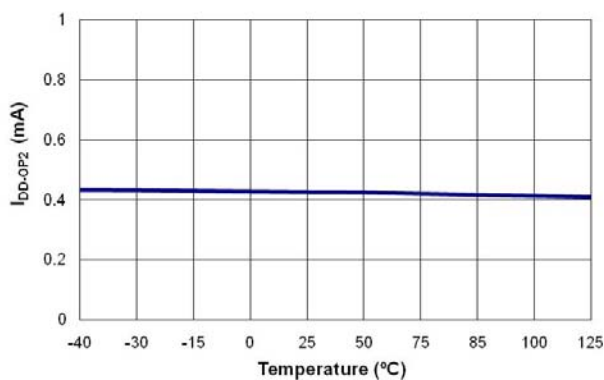


Figure 8. Operating Current (I_{DD-OP2}) vs. Temperature

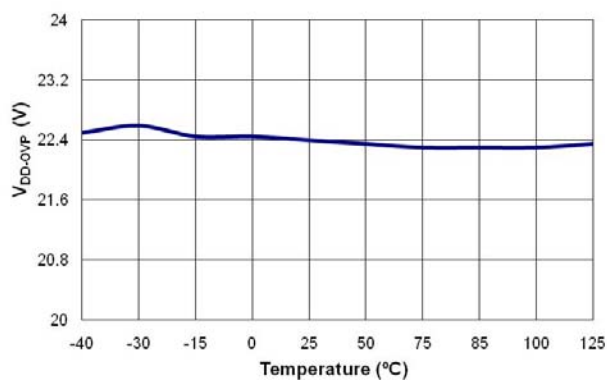


Figure 9. V_{DD} Over-Voltage Protection (V_{DD-OVP}) vs. Temperature

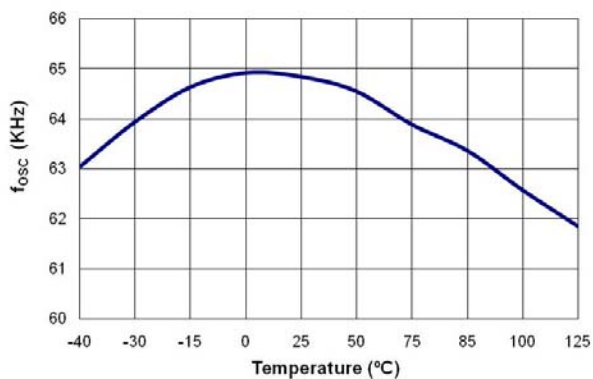


Figure 10. Center Frequency (f_{OSC}) vs. Temperature

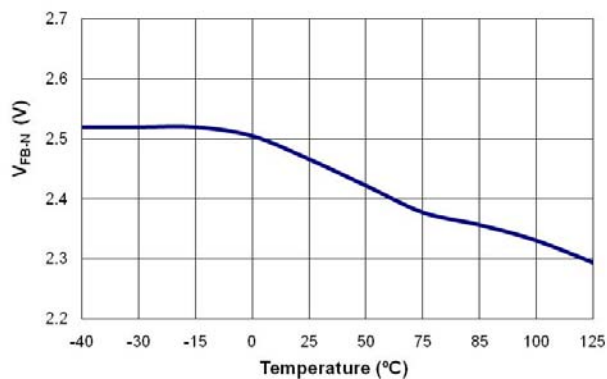
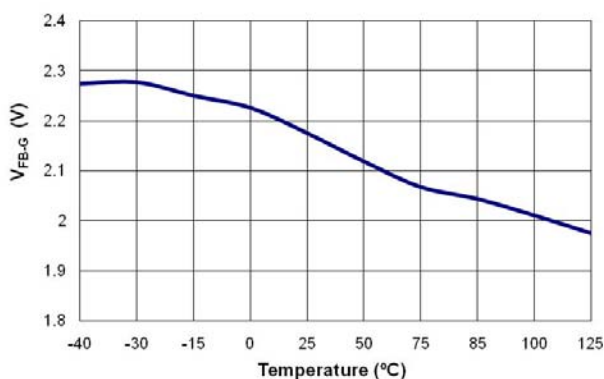
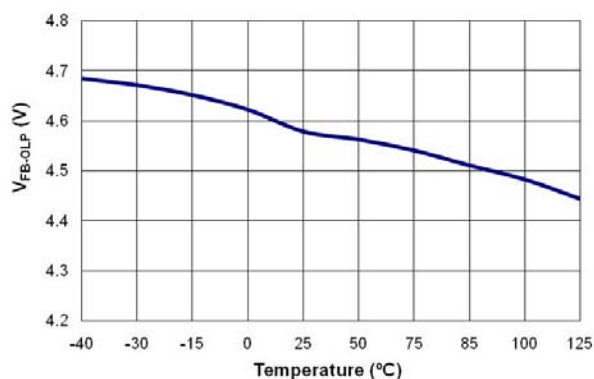
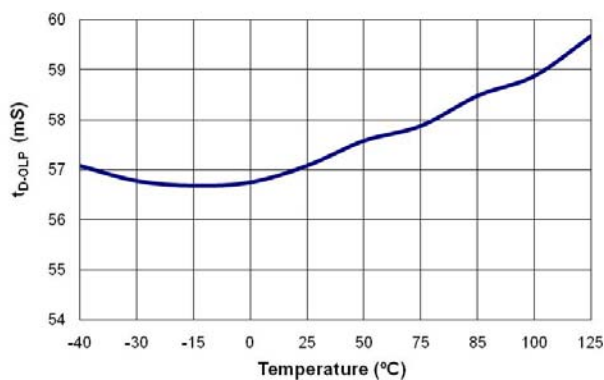
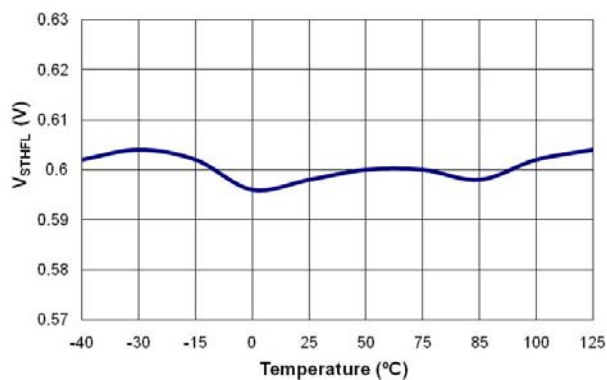
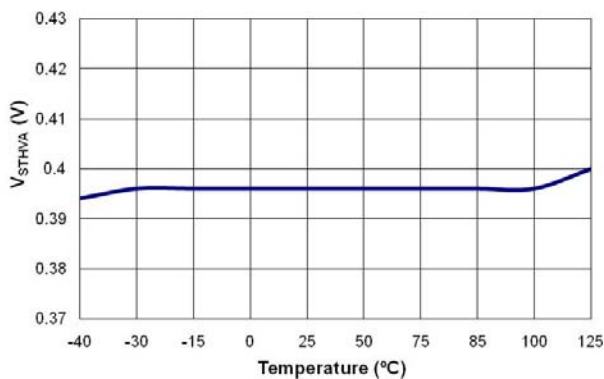
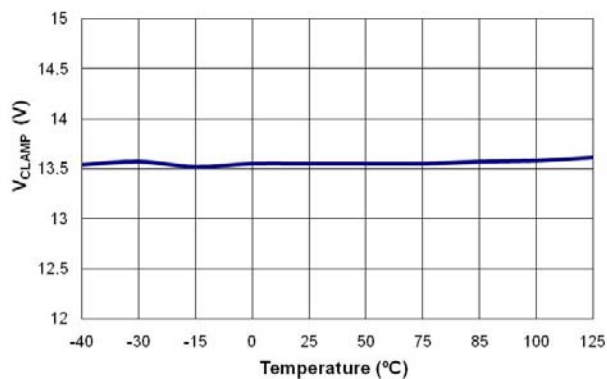


Figure 11. FB Threshold Voltage for Frequency Reduction (V_{FB-N}) vs. Temperature

Typical Performance Characteristics (Continued)

Figure 12. FB Voltage at f_{OSC-G} (V_{FB-G}) vs. TemperatureFigure 13. Threshold Voltage for Open-Loop Protection (V_{FB-OLP}) vs. TemperatureFigure 14. Open-Loop Protection Delay (t_{D-OLP}) vs. TemperatureFigure 15. Flat Threshold Voltage for Current Limit ($V_{LIMIT-H}$) vs. TemperatureFigure 16. Valley Threshold Voltage for Current Limit ($V_{LIMIT-L}$) vs. TemperatureFigure 17. GATE Output Clamping Voltage (V_{CLAMP}) vs. Temperature

Typical Performance Characteristics (Continued)

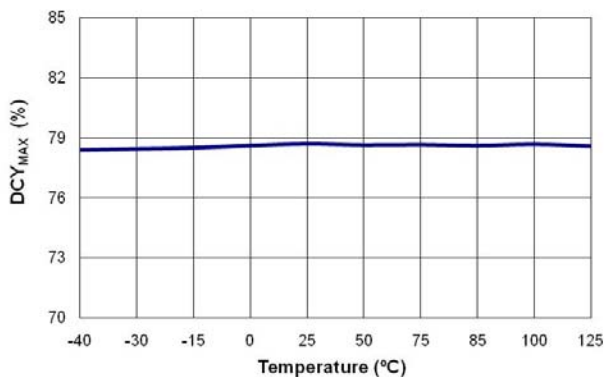


Figure 18. Maximum Duty Cycle (DCY_{MAX}) vs. Temperature

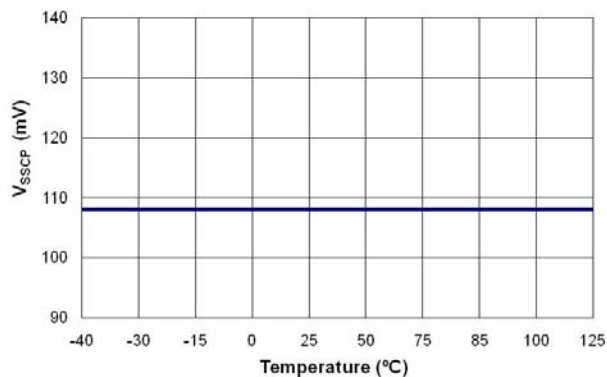


Figure 19. Threshold Voltage for SENSE Short-Circuit Protection (V_{SSCP}) vs. Temperature

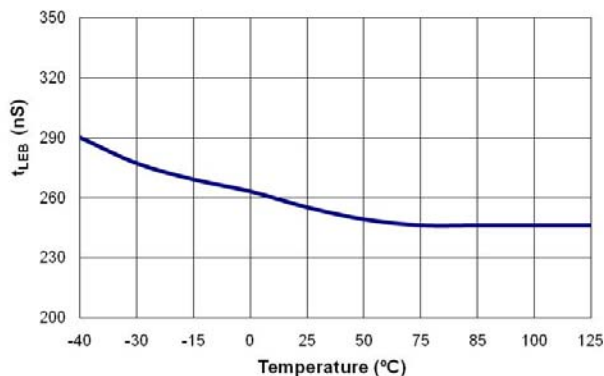


Figure 20. Leading-Edge Blanking Time (t_{LEB}) vs. Temperature

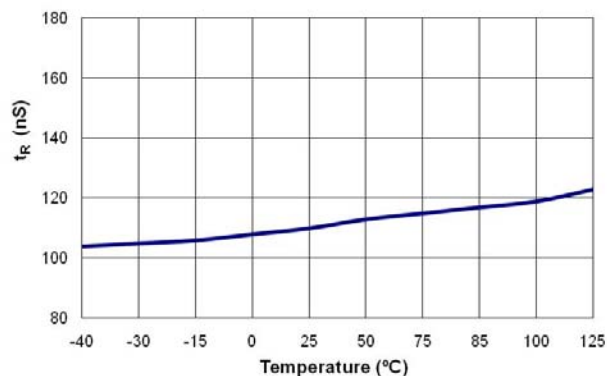


Figure 21. Rising Time (t_R) vs. Temperature

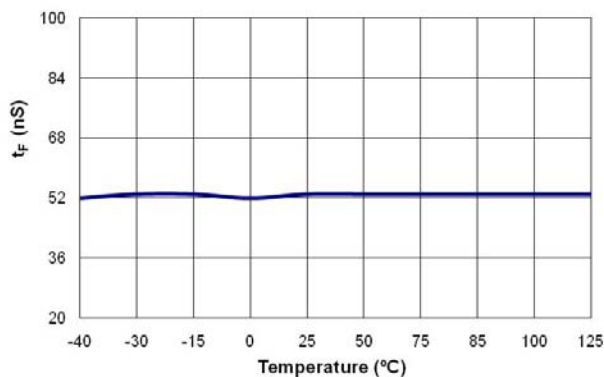


Figure 22. Falling Time (t_F) vs. Temperature

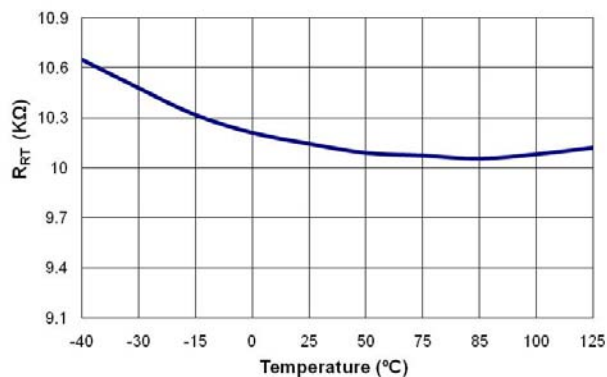


Figure 23. Maximum External Resistance of RT Pin to Trigger Protection (R_{RT}) vs. Temperature

Operation Description

Startup Operation

Figure 24 shows a typical startup circuit and transformer auxiliary winding for a typical application. Before switching operation begins, FAN6863WTY consumes only startup current (typically 8 μA) and the current supplied through the startup resistor charges the V_{DD} capacitor (C_{DD}). When V_{DD} reaches turn-on voltage of 16 V (V_{DD-ON}), switching begins and the current consumed increases to 2 mA. Power is then supplied from the transformer auxiliary winding. The large hysteresis of V_{DD} (7 V) provides more holdup time, which allows using a small capacitor for V_{DD} . The startup resistor is typically connected to AC line for a fast reset of latch protection.

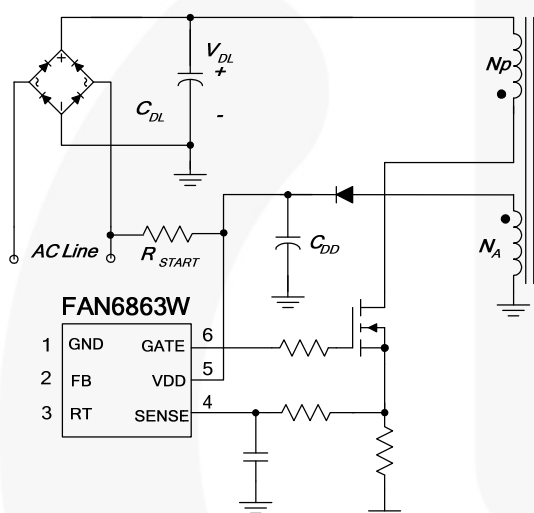


Figure 24. Startup Circuit

Green-Mode Operation

The FAN6863WTY uses feedback voltage (V_{FB}) as an indicator of the output load and modulates the PWM frequency, as shown in Figure 25, such that the switching frequency decreases as load decreases. In heavy-load conditions, the switching frequency is 65 kHz. Once V_{FB} decreases below V_{FB-N} (2.5 V), the PWM frequency starts to linearly decrease from 65 kHz to 22.5 kHz to reduce the switching losses. As V_{FB} decreases below V_{FB-G} (2.2 V), the switching frequency is fixed at 22.5 kHz and FAN6863WTY enters "deep" Green Mode, where the operating current decreases to 600 μA (maximum), further reducing the standby power consumption. As V_{FB} decreases below V_{FB-ZDC} (1.6 V), FAN6863WTY enters Burst-Mode operation. When V_{FB} drops below V_{FB-ZDC} , switching stops and the output voltage starts to drop, which causes the feedback voltage to rise. Once V_{FB} rises above V_{FB-ZDC} , switching resumes. Burst Mode alternately enables and disables switching, thereby reducing switching loss in Standby Mode, as shown in Figure 26.

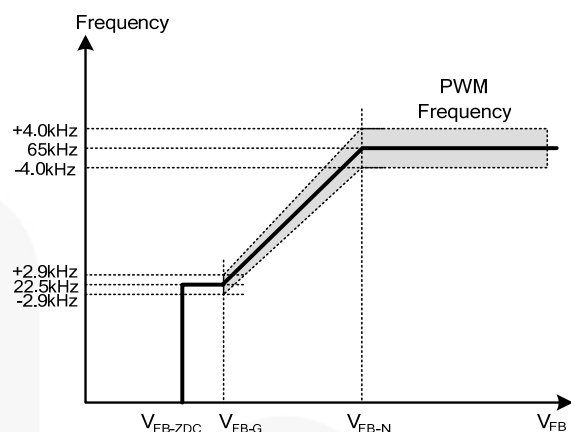


Figure 25. PWM Frequency

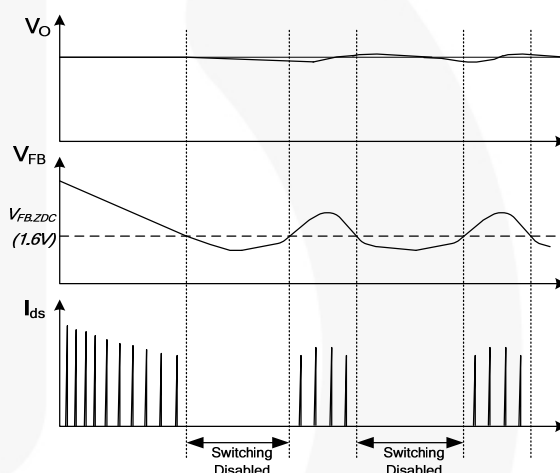


Figure 26. Burst-Mode Operation

Frequency Hopping

EMI reduction is accomplished by frequency hopping, which spreads the energy over a wider frequency range than the bandwidth measured by the EMI test equipment. An internal frequency hopping circuit changes the switching frequency between 61.0 kHz and 69.0 kHz with a period of 4.4 ms, as shown in Figure 27. It covers the whole frequency range in hopping function and shrinks the period with operation frequency proportionally.

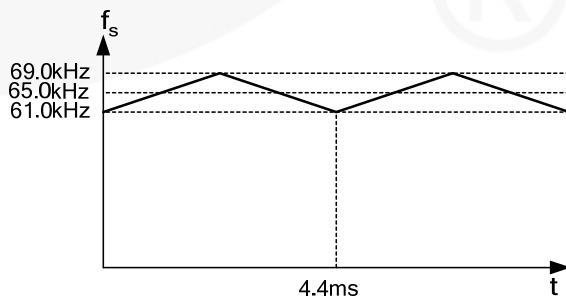


Figure 27. Frequency Hopping

Protections

Self-protective functions include V_{DD} Over-Voltage Protection (OVP), Open-Loop / Overload Protection (OLP), Over-Current Protection (OCP), Short-Circuit Protection (SCP), SENSE pin Short-Circuit Protection (SSCP), and Over-Temperature Protection (OTP). OLP, OCP, SCP, and SSCP are Auto-Restart Mode protections; OVP and OTP are Latch-Mode protections.

Auto-Restart Mode Protection

Once a fault condition is detected, switching is terminated and the MOSFET remains off. This causes V_{DD} to fall because no more power is delivered from auxiliary winding. When V_{DD} falls to V_{DD-OFF} (7 V), the protection is reset and the operating current reduces to startup current, which causes V_{DD} to rise. FAN6863WTY resumes normal operation when V_{DD} reaches V_{DD-ON} (16 V). In this manner, the auto-restart can alternately enable and disable the switching of the MOSFET until the fault condition is eliminated (see Figure 28).

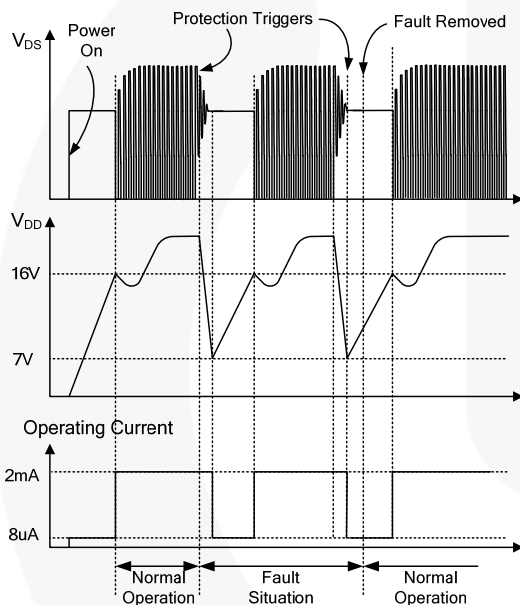


Figure 28. Auto Restart Operation

Latch-Mode Protection

Once this protection is triggered, switching is terminated and the MOSFET remains off. The latch is reset only when V_{DD} is discharged below 4 V by unplugging the AC power line.

Over-Current Protection (OCP)

The FAN6863WTY over-current protection threshold is a pulse-by-pulse current limit (V_{LIMIT}), which turns off MOSFET for the remainder of the switching cycle when the sensing voltage of MOSFET drain current reaches the threshold. The V_{LIMIT} compensates the power limit variation over universal input range and adaptively keeps the power limit substantially constant.

Open-Loop / Overload Protection (OLP)

When the upper branch of the voltage divider for the shunt regulator (KA431 shown in Figure 29) is broken, no current flows through the photo-coupler transistor, which pulls up the feedback voltage to 5.4 V.

When feedback voltage is above 4.6 V for longer than 60 ms, OLP is triggered. This protection is also triggered when the SMPS output drops below the nominal value for longer than 60 ms due to the overload condition.

If the secondary output-short situation occurs when the feedback voltage is above 4.6 V, protection time is 7 ms for shorter debounce time.

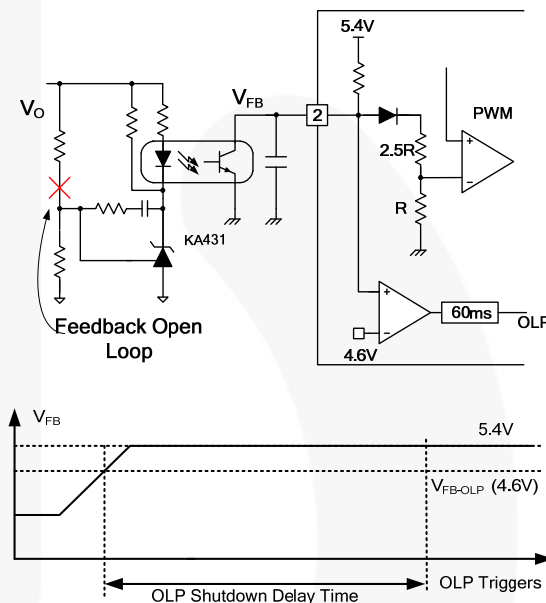


Figure 29. OLP Operation

V_{DD} Over-Voltage Protection (OVP)

V_{DD} over-voltage protection prevents IC damage caused by over voltage on the V_{DD} pin. The OVP is triggered when V_{DD} reaches 22.2 V. A debounce time (typically 50 μ s) prevents false triggering by switching noise.

Over-Temperature Protection (OTP)

The OTP circuit is composed of current source and voltage comparators. Typically, an NTC thermistor is connected between the RT and GND pins. If the voltage of this pin drops below a threshold of 1.0 V, PWM output is disabled after t_{DOTP} debounce time. If this pin voltage drops below 0.7 V, it triggers the latch-off protection immediately after t_{DOTP2} debounce time.

Typical Application Circuit (Netbook Adapter by Flyback)

| Application | Fairchild Devices | Input Voltage Range | Output |
|-----------------|-------------------|------------------------|------------------|
| Netbook Adapter | FAN6863WTY | 90~265 V _{AC} | 19 V/2.1 A (40W) |

Features

- High efficiency (>85.3% at full-load condition) meeting EPS regulation with enough margin
- Low standby (pin<0.1 W at no-load condition)
- Soft-start time: 5 ms

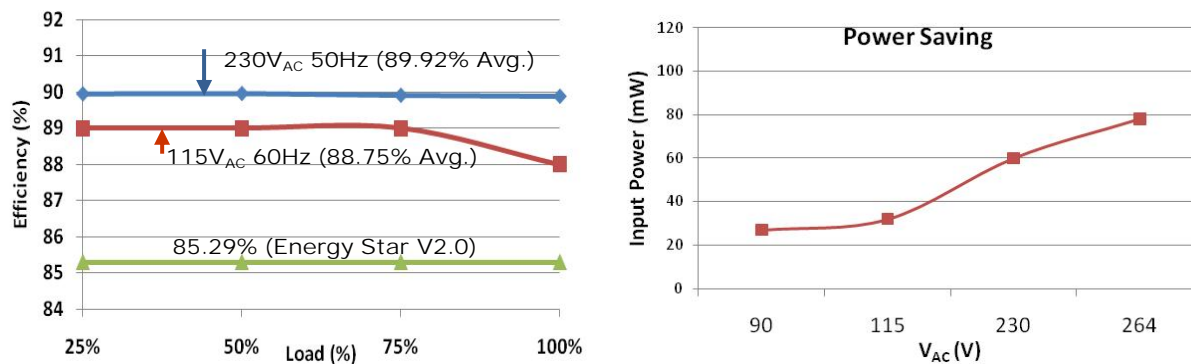


Figure 30. Measured Efficiency and Power Saving

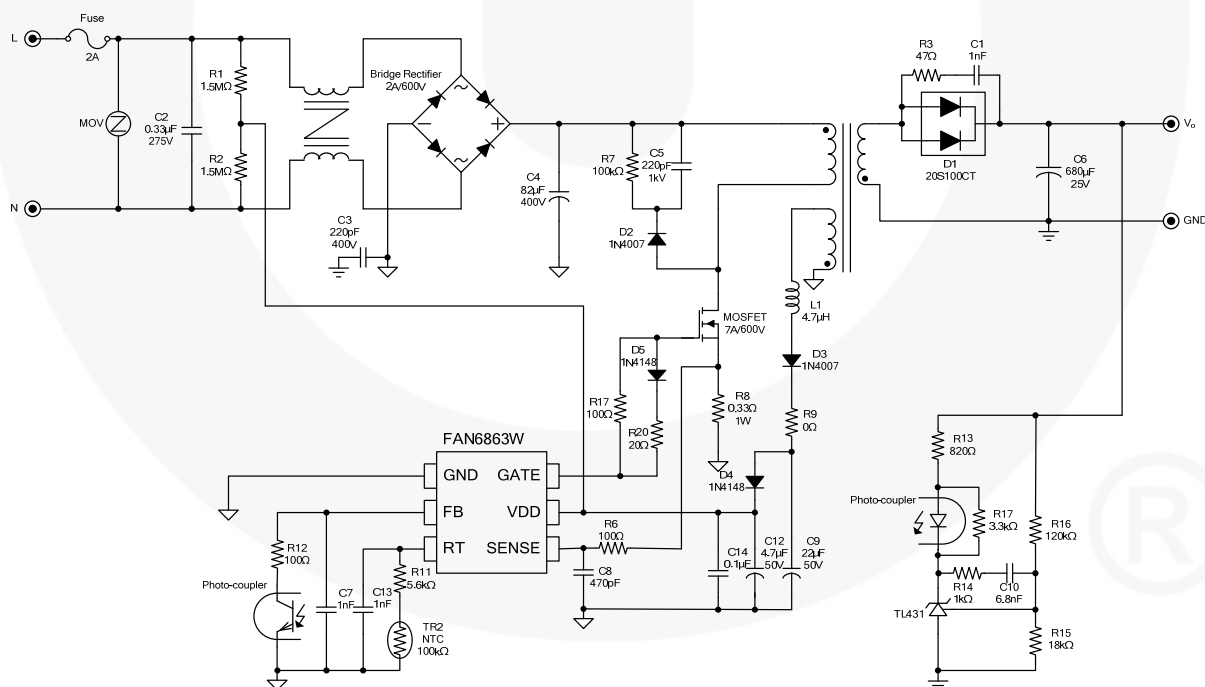


Figure 31. Schematic of Typical Application Circuit

Transformer Specification

- Core: RM 8
- Bobbin: RM 8

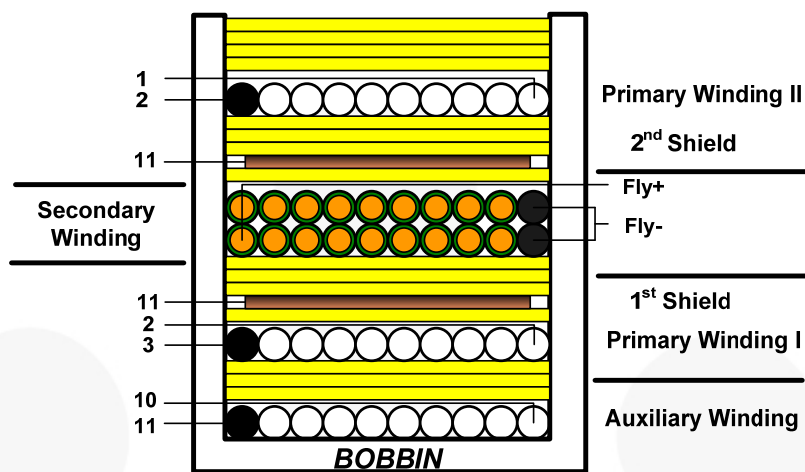


Figure 32. Transformer

| NO | Terminal | | Wire | Ts | Insulation | Barrier | |
|----|----------|------|--------------------|-----|------------|---------|-----------|
| | S | F | | | Ts | Primary | Secondary |
| N1 | 11 | 10 | 0.25 • 1 | 9 | 3 | | |
| N2 | 3 | 2 | 0.25 • 1 | 33 | 1 | | |
| | 11 | | COPPER SHIELD | 1.2 | 3 | | |
| N3 | Fly- | Fly+ | 0.5 • 2 | 12 | 1 | | |
| | 11 | | COPPER SHIELD | 1.2 | 3 | | |
| N4 | 2 | 1 | 0.25 • 1 | 33 | 4 | | |
| | | | CORE ROUNDING TAPE | | 3 | | |

| | Pin | Specification | Remark |
|--------------------------------|-----|----------------------|-------------------------------------|
| Primary-Side Inductance | 3—1 | 920 μ H \pm 5% | 100 kHz, 1 V |
| Primary-Side Effective Leakage | 3—1 | 15 μ H Maximum | Short One of the Secondary Windings |

Physical Dimensions

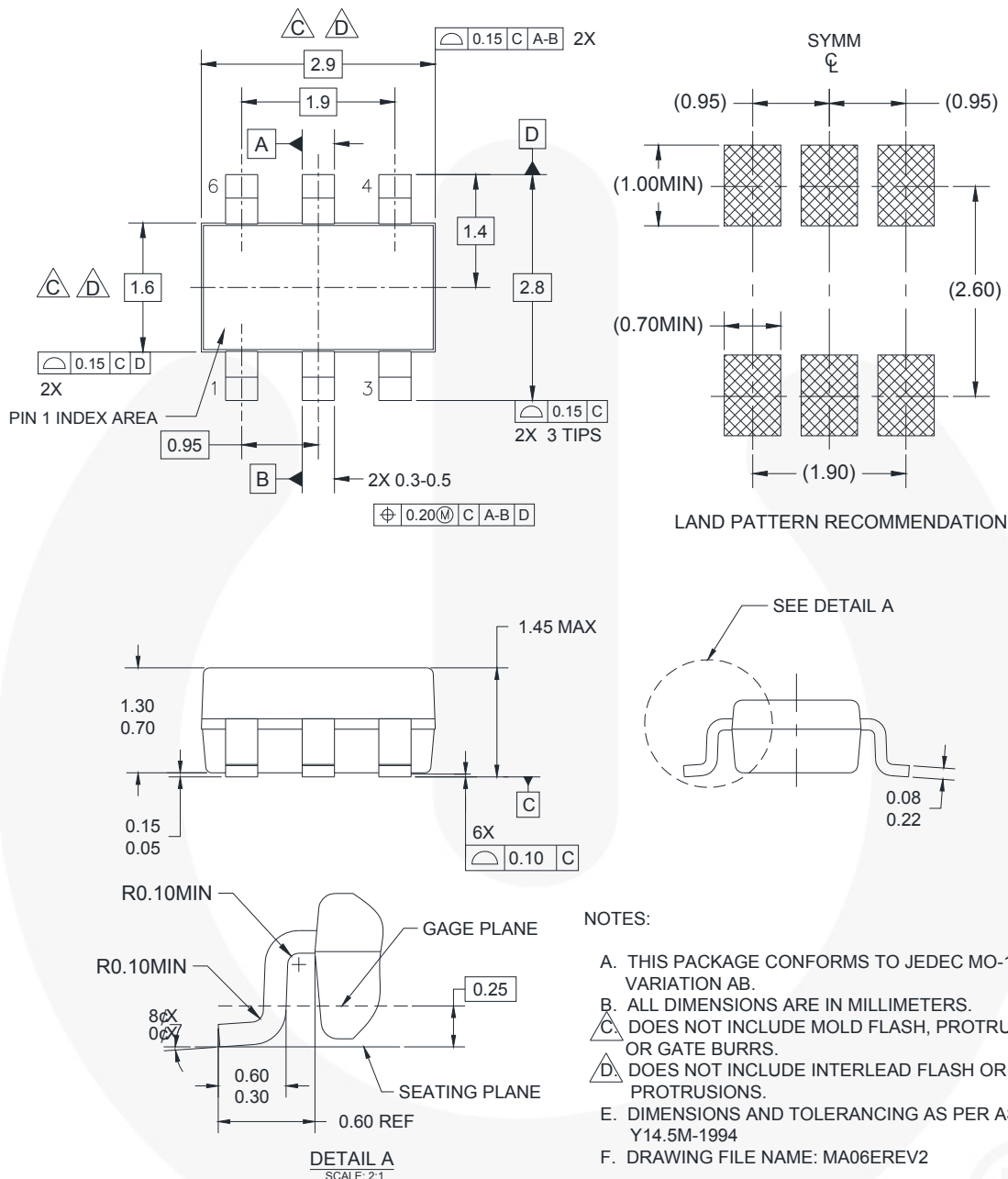


Figure 33. 6-Lead, SuperSOT™-6 JEDEC, M0-193 1.6mm Wide




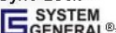
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