

SiC JFET Division

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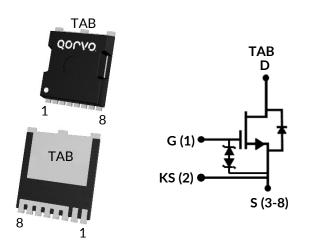








UJ4C075033L8S



Part Number	Package	Marking
UJ4C075033L8S	MO-229	UJ4C075033







Silicon Carbide (SiC) Cascode JFET -EliteSiC, Power N-Channel, TOLL, 750 V, 33 mohm

Rev. D, January 2025

Description

The UJ4C075033L8S is a 750V, $33m\Omega$ G4 SiC FET. It is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows use of off-the-shelf gate drivers hence requiring minimal re-design when replacing Si IGBTs, Si superjunction devices or SiC MOSFETs. Available in the space-saving MO-229 package which enables automated assembly, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

Features

- On-resistance R_{DS(on)}: 33mΩ (typ)
- Operating temperature: 175°C (max)
- Excellent reverse recovery: Q_{rr} = 89nC
- Low body diode V_{FSD}: 1.26V
- Low gate charge: Q_G = 37.8nC
- Threshold voltage V_{G(th)}: 4.8V (typ) allowing 0 to 15V drive
- Low intrinsic capacitance
- ESD protected: HBM class 2 and CDM class C3
- MO-229 package for faster switching, clean gate waveforms

Typical applications

- Line rectification and active-bridge rectification circuits in AC/DC front-ends
- EV charging
- PV inverters
- Switch mode power supplies
- Power factor correction modules
- Motor drives
- Induction heating













Maximum Ratings

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	V_{DS}		750	V
Gate-source voltage	V	DC	-20 to +20	V
Gate-source voltage	V_{GS}	AC (f > 1Hz)	-25 to +25	V
Continuous drain current ¹		T _C = 25°C	44	Α
Continuous drain current	I _D	T _C =100°C	33	Α
Pulsed drain current ²	I _{DM}	T _C = 25°C	132	Α
Single pulsed avalanche energy ³	E _{AS}	L=15mH, I _{AS} =2.4A	43	mJ
SiC FET dv/dt ruggedness	dv/dt	V _{DS} ≤ 500V	200	V/ns
Power dissipation	P _{tot}	T _C = 25°C	205	W
Maximum junction temperature	$T_{J,max}$		175	°C
Operating and storage temperature	T _J , T _{STG}		-55 to 175	°C
Reflow soldering temperature	T _{solder}	reflow MSL 1	260	°C

- 1. Limited by $T_{J,max}$
- 2. Pulse width $t_{\rm p}$ limited by $T_{\rm J,max}$
- 3. Starting $T_J = 25$ °C

Thermal Characteristics

Darameter	Symbol	Test Conditions		Value		Units
Parameter	Symbol		Min	Тур	Max	Onits
Thermal resistance, junction-to-case	$R_{\theta JC}$			0.56	0.73	°C/W













Electrical Characteristics (T_J = +25°C unless otherwise specified)

Typical Performance - Static

Parameter	Symbol	Test Conditions		Value			
	Syllibol	Test Collutions	Min	Тур	Max	Units	
Drain-source breakdown voltage	BV _{DS}	V_{GS} =0V, I_D =1mA	750			V	
		V _{DS} =750V,		2	20		
Total drain leakage current	ı	$V_{GS}=0V, T_J=25$ °C			20	^	
Total drain leakage current	I _{DSS}	V _{DS} =750V,		20		μА	
		V _{GS} =0V, T _J =175°C		20			
Tatal aska laslassa suumust		V _{DS} =0V, T _J =25°C,	,	, , , , ,	6 ±20	μА	
Total gate leakage current	I _{GSS}	V _{GS} =-20V / +20V		6			
		V_{GS} =12V, I_{D} =30A,		22	44		
		T _J =25°C		33	41		
Drain-source on-resistance	D	V _{GS} =12V, I _D =30A,		E-7		m()	
Drain-source on-resistance	R _{DS(on)}	T _J =125°C		57		mΩ	
		V _{GS} =12V, I _D =30A,		75		1	
		T _J =175°C		75			
Gate threshold voltage	$V_{G(th)}$	V_{DS} =5V, I_D =10mA	4	4.8	6	V	
Gate resistance	R_G	f=1MHz, open drain		4.5		Ω	

Typical Performance - Reverse Diode

Parameter	Symbol	Test Conditions	Value			Units
Parameter	Syllibol	Test Conditions	Min	Тур	Max	UTILS
Diode continuous forward current ¹	I _S	T _C = 25°C			44	Α
Diode pulse current ²	$I_{S,pulse}$	T _C = 25°C			132	Α
Command vallence	V	V _{GS} =0V, I _S =15A, T _J =25°C		1.26	1.42	V
Forward voltage	V _{FSD}	V _{GS} =0V, I _S =15A, T _J =175°C		1.59		V
Reverse recovery charge	Q _{rr}	V_R =400V, I_S =30A, V_{GS} =0V, R_{G_EXT} =50 Ω		89		nC
Reverse recovery time	t _{rr}	di/dt=1100A/μs, Τ _J =25°C		20.8		ns
Reverse recovery charge	Q _{rr}	V_R =400V, I_S =30A, V_{GS} =0V, R_{G_EXT} =50 Ω		97		nC
Reverse recovery time	t _{rr}	di/dt=1100A/μs, Τ _J =150°C		21.6		ns













Typical Performance - Dynamic

5		T + C - I''		Value		l	
Parameter	Symbol	Test Conditions	Min	Тур	Max	- Units	
Input capacitance	C _{iss}	V _{DS} =400V, V _{GS} =0V		1400			
Output capacitance	C _{oss}	f=100kHz		68		pF	
Reverse transfer capacitance	C _{rss}	I-100KHZ		2.5		1	
Effective output capacitance, energy related	C _{oss(er)}	V_{DS} =0V to 400V, V_{GS} =0V		83		pF	
Effective output capacitance, time related	C _{oss(tr)}	V _{DS} =0V to 400V, V _{GS} =0V		162		pF	
C _{OSS} stored energy	E _{oss}	V _{DS} =400V, V _{GS} =0V		6.6		μJ	
Total gate charge	Q_{G}	V _{DS} =400V, I _D =30A,		37.8			
Gate-drain charge	Q_{GD}	$V_{DS} = 400V, I_D = 50A,$ $V_{GS} = 0V \text{ to } 15V$		8		nC	
Gate-source charge	Q_{GS}			11.8			
Turn-on delay time	t _{d(on)}	Notes 4,		11			
Rise time	t _r	$V_{DS}\text{=}400V, I_{D}\text{=}30A, Gate}$ $Driver \text{=}0V \text{ to } +15V,$ $Turn\text{-}on R_{G,EXT}\text{=}1\Omega,$ $Turn\text{-}off R_{G,EXT}\text{=}50\Omega,$		21			
Turn-off delay time	t _{d(off)}			126		- ns	
Fall time	t _f			12			
Turn-on energy	E _{ON}	inductive Load, FWD: same device with		176			
Turn-off energy	E _{OFF}	$V_{GS} = OV$ and $R_G = 50\Omega$,		79		μЈ	
Total switching energy	E _{TOTAL}	T _J =25°C		255			
Turn-on delay time	t _{d(on)}	Notes 4,		11			
Rise time	t _r	V _{DS} =400V, I _D =30A, Gate		22		- ns	
Turn-off delay time	$t_{d(off)}$	Driver =0V to +15V, Turn-on $R_{G,EXT}$ =1 Ω , Turn-off $R_{G,EXT}$ =50 Ω ,		137		115	
Fall time	t _f			14			
Turn-on energy	E _{ON}	inductive Load, FWD: same device with		189			
Turn-off energy	E _{OFF}	$V_{GS} = 0V$ and $R_G = 50\Omega$,		86		μJ	
Total switching energy	E _{TOTAL}	T _J =150°C		275			

^{4.} Measured with the switching test circuit in Figure 23.













Typical Performance - Dynamic (continued)

Parameter	Symbol	Test Conditions		Value		Units
Parameter	Symbol	rest Conditions	Min	Тур	Max	Units
Turn-on delay time	t _{d(on)}	Notes 5 and 6,		12		
Rise time	t _r	V _{DS} =400V, I _D =30A, Gate		22		
Turn-off delay time	t _{d(off)}	Driver =0V to +15V, Turn-on $R_{G,EXT}=1\Omega$,		42		ns
Fall time	t _f	Turn-off $R_{G,EXT} = 5\Omega$,		6		
Turn-on energy including R _S energy	E _{ON}	inductive Load,		178		
Turn-off energy including R _S energy	E _{OFF}	FWD: same device with		30		
Total switching energy	E _{TOTAL}	V_{GS} = 0V and R_{G} = 5Ω , RC snubber: R_{S} =10 Ω and C_{S} =100pF, T_{J} =25°C		208		μЈ
Snubber R _S energy during turn-on	E _{RS_ON}			0.95		
Snubber R _S energy during turn-off	E _{RS_OFF}			1.43		
Turn-on delay time	t _{d(on)}	Notes 5 and 6,		11		
Rise time	t _r	V_{DS} =400V, I_{D} =30A, Gate		22		ns
Turn-off delay time	$t_{d(off)}$	Driver =0V to +15V, $Turn-on R_{G,EXT}=1\Omega,$		42		115
Fall time	t _f	Turn-off $R_{G,EXT} = 5\Omega$,		7		
Turn-on energy including R _S energy	E _{ON}	inductive Load,		185		
Turn-off energy including R _S energy	E _{OFF}	FWD: same device with V_{GS} = 0V and R_{G} = 5 Ω , RC snubber: R_{S} =10 Ω and C_{S} =100pF,		31		
Total switching energy	E _{TOTAL}			216		μЈ
Snubber R _S energy during turn-on	E _{RS_ON}			1		
Snubber R _S energy during turn-off	E _{RS_OFF}	T _J =150°C		1.41		

^{5.} Measured with the switching test circuit in Figure 24.

^{6.} In this table, the switching energies (turn-on energy, turn-off energy and total energy) presented include the device RC snubber energy losses.





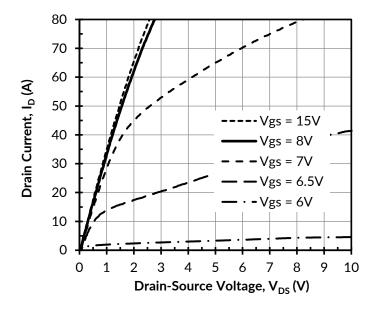








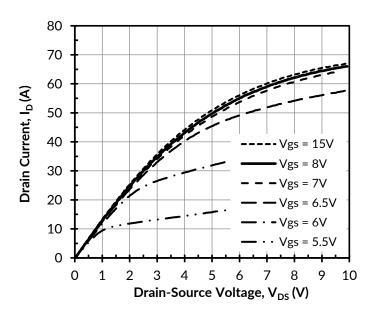
Typical Performance Diagrams



80 70 60 Drain Current, I_D (A) 50 40 Vgs = 15V 30 Vgs = 8V Vgs = 7V20 - Vgs = 6.5V 10 Vgs = 6V 0 0 2 8 9 10 1 Drain-Source Voltage, V_{DS} (V)

Figure 1. Typical output characteristics at T_J = - 55°C, tp < 250 μ s

Figure 2. Typical output characteristics at T_J = 25°C, tp < 250 μ s



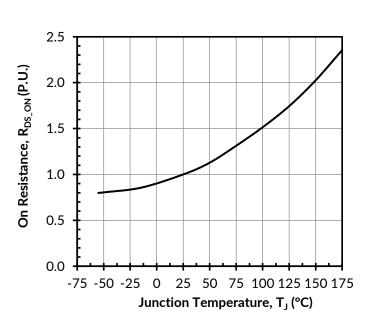


Figure 3. Typical output characteristics at T_J = 175°C, tp < 250 μs

Figure 4. Normalized on-resistance vs. temperature at $V_{\text{GS}} = 12V$



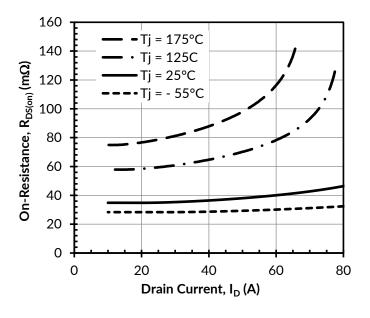








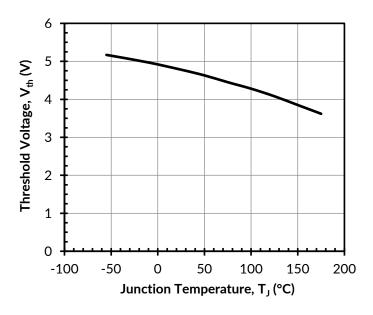




60 Tj = -55°C 50 Tj = 25°C Tj = 175°C Drain Current, I_D (A) 40 30 20 10 0 0 1 2 3 4 5 6 9 10 Gate-Source Voltage, V_{GS} (V)

Figure 5. Typical drain-source on-resistances at V_{GS} = 12V

Figure 6. Typical transfer characteristics at $V_{DS} = 5V$



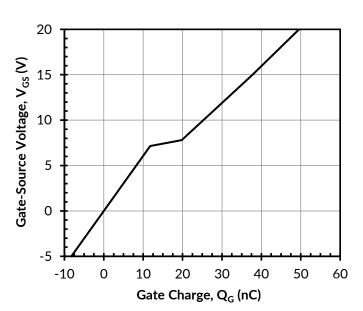


Figure 7. Threshold voltage vs. junction temperature at Figure 8. Typical gate charge at I_D = 30A V_{DS} = 5V and I_{D} = 10mA













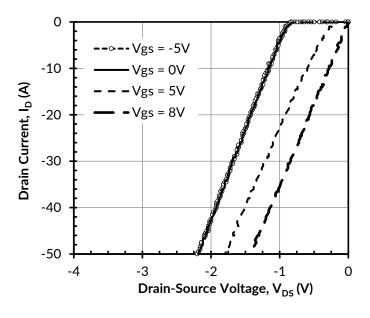


Figure 9. 3rd quadrant characteristics at $T_J = -55$ °C

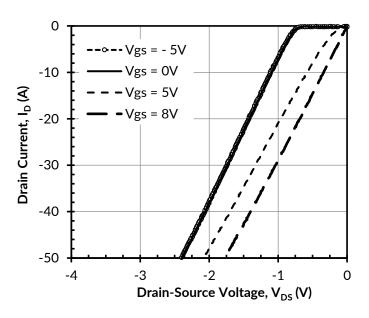


Figure 10. 3rd quadrant characteristics at $T_J = 25$ °C

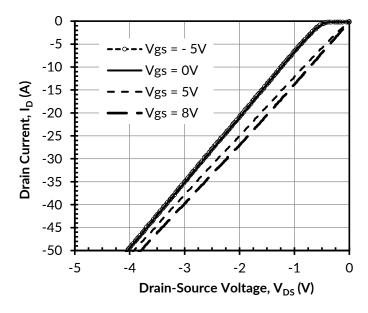


Figure 11. 3rd quadrant characteristics at $T_J = 175$ °C

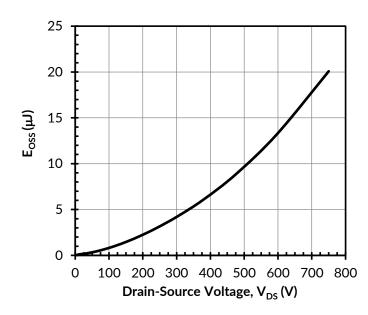


Figure 12. Typical stored energy in C_{OSS} at V_{GS} = 0V



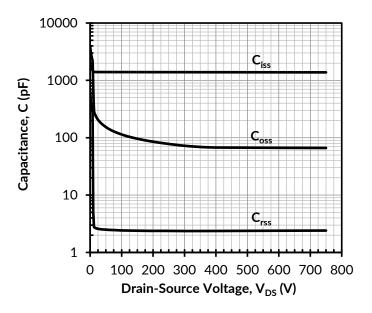








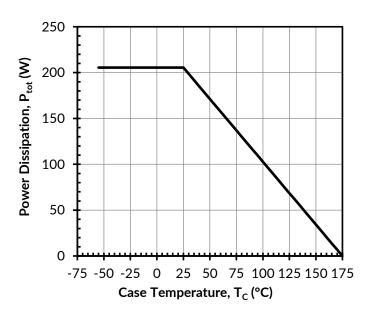




50 40 -75 -50 -25 0 25 50 75 100 125 150 175 Case Temperature, T_c (°C)

Figure 13. Typical capacitances at f = 100kHz and V_{GS} = 0V

Figure 14. DC drain current derating



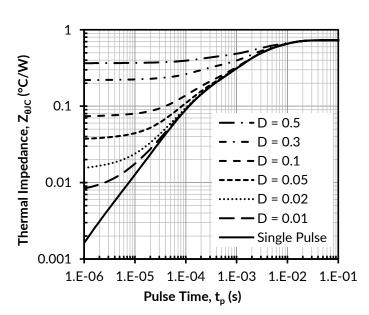


Figure 15. Total power dissipation

Figure 16. Maximum transient thermal impedance













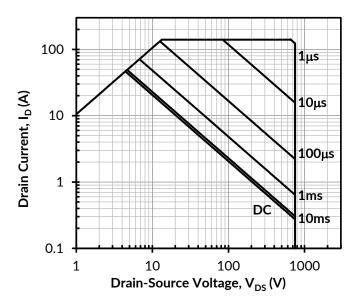


Figure 17. Safe operation area at T_C = 25°C, D = 0, Parameter t_p

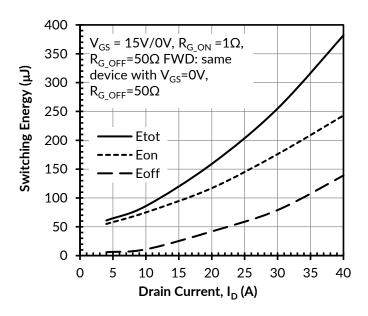


Figure 19. Clamped inductive switching energy $\,$ vs. drain current at V_{DS} = 400V and T_J = 25°C

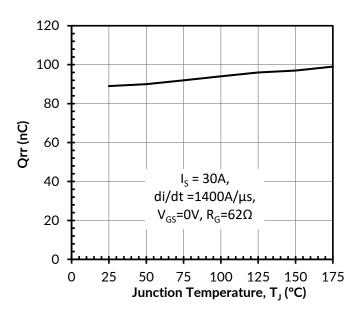


Figure 18. Reverse recovery charge Qrr vs. junction temperature at V_{DS} = 400V

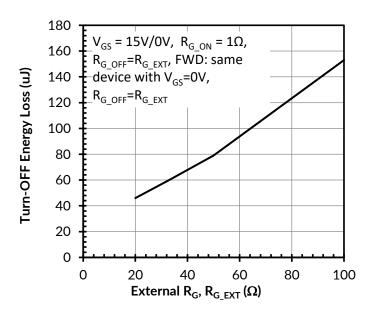


Figure 20. Turn-OFF clamped inductive switching energies vs. $R_{G,EXT}$ at V_{DS} = 400V, I_{D} =30A, and T_{J} = 25°C













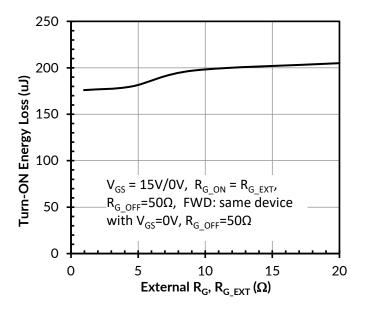


Figure 21. Turn-ON Clamped inductive switching energies vs. $R_{G,EXT}$ at V_{DS} = 400V, I_D = 30A, and T_J = 25°C

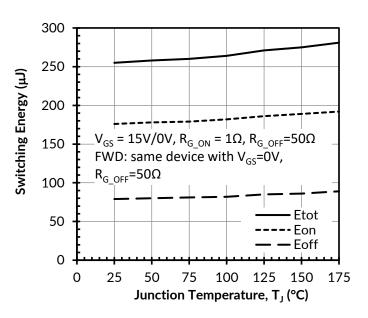


Figure 22. Clamped inductive switching energy vs. junction temperature at V_{DS} = 400V and I_{D} = 30A

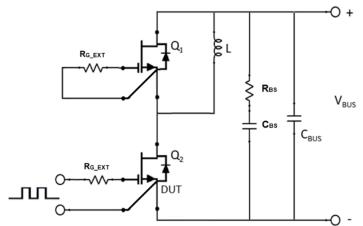


Figure 23. Schematic of the half-bridge mode switching test circuit. Note, a bus RC snubber (R_{BS} = 2.5 Ω , C_{BS} =100nF) is used to reduce the power loop high frequency oscillations.

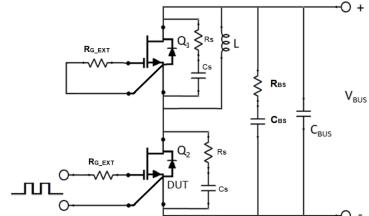


Figure 24. Schematic of the half-bridge mode switching test circuit with device RC snubbers (R $_{s}$ =10 Ω , C $_{s}$ = 100pF) and a bus RC snubber (R $_{BS}$ = 2.5 Ω , C $_{BS}$ =100nF).





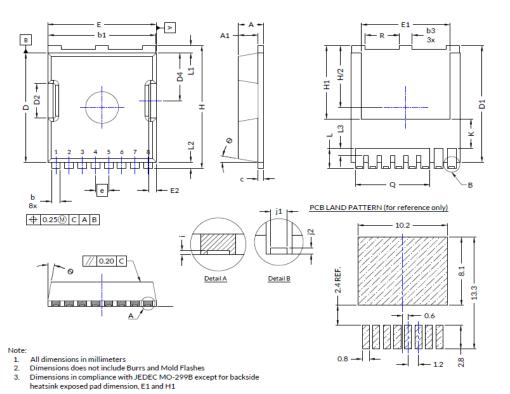








Package Outlines



TO-IL				
O/A #DOL		Value		
SYMBOL	Min	Nom	Max	
Α	2.15	2.30	2.45	
A1		1.80 REF		
ь	0.70	0.80	0.90	
b1	9.65	9.80	9.95	
b3	1.10	1.20	1.30	
c	0.40	0.50	0.60	
D	10.18	10.38	10.58	
D1	10.98	11.08	11.18	
D2	3.15	3.30	3.45	
D4	4.40	4.55	4.70	
Е	9.70	9.90	10.10	
E1	7.95	8.10	8.25	
E2	0.60	0.70	0.80	
e		1.20 BSC		
Н	11.48	11.68	11.88	
H1	6.80	6.95	7.10	
i		0.10 REF		
j1		0.46 REF		
j2		0.20 REF		
K		2.80 REF		
L	1.40	1.90	2.10	
L1	0.50	0.70	0.90	
L2	0.48	0.60	0.72	
L3	0.30	0.70	0.80	
Q		6.80 REF		
R	3.00 3.10 3.20			
θ		10°		

Pin Designations:

1:Gate

2 : Source Kelvii 3-8 : Source

Applications Information

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Q_{rr}) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see https://www.qorvo.com/design-hub.

A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the UnitedSiC website at https://www.qorvo.com/design-hub.













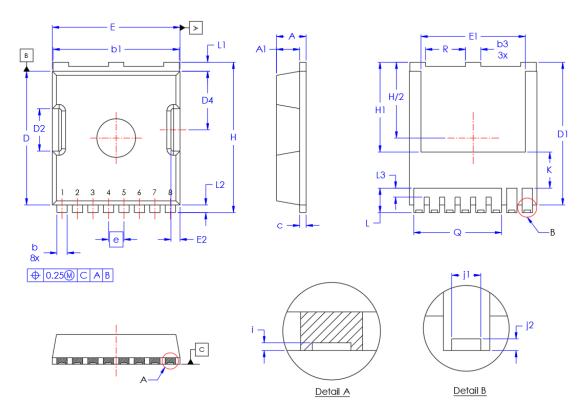
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DS TOLL	Rev B

PACKAGE OUTLINE



	TO-LL			
SYMBOL	Va			
01111002	Min	Max		
A	2.15	2.45		
A1	1.80	REF		
b	0.65	0.90		
b1	9.65	9.95		
b3	1.10	1.30		
С	0.40	0.60		
D	10.18	10.58		
DI	10.88	11.28		
D2	3.15	3.45		
D4	4.40	4.70		
E	9.70	10.10		
El	7.95	8.25		
E2	0.60	0.80		
е	1.20	BSC		
Н	11.48	11.88		
Hl	6.80	7.10		
i	0.10	REF		
jl	0.46	REF		
j2	0.20	REF		
K	2.80	REF		
L	1.40	2.10		
L1	0.50	0.90		
L2	0.48	0.72		
L3	0.30	0.80		
Q	6.80	REF		
R	3.00	3.20		

Note:

- 1. All dimensions in millimeters
- 2. Dimensions does not include Burrs and Mold Flashes

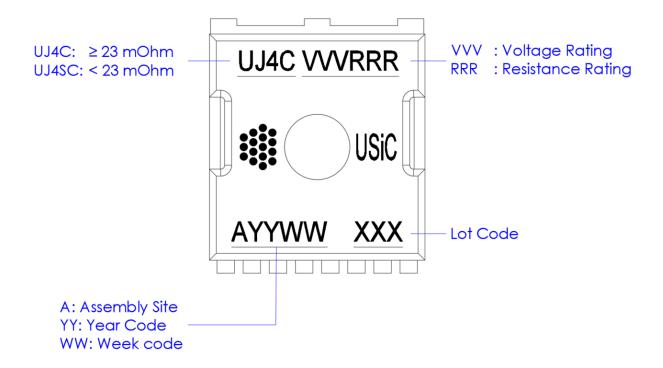
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PART MARKING



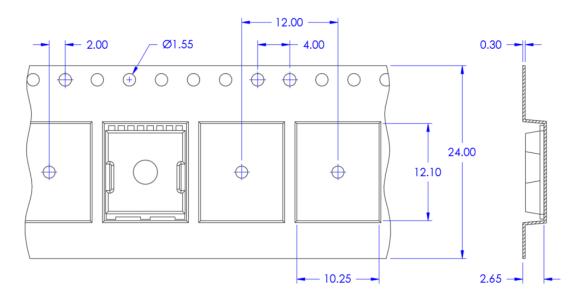
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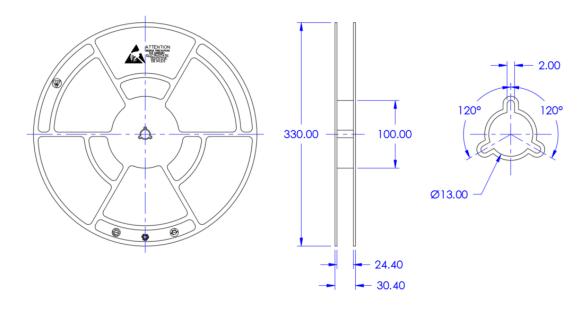
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DS_TOLL	Rev B

PACKING TYPE

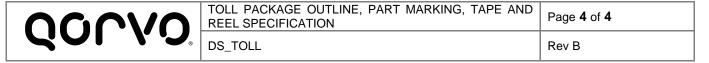
Carrier Tape



Reel



All dimensions in millimeters Quantity per Reel: 2000 units



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REVISION HISTORY

Revision	Create Date (mm/dd/yyyy)	Description of Change	Initiator of Change
А	10/13/2023	Initial Production Release	Glenn Galang
В	01/31/2024	Corrected device orientation inside carrier tape pocket (Page 3)	Glenn Galang

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