Power MOSFET 2 Amps, 25 Volts

N-Channel SO-8, Dual

These miniature surface mount MOSFETs feature ultra low $R_{DS(on)}$ and true logic level performance. They are capable of withstanding high energy in the avalanche and commutation modes and the drain–to–source diode has a low reverse recovery time. These devices are designed for use in low voltage, high speed switching applications where power efficiency is important. Typical applications are dc–dc converters, and power management in portable and battery powered products such as computers, printers, cellular and cordless phones. They can also be used for low voltage motor controls in mass storage products such as disk drives and tape drives. The avalanche energy is specified to eliminate the guesswork in designs where inductive loads are switched and offer additional safety margin against unexpected voltage transients.

Features

- Ultra Low R_{DS(on)} Provides Higher Efficiency and Extends Battery Life
- Logic Level Gate Drive Can Be Driven by Logic ICs
- Miniature SO-8 Surface Mount Package Saves Board Space
- Diode Is Characterized for Use In Bridge Circuits
- I_{DSS} Specified at Elevated Temperatures
- Avalanche Energy Specified
- Mounting Information for SO-8 Package Provided
- This is a Pb-Free Device

MAXIMUM RATINGS (T_J = 25°C unless otherwise noted)

Rating	Symbol	Value	Unit
Drain-to-Source Voltage	V _{DSS}	25	Vdc
Gate-to-Source Voltage - Continuous	V _{GS}	± 20	Vdc
	I _D I _D I _{DM}	3.6 2.5 18	Adc Apk
Total Power Dissipation @ T _A = 25°C (Note 1)	P _D	2.0	W
Operating and Storage Temperature Range	T _J , T _{stg}	-55 to 150	°C
Single Pulse Drain–to–Source Avalanche Energy – Starting T_J = 25°C (V_{DD} = 20 Vdc, V_{GS} = 10 Vdc, Peak I_L = 9.0 Apk, L = 6.0 mH, R_G = 25 Ω)	E _{AS}	245	mJ
Thermal Resistance, Junction-to-Ambient (Note 1)	$R_{\theta JA}$	62.5	°C/W
Maximum Lead Temperature for Soldering Purposes, 0.0625" from case for 10 seconds	TL	260	°C

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

1. Mounted on 2" square FR4 board (1" sq. 2 oz. Cu 0.06" thick single sided) with one die operating, 10 sec. max.

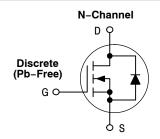
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2 AMPERES, 25 VOLTS $R_{DS(on)} = 100 \text{ m}\Omega$



MARKING DIAGRAM



SO-8 CASE 751 STYLE 11

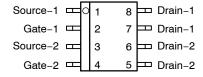


F2N02 = Device Code A = Assembly Location

Y = Year
WW = Work Week
Pb-Free Package

(Note: Microdot may be in either location)

PIN ASSIGNMENT



ORDERING INFORMATION

Device	Package	Shipping [†]
MMDF2N02ER2G	SO-8 (Pb-Free)	2500 Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

ELECTRICAL CHARACTERISTICS ($T_A = 25^{\circ}C$ unless otherwise noted)

Characteristic	Symbol	Min	Тур	Max	Unit	
OFF CHARACTERISTICS		<u> </u>	I		1	
Drain-to-Source Breakdown Voltage (V_{GS} = 0 Vdc, I_D = 250 μ Adc)	V _{(BR)DSS}	25	-	-	Vdc	
Zero Gate Voltage Drain Current $(V_{DS} = 20 \text{ Vdc}, V_{GS} = 0 \text{ Vdc})$ $(V_{DS} = 20 \text{ Vdc}, V_{GS} = 0 \text{ Vdc}, T_{J} =$	125°C)	I _{DSS}	<u>-</u>	- -	1.0 10	μAdc
Gate-Body Leakage Current (V _{GS} = ±	20 Vdc, V _{DS} = 0)	I _{GSS}	-	-	100	nAdc
ON CHARACTERISTICS (Note 2)		•				
Gate Threshold Voltage (V _{DS} = V _{GS} , I _D = 250 μAdc		V _{GS(th)}	1.0	2.0	3.0	Vdc
Static Drain-to-Source On-Resistance ($V_{GS} = 10 \text{ Vdc}$, $I_D = 2.2 \text{ Adc}$) ($V_{GS} = 4.5 \text{ Vdc}$, $I_D = 1.0 \text{ Adc}$)	R _{DS(on)}	<u>-</u>	0.083 0.110	0.100 0.200	Ω	
Forward Transconductance (V _{DS} = 3.0) Vdc, I _D = 1.0 Adc)	9FS	1.0	2.6	-	Mhos
DYNAMIC CHARACTERISTICS		•				
Input Capacitance		C _{iss}	-	380	532	pF
Output Capacitance	$(V_{DS} = 16 \text{ Vdc}, V_{GS} = 0 \text{ Vdc}, f = 1.0 \text{ MHz})$	C _{oss}	-	235	329	
Transfer Capacitance	,,	C _{rss}	-	55	110	
SWITCHING CHARACTERISTICS (N	ote 3)	•				
Turn-On Delay Time		t _{d(on)}	-	7.0	21	ns
Rise Time	(V _{DD} = 10 Vdc, I _D = 2.0 Adc,	t _r	-	17	30	
Turn-Off Delay Time	$V_{GS} = 10 \text{ Vdc}, R_G = 6.0 \Omega$	t _{d(off)}	-	27	48	
Fall Time		t _f	-	18	30	
Turn-On Delay Time		t _{d(on)}	-	10	30	
Rise Time	(V _{DD} = 10 Vdc, I _D = 2.0 Adc,	t _r	-	35	70	
Turn-Off Delay Time	$V_{GS} = 4.5 \text{ Vdc}, R_G = 9.1 \Omega$	t _{d(off)}	-	19	38	
Fall Time		t _f	-	25	50	
Gate Charge		Q _T	-	10.6	30	nC
	(V _{DS} = 16 Vdc, I _D = 2.0 Adc,	Q ₁	-	1.3	_	
	V _{GS} = 10 Vdc)	Q_2	-	2.9	_	
		Q_3	_	2.7	_	
SOURCE-DRAIN DIODE CHARACTI	ERISTICS	•		•	•	•
Forward On-Voltage (Note 2)	(I _S = 2.0 Adc, V _{GS} = 0 Vdc)	V _{SD}	-	1.0	1.4	Vdc
Reverse Recovery Time		t _{rr}	-	34	66	ns
See Figure 11	(I _S = 2.0 Adc, V _{GS} = 0 Vdc,	t _a	-	17 –		
	$dI_{S}/dt = 100 \text{ A}/\mu\text{s})$	t _b	-	17	-	
Reverse Recovery Storage Charge		Q _{RR}	_	0.03	_	μC

Pulse Test: Pulse Width ≤ 300 μs, Duty Cycle ≤ 2%.
 Switching characteristics are independent of operating junction temperature.

TYPICAL ELECTRICAL CHARACTERISTICS

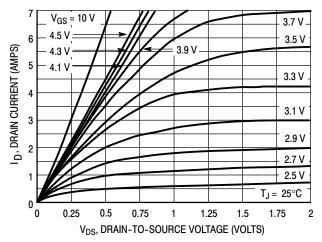


Figure 1. On-Region Characteristics

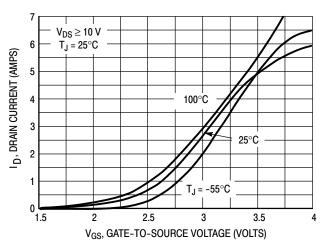


Figure 2. Transfer Characteristics

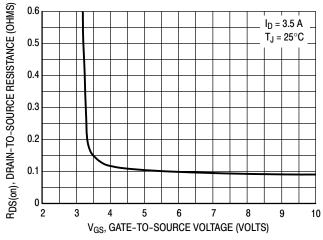


Figure 3. On-Resistance versus Gate-to-Source Voltage

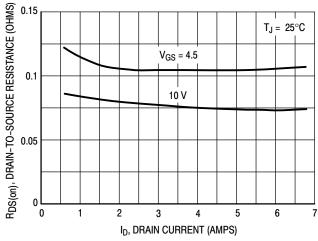


Figure 4. On-Resistance versus Drain Current and Gate Voltage

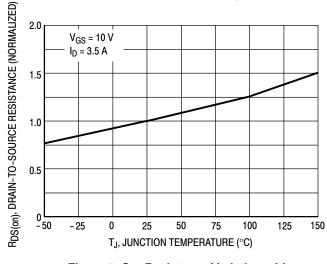


Figure 5. On–Resistance Variation with Temperature

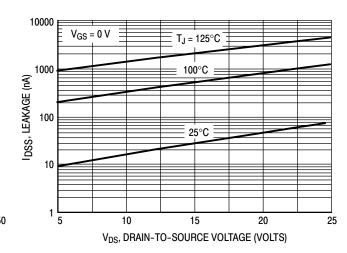


Figure 6. Drain-to-Source Leakage Current versus Voltage

POWER MOSFET SWITCHING

Switching behavior is most easily modeled and predicted by recognizing that the power MOSFET is charge controlled. The lengths of various switching intervals (Δt) are determined by how fast the FET input capacitance can be charged by current from the generator.

The published capacitance data is difficult to use for calculating rise and fall because drain–gate capacitance varies greatly with applied voltage. Accordingly, gate charge data is used. In most cases, a satisfactory estimate of average input current ($I_{G(AV)}$) can be made from a rudimentary analysis of the drive circuit so that:

$$t = Q/I_{G(AV)}$$

During the rise and fall time interval when switching a resistive load, V_{GS} remains virtually constant at a level known as the plateau voltage, V_{SGP} . Therefore, rise and fall times may be approximated by the following:

$$t_r = Q_2 \times R_G/(V_{GG} - V_{GSP})$$

$$t_f = Q_2 \times R_G/V_{GSP}$$

where

 V_{GG} = the gate drive voltage, which varies from zero to V_{GG}

 R_G = the gate drive resistance and Q_2 and V_{GSP} are read from the gate charge curve.

During the turn-on and turn-off delay times, gate current is not constant. The simplest calculation uses appropriate values from the capacitance curves in a standard equation for voltage change in an RC network. The equations are:

$$t_{d(on)} = R_G C_{iss} In \left[V_{GG} / (V_{GG} - V_{GSP}) \right]$$

$$t_{d(off)} = R_G C_{iss} In (V_{GG}/V_{GSP})$$

The capacitance (C_{iss}) is read from the capacitance curve at a voltage corresponding to the off–state condition when calculating $t_{d(on)}$ and is read at a voltage corresponding to the on–state when calculating $t_{d(off)}$.

At high switching speeds, parasitic circuit elements complicate the analysis. The inductance of the MOSFET source lead, inside the package and in the circuit wiring which is common to both the drain and gate current paths, produces a voltage at the source which reduces the gate drive current. The voltage is determined by Ldi/dt, but since di/dt is a function of drain current, the mathematical solution is complex. The MOSFET output capacitance also complicates the mathematics. And finally, MOSFETs have finite internal gate resistance which effectively adds to the resistance of the driving source, but the internal resistance is difficult to measure and, consequently, is not specified.

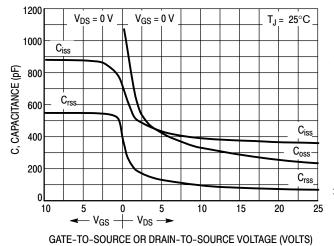
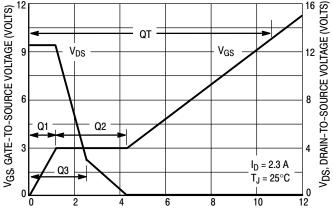


Figure 7. Capacitance Variation



Q₀, TOTAL GATE CHARGE (nC)

Figure 8. Gate-to-Source and Drain-to-Source Voltage versus Total Charge

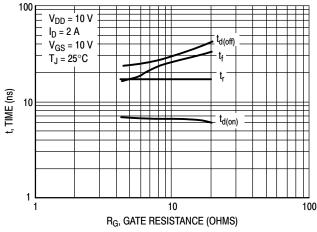


Figure 9. Resistive Switching Time Variation versus Gate Resistance

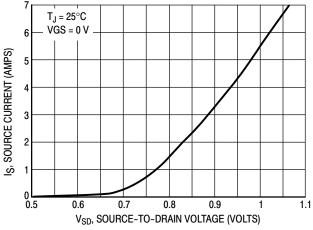


Figure 10. Diode Forward Voltage versus Current

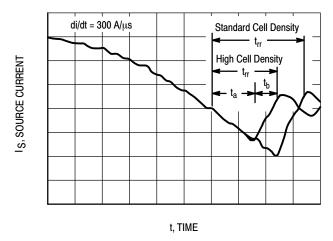


Figure 11. Reverse Recovery Time (t_{rr})

SAFE OPERATING AREA

The Forward Biased Safe Operating Area curves define the maximum simultaneous drain-to-source voltage and drain current that a transistor can handle safely when it is forward biased. Curves are based upon maximum peak junction temperature and a case temperature (T_C) of 25°C. Peak repetitive pulsed power limits are determined by using the thermal response data in conjunction with the procedures discussed in AN569, "Transient Thermal Resistance – General Data and Its Use."

Switching between the off-state and the on-state may traverse any load line provided neither rated peak current (I_{DM}) nor rated voltage (V_{DSS}) is exceeded, and that the transition time (t_r , t_f) does not exceed 10 μ s. In addition the total power averaged over a complete switching cycle must not exceed ($T_{J(MAX)} - T_C$)/($R_{\theta JC}$).

A power MOSFET designated E-FET can be safely used in switching circuits with unclamped inductive loads. For reliable operation, the stored energy from circuit inductance dissipated in the transistor while in avalanche must be less than the rated limit and must be adjusted for operating conditions differing from those specified. Although industry practice is to rate in terms of energy, avalanche energy capability is not a constant. The energy rating decreases non–linearly with an increase of peak current in avalanche and peak junction temperature.

Although many E–FETs can withstand the stress of drain–to–source avalanche at currents up to rated pulsed current (I_{DM}), the energy rating is specified at rated continuous current (I_{D}), in accordance with industry custom. The energy rating must be derated for temperature as shown in the accompanying graph (Figure 13). Maximum energy at currents below rated continuous I_{D} can safely be assumed to equal the values indicated.

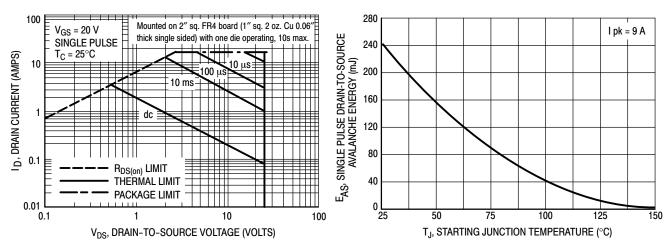


Figure 12. Maximum Rated Forward Biased Safe Operating Area

Figure 13. Maximum Avalanche Energy versus Starting Junction Temperature

TYPICAL ELECTRICAL CHARACTERISTICS

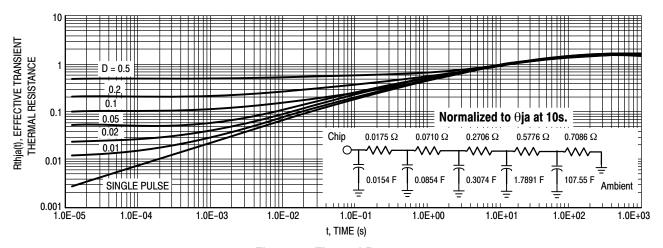


Figure 14. Thermal Response

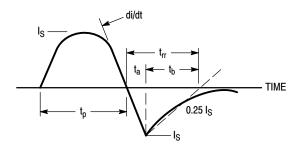


Figure 15. Diode Reverse Recovery Waveform



SOIC-8 NB CASE 751-07 **ISSUE AK**

DATE 16 FEB 2011



- NOTES:
 1. DIMENSIONING AND TOLERANCING PER
- ANSI Y14.5M, 1982.
 CONTROLLING DIMENSION: MILLIMETER.
- DIMENSION A AND B DO NOT INCLUDE MOLD PROTRUSION.
- MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE
- DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.
- 751-01 THRU 751-06 ARE OBSOLETE. NEW STANDARD IS 751-07.

	MILLIMETERS		MILLIMETERS INCHES	
DIM	MIN	MAX	MIN	MAX
Α	4.80	5.00	0.189	0.197
В	3.80	4.00	0.150	0.157
С	1.35	1.75	0.053	0.069
D	0.33	0.51	0.013	0.020
G	1.27 BSC		0.050 BSC	
Н	0.10	0.25	0.004	0.010
J	0.19	0.25	0.007	0.010
K	0.40	1.27	0.016	0.050
М	0 °	8 °	0 °	8 °
N	0.25	0.50	0.010	0.020
S	5.80	6.20	0.228	0.244

SOLDERING FOOTPRINT*



^{*}For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

GENERIC MARKING DIAGRAM*



XXXXX = Specific Device Code = Assembly Location

= Wafer Lot = Year = Work Week

= Pb-Free Package



XXXXXX = Specific Device Code = Assembly Location Α

= Year ww = Work Week

= Pb-Free Package

*This information is generic. Please refer to device data sheet for actual part marking. Pb-Free indicator, "G" or microdot "•", may or may not be present. Some products may not follow the Generic Marking.

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SOIC-8 NB CASE 751-07 ISSUE AK

DATE 16 FEB 2011

STYLE 4: PIN 1. ANODE 1 2. ANODE 2 3. ANODE 2 4. ANODE 5. ANODE #2 6. ANODE #2 7. ANODE #1 8. COMMON CATHODE
STYLE 8: PIN 1. COLLECTOR, DIE #1 2. BASE, #1 3. BASE, #2 4. COLLECTOR, #2 5. COLLECTOR, #2 6. EMITTER, #2 STAGE Vd 7. EMITTER, #1 AGE Vd 8. COLLECTOR, #1
STYLE 12: 1 PIN 1. SOURCE 2 SOURCE 2 3. SOURCE 4. GATE 5. DRAIN 6. DRAIN 7. DRAIN 8. DRAIN
STYLE 16: PIN 1. EMITTER, DIE #1 2. BASE, DIE #1 3. EMITTER, DIE #2 4. BASE, DIE #2 5. COLLECTOR, DIE #2 6. COMMON 6. COLLECTOR, DIE #2 6. COMMON 7. COLLECTOR, DIE #1 6. COMMON 8. COLLECTOR, DIE #1
STYLE 20: 1 PIN 1. SOURCE (N) 2. GATE (N) 2 3. SOURCE (P) 4. GATE (P) 5. DRAIN 2 6. DRAIN 7. DRAIN 1 8. DRAIN
STYLE 24:
STYLE 28: PIN 1. SW_TO_GND 2. DASIC_OFF 3. DASIC_SW_DET 4. GND E 5. V_MON E 6. VBULK E 7. VBULK 8. VIN

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