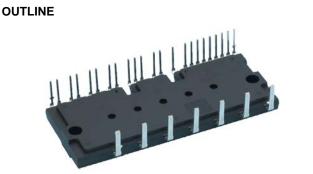


## < DIPIPM > **PSS15SA2FT** TRANSFER MOLDING TYPE

INSULATED TYPE



### MAIN FEATURES AND RATINGS

- 3 phase DC/AC inverter
- 1200V / 15A
- Built-in LPT-CSTBT (6th generation IGBT)
- Built-in bootstrap diodes with current limiting resistor
- Insulated transfer molding package
- N-side IGBT open emitter

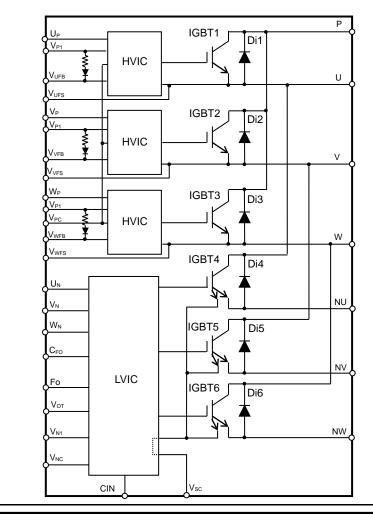
### APPLICATION

• AC 400V class motor control

## INTEGRATED DRIVE, PROTECTION AND SYSTEM CONTROL FUNCTIONS

- For P-side : Drive circuit, High voltage high-speed level shifting, Control supply under-voltage (UV) protection
- For N-side : Drive circuit, Control supply under-voltage protection (UV), Short circuit protection (SC)
- Fault signaling : Corresponding to SC fault (N-side IGBT), UV fault (N-side supply)
- Temperature output : Outputting LVIC temperature by analog signal
- Input interface : 5V line, Schmitt trigger receiver circuit (High Active)
- UL Recognized : UL1557 File E80276

### **INTERNAL CIRCUIT**



## MITSUBISHI ELECTRIC CORPORATION

## MAXIMUM RATINGS (T<sub>j</sub> = 25°C, unless otherwise noted)

**INVERTER PART** 

	Condition		Ratings	Unit
Supply voltage	Applied between P-NU,NV,NW		900	V
Supply voltage (surge)	Applied between P-NU,NV,NW		1000	V
Collector-emitter voltage			1200	V
Each IGBT collector current	T <sub>C</sub> = 25°C	(Note 1)	15	Α
Each IGBT collector current (peak)	$T_{c}$ = 25°C, up to 1ms		30	Α
Collector dissipation	T <sub>c</sub> = 25°C, per 1 chip		86.9	W
Junction temperature			-30~+150	°C
	Supply voltage (surge) Collector-emitter voltage Each IGBT collector current Each IGBT collector current (peak) Collector dissipation Junction temperature	Supply voltage (surge)Applied between P-NU,NV,NWCollector-emitter voltageEach IGBT collector current $T_c= 25^{\circ}C$ Each IGBT collector current (peak) $T_c= 25^{\circ}C$ , up to 1msCollector dissipation $T_c= 25^{\circ}C$ , per 1 chipJunction temperatureImage: Collector current curre	Supply voltage (surge)Applied between P-NU,NV,NWCollector-emitter voltage $T_c= 25^{\circ}C$ (Note 1)Each IGBT collector current (peak) $T_c= 25^{\circ}C$ , up to 1msCollector dissipation $T_c= 25^{\circ}C$ , per 1 chipJunction temperature $T_c= 25^{\circ}C$ , per 1 chip	Supply voltage (surge)Applied between P-NU,NV,NW1000Collector-emitter voltage1200Each IGBT collector current $T_c= 25^{\circ}C$ (Note 1)15Each IGBT collector current (peak) $T_c= 25^{\circ}C$ , up to 1ms30Collector dissipation $T_c= 25^{\circ}C$ , per 1 chip86.9

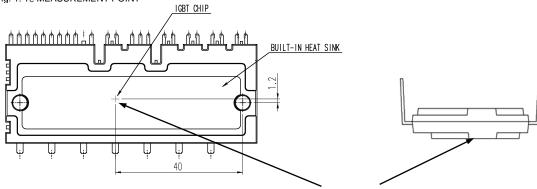
### **CONTROL (PROTECTION) PART**

Symbol	Parameter	Condition	Ratings	Unit
VD	Control supply voltage	Applied between $V_{P1}$ - $V_{PC}$ , $V_{N1}$ - $V_{NC}$	20	V
V <sub>DB</sub>	Control supply voltage	Applied between VUFB-VUFS, VVFB-VVFS, VWFB-VWFS	20	V
V <sub>IN</sub>	Input voltage	Applied between U <sub>P</sub> , V <sub>P</sub> , W <sub>P</sub> -V <sub>PC</sub> , U <sub>N</sub> , V <sub>N</sub> , W <sub>N</sub> -V <sub>NC</sub>	-0.5~V <sub>D</sub> +0.5	V
$V_{\text{FO}}$	Fault output supply voltage	Applied between Fo-VNC	-0.5~V <sub>D</sub> +0.5	V
I <sub>FO</sub>	Fault output current	Sink current at Fo terminal	5	mA
V <sub>SC</sub>	Current sensing input voltage	Applied between CIN-V <sub>NC</sub>	-0.5~V <sub>D</sub> +0.5	V

#### TOTAL SYSTEM

Symbol	Parameter	Condition	Ratings	Unit
V <sub>CC(PROT)</sub>	Self protection supply voltage limit (Short circuit protection capability)	$V_D$ = 13.5~16.5V, Inverter Part T <sub>i</sub> = 125°C, non-repetitive, up to 2µs	800	V
Tc	Module case operation temperature	Tc measurement point is defined in Fig.1.	-30~+100	°C
T <sub>stg</sub>	Storage temperature		-40~+125	°C
V <sub>iso</sub>	Isolation voltage	60Hz, Sinusoidal, AC 1min, between connected all pins and heat sink plate	2500	V <sub>rms</sub>

Fig. 1: Tc MEASUREMENT POINT



Measurement point for Tc

#### THERMAL RESISTANCE

Symbol	Parameter	Condition		Limits		
Symbol	Falametei			Тур.	Max.	Unit
R <sub>th(j-c)Q</sub>	Junction to case thermal Inverter IGBT part (per 1/6 module)		-	-	1.15	K/W
R <sub>th(j-c)F</sub>	resistance (Note 2) Inverter FWDi part (per 1/6 module)		-	-	1.60	K/W

Note 2: Grease with good thermal conductivity and long-term endurance should be applied evenly with about +100µm~+200µm on the contacting surface of DIPIPM and heat sink. The contacting thermal resistance between DIPIPM case and heat sink Rth(c-f) is determined by the thickness and the thermal conductivity of the applied grease. For reference, Rth(c-f) is about 0.2K/W (per 1/6 module, grease thickness: 20µm, thermal conductivity: 1.0W/m•k).

#### ELECTRICAL CHARACTERISTICS (T<sub>j</sub> = 25°C, unless otherwise noted) INVERTER PART

Cumbal	Deremeter	Condition			Limits			
Symbol	Parameter	Condi	Condition		Тур.	Max.	Unit	
M	Collector-emitter saturation		T <sub>j</sub> = 25°C	-	1.50	2.20	v	
V <sub>CE(sat)</sub>	voltage	$V_{D}=V_{DB} = 15V, V_{IN}=5V, I_{C}=15A$	T <sub>j</sub> = 125°C	-	1.70	2.40	v	
$V_{\text{EC}}$	FWDi forward voltage	$V_{IN}=0V$ , $-I_C=15A$		-	2.20	2.80	V	
t <sub>on</sub>				1.20	1.90	2.60	μs	
t <sub>C(on)</sub>		V <sub>CC</sub> = 600V, V <sub>D</sub> = V <sub>DB</sub> = 15V		-	0.50	0.80	μs	
t <sub>off</sub>	Switching times	mes I <sub>C</sub> = 15A, T <sub>j</sub> = 125°C, V <sub>IN</sub> = 0↔5V	-	2.60	3.60	μs		
t <sub>C(off)</sub>		Inductive Load (upper-lower arm)		-	0.50	0.90	μs	
t <sub>rr</sub>				-	0.50	-	μs	
1	Collector-emitter cut-off	ut-off V <sub>CE</sub> =V <sub>CES</sub>	T <sub>j</sub> = 25°C	-	-	1	mA	
ICES	current		T <sub>i</sub> = 125°C	-	-	10	IIIA	

## **CONTROL (PROTECTION) PART**

Symbol	Deremeter	Condition		Limits			Unit	
Symbol	Parameter		Condition		Min.	Тур.	Max.	Unit
			V <sub>D</sub> =15V, V <sub>IN</sub> =0V		-	-	5.60	
ID	Circuit current	Total of $V_{P1}$ - $V_{PC}$ , $V_{N1}$ - $V_{NC}$	$V_D=15V, V_{IN}=5V$		-	-	5.60	
1		Each part of V <sub>UFB</sub> -V <sub>UFS</sub> ,	$V_{DB}$ =15V, $V_{IN}$ =0V		-	-	1.10	mA
DB			$V_{DB}$ =15V, $V_{IN}$ =5V		-	-	1.10	
I <sub>SC</sub>	Short circuit trip level	-30°C≤Tj≤125°C, Rs=100Ω Not connecting outer shunt NU,NV,NW terminals		(Note 3)	25.5	-	-	A
$UV_{DBt}$	P-side Control supply	T <10500	Trip level		10.0	-	12.0	V
$UV_{DBr}$	under-voltage protection(UV)	T <sub>j</sub> ≤125°C	Reset level		10.5	-	12.5	V
UV <sub>Dt</sub>	N-side Control supply	T <405%	Trip level		10.3	-	12.5	V
$UV_{Dr}$	under-voltage protection(UV)	T <sub>j</sub> ≤125°C	Reset level		10.8	-	13.0	V
$V_{\text{FOH}}$	Fault output voltage	V <sub>SC</sub> = 0V, F <sub>0</sub> terminal pulled	l up to 5V by 10kΩ		4.9	-	-	V
V <sub>FOL</sub>	Fault output voltage	$V_{SC} = 1V$ , $I_{FO} = 1mA$			-	-	0.95	V
t <sub>FO</sub>	Fault output pulse width	C <sub>FO</sub> =22nF		(Note 4)	1.6	2.4	-	ms
I <sub>IN</sub>	Input current	$V_{IN} = 5V$			0.70	1.00	1.50	mA
V <sub>th(on)</sub>	ON threshold voltage	Applied between U <sub>P</sub> , V <sub>P</sub> , W			-	-	3.5	v
V <sub>th(off)</sub>	OFF threshold voltage	Applied between Op, Vp, W	, UN, VN, VVN-VNC		0.8	-	-	v
V <sub>OT</sub>	Temperature output	LVIC temperature = 75°C (Note 5)		2.26	2.38	2.51	V	
V <sub>F</sub>	Bootstrap Di forward voltage	I <sub>F</sub> =10mA including voltage drop by limiting resistor (Note 6)		0.5	0.9	1.3	V	
R	Built-in limiting resistance	Included in bootstrap Di			16	20	24	Ω

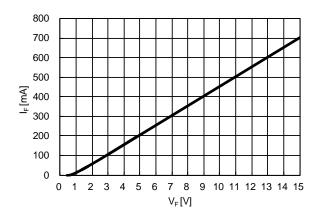
Note 3: Short circuit protection detects sense current divided from main current at N-side IGBT and works for N-side IGBT only. In the case that outer shunt resistor is inserted into main current path, protection current level Isc changes. For details, please refer the application note for this DIPIPM.
 4: Fault signal is output when short circuit or N-side control supply under-voltage protection works. The fault output pulse-width t<sub>FO</sub> depends on the capacitance of

CFO. (CFO (typ.) = tFO x 9.1 x 10<sup>-6</sup>) [F])

5: DIPIPM doesn't shut down IGBTs and output fault signal automatically when temperature rises excessively. When temperature exceeds the protective level that user defined, controller (MCU) should stop immediately. Temperature of LVIC vs. Vor output characteristics is described in Fig.3

6: The characteristics of bootstrap Di is described in Fig.2.

Fig. 2 Characteristics of bootstrap Di V<sub>F</sub>-I<sub>F</sub> curve (@Ta=25°C) including voltage drop by limiting resistor (Right chart is enlarged chart.)



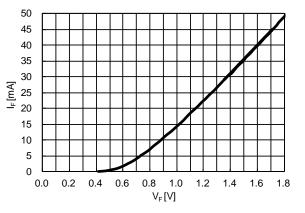
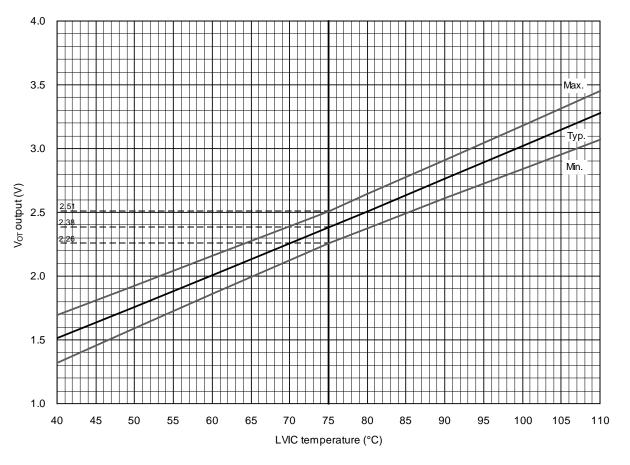
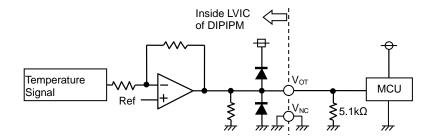


Fig. 3 Temperature of LVIC vs. Vot Output Characteristics





 (1) It is recommended to insert 5.1kΩ pull down resistor for getting linear output characteristics at low temperature below room temperature. When the pull down resistor is inserted between V<sub>OT</sub> and V<sub>NC</sub>(control GND), the extra circuit current, which is calculated approximately by V<sub>OT</sub> output voltage divided by pull down resistance, flows as LVIC circuit current continuously. In the case of using V<sub>OT</sub> for detecting high temperature over room temperature only, it is unnecessary to insert the pull down resistor.
 (2) In the case of not using V<sub>OT</sub>, leave V<sub>OT</sub> output NC (No Connection).

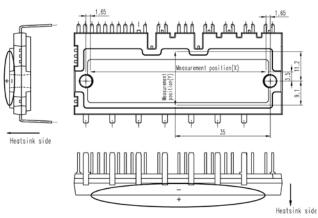
Refer the application note for this product about the usage of  $V_{\text{OT}}$ .

## MECHANICAL CHARACTERISTICS AND RATINGS

Parameter	Condition			Limits			
Parameter	Conc	Min.	Тур.	Max.	Unit		
Mounting torque	Mounting screw : M4 (Note 7)	Recommended 1.18N·m	0.98	1.18	1.47	N∙m	
Terminal pulling strength	Load 19.6N	JEITA-ED-4701	10	-	-	s	
Terminal bending strength	Load 9.8N, 90deg. bend	JEITA-ED-4701	2	-	-	times	
Weight			-	46	-	g	
Heat-sink flatness		-50	-	100	μm		

Note 7: Plain washers (ISO 7089~7094) are recommended.

Note 8: Measurement point of heat-sink flatness



## **RECOMMENDED OPERATION CONDITIONS**

Cumple al	Deveryoter	Condition			Limits		1.1.4.14
Symbol	Parameter	Condition		Min.	Тур.	Max.	Unit
V <sub>CC</sub>	Supply voltage	Applied between P-NU, NV, NW		350	600	800	V
VD	Control supply voltage	Applied between VP1-VPC, VN1-VNC		13.5	15.0	16.5	V
V <sub>DB</sub>	Control supply voltage	Applied between V <sub>UFB</sub> -V <sub>UFS</sub> , V <sub>VFB</sub> -V <sub>VFS</sub> , V <sub>V</sub>	VFB-VWFS	13.0	15.0	18.5	V
$\Delta V_D,\Delta V_{DB}$	Control supply variation				-	+1	V/µs
t <sub>dead</sub>	Arm shoot-through blocking time	For each input signal		3.0	-	-	μs
f <sub>PWM</sub>	PWM input frequency	T <sub>C</sub> ≤ 100°C, T <sub>j</sub> ≤ 125°C		-	-	20	kHz
		Sinusoidal PWM	f <sub>PWM</sub> = 5kHz	-	-	8.1	
I <sub>O</sub>	Allowable r.m.s. current			f <sub>PWM</sub> = 15kHz	-	-	5.5
PWIN(on)			(Note 10)	1.5	-	-	
PWIN(off)	Minimum input pulse width	$350 \le V_{CC} \le 800V, 13.5 \le V_D \le 16.5V,$ $13.0 \le V_{DB} \le 18.5V, -20^{\circ}C \le T_C \le 100^{\circ}C,$	I <sub>C</sub> ≤15A	3.0	-	-	μs
F WIN(OII)	N line wiring inductance less than 10nH (Note11)	15 <i<sub>C≤25.5A</i<sub>	3.5	-	-		
V <sub>NC</sub>	V <sub>NC</sub> variation	Between V <sub>NC</sub> -NU, NV, NW (including sur	ge)	-5.0	-	+5.0	V
Tj	Junction temperature			-20	-	+125	°C

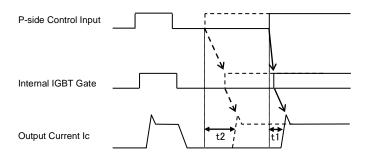
Note 9: The allowable r.m.s. current value depends on the actual application conditions.

10: DIPIPM might not make response to the input on signal with pulse width less than PWIN (on).

11: IPM might make no response or delayed response (P-side IGBT only) for the input signal with off pulse width less than PWIN(off).

Please refer below figure about delayed response.

Fig. 4 About Delayed Response Against Shorter Input Off Signal Than PWIN(off) (P-side only)



 Solid line
 ···Off pulse width ≥ PWIN(off);

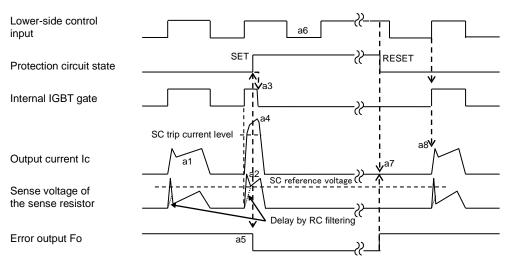
 Turn on time t1 (Normal delay)

 Broken line
 ···Off pulse width < PWIN(off);</td>

Turn on time t2 (Longer delay in some cases)

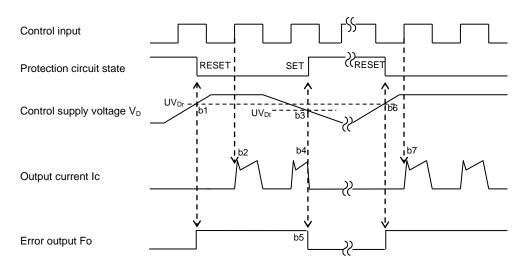
#### Fig. 5 Timing Charts of DIPIPM Protective Functions

- [A] Short-Circuit Protection (N-side only with the external sense resistor and RC filter)
- a1. Normal operation: IGBT ON and outputs current.
- a2. Short circuit current detection (SC trigger)
  - (It is recommended to set RC time constant 1.5~2.0µs so that IGBT shut down within 2.0µs when SC occurs.)
- a3. All N-side IGBT's gates are hard interrupted.
- a4. All N-side IGBTs turn OFF.
- a5.  $F_{\text{O}}$  outputs with a fixed pulse width determined by the external capacitor  $C_{\text{FO}}$
- a6. Input = "L": IGBT OFF
- a7. Fo finishes output, but IGBTs don't turn on until inputting next ON signal ( $L \rightarrow H$ ).
- (IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- a8. Normal operation: IGBT ON and outputs current.



[B] Under-Voltage Protection (N-side, UV<sub>D</sub>)

- b1. Control supply voltage  $V_D$  exceeds under voltage reset level (UV<sub>Dr</sub>), but IGBT turns ON by next ON signal (L $\rightarrow$ H). (IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- b2. Normal operation: IGBT ON and outputs current.
- b3.  $V_D$  level drops to under voltage trip level. (UV<sub>Dt</sub>).
- b4. All N-side IGBTs turn OFF in spite of control input condition.
- b5. Fo outputs for the period determined by the capacitance  $C_{FO}$ , but output is extended during  $V_D$  keeps below  $UV_{Dr}$ .
- b6. V<sub>D</sub> level reaches UV<sub>Dr</sub>.
- b7. Normal operation: IGBT ON and outputs current by next ON signal (L $\rightarrow$ H).



[C] Under-Voltage Protection (P-side, UV<sub>DB</sub>)

- c1. Control supply voltage  $V_{DB}$  rises. After the voltage reaches under voltage reset level  $UV_{DBr}$ , IGBT turns on by next ON signal (L $\rightarrow$ H).
- c2. Normal operation: IGBT ON and outputs current.
- c3.  $V_{DB}$  level drops to under voltage trip level (UV<sub>DBt</sub>).
- c4. IGBT of corresponding phase only turns OFF in spite of control input signal level, but there is no Fo signal output.
- c5.  $V_{DB}$  level reaches  $UV_{DBr}$ .
- c6. Normal operation: IGBT ON and outputs current by next ON signal (L $\rightarrow$ H).

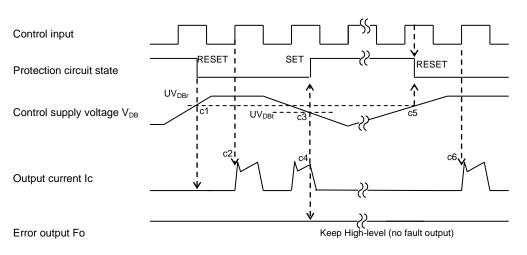
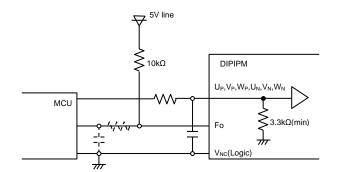


Fig. 6 MCU I/O Interface Circuit



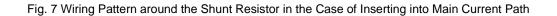
Note)

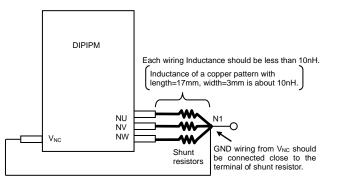
Design for input RC filter depends on the PWM control scheme used in the application and the wiring impedance of the printed circuit board.

But because noisier in the application for 1200V, it is strongly recommended to insert RC filter. (Time constant: over 100ns. e.g.  $100\Omega$ , 1000pF)

The DIPIPM input signal interface integrates a min.  $3.3k\Omega$  pull-down resistor. Therefore, when using RC filter, be careful to satisfy turn-on threshold voltage requirement.

Fo output is open drain type. It should be pulled up to the positive side of 5V or 15V power supply with the resistor that limits Fo sink current  $I_{Fo}$  under 1mA. In the case of pulling up to 5V supply, over  $5.1k\Omega$  is needed. (10k $\Omega$  is recommended.)

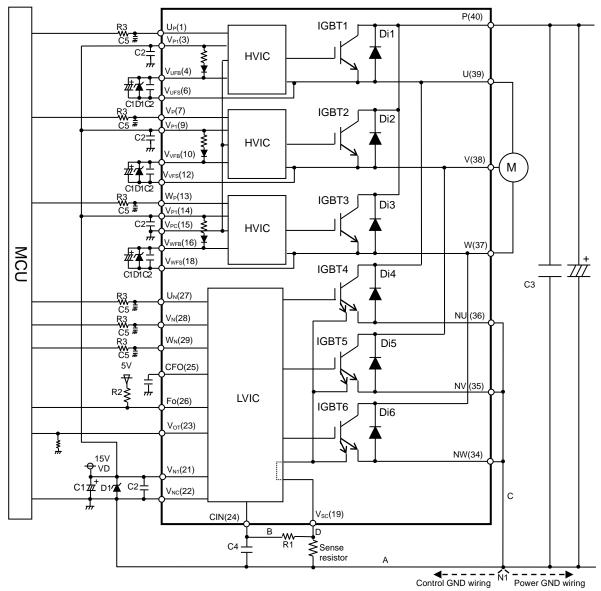




Low inductance shunt resistor like surface mounted (SMD) type is recommended. Protection current level  $I_{SC}$  changes by inserting shunt resistor.

## < DIPIPM > **PSS15SA2FT** TRANSFER MOLDING TYPE INSULATED TYPE

## Fig. 8 Example of Application Circuit



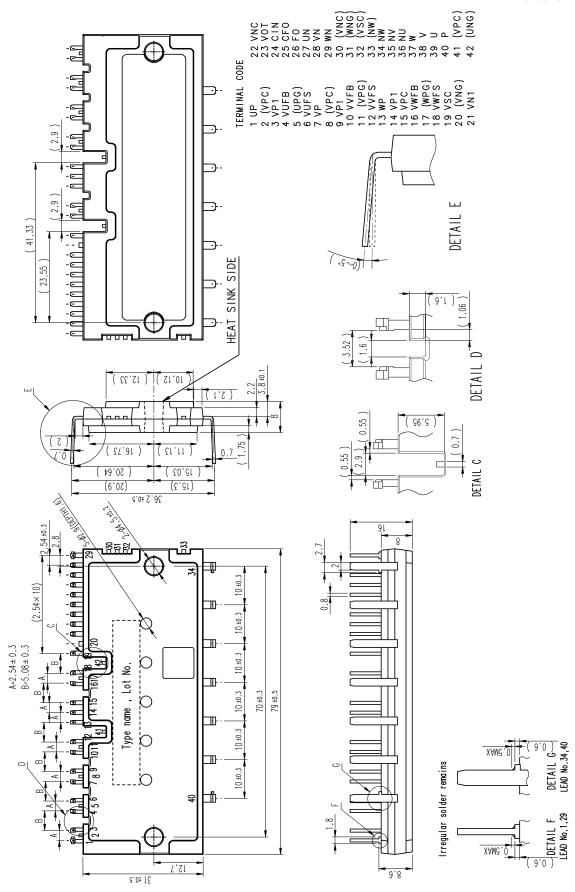
#### Note

- 1 :If control GND and power GND are patterned by common wiring, it may cause malfunction by fluctuation of power GND level. It is recommended to connect control GND and power GND at only a N1 point at which NU, NV, NW are connected to power GND line.
- 2 :It is recommended to insert a Zener diode D1 (24V/1W) between each pair of control supply terminals to prevent surge destruction.
- 3 :To prevent surge destruction, the wiring between the smoothing capacitor and the P, N1 terminals should be as short as possible. Generally inserting a 0.1µ~0.22µF snubber capacitor C3 between the P-N1 terminals is recommended.
- 4 :R1, C4 of RC filter for preventing protection circuit malfunction is recommended to select tight tolerance, temp-compensated type. The time constant R1C4 should be set so that SC current is shut down within 2µs. (1.5µs~2µs is general value.) SC interrupting time might vary with the wiring pattern, so the enough evaluation on the real system is recommended. If R1 is too small, it may leads to delay of protection. So R1 should be min. 10 times larger resistance than Rs. (100 times is recommended.)
- 5 :To prevent erroneous operation, the wiring of A, B, C should be as short as possible.
- 6 :For sense resistor, the variation within 1% (including temperature characteristics), low inductance type is recommended. And the over 0.03W is recommended, but it is necessary to evaluate in your real system finally.
- 7 :To prevent erroneous SC protection, the wiring from V<sub>SC</sub> terminal to CIN filter should be divided at the point D that is close to the terminal of sense resistor. And the wiring should be patterned as short as possible.
- 8 :All capacitors should be mounted as close to the terminals of the DIPIPM as possible. (C1: good temperature, frequency characteristic electrolytic type, and C2: 0.01µ~2.0µF, good temperature, frequency and DC bias characteristic ceramic type are recommended.)
- 9 :Input drive is High-active type. There is a min. 3.3kΩ pull-down resistor in the input circuit of IC. To prevent malfunction, the wiring of each input should be as short as possible. And it is recommended to insert RC filter (e.g. R3=100Ω and C5=1000pF) and confirm the input signal level to meet the turn-on and turn-off threshold voltage. Thanks to HVIC inside the module, direct coupling to MCU without any opto-coupler or transformer isolation is possible.
- 10 :Fo output is open drain type. Fo output will be max 0.95V(@IFo=1mA,25°C), so it should be pulled up to MCU or control power supply (e.g. 5V,15V) by a resistor that makes IFo up to 1mA. (In the case of pulled up to 5V, 10kΩ is recommended.)
- 11 :Error signal output width (tro) can be set by the capacitor connected to CFO terminal. CFO(typ.) = tFo x 9.1 x 10<sup>-6</sup> (F)
- 12: If high frequency noise superimposed to the control supply line, IC malfunction might happen and cause erroneous operation. To avoid such problem, voltage ripple of control supply line should meet dV/dt <+/-1V/µs, Vripple≤2Vp-p.
- 13 :For DIPIPM, it isn't recommended to drive same load by parallel connection with other phase IGBT or other DIPIPM.

## < DIPIPM > **PSS15SA2FT** TRANSFER MOLDING TYPE INSULATED TYPE

Fig. 9 Package Outlines

Dimensions in mm



## Important Notice

The information contained in this datasheet shall in no event be regarded as a guarantee of conditions or characteristics. This product has to be used within its specified maximum ratings, and is subject to customer's compliance with any applicable legal requirement, norms and standards.

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## Keep safety first in your circuit designs!

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