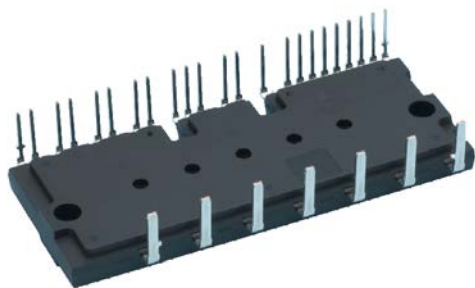


< DIIPM >

PSS50SA2FT

TRANSFER MOLDING TYPE
INSULATED TYPE

OUTLINE



MAIN FEATURES AND RATINGS

- 3 phase DC/AC inverter
- 1200V / 50A
- Built-in LPT-CSTBT (6th generation IGBT)
- Built-in bootstrap diodes with current limiting resistor
- Insulated transfer molding package
- N-side IGBT open emitter

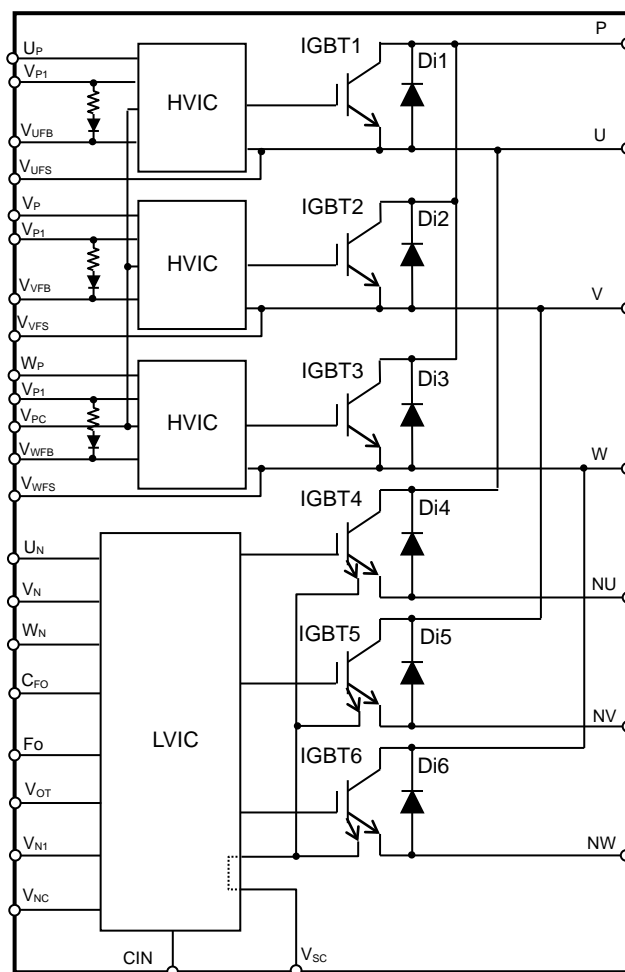
APPLICATION

- AC 400V class motor control

INTEGRATED DRIVE, PROTECTION AND SYSTEM CONTROL FUNCTIONS

- For P-side : Drive circuit, High voltage high-speed level shifting, Control supply under-voltage (UV) protection
- For N-side : Drive circuit, Control supply under-voltage protection (UV), Short circuit protection (SC)
- Fault signaling : Corresponding to SC fault (N-side IGBT), UV fault (N-side supply)
- Temperature output : Outputting LVIC temperature by analog signal
- Input interface : 5V line, Schmitt trigger receiver circuit (High Active)
- UL Recognized : UL1557 File E80276

INTERNAL CIRCUIT



MAXIMUM RATINGS ($T_j = 25^\circ\text{C}$, unless otherwise noted)**INVERTER PART**

| Symbol | Parameter | Condition | Ratings | Unit |
|-----------------|------------------------------------|---------------------------------------|----------|------------------|
| V_{CC} | Supply voltage | Applied between P-NU, NV, NW | 900 | V |
| $V_{CC(surge)}$ | Supply voltage (surge) | Applied between P-NU, NV, NW | 1000 | V |
| V_{CES} | Collector-emitter voltage | | 1200 | V |
| $\pm I_C$ | Each IGBT collector current | $T_C = 25^\circ\text{C}$ (Note 1) | 50 | A |
| $\pm I_{CP}$ | Each IGBT collector current (peak) | $T_C = 25^\circ\text{C}$, up to 1ms | 100 | A |
| P_C | Collector dissipation | $T_C = 25^\circ\text{C}$, per 1 chip | 123.4 | W |
| T_j | Junction temperature | | -30~+150 | $^\circ\text{C}$ |

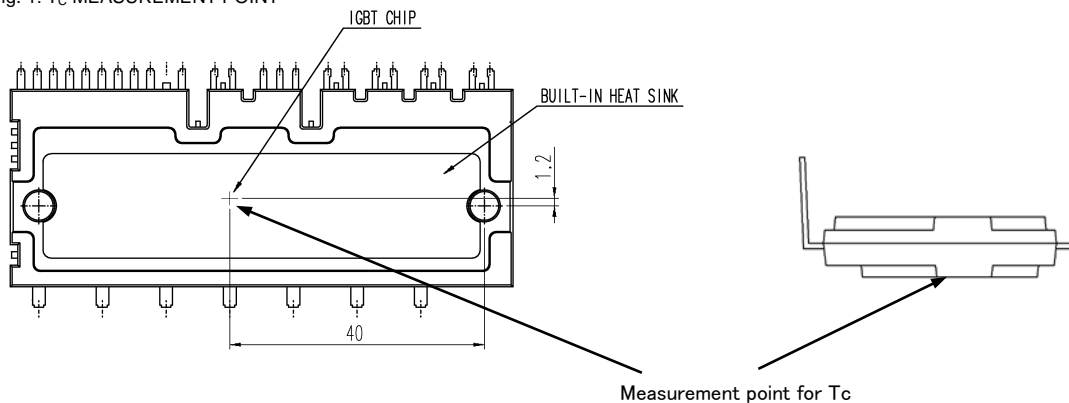
Note 1: Pulse width and period are limited due to junction temperature.

CONTROL (PROTECTION) PART

| Symbol | Parameter | Condition | Ratings | Unit |
|----------|-------------------------------|-----------------------------------------------------------------------------|------------------|------|
| V_D | Control supply voltage | Applied between $V_{P1}-V_{PC}$, $V_{N1}-V_{NC}$ | 20 | V |
| V_{DB} | Control supply voltage | Applied between $V_{UFB}-V_{UFS}$, $V_{VFB}-V_{VFS}$, $V_{WFB}-V_{WFS}$ | 20 | V |
| V_{IN} | Input voltage | Applied between U_P , V_P , W_P-V_{PC} , U_N , V_N , W_N-V_{NC} | -0.5~ V_D +0.5 | V |
| V_{FO} | Fault output supply voltage | Applied between F_O-V_{NC} | -0.5~ V_D +0.5 | V |
| I_{FO} | Fault output current | Sink current at F_O terminal | 5 | mA |
| V_{SC} | Current sensing input voltage | Applied between $CIN-V_{NC}$ | -0.5~ V_D +0.5 | V |

TOTAL SYSTEM

| Symbol | Parameter | Condition | Ratings | Unit |
|----------------|----------------------------------------------------------------------------|--------------------------------------------------------------------------------------------------------------------|----------|------------------|
| $V_{CC(PROT)}$ | Self protection supply voltage limit (Short circuit protection capability) | $V_D = 13.5\sim 16.5\text{V}$, Inverter Part $T_j = 125^\circ\text{C}$, non-repetitive, up to 2 μs | 800 | V |
| T_C | Module case operation temperature | T_C measurement point is defined in Fig.1. | -30~+100 | $^\circ\text{C}$ |
| T_{stg} | Storage temperature | | -40~+125 | $^\circ\text{C}$ |
| V_{iso} | Isolation voltage | 60Hz, Sinusoidal, AC 1min, between connected all pins and heat sink plate | 2500 | V_{rms} |

Fig. 1: T_C MEASUREMENT POINT**THERMAL RESISTANCE**

| Symbol | Parameter | Condition | Limits | | | Unit |
|----------------|-------------------------------------|-------------------------------------|--------|------|------|------|
| | | | Min. | Typ. | Max. | |
| $R_{th(j-c)Q}$ | Junction to case thermal resistance | Inverter IGBT part (per 1/6 module) | - | - | 0.81 | K/W |
| $R_{th(j-c)F}$ | resistance (Note 2) | Inverter FWDi part (per 1/6 module) | - | - | 1.25 | K/W |

Note 2: Grease with good thermal conductivity and long-term endurance should be applied evenly with about +100 μm ~+200 μm on the contacting surface of DIIPM and heat sink. The contacting thermal resistance between DIIPM case and heat sink $R_{th(c-f)}$ is determined by the thickness and the thermal conductivity of the applied grease. For reference, $R_{th(c-f)}$ is about 0.2K/W (per 1/6 module, grease thickness: 20 μm , thermal conductivity: 1.0W/m \cdot K).

ELECTRICAL CHARACTERISTICS ($T_j = 25^\circ\text{C}$, unless otherwise noted)**INVERTER PART**

| Symbol | Parameter | Condition | | Limits | | | Unit |
|---------------|--------------------------------------|------------------------------------------------------------------------------------------------------------------------------------|---------------------|--------|------|------|---------|
| | | | | Min. | Typ. | Max. | |
| $V_{CE(sat)}$ | Collector-emitter saturation voltage | $V_D=V_{DB} = 15V, V_{IN}= 5V, I_C= 50A$ | $T_J= 25^{\circ}C$ | - | 1.70 | 2.40 | V |
| | | | $T_J= 125^{\circ}C$ | - | 2.00 | 2.50 | |
| V_{EC} | FWDi forward voltage | $V_{IN}= 0V, -I_C= 50A$ | | - | 2.50 | 3.20 | V |
| t_{on} | Switching times | $V_{CC}= 600V, V_D= V_{DB}= 15V$ $I_C= 50A, T_J= 125^{\circ}C, V_{IN}= 0\leftrightarrow 5V$ Inductive Load (upper-lower arm) | | 0.70 | 1.80 | 2.50 | μs |
| $t_{C(on)}$ | | | | - | 0.50 | 0.80 | μs |
| t_{off} | | | | - | 2.60 | 3.80 | μs |
| $t_{C(off)}$ | | | | - | 0.50 | 0.90 | μs |
| t_{rr} | | | | - | 0.50 | - | μs |
| I_{CES} | Collector-emitter cut-off current | $V_{CE}=V_{CES}$ | $T_J= 25^{\circ}C$ | - | - | 1 | mA |
| | | | $T_J= 125^{\circ}C$ | - | - | 10 | |

CONTROL (PROTECTION) PART

| Symbol | Parameter | Condition | | Limits | | | Unit |
|----------------------|----------------------------------------------------------|--------------------------------------------------------------------------------------------------------------------------------------|-------------------------------------------|--------|------|------|------|
| | | | | Min. | Typ. | Max. | |
| I _D | Circuit current | Total of V _{P1} -V _{PC} , V _{N1} -V _{NC} | V _D =15V, V _{IN} =0V | - | - | 5.60 | mA |
| | | | V _D =15V, V _{IN} =5V | - | - | 5.60 | |
| I _{DB} | | Each part of V _{UFB} -V _{UFS} , V _{VFB} -V _{VFS} , V _{WFB} -V _{WFS} | V _{DB} =15V, V _{IN} =0V | - | - | 1.10 | |
| | | | V _{DB} =15V, V _{IN} =5V | - | - | 1.10 | |
| I _{SC} | Short circuit trip level | -30°C≤T _J ≤125°C, R _S =34.0Ω (±1%), Not connecting outer shunt resistors to NU,NV,NW terminals (Note 3) | | 85 | - | - | A |
| UV _{DBt} | P-side Control supply under-voltage protection(UV) | T _J ≤125°C | Trip level | 10.0 | - | 12.0 | V |
| UV _{DBr} | | | Reset level | 10.5 | - | 12.5 | V |
| UV _{Dt} | N-side Control supply under-voltage protection(UV) | T _J ≤125°C | Trip level | 10.3 | - | 12.5 | V |
| UV _{Dr} | | | Reset level | 10.8 | - | 13.0 | V |
| V _{FOH} | Fault output voltage | V _{SC} = 0V, F _O terminal pulled up to 5V by 10kΩ | | 4.9 | - | - | V |
| V _{FOL} | | V _{SC} = 1V, I _{FO} = 1mA | | - | - | 0.95 | V |
| t _{FO} | Fault output pulse width | C _{FO} =22nF (Note 4) | | 1.6 | 2.4 | - | ms |
| I _{IN} | Input current | V _{IN} = 5V | | 0.70 | 1.00 | 1.50 | mA |
| V _{th(on)} | ON threshold voltage | Applied between U _P , V _P , W _P , U _N , V _N , W _N -V _{NC} | | - | - | 3.5 | V |
| V _{th(off)} | OFF threshold voltage | | | 0.8 | - | - | |
| V _{OT} | Temperature output | LVIC temperature = 75°C (Note 5) | | 2.26 | 2.38 | 2.51 | V |
| V _F | Bootstrap Di forward voltage | I _F =10mA including voltage drop by limiting resistor (Note 6) | | 0.5 | 0.9 | 1.3 | V |
| R | Built-in limiting resistance | Included in bootstrap Di | | 16 | 20 | 24 | Ω |

Note 3: Short circuit protection detects sense current divided from main current at N-side IGBT and works for N-side IGBT only. In the case that outer shunt resistor is inserted into main current path, protection current level I_{SC} changes. For details, please refer the application note for this DIIPM.

4: Fault signal is output when short circuit or N-side control supply under-voltage protection works. The fault output pulse-width t_{FO} depends on the capacitance of C_{FO} . ($C_{FO}(\text{typ.}) = t_{FO} \times 9.1 \times 10^{-6}$) [F]

5: DIIPM doesn't shut down IGBTs and output fault signal automatically when temperature rises excessively. When temperature exceeds the protective level that user defined, controller (MCU) should stop immediately. Temperature of LVIC vs. V_{OT} output characteristics is described in Fig.3

6: The characteristics of bootstrap Di is described in Fig.2.

Fig. 2 Characteristics of bootstrap Di V_F - I_F curve (@ $T_a=25^\circ\text{C}$) including voltage drop by limiting resistor (Right chart is enlarged chart.)

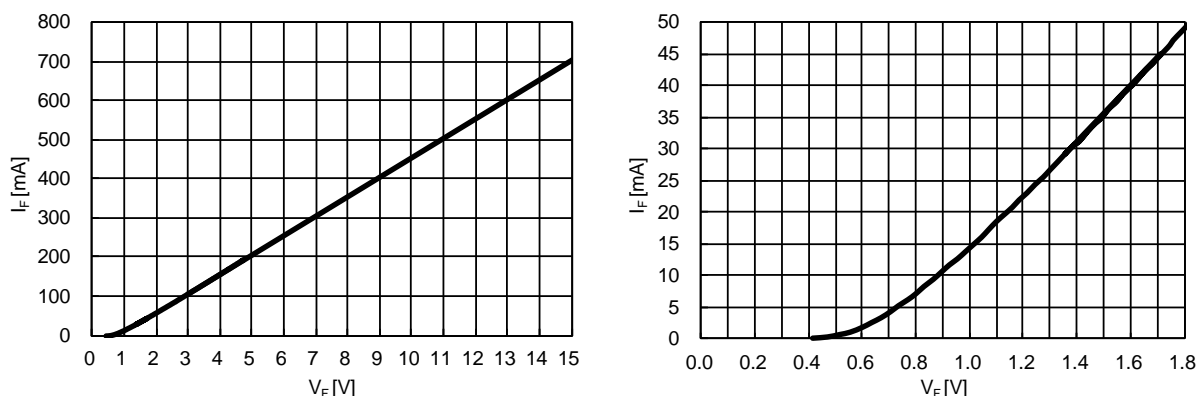
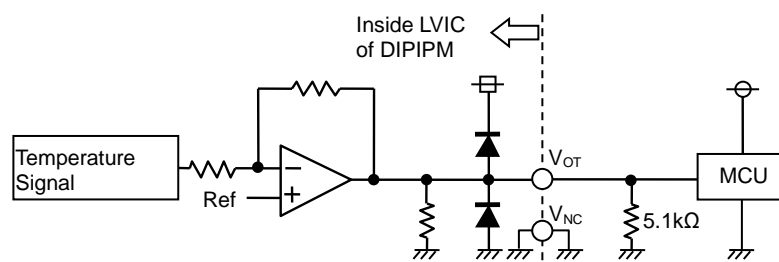
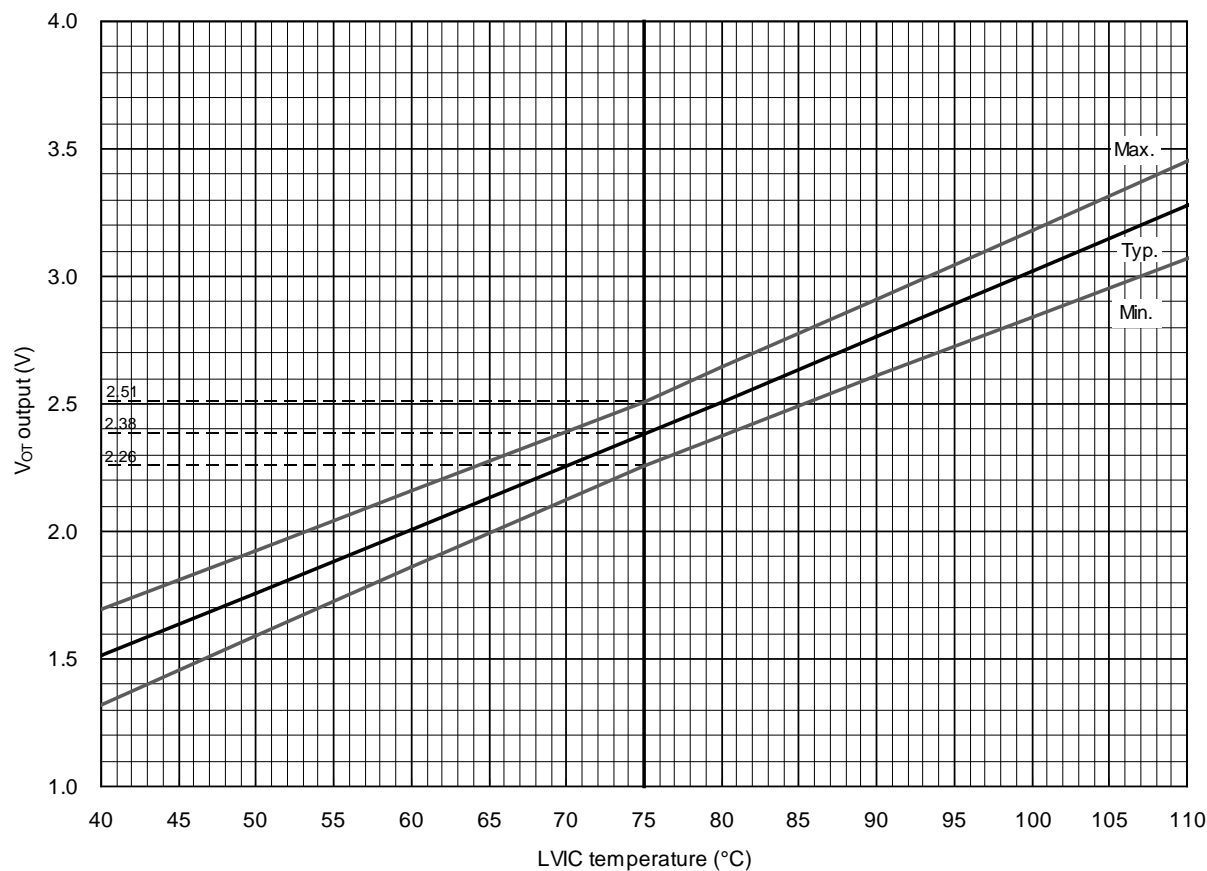


Fig. 3 Temperature of LVIC vs. V_{OT} Output Characteristics

- (1) It is recommended to insert 5.1kΩ pull down resistor for getting linear output characteristics at low temperature below room temperature. When the pull down resistor is inserted between V_{OT} and V_{NC} (control GND), the extra circuit current, which is calculated approximately by V_{OT} output voltage divided by pull down resistance, flows as LVIC circuit current continuously. In the case of using V_{OT} for detecting high temperature over room temperature only, it is unnecessary to insert the pull down resistor.
- (2) In the case of not using V_{OT} , leave V_{OT} output NC (No Connection).

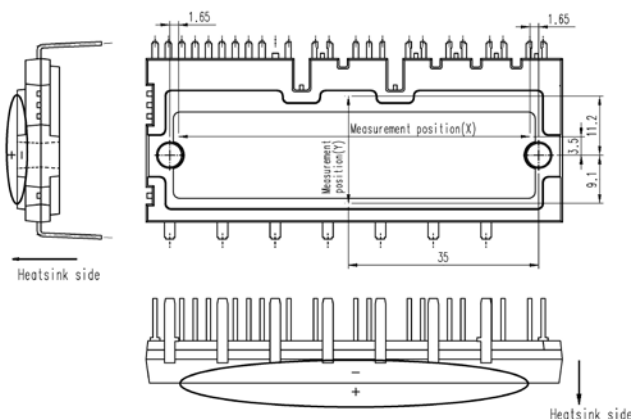
Refer the application note for this product about the usage of V_{OT} .

MECHANICAL CHARACTERISTICS AND RATINGS

| Parameter | Condition | | Limits | | | Unit |
|---------------------------|------------------------------|---------------------|--------|------|------|-------|
| | | | Min. | Typ. | Max. | |
| Mounting torque | Mounting screw : M4 (Note 7) | Recommended 1.18N·m | 0.98 | 1.18 | 1.47 | N·m |
| Terminal pulling strength | Load 19.6N | JEITA-ED-4701 | 10 | - | - | s |
| Terminal bending strength | Load 9.8N, 90deg. bend | JEITA-ED-4701 | 2 | - | - | times |
| Weight | | | - | 46 | - | g |
| Heat-sink flatness | | (Note 8) | -50 | - | 100 | μm |

Note 7: Plain washers (ISO 7089~7094) are recommended.

Note 8: Measurement point of heat-sink flatness

**RECOMMENDED OPERATION CONDITIONS**

| Symbol | Parameter | Condition | Limits | | | Unit |
|--------------------------------|---------------------------------|-------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|-----------------------------------|---------------------------------|---------------------------------|------|
| | | | Min. | Typ. | Max. | |
| V_{CC} | Supply voltage | Applied between P-NU, NV, NW | 350 | 600 | 800 | V |
| V_D | Control supply voltage | Applied between $V_{P1}-V_{PC}$, $V_{N1}-V_{NC}$ | 13.5 | 15.0 | 16.5 | V |
| V_{DB} | Control supply voltage | Applied between $V_{UFB}-V_{UFS}$, $V_{VFB}-V_{VFS}$, $V_{WFB}-V_{WFS}$ | 13.0 | 15.0 | 18.5 | V |
| ΔV_D , ΔV_{DB} | Control supply variation | | -1 | - | +1 | V/μs |
| t_{dead} | Arm shoot-through blocking time | For each input signal | 3.0 | - | - | μs |
| f_{PWM} | PWM input frequency | $T_C \leq 100^\circ\text{C}$, $T_J \leq 125^\circ\text{C}$ | - | - | 20 | kHz |
| I_O | Allowable r.m.s. current | $V_{CC} = 600\text{V}$, $V_D = 15\text{V}$, P.F = 0.8, Sinusoidal PWM $T_C \leq 100^\circ\text{C}$, $T_J \leq 125^\circ\text{C}$ (Note 9) | $f_{PWM} = 5\text{kHz}$ - | $f_{PWM} = 5\text{kHz}$ - | $f_{PWM} = 5\text{kHz}$ 25 | Arms |
| I_O | Allowable r.m.s. current | $T_C \leq 100^\circ\text{C}$, $T_J \leq 125^\circ\text{C}$ (Note 9) | $f_{PWM} = 15\text{kHz}$ - | $f_{PWM} = 15\text{kHz}$ - | $f_{PWM} = 15\text{kHz}$ 13 | |
| PWIN(on) | Minimum input pulse width | (Note 10) | 1.5 | - | - | μs |
| PWIN(off) | Minimum input pulse width | $350 \leq V_{CC} \leq 800\text{V}$, $13.5 \leq V_D \leq 16.5\text{V}$, $13.0 \leq V_{DB} \leq 18.5\text{V}$, $-20^\circ\text{C} \leq T_C \leq 100^\circ\text{C}$, N line wiring inductance less than 10nH (Note 11) | $I_C \leq 50\text{A}$ 3.0 | $I_C \leq 50\text{A}$ - | $I_C \leq 50\text{A}$ - | |
| | | | $50 < I_C \leq 85\text{A}$ 3.5 | $50 < I_C \leq 85\text{A}$ - | $50 < I_C \leq 85\text{A}$ - | |
| V_{NC} | V_{NC} variation | Between $V_{NC}-\text{NU}$, NV, NW (including surge) | -5.0 | - | +5.0 | V |
| T_J | Junction temperature | | -20 | - | +125 | °C |

Note 9: The allowable r.m.s. current value depends on the actual application conditions.

Note 10: DIIPM might not make response to the input on signal with pulse width less than PWIN (on).

Note 11: IPM might make no response or delayed response (P-side IGBT only) for the input signal with off pulse width less than PWIN(off).

Please refer below figure about delayed response.

Fig. 4 About Delayed Response Against Shorter Input Off Signal Than PWIN(off) (P-side only)

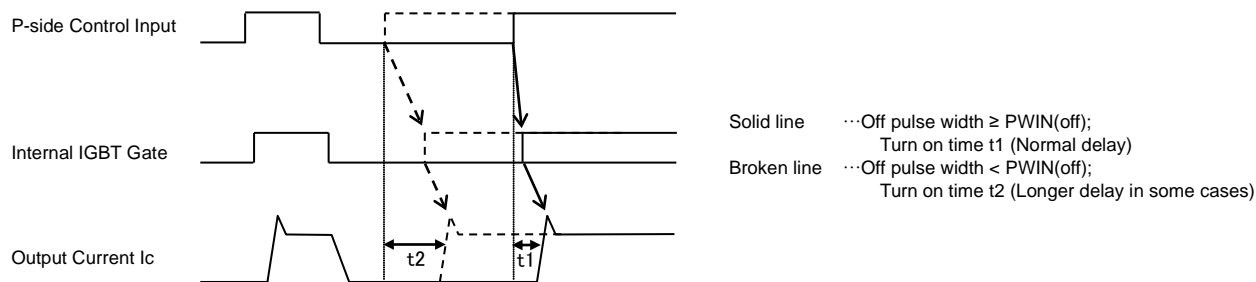
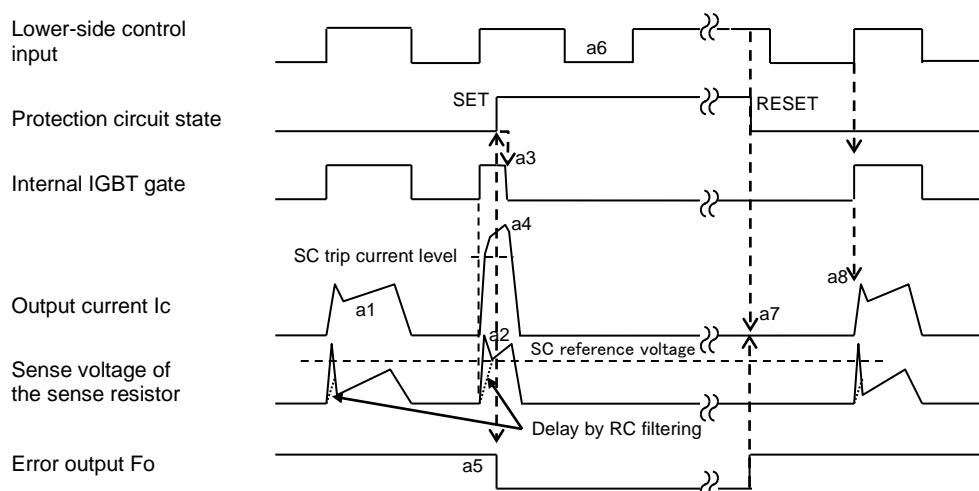


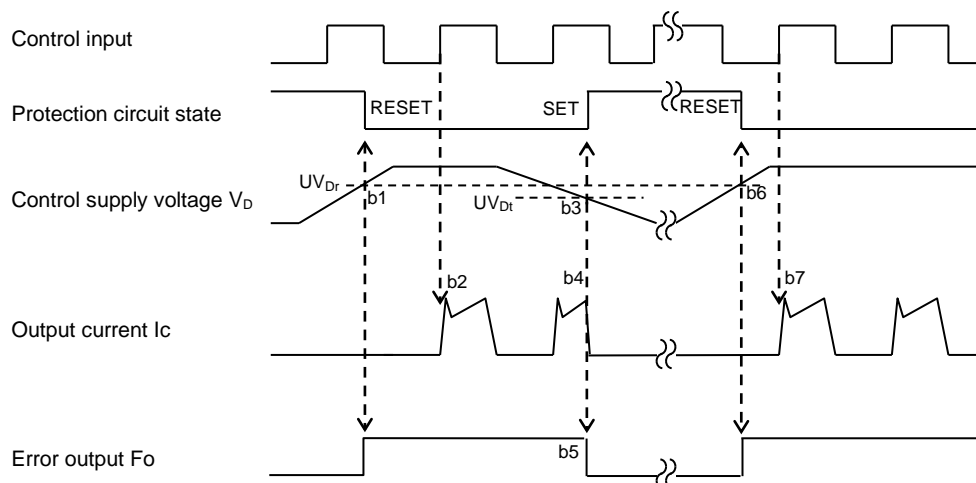
Fig. 5 Timing Charts of DIIPM Protective Functions

[A] Short-Circuit Protection (N-side only with the external sense resistor and RC filter)

- a1. Normal operation: IGBT ON and outputs current.
- a2. Short circuit current detection (SC trigger)
(It is recommended to set RC time constant 1.5~2.0 μ s so that IGBT shut down within 2.0 μ s when SC occurs.)
- a3. All N-side IGBT's gates are hard interrupted.
- a4. All N-side IGBTs turn OFF.
- a5. F_O outputs with a fixed pulse width determined by the external capacitor C_{FO} .
- a6. Input = "L": IGBT OFF
- a7. F_O finishes output, but IGBTs don't turn on until inputting next ON signal (L \rightarrow H).
(IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- a8. Normal operation: IGBT ON and outputs current.

**[B] Under-Voltage Protection (N-side, UV_D)**

- b1. Control supply voltage V_D exceeds under voltage reset level (UV_{Dr}), but IGBT turns ON by next ON signal (L \rightarrow H).
(IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- b2. Normal operation: IGBT ON and outputs current.
- b3. V_D level drops to under voltage trip level. (UV_{Dt}).
- b4. All N-side IGBTs turn OFF in spite of control input condition.
- b5. F_O outputs for the period determined by the capacitance C_{FO} , but output is extended during V_D keeps below UV_{Dr} .
- b6. V_D level reaches UV_{Dr} .
- b7. Normal operation: IGBT ON and outputs current by next ON signal (L \rightarrow H).



- c1. Control supply voltage V_{DB} rises. After the voltage reaches under voltage reset level UV_{DBr} , IGBT turns on by next ON signal (L→H).
- c2. Normal operation: IGBT ON and outputs current.
- c3. V_{DB} level drops to under voltage trip level (UV_{DBt}).
- c4. IGBT of corresponding phase only turns OFF in spite of control input signal level, but there is no F_O signal output.
- c5. V_{DB} level reaches UV_{DBr} .
- c6. Normal operation: IGBT ON and outputs current by next ON signal (L→H).

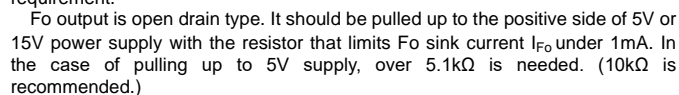


Diagram illustrating the connection of a shunt resistor for a DIPIPM (Digital Input/Output Power Interface Module).

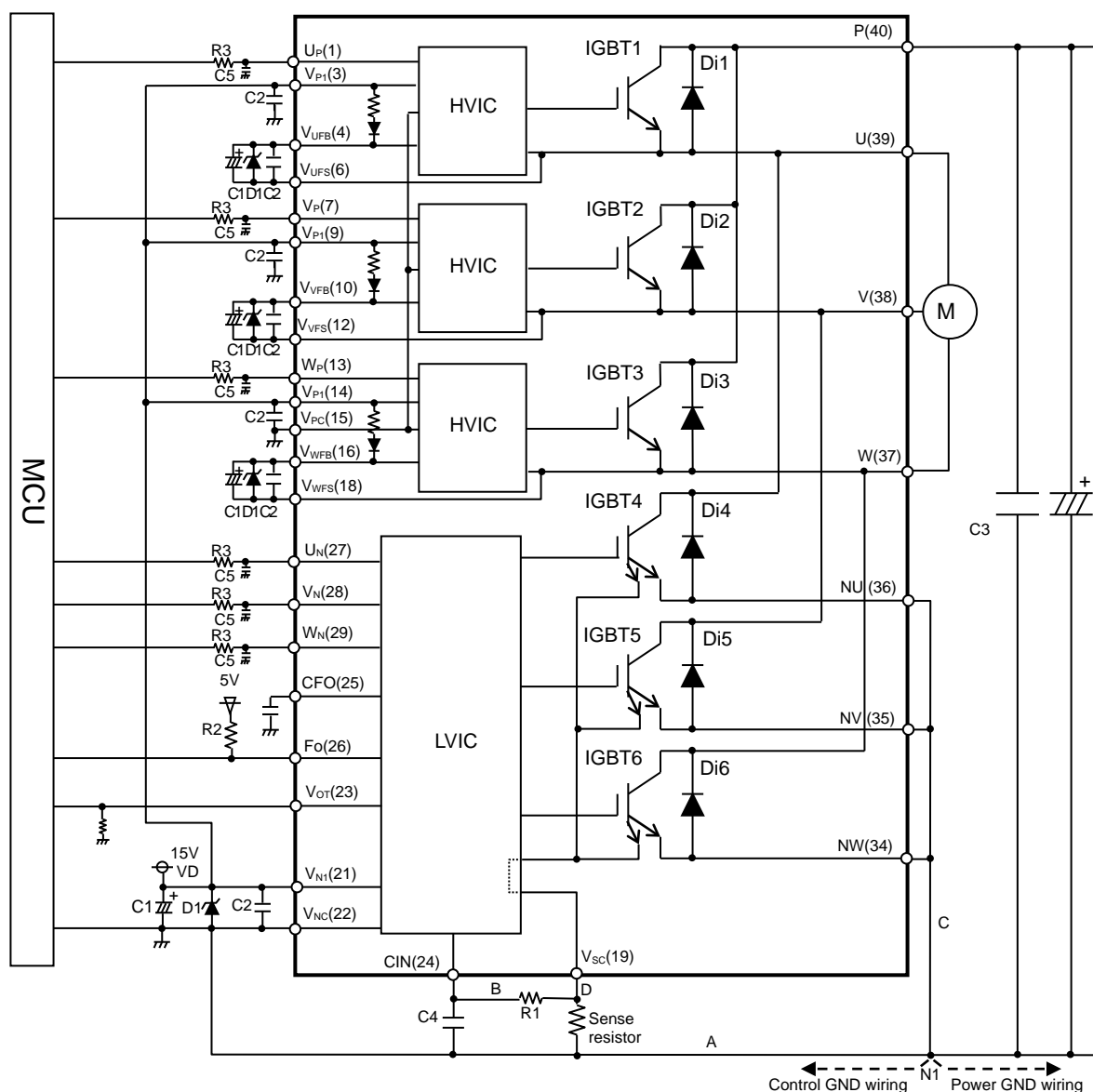
The DIPIPM is connected to a voltage source V_{NC} . The output terminals are labeled NU, NV, and NW. These terminals are connected to a common node N1 through shunt resistors.

Each wiring inductance should be less than 10nH. Inductance of a copper pattern with length=17mm, width=3mm is about 10nH.

The shunt resistors are connected to a common node N1, which is connected to ground (GND). The GND wiring from V_{NC} should be connected close to the terminal of shunt resistor.

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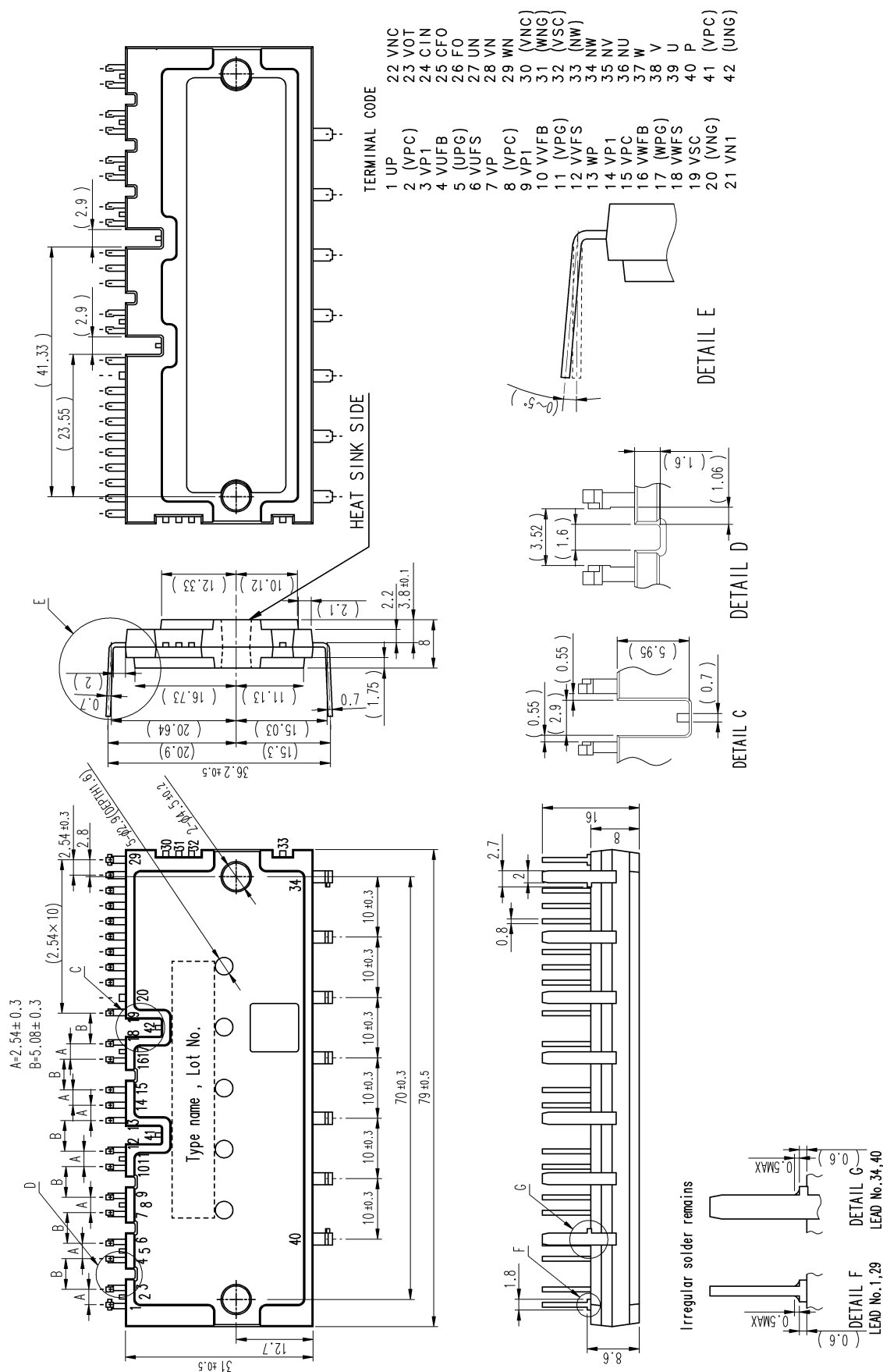
Fig. 8 Example of Application Circuit

**Note**

- 1 :If control GND and power GND are patterned by common wiring, it may cause malfunction by fluctuation of power GND level. It is recommended to connect control GND and power GND at only a N1 point at which NU, NV, NW are connected to power GND line.
- 2 :It is recommended to insert a Zener diode D1 (24V/1W) between each pair of control supply terminals to prevent surge destruction.
- 3 :To prevent surge destruction, the wiring between the smoothing capacitor and the P, N1 terminals should be as short as possible. Generally inserting a 0.1 μ F~0.22 μ F snubber capacitor C3 between the P-N1 terminals is recommended.
- 4 :R1, C4 of RC filter for preventing protection circuit malfunction is recommended to select tight tolerance, temp-compensated type. The time constant R1C4 should be set so that SC current is shut down within 2 μ s. (1.5 μ s~2 μ s is general value.) SC interrupting time might vary with the wiring pattern, so the enough evaluation on the real system is recommended. If R1 is too small, it may leads to delay of protection. So R1 should be min. 10 times larger resistance than Rs. (100 times is recommended.)
- 5 :To prevent erroneous operation, the wiring of A, B, C should be as short as possible.
- 6 :For sense resistor, the variation within 1%(including temperature characteristics), low inductance type is recommended. And the over 0.03W is recommended, but it is necessary to evaluate in your real system finally.
- 7 :To prevent erroneous SC protection, the wiring from Vsc terminal to CIN filter should be divided at the point D that is close to the terminal of sense resistor. And the wiring should be patterned as short as possible.
- 8 :All capacitors should be mounted as close to the terminals of the DIIPM as possible. (C1: good temperature, frequency characteristic electrolytic type, and C2: 0.01 μ F~2.0 μ F, good temperature, frequency and DC bias characteristic ceramic type are recommended.)
- 9 :Input drive is High-active type. There is a min. 3.3k Ω pull-down resistor in the input circuit of IC. To prevent malfunction, the wiring of each input should be as short as possible. And it is recommended to insert RC filter (e.g. R3=100 Ω and C5=1000pF) and confirm the input signal level to meet the turn-on and turn-off threshold voltage. Thanks to HVIC inside the module, direct coupling to MCU without any opto-coupler or transformer isolation is possible.
- 10 :Fo output is open drain type. Fo output will be max 0.95V (@IfO=1mA, 25 $^{\circ}$ C), so it should be pulled up to MCU or control power supply (e.g. 5V, 15V) by a resistor that makes IfO up to 1mA. (In the case of pulled up to 5V, 10k Ω is recommended.)
- 11 :Error signal output width (tFO) can be set by the capacitor connected to CFO terminal. CFO(typ.) = tFO \times 9.1 \times 10 $^{-6}$ (F)
- 12 :If high frequency noise superimposed to the control supply line, IC malfunction might happen and cause erroneous operation. To avoid such problem, voltage ripple of control supply line should meet dV/dt \leq \pm 1V/ μ s, Vripple \leq 2Vp-p.
- 13 :For DIIPM, it isn't recommended to drive same load by parallel connection with other phase IGBT or other DIIPM.

Fig. 9 Package Outlines

Dimensions in mm



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