

## Four Output Ultra-Low Additive Phase Noise PCIe, Gen 1 to 6, and UPI/QPI Fanout Buffer

#### **Features**

- One Differential Input which Accepts Any Differential Format
- · Four Differential HCSL Outputs
- Ultra-Low Additive Jitter: 33 fs (in 12 kHz to 20 MHz Integration Band at 491.52 MHz Clock Frequency)
- Supports Clock Frequencies from 0 MHz to 600 MHz
- Supports 2.5V or 3.3V Power Supplies for HCSL Outputs
- Embedded Low Drop Out (LDO) Voltage Regulator Provides Superior Power Supply Noise Rejection
- · Maximum Output to Output Skew of 50 ps
- Individual Output Enable Pin for Each Differential Pair
- · Transfers Spread-Spectrum without Attenuation

#### **Applications**

- PCI Express Generation 1/2/3/4/5/6 Clock Distribution
- · UPI/QPI Clock Distribution
- · Low Jitter Clock Trees
- · Logic Translation
- · Clock and Data Signal Restoration
- High-Performance Microprocessor Clock Distribution
- · Test Equipment

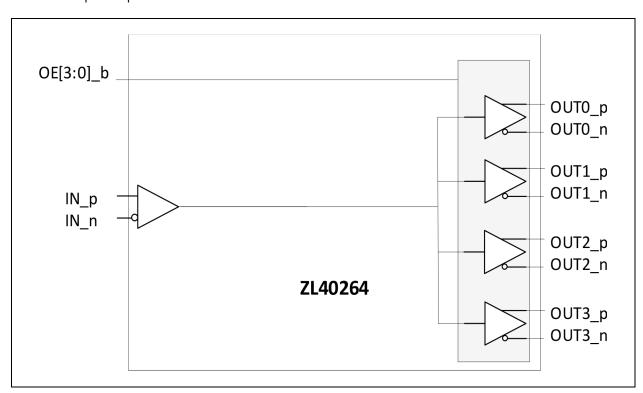


FIGURE 0-1: Functional Block Diagram.

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## 1.0 PIN DESCRIPTION AND CONFIGURATION

The device is packaged in a 4 mm × 4 mm 20-lead VQFN with 2.125 mm exposed pad.

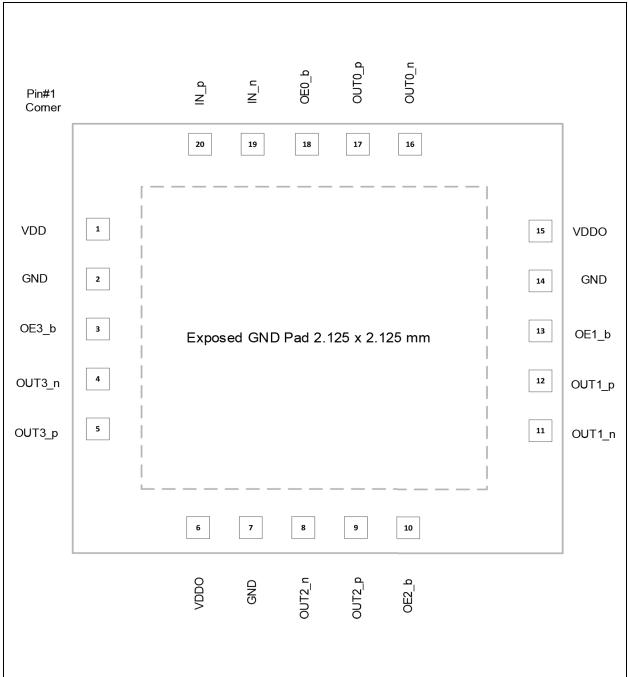


FIGURE 1-1: 20-Lead 4 mm x 4 mm VQFN Pin Diagram.

## 1.1 Pin Descriptions

All device inputs and outputs are HCSL unless described otherwise.

The I/O column uses the following symbols: I – input, IPU – input with 300 k $\Omega$  internal pull-up resistor, I<sub>PD</sub> – input with 300 k $\Omega$  internal pull-down resistor, I<sub>APU</sub> – input with 31 k $\Omega$  internal pull-up resistor, I<sub>APD</sub> – input with 30 k $\Omega$  internal pull-down resistor, I<sub>APU/APD</sub> – input biased to VDD/2 with 60 k $\Omega$  internal pull-up and pull-down resistors (30 k $\Omega$  equivalent), O – output, I/O – Input/Output pin, NC – No connect pin, P – power supply pin.

TABLE 1-1: PIN DESCRIPTION

Pin Number	Pin Name	Туре	Description				
Input Refe	erence	•					
20	IN0_p		Input Differential or Single Ended Reference				
19	INO_n	I <sub>APD</sub> I <sub>APU/APD</sub>	Input frequency range 0 Hz to 600 MHz. Non-inverting inputs (_p) are pulled down with internal 30 k $\Omega$ pull-down resistors. Inverting inputs (_n) are pulled up and pulled down with 60 k $\Omega$ internal resistors (30 k $\Omega$ equivalent) to keep inverting input voltages at VDD/2 when inverting inputs are left floating (device fed with a single ended reference).				
Output CI	ocks		1				
17	OUT0_p						
16	OUT0_n						
12	OUT1_p						
11	OUT1_n	0	Ultra-Low Additive	e Jitter Differential HCSL Outputs 0 to 1			
9	OUT2_p		Output frequency range 0 MHz to 600 MHz.				
8	OUT2_n	=					
5	OUT3_p						
4	OUT3_n						
Control							
18	OE0_b			gic level on these pins enables/disables corresponding			
13	OE1_b		outputs.				
10	OE2 h		OEn_b	OUTn_p/n			
10	OE2_b	I <sub>PD</sub>	0	Active			
3	OE3_b		High-Z (outputs p/n will be low/low because of 50 Ω shunt resistors—see recommended output termination)				
Power an	d Ground		•				
1	VDD	Р	Positive Supply V	oltage. Connect to 3.3V or 2.5V supply.			
6			Positive Supply Voltage for Differential Outputs. Connect 3.3V or 2.5V				
15	VDDO	Р	power supply. VDDO does not have to be connected to the same voltage level as VDD.				
2 7 14 E-Pad	GND	Р	Ground. Connect t	o the ground.			

7	1 4	02	64	
		UL	$\mathbf{v}$	•

NOTES:

#### 2.0 FUNCTIONAL DESCRIPTION

The ZL40264 is an ultra-low additive jitter, low power 1 to 4 HCSL fanout buffer.

The device operates from  $2.5V\pm5\%$  or  $3.3V\pm5\%$  supply. Its operation is guaranteed over the industrial temperature range -40°C to +85°C.

#### 2.1 Clock Inputs

The following blocks diagram shows how to terminate different signals fed to the ZL40264 inputs.

Figure 2-1 and Figure 2-2 show how to terminate the input when driven from an HCSL driver.

The input buffer in ZL40264 in a native HCSL receiver so other differential formats need to be AC coupled as shown in Figure 2-3 and Figure 2-4 for LVPECL and LVDS signals respectively.

The resistors R1, R2, R3, and R4 in Figure 2-3 are used for biasing, termination and hysteresis. The resistor pairs (R1, R2) and (R3, R4) are different to add 20 mV to 30 mV of hysteresis. The hysteresis prevents the device from generating random transitions when input is not driven. If the input is driven all the time with a clock, then pairs (R1, R2) and (R2, R4) can be identical. That is, R3 = R1 and R4 = R2.

The resistors R1, R2, R3, and R4 in Figure 2-4 are used for biasing and hysteresis. Same as above, the hysteresis prevents the device from generating random transitions when input is not driven. If the input is driven all the time with a clock, then pairs (R1, R2) and (R2, R4) can be identical. That is, R3 = R1 and R4 = R2. The  $100\Omega$  termination resistor is placed before AC coupling capacitors, because some LVDS drivers need a DC path between p and n outputs. If the LVDS driver does not need the DC path, the  $100\Omega$  resistor can be placed after the AC coupling capacitors.

Figure 2-5 shows how to terminate a single ended output such as LVCMOS. Ideally, resistors R1 and R2 should be  $100\Omega$  each and  $R_O$  +  $R_S$  should be  $50\Omega$  so that the transmission line is terminated at both ends with characteristic impedance.

If the driving strength of the output driver is not sufficient to drive low impedance, the value of series resistor  $R_S$  should be increased. This will reduce the voltage swing at the input but this should be fine as long as the input voltage swing requirement is not violated (Table 3-4). The source resistors of  $R_S = 270\Omega$  could be used for standard LVCMOS driver. This will provide 516 mV of voltage swing for 3.3V LVCMOS driver with load current of  $(3.3V/2) \times (1/(270\Omega + 50\Omega)) = 5.16$  mA.

For optimum performance, both differential input pins (\_p and \_n) need to be DC biased to the same voltage. Hence, the ratio R1/R2 should be equal to the ratio R3/R4.

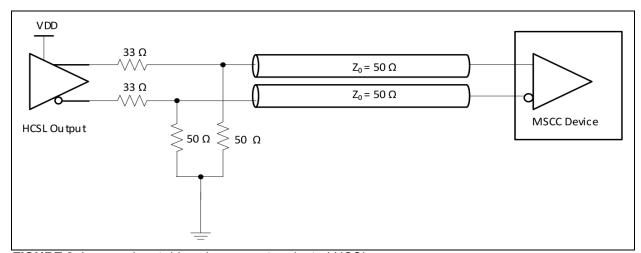


FIGURE 2-1: Input driven by source terminated HCSL.

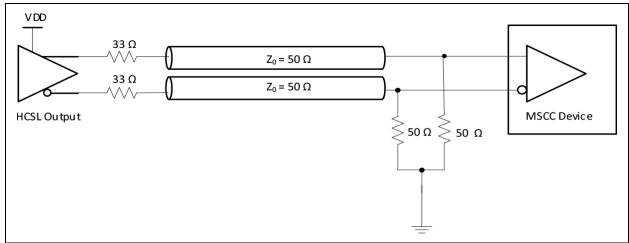


FIGURE 2-2: Input driven by receiver terminated HCSL.

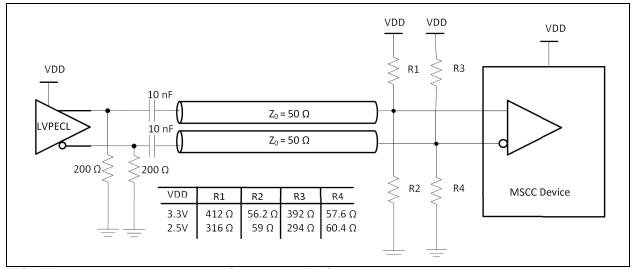


FIGURE 2-3: Input driven by AC-coupled LVPECL output.

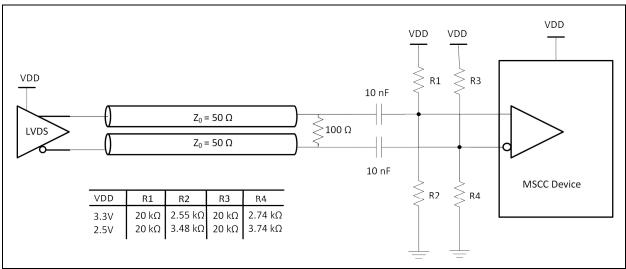


FIGURE 2-4: Input driven by AC-coupled LVDS.

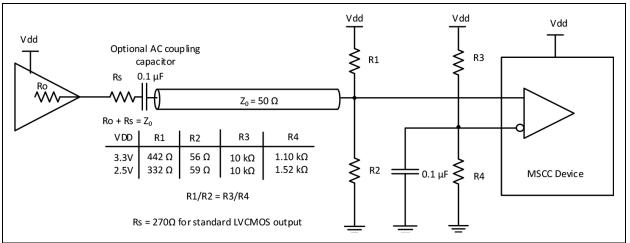


FIGURE 2-5: Input driven by a single-ended output.

## 2.2 Clock Outputs

Differential HCSL outputs should be terminated as shown in Figure 2-6 or Figure 2-7.

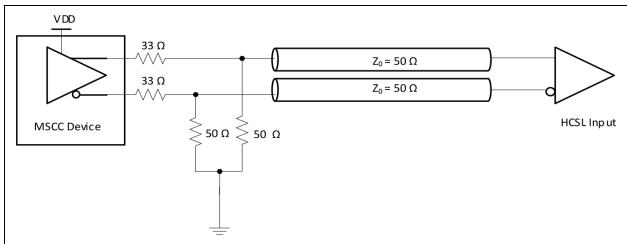


FIGURE 2-6: Source-terminated HCSL.

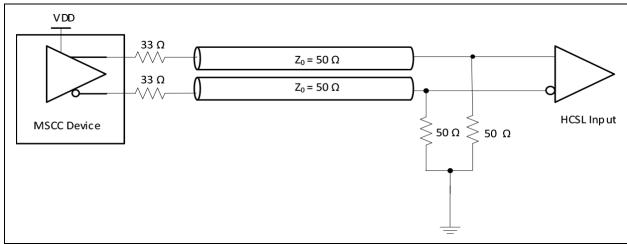


FIGURE 2-7: Receiver-terminated HCSL.

### 2.3 Termination of Unused Inputs and Outputs

Unused outputs should be left unconnected.

### 2.4 Power Consumption

The device total power consumption can be calculated as:

#### **EQUATION 2-1:**

$$P_T = P_S + P_C + P_{O\_DIFF}$$

Where:

$P_S = V_{DD} \times I_S$	Core power consumed by the input buffer. The static current (IS) is specified in Table 3-3.
$P_{C} = V_{DDO} \times I_{DD\_CM}$	Common output power shared among four outputs. The current IDD_CM is specified in Table 3-3.
$P_{O\_DIF} = V_{DDO} \times I_{DD\_HCSL} \times N$	Output power where output current per output (IDD_HCSL) is specified in Table 3-3. N is number of enabled outputs.

Power dissipated inside the device can be calculated by subtracting power dissipated in termination/biasing resistors from the power consumption:

#### **EQUATION 2-2:**

$$P_D = P_T - N \times P_{HCSL}$$

Where:

	$V_{SW}$ is voltage swing of HCSL output. $50\Omega$ is termination
$P_{HCSL} = (V_{SW} / 50\Omega)^2 \times (50\Omega + 33\Omega)$	resistance and $33\Omega$ is series resistance of the HCSL out-
	put.

#### 2.5 Power Supply Filtering

Each power pin (VDD and VDDO) should be decoupled with 0.1  $\mu$ F capacitor with minimum equivalent series resistance (ESR) and minimum series inductance (ESL). For example, 0402 X5R Ceramic Capacitors with 6.3V minimum rating could be used. These capacitors should be placed as close as possible to the power pins. To reduce the power noise from adjacent digital components on the board, each power supply could optionally be further insulated with low resistance ferrite bead with 10  $\mu$ F and 1  $\mu$ F capacitors. Figure 2-8 shows the standard and optional decoupling method.

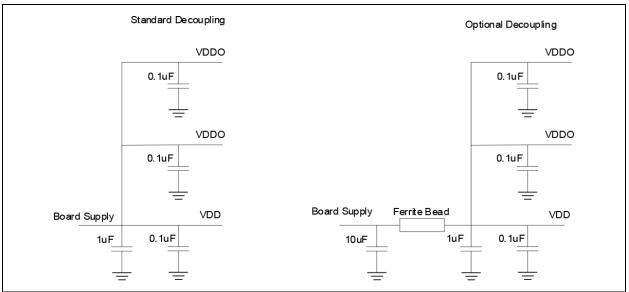


FIGURE 2-8: Power Supply Filtering.

## 2.6 Power Supplies and Power-up Sequence

The device has two different power supplies: VDD and VDDO which should always be connected to the same voltage supply. Voltages supported by each of these power supplies are specified in Table 3-2.

VDD and VDDO should always be turned on and off at the same time.

#### 2.7 Device Control

ZL40264 outputs are controlled via OE[3:0]\_b pins. When an OE\_b pin is low the corresponding outputs will be active and when this pin is high the output will be high-Z. When the output driver is in high-Z mode, the output pins will be pulled low via external  $50\Omega$  HCSL termination resistors.

### 2.8 Typical phase noise performance

The following plots show typical phase noise performance for 100 MHz, 133 MHz, 400 MHz, and 491.52 MHz clocks respectively.

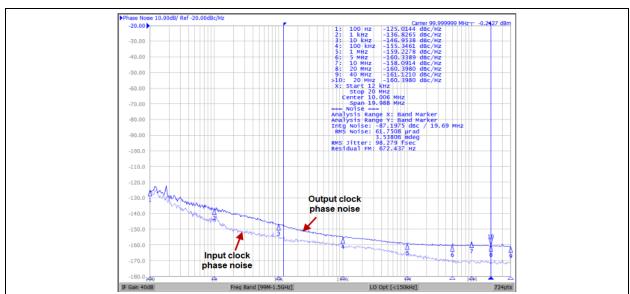


FIGURE 2-9: 100 MHz HCSL Phase Noise.

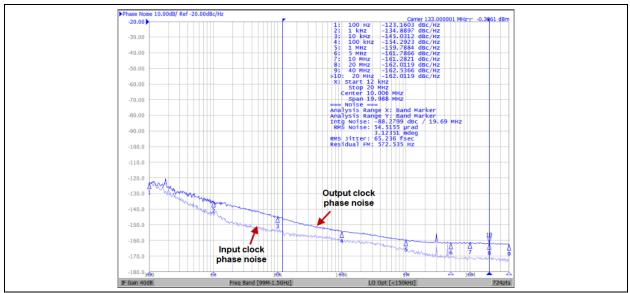


FIGURE 2-10: 133 MHz HCSL Phase Noise.

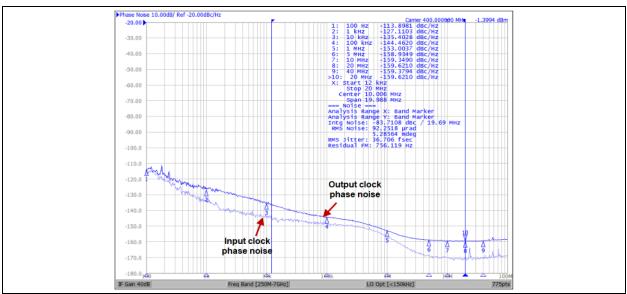


FIGURE 2-11: 400 MHz HCSL Phase Noise.

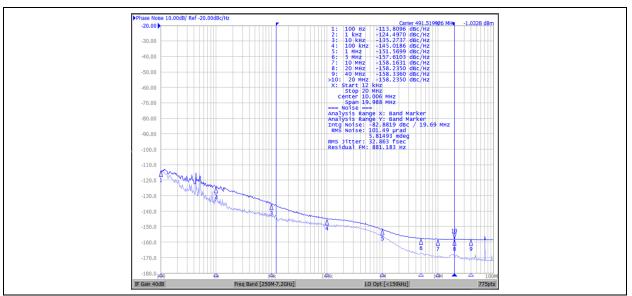


FIGURE 2-12: 491.52 MHz HCSL Phase Noise.

7	1 4	02	64	
		UL	$\mathbf{v}$	•

NOTES:

#### 3.0 AC AND DC ELECTRICAL CHARACTERISTICS

TABLE 3-1: ABSOLUTE MAXIMUM RATINGS

(Note 1, Note 2, Note 3)							
Parameter	Symbol	Min.	Max.	Units			
Supply Voltage, 3.3V	V <sub>DD</sub> /V <sub>DDO</sub>	-0.5	4.6	V			
Supply Voltage, 2.5V	V <sub>DD</sub> /V <sub>DDO</sub>	-0.5	3.5	V			
Storage Temperature Range	T <sub>ST</sub>	<b>–</b> 55	125	°C			

Note 1: Exceeding these values may cause permanent damage.

2: Functional operation under these conditions is not implied.

3: Voltages are with respect to ground (GND) unless otherwise stated.

TABLE 3-2: RECOMMENDED OPERATING RATINGS

(Note 1, Note 2)						
Parameter	Symbol	Min.	Тур.	Max.	Units	
Supply Voltage, 3.3V	$V_{DD}/V_{DDO}$	3.135	3.30	3.465	V	
Supply Voltage, 2.5V	$V_{DD}/V_{DDO}$	2.375	2.50	2.625	V	
Operating Temperature	T <sub>A</sub>	-40	25	85	°C	
Input Voltage	V <sub>DD-IN</sub>	-0.3	_	V <sub>DD</sub> + 0.3	V	

Note 1: Voltages are with respect to ground (GND) unless otherwise stated.

2: The device core supports two power supply modes (3.3V and 2.5V).

TABLE 3-3: CURRENT CONSUMPTION

Parameter	Symbol	Min	Тур.	Max	Units	Condition
Core device current	I <sub>S_3.3V</sub>	_	49	53	mA	V <sub>DD</sub> = 3.3V+5%
Core device current	I <sub>S_2.5V</sub>	_	48	53	mA	V <sub>DD</sub> = 2.5V+5%
Common output current	I <sub>DD_CM_3.3V</sub>	_	5.24	5.82	mA	V <sub>DDO</sub> = 3.3V+5%
	I <sub>DD_CM_2.5V</sub>	_	4.72	5.32	mA	V <sub>DDO</sub> = 2.5V+5%
Current dissipation per HCSL output	I <sub>DD_HCSL_3.3V</sub>	_	14.92	17.18	mA	V <sub>DDO</sub> = 3.3V+5%
Current dissipation per most output	I <sub>DD_HCSL_2.5V</sub>	_	14.61	16.62	mA	V <sub>DDO</sub> = 2.5V+5%

**TABLE 3-4: INPUT CHARACTERISTICS** 

Note 1, Note 2, Note 3						
Parameter	Symbol	Min	Тур.	Max	Units	Condition
CMOS high-level input voltage for control	V <sub>CIH_3.3V</sub>	0.7 × V <sub>DD</sub>	_	_	V	V <sub>DD</sub> = 3.3V
inputs	V <sub>CIH_2.5V</sub>	0.8 × V <sub>DD</sub>	_	_	V	V <sub>DD</sub> = 3.3V
CMOS low-level input voltage for control inputs	V <sub>CIL</sub>		_	0.32 × V <sub>DD</sub>	V	_
CMOS input leakage current for control inputs (includes current due to pull down resistors)	I <sub>IL</sub>	-25	_	50	μA	$V_I = V_{DD}$ or $0V$
Differential input common mode voltage for IN_p/n	V <sub>CM</sub>	0.1	_	0.8	V	_
Differential input voltage for IN_p/n	V <sub>ID</sub>	0.2	_	VDD + 0.3	V	_
Differential input leakage current for IN_p/n (includes current due to pull-up and pull-down resistors)	I <sub>IL</sub>	-150	_	150	μA	V <sub>I</sub> = 2V or 0V
Single ended input voltage for IN_p	V <sub>SI</sub>	-0.3	_	2.7	V	V <sub>DD</sub> = 3.3V or 2.5V
Single ended input common mode voltage IN_p	V <sub>SIC</sub>	0.1	_	0.8	V	V <sub>DD</sub> = 3.3V or 2.5V
Single ended input voltage swing for IN_p	$V_{SID}$	0.3		1.3	V	V <sub>DD</sub> = 3.3V or 2.5V
Input frequency (differential)	f <sub>IN</sub>	0	_	600	MHz	_
Input duty cycle (400 MHz input clock)	dc	35		65	%	Note 4
Input slew rate	slew	0.6	2		V/ns	_
Input pull-up/pull-down resistance	R <sub>PU</sub> /R <sub>PD</sub>		60	_	kΩ	_
Input pull-down resistance for IN_p	R <sub>PD</sub>	_	30	_	kΩ	_
Control input (OE_b) pull-down resistance	R <sub>PDOE</sub>	_	300	_	kΩ	_

Note 1: Values are over recommended operating conditions.

<sup>2:</sup> Values are over all two power supply modes ( $V_{DD}$  = 3.3V and  $V_{DD}$  = 2.5V).

<sup>3:</sup> Low frequency only.

**<sup>4:</sup>** Minimum and maximum duty cycles should be scaled for different input frequencies. For example, a 10 MHz input clock would have the minimum duty cycle of 1% and the maximum duty cycle of 99%.

TABLE 3-5: POWER SUPPLY REJECTION RATIO FOR VDD = VDDO = 3.3V

Note 1, Note 2, Note 3								
Parameter	Symbol	Min	Тур.	Max	Units	Condition		
			-80.7			f <sub>IN</sub> = 100 MHz		
PSRR for HCSL output	PSRR <sub>HCSL</sub>	_	-76.4	_	dBc	f <sub>IN</sub> = 133 MHz		
			-66.5			f <sub>IN</sub> = 400 MHz		

- Note 1: Values are over recommended operating conditions.
  - 2: Noise injected to  $V_{DD}/V_{DDO}$  power supply with frequency 100 kHz and amplitude 100 mVpp.
  - 3: PSRR is measured as amplitude of 100 kHz spur in dBc on the output clock phase noise plot.

TABLE 3-6: POWER SUPPLY REJECTION RATIO FOR VDD = VDDO = 2.5V

Note 1, Note 2, Note 3								
Parameter	Symbol	Min	Тур.	Max	Units	Condition		
			-73.5			f <sub>IN</sub> = 100 MHz		
PSRR for HCSL output	output PSRR <sub>HCSL</sub> — -69.8 — dBc	dBc	f <sub>IN</sub> = 133 MHz					
			-61.2			f <sub>IN</sub> = 400 MHz		

- **Note 1:** Values are over recommended operating conditions.
  - 2: Noise injected to  $V_{DD}/V_{DDO}$  power supply with frequency 100 kHz and amplitude 100 mVpp.
  - 3: PSRR is measured as amplitude of 100 kHz spur in dBc on the output clock phase noise plot.

TABLE 3-7: HCSL OUTPUTS FOR VDDO = 3.3V

Note 1						
Parameter	Symbol	Min	Тур.	Max	Units	Condition
Rising edge rate	rise_rate	1.3	1.7	2	V/ns	Note 3, Note 4
Falling edge rate	fall_rate	1.3	1.7	2	V/ns	Note 3, Note 4
Differential high voltage	$V_{IH}$	0.6	_	0.9	V	Note 3
Differential low voltage	$V_{IL}$	-0.9	_	-0.6	V	Note 3
Single ended high voltage	V <sub>SIH</sub>	0.6	0.74	0.85	V	DC Measurement
Single ended low voltage	$V_{SIL}$	-0.01	0	0.01	V	DC Measurement
Absolute crossing voltage	V <sub>CROSS</sub>	0.26	0.32	0.38	V	Note 2, Note 5, Note 6
Variation of V <sub>CROSS</sub> over all rising clock edges	ΔV <sub>CROSS</sub>	0.039	0.050	0.061	V	Note 2, Note 5, Note 10
Ring back voltage margin	$V_{RB}$	0.534	0.674	0.809	V	Note 3, Note 12
Time before V <sub>RB</sub> is allowed	t <sub>STABLE</sub>	4.6	_		ns	Note 3, Note 12
Cycle-to-cycle additive jitter	T <sub>JCC</sub>	_	6.5	8.1	ps peak- to-peak	Note 3
Absolute maximum voltage	$V_{MAX}$	_	_	0.92	_	Note 2, Note 8
Absolute minimum voltage	$V_{MIN}$	-0.05	_		_	Note 2, Note 9
Output duty-cycle (when input has 50% duty-cycle)	duty_cycle	48	50	52	%	Note 3
Rising to falling edge matching	r/f match			15	%	Note 2, Note 13
Clock source DC impedance (CK)	Z <sub>C-DC_CK</sub>	49	50	51	Ω	DC Measurement Note 2, Note 14)
Clock source DC impedance (CK#)	Z <sub>C-DC_CK#</sub>	49	50	51	Ω	DC Measurement Note 2, Note 14
Output frequency	F <sub>MAX</sub>	0	_	600	MHz	_

TABLE 3-7: HCSL OUTPUTS FOR VDDO = 3.3V (CONTINUED)

Note 1								
Parameter	Symbol	Min	Тур.	Max	Units	Condition		
Output-to-output skew	toosk	_	_	50	ps	_		
Device-to-device output skew	t <sub>DOOSK</sub>	_	_	129	ps	_		
Input-to-output delay	t <sub>IOD</sub>	0.75	0.84	1	ns	_		
Output enable time	t <sub>EN</sub>	_	_	3	cycles	_		
Output disable time	t <sub>DIS</sub>	_	_	2	cycles	_		

- Note 1: Values are over recommended operating conditions.
  - 2: Measurement taken from single ended waveform.
  - 3: Measurement taken from differential waveform.
  - **4:** Measured from -150 mV to +150 mV on the differential waveform (derived from CK minus CK#) The signal must be monotonic through the measurement region for rise and fall time. The 300 mV measurement window is centered on the differential zero crossing. See Figure 3-4.
  - 5: Measured at crossing point where the instantaneous voltage value of the rising edge of CK equals the falling edge of CK#. See Figure 3-1.
  - **6:** Refers to the total variation from the lowest crossing point to the highest, regardless of which edge is crossing. Refers to all crossing points for this measurement. See Figure 3-1.
  - 7: This requirement—from PCI Express Base Specification, Revision 4.0—is applicable only to clock generators and not to buffers. A clock buffer is a transparent device whose output clock period follows the input clock period.
  - 8: Defined as the maximum instantaneous voltage including overshoot. See Figure 3-1.
  - 9: Defined as the minimum instantaneous voltage including undershoot. See Figure 3-1.
  - **10:** Defined as the total variation of all crossing voltages of Rising CK and Falling CK# This is the maximum allowed variance in V<sub>CROSS</sub> for any particular system. See Figure 3-2.
  - **11:** The PPM requirement—from PCIe Express Base Specification, Revision 4.0—is related to clock generation devices. This requirement is not applicable to buffers because buffer's output frequency accuracy is identical to the frequency accuracy of the source driving the buffer.
  - 12: The t<sub>STABLE</sub> is the time the differential clock must maintain a minimum ±150 mV differential voltage after 20 rising/falling edges before it is allowed to droop back into the V<sub>RB</sub> ±100 mV differential range. See Figure 3-5.
  - 13: Matching applies to rising edge rate for CKx and falling edge rate for CK#x. It is measured using a ±75 mV window centered on the median cros point where CKx rising meets CK#x falling. The median cross point is used to calculate the voltage thresholds the oscilloscope is to use for the edge rate calculations. The Rise Edge Rate of CKx should be compared to the Fall Edge Rate of CK#x the maximum allowed difference should not exceed 20% of the slowest edge rate. See Figure 3-3.
  - **14:** Clock DC impedance tolerance depends only on the tolerance of external 50Ω shunt resistors used in HCSL. The test used resistors with ±1% tolerance.

TABLE 3-8: HCSL (PCIE) JITTER PERFORMANCE FOR VDDO = 3.3V

Note 1									
Parameter	Symbol	Min	Тур.	Max	Units	Condition			
Additive Jitter as per PCle 1.0 (1.5 MHz to 22 MHz)	T <sub>PCle_1.0</sub>	_	1.2	1.45	ps pk-pk	Input clock: 100 MHz			
Additive Jitter as per PCle 2.0 high band (1.5 MHz to 50 MHz)	T <sub>PCle_2.0_high</sub>	_	134	163	fs RMS	Input clock: 100 MHz			
Additive Jitter as per PCle 2.0 low band (10 kHz to 1.5 MHz)	T <sub>PCle_2.0_low</sub>	_	31	48	fs RMS	Input clock: 100 MHz			
Additive Jitter as per PCle 2.0 mid band (5 MHz to 16 MHz)	T <sub>PCle_2.0_mid</sub>		105	130	fs RMS	Input clock: 100 MHz			

TABLE 3-8: HCSL (PCIE) JITTER PERFORMANCE FOR VDDO = 3.3V (CONTINUED)

T <sub>PCle_3.0</sub>	_	33	41	fs RMS	Input clock: 100 MHz
T <sub>PCle_4.0</sub>	_	33	41	fs RMS	Input clock: 100 MHz
T <sub>PCle_5.0</sub>	_	13	16	fs RMS	Input clock: 100 MHz
T <sub>PCle_6.0</sub>	_	8	10	fs RMS	Input clock: 100 MHz
T <sub>PCle</sub>	_	61	75	fs RMS	Input clock: 100 MHz
T <sub>j_1M_20M</sub>	_	87	106		Input clock: 100 MHz
		56	68	fe DMS	Input clock: 133 MHz
		26	34	18 IXIVIS	Input clock: 400 MHz
		25	32		Input clock: 491.52 MHz
		91	112		Input clock: 100 MHz
т	_	60	75	fe DMS	Input clock: 133 MHz
' j_12k_20M		32	48	13 1(1010	Input clock: 400 MHz
		30	43		Input clock: 491.52 MHz
	_	-161	-159		Input clock: 100 MHz
NE		-162	-161	dBc/Hz	Input clock: 133 MHz
INF		-160	-157	UDG/11Z	Input clock: 400 MHz
	_	-158	-155		Input clock: 491.52 MHz
	T <sub>PCle_4.0</sub> T <sub>PCle_5.0</sub> T <sub>PCle_6.0</sub>	T <sub>PCle_4.0</sub> —  T <sub>PCle_5.0</sub> —  T <sub>PCle_6.0</sub> —  T <sub>pCle_6.0</sub> —  T <sub>j_1M_20M</sub> —  T <sub>j_12k_20M</sub> —  NF	T <sub>PCle_4.0</sub> — 33  T <sub>PCle_5.0</sub> — 13  T <sub>PCle_6.0</sub> — 8  T <sub>PCle_6.0</sub> — 61	T <sub>PCle_4.0</sub> — 33 41  T <sub>PCle_5.0</sub> — 13 16  T <sub>PCle_6.0</sub> — 8 10  T <sub>PCle_6.0</sub> — 61 75  — 87 106  — 56 68  — 26 34  — 25 32  — 91 112  — 60 75  — 32 48  — 30 43  NF  NF  NF	T <sub>PCle_4.0</sub> — 33 41 fs RMS  T <sub>PCle_5.0</sub> — 13 16 fs RMS  T <sub>PCle_6.0</sub> — 8 10 fs RMS  T <sub>PCle_6.0</sub> — 61 75 fs RMS  T <sub>pCle</sub> — 61 75 fs RMS  - 87 106 - 56 68 - 26 34 - 25 32  T <sub>j_11k_20M</sub> — 91 112 - 60 75 - 32 48 - 30 43  NF  NF  - 162 -161 - 159160 -157

Note 1: Values are over recommended operating conditions.

TABLE 3-9: HCSL OUTPUTS FOR VDDO = 2.5V

Note 1								
Parameter	Symbol	Min	Тур.	Max	Units	Condition		
Rising edge rate	rise_rate	1.3	1.6	1.9	V/ns	Note 2, Note 3		
Falling edge rate	fall_rate	1.3	1.6	1.9	Vms	Note 2, Note 3		
Differential high voltage	V <sub>IH</sub>	0.6	_	0.9	V	Note 2		
Differential low voltage	$V_{IL}$	-0.9	_	-0.6	V	Note 2		
Single-ended high voltage	V <sub>SIH</sub>	0.58	0.71	0.84	V	DC measurement		
Single-ended low voltage	V <sub>SIL</sub>	-0.01	0	0.01	V	DC measurement		
Absolute crossing voltage	V <sub>CROSS</sub>	0.25	0.31	0.37	V	Note 1, Note 4, Note 5		
Variation of V <sub>CROSS</sub> over all rising clock edges	ΔV <sub>CROSS</sub>	0.04	0.05	0.06	V	Note 1, Note 4, Note 9		
Ring back voltage margin	$V_{RB}$	0.514	0.660	0.791	V	Note 2, Note 11		
Time before VRB is allowed	t <sub>STABLE</sub>	4.6	_	_	ns	Note 2, Note 11		
Additive cycle-to-cycle jitter	T <sub>JCC</sub>	_	5.5	7.1	ps peak- to-peak	Note 2		
Absolute maximum voltage	V <sub>MAX</sub>	_	_	0.90		Note 1, Note 7		
Absolute minimum voltage	V <sub>MIN</sub>	-0.05	_	_		Note 1, Note 8		
Output duty-cycle (when input has 50% duty-cycle)	duty_cycle	48	50	52	%	Note 2		

TABLE 3-9: HCSL OUTPUTS FOR VDDO = 2.5V (CONTINUED)

Note 1									
Parameter	Symbol	Min	Тур.	Max	Units	Condition			
Rising to falling edge matching	r/f match	_	_	15	%	Note 1, Note 12			
Clock source DC impedance (CK)	Z <sub>C-DC_CK</sub>	49	50	51	Ω	DC measurement, Note 1, Note 13			
Clock source DC impedance (CK#)	Z <sub>C-DC_CK#</sub>	0	50	51	Ω	DC measurement, Note 1, Note 13			
Output frequency	F <sub>MAX</sub>	_	_	600	MHz	_			
Output-to-output skew	t <sub>oosk</sub>	_		50	ps	_			
Device-to-device output skew	t <sub>DOOSK</sub>	0.75	_	129	ps	_			
Input-to-output delay	t <sub>OPD</sub>	_	0.85	1	ns	_			
Output enable time	t <sub>EN</sub>	_	_	3	cycles	_			
Output disable time	t <sub>DIS</sub>	_	_	3	cycles	_			

- Note 1: Values are over recommended operating conditions.
  - 2: Measurement taken from single ended waveform.
  - 3: Measurement taken from differential waveform.
  - 4: Measured from -150 mV to +150 mV on the differential waveform (derived from CK minus CK#) The signal must be monotonic through the measurement region for rise and fall time. The 300 mV measurement window is centered on the differential zero crossing. See Figure 3-4.
  - 5: Measured at crossing point where the instantaneous voltage value of the rising edge of CK equals the falling edge of CK#. See Figure 3-1.
  - **6:** Refers to the total variation from the lowest crossing point to the highest, regardless of which edge is crossing. Refers to all crossing points for this measurement. See Figure 3-1.
  - 7: This requirement—from PCI Express Base Specification, Revision 4.0—is applicable only to clock generators and not to buffers. A clock buffer is a transparent device whose output clock period follows the input clock period.
  - 8: Defined as the maximum instantaneous voltage including overshoot. See Figure 3-1.
  - 9: Defined as the minimum instantaneous voltage including undershoot. See Figure 3-1.
  - **10:** Defined as the total variation of all crossing voltages of Rising CK and Falling CK# This is the maximum allowed variance in V<sub>CROSS</sub> for any particular system. See Figure 3-2.
  - **11:** The PPM requirement—from PCIe Express Base Specification, Revision 4.0—is related to clock generation devices. This requirement is not applicable to buffers because buffer's output frequency accuracy is identical to the frequency accuracy of the source driving the buffer.
  - 12: The t<sub>STABLE</sub> is the time the differential clock must maintain a minimum ±150 mV differential voltage after 20 rising/falling edges before it is allowed to droop back into the V<sub>RB</sub> ±100 mV differential range. See Figure 3-5.
  - 13: Matching applies to rising edge rate for CKx and falling edge rate for CK#x. It is measured using a ±75 mV window centered on the median cross point where CKx rising meets CK#x falling. The median cross point is used to calculate the voltage thresholds the oscilloscope is to use for the edge rate calculations. The Rise Edge Rate of CKx should be compared to the Fall Edge Rate of CK#x the maximum allowed difference should not exceed 20% of the slowest edge rate. See Figure 3-3.
  - 14: Clock DC impedance tolerance depends only on the tolerance of external  $50\Omega$  shunt resistors used in HCSL. The test used resistors with  $\pm 1\%$  tolerance.

TABLE 3-10: HCSL (PCIE) JITTER PERFORMANCE FOR VDDO = 2.5V

Note 1						- <del>-</del>
Parameter	Symbol	Min	Тур.	Max	Units	Condition
Additive Jitter as per PCle 1.0 (1.5 MHz to 22 MHz)	T <sub>jPCle_1.0</sub>	_	1.03	1.27	ps pk-pk	Input clock: 100 MHz
Additive Jitter as per PCle 2.0 high band (1.5 MHz to 50 MHz)	T <sub>jPCle_2.0_high</sub>	_	115	143	fs RMS	Input clock: 100 MHz
Additive Jitter as per PCle 2.0 low band (10 kHz to 1.5 MHz)	T <sub>jPCle_2.0_low</sub>	_	28	46	fs RMS	Input clock: 100 MHz
Additive Jitter as per PCle 2.0 mid band (5 MHz to 16 MHz)	T <sub>jPCle_2.0_mid</sub>	_	91	113	fs RMS	Input clock: 100 MHz
Additive Jitter as per PCle 3.0 (PLL_BW = 2 MHz to 5 MHz, CDR = 10 MHz)	T <sub>jPCle_3.0</sub>	_	29	36	fs RMS	Input clock: 100 MHz
Additive Jitter as per PCle 4.0 (PLL_BW = 2 MHz to 5 MHz, CDR = 10 MHz)	T <sub>jPCle_4.0</sub>	_	29	36	fs RMS	Input clock: 100 MHz
Additive Jitter as per PCle 5.0 (PLL_BW = 0.5 MHz to 1.8 MHz, CDR for 32 GT/s CC)	T <sub>jPCle_5.0</sub>	_	11	14	fs RMS	Input clock: 100 MHz
Additive Jitter as per PCle 6.0 (PLL_BW = 0.5 MHz to 1 MHz, CDR for 64 GT/s CC)	T <sub>jPCle_6.0</sub>	_	7	9	fs RMS	Input clock: 100 MHz
Additive Jitter as per Intel QPI 9.6 Gbps	T <sub>jQPI</sub>	_	53	67	fs RMS	Input clock: 100 MHz
		_	75	94		Input clock: 100 MHz
Additive RMS Jitter in 1 MHz to	_	_	51	64	fs RMS	Input clock: 133 MHz
20 MHz band	T <sub>j_1M_20M</sub>	_	26	33	15 INIO	Input clock: 400 MHz
		_	25	32		Input clock: 491.52 MHz
		_	79	99		Input clock: 100 MHz
Additive RMS Jitter in 12 kHz to	T	_	55	68	fs RMS	Input clock: 133 MHz
20 MHz band	T <sub>j_12k_20M</sub>	_	32	47	IS INIVIO	Input clock: 400 MHz
			30	43		Input clock: 491.52 MHz
		_	-162	-159		Input clock: 100 MHz
Noise floor	N <sub>F</sub>	_	-163	-161	dBc/Hz	Input clock: 133 MHz
INOISE HOOF	INE	_	-160	-158	UDG/TIZ	Input clock: 400 MHz
		_	-158	-155		Input clock: 491.52 MHz

**Note 1:** Values are over recommended operating conditions.

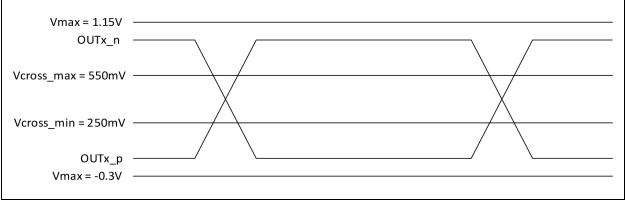


FIGURE 3-1: Single-Ended Measurement Points for Absolute Cross Point and Swing.

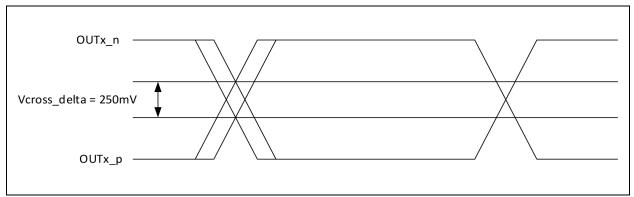


FIGURE 3-2: Single-Ended Measurement Points for Delta Cross Point.

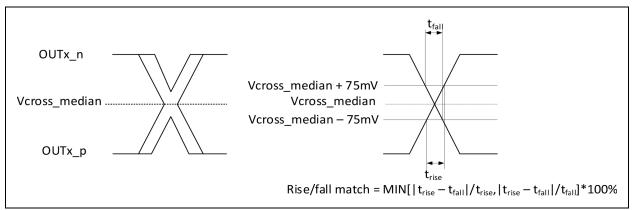


FIGURE 3-3: Single-Ended Measurement Points for Rise and Fall Time Matching.

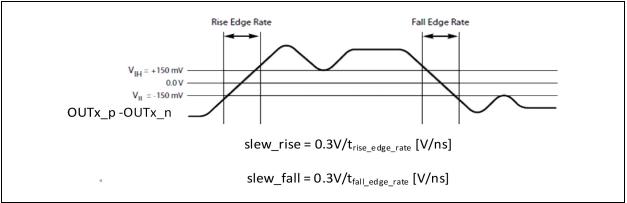


FIGURE 3-4: Differential Measurement Points for Rise and Fall Time.

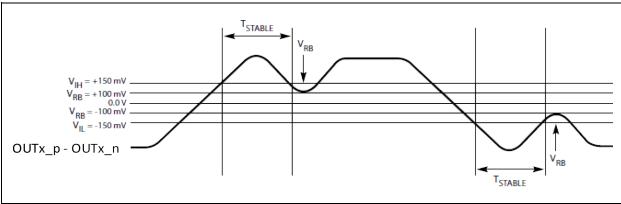


FIGURE 3-5: Differential Measurement Points for Ringback.

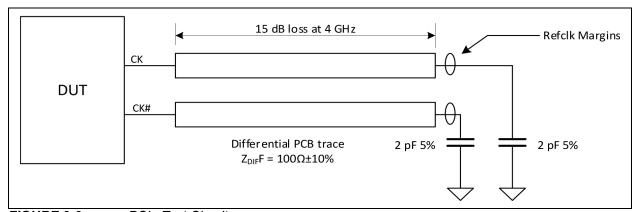


FIGURE 3-6: PCIe Test Circuit.

TABLE 3-11: 4 MM × 4 MM VQFN PACKAGE THERMAL PROPERTIES

Parameter	Symbol	Value	Units	Condition
Maximum ambient temperature	T <sub>A</sub>	85	°C	_
Maximum junction temperature	$T_{JMAX}$	125	°C	_
		34		Still air
Junction-to-ambient thermal resistance (Note 1)	$\theta_{JA}$	28.9	°C/W	1 m/s airflow
		27.0		2.5 m/s airflow
Junction-to-board thermal resistance	$\theta_{JB}$	15.4	°C/W	_
Junction-to-case thermal resistance	$\theta_{JC}$	25.9	°C/W	_
Junction-to-pad thermal resistance	$\theta_{JP}$	8.1	°C/W	Still air
Junction-to-top-center thermal characterization parameter	$\Psi_{JT}$	1.0	°C/W	Still air

Note 1: Theta-JA  $(\theta_{JA})$  is the thermal resistance from junction to ambient when the package is mounted on a 4-layer JEDEC standard test board and dissipating maximum power.

TABLE 3-12: HCSL OUTPUT PHASE NOISE WITH 25 MHZ XTAL

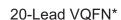
Note 1								
Parameter	Symbol	Min	Тур.	Max	Units	Condition		
Jitter RMS in 12 kHz to 5 MHz	т	1	235	_	fs	V <sub>DD</sub> = 3.3V, V <sub>DDO</sub> = 3.3V		
band	T <sub>J_12M_5M</sub>		143		2	$V_{DD} = 2.5V, V_{DDO} = 2.5V$		
			-102	_		V <sub>DD</sub> = 3.3V, V <sub>DDO</sub> = 3.3V @100 Hz		
		_	-126	_		$V_{DD} = 3.3V, V_{DDO} = 3.3V @1 \text{ kHz}$		
		_	-153	_		V <sub>DD</sub> = 3.3V, V <sub>DDO</sub> = 3.3V @10 kHz		
		_	-158	_		V <sub>DD</sub> = 3.3V, V <sub>DDO</sub> = 3.3V @100 kHz		
		_	-159	_		V <sub>DD</sub> = 3.3V, V <sub>DDO</sub> = 3.3V @1 MHz		
Noise floor	NI	_	-158	_	dBc/Hz	V <sub>DD</sub> = 3.3V, V <sub>DDO</sub> = 3.3V @5 MHz		
Noise 11001	N <sub>F</sub>	_	-97	_	ubc/HZ	V <sub>DD</sub> = 2.5V, V <sub>DDO</sub> = 2.5V @100 Hz		
		_	-123	_		V <sub>DD</sub> = 2.5V, V <sub>DDO</sub> = 2.5V @1 kHz		
		_	-153	_		V <sub>DD</sub> = 2.5V, V <sub>DDO</sub> = 2.5V @10 kHz		
		_	-162	_		V <sub>DD</sub> = 2.5V, V <sub>DDO</sub> = 2.5V @100 kHz		
		_	-162	_		V <sub>DD</sub> = 2.5V, V <sub>DDO</sub> = 2.5V @1 MHz		
		_	-163	_		V <sub>DD</sub> = 2.5V, V <sub>DDO</sub> = 2.5V @5 MHz		

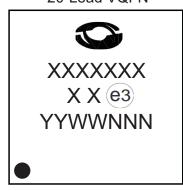
Note 1: Values are over recommended operating conditions.

<sup>2:</sup> Theta-JP ( $\theta_{JP}$ ) is the thermal resistance from junction to the center exposed pad on the bottom of the package).

#### 4.0 PACKAGE OUTLINE

### 4.1 Package Marking Information





#### Example



Legend: XX...X Product code or customer-specific information

Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')

NNN Alphanumeric traceability code

e3 Pb-free JEDEC® designator for Matte Tin (Sn)

This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

•, ▲, ▼ Pin one index is identified by a dot, delta up, or delta down (triangle mark).

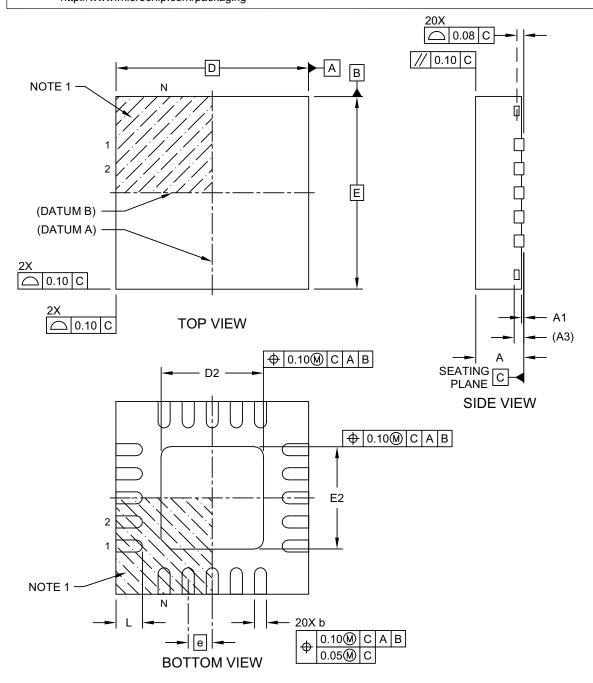
bte: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information. Package may or may not include the corporate logo.

Underbar ( ) and/or Overbar ( ) symbol may not be to scale.

**Note:** If the full seven-character YYWWNNN code cannot fit on the package, the following truncated codes are used based on the available marking space: 6 Characters = YWWNNN; 5 Characters = WWNNN; 4 Characters = WNNN; 3 Characters = NNN; 2 Characters = NN; 1 Character = N.

# 20-Lead 4 mm $\times$ 4 mm VQFN with 2.125 mm Exposed Pad (LWC) Package Outline and Recommended Landing Pattern

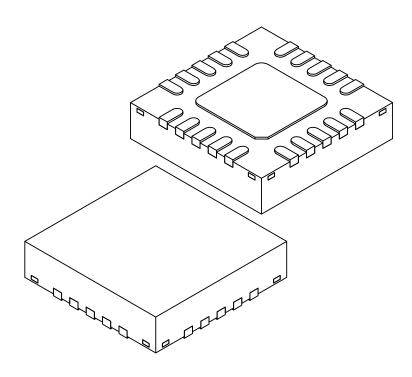
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-25395 Rev A Sheet 1 of 2

# 20-Lead 4 mm × 4 mm VQFN with 2.125 mm Exposed Pad (LWC) Package Outline and Recommended Landing Pattern

For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS				
Dimension	Limits	MIN	NOM	MAX		
Number of Terminals	N		20			
Pitch	е		0.50 BSC			
Overall Height	Α	0.80	0.90	1.00		
Standoff	A1	0.00	0.02	0.05		
Terminal Thickness	A3	0.20 REF				
Overall Length	D		4.00 BSC			
Exposed Pad Length	D2	2.025	2.125	2.225		
Overall Width	Е		4.00 BSC			
Exposed Pad Width	E2	2.025	2.125	2.225		
Terminal Width	b	0.20	0.25	0.32		
Terminal Length	L	0.45	0.55	0.65		

#### Notes:

Note:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Package is saw singulated
- 3. Dimensioning and tolerancing per ASME Y14.5M

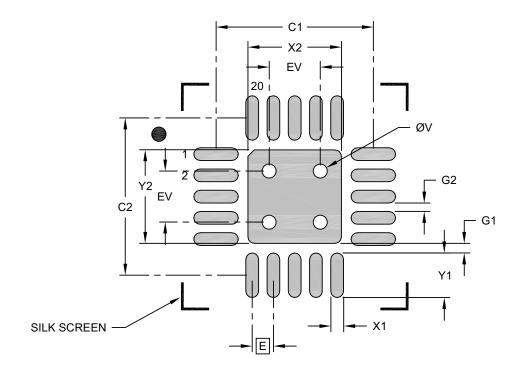
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-25395 Rev A Sheet 2 of 2

# 20-Lead 4 mm × 4 mm VQFN with 2.125 mm Exposed Pad (LWC) Package Outline and Recommended Landing Pattern

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



#### RECOMMENDED LAND PATTERN

	MILLIMETERS			
Dimension	MIN	NOM	MAX	
Contact Pitch	E	0.50 BSC		
Optional Center Pad Width	X2			2.20
Optional Center Pad Length	Y2			2.20
Contact Pad Spacing	C1		3.70	
Contact Pad Spacing	C2		3.70	
Contact Pad Width (X20)	X1			0.30
Contact Pad Length (X20)	Y1			1.05
Contact Pad to Center Pad (X20)	G1	0.23		
Contact Pad to Contact Pad (X16)	G2	0.20		
Thermal Via Diameter	V		0.33	
Thermal Via Pitch	EV		1.20	

#### Notes:

- Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.
- 2. For best soldering results, thermal vias, if used, should be filled or tented to avoid solder loss during reflow process

Microchip Technology Drawing C04-27395 Rev A

NOTES:

## APPENDIX A: DATA SHEET REVISION HISTORY

## **TABLE A-1: REVISION HISTORY**

Revision	Section/Figure/Entry	Correction		
DS20006788A (07-2024)	_	Converted Microsemi data sheet ZL40264 to Microchip DS20006788A. Figures 2-4 and 2-5 updated. Added Figure 2-12. Updated multiple tables throughout Section 3. Minor text and table changes throughout.		

NOTES:

## PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO. Device	XX X       Chip Carrier Package Type	XX Media Type	X       Finish	Examples	<b>s</b> :	
Device:	ZL40264: Four Output Ultra-Lo Gen 1 to 5, and UPI/ L = Leadless Chip Carrier		oise PCle,	a) ZL40264	LDG1:	Low Skew, Low Additive Jitter, Leadless Chip Carrier, 48-Lead VQFN Package, 490/Tray and Pb Free with Matte Sn Lead Finish Equating to RoHS e3
Type: Package: Media Type:	D = 20-Lead VQFN Package  G = 490/Tray	with E-Pad		b) ZL40264	LDF1:	Low Skew, Low Additive Jitter, Leadless Chip Carrier, 48-Lead VQFN Package, 4,000/Reel and Pb Free with Matte Sn Lead Fin- ish Equating to RoHS e3
Finish:	F = 4,000/Tape & Reel  1 = Pb Free with Matte Sn Lead Finish Equating to RoHS e3			Note 1: Tape and Reel identifier only appears in the catalog part number description. This identifier is used for ordering purposes and is not printed on the device package. Check with your Microchip Sales Office for package availability with the Tape and Reel option.		

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- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of its code. Code protection does not
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  continuously improving the code protection features of our products.

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