

### 6/8-Pin Flash-Based, 8-Bit Microcontrollers

#### High-Performance RISC CPU

- Only 35 Instructions to Learn:
- All single-cycle instructions, except branches
- Operating Speed:
- DC 16 MHz clock input
- DC 250 ns instruction cycle
- Eight-Level Deep Hardware Stack
- Interrupt Capability
- Processor Self-Write/Read access to Program Memory
- Pinout Compatible to other 6-Pin PIC10FXXX Microcontrollers

#### Memory

- Up to 512 Words of Flash Program Memory
- · 64 Bytes Data Memory
- High-Endurance Flash Data Memory (HEF)
  - 128B of nonvolatile data storage
  - 100K erase/write cycles

#### **Special Microcontroller Features**

- Low-Power 16 MHz Internal Oscillator:
  - Software selectable frequency range from 16 MHz to 31 kHz
  - Factory calibrated to  $\pm$  1%, typical
- Wide Operating Range:
  - 1.8V to 3.6V (PIC10LF320/322)
  - 2.3V to 5.5V (PIC10F320/322)
- Power-on Reset (POR)
- Power-up Timer (PWRT)
- Brown-out Reset (BOR)
- Ultra Low-Power Sleep Regulator
- Extended Watchdog Timer (WDT)
- Programmable Code Protection
- Power-Saving Sleep mode
- Selectable Oscillator Options (EC mode or Internal Oscillator)
- In-Circuit Serial Programming<sup>™</sup> (ICSP<sup>™</sup>) (via Two Pins)
- In-Circuit Debugger Support
- Fixed Voltage Reference (FVR) with 1.024V, 2.048V and 4.096V ('F' variant only) Output Levels
- Integrated Temperature Indicator
- 40-year Flash Data Retention

### eXtreme Low-Power (XLP) Features (PIC10LF320/322)

- · Sleep Current:
- 20 nA @ 1.8V, typical
- Operating Current:
- 25 μA @ 1 MHz, 1.8V, typical
- Watchdog Timer Current:
- 500 nA @ 1.8V, typical

#### **Peripheral Features**

- Four I/O Pins:
  - One input-only pin
  - High current sink/source for LED drivers
  - Individually selectable weak pull-ups
  - Interrupt-on-Change
- Timer0: 8-Bit Timer/Counter with 8-Bit Programmable Prescaler
- Timer2: 8-Bit Timer/Counter with 8-Bit Period Register, Prescaler and Postscaler
- Two PWM Modules:
- 10-bit PWM, max. frequency 16 kHz
- Combined to single two-phase output
- A/D Converter:
  - 8-bit resolution with three channels
- Configurable Logic Cell (CLC):
  - Eight selectable input source signals
  - Two inputs per module
  - Software selectable logic functions including: AND/OR/XOR/D Flop/D Latch/SR/JK
  - External or internal inputs/outputs
  - Operation while in Sleep
- Numerically Controlled Oscillator (NCO):
  - 20-bit accumulator
  - 16-bit increment
  - Linear frequency control
  - High-speed clock input
  - Selectable Output modes
    - Fixed Duty Cycle (FDC)
    - Pulse Frequency (PF) mode
- Complementary Waveform Generator (CWG):
- Selectable falling and rising edge dead-band control
  - Polarity control
- Two auto-shutdown sources
- Multiple input sources: PWM, CLC, NCO

#### PIC10(L)F320/322 Family Types

Device	Data Sheet Index	Program Memory Flash (words)	Data SRAM (bytes)	High Endurance Flash (bytes)	I/O's <sup>(2)</sup>	8-Bit ADC (ch)	Timers (8-Bit)	PWM	Complementary Wave Generator (CWG)	Configurable Logic Cell (CLC)	Fixed Voltage Reference (FVR)	Numerically Controlled Oscillator (NCO)	Debug <sup>(1)</sup>	XLP
PIC10(L)F320	(1)	256	64	128	4	3	2	2	1	1	1	1	Н	Y
PIC10(L)F322	(1)	512	64	128	4	3	2	2	1	1	1	1	Н	Y

**Note 1:** I - Debugging, Integrated on Chip; H - Debugging, Available using Debug Header; E - Emulation, Available using Emulation Header.

2: One pin is input-only.

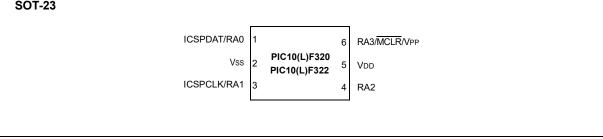
#### Data Sheet Index:

1: DS40001585 PIC10(L)F320/322 Data Sheet, 6/8 Pin High Performance, Flash Microcontrollers.

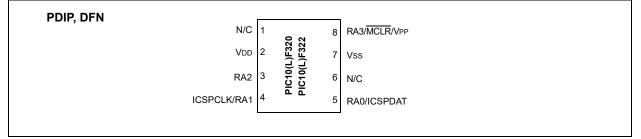
**Note:** For other small form-factor package availability and marking information, please visit http://www.microchip.com/packaging or contact your local sales office.

#### FIGURE 1: 6-PIN DIAGRAM, PIC10(L)F320/322

#### SOT-23



#### FIGURE 2: 8-PIN DIAGRAM, PIC10(L)F320/322



#### 6 AND 8-PIN ALLOCATION TABLE, PIC10(L)F320/322 TABLE 1:

I/O	6-Pin	8-Pin	Analog	Timer	PWM	Interrupts	Pull-ups	CWG	NCO	CLC	Basic	ICSP
RA0	1	5	AN0		PWM1	IOC0	Y	CWG1A	—	CLC1IN0	—	ICSPDAT
RA1	3	4	AN1		PWM2	IOC1	Y	CWG1B	NCO1CLK	CLC1	CLKIN	ICSPCLK
RA2	4	3	AN2	<b>T0CKI</b>		INT/IOC2	Y	CWG1FLT	NCO1	CLC1IN1	CLKR	
RA3	6	8				IOC3	Y		_		MCLR	Vpp
N/C	_	1	_		_	—	_		—	_	_	_
N/C	_	6	_	_	_	—	—	—	—	_	—	—
Vdd	5	2	_	_	_	_	_	_	_	_	Vdd	_
Vss	2	7	—	_	—	—	_	_	—	—	Vss	—

#### **Table of Contents**

1.0	Device Overview	6
2.0	Memory Organization	9
3.0	Device Configuration	. 19
4.0	Oscillator Module	. 24
5.0	Resets	. 28
6.0	Interrupts	. 35
7.0	Power-Down Mode (Sleep)	. 44
8.0	Watchdog Timer	. 46
9.0	Flash Program Memory Control	. 50
10.0	I/O Port	. 67
11.0	Interrupt-On-Change	. 73
12.0	Fixed Voltage Reference (FVR)	. 77
13.0	Internal Voltage Regulator (IVR)	. 79
14.0	Temperature Indicator Module	. 81
15.0	Analog-to-Digital Converter (ADC) Module	. 83
16.0	Timer0 Module	. 93
17.0	Timer2 Module	. 96
18.0	Pulse-Width Modulation (PWM) Module	. 98
19.0	Configurable Logic Cell (CLC)	104
20.0	Numerically Controlled Oscillator (NCO) Module	119
21.0	Complementary Waveform Generator (CWG) Module	129
22.0	In-Circuit Serial Programming ™ (ICSP ™)	144
23.0	Instruction Set Summary	
24.0	Electrical Specifications	156
25.0	DC and AC Characteristics Graphs and Charts	176
26.0	Development Support	177
27.0	Packaging Information	181
Appe	ndix A: Data Sheet Revision History	189

### TO OUR VALUED CUSTOMERS

It is our intention to provide our valued customers with the best documentation possible to ensure successful use of your Microchip products. To this end, we will continue to improve our publications to better suit your needs. Our publications will be refined and enhanced as new volumes and updates are introduced.

If you have any questions or comments regarding this publication, please contact the Marketing Communications Department via E-mail at docerrors@microchip.com. We welcome your feedback.

#### Most Current Data Sheet

To obtain the most up-to-date version of this data sheet, please register at our Worldwide Website at:

#### http://www.microchip.com

You can determine the version of a data sheet by examining its literature number found on the bottom outside corner of any page. The last character of the literature number is the version number, (e.g., DS30000000A is version A of document DS30000000).

#### Errata

An errata sheet, describing minor operational differences from the data sheet and recommended workarounds, may exist for current devices. As device/documentation issues become known to us, we will publish an errata sheet. The errata will specify the revision of silicon and revision of document to which it applies.

To determine if an errata sheet exists for a particular device, please check with one of the following:

- Microchip's Worldwide Website; http://www.microchip.com
- Your local Microchip sales office (see last page)

When contacting a sales office, please specify which device, revision of silicon and data sheet (include literature number) you are using.

#### **Customer Notification System**

Register on our website at www.microchip.com to receive the most current information on all of our products.

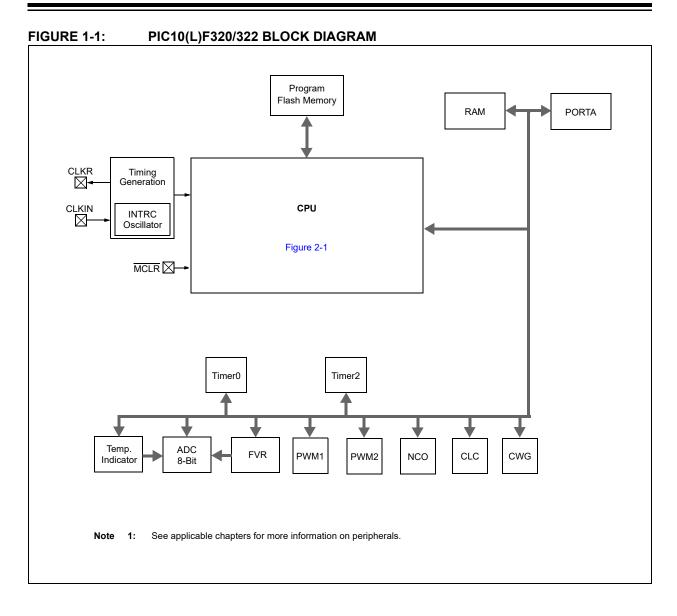
#### 1.0 DEVICE OVERVIEW

The PIC10(L)F320/322 are described within this data sheet. They are available in 6/8-pin packages. Figure 1-1 shows a block diagram of the PIC10(L)F320/322 devices. Table 1-2 shows the pinout descriptions.

Reference Table 1-1 for peripherals available per device.

#### TABLE 1-1: DEVICE PERIPHERAL SUMMARY

Peripheral		PIC10(L)F320	PIC10(L)F322
Analog-to-Digital Converter	(ADC)	•	•
Configurable Logic Cell (CL	C)	•	•
Complementary Wave Gene	erator (CWG)	•	•
Fixed Voltage Reference (F	VR)	٠	•
Numerically Controlled Osci	illator (NCO)	٠	•
Temperature Indicator		٠	•
PWM Modules			
	PWM1	٠	•
	PWM2	٠	•
Timers			
	Timer0	•	•
	Timer2	•	•



#### **TABLE 1-2:** PIC10(L)F320/322 PINOUT DESCRIPTION

Name	Function	Input Type	Output Type	Description
RA0/PWM1/CLC1IN0/CWG1A/	RA0	TTL	CMOS	General purpose I/O with IOC and WPU.
AN0/ICSPDAT	PWM1		CMOS	PWM output.
	CLC1IN0	ST	_	CLC input.
	CWG1A	_	CMOS	CWG primary output.
	AN0	AN		A/D Channel input.
	ICSPDAT	ST	CMOS	ICSP™ Data I/O.
RA1/PWM2/CLC1/CWG1B/AN1/	RA1	TTL	CMOS	General purpose I/O with IOC and WPU.
CLKIN/ICSPCLK/NCO1CLK	PWM2	—	CMOS	PWM output.
	CLC1	—	CMOS	CLC output.
	CWG1B	_	CMOS	CWG complementary output.
	AN1	AN	_	A/D Channel input.
	CLKIN	ST	—	External Clock input (EC mode).
	ICSPCLK	ST	—	ICSP™ Programming Clock.
	NCO1CLK	ST	_	Numerical Controlled Oscillator external clock input.
RA2/INT/T0CKI/NCO1/CLC1IN1/	RA2	TTL	CMOS	General purpose I/O with IOC and WPU.
CLKR/AN2/CWG1FLT	INT	ST	_	External interrupt.
	TOCKI	ST	_	Timer0 clock input.
	NCO1	_	CMOS	Numerically Controlled Oscillator output.
	CLC1IN1	ST	_	CLC input.
	CLKR	_	CMOS	Clock Reference output.
	AN2	AN	_	A/D Channel input.
	CWG1FLT	ST	_	Complementary Waveform Generator Fault 1 source input.
RA3/MCLR/VPP	RA3	TTL		General purpose input.
	MCLR	ST		Master Clear with internal pull-up.
	Vpp	HV	—	Programming voltage.
Vdd	Vdd	Power		Positive supply.
Vss	Vss	Power		Ground reference.

**Legend:** AN = Analog input or output

CMOS = CMOS compatible input or output

TTL = CMOS input with TTL levels

ST = CMOS input with Schmitt Trigger levels

HV = High Voltage

#### 2.0 MEMORY ORGANIZATION

These devices contain the following types of memory:

- Program Memory
  - Configuration Word
  - Device ID
  - User ID
  - Flash Program Memory
- Data Memory
  - Core Registers
  - Special Function Registers
  - General Purpose RAM
  - Common RAM

The following features are associated with access and control of program memory and data memory:

- PCL and PCLATH
- Stack
- Indirect Addressing

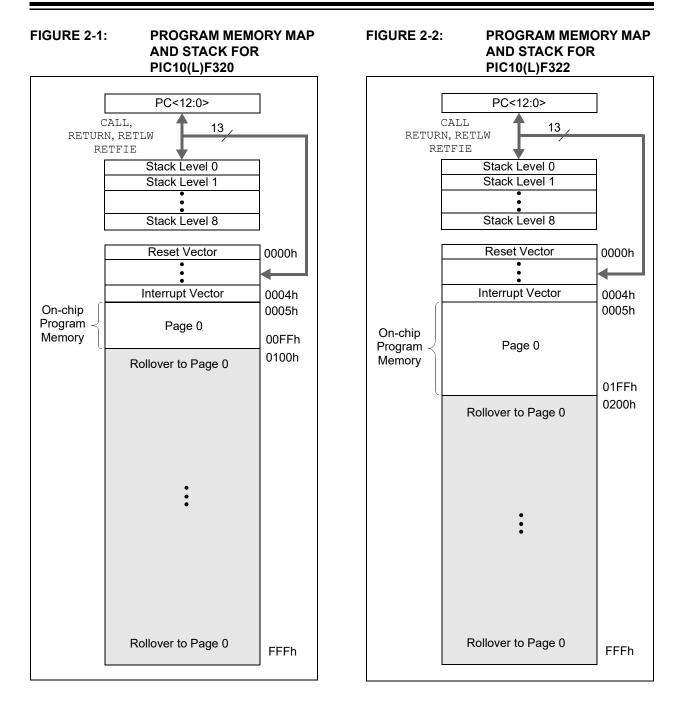
#### TABLE 2-1: DEVICE SIZES AND ADDRESSES

#### 2.1 Program Memory Organization

The mid-range core has a 13-bit program counter capable of addressing 8K x 14 program memory space. This device family only implements up to 512 words of the 8K program memory space. Table 2-1 shows the memory sizes implemented for the PIC10(L)F320/322 family. Accessing a location above these boundaries will cause a wrap-around within the implemented memory space. The Reset vector is at 0000h and the interrupt vector is at 0004h (see Figures 2-1, and 2-2).

Device	Program Memory Space (Words)	Last Program Memory Address	High-Endurance Flash Memory Address Range <sup>(1)</sup>
PIC10(L)F320	256	00FFh	0080h-00FFh
PIC10(L)F322	512	01FFh	0180h-01FFh

**Note 1:** High-endurance Flash applies to the low byte of each address in the range.



#### 2.2 Data Memory Organization

The data memory is in one bank, which contains the General Purpose Registers (GPR) and the Special Function Registers (SFR). The RP<1:0> bits of the STATUS register are the bank select bits.

#### <u>RP1</u> <u>RP0</u>

 $0 \quad 0 \quad \rightarrow \text{Bank 0 is selected}$ 

The bank extends up to 7Fh (128 bytes). The lower locations of the bank are reserved for the Special Function Registers. Above the Special Function Registers are the General Purpose Registers, implemented as Static RAM.

#### 2.2.1 GENERAL PURPOSE REGISTER FILE

The register file is organized as  $64 \times 8$  in the PIC10(L)F320/322. Each register is accessed, either directly or indirectly, through the File Select Register (FSR) (see Section 2.4 "Indirect Addressing, INDF and FSR Registers").

#### 2.2.2 SPECIAL FUNCTION REGISTERS

The Special Function Registers are registers used by the CPU and peripheral functions for controlling the desired operation of the device (see Table 2-3). These registers are static RAM.

The special registers can be classified into two sets: core and peripheral. The Special Function Registers associated with the "core" are described in this section. Those related to the operation of the peripheral features are described in the section of that peripheral feature.

#### 2.2.2.1 STATUS Register

The STATUS register, shown in Register 2-1, contains:

- the arithmetic status of the ALU
- · the Reset status
- the bank select bits for data memory (SRAM)

The STATUS register can be the destination for any instruction, like any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the TO and PD bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, CLRF STATUS will clear the upper three bits and set the Z bit. This leaves the STATUS register as `000u uluu' (where u = unchanged).

It is recommended, therefore, that only BCF, BSF, SWAPF and MOVWF instructions are used to alter the STATUS register, because these instructions do not affect any Status bits. For other instructions not affecting any Status bits (see Section 23.0 "Instruction Set Summary").

- **Note 1:** Bits IRP and RP1 of the STATUS register are not used by the PIC10(L)F320 and will be maintained as clear. Use of these bits is not recommended, since this may affect upward compatibility with future products.
  - 2: The <u>C and DC bits</u> operate as a Borrow and Digit Borrow out bit, respectively, in subtraction.

R/W-0/0	R/W-0/0	R/W-0/0	R-1/q	R-1/q	R/W-x/u	R/W-x/u	R/W-x/u
IRP	RP1	RP0	TO	PD	Z	DC	С
bit 7							bit 0
Legend:							
R = Readabl	e bit	W = Writable I	oit	U = Unimple	mented bit, read	as '0'	
u = Bit is und	changed	x = Bit is unkn	own	-n/n = Value	at POR and BO	R/Value at all o	ther Resets
'1' = Bit is se	t	'0' = Bit is clea	ared	q = Value de	pends on condit	ion	
bit 7	IRP: Reserve	ed <sup>(2)</sup>					
bit 6-5	<b>RP&lt;1:0&gt;:</b> Re	served <sup>(2)</sup>					
bit 4	TO: Time-out	bit					
		ver-up, CLRWDT		or SLEEP instru	ction		
		ime-out occurre	d				
bit 3	PD: Power-D						
		ver-up or by the ution of the <code>SLE</code>					
bit 2	<b>Z</b> : Zero bit			ļ.			
		Ilt of an arithme	tic or loaic or	peration is zero	1		
		llt of an arithme	• •				
bit 1	DC: Digit Car	rry/Borrow bit (A	DDWF, ADDL	W, SUBLW, SUB	WF instructions) <sup>(</sup>	1)	
		out from the 4th			ccurred		
	•	-out from the 4t					
bit 0	•	row bit (ADDWF,			,		
		out from the Mos	0				
	0 = ino carry	-out from the M	ust Significat	IL DIL OI LITE IES	uit occurred		
Note 1: For	Borrow, the pola	arity is reversed.	A subtractio	n is executed b	by adding the two	o's complemen	t of the

#### REGISTER 2-1: STATUS: STATUS REGISTER

- **Note 1:** For Borrow, the polarity is reversed. A subtraction is executed by adding the two's complement of the second operand. For rotate (RRF, RLF) instructions, this bit is loaded with either the high or low-order bit of the source register.
  - 2: Maintain as '0'.

#### 2.2.3 DEVICE MEMORY MAPS

The memory maps for PIC10(L)F320/322 are as shown in Table 2-2.

#### TABLE 2-2: PIC10(L)F320/322 MEMORY MAP (BANK 0)

	7		-		-		-
INDF <sup>(*)</sup>	00h	PMADRL	20h		40h		60h
TMR0	01h	PMADRH	21h				
PCL	02h	PMDATL	22h				
STATUS	03h	PMDATH	23h				
FSR	04h	PMCON1	24h				
PORTA	05h	PMCON2	25h				
TRISA	06h	CLKRCON	26h				
LATA	07h	NCO1ACCL	27h				
ANSELA	08h	NCO1ACCH	28h				
WPUA	09h	NCO1ACCU	29h				
PCLATH	0Ah	NCO1INCL	2Ah				
INTCON	0Bh	NCO1INCH	2Bh				
PIR1	0Ch	Reserved	2Ch				
PIE1	0Dh	NCO1CON	2Dh				
OPTION_REG	0Eh	NCO1CLK	2Eh	General		General	
PCON	0Fh	Reserved	2Fh	Purpose Registers		Purpose Registers	
OSCCON	10h	WDTCON	30h	rtegisters		Registers	
TMR2	11h	CLC1CON	31h	32 Bytes		32 Bytes	
PR2	12h	CLC1SEL1	32h				
T2CON	13h	CLC1SEL2	33h				
PWM1DCL	14h	CLC1POL	34h				
PWM1DCH	15h	CLC1GLS0	35h				
PWM1CON	16h	CLC1GLS1	36h				
PWM2DCL	17h	CLC1GLS2	37h				
PWM2DCH	18h	CLC1GLS3	38h				
PWM2CON	19h	CWG1CON0	39h				
IOCAP	1Ah	CWG1CON1	3Ah				
IOCAN	1Bh	CWG1CON2	3Bh				
IOCAF	1Ch	CWG1DBR	3Ch				
FVRCON	1Dh	CWG1DBF	3Dh				
ADRES	1Eh	VREGCON	3Eh				
ADCON	1Fh	BORCON	3Fh		5Fh		7Fh

Legend: = Unimplemented data memory locations, read as '0'.

\* = Not a physical register.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other resets
Bank 0											
00h	INDF	Address	sing this loca	tion uses co	ntents of FSI	R to address	data memory	(not a physical	register)	XXXX XXXX	XXXX XXXX
01h	TMR0				Timer0 M	lodule Registe	er			XXXX XXXX	uuuu uuuu
02h	PCL			Progra	m Counter (F	PC) Least Sig	nificant Byte			0000 0000	0000 0000
03h	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	000q quuu
04h	FSR			Indi	rect Data Me	mory Address	Pointer			XXXX XXXX	uuuu uuuu
05h	PORTA	_	_	_	_	RA3	RA2	RA1	RA0	xxxx	uuuu
06h	TRISA	_	_	_	_	_(1)	TRISA2	TRISA1	TRISA0	1111	1111
07h	LATA	_	_	_	_	_	LATA2	LATA1	LATA0	xxx	uuu
08h	ANSELA	_	_	_	_	_	ANSA2	ANSA1	ANSA0	111	111
09h	WPUA	_	_	_	_	WPUA3	WPUA2	WPUA1	WPUA0	1111	1111
0Ah	PCLATH	_	_	_	_	_	_	—	PCLH0	0	0
0Bh	INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	0000 0000	0000 000u
0Ch	PIR1	_	ADIF	_	NCO1IF	CLC1IF	—	TMR2IF	—	-0-0 0-0-	-0-0 0-0-
0Dh	PIE1		ADIE		NCO1IE	CLC1IE	_	TMR2IE	_	-0-0 0-0-	-0-0 0-0-
0Eh	OPTION_REG	WPUEN	INTEDG	TOCS	T0SE	PSA		PS<2:0>	•	1111 1111	uuuu uuuu
0Fh	PCON	_	_	_	_	_	_	POR	BOR	dd	uu
10h	OSCCON	_		IRCF<2:0>		HFIOFR	_	LFIOFR	HFIOFS	-110 0-00	-110 0-00
11h	TMR2				Timer2 M	lodule Registe	er			0000 0000	0000 0000
12h	PR2				Timer2 P	Period Registe	r			1111 1111	1111 1111
13h	T2CON	_		TOUT	PS<3:0>		TMR2ON	T2CKP	S<1:0>	-000 0000	-000 0000
14h	PWM1DCL	PWM1D	CL<1:0>	_	_	—	_	—	_	xx	uu
15h	PWM1DCH				PWM	1DCH<7:0>				XXXX XXXX	uuuu uuuu
16h	PWM1CON	PWM1EN	PWM10E	PWM10UT	PWM1POL	_	_	_	_	0000	0000
17h	PWM2DCL	PWM2D	CL<1:0>	_	_	_	_	—	_	xx	uu
18h	PWM2DCH				PWM	2DCH<7:0>				XXXX XXXX	uuuu uuuu
19h	PWM2CON	PWM2EN	PWM2OE	PWM2OUT	PWM2POL	—	—	—	_	0000	0000
1Ah	IOCAP	_	_	_	_	IOCAP3	IOCAP2	IOCAP1	IOCAP0	0000	0000
1Bh	IOCAN		_		_	IOCAN3	IOCAN2	IOCAN1	IOCAN0	0000	0000
1Ch	IOCAF	_	_	_	_	IOCAF3	IOCAF2	IOCAF1	IOCAF0	0000	0000
1Dh	FVRCON	FVREN	FVRRDY	TSEN	TSRNG	—	—	ADFVF	R<1:0>	0x0000	0x0000
1Eh	ADRES		•		A/D Re	sult Register		•		XXXX XXXX	uuuu uuuu
1Fh	ADCON		ADCS<2:0>			CHS<2:0>		<u>GO/</u> DONE	ADON	0000 0000	0000 0000

TABLE 2-3: SPECIAL FUNCTION REGISTER SUMMARY (BANK 0)

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other resets
Bank 0 (	(Continued)										
20h	PMADRL				PMA	ADR<7:0>				0000 0000	0000 0000
21h	PMADRH	—	_	_			_	-	PMADR8	0	0
22h	PMDATL				PMI	DAT<7:0>				XXXX XXXX	uuuu uuuu
23h	PMDATH	—	—			PMDA	AT<13:8>			xx xxxx	uu uuuu
24h	PMCON1	_	CFGS	LWLO	FREE	WRERR	WREN	WR	RD	1000 0000	1000 q000
25h	PMCON2		Pr	ogram Mem	ory Control R	egister 2 (not	a physical ree	gister)	-	0000 0000	0000 0000
26h	CLKRCON	_	CLKROE	—	—	_	—	—	—	-0	-0
27h	NCO1ACCL				NCO1 Acc	cumulator <7:0	)>			0000 0000	0000 0000
28h	NCO1ACCH				NCO1 Acc	umulator <15:	8>			0000 0000	0000 0000
29h	NCO1ACCU		-	_			NCO1 Accum	ulator <1916>		0000	0000
2Ah	NCO1INCL				NCO1 In	crement <7:0>	•			0000 0001	0000 0001
2Bh	NCO1INCH				NCO1 Inc	crement <15:8	>			0000 0000	0000 0000
2Ch	_	Unimpleme	nted							_	_
2Dh	NCO1CON	N1EN	N10E	N1OUT	N1POL	_		_	N1PFM	00000	00x00
2Eh	NCO1CLK		N1PWS<2:0>		_	_	_	N1CKS	S<1:0>	00000	00000
2Fh	Reserved				R	eserved				XXXX XXXX	uuuu uuuu
30h	WDTCON	—	—		r	WDTPS<4:0	)>		SWDTEN	01 0110	01 0110
31h	CLC1CON	LC1EN	LC10E	LC10UT	LC1INTP	LC1INTN	L	.C1MODE<2:0	>	00x0 -000	00x0 -000
32h	CLC1SEL0	—	L	.C1D2S<2:0	>	—		LC1D1S<2:0>		-xxx -xxx	-uuu -uuu
33h	CLC1SEL1	—	L	.C1D4S<2:0	>	_		LC1D3S<2:0>		-xxx -xxx	-uuu -uuu
34h	CLC1POL	LC1POL	_	_	_	LC1G4POL	LC1G3POL	LC1G2POL	LC1G1POL	0 xxxx	0 uuuu
35h	CLC1GLS0	LC1G1D4T	LC1G1D4N	LC1G1D3T	LC1G1D3N	LC1G1D2T	LC1G1D2N	LC1G1D1T	LC1G1D1N	XXXX XXXX	uuuu uuuu
36h	CLC1GLS1	LC1G2D4T	LC1G2D4N	LC1G2D3T	LC1G2D3N	LC1G2D2T	LC1G2D2N	LC1G2D1T	LC1G2D1N	XXXX XXXX	uuuu uuuu
37h	CLC1GLS2	LC1G3D4T	LC1G3D4N	LC1G3D3T	LC1G3D3N	LC1G3D2T	LC1G3D2N	LC1G3D1T	LC1G3D1N	XXXX XXXX	uuuu uuuu
38h	CLC1GLS3	LC1G4D4T	LC1G4D4N	LC1G4D3T	LC1G4D3N	LC1G4D2T	LC1G4D2N	LC1G4D1T	LC1G4D1N	XXXX XXXX	uuuu uuuu
39h	CWG1CON0	G1EN	G10EB	G10EA	G1POLB	G1POLA	_	—	G1CS0	0000 00	0000 00
3Ah	CWG1CON1	G1ASD	LB<1:0>	G1ASD	LA<1:0>	_	_	G1IS•	<1:0>	xxxxxx	uuuuuu
3Bh	CWG1CON2	G1ASE	G1ARSEN	_	_	_	_	G1ASDCLC1	G1ASDFLT	xxxx	uuuu
3Ch	CWG1DBR	_	_			CWG1	DBR<5:0>			xx xxxx	uu uuuu
3Dh	CWG1DBF					CWG1	DBF<5:0>			xx xxxx	uu uuuu
3Eh	VREGCON	_	—	—	—	_	_	VREGPM1	Reserved	01	01
3Fh	BORCON	SBOREN	BORFS	_	_			_	BORRDY	10q	uuu

#### SPECIAL FUNCTION REGISTER SUMMARY (BANK 0) (CONTINUED) **TABLE 2-3**:

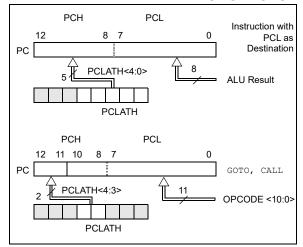
x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: Unimplemented, read as '1'.

#### 2.3 PCL and PCLATH

The Program Counter (PC) is 13 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<12:8>) is not directly readable or writable and comes from PCLATH. On any Reset, the PC is cleared. Figure 2-3 shows the two situations for the loading of the PC. The upper example in Figure 2-3 shows how the PC is loaded on a write to PCL (PCLATH<4:0>  $\rightarrow$  PCH). The lower example in Figure 2-3 shows how the PC is loaded during a CALL or GOTO instruction (PCLATH<4:3>  $\rightarrow$  PCH).

#### FIGURE 2-3: LOADING OF PC IN DIFFERENT SITUATIONS



#### 2.3.1 MODIFYING PCL

Executing any instruction with the PCL register as the destination simultaneously causes the Program Counter PC<12:8> bits (PCH) to be replaced by the contents of the PCLATH register. This allows the entire contents of the program counter to be changed by writing the desired upper five bits to the PCLATH register. When the lower eight bits are written to the PCL register, all 13 bits of the program counter will change to the values contained in the PCLATH register and those being written to the PCL register.

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). Care must be exercised when jumping into a look-up table or program branch table (computed GOTO) by modifying the PCL register. Assuming that PCLATH is set to the table start address, if the table length is greater than 255 instructions or if the lower eight bits of the memory address rolls over from 0xFF to 0x00 in the middle of the table, then PCLATH must be incremented for each address rollover that occurs between the table beginning and the target location within the table.

For more information refer to Application Note AN556, *"Implementing a Table Read"* (DS00556).

#### 2.3.2 STACK

All devices have an 8-level x 13-bit wide hardware stack (see Figure 2-1). The stack space is not part of either program or data space and the Stack Pointer is not readable or writable. The PC is PUSHed onto the stack when a CALL instruction is executed or an interrupt causes a branch. The stack is POPed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth push overwrites the value that was stored from the first push. The tenth push overwrites the second push (and so on).

Note 1:	There are no Status bits to indicate Stack Overflow or Stack Underflow conditions.
2:	There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.

#### 2.4 Indirect Addressing, INDF and FSR Registers

The INDF register is not a physical register. Addressing the INDF register will cause indirect addressing.

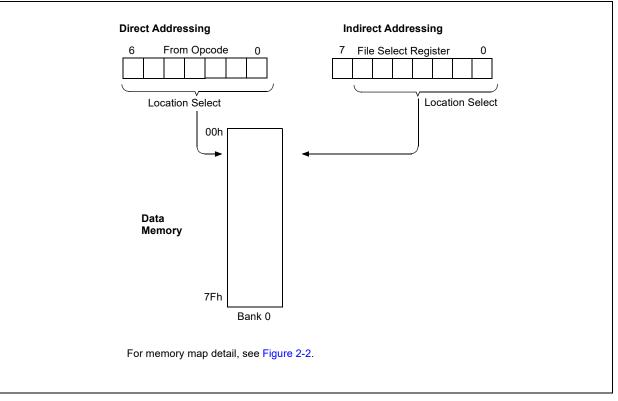
Indirect addressing is possible by using the INDF register. Any instruction using the INDF register actually accesses data pointed to by the File Select Register (FSR). Reading INDF itself indirectly will produce 00h. Writing to the INDF register indirectly results in a no operation (although Status bits may be affected). An effective 9-bit address is obtained by concatenating the 8-bit FSR and the IRP bit of the STATUS register, as shown in Figure 2-4.

A simple program to clear RAM location 40h-7Fh using indirect addressing is shown in Example 2-1.

EXAMPLE 2-1: INDIRECT ADDRESSING
----------------------------------

	MOVLW	0x40	;initialize pointer
	MOVWF	FSR	;to RAM
NEXT	CLRF	INDF	;clear INDF register
	INCF	FSR	;inc pointer
	BTFSS	FSR,7	;all done?
	GOTO	NEXT	;no clear next
CONTIN	UE		;yes continue





#### 3.0 DEVICE CONFIGURATION

Device configuration consists of Configuration Word and Device ID.

#### 3.1 Configuration Word

There are several Configuration Word bits that allow different oscillator and memory protection options. These are implemented as Configuration Word at 2007h.

### 3.2 Register Definitions: Configuration Word

#### REGISTER 3-1: CONFIG: CONFIGURATION WORD

		U-1	R/P-1/1	R/P-1/1	R/P-1/1	R/P-1/1	R/P-1/1			
			i	Γ<1:0>	BORV	LPBOR	LVP			
		bit 13	VII	1 41.02	Borty	Libolt	bit 8			
R/P-1/1	R/P-1/1	R/P-1/1	R/P-1/1	R/P-1/1	R/P-1/1	R/P-1/1	R/P-1/1			
CP	MCLRE	PWRTE	WDT	E<1:0>	BORE	N<1:0>	FOSC			
bit 7							bit			
L <b>egend:</b> R = Readable	hit	W = Writable	ait	LI – Unimpion	nantad hit raad					
					nented bit, read		thar Deceta			
u = Bit is unchangedx = Bit is unknown-n/n = Value at POR and BOR/Value at all other I'1' = Bit is set'0' = Bit is clearedP = Programmable bit					Inel Resels					
I – DILIS SEL			areu	F – Flogrann						
bit 13	Unimplemer	nted: Read as '1	,							
bit 12-11	-	Flash Memory S		ection bits						
		memory: PIC10								
		e protection off								
		h to 03Fh write-								
		01 =000h to 07Fh write-protected, 080h to 0FFh may be modified by PMCON control								
		to 0FFh write-protected, no addresses may be modified by PMCON control nemory: PIC10(L)F322:								
		e protection off	<u>(L) 322.</u>							
		h to 07Fh write-	protected, 080	h to 1FFh may	be modified by	PMCON contro	bl			
		h to 0FFh write-								
	00 =000	h to 1FFh write-	protected, no	addresses may	be modified by	PMCON contro	ol			
oit 10	BORV: Brow	n-out Reset Volt	age Selection	bit						
		ut Reset voltage								
	0 = Brown-ou	ut Reset voltage	(VBOR), high	trip point selecte	ed.					
bit 9	LPBOR: Low	-Power Brown-	out Reset Ena	ble bit						
	•	wer Brown-out F								
	0 = Low-po	wer Brown-out	Reset is disab	led						
oit 8		ltage Programm	•			-				
		age Pro <u>gramm</u> i				1 4-				
		tage on MCLR/	/PP must be u	sed for program	nming					
bit 7	CP: Code Pr									
		memory code p memory code p								
oit 6	If LVP bit = 1	LR/VPP Pin Fun	ction Select D	τ						
	This bit is									
	If LVP bit = 0									
	1 = MCLF	R/VPP pin functio								
		R/VPP pin functio				Veak pull-up un	der control (			
	WPU	A3 bit.								
loto 1. ⊏n	abling Brown	out Reset does i	not automatic	ally enable Pow	er un Timer					
	en 100 (1 1 1 1 1 1 VVI I=(	AN INCOCLUDES								

#### **Note 1:** Enabling Brown-out Reset does not automatically enable Power-up Timer.

- 2: Once enabled, code-protect can only be disabled by bulk erasing the device.
- 3: See VBOR parameter for specific trip point voltages.

#### REGISTER 3-1: CONFIG: CONFIGURATION WORD (CONTINUED)

bit 5	<b>PWRTE:</b> Power-up Timer Enable bit <sup>(1)</sup> 1 = PWRT disabled 0 = PWRT enabled
bit 4-3	WDTE<1:0>: Watchdog Timer Enable bit 11 = WDT enabled 10 = WDT enabled while running and disabled in Sleep 01 = WDT controlled by the SWDTEN bit in the WDTCON register 00 = WDT disabled
bit 2-1	BOREN<1:0>: Brown-out Reset Enable bits 11 = Brown-out Reset enabled; SBOREN bit is ignored 10 = Brown-out Reset enabled while running, disabled in Sleep; SBOREN bit is ignored 01 = Brown-out Reset controlled by the SBOREN bit in the BORCON register 00 = Brown-out Reset disabled; SBOREN bit is ignored
bit 0	FOSC: Oscillator Selection bit 1 = EC on CLKIN pin 0 = INTOSC oscillator I/O function available on CLKIN pin
Note 1:	Enabling Brown-out Reset does not automatically enable Power-up Timer.

- **2:** Once enabled, code-protect can only be disabled by bulk erasing the device.
- **3:** See VBOR parameter for specific trip point voltages.

#### 3.3 Code Protection

Code protection allows the device to be protected from unauthorized access. Program memory protection and data memory protection are controlled independently. Internal access to the program memory and data memory are unaffected by any code protection setting.

#### 3.3.1 PROGRAM MEMORY PROTECTION

The entire program memory space is protected from external reads and writes by the  $\overline{CP}$  bit in Configuration Word. When  $\overline{CP} = 0$ , external reads and writes of program memory are inhibited and a read will return all '0's. The CPU can continue to read program memory, regardless of the protection bit settings. Writing the program memory is dependent upon the write protection setting. See Section 3.4 "Write Protection" for more information.

#### 3.4 Write Protection

Write protection allows the device to be protected from unintended self-writes. Applications, such as boot loader software, can be protected while allowing other regions of the program memory to be modified.

The WRT<1:0> bits in Configuration Word define the size of the program memory block that is protected.

#### 3.5 User ID

Four memory locations (2000h-2003h) are designated as ID locations where the user can store checksum or other code identification numbers. These locations are readable and writable during normal execution. See **Section 3.6 "Device ID and Revision ID**" for more information on accessing these memory locations. For more information on checksum calculation, see the "PIC10(L)F320/322 Flash Memory Programming Specification" (DS41572).

#### 3.6 Device ID and Revision ID

The memory location 2006h is where the Device ID and Revision ID are stored. The upper nine bits hold the Device ID. The lower five bits hold the Revision ID. See Section 9.4 "User ID, Device ID and Configuration Word Access" for more information on accessing these memory locations.

Development tools, such as device programmers and debuggers, may be used to read the Device ID and Revision ID.

#### 3.7 Register Definitions: Device and Revision

#### REGISTER 3-2: DEVID: DEVICE ID REGISTER<sup>(1)</sup>

		R	R	R	R	R	R
				DEV	<8:3>		
		bit 13					bit 8
R	R	R	R	R	R	R	R
	DEV<2:0>				REV<4:0>		
bit 7							bit 0

#### Legend:

R = Readable bit '1' = Bit is set

bit 13-5 **DEV<8:0>:** Device ID bits

Device	DEVID<13:0> Values						
Device	DEV<8:0>	REV<4:0>					
PIC10F320	10 1001 101	x xxxx					
PIC10LF320	10 1001 111	x xxxx					
PIC10F322	10 1001 100	x xxxx					
PIC10LF322	10 1001 110	x xxxx					

'0' = Bit is cleared

#### bit 4-0 **REV<4:0>:** Revision ID bits

These bits are used to identify the revision.

**Note 1:** This location cannot be written.

### 4.0 OSCILLATOR MODULE

#### 4.1 Overview

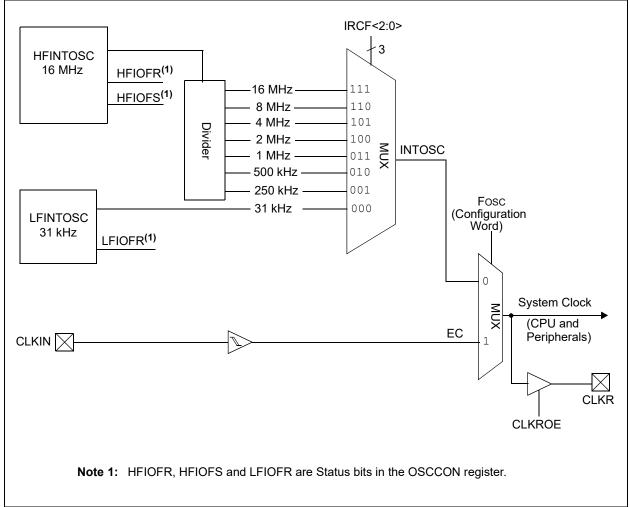
The oscillator module has a variety of clock sources and selection features that allow it to be used in a range of applications while maximizing performance and minimizing power consumption. Figure 4-1 illustrates a block diagram of the oscillator module.

The system can be configured to use an internal calibrated high-frequency oscillator as clock source, with a choice of selectable speeds via software.

Clock source modes are configured by the FOSC bit in Configuration Word (CONFIG).

- 1. EC oscillator from CLKIN.
- 2. INTOSC oscillator, CLKIN not enabled.





#### 4.2 Clock Source Modes

Clock source modes can be classified as external or internal.

- Internal clock source (INTOSC) is contained within the oscillator module, which has eight selectable output frequencies, with a maximum internal frequency of 16 MHz.
- The External Clock mode (EC) relies on an external signal for the clock source.

The system clock can be selected between external or internal clock sources via the FOSC bit of the Configuration Word.

#### 4.3 Internal Clock Modes

The internal clock sources are contained within the oscillator module. The internal oscillator block has two internal oscillators that are used to generate all internal system clock sources: the 16 MHz High-Frequency Internal Oscillator (HFINTOSC) and the 31 kHz (LFINTOSC).

The HFINTOSC consists of a primary and secondary clock. The secondary clock starts first with rapid startup time, but low accuracy. The secondary clock ready signal is indicated with the HFIOFR bit of the OSCCON register. The primary clock follows with slower start-up time and higher accuracy. The primary clock is stable when the HFIOFS bit of the OSCCON register bit goes high.

#### 4.3.1 INTOSC MODE

When the FOSC bit of the Configuration Word is cleared, the INTOSC mode is selected. When INTOSC is selected, CLKIN pin is available for general purpose I/O. See **Section 3.0** "**Device Configuration**" for more information.

#### 4.3.2 FREQUENCY SELECT (IRCF) BITS

The output of the 16 MHz HFINTOSC is connected to a divider and multiplexer (see Figure 4-1). The Internal Oscillator Frequency Select (IRCF) bits of the OSCCON register select the frequency output of the internal oscillator:

- HFINTOSC
  - 16 MHz
  - 8 MHz (default after Reset)
  - 4 MHz
  - 2 MHz
  - 1 MHz
  - 500 kHz
  - 250 kHz
- LFINTOSC
  - 31 kHz

Note:	Following any Reset, the IRCF<2:0> bits
	of the OSCCON register are set to '110'
	and the frequency selection is set to
	8 MHz. The user can modify the IRCF bits
	to select a different frequency.

There is no delay when switching between HFINTOSC frequencies with the IRCF bits. This is because the switch involves only a change to the frequency output divider.

Start-up delay specifications are located in **Section 24.0 "Electrical Specifications"**.

#### 4.4 Register Definitions: Reference Clock Control

U-0	R/W-0/0	U-0	U-0	U-0	U-0	U-0	U-0			
—	CLKROE	—		—		—	—			
bit 7							bit 0			
Legend:	Legend:									
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'										
-n = Value at P	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown			
q = Value depends on condition										

#### REGISTER 4-1: CLKRCON – REFERENCE CLOCK CONTROL REGISTER

bit 7	Unimplemented: Read as '0'
bit 6	CLKROE: Reference Clock Output Enable bit
	1 = Reference Clock output (CLKR), regardless of TRIS
	0 = Reference Clock output disabled
bit 5-0	Unimplemented: Read as '0'

#### 4.5 Register Definitions: Oscillator Control

#### REGISTER 4-2: OSCCON: OSCILLATOR CONTROL REGISTER

U-0	R/W-1/1	R/W-1/1	R/W-0/0	R-0/0	U-0	R-0/0	R-0/0
—		IRCF<2:0>		HFIOFR	—	LFIOFR	HFIOFS
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7	Unimplemented: Read as '0'
bit 6-4	IRCF<2:0>: INTOSC (Fosc) Frequency Select bits
	111 = 16 MHz
	110 = 8 MHz (default value)
	101 = 4 MHz
	100 = 2 MHz
	011 = 1 MHz
	010 = 500 kHz
	001 = 250 kHz
	000 = 31 kHz (LFINTOSC)
bit 3	<b>HFIOFR:</b> High-Frequency Internal Oscillator Ready bit 1 = 16 MHz Internal Oscillator (HFINTOSC) is ready 0 = 16 MHz Internal Oscillator (HFINTOSC) is not ready
bit 2	Unimplemented: Read as '0'
bit 1	LFIOFR: Low-Frequency Internal Oscillator Ready bit 1 = 31 kHz Internal Oscillator (LFINTOSC) is ready 0 = 31 kHz Internal Oscillator (LFINTOSC) is not ready
bit 0	HFIOFS: High-Frequency Internal Oscillator Stable bit
	1 = 16 MHz Internal Oscillator (HFINTOSC) is stable
	0 = 16 MHz Internal Oscillator (HFINTOSC) is not stable

#### 4.6 External Clock Mode

#### 4.6.1 EC MODE

The External Clock (EC) mode allows an externally generated logic level as the system clock source. When operating in this mode, an external clock source is connected to the CLKIN input.

#### TABLE 4-1: SUMMARY OF REGISTERS ASSOCIATED WITH CLOCK SOURCES

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
CLKRCON	_	CLKROE	_	_	—	_	_	_	26
OSCCON			IRCF<2:0>		HFIOFR	_	LFIOFR	HFIOFS	26

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by ECWG.

#### TABLE 4-2: SUMMARY OF CONFIGURATION WORD WITH CLOCK SOURCES

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
	13:8	_	_	—	WRT<1:0>		BORV	LPBOR	LVP	20
CONFIG	7:0	CP	MCLRE	PWRTE	WDTE<1:0>		BOREI	N<1:0>	FOSC	20

**Legend:** — = unimplemented location, read as '0'. Shaded cells are not used by clock sources.

### 5.0 RESETS

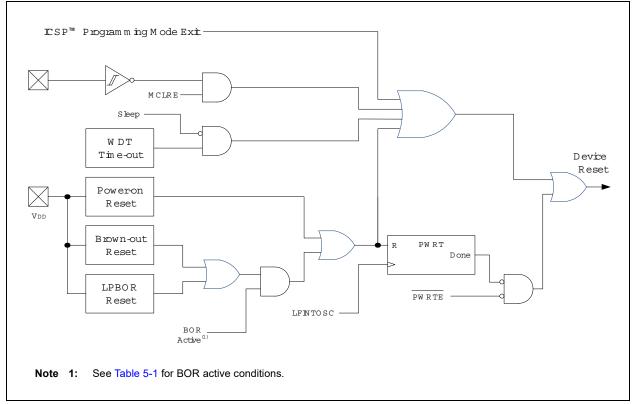
There are multiple ways to reset this device:

- Power-On Reset (POR)
- Brown-Out Reset (BOR)
- Low-Power Brown-Out Reset (LPBOR)
- MCLR Reset
- WDT Reset
- Programming mode exit

To allow VDD to stabilize, an optional Power-up Timer can be enabled to extend the Reset time after a BOR or POR event.

A simplified block diagram of the On-Chip Reset Circuit is shown in Figure 5-1.

#### FIGURE 5-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



#### 5.1 Power-On Reset (POR)

The POR circuit holds the device in Reset until VDD has reached an acceptable level for minimum operation. Slow rising VDD, fast operating speeds or analog performance may require greater than minimum VDD. The PWRT, BOR or MCLR features can be used to extend the start-up period until all device operation conditions have been met.

#### 5.1.1 POWER-UP TIMER (PWRT)

The Power-up Timer provides a nominal 64 ms timeout on POR or Brown-out Reset.

The device is held in Reset as long as PWRT is active. The PWRT delay allows additional time for the VDD to rise to an acceptable level. The Power-up Timer is enabled by clearing the PWRTE bit in Configuration Word.

The Power-up Timer starts after the release of the POR and BOR.

For additional information, refer to Application Note AN607, *"Power-up Trouble Shooting"* (DS00607).

#### 5.2 Brown-Out Reset (BOR)

The BOR circuit holds the device in Reset when VDD reaches a selectable minimum level. Between the POR and BOR, complete voltage range coverage for execution protection can be implemented.

The Brown-out Reset module has four operating modes controlled by the BOREN<1:0> bits in Configuration Word. The four operating modes are:

- BOR is always on
- BOR is off when in Sleep
- · BOR is controlled by software
- · BOR is always off

Refer to Table 5-1 for more information.

The Brown-out Reset voltage level is selectable by configuring the BORV bit in Register 3-1.

A VDD noise rejection filter prevents the BOR from triggering on small events. If VDD falls below VBOR for a duration greater than parameter TBORDC, the device will reset. See Figure 5-2 for more information.

BOREN<1:0>	SBOREN	Device Mode	BOR Mode	Device Operation upon: Release of POR/Wake- up from Sleep			
11	Х	Х	X Active Waits for BOR ready <sup>(1)</sup> (BORRD)				
1.0	37	Awake	Active	Weite for POP ready (POPPDV = 1)			
10	Х	Sleep	Disabled	Waits for BOR ready (BORRDY = 1)			
0.1	1	х	Active	Waits for BOR ready <sup>(1)</sup> (BORRDY = 1)			
01 -	0	х	Disabled	Paging immediately (POPPDV =)			
00	Х	х	Disabled	Begins immediately (BORRDY = x)			

TABLE 5-1: BOR OPERATING MODES

**Note 1:** In these specific cases, "Release of POR" and "Wake-up from Sleep", there is no delay in start-up. The BOR ready flag, (BORRDY = 1), will be set before the CPU is ready to execute instructions because the BOR circuit is forced on by the BOREN<1:0> bits.

#### 5.2.1 BOR IS ALWAYS ON

When the BOREN bits of Configuration Word are programmed to '11', the BOR is always on. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is active during Sleep. The BOR does not delay wake-up from Sleep.

#### 5.2.2 BOR IS OFF IN SLEEP

When the BOREN bits of Configuration Word are programmed to '10', the BOR is on, except in Sleep. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is not active during Sleep. The device wake-up will be delayed until the BOR is ready.

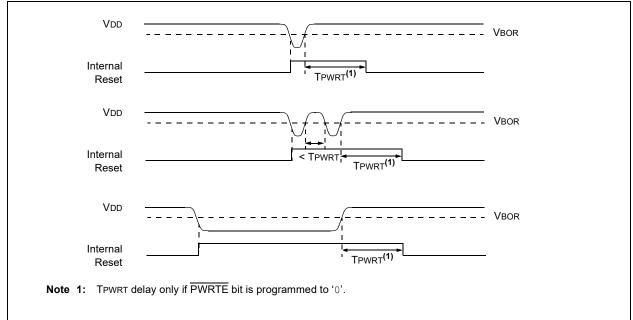
#### 5.2.3 BOR CONTROLLED BY SOFTWARE

When the BOREN bits of Configuration Word are programmed to '01', the BOR is controlled by the SBOREN bit of the BORCON register. The device startup is not delayed by the BOR ready condition or the VDD level.

BOR protection begins as soon as the BOR circuit is ready. The status of the BOR circuit is reflected in the BORRDY bit of the BORCON register.

BOR protection is unchanged by Sleep.





#### 5.3 Register Definition: BOR Control

#### REGISTER 5-1: BORCON: BROWN-OUT RESET CONTROL REGISTER

R/W-1/u	R/W-0/u	U-0	U-0	U-0	U-0	U-0	R-q/u
SBOREN	BORFS <sup>(1)</sup>	—	—	—	—	—	BORRDY
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7	<pre>SBOREN: Software Brown-out Reset Enable bit If BOREN &lt;1:0&gt; in Configuration Word ≠ 01: SBOREN is read/write, but has no effect on the BOR. If BOREN &lt;1:0&gt; in Configuration Word = 01: 1 = BOR enabled 0 = BOR disabled</pre>
bit 6	BORFS: Brown-out Reset Fast Start bit <sup>(1)</sup> <u>If BOREN&lt;1:0&gt; = 11 (Always on) or BOREN&lt;1:0&gt; = 00 (Always off)</u> BORFS is Read/Write, but has no effect.
	<u>If BOREN &lt;1:0&gt; = 10 (Disabled in Sleep) or BOREN&lt;1:0&gt; = 01 (Under software control):</u> 1 = Band gap is forced on always (covers Sleep/wake-up/operating cases) 0 = Band gap operates normally, and may turn off
bit 5-1	Unimplemented: Read as '0'
bit 0	<b>BORRDY:</b> Brown-out Reset Circuit Ready Status bit 1 = The Brown-out Reset circuit is active 0 = The Brown-out Reset circuit is inactive

**Note 1:** BOREN<1:0> bits are located in Configuration Word.

### 5.4 Low-Power Brown-out Reset (LPBOR)

The Low-Power Brown-Out Reset (LPBOR) is an essential part of the Reset subsystem. Refer to Figure 5-1 to see how the BOR interacts with other modules.

The LPBOR is used to monitor the external VDD pin. When too low of a voltage is detected, the device is held in Reset. When this occurs, a register bit ( $\overline{\text{BOR}}$ ) is changed to indicate that a BOR Reset has occurred. The same bit is set for both the BOR and the LPBOR. Refer to Register 5-2.

#### 5.4.1 ENABLING LPBOR

The LPBOR is controlled by the LPBOR bit of Configuration Word. When the device is erased, the LPBOR module defaults to enabled.

#### 5.4.1.1 LPBOR Module Output

The output of the LPBOR module is a signal indicating whether or not a Reset is to be asserted. This signal is OR'd together with the Reset signal of the BOR module to provide the generic BOR signal which goes to the PCON register and to the power control block.

#### 5.5 MCLR

The  $\overline{\text{MCLR}}$  is an optional external input that can reset the device. The  $\overline{\text{MCLR}}$  function is controlled by the MCLRE and the LVP bit of Configuration Word (Table 5-2).

TABLE 5-2: MCLR CONFIGURATION

MCLRE	LVP	MCLR
0	0	Disabled
1	0	Enabled
x	1	Enabled

#### 5.5.1 MCLR ENABLED

When MCLR is enabled and the pin is held low, the device is held in Reset. The MCLR pin is connected to VDD through an internal weak pull-up.

The device has a noise filter in the  $\overline{\text{MCLR}}$  Reset path. The filter will detect and ignore small pulses.

```
Note: A Reset does not drive the \overline{MCLR} pin low.
```

#### 5.5.2 MCLR DISABLED

When MCLR is disabled, the pin functions as a general purpose input and the internal weak pull-up is under software control.

#### 5.6 Watchdog Timer (WDT) Reset

The Watchdog Timer generates a Reset if the firmware does not issue a CLRWDT instruction within the time-out period. The TO and PD bits in the STATUS register are changed to indicate the WDT Reset. See Section 8.0 "Watchdog Timer" for more information.

#### 5.7 Programming Mode ICSP Exit

Upon exit of Programming mode, the device will behave as if a POR had just occurred.

#### 5.8 Power-Up Timer

The Power-up Timer optionally delays device execution after a BOR or POR event. This timer is typically used to allow VDD to stabilize before allowing the device to start running.

The Power-up Timer is controlled by the  $\overrightarrow{\text{PWRTE}}$  bit of Configuration Word.

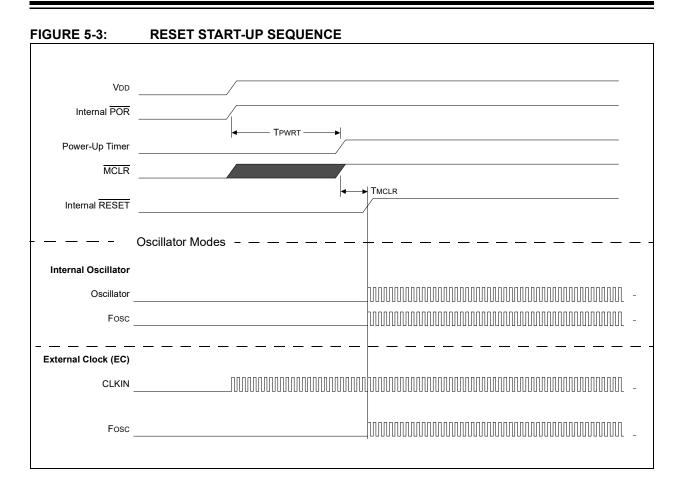
#### 5.9 Start-up Sequence

Upon the release of a POR or BOR, the following must occur before the device will begin executing:

- 1. Power-up Timer runs to completion (if enabled).
- 2. MCLR must be released (if enabled).

The total time-out will vary based on oscillator configuration and Power-up Timer configuration. See **Section 4.0 "Oscillator Module"** for more information.

The Power-up Timer runs independently of MCLR Reset. If MCLR is kept low long enough, the Power-up Timer will expire. Upon bringing MCLR high, the device will begin execution after 10 FOSC cycles (see Figure 5-3). This is useful for testing purposes or to synchronize more than one device operating in parallel.



#### 5.10 Determining the Cause of a Reset

Upon any Reset, multiple bits in the STATUS and PCON registers are updated to indicate the cause of the Reset. Table 5-3 and Table 5-4 show the Reset conditions of these registers.

POR	BOR	то	PD	Condition
0	х	1	1	Power-on Reset
u	0	1	1	Brown-out Reset
u	u	0	u	WDT Reset
u	u	0	0	WDT Wake-up from Sleep
u	u	u	u	MCLR Reset during normal operation
u	u	1	0	MCLR Reset during Sleep

#### TABLE 5-3: RESET STATUS BITS AND THEIR SIGNIFICANCE

#### TABLE 5-4:RESET CONDITION FOR SPECIAL REGISTERS

Condition	Program Counter	STATUS Register	PCON Register
Power-on Reset	0000h	0001 1000	0x
MCLR Reset during normal operation	0000h	000u uuuu	uu
MCLR Reset during Sleep	0000h	0001 Ouuu	uu
WDT Reset	0000h	0000 uuuu	uu
WDT Wake-up from Sleep	PC + 1	0000 Ouuu	uu
Brown-out Reset	0000h	0001 luuu	u0
Interrupt Wake-up from Sleep	PC + 1 <sup>(1)</sup>	0001 Ouuu	uu

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, reads as '0'.

**Note 1:** When the wake-up is due to an interrupt and Global Enable (GIE) bit is set, the return address is pushed on the stack and PC is loaded with the interrupt vector (0004h) after execution of PC + 1.

#### 5.11 Power Control (PCON) Register

The Power Control (PCON) register contains flag bits to differentiate between a:

- Power-On Reset (POR)
- Brown-Out Reset (BOR)

The PCON register bits are shown in Register 5-2.

#### 5.12 Register Definition: Power Control

#### REGISTER 5-2: PCON: POWER CONTROL REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	R/W/HC-q/u	R/W/HC-q/u
—	—	—	—	—	—	POR	BOR
bit 7							bit 0

Legend:		
HC = Bit is cleared by ha	dware	
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7-2	Unimplemented: Read as '0'
bit 1	POR: Power-on Reset Status bit
bit 0	1 = No Power-on Reset occurred 0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs) <b>BOR:</b> Brown-out Reset Status bit
	<ul> <li>1 = No Brown-out Reset occurred</li> <li>0 = A Brown-out Reset occurred (must be set in software after a Power-on Reset or Brown-out Reset occurs)</li> </ul>

#### TABLE 5-5: SUMMARY OF REGISTERS ASSOCIATED WITH RESETS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
BORCON	SBOREN	BORFS						BORRDY	30
PCON	_	_			_	_	POR	BOR	34
STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	13
WDTCON				V	SWDTEN	48			

**Legend:** — = unimplemented location, read as '0'. Shaded cells are not used by Resets.

#### TABLE 5-6: SUMMARY OF CONFIGURATION WORD WITH RESETS

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
CONFIG	13:8	_		_	WRT<1:0>		BORV	LPBOR	LVP	20
	7:0	CP	MCLRE	PWRTE	WDTE<1:0>		BOREN<1:0>		FOSC	20

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by Reset.

#### 6.0 INTERRUPTS

The interrupt feature allows certain events to preempt normal program flow. Firmware is used to determine the source of the interrupt and act accordingly. Some interrupts can be configured to wake the MCU from Sleep mode.

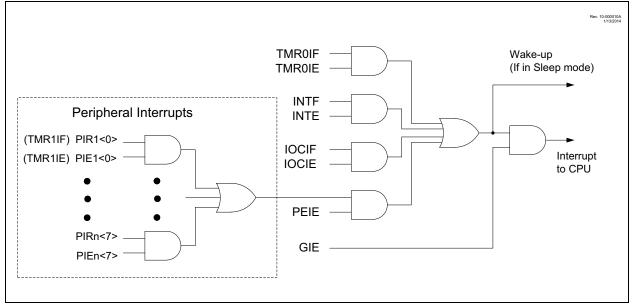
This chapter contains the following information for Interrupts:

- Operation
- Interrupt Latency
- Interrupts During Sleep
- INT Pin
- · Context Saving during Interrupts

Many peripherals produce interrupts. Refer to the corresponding chapters for details.

A block diagram of the interrupt logic is shown in Figure 6-1.





#### 6.1 Operation

Interrupts are disabled upon any device Reset. They are enabled by setting the following bits:

- · GIE bit of the INTCON register
- Interrupt Enable bit(s) for the specific interrupt event(s)
- PEIE bit of the INTCON register (if the Interrupt Enable bit of the interrupt event is contained in the PIE1 register)

The INTCON and PIR1 registers record individual interrupts via interrupt flag bits. Interrupt flag bits will be set, regardless of the status of the GIE, PEIE and individual interrupt enable bits.

The following events happen when an interrupt event occurs while the GIE bit is set:

- · Current prefetched instruction is flushed
- · GIE bit is cleared
- Current Program Counter (PC) is pushed onto the stack
- · PC is loaded with the interrupt vector 0004h

The firmware within the Interrupt Service Routine (ISR) will determine the source of the interrupt by polling the interrupt flag bits. The interrupt flag bits must be cleared before exiting the ISR to avoid repeated interrupts. Because the GIE bit is cleared, any interrupt that occurs while executing the ISR will be recorded through its interrupt flag, but will not cause the processor to redirect to the interrupt vector.

The  ${\tt RETFIE}$  instruction exits the ISR by popping the previous address from the stack, and setting the GIE bit.

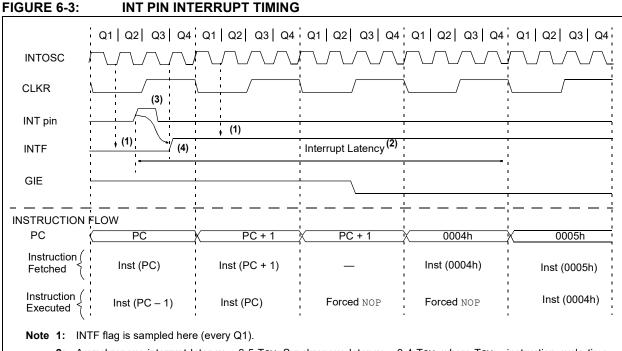
For additional information on a specific interrupt's operation, refer to its peripheral chapter.

- Note 1: Individual interrupt flag bits are set, regardless of the state of any other enable bits.
  - 2: All interrupts will be ignored while the GIE bit is cleared. Any interrupt occurring while the GIE bit is clear will be serviced when the GIE bit is set again.

#### 6.2 Interrupt Latency

Interrupt latency is defined as the time from when the interrupt event occurs to the time code execution at the interrupt vector begins. The latency for synchronous interrupts is three or four instruction cycles. For asynchronous interrupts, the latency is three to five instruction cycles, depending on when the interrupt occurs. See Figure 6-2 and Section 6.3 "Interrupts During Sleep" for more details.

FIGURE 6	6-2: IN	NTERRUPT	LATENCY					
							Q 1 Q 2 Q 3 Q 4	
CLKR			Internu during	ptSampled Q1				
Interrupt								
GE								
PC	PC-1	PC	PC	+1	0004h	0005h		()
Execute	1 Cycle Inst	uction at PC	Inst(PC)	NOP	NOP	Inst(0004h)		
Interrupt			l					
GE								
PC	PC-1	PC	PC+1/FSR ADDR	New PC/ PC+1	0004h	0005h		
Execute-	2 Cycle Inst	uction at PC	Inst(PC)	NOP	NOP	Inst(0004h)		
				1				
Interrupt								
GE								
PC	PC -1	PC	FSR ADDR	PC+1	PC+2	0004h	0005h	
Execute	3 Cycle Inst	uction at PC	INST (PC)	NOP	NOP	NOP	lhst(0004h)	Inst(0005h)
Interrupt								
GE								
PC	PC-1	PC	FSR ADDR	PC+1	PC	2+2	0004h	0005h
Execute	3 Cycle Inst	uction at PC	INST (PC)	NOP	NOP	NOP	NOP	Inst(0004h)



2: Asynchronous interrupt latency = 3-5 TCY. Synchronous latency = 3-4 TCY, where TCY = instruction cycle time. Latency is the same whether Inst (PC) is a single cycle or a 2-cycle instruction.

- 3: For minimum width of INT pulse, refer to AC specifications in Section 24.0 "Electrical Specifications".
- 4: INTF is enabled to be set any time during the Q4-Q1 cycles.

## 6.3 Interrupts During Sleep

Some interrupts can be used to wake from Sleep. To wake from Sleep, the peripheral must be able to operate without the system clock. The interrupt source must have the appropriate Interrupt Enable bit(s) set prior to entering Sleep.

On waking from Sleep, if the GIE bit is also set, the processor will branch to the interrupt vector. Otherwise, the processor will continue executing instructions after the SLEEP instruction. The instruction directly after the SLEEP instruction will always be executed before branching to the ISR. Refer to the Section 7.0 "Power-Down Mode (Sleep)" for more details.

# 6.4 INT Pin

The INT pin can be used to generate an asynchronous edge-triggered interrupt. This interrupt is enabled by setting the INTE bit of the INTCON register. The INTEDG bit of the OPTION\_REG register determines on which edge the interrupt will occur. When the INTEDG bit is set, the rising edge will cause the interrupt. When the INTEDG bit is clear, the falling edge will cause the interrupt. The INTF bit of the INTCON register will be set when a valid edge appears on the INT pin. If the GIE and INTE bits are also set, the processor will redirect program execution to the interrupt vector.

#### 6.5 Context Saving During Interrupts

During an interrupt, only the return PC value is saved on the stack. Typically, users may wish to save key registers during an interrupt (e.g., W and STATUS registers). This must be implemented in software.

Temporary holding registers W\_TEMP and STATUS\_TEMP must be placed in the last 16 bytes of GPR (see Table 1-2). This makes context save and restore operations simpler. The code shown in Example 6-1 can be used to:

- Store the W register
- Store the STATUS register
- · Execute the ISR code
- Restore the Status (and Bank Select Bit register)
- · Restore the W register
- Note: These devices do not require saving the PCLATH. However, if computed GOTOS are used in both the ISR and the main code, the PCLATH must be saved and restored in the ISR.

#### EXAMPLE 6-1: SAVING STATUS AND W REGISTERS IN RAM

MOVWF SWAPF	W_TEMP STATUS,W	;Copy W to TEMP register ;Swap status to be saved into W ;Swaps are used because they do not affect the status bits
MOVWF	STATUS_TEMP	;Save status to bank zero STATUS_TEMP register
:(ISR) :		;Insert user code here
SWAPF	STATUS_TEMP,W	;Swap STATUS_TEMP register into W ;(sets bank to original state)
MOVWF	STATUS	;Move W into STATUS register
SWAPF	W TEMP,F	;Swap W TEMP
SWAPF	W_TEMP,W	;Swap W_TEMP into W

# 6.6 Interrupt Control Registers

#### **REGISTER 6-1:** INTCON: INTERRUPT CONTROL REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R-0/0			
GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF <sup>(1)</sup>			
bit 7							bit 0			
Legend:										
R = Readab	le bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'				
u = Bit is un	changed	x = Bit is unkr	nown	-n/n = Value a	at POR and BO	R/Value at all o	ther Resets			
'1' = Bit is se	et	'0' = Bit is clea	ared							
bit 7		nterrupt Enable								
		all active interru	ipts							
h:4 0	0 = Disables	-								
bit 6	<b>PEIE:</b> Peripheral Interrupt Enable bit 1 = Enables all active peripheral interrupts									
		all peripheral in								
bit 5	TMR0IE: Tim	er0 Overflow Ir	iterrupt Enabl	e bit						
		the Timer0 inter								
		the Timer0 inte	•							
bit 4		tternal Interrupt the INT externa								
		the INT externa								
bit 3		upt-on-Change	•	ble bit						
	1 = Enables f	the interrupt-on	-change interi	rupt						
	0 = Disables	the interrupt-or	-change inter	rupt						
bit 2		er0 Overflow In		bit						
		gister has overf gister did not ov								
bit 1		ternal Interrupt								
		external interru								
	0 = The INT e	external interru	ot did not occu	ur						
bit 0		upt-on-Change								
		least one of the he interrupt-on-								
			•	C C						
	he IOCIF Flag bi ave been cleared		nd cleared wh	en all the Inter	rupt-on-Change	flags in the IO	CAF register			

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Interrupt Enable bit, GIE, of the INTCON register. User software must ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

U-0	R/W-0/0	U-0	R/W-0/0	R/W-0/0	U-0	R/W-0/0	U-0					
_	ADIE		NCO1IE	CLC1IE		TMR2IE						
bit 7							bit (					
Legend:												
R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit, read	l as '0'						
u = Bit is unc	hanged	x = Bit is unk	nown	-n/n = Value a	at POR and BO	R/Value at all ot	her Resets					
'1' = Bit is set	t	'0' = Bit is cle	ared									
bit 7	Unimplemer	nted: Read as '	0'									
bit 6	ADIE: A/D Co	ADIE: A/D Converter Interrupt Enable bit										
	1 = Enables the A/D converter interrupt											
	0 = Disables the A/D converter interrupt											
bit 5	Unimplemer	nted: Read as '	0'									
bit 4	NCO1IE: Nui	merically Contr	olled Oscillato	r Interrupt Ena	ble bit							
		1 = Enables the NCO overflow interrupt										
		the NCO over	-									
bit 3		figurable Logic		pt Enable bit								
		the CLC interre										
<b>h</b> it 0		the CLC interr	•									
bit 2	-	nted: Read as '										
bit 1		R2 to PR2 Mat										
		the TMR2 to P the TMR2 to F										
hit 0				enupt								
bit 0	Unimplemen	nted: Read as '	U									
Noto: Bi	t DEIE of the IN		must be									

### REGISTER 6-2: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

**Note:** Bit PEIE of the INTCON register must be set to enable any peripheral interrupt.

its corresponding enable bit or the Global Interrupt Enable bit, GIE, of the INTCON register. User software must ensure the appropriate interrupt flag bits are clear prior

to enabling an interrupt.

U-0	R/W-0/0	U-0	R/W-0/0	R/W-0/0	U-0	R/W-0/0	U-0						
_	ADIF	—	NCO1IF	CLC1IF	—	TMR2IF	_						
bit 7							bit 0						
Legend:													
R = Readat	ala hit	W = Writable	hit	II – Unimplor	nonted hit rea	d oo 'O'							
u = Bit is ur		x = Bit is unk		•	nented bit, rea	o as o DR/Value at all ot	hor Posota						
'1' = Bit is s	•	0' = Bit is cle			at FOR and BC	n, value at all oti	IEI NESEIS						
	el		aleu										
bit 7	Unimplemer	nted: Read as	'O'										
bit 6	ADIF: A/D Co	ADIF: A/D Converter Interrupt Flag bit											
		<ul> <li>1 = The A/D conversion completed</li> <li>0 = The A/D conversion is not complete</li> </ul>											
			•										
bit 5	Unimplemer	Unimplemented: Read as '0'											
bit 4		•		r Interrupt Flag									
	1 = NCO1 o 0 = No NCO		ed (must be cle	eared in softwa	re)								
bit 3			c Block Rising	Edge Interrupt	Elac bit								
		• •	•	red in software	•								
	0 = No CLC		(		,								
bit 2	Unimplemer	nted: Read as	'0'										
bit 1	TMR2IF: TM	R2 to PR2 Ma	tch Interrupt FI	ag bit									
				be cleared in s	oftware)								
		2 to PR2 matc		e									
				of times specifi	ed by the TMR	2 postscaler (Re	gister 17-1).						
bit 0	Unimplemen	nted: Read as	•0'										
	Interrupt flag bits a												
(	condition occurs, r	egardless of th	ne state of										

# REGISTER 6-3: PIR1: PERIPHERAL INTERRUPT REQUEST REGISTER 1

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	40
IOCAF	_	—	—	—	IOCAF3	IOCAF2	IOCAF1	IOCAF0	76
IOCAN	—	_	_	_	IOCAN3	IOCAN2	IOCAN1	IOCAN0	75
IOCAP	_	—	—	—	IOCAP3	IOCAP2	IOCAP1	IOCAP0	75
OPTION_REG	WPUEN	INTEDG	TOCS	T0SE	PSA	PS<2:0>		95	
PIE1	—	ADIE	_	NCO1IE	CLC1IE	—	TMR2IE	—	41
PIR1		ADIF		NCO1IF	CLC1IF	_	TMR2IF	_	42

 TABLE 6-1:
 SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPTS

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by Interrupts.

# 7.0 POWER-DOWN MODE (SLEEP)

The Power-Down mode is entered by executing a SLEEP instruction.

Upon entering Sleep mode, the following conditions exist:

- 1. WDT will be cleared but keeps running, if enabled for operation during Sleep.
- 2. PD bit of the STATUS register is cleared.
- 3. TO bit of the STATUS register is set.
- 4. CPU clock is disabled.
- 5. 31 kHz LFINTOSC is unaffected and peripherals that operate from it may continue operation in Sleep.
- 6. ADC is unaffected, if the dedicated FRC clock is selected.
- I/O ports maintain the status they had before SLEEP was executed (driving high, low or highimpedance).
- 8. Resets other than WDT are not affected by Sleep mode.

Refer to individual chapters for more details on peripheral operation during Sleep.

To minimize current consumption, the following conditions need to be considered:

- I/O pins must not be floating
- External circuitry sinking current from I/O pins
- Internal circuitry sourcing current from I/O pins
- · Current draw from pins with internal weak pull-ups
- Modules using 31 kHz LFINTOSC
- · CWG and NCO modules using HFINTOSC

I/O pins that are high-impedance inputs will be pulled to VDD or VSS externally to avoid switching currents caused by floating inputs.

Examples of internal circuitry that might be sourcing current include the FVR module. See Section 12.0 "Fixed Voltage Reference (FVR)" for more information on these modules.

# 7.1 Wake-up from Sleep

The device can wake-up from Sleep through one of the following events:

- 1. External Reset input on MCLR pin, if enabled
- 2. BOR Reset, if enabled
- 3. POR Reset
- 4. Watchdog Timer, if enabled
- 5. Any external interrupt
- 6. Interrupts by peripherals capable of running during Sleep (see individual peripheral for more information)

The first three events will cause a device Reset. The last three events are considered a continuation of program execution. To determine whether a device Reset or wake-up event occurred, refer to **Section 5.10 "Determining the Cause of a Reset"**.

When the SLEEP instruction is being executed, the next instruction (PC + 1) is prefetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be enabled. Wake-up will occur regardless of the state of the GIE bit. If the GIE bit is disabled, the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is enabled, the device executes the instruction after the SLEEP instruction, the device will then call the Interrupt Service Routine. In cases where the execution of the instruction following SLEEP is not desirable, the user will have a NOP after the SLEEP instruction.

The WDT is cleared when the device wakes up from Sleep, regardless of the source of wake-up.

The Complementary Waveform Generator (CWG) and the Numerically Controlled Oscillator (NCO) modules can utilize the HFINTOSC oscillator as their respective clock source. Under certain conditions, when the HFINTOSC is selected for use with the CWG or NCO modules, the HFINTOSC will remain active during Sleep. This will have a direct effect on the Sleep mode current. Please refer to 21.0 "Complementary Waveform Generator (CWG) Module" and 20.0 "Numerically Controlled Oscillator (NCO) Module" for more information.

# 7.1.1 WAKE-UP USING INTERRUPTS

When global interrupts are disabled (GIE cleared) and any interrupt source has both its interrupt enable bit and interrupt flag bit set, one of the following will occur:

- If the interrupt occurs **before** the execution of a SLEEP instruction
  - SLEEP instruction will execute as a NOP.
  - WDT and WDT prescaler will not be cleared
  - TO bit of the STATUS register will not be set
  - PD bit of the STATUS register will not be cleared.

- If the interrupt occurs **during or after** the execution of a **SLEEP** instruction
  - SLEEP instruction will be completely executed
  - Device will immediately wake-up from Sleep
  - WDT and WDT prescaler will be cleared
  - TO bit of the STATUS register will be set
  - PD bit of the STATUS register will be cleared.

Even if the flag bits were checked before executing a SLEEP instruction, it may be possible for flag bits to become set before the SLEEP instruction completes. To determine whether a SLEEP instruction executed, test the PD bit. If the PD bit is set, the SLEEP instruction was executed as a NOP.

			-			-		
					i			
CLKIN <sup>(1</sup>	Q1 Q2 Q3 Q4	Q1 Q2 Q3  Q4	Q1	1 1 1	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4
CLKOUT <sup>(2</sup>	1			Tost(3)				
Interrupt flag	1 1 1	I I I	/	, , , ≠	Interrupt Laten	су <sup>(4)</sup>	· · · · · · · · · · · · · · · · · · ·	
GIE bit (INTCON reg.	 .); 	     	Processor in	· · ·		<u></u>	· · ·	
struction Flow PC	/ X PC	X PC + 1	ХРС	+ 2	X PC + 2	PC + 2	X 0004h	X 0005h
Instruction {	Inst(PC) = Sleep	Inst(PC + 1)			Inst(PC + 2)	   	Inst(0004h)	Inst(0005h)
Instruction { Executed {	Inst(PC - 1)	Sleep	1 1		Inst(PC + 1)	Forced NOP	Forced NOP	Inst(0004h)
2: 3:	External clock. Hig CLKOUT is shown Tost= 1024 Tosc; Power Brown-out GIE = 1 assumed.	here for timing ref This delay does no <b>Reset (LPBOR)</b> "	erence. ot apply to EC .).	, RC and				

## FIGURE 7-1: WAKE-UP FROM SLEEP THROUGH INTERRUPT

# TABLE 7-1: SUMMARY OF REGISTERS ASSOCIATED WITH POWER-DOWN MODE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	13
WDTCON	_	_			SWDTEN	48			

Legend: — = unimplemented location, read as '0'. Shaded cells are not used in Power-down mode.

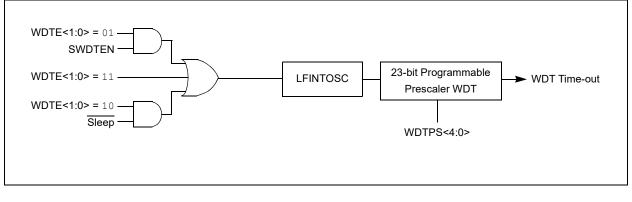
# 8.0 WATCHDOG TIMER

The Watchdog Timer is a system timer that generates a Reset if the firmware does not issue a CLRWDT instruction within the time-out period. The Watchdog Timer is typically used to recover the system from unexpected events.

The WDT has the following features:

- Independent clock source
- · Multiple operating modes
  - WDT is always on
  - WDT is off when in Sleep
  - WDT is controlled by software
  - WDT is always off
- Configurable time-out period is from 1 ms to 256 seconds (typical)
- Multiple Reset conditions
- Operation during Sleep





#### 8.1 Independent Clock Source

The WDT derives its time base from the 31 kHz LFINTOSC internal oscillator. Time intervals in this chapter are based on a nominal interval of 1ms. See **Section 24.0** "Electrical Specifications" for the LFINTOSC tolerances.

### 8.2 WDT Operating Modes

The Watchdog Timer module has four operating modes controlled by the WDTE<1:0> bits in Configuration Word. See Table 8-1.

#### 8.2.1 WDT IS ALWAYS ON

When the WDTE bits of Configuration Word are set to '11', the WDT is always on.

WDT protection is active during Sleep.

#### 8.2.2 WDT IS OFF IN SLEEP

When the WDTE bits of Configuration Word are set to '10', the WDT is on, except in Sleep.

WDT protection is not active during Sleep.

#### 8.2.3 WDT CONTROLLED BY SOFTWARE

When the WDTE bits of Configuration Word are set to '01', the WDT is controlled by the SWDTEN bit of the WDTCON register.

WDT protection is unchanged by Sleep. See Table 8-1 for more details.

#### TABLE 8-1: WDT OPERATING MODES

WDTE<1:0>	SWDTEN	Device Mode	WDT Mode
11	Х	Х	Active
1.0	V	Awake	Active
10	Х	Sleep	Disabled
0.1	1		Active
01	0	Х	Disabled
00	Х	Х	Disabled

#### TABLE 8-2:WDT CLEARING CONDITIONS

Conditions	WDT		
WDTE<1:0> = 00			
WDTE<1:0> = 01 and SWDTEN = 0			
WDTE<1:0> = 10 and enter Sleep	Cleared		
CLRWDT Command			
Exit Sleep			
Change INTOSC divider (IRCF bits)	Unaffected		

### 8.3 Time-Out Period

The WDTPS bits of the WDTCON register set the timeout period from 1 ms to 256 seconds (nominal). After a Reset, the default time-out period is 2 seconds.

## 8.4 Clearing the WDT

The WDT is cleared when any of the following conditions occur:

- Any Reset
- CLRWDT instruction is executed
- · Device enters Sleep
- · Device wakes up from Sleep
- Oscillator fail
- WDT is disabled

See Table 8-2 for more information.

## 8.5 Operation During Sleep

When the device enters Sleep, the WDT is cleared. If the WDT is enabled during Sleep, the WDT resumes counting.

When the device exits Sleep, the WDT is cleared again.

When a WDT time-out occurs while the device is in Sleep, no Reset is generated. Instead, the device wakes up and resumes operation. The TO and PD bits in the STATUS register are changed to indicate the event. See Section 2.0 "Memory Organization" and *Register 2-1* for more information.

# 8.6 Watchdog Control Register

# REGISTER 8-1: WDTCON: WATCHDOG TIMER CONTROL REGISTER

U-0	U-0	R/W-0/0	R/W-1/1	R/W-0/0	R/W-1/1	R/W-1/1	R/W-0/0
_	—			WDTPS<4:02	>		SWDTEN
bit 7		•					bit (
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'	
u = Bit is unch	anged	x = Bit is unki	nown	-n/n = Value a	at POR and BO	R/Value at all	other Resets
'1' = Bit is set		'0' = Bit is cle	ared				
bit 7-6	Unimplome	ented: Read as '	0'				
bit 5-1	-	0>: Watchdog Ti		alact hits(1)			
bit 0-1		Prescale Rate					
		leserved. Result	e in minimum	interval (1.32)			
	•	leserveu. Nesuit	5	intervar(1.52)			
	•						
	•						
	10011 = R	eserved. Result	s in minimum	interval (1:32)			
	10010 <b>= 1</b>	:8388608 (2 <sup>23</sup> ) (	Interval 256s	nominal)			
		:4194304 (2 <sup>22</sup> ) (					
	10000 <b>= 1</b>	:2097152 (2 <sup>21</sup> ) (	Interval 64s n	ominal)			
	01111 <b>= 1</b>	:1048576 (2 <sup>20</sup> ) (	Interval 32s n	ominal)			
	01110 = 1	:524288 (2 <sup>19</sup> ) (Ir :262144 (2 <sup>18</sup> ) (Ir	nterval 16s no	minal)			
	01101 = 1	:262144 (2 <sup>10</sup> ) (Ir	nterval 8s non	ninal)			
		:131072 (2 <sup>17</sup> ) (Ir :65536  (Interval					
		:32768 (Interval	,	(Reset value)			
		:16384 (Interval		nal)			
		:8192 (Interval 2		,			
	00111 <b>= 1</b>	:4096 (Interval 1	28 ms nomina	al)			
		:2048 (Interval 6					
		:1024 (Interval 3		)			
		:512 (Interval 16	,				
		:256 (Interval 8 ı :128 (Interval 4 ı					
		:64 (Interval 2 m					
		:32 (Interval 1 m	,				
bit 0		Software Enable		/atchdog Timer	bit		
	If WDTE<1:			5			
	This bit is ig						
	<u>If WDTE&lt;1:</u>	<b>0&gt; =</b> <u>01</u> :					
	1 = WDT is						
	0 = WDT is						
	If WDTE<1:						
	This bit is ig	norea.					



Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
OSCCON	—		IRCF<2:0>		HFIOFR	—	LFIOFR	HFIOFS	26
STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	13
WDTCON	—	—		١	WDTPS<4:0	>		SWDTEN	48

 TABLE 8-3:
 SUMMARY OF REGISTERS ASSOCIATED WITH WATCHDOG TIMER

**Legend:** x = unknown, u = unchanged, – = unimplemented locations read as '0'. Shaded cells are not used by Watchdog Timer.

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
CONFIG	13:8			_	WRT<1:0>		BORV	LPBOR	LVP	20
CONFIG	7:0	CP	MCLRE	PWRTE	WDTE<1:0>		BOREI	N<1:0>	FOSC	20

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by Watchdog Timer.

# 9.0 FLASH PROGRAM MEMORY CONTROL

The Flash program memory is readable and writable during normal operation over the full VDD range. Program memory is indirectly addressed using Special Function Registers (SFRs). The SFRs used to access program memory are:

- PMCON1
- PMCON2
- PMDATL
- PMDATH
- PMADRL
- PMADRH

When accessing the program memory, the PMDATH:PMDATL register pair forms a 2-byte word that holds the 14-bit data for read/write, and the PMADRH:PMADRL register pair forms a 2-byte word that holds the 9-bit address of the program memory location being read.

The write time is controlled by an on-chip timer. The write/ erase voltages are generated by an on-chip charge pump rated to operate over the operating voltage range of the device.

The Flash program memory can be protected in two ways; by code protection (CP bit in Configuration Word) and write protection (WRT<1:0> bits in Configuration Word).

Code protection  $(\overline{CP} = 0)^{(1)}$ , disables access, reading and writing, to the Flash program memory via external device programmers. Code protection does not affect the self-write and erase functionality. Code protection can only be reset by a device programmer performing a Bulk Erase to the device, clearing all Flash program memory, Configuration bits and User IDs.

Write protection prohibits self-write and erase to a portion or all of the Flash program memory as defined by the bits WRT<1:0>. Write protection does not affect a device programmers ability to read, write or erase the device.

Note 1:	Code	protection	of	the	entire	Flash
	progra	m m <u>em</u> ory	ar	ray i	s enab	led by
	clearin	g the CP bit	of C	Config	uration	Word.

# 9.1 PMADRL and PMADRH Registers

The PMADRH:PMADRL register pair can address up to a maximum of 512 words of program memory. When selecting a program address value, the MSB of the address is written to the PMADRH register and the LSB is written to the PMADRL register.

### 9.1.1 PMCON1 AND PMCON2 REGISTERS

PMCON1 is the control register for Flash program memory accesses.

Control bits RD and WR initiate read and write, respectively. These bits cannot be cleared, only set, in software. They are cleared by hardware at completion of the read or write operation. The inability to clear the WR bit in software prevents the accidental, premature termination of a write operation.

The WREN bit, when set, will allow a write operation to occur. On power-up, the WREN bit is clear. The WRERR bit is set when a write operation is interrupted by a Reset during normal operation. In these situations, following Reset, the user can check the WRERR bit and execute the appropriate error handling routine.

The PMCON2 register is a write-only register. Attempting to read the PMCON2 register will return all '0's.

To enable writes to the program memory, a specific pattern (the unlock sequence), must be written to the PMCON2 register. The required unlock sequence prevents inadvertent writes to the program memory write latches and Flash program memory.

# 9.2 Flash Program Memory Overview

It is important to understand the Flash program memory structure for erase and programming operations. Flash program memory is arranged in rows. A row consists of a fixed number of 14-bit program memory words. A row is the minimum size that can be erased by user software.

After a row has been erased, the user can reprogram all or a portion of this row. Data to be written into the program memory row is written to 14-bit wide data write latches. These write latches are not directly accessible to the user, but may be loaded via sequential writes to the PMDATH:PMDATL register pair.

See Table 9-1 for Erase Row size and the number of write latches for Flash program memory.

**Note:** If the user wants to modify only a portion of a previously programmed row, then the contents of the entire row must be read and saved in RAM prior to the erase. Then, new data and retained data can be written into the write latches to reprogram the row of Flash program memory. How-ever, any unprogrammed locations can be written without first erasing the row. In this case, it is not necessary to save and rewrite the other previously programmed locations.

# TABLE 9-1:FLASH MEMORY<br/>ORGANIZATION BY DEVICE

Device	Row Erase (words)	Write Latches (words)
PIC10(L)F320	16	16
PIC10(L)F322	10	10

# 9.2.1 READING THE FLASH PROGRAM MEMORY

To read a program memory location, the user must:

- 1. Write the desired address to the PMADRH:PMADRL register pair.
- 2. Clear the CFGS bit of the PMCON1 register.
- 3. Then, set control bit RD of the PMCON1 register.

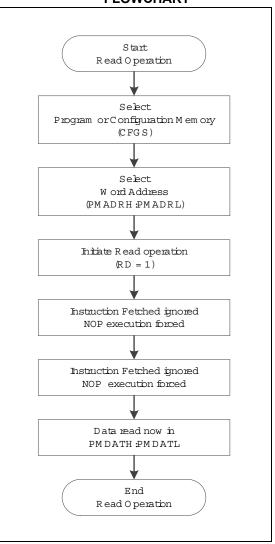
Once the read control bit is set, the program memory Flash controller will use the second instruction cycle to read the data. This causes the second instruction immediately following the "BSF PMCON1, RD" instruction to be ignored. The data is available in the very next cycle, in the PMDATH:PMDATL register pair; therefore, it can be read as two bytes in the following instructions.

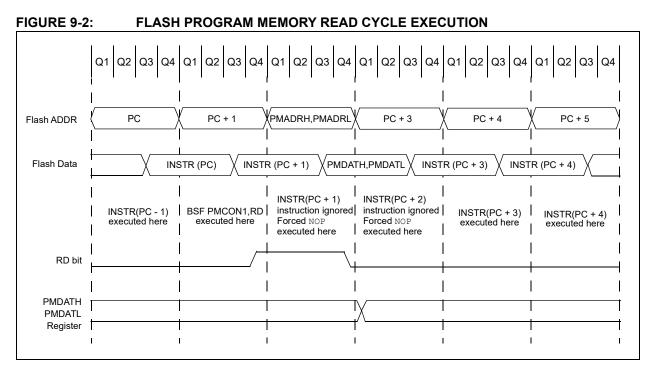
PMDATH:PMDATL register pair will hold this value until another read or until it is written to by the user.

Note:	The two instructions following a program
	memory read are required to be NOPs.
	This prevents the user from executing a
	2-cycle instruction on the next instruction
	after the RD bit is set.

#### FIGURE 9-1:

#### FLASH PROGRAM MEMORY READ FLOWCHART





# EXAMPLE 9-1: FLASH PROGRAM MEMORY READ

\* This code block will read 1 word of program \* memory at the memory address: PROG ADDR HI: PROG ADDR LO data will be returned in the variables; PROG\_DATA\_HI, PROG\_DATA\_LO BANKSEL PMADRL ; not required on devices with 1 Bank of SFRs MOVLW PROG ADDR LO ; MOVWF PMADRL ; Store LSB of address PROG ADDR HI MOVLW ; MOVWF PMADRH ; Store MSB of address BCF PMCON1,CFGS ; Do not select Configuration Space PMCON1,RD BSF ; Initiate read NOP ; Ignored (Figure 9-2) NOP ; Ignored (Figure 9-2) MOVF PMDATL,W ; Get LSB of word MOVWF PROG\_DATA\_LO ; Store in user location ; Get MSB of word MOVF PMDATH,W MOVWF PROG DATA HI ; Store in user location

## 9.2.2 FLASH MEMORY UNLOCK SEQUENCE

Note: A delay of at least 100 μs is required after Power-On Reset (POR) before executing a Flash memory unlock sequence.

The unlock sequence is a mechanism that protects the Flash program memory from unintended self-write programming or erasing. The sequence must be executed and completed without interruption to successfully complete any of the following operations:

- Row Erase
- · Load program memory write latches
- Write of program memory write latches to program memory
- Write of program memory write latches to User IDs

The unlock sequence consists of the following steps:

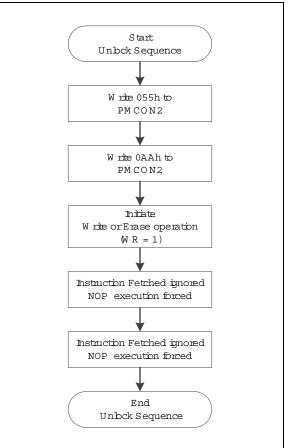
- 1. Write 55h to PMCON2
- 2. Write AAh to PMCON2
- 3. Set the WR bit in PMCON1
- 4. NOP instruction
- 5. NOP instruction

Once the WR bit is set, the processor will always force two NOP instructions. When an Erase Row or Program Row operation is being performed, the processor will stall internal operations (typical 2 ms), until the operation is complete and then resume with the next instruction. When the operation is loading the program memory write latches, the processor will always force the two NOP instructions and continue uninterrupted with the next instruction.

Since the unlock sequence must not be interrupted, global interrupts must be disabled prior to the unlock sequence and re-enabled after the unlock sequence is completed.

#### FIGURE 9-3:

#### FLASH PROGRAM MEMORY UNLOCK SEQUENCE FLOWCHART



#### 9.2.3 ERASING FLASH PROGRAM MEMORY

While executing code, program memory can only be erased by rows. To erase a row:

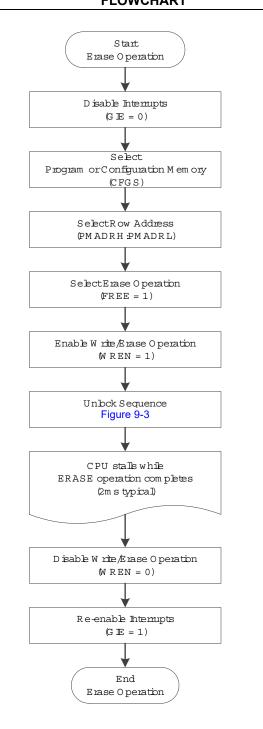
- 1. Load the PMADRH:PMADRL register pair with any address within the row to be erased.
- 2. Clear the CFGS bit of the PMCON1 register.
- 3. Set the FREE and WREN bits of the PMCON1 register.
- 4. Write 55h, then AAh, to PMCON2 (Flash programming unlock sequence).
- 5. Set control bit WR of the PMCON1 register to begin the erase operation.

#### See Example 9-2.

After the "BSF PMCON1, WR" instruction, the processor requires two cycles to set up the erase operation. The user must place two NOP instructions after the WR bit is set. The processor will halt internal operations for the typical 2 ms erase time. This is not Sleep mode as the clocks and peripherals will continue to run. After the erase cycle, the processor will resume operation with the third instruction after the PMCON1 write instruction.

#### FIGURE 9-4:

#### FLASH PROGRAM MEMORY ERASE FLOWCHART



#### EXAMPLE 9-2: **ERASING ONE ROW OF PROGRAM MEMORY** ; This row erase routine assumes the following: ; 1. A valid address within the erase row is loaded in ADDRH:ADDRL ; 2. ADDRH and ADDRL are located in shared data memory $0 \times 70$ - $0 \times 7F$ (common RAM) BCF INTCON,GIE ; Disable ints so required sequences will execute properly BANKSEL PMADRI. ; not required on devices with 1 Bank of SFRs ; Load lower 8 bits of erase address boundary MOVF ADDRL,W MOVWF PMADRL MOVF ADDRH,W ; Load upper 6 bits of erase address boundary MOVWF PMADRH BCF PMCON1,CFGS ; Not configuration space ; Specify an erase operation PMCON1, FREE BSF BSF PMCON1,WREN ; Enable writes MOVLW 55h ; Start of required sequence to initiate erase MOVWF PMCON2 ; Write 55h Required Sequence MOVLW 0AAh : MOVWF PMCON2 ; Write AAh BSF PMCON1,WR ; Set WR bit to begin erase NOP ; NOP instructions are forced as processor starts NOP ; row erase of program memory. ; The processor stalls until the erase process is complete ; after erase processor continues with 3rd instruction BCF PMCON1,WREN ; Disable writes BSF INTCON, GIE ; Enable interrupts

# 9.2.4 WRITING TO FLASH PROGRAM MEMORY

Program memory is programmed using the following steps:

- 1. Load the address in PMADRH:PMADRL of the row to be programmed.
- 2. Load each write latch with data.
- 3. Initiate a programming operation.
- 4. Repeat steps 1 through 3 until all data is written.

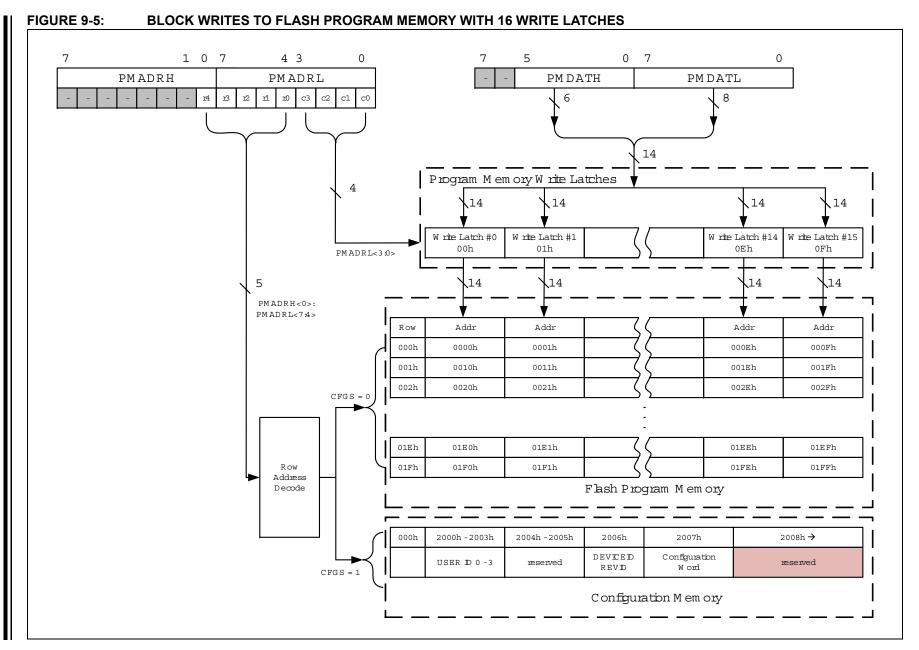
Before writing to program memory, the word(s) to be written must be erased or previously unwritten. Program memory can only be erased one row at a time. No automatic erase occurs upon the initiation of the write.

Program memory can be written one or more words at a time. The maximum number of words written at one time is equal to the number of write latches. See Figure 9-5 (row writes to program memory with 16 write latches) for more details.

The write latches are aligned to the Flash row address boundary defined by the upper ten bits of PMADRH:PMADRL, (PMADRH<6:0>:PMADRL<7:5>) with the lower five bits of PMADRL, (PMADRL<7:5>) determining the write latch being loaded. Write operations do not cross these boundaries. At the completion of a program memory write operation, the data in the write latches is reset to contain 0x3FFF. The following steps must be completed to load the write latches and program a row of program memory. These steps are divided into two parts. First, each write latch is loaded with data from the PMDATH:PMDATL using the unlock sequence with LWLO = 1. When the last word to be loaded into the write latch is ready, the LWLO bit is cleared and the unlock sequence executed. This initiates the programming operation, writing all the latches into Flash program memory.

- Note: The special unlock sequence is required to load a write latch with data or initiate a Flash programming operation. If the unlock sequence is interrupted, writing to the latches or program memory will not be initiated.
- 1. Set the WREN bit of the PMCON1 register.
- 2. Clear the CFGS bit of the PMCON1 register.
- Set the LWLO bit of the PMCON1 register. When the LWLO bit of the PMCON1 register is '1', the write sequence will only load the write latches and will not initiate the write to Flash program memory.
- 4. Load the PMADRH:PMADRL register pair with the address of the location to be written.
- 5. Load the PMDATH:PMDATL register pair with the program memory data to be written.
- Execute the unlock sequence (Section 9.2.2 "Flash Memory Unlock Sequence"). The write latch is now loaded.
- 7. Increment the PMADRH:PMADRL register pair to point to the next location.
- 8. Repeat steps 5 through 7 until all but the last write latch has been loaded.
- Clear the LWLO bit of the PMCON1 register. When the LWLO bit of the PMCON1 register is '0', the write sequence will initiate the write to Flash program memory.
- 10. Load the PMDATH:PMDATL register pair with the program memory data to be written.
- 11. Execute the unlock sequence (Section 9.2.2 "Flash Memory Unlock Sequence"). The entire program memory latch content is now written to Flash program memory.
  - **Note:** The program memory write latches are reset to the blank state (0x3FFF) at the completion of every write or erase operation. As a result, it is not necessary to load all the program memory write latches. Unloaded latches will remain in the blank state.

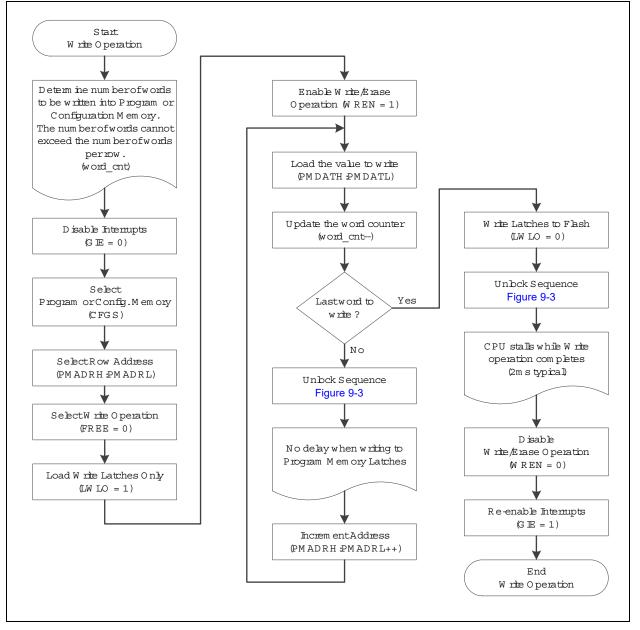
An example of the complete write sequence is shown in Example 9-3. The initial address is loaded into the PMADRH:PMADRL register pair; the data is loaded using indirect addressing.



DS40001585E-page 57

PIC10(L)F320/322





#### EXAMPLE 9-3: WRITING TO FLASH PROGRAM MEMORY

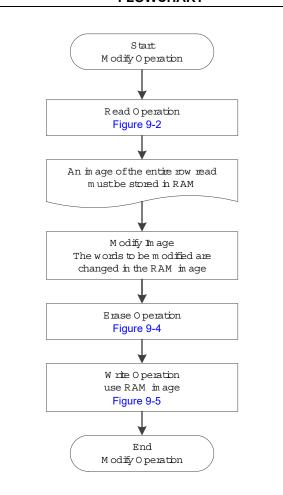
```
; This write routine assumes the following:
         A valid starting address (the least significant bits = '00')
  ;
         is loaded in ADDRH:ADDRL
  :
         ADDRH, ADDRL and DATADDR are all located in data memory
  ;
  BANKSEL
              PMADRH
  MOVF ADDRH,W ;Load initial address
  MOVWF PMADRH
                  ;
  MOVF
        ADDRL,W
                  ;
  MOVWF PMADRL
                  ;
       DATAADDR,W ;Load initial data address
  MOVE
 MOVWF FSR
                  ;
LOOP MOVE INDE,W
                 ;Load first data byte into lower
  MOVWF PMDATL
                 ;
  INCF FSR, F
                 ;Next byte
        INDF,W
 MOVF
                 ;Load second data byte into upper
 MOVWF PMDATH
                 ;
  INCF
        FSR,F
                  ;
  BANKSEL PMCON1
  BSF PMCON1, WREN ; Enable writes
  BCF
      INTCON,GIE ;Disable interrupts (if using)
  BTFSC INTCON, GIE ; See AN576
  GOTO
       $-2
  ;
        Required Sequence
               ;Start of required write sequence:
  MOVLW 55h
                 ;Write 55h
  MOVWF
        PMCON2
  MOVLW
        0AAh
  MOVWF PMCON2
                  ;Write OAAh
        PMCON1,WR ;Set WR bit to begin write
  BSF
  NOP
                  ;Required to transfer data to the buffer
  NOP
                  ;registers
  BCF
       PMCON1,WREN ;Disable writes
        INTCON,GIE ;Enable interrupts (comment out if not using interrupts)
  BSF
  BANKSEL PMADRL
  MOVF
      PMADRL, W
  INCF
        PMADRL,F ;Increment address
  ANDLW 0x03
                  ; Indicates when sixteen words have been programmed
  SUBLW 0x03
                  ;Change value for different size write blocks
                  ;0x0F = 16 words
                  ;0x0B = 12 words
                  ;0x07 = 8 words
                  ;0x03 = 4 words
  BTFSS STATUS,Z
                 ;Exit on a match,
  GOTO
        LOOP
                  ;Continue if more data needs to be written
```

# 9.3 Modifying Flash Program Memory

When modifying existing data in a program memory row, and data within that row must be preserved, it must first be read and saved in a RAM image. Program memory is modified using the following steps:

- 1. Load the starting address of the row to be modified.
- 2. Read the existing data from the row into a RAM image.
- 3. Modify the RAM image to contain the new data to be written into program memory.
- 4. Load the starting address of the row to be rewritten.
- 5. Erase the program memory row.
- 6. Load the write latches with data from the RAM image.
- 7. Initiate a programming operation.

#### FIGURE 9-7: FLASH PROGRAM MEMORY MODIFY FLOWCHART



# 9.4 User ID, Device ID and Configuration Word Access

Instead of accessing program memory, the User ID's, Device ID/Revision ID and Configuration Word can be accessed when CFGS = 1 in the PMCON1 register. This is the region that would be pointed to by PC<13> = 1, but not all addresses are accessible. Different access may exist for reads and writes. Refer to Table 9-2.

When read access is initiated on an address outside the parameters listed in Table 9-2, the PMDATH:PMDATL register pair is cleared, reading back '0's.

Address	Function	Read Access	Write Access
2000h-2003h	User IDs	Yes	Yes
2006h	Device ID/Revision ID	Yes	No
2007h	Configuration Word	Yes	No

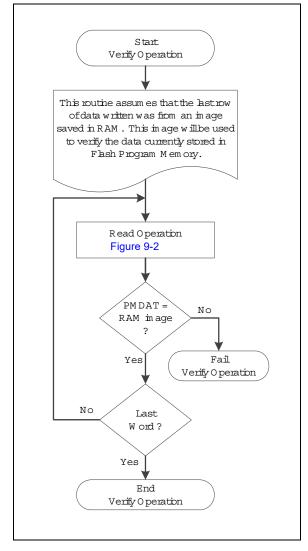
#### EXAMPLE 9-4: CONFIGURATION WORD AND DEVICE ID ACCESS

* T * *	This code block will read 1 word of program memory at the memory address: PROG_ADDR_LO (must be 00h-08h) data will be returned in the variables; PROG_DATA_HI, PROG_DATA_LO						
	BANKSEL	PMADRL	; not required on devices with 1 Bank of SFRs				
	MOVLW	PROG_ADDR_LO	;				
	MOVWF	PMADRL	; Store LSB of address				
	CLRF	PMADRH	; Clear MSB of address				
	BSF	PMCON1,CFGS	; Select Configuration Space				
	BCF	INTCON, GIE	; Disable interrupts				
	BSF	PMCON1,RD	; Initiate read				
	NOP		; Executed (See Figure 9-2)				
	NOP		; Ignored (See Figure 9-2)				
	BSF	INTCON, GIE	; Restore interrupts				
	MOVF	PMDATL,W	; Get LSB of word				
	MOVWF	PROG DATA LO	; Store in user location				
	MOVF	PMDATH,W	; Get MSB of word				
	MOVWF	PROG_DATA_HI	; Store in user location				

# 9.5 Write Verify

It is considered good programming practice to verify that program memory writes agree with the intended value. Since program memory is stored as a full page then the stored program memory contents are compared with the intended data stored in RAM after the last write is complete.

FIGURE 9-8: FLASH PROGRAM MEMORY VERIFY FLOWCHART



# 9.6 Flash Program Memory Control Registers

# REGISTER 9-1: PMDATL: PROGRAM MEMORY DATA LOW

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			PMDA	AT<7:0>			
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	l as '0'	
u = Bit is uncha	anged	x = Bit is unkn	nown	-n/n = Value a	at POR and BO	R/Value at all c	ther Resets
'1' = Bit is set		'0' = Bit is clea	ared				

bit 7-0 **PMDAT<7:0>**: The value of the program memory word pointed to by PMADRH and PMADRL after a Program Memory Read command.

#### REGISTER 9-2: PMDATH: PROGRAM MEMORY DATA HIGH

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
—	—			PMDA	T<13:8>		
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 Unimplemented: Read as '0'

bit 5-0 **PMDAT<13:8>**: The value of the program memory word pointed to by PMADRH and PMADRL after a Program Memory Read command.

# REGISTER 9-3: PMADRL: PROGRAM MEMORY ADDRESS LOW

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
			PMAD	R<7:0>			
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable I	bit	U = Unimpler	mented bit, read	l as '0'	
u = Bit is unch	anged	x = Bit is unkn	nown	-n/n = Value a	at POR and BO	R/Value at all o	ther Resets
'1' = Bit is set		'0' = Bit is clea	ared				

bit 7-0 **PMADR<7:0>**: Program Memory Read Address low bits

# REGISTER 9-4: PMADRH: PROGRAM MEMORY ADDRESS HIGH

U-0	U-0	U-0	U-0	U-0	U-0	U-0	R/W-0/0
—	—	—	_	_	—	_	PMADR8
bit 7							bit 0
Legend:							
R = Readable b	bit	W = Writable	bit	U = Unimpler	mented bit, read	as '0'	
u = Bit is uncha	inged	x = Bit is unkn	iown	-n/n = Value a	at POR and BO	R/Value at all o	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				

bit 7-1 Unimplemented: Read as '0'

bit 0 PMADR8: Program Memory Read Address High bit

U-1(1	1) R/W-0/0	R/W-0/0	R/W/HC-0/0	R/W/HC-0/q <sup>(2)</sup>	R/W-0/0	R/S/HC-0/0	R/S/HC-0/0
	CFGS	LWLO	FREE	WRERR	WREN	WR	RD
bit 7							bit 0
Legend:							
R = Read	lable bit	W = Writable	bit	U = Unimplemer	nted bit, read a	as '0'	
S = Bit ca	an only be set	x = Bit is unk	nown	-n/n = Value at F	POR and BOR	/Value at all othe	er Resets
'1' = Bit is	s set	'0' = Bit is cle	ared	HC = Bit is clear	ed by hardwar	re	
bit 7	Unimplemen	ted: Read as	'1'				
bit 6	CFGS: Config	guration Selec	t bit				
				evice ID Registers	6		
		-lash program	-				
bit 5		Write Latches					
				write latch is load			
			•	e latch is loaded/ ext WR command	•	a write of all pro	gram memor
bit 4	FREE: Progra	am Flash Eras	e Enable bit				
				ext WR command	d (hardware cle	eared upon com	pletion)
	0 = Performs	s an write oper	ation on the ne	ext WR command			
bit 3		gram/Erase E					
				gram or erase s		npt or terminat	ion (bit is se
		• •	• •	e '1') of the WR b bleted normally.	it).		
bit 2		ram/Erase Ena		neteu normany.			
	•	rogram/erase cha					
			rasing of progr	am Flash			
bit 1	WR: Write Co	ontrol bit					
	1 = Initiates a	a program Fla	sh program/era	se operation.			
				is cleared by har	dware once o	peration is com	olete.
				ed) in software.	inactivo		
L:4 0		•	on to the Flash	is complete and i	inactive.		
bit 0	RD: Read Co		sh road Road	takes one cycle.	PD is cleared	in hardwara. T	ha PD hit ca
		set (not cleared		lakes one cycle.	RD IS cleared	in naruware. T	
	-	•	ram Flash read	d.			
Note 1:	Unimplemented bi						
2:	The WRERR bit is		set by hardwar	e when a progran	n memory write	e or erase opera	tion is started
	(WR = 1).						
2.	The LVA/LO bit is in	marad during a	program man	on aroon on or ot	ian / EDEE = 1	1	

# REGISTER 9-5: PMCON1: PROGRAM MEMORY CONTROL 1 REGISTER

**3**: The LWLO bit is ignored during a program memory erase operation (FREE = 1).

W-0/0	W-0/0	W-0/0	W-0/0	W-0/0	W-0/0	W-0/0	W-0/0
		Prog	gram Memory	Control Regist	er 2		
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	l as '0'	
S = Bit can onl	y be set	x = Bit is unkn	nown	-n/n = Value a	at POR and BO	R/Value at all c	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				

# REGISTER 9-6: PMCON2: PROGRAM MEMORY CONTROL 2 REGISTER

#### bit 7-0 Flash Memory Unlock Pattern bits

To unlock writes, a 55h must be written first, followed by an AAh, before setting the WR bit of the PMCON1 register. The value written to this register is used to unlock the writes. There are specific timing requirements on these writes.

#### TABLE 9-3: SUMMARY OF REGISTERS ASSOCIATED WITH FLASH PROGRAM MEMORY

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	40
PMCON1	_	CFGS	LWLO	FREE	WRERR	WREN	WR	RD	65
PMCON2	Program Memory Control Register 2							66	
PMADRL				PMAD	R<7:0>				64
PMADRH	_	_	_	_	_	_	_	PMADR8	64
PMDATL	PMDAT<7:0>							63	
PMDATH	_	— — PMDAT<13:8>							63
Lawawala		فمحما المحفيت مع							

**Legend:** — = unimplemented location, read as '0'. Shaded cells are not used by Flash program memory module.

#### TABLE 9-4: SUMMARY OF CONFIGURATION WORD WITH FLASH PROGRAM MEMORY

Name	Bits	Bit -/7	Bit -/6	Bit 13/5	Bit 12/4	Bit 11/3	Bit 10/2	Bit 9/1	Bit 8/0	Register on Page
	13:8	_	-	—	WRT	WRT<1:0>		LPBOR	LVP	20
CONFIG	7:0	CP	MCLR	PWRTE	WDTE	WDTE<1:0>		N<1:0>	FOSC	20

Legend: — = unimplemented location, read as '0'. Shaded cells are not used by Flash program memory.

# 10.0 I/O PORT

Depending on which peripherals are enabled, some or all of the pins may not be available as general purpose I/O. In general, when a peripheral is enabled on a port pin, that pin cannot be used as a general purpose output. However, the pin can still be read.

PORTA has three standard registers for its operation. These registers are:

- TRISA register (data direction)
- PORTA register (reads the levels on the pins of the device)
- LATA register (output latch)

Some ports may have one or more of the following additional registers. These registers are:

- ANSELA (analog select)
- WPUA (weak pull-up)

The Data Latch (LATA) register is useful for readmodify-write operations on the value that the I/O pins are driving.

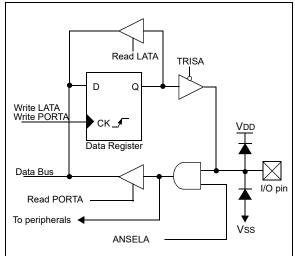
A write operation to the LATA register has the same effect as a write to the corresponding PORTA register. A read of the LATA register reads of the values held in the I/O PORT latches, while a read of the PORTA register reads the actual I/O pin value.

Ports that support analog inputs have an associated ANSELA register. When an ANSEL bit is set, the digital input buffer associated with that bit is disabled. Disabling the input buffer prevents analog signal levels on the pin between a logic high and low from causing excessive current in the logic input circuitry. A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 10-1.

#### EXAMPLE 10-1: INITIALIZING PORTA

```
; This code example illustrates
; initializing the PORTA register. The
; other ports are initialized in the same
; manner.
BANKSEL
          PORTA
                         ;not required on devices with 1 Bank of SFRs
          PORTA
CLRF
                         ;Init PORTA
BANKSEL
          LATA
                          ;not required on devices with 1 Bank of SFRs
CLRF
          LATA
                         ;
BANKSEL
          ANSELA
                         ;not required on devices with 1 Bank of SFRs
CLRF
          ANSELA
                         ;digital I/O
BANKSEL
          TRISA
                         ;not required on devices with 1 Bank of SFRs
MOVLW
          B'00000011'
                         ;Set RA<1:0> as inputs
MOVWF
          TRISA
                         ;and set RA<2:3> as
                          ;outputs
```

#### FIGURE 10-1: I/O PORT OPERATION



# 10.1 PORTA Registers

PORTA is a 8-bit wide, bidirectional port. The corresponding data direction register is TRISA (Register 10-2). Setting a TRISA bit (= 1) will make the corresponding PORTA pin an input (i.e., disable the output driver). Clearing a TRISA bit (= 0) will make the corresponding PORTA pin an output (i.e., enables output driver and puts the contents of the output latch on the selected pin). Example 10-1 shows how to initialize PORTA.

Reading the PORTA register (Register 10-1) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATA).

The TRISA register (Register 10-2) controls the PORTA pin output drivers, even when they are being used as analog inputs. The user will ensure the bits in the TRISA register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

#### 10.1.1 WEAK PULL-UPS

Each of the PORTA pins has an individually configurable internal weak pull-up. Control bits WPUA<3:0> enable or disable each pull-up (see Register 10-5). Each weak pull-up is automatically turned off when the port pin is configured as an output. All pull-ups are disabled on a Power-on Reset by the WPUEN bit of the OPTION\_REG register.

#### 10.1.2 ANSELA REGISTER

The ANSELA register (Register 10-4) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELA bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELA bits has no effect on digital output functions. A pin with TRIS clear and ANSEL set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note: The ANSELA bits default to the Analog mode after Reset. To use any pins as digital general purpose or peripheral inputs, the corresponding ANSEL bits must be initialized to '0' by user software.

#### 10.1.3 PORTA FUNCTIONS AND OUTPUT PRIORITIES

Each PORTA pin is multiplexed with other functions. The pins, their combined functions and their output priorities are shown in Table 10-1.

When multiple outputs are enabled, the actual pin control goes to the peripheral with the highest priority.

Digital output functions may control the pin when it is in Analog mode with the priority shown in Table 10-1.

Pin Name	Function Priority <sup>(1)</sup>
RA0	ICSPDAT
	CWG1A
	PWM1
	RA0
RA1	CWG1B
	PWM2
	CLC1
	RA1
RA2	NCO1
	CLKR
	RA2
RA3	None

#### TABLE 10-1: PORTA OUTPUT PRIORITY

**Note 1:** Priority listed from highest to lowest.

# 10.2 Register Definitions: PORTA

# **REGISTER 10-1: PORTA: PORTA REGISTER**

U-0	U-0	U-0	U-0	R-x/x	R/W-x/x	R/W-x/x	R/W-x/x
—	—	—	—	RA3	RA2	RA1	RA0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4 Unimplemented: Read as '0'

bit 3-0 RA<3:0>: PORTA I/O Value bits (RA3 is read-only)

**Note 1:** Writes to PORTx are actually written to the corresponding LATx register. Reads from PORTx register return actual I/O pin values.

### REGISTER 10-2: TRISA: PORTA TRI-STATE REGISTER

U-0	U-0	U-0	U-0	U-1	R/W-1/1	R/W-1/1	R/W-1/1
—	—	—	—	(1)	TRISA2	TRISA1	TRISA0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4	Unimplemented: Read as '0'.
bit 3	Unimplemented: Read as '1'.
bit 2-0	TRISA<2:0>: RA<2:0> Port I/O Tri-State Control bits
	<ul><li>1 = Port output driver is disabled</li><li>0 = Port output driver is enabled</li></ul>

**Note 1:** Unimplemented, read as '1'.

### REGISTER 10-3: LATA: PORTA DATA LATCH REGISTER

U-0	U-0	U-0	U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	
—	_	—		—	LATA2	LATA1	LATA0	
bit 7							bit 0	
Legend:								
R = Readable b	bit	W = Writable I	oit	U = Unimpler	nented bit, read	l as '0'		
u = Bit is unchanged x = Bit is unknown				-n/n = Value at POR and BOR/Value at all other Resets				
'1' = Bit is set		'0' = Bit is clea	ared					

bit 7-3 Unimplemented: Read as '0'.

bit 2-0 LATA<2:0>: RA<2:0> Output Latch Value bits

**Note 1:** Writes to PORTx are actually written to the corresponding LATx register. Reads from LATx register return register values, not I/O pin values.

# REGISTER 10-4: ANSELA: PORTA ANALOG SELECT REGISTER

U-0	U-0	U-0	U-0	U-0	R/W-1/1	R/W-1/1	R/W-1/1		
—			_	—	ANSA2	ANSA1	ANSA0		
bit 7 bit 0									

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-3 Unimplemented: Read as '0'.

bit 2-0 ANSA<2:0>: Analog Select between Analog or Digital Function on Pins RA<2:0>, respectively

- 1 = Analog input. Pin is assigned as analog input<sup>(1)</sup>. Digital Input buffer disabled.
- 0 = Digital I/O. Pin is assigned to port or Digital special function.
- **Note 1:** Setting a pin to an analog input automatically disables the digital input circuitry. Weak pull-ups, if available, are unaffected. The corresponding TRIS bit must be set to Input mode by the user in order to allow external control of the voltage on the pin.

REGISTER 10-5: WPUA: WEAK PULL-UP PORTA REGISTER									
U-0	U-0	U-0 U-0		R/W-1/1 R/W-1/1		R/W-1/1	R/W-1/1		
—	—	—	—	WPUA3	WPUA2	WPUA1	WPUA0		
bit 7							bit 0		
Legend:									
R = Readable bit		W = Writable bit		U = Unimplemented bit, read as '0'					
u = Bit is unchanged		x = Bit is unknown		-n/n = Value at POR and BOR/Value at all other Resets					
'1' = Bit is set		'0' = Bit is clea	ared						
bit 7-4	Unimplemen	Unimplemented: Read as '0'.							
bit 3-0	1 = Weak Pul	WPUA<3:0>: Weak Pull-up PORTA Control bits 1 = Weak Pull-up enabled <sup>(1)</sup> 0 = Weak Pull-up disabled.							

# **Note 1:** Enabling weak pull-ups also requires that the WPUEN bit of the OPTION\_REG register be cleared (Register 16-1).

## TABLE 10-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	_	—	_	—	ANSA2	ANSA1	ANSA0	70
IOCAF	—	_	—	_	IOCAF3	IOCAF2	IOCAF1	IOCAF0	76
IOCAN	—	_	—	_	IOCAN3	IOCAN2	IOCAN1	IOCAN0	75
IOCAP	—	_	—	_	IOCAP3	IOCAP2	IOCAP1	IOCAP0	75
LATA	—	_	—	_	—	LATA2	LATA1	LATA0	70
PORTA	—	_	—	_	RA3	RA2	RA1	RA0	69
TRISA	—	—	—	—	(1)	TRISA2	TRISA1	TRISA0	69
WPUA	—		_		WPUA3	WPUA2	WPUA1	WPUA0	71

Legend: x = unknown, u = unchanged, – = unimplemented locations read as '0'. Shaded cells are not used by PORTA.

Note 1: Unimplemented, read as '1'.

# 11.0 INTERRUPT-ON-CHANGE

The PORTA pins can be configured to operate as Interrupt-On-Change (IOC) pins. An interrupt can be generated by detecting a signal that has either a rising edge or a falling edge. Any individual PORTA pin, or combination of PORTA pins, can be configured to generate an interrupt. The Interrupt-on-change module has the following features:

- Interrupt-on-Change enable (Main Switch)
- Individual pin configuration
- · Rising and falling edge detection
- Individual pin interrupt flags

Figure 11-1 is a block diagram of the IOC module.

## 11.1 Enabling the Module

To allow individual PORTA pins to generate an interrupt, the IOCIE bit of the INTCON register must be set. If the IOCIE bit is disabled, the edge detection on the pin will still occur, but an interrupt will not be generated.

### 11.2 Individual Pin Configuration

For each PORTA pin, a rising edge detector and a falling edge detector are present. To enable a pin to detect a rising edge, the associated IOCAPx bit of the IOCAP register is set. To enable a pin to detect a falling edge, the associated IOCANx bit of the IOCAN register is set.

A pin can be configured to detect rising and falling edges simultaneously by setting both the IOCAPx bit and the IOCANx bit of the IOCAP and IOCAN registers, respectively.

## 11.3 Interrupt Flags

The IOCAFx bits located in the IOCAF register are status flags that correspond to the interrupt-on-change pins of PORTA. If an expected edge is detected on an appropriately enabled pin, then the status flag for that pin will be set, and an interrupt will be generated if the IOCIE bit is set. The IOCIF bit of the INTCON register reflects the status of all IOCAFx bits.

# 11.4 Clearing Interrupt Flags

The individual status flags, (IOCAFx bits), can be cleared by resetting them to zero. If another edge is detected during this clearing operation, the associated status flag will be set at the end of the sequence, regardless of the value actually being written.

To ensure that no detected edge is lost while clearing flags, only AND operations masking out known changed bits must be performed. The following sequence is an example of what must be performed.

### EXAMPLE 11-1: CLEARING INTERRUPT FLAGS

MOVLW 0xff XORWF IOCAF, W ANDWF IOCAF, F

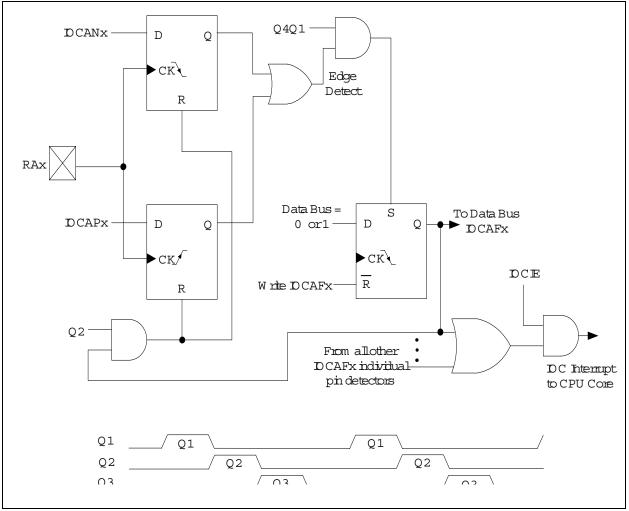
# 11.5 Operation in Sleep

The interrupt-on-change interrupt sequence will wake the device from Sleep mode, if the IOCIE bit is set.

If an edge is detected while in Sleep mode, the IOCAF register will be updated prior to the first instruction executed out of Sleep.

# PIC10(L)F320/322





# 11.6 Interrupt-On-Change Registers

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
_	—	—	_	IOCAP3	IOCAP2	IOCAP1	IOCAP0	
bit 7	•						bit 0	
Legend:								
R = Readable bit W = Writable bit			U = Unimplemented bit, read as '0'					
u = Bit is unchanged x = Bit is unki			iown	-n/n = Value a	at POR and BO	R/Value at all c	ther Resets	
'1' = Bit is set '0' = Bit is cleared								

### **REGISTER 11-1: IOCAP: INTERRUPT-ON-CHANGE PORTA POSITIVE EDGE REGISTER**

bit 7-4 Unimplemented: Read as '0'.	
-------------------------------------	--

bit 3-0 **IOCAP<3:0>:** Interrupt-on-Change PORTA Positive Edge Enable bits

- 1 = Interrupt-on-Change enabled on the pin for a positive going edge. Associated Status bit and interrupt flag will be set upon detecting an edge.<sup>(1)</sup>
- 0 = Interrupt-on-Change disabled for the associated pin.

Note 1: Interrupt-on-change also requires that the IOCIE bit of the INTCON register be set (Register 6-1).

# REGISTER 11-2: IOCAN: INTERRUPT-ON-CHANGE PORTA NEGATIVE EDGE REGISTER

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	_	IOCAN3	IOCAN2	IOCAN1	IOCAN0
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4	Unimplemented: Read as '0'.
bit 3-0	IOCAN<3:0>: Interrupt-on-Change PORTA Negative Edge Enable bits
	<ul> <li>1 = Interrupt-on-Change enabled on the pin for a negative going edge. Associated Status bit and interrupt flag will be set upon detecting an edge.<sup>(1)</sup></li> <li>0 = Interrupt-on-Change disabled for the associated pin.</li> </ul>

Note 1: Interrupt-on-change also requires that the IOCIE bit of the INTCON register be set (Register 6-1).

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0		
—	—	—	—	IOCAF3	IOCAF2	IOCAF1	IOCAF0		
bit 7							bit 0		
Legend:									
R = Readable bit W = Writable bit			bit	U = Unimplemented bit, read as '0'					
u = Bit is unchanged x = Bit is unknown			nown	-n/n = Value at POR and BOR/Value at all other Resets					
'1' = Bit is set		'0' = Bit is clea	ared	HS - Bit is set in hardware					

# REGISTER 11-3: IOCAF: INTERRUPT-ON-CHANGE PORTA FLAG REGISTER

bit 7-4 Unimplemented: Read as '0'.

bit 3-0 **IOCAF<3:0>:** Interrupt-on-Change PORTA Flag bits

1 = An enable change was detected on the associated pin.

Set when IOCAPx = 1 and a rising edge was detected on RAx, or when IOCANx = 1 and a falling edge was detected on RAx.<sup>(1)</sup>

0 = No change was detected, or the user cleared the detected change.

Note 1: Interrupt-on-change also requires that the IOCIE bit of the INTCON register be set (Register 6-1).

TABLE 11-1: SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPT-ON-CHA	ANGE
---	------

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	40
IOCAF	—	_	—	_	IOCAF3	IOCAF2	IOCAF1	IOCAF0	76
IOCAN	—	_	—	_	IOCAN3	IOCAN2	IOCAN1	IOCAN0	75
IOCAP	—	_	—	_	IOCAP3	IOCAP2	IOCAP1	IOCAP0	75
TRISA	—	_	—	_	(1)	TRISA2	TRISA1	TRISA0	69

**Legend:** — = unimplemented location, read as '0'. Shaded cells are not used by Interrupt-on-Change.

Note 1: Unimplemented, read as '1'.

# 12.0 FIXED VOLTAGE REFERENCE (FVR)

The Fixed Voltage Reference, or FVR, is a stable voltage reference, independent of VDD, with 1.024V, 2.048V or 4.096V selectable output levels. The output of the FVR can be configured to supply a reference voltage to the following:

ADC input channel

The FVR can be enabled by setting the FVREN bit of the FVRCON register.

# 12.1 Independent Gain Amplifiers

The output of the FVR supplied to the ADC is routed through an independent programmable gain amplifier. The amplifier can be configured to amplify the reference voltage by 1x, 2x or 4x, to produce the three possible voltage levels.

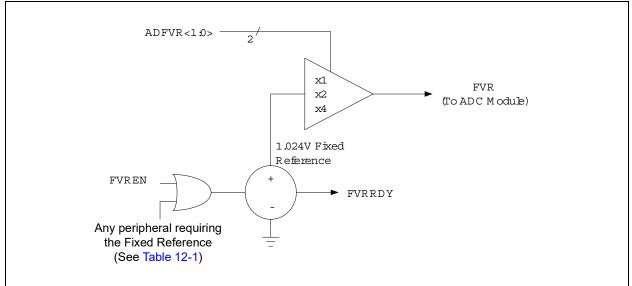
The ADFVR<1:0> bits of the FVRCON register are used to enable and configure the gain amplifier settings for the reference supplied to the ADC module. Reference Section 15.0 "Analog-to-Digital Converter (ADC) Module" for additional information.

To minimize current consumption when the FVR is disabled, the FVR buffers must be turned off by clearing the ADFVR<1:0> bits.

## 12.2 FVR Stabilization Period

When the Fixed Voltage Reference module is enabled, it requires time for the reference and amplifier circuits to stabilize. Once the circuits stabilize and are ready for use, the FVRRDY bit of the FVRCON register will be set. See **Section 24.0** "**Electrical Specifications**" for the minimum delay requirement.

### FIGURE 12-1: VOLTAGE REFERENCE BLOCK DIAGRAM



### TABLE 12-1: PERIPHERALS REQUIRING THE FIXED VOLTAGE REFERENCE (FVR)

Peripheral	Conditions	Description				
HFINTOSC	FOSC = 1	EC on CLKIN pin.				
	BOREN<1:0> = 11	BOR always enabled.				
BOR	BOREN<1:0> = 10 and BORFS = 1	BOR disabled in Sleep mode, BOR Fast Start enabled.				
	BOREN<1:0> = 01 and BORFS = 1	BOR under software control, BOR Fast Start enabled.				
IVR	All PIC10F320/322 devices, when VREGPM1 = 1 and not in Sleep	The device runs off of the Power-Save mode regulator when in Sleep mode.				

# 12.3 FVR Control Registers

### REGISTER 12-1: FVRCON: FIXED VOLTAGE REFERENCE CONTROL REGISTER

R/W-0/0	R-q/q	R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0		
FVREN FVRRDY <sup>(</sup>		TSEN <sup>(3)</sup>	TSRNG <sup>(3)</sup>	—	_	ADFV	R<1:0>		
bit 7							bit 0		
Legend:									
R = Readable	e bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'			
u = Bit is unc	hanged	x = Bit is unk	nown	-n/n = Value	at POR and BO	R/Value at all o	ther Resets		
'1' = Bit is set		'0' = Bit is cle	ared	q = Value de	pends on condit	ion			
bit 7	1 = Fixed Vo 0 = Fixed Vo	d Voltage Refe Itage Referenc Itage Referenc	e is enabled e is disabled						
bit 6	1 = Fixed Vo	ed Voltage Re Itage Referenc Itage Referenc	e output is rea		enabled				
bit 5	1 = Tempera	erature Indicate ture Indicator i ture Indicator i	s enabled	)					
bit 4	1 = VOUT = V	<b>TSRNG:</b> Temperature Indicator Range Selection bit <sup>(3)</sup> 1 = VOUT = VDD - 4VT (High Range) 0 = VOUT = VDD - 2VT (Low Range)							
bit 3-2	Unimplemen	<b>ted</b> : Read as '	0'						
bit 1-0	ADFVR<1:0>: ADC Fixed Voltage Reference Selection bit 11 = ADC Fixed Voltage Reference Peripheral output is 4x (4.096V) <sup>(2)</sup> 10 = ADC Fixed Voltage Reference Peripheral output is 2x (2.048V) <sup>(2)</sup> 01 = ADC Fixed Voltage Reference Peripheral output is 1x (1.024V) 00 = ADC Fixed Voltage Reference Peripheral output is off.								
	RRDY indicates	the true state	of the FVR.						

- 2: Fixed Voltage Reference output cannot exceed VDD.
- 3: See Section 14.0 "Temperature Indicator Module" for additional information.

<b>TABLE 12-2</b> :	SUMMARY OF REGISTERS ASSOCIATED WITH FIXED VOLTAGE REFERENCE
---------------------	--

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on page
FVRCON	FVREN	FVRRDY	TSEN	TSRNG			ADFVR<1:0>		78

**Legend:** Shaded cells are not used with the Fixed Voltage Reference.

# 13.0 INTERNAL VOLTAGE REGULATOR (IVR)

The Internal Voltage Regulator (IVR), which provides operation above 3.6V is available on:

- PIC10F320
- PIC10F322

This circuit regulates a voltage for the internal device logic while permitting the VDD and I/O pins to operate at a higher voltage. When VDD approaches the regulated voltage, the IVR output automatically tracks the input voltage.

The IVR operates in one of three power modes based on user configuration and peripheral selection. The operating power modes are:

- High
- Low
- Power-Save Sleep mode

Power modes are selected automatically depending on the device operation, as shown in Table 13-1. Tracking mode is selected automatically when VDD drops below the safe operating voltage of the core.

Note: IVR is disabled in Tracking mode, but will consume power. See Section 24.0 "Electrical Specifications" for more information.

### TABLE 13-1: IVR POWER MODES - REGULATED

VREGPM1 Bit	Sleep Mode	Memory Bias Power Mode	IVR Power Mode	
		EC Mode or INTOSC = 16 MHz (HP Bias)	Jida	
х	No	INTOSC = 1 to 8 MHz (MP Bias)	High	
		INTOSC = 31 kHz to 500 kHz (LP Bias)	Low	
0	Yes	Don't Care	Low	
1	Vee	No HFINTOSC	Power Save <sup>(1)</sup>	
1	Yes	No Peripherals	Power Save	

**Note 1:** Forced to Low-Power mode by any of the following conditions:

- BOR is enabled
- HFINTOSC is an active peripheral source
- Self-write is active
- ADC is in an active conversion

### REGISTER 13-1: VREGCON: VOLTAGE REGULATOR CONTROL REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-1/1
_	_		_	_	—	VREGPM1	Reserved
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-2 Unimplemented: Read as '0'.

bit 1 **VREGPM1:** Voltage Regulator Power Mode Selection bit

1 = Power-Save Sleep mode enabled in Sleep. Draws lowest current in Sleep, slower wake-up.

0 = Low-Power mode enabled in Sleep. Draws higher current in Sleep, faster wake-up.

bit 0 Reserved: Maintain this bit set.

# 14.0 TEMPERATURE INDICATOR MODULE

This family of devices is equipped with a temperature circuit designed to measure the operating temperature of the silicon die. The circuit's range of operating temperature falls between of  $-40^{\circ}$ C and  $+85^{\circ}$ C. The output is a voltage that is proportional to the device temperature. The output of the temperature indicator is internally connected to the device ADC.

The circuit may be used as a temperature threshold detector or a more accurate temperature indicator, depending on the level of calibration performed. A one-point calibration allows the circuit to indicate a temperature closely surrounding that point. A two-point calibration allows the circuit to sense the entire range of temperature more accurately. Reference Application Note AN1333, "Use and Calibration of the Internal Temperature Indicator" (DS01333) for more details regarding the calibration process.

# 14.1 Circuit Operation

Figure 14-1 shows a simplified block diagram of the temperature circuit. The proportional voltage output is achieved by measuring the forward voltage drop across multiple silicon junctions.

Equation 14-1 describes the output characteristics of the temperature indicator.

# EQUATION 14-1: VOUT RANGES

High Range: VOUT = VDD - 4VT

Low Range: VOUT = VDD - 2VT

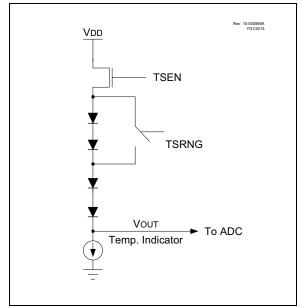
The temperature sense circuit is integrated with the Fixed Voltage Reference (FVR) module. See **Section 12.0 "Fixed Voltage Reference (FVR)"** for more information.

The circuit is enabled by setting the TSEN bit of the FVRCON register. When disabled, the circuit draws no current.

The circuit operates in either high or low range. The high range, selected by setting the TSRNG bit of the FVRCON register, provides a wider output voltage. This provides more resolution over the temperature range, but may be less consistent from part to part. This range requires a higher bias voltage to operate and thus, a higher VDD is needed.

The low range is selected by clearing the TSRNG bit of the FVRCON0 register. The low range generates a lower voltage drop and thus, a lower bias voltage is needed to operate the circuit. The low range is provided for low voltage operation.

### FIGURE 14-1: TEMPERATURE CIRCUIT DIAGRAM



# 14.2 Minimum Operating VDD vs. Minimum Sensing Temperature

When the temperature circuit is operated in low range, the device may be operated at any operating voltage that is within specifications.

When the temperature circuit is operated in high range, the device operating voltage, VDD, must be high enough to ensure that the temperature circuit is correctly biased.

Table 14-1 shows the recommended minimum VDD vs.range setting.

TABLE 14-1: RECOMMENDED VDD VS. RANGE

Min. VDD, TSRNG = 1	Min. VDD, TSRNG = 0
3.6V	1.8V

# 14.3 Temperature Output

The output of the circuit is measured using the internal Analog-to-Digital Converter. A channel is reserved for the temperature circuit output. Refer to Section 15.0 "Analog-to-Digital Converter (ADC) Module" for detailed information.

# 14.4 ADC Acquisition Time

To ensure accurate temperature measurements, the user must wait at least 200  $\mu$ s after the ADC input multiplexer is connected to the temperature indicator output before the conversion is performed. In addition, the user must wait 200  $\mu$ s between sequential conversions of the temperature indicator output.

# PIC10(L)F320/322

	0011111/								OAION
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
FVRCON	FVREN	FVRRDY	TSEN	TSRNG — — ADFVR<				R<1:0>	78
ADCON		ADCS<2:0>		CHS<2:0> GO/ DONE ADON				ADON	88
ADRES		A/D Result Register							89

## TABLE 14-2: SUMMARY OF REGISTERS ASSOCIATED WITH THE TEMPERATURE INDICATOR

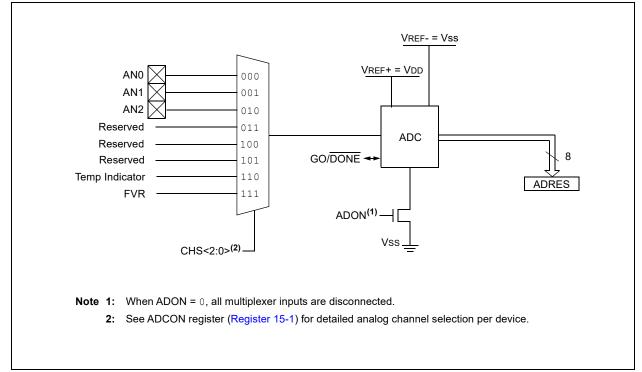
**Legend:** — = unimplemented location, read as '0'. Shaded cells are not used by the temperature indicator module.

# 15.0 ANALOG-TO-DIGITAL CONVERTER (ADC) MODULE

The Analog-to-Digital Converter (ADC) converts an analog input signal to an 8-bit binary representation of that signal. This device uses three analog input channels, which are multiplexed into a single sample and hold circuit. The output of the sample and hold is connected to the input of the converter. The converter generates an 8-bit binary result via successive approximation and stores the conversion result into the ADC Result (ADRES) register. Figure 15-1 shows the block diagram of the ADC.

The ADC voltage reference is software selectable to be internally generated.

The ADC can generate an interrupt upon completion of a conversion. This interrupt can be used to wake-up the device from Sleep.



## FIGURE 15-1: ADC SIMPLIFIED BLOCK DIAGRAM

# 15.1 ADC Configuration

When configuring and using the ADC the following functions must be considered:

- Port configuration
- · Channel selection
- ADC conversion clock source
- · Interrupt control

### 15.1.1 PORT CONFIGURATION

The ADC can be used to convert both analog and digital signals. When converting analog signals, the I/O pin must be configured for analog by setting the associated TRIS and ANSEL bits. Refer to Section 10.0 "I/O Port" for more information.

Note:	Analog voltages on any pin that is defined
	as a digital input may cause the input buf-
	fer to conduct excess current.

### 15.1.2 CHANNEL SELECTION

There are up to five channel selections available:

- AN<2:0> pins
- Temperature Indicator
- FVR (Fixed Voltage Reference) Output

Refer to Section 12.0 "Fixed Voltage Reference (FVR)" and Section 14.0 "Temperature Indicator Module" for more information on these channel selections.

The CHS bits of the ADCON register determine which channel is connected to the sample and hold circuit.

When changing channels, a delay is required before starting the next conversion. Refer to **Section 15.2 "ADC Operation"** for more information.

### 15.1.3 ADC VOLTAGE REFERENCE

There is no external voltage reference connections to the ADC. Only VDD can be used as a reference source. The FVR is only available as an input channel and not a VREF+ input to the ADC.

### 15.1.4 CONVERSION CLOCK

The source of the conversion clock is software selectable via the ADCS bits of the ADCON register (Register 15-1). There are seven possible clock options:

- · Fosc/2
- Fosc/4
- Fosc/8
- Fosc/16
- Fosc/32
- Fosc/64
- FRC (dedicated internal RC oscillator)

The time to complete one bit conversion is defined as TAD. One full 8-bit conversion requires 9.5 TAD periods as shown in Figure 15-2.

For correct conversion, the appropriate TAD specification must be met. Refer to the A/D conversion requirements in **Section 24.0 "Electrical Specifications"** for more information. Table 15-1 gives examples of appropriate ADC clock selections.

**Note:** Unless using the FRC, any changes in the system clock frequency will change the ADC clock frequency, which may adversely affect the ADC result.

### TABLE 15-1: ADC CLOCK PERIOD (TAD) Vs. DEVICE OPERATING FREQUENCIES

ADC Clock P	eriod (TAD)	Device Frequency (Fosc)					
ADC Clock Source	ADCS<2:0>	16 MHz	8 MHz	4 MHz	1 MHz		
Fosc/2	000	125 ns <sup>(1)</sup>	250 ns <sup>(1)</sup>	500 ns <sup>(1)</sup>	2.0 μs		
Fosc/4	100	250 ns <sup>(1)</sup>	500 ns <sup>(1)</sup>	1.0 μs	4.0 μs		
Fosc/8	001	0.5 μs <sup>(1)</sup>	1.0 μs	2.0 μs	8.0 μs <sup>(2)</sup>		
Fosc/16	101	1.0 μs	2.0 μs	4.0 μs	16.0 μs <sup>(2)</sup>		
Fosc/32	010	2.0 μs	4.0 μs	8.0 μs <sup>(2)</sup>	32.0 μs <sup>(2)</sup>		
Fosc/64	110	4.0 μs	8.0 μs <sup>(2)</sup>	16.0 μs <sup>(2)</sup>	64.0 μs <sup>(2)</sup>		
FRC	x11	1.0-6.0 μs <sup>(1,3)</sup>	1.0-6.0 μs <sup>(1,3)</sup>	1.0-6.0 μs <sup>(1,3)</sup>	1.0-6.0 μs <sup>(1,3)</sup>		

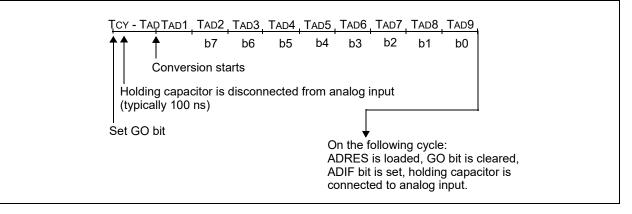
Legend: Shaded cells are outside of recommended range.

**Note 1:** These values violate the minimum required TAD time.

2: For faster conversion times, the selection of another clock source is recommended.

3: The ADC clock period (TAD) and total ADC conversion time can be minimized when the ADC clock is derived from the system clock FOSC. However, the FRC clock source must be used when conversions are to be performed with the device in Sleep mode.





### 15.1.5 INTERRUPTS

The ADC module allows for the ability to generate an interrupt upon completion of an Analog-to-Digital conversion. The ADC Interrupt Flag is the ADIF bit in the PIR1 register. The ADC Interrupt Enable is the ADIE bit in the PIE1 register. The ADIF bit must be cleared in software.

Note:	The ADIF bit is set at the completion of
	every conversion, regardless of whether
	or not the ADC interrupt is enabled.

This interrupt can be generated while the device is operating or while in Sleep. If the device is in Sleep, the interrupt will wake-up the device. Upon waking from Sleep, the next instruction following the SLEEP instruction is always executed. If the user is attempting to wake-up from Sleep and resume in-line code execution, the GIE and PEIE bits of the INTCON register must be disabled. If the GIE and PEIE bits of the INTCON register are enabled, execution will switch to the Interrupt Service Routine.

### 15.2 ADC Operation

### 15.2.1 STARTING A CONVERSION

To enable the ADC module, the ADON bit of the ADCON register must be set to a '1'. Setting the GO/ DONE bit of the ADCON register to a '1' will start the Analog-to-Digital conversion.

Note:	The GO/DONE bit must not be set in the
	same instruction that turns on the ADC.
	Refer to Section 15.2.5 "A/D Conver-
	sion Procedure".

### 15.2.2 COMPLETION OF A CONVERSION

When the conversion is complete, the ADC module will:

- Clear the GO/DONE bit
- · Set the ADIF Interrupt Flag bit
- Update the ADRES register with new conversion result

### 15.2.3 TERMINATING A CONVERSION

If a conversion must be terminated before completion, the GO/DONE bit can be cleared in software. The ADRES register will be updated with the partially complete Analog-to-Digital conversion sample. Incomplete bits will match the last bit converted.

Note:	A device Reset forces all registers to their
	Reset state. Thus, the ADC module is
	turned off and any pending conversion is
	terminated.

### 15.2.4 ADC OPERATION DURING SLEEP

The ADC module can operate during Sleep. This requires the ADC clock source to be set to the FRC option. When the FRC clock source is selected, the ADC waits one additional instruction before starting the conversion. This allows the SLEEP instruction to be executed, which can reduce system noise during the conversion. If the ADC interrupt is enabled, the device will wake-up from Sleep when the conversion completes. If the ADC interrupt is disabled, the ADC module is turned off after the conversion completes, although the ADON bit remains set.

When the ADC clock source is something other than FRC, a SLEEP instruction causes the present conversion to be aborted and the ADC module is turned off, although the ADON bit remains set.

### 15.2.5 A/D CONVERSION PROCEDURE

This is an example procedure for using the ADC to perform an Analog-to-Digital conversion:

- 1. Configure Port:
  - Disable pin output driver (Refer to the TRIS register)
  - Configure pin as analog (Refer to the ANSEL register)
  - Disable weak pull-ups either globally (Refer to the OPTION\_REG register) or individually (Refer to the appropriate WPUX register)
- 2. Configure the ADC module:
  - Select ADC conversion clock
  - Select ADC input channel
  - Turn on ADC module
- 3. Configure ADC interrupt (optional):
  - Clear ADC interrupt flag
  - Enable ADC interrupt
  - Enable peripheral interrupt
  - Enable global interrupt<sup>(1)</sup>
- 4. Wait the required acquisition time<sup>(2)</sup>.
- 5. Start conversion by setting the GO/DONE bit.
- 6. Wait for ADC conversion to complete by one of the following:
  - Polling the GO/DONE bit
  - Waiting for the ADC interrupt (interrupts enabled)
- 7. Read ADC Result.
- 8. Clear the ADC interrupt flag (required if interrupt is enabled).

**Note 1:** The global interrupt can be disabled if the user is attempting to wake-up from Sleep and resume in-line code execution.

2: Refer to Section 15.4 "A/D Acquisition Requirements".

# **15.3 ADC Register Definitions**

The following registers are used to control the operation of the ADC.

### REGISTER 15-1: ADCON: A/D CONTROL REGISTER 0

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
	ADCS<2:0>			CHS<2:0>		GO/DONE	ADON
bit 7							bit 0
Legend:							
R = Readable		W = Writable		•	nented bit, rea		
u = Bit is unch	•	x = Bit is unkn		-n/n = Value a	at POR and BC	OR/Value at all o	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				
bit 7-5	ADCS<2:0>: 111 = FRC 110 = Fosc/4 101 = Fosc/4 011 = FRC 010 = Fosc/2 011 = FRC 001 = Fosc/2 001 = Fosc/2	16 4 32 3	n Clock Selec	t bits			
bit 4-2	111 = FVR (I 110 = Tempe 101 = Reser 100 = Reser	nalog Channel Fixed Voltage R erature Indicatol ved. No channe ved. No channe ved. No channe	eference) Bu (1) I connected. I connected.	ffer Output <sup>(2)</sup>			
bit 1	<b>GO/DONE:</b> A <u>If ADON = 1:</u> 1 = A/D con 0 = A/D cor convers	/D Conversion oversion in prog oversion not in sion is complete leared while a c	ress (Setting <sup>,</sup> progress (Th .)	is bit is automa	atically cleared	by hardware	
	conversion up not be set. <u>If ADON = </u> 0:	p to this point w	ill be transferi				
bit 0	<b>ADON:</b> ADC 1 = ADC is er		-	erating current			
		"Temperature "Fixed Voltage					

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			ADRE	S<7:0>			
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable I	bit	U = Unimpler	nented bit, read	d as '0'	
u = Bit is unch	anged	x = Bit is unkn	own	-n/n = Value a	at POR and BC	R/Value at all	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				

### REGISTER 15-2: ADRES: ADC RESULT REGISTER

bit 7-0 ADRES<7:0>: ADC Result Register bits 8-bit result PIC10(L)F320/322

### **15.4** A/D Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in Figure 15-3. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD), refer to Figure 15-3. The maximum recommended impedance for analog sources is 10 k $\Omega$ . As the

source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an A/D acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 15-1 may be used. This equation assumes that 1/2 LSb error is used (511 steps for the ADC). The 1/2 LSb error is the maximum error allowed for the ADC to meet its specified resolution.

### EQUATION 15-1: ACQUISITION TIME EXAMPLE

Assumptions: Temperature = 50°C and external impedance of 10k
$$\Omega$$
 5.0V VDD  
TACQ = Amplifier Settling Time + Hold Capacitor Charging Time + Temperature Coefficient  
= TAM P + TC + TCOFF  
= 2µs + TC + [(Temperature - 25°C)(0.05µs/°C)]

The value for TC can be approximated with the following equations:

$$V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) = V_{CHOLD} ; [1] VCHOLD charged to within 1/2 lsb$$

$$V_{APPLIED}\left(1 - e^{\frac{-Tc}{RC}}\right) = V_{CHOLD} ; [2] VCHOLD charge response to VAPPLIED$$

$$V_{APPLIED}\left(1 - e^{\frac{-Tc}{RC}}\right) = V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) ; combining [1] and [2]$$

*Note:* Where n = number of bits of the ADC.

Solving for TC:

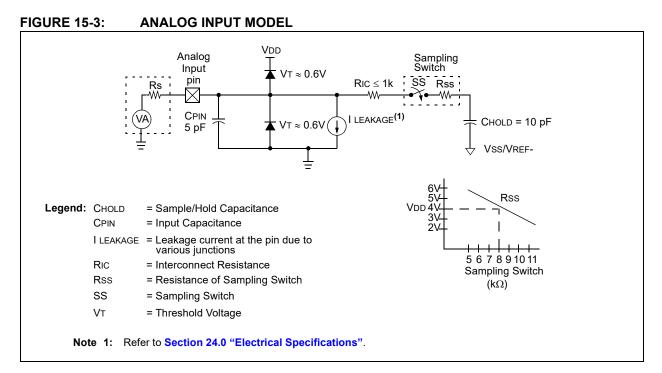
$$Tc = -C_{HOLD}(RIC + RSS + RS) \ln(1/511)$$
  
= -10pF(1k\Omega + 7k\Omega + 10k\Omega) \ln(0.001957)  
= 1.12\u03cc

Therefore:

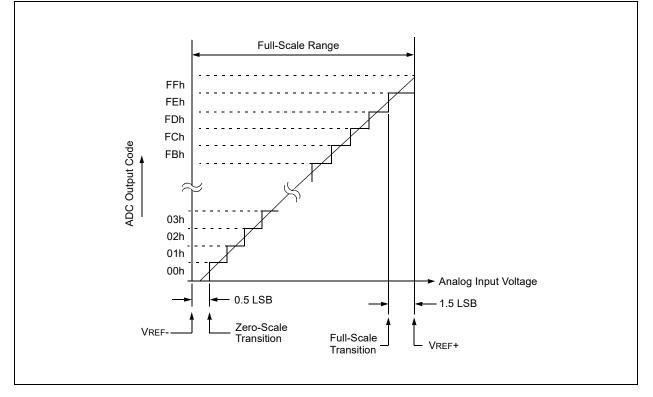
**Note 1:** The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

- 2: The charge holding capacitor (CHOLD) is not discharged after each conversion.
- **3:** The maximum recommended impedance for analog sources is  $10 \text{ k}\Omega$ . This is required to meet the pin leakage specification.

# PIC10(L)F320/322







### TABLE 15-2: SUMMARY OF REGISTERS ASSOCIATED WITH ADC

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ADCON		ADCS<2:0>			CHS<2:0>		GO/DONE	ADON	88
ADRES	ADRES<7:0>				89				
ANSELA	—	_	_	_	_	ANSA2	ANSA1	ANSA0	70
FVRCON	FVREN	FVRRDY	TSEN	TSRNG	_	_	ADFVF	۲<1:0>	78
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	40
PIE1	—	ADIE	_	NCO1IE	CLC1IE	_	TMR2IE	_	41
PIR1	—	ADIF	—	NCO1IF	CLC1IF	—	TMR2IF	_	42
TRISA	_		_	_	_	TRISA2	TRISA1	TRISA0	69

Legend: x = unknown, u = unchanged, — = unimplemented read as '0', q = value depends on condition. Shaded cells are not used for ADC module.

#### TIMER0 MODULE 16.0

The Timer0 module is an 8-bit timer/counter with the following features:

- 8-bit timer/counter register (TMR0)
- 8-bit prescaler (independent of Watchdog Timer)
- · Programmable internal or external clock source
- · Programmable external clock edge selection
- · Interrupt on overflow

**FIGURE 16-1:** 

Figure 16-1 is a block diagram of the Timer0 module.

#### 16.1 **Timer0** Operation

The Timer0 module can be used as either an 8-bit timer or an 8-bit counter.

#### 8-BIT TIMER MODE 16.1.1

The Timer0 module will increment every instruction cycle, if used without a prescaler. 8-Bit Timer mode is selected by clearing the T0CS bit of the OPTION REG register.

When TMR0 is written, the increment is inhibited for two instruction cycles immediately following the write.

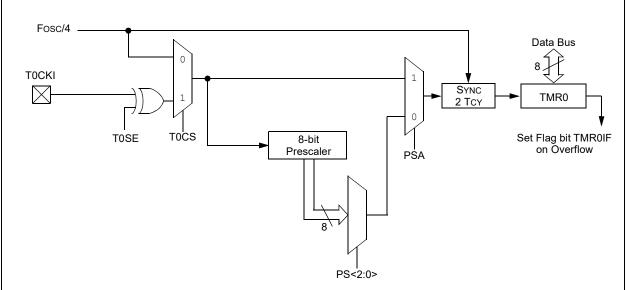
The value written to the TMR0 register Note: can be adjusted, in order to account for the two instruction cycle delay when TMR0 is written.

#### 8-BIT COUNTER MODE 16.1.2

In 8-Bit Counter mode, the Timer0 module will increment on every rising or falling edge of the T0CKI pin.

8-Bit Counter mode using the TOCKI pin is selected by setting the TOCS bit in the OPTION REG register to '1'.

The rising or falling transition of the incrementing edge for the external input source is determined by the T0SE bit in the OPTION\_REG register.



**BLOCK DIAGRAM OF THE TIMER0 PRESCALER** 

### 16.1.3 SOFTWARE PROGRAMMABLE PRESCALER

A single software programmable prescaler is available for use with Timer0. The prescaler assignment is controlled by the PSA bit of the OPTION\_REG register. To assign the prescaler to Timer0, the PSA bit must be cleared to a '0'.

There are eight prescaler options for the Timer0 module ranging from 1:2 to 1:256. The prescale values are selectable via the PS<2:0> bits of the OPTION REG register.

The prescaler is not readable or writable. When assigned to the Timer0 module, all instructions writing to the TMR0 register will clear the prescaler.

### 16.1.4 TIMER0 INTERRUPT

Timer0 will generate an interrupt when the TMR0 register overflows from FFh to 00h. The TMR0IF interrupt flag bit of the INTCON register is set every time the TMR0 register overflows, regardless of whether or not the Timer0 interrupt is enabled. The TMR0IF bit can only be cleared in software. The Timer0 interrupt enable is the TMR0IE bit of the INTCON register.

Note:	The Timer0 interrupt cannot wake the
	processor from Sleep since the timer is
	frozen during Sleep.

### 16.1.5 8-BIT COUNTER MODE SYNCHRONIZATION

When in 8-Bit Counter mode, the incrementing edge on the T0CKI pin must be synchronized to the instruction clock. Synchronization can be accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the instruction clock. The high and low periods of the external clocking source must meet the timing requirements as shown in Section 24.0 "Electrical Specifications".

		_									
R/W-1/u	R/W-1/u	R/W-1/u	R/W-1/u	R/W-1/u	R/W-1/u	R/W-1/u	R/W-1/u				
WPUEN <sup>(1)</sup>	INTEDG	TOCS	TOSE	PSA		PS<2:0>					
bit 7							bit (				
Legend:											
R = Readable	bit	W = Writabl	e bit	U = Unimplei	mented bit, read	d as '0'					
u = Bit is unch	anged	x = Bit is un	known	-n/n = Value	at POR and BC	R/Value at all c	other Resets				
'1' = Bit is set		'0' = Bit is c	eared								
bit 7		eak Pull-up Er									
		III-ups are disa			veluee						
L.H.C.	•	•	bled by individu		values						
bit 6		NTEDG: Interrupt Edge Select bit . = Interrupt on rising edge of INT pin									
	1 = Interrupt on falling edge of INT pin 0 = Interrupt on falling edge of INT pin										
bit 5	TOCS: TMR0 Clock Source Select bit										
	1 = Transitio	n on T0CKI pi	n								
	0 = Internal i	nstruction cyc	le clock (Fosc/	4)							
bit 4	T0SE: TMR0 Source Edge Select bit										
	<ul> <li>1 = Increment on high-to-low transition on T0CKI pin</li> <li>0 = Increment on low-to-high transition on T0CKI pin</li> </ul>										
			•	10CKI pin							
bit 3		<b>PSA:</b> Prescaler Assignment bit 1 = Prescaler is inactive and has no effect on the Timer 0 module									
			to the Timer0 m		u module						
bit 2-0		escaler Rate									
	Bit	Value TMR	) Rate								
		000 1:	2								
		001 1:									
		010 1:	8								
			16								
			32								
			64 128								
			120								

### REGISTER 16-1: OPTION\_REG: OPTION REGISTER

**Note 1:**  $\overline{\text{WPUEN}}$  does not disable the pull-up for the  $\overline{\text{MCLR}}$  input when  $\overline{\text{MCLR}}$  = 1.

1 : 256

111

### TABLE 16-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER0

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	40
OPTION_REG	WPUEN	INTEDG	TOCS	T0SE	PSA		PS<2:0>		95
TMR0			٦	imer0 mo	dule Regist	er		40	
TRISA	—	_	—	—	—	TRISA2	TRISA1	TRISA0	69

**Legend:** – = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the Timer0 module.

# 17.0 TIMER2 MODULE

The Timer2 module is an 8-bit timer with the following features:

- 8-bit timer register (TMR2)
- 8-bit period register (PR2)
- · Interrupt on TMR2 match with PR2
- Software programmable prescaler (1:1, 1:4, 1:16, 1:64)
- Software programmable postscaler (1:1 to 1:16)

See Figure 17-1 for a block diagram of Timer2.

# 17.1 Timer2 Operation

The clock input to the Timer2 module is the system instruction clock (Fosc/4). The clock is fed into the Timer2 prescaler, which has prescale options of 1:1, 1:4 or 1:64. The output of the prescaler is then used to increment the TMR2 register.

The values of TMR2 and PR2 are constantly compared to determine when they match. TMR2 will increment from 00h until it matches the value in PR2. When a match occurs, two things happen:

- TMR2 is reset to 00h on the next increment cycle.
- The Timer2 postscaler is incremented.

The match output of the Timer2/PR2 comparator is then fed into the Timer2 postscaler. The postscaler has postscale options of 1:1 to 1:16 inclusive. The output of the Timer2 postscaler is used to set the TMR2IF interrupt flag bit in the PIR1 register.

### FIGURE 17-1: TIMER2 BLOCK DIAGRAM

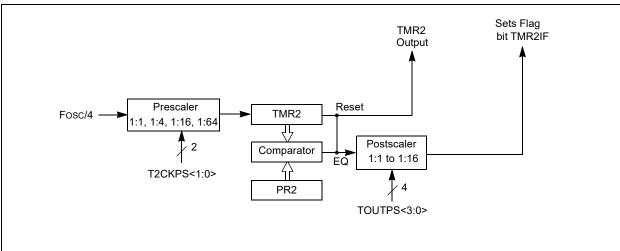
The TMR2 and PR2 registers are both fully readable and writable. On any Reset, the TMR2 register is set to 00h and the PR2 register is set to FFh.

Timer2 is turned on by setting the TMR2ON bit in the T2CON register to a '1'. Timer2 is turned off by clearing the TMR2ON bit to a '0'.

The Timer2 prescaler is controlled by the T2CKPS bits in the T2CON register. The Timer2 postscaler is controlled by the TOUTPS bits in the T2CON register. The prescaler and postscaler counters are cleared when:

- A write to TMR2 occurs.
- · A write to T2CON occurs.
- Any device Reset occurs (Power-on Reset, MCLR Reset, Watchdog Timer Reset, or Brown-out Reset).

Note: TMR2 is not cleared when T2CON is written.



U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0				
_		TOUTP	S<3:0>		TMR2ON	T2CKF	PS<1:0>				
bit 7							bit 0				
Legend: R = Readabl	a hit	M = Mritable	- it	II – Unimploy	monted hit read	aa '0'					
		W = Writable		•	mented bit, read						
u = Bit is und	0	x = Bit is unkn		-n/n = value	at POR and BO	R/Value at all	other Resets				
'1' = Bit is se	t	'0' = Bit is clea	ared								
			- 1								
bit 7	-	nted: Read as '									
bit 6-3		:0>: Timer2 Outp	out Postscaler	Select bits							
	1111 = 1:16										
		1110 = 1:15 Postscaler									
		= 1:14 Postscaler									
		1100 = 1:13 Postscaler 1011 = 1:12 Postscaler									
		011 = 1.12 Postscaler									
		0 Postscaler									
	1000 = 1:9										
	0111 = 1:8										
	0110 = <b>1</b> : <b>7</b>										
	0101 = <b>1:6</b>	Postscaler									
	0100 = <b>1</b> :5										
	0011 = <b>1</b> :4	Postscaler									
	0010 = <b>1</b> :3										
	0001 = <b>1:2</b>										
	0000 = 1:1										
bit 2	_	ïmer2 On bit									
	1 = Timer2 0 = Timer2										
bit 1-0	T2CKPS<1:	<b>0&gt;:</b> Timer2 Cloc	k Prescale Se	elect bits							
	11 = Presc	aler is 64									
	10 = Presc	aler is 16									
	01 = Presc	aler is 4									
	00 = Presc	aler is 1									
<b>TABLE 17-</b> 1	I: SUMMAF	RY OF REGIST	ERS ASSO	CIATED WIT	H TIMER2						

### REGISTER 17-1: T2CON: TIMER2 CONTROL REGISTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	40
PIE1	_	ADIE	_	NCO1IE	CLC1IE	_	TMR2IE	—	41
PIR1	_	ADIF	_	NCO1IF	CLC1IF	_	TMR2IF	—	42
PR2				Timer2 mo	odule Period Reg	ister			96
TMR2	Timer2 module Register							96	
T2CON			TOUTPS<3:0> TMR2ON T2CKPS<1:0>						97

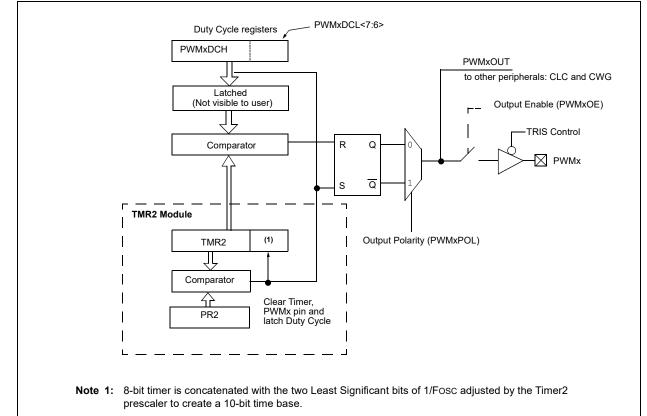
Legend: x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are not used for Timer2 module.

# 18.0 PULSE-WIDTH MODULATION (PWM) MODULE

The PWM module generates a Pulse-Width Modulated signal determined by the duty cycle, period, and resolution that are configured by the following registers:

- PR2
- T2CON
- PWMxDCH
- PWMxDCL
- PWMxCON

### FIGURE 18-1: SIMPLIFIED PWM BLOCK DIAGRAM



For a step-by-step procedure on how to set up this module for PWM operation, refer to Section 18.1.9 "Setup for PWM Operation using PWMx Pins".

### FIGURE 18-2: PWM OUTPUT

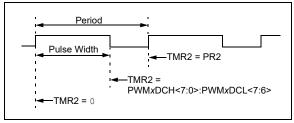


Figure 18-1 shows a simplified block diagram of PWM operation.

Figure 18-2 shows a typical waveform of the PWM signal.

## **18.1 PWMx Pin Configuration**

All PWM outputs are multiplexed with the PORT data latch. The user must configure the pins as outputs by clearing the associated TRIS bits.

**Note:** Clearing the PWMxOE bit will relinquish control of the PWMx pin.

### 18.1.1 FUNDAMENTAL OPERATION

The PWM module produces a 10-bit resolution output. Timer2 and PR2 set the period of the PWM. The PWMxDCL and PWMxDCH registers configure the duty cycle. The period is common to all PWM modules, whereas the duty cycle is independently controlled.

**Note:** The Timer2 postscaler is not used in the determination of the PWM frequency. The postscaler could be used to have a servo update rate at a different frequency than the PWM output.

All PWM outputs associated with Timer2 are set when TMR2 is cleared. Each PWMx is cleared when TMR2 is equal to the value specified in the corresponding PWMxDCH (8 MSb) and PWMxDCL<7:6> (2 LSb) registers. When the value is greater than or equal to PR2, the PWM output is never cleared (100% duty cycle).

**Note:** The PWMxDCH and PWMxDCL registers are double buffered. The buffers are updated when Timer2 matches PR2. Care must be taken to update both registers before the timer match occurs.

### 18.1.2 PWM OUTPUT POLARITY

The output polarity is inverted by setting the PWMxPOL bit of the PWMxCON register.

### 18.1.3 PWM PERIOD

The PWM period is specified by the PR2 register of Timer2. The PWM period can be calculated using the formula of Equation 18-1.

### EQUATION 18-1: PWM PERIOD

 $PWM Period = [(PR2) + 1] \bullet 4 \bullet Tosc \bullet$ 

(TM R2 Prescale Value)

Note: Tosc = 1/Fosc

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- · TMR2 is cleared
- The PWM output is active. (Exception: When the PWM duty cycle = 0%, the PWM output will remain inactive.)
- The PWMxDCH and PWMxDCL register values are latched into the buffers.

Note:	The Timer2 postscaler has no effect on the
	PWM operation.

### 18.1.4 PWM DUTY CYCLE

The PWM duty cycle is specified by writing a 10-bit value to the PWMxDCH and PWMxDCL register pair. The PWMxDCH register contains the eight MSbs and the PWMxDCL<7:6>, the two LSbs. The PWMxDCH and PWMxDCL registers can be written to at any time.

Equation 18-2 is used to calculate the PWM pulse width.

Equation 18-3 is used to calculate the PWM duty cycle ratio.

### EQUATION 18-2: PULSE WIDTH

Pulse W idth = (PW M xD CH : PW M xD CL<7:6>) •

Tosc • (TMR2 Prescale Value)

Note: Tosc = 1/Fosc

# EQUATION 18-3: DUTY CYCLE RATIO

 $\texttt{Duty Cycle Ratio} = \frac{(\texttt{PW M xD CH : \texttt{PW M xD CL<7:6>})}{4(\texttt{PR2+1})}$ 

The 8-bit timer TMR2 register is concatenated with the two Least Significant bits of 1/Fosc, adjusted by the Timer2 prescaler to create the 10-bit time base. The system clock is used if the Timer2 prescaler is set to 1:1.

### 18.1.5 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is ten bits when PR2 is 255. The resolution is a function of the PR2 register value as shown by Equation 18-4.

### EQUATION 18-4: PWM RESOLUTION

Resolution =  $\frac{\log[4(PR2 + 1)]}{\log(2)}$  bits

Note: If the pulse-width value is greater than the period the assigned PWM pin(s) will remain unchanged.

TABLE 18-1:	EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 20 MHz)
-------------	---

PWM Frequency	0.31 kHz	4.88 kHz	19.53 kHz	78.12 kHz	156.3 kHz	208.3 kHz
Timer Prescale (1, 4, 64)	64	4	1	1	1	1
PR2 Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x17
Maximum Resolution (bits)	10	10	10	8	7	6.6

TABLE 18-2:	<b>EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (</b>	Fosc = 8 MHz)	1
-------------	--	---------------	---

PWM Frequency	0.31 kHz	4.90 kHz	19.61 kHz	76.92 kHz	153.85 kHz	200.0 kHz
Timer Prescale (1, 4, 64)	64	4	1	1	1	1
PR2 Value	0x65	0x65	0x65	0x19	0x0C	0x09
Maximum Resolution (bits)	8	8	8	6	5	5

### 18.1.6 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the PWMx pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

### 18.1.7 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency (Fosc). Any changes in the system clock frequency will result in changes to the PWM frequency. Refer to **Section 4.0** "**Oscillator Module**" for additional details.

### 18.1.8 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the PWM registers to their Reset states.

### 18.1.9 SETUP FOR PWM OPERATION USING PWMx PINS

The following steps must be taken when configuring the module for PWM operation using the PWMx pins:

- 1. Disable the PWMx pin output driver(s) by setting the associated TRIS bit(s).
- 2. Clear the PWMxCON register.
- 3. Load the PR2 register with the PWM period value.
- 4. Clear the PWMxDCH register and bits <7:6> of the PWMxDCL register.
- 5. Configure and start Timer2:
  - Clear the TMR2IF interrupt flag bit of the PIR1 register. See Note below.
  - Configure the T2CKPS bits of the T2CON register with the Timer2 prescale value.
  - Enable Timer2 by setting the TMR2ON bit of the T2CON register.
- Enable PWM output pin and wait until Timer2 overflows, TMR2IF bit of the PIR1 register is set. See Note below.
- Enable the PWMx pin output driver(s) by clearing the associated TRIS bit(s) and setting the PWMxOE bit of the PWMxCON register.
- 8. Configure the PWM module by loading the PWMxCON register with the appropriate values.
  - Note 1: To send a complete duty cycle and period on the first PWM output, the above steps must be followed in the order given. If it is not critical to start with a complete PWM signal, then move Step 8 to replace Step 4.
    - **2:** For operation with other peripherals only, disable PWMx pin outputs.

# 18.2 **PWM Register Definitions**

## REGISTER 18-1: PWMxCON: PWM CONTROL REGISTER

R/W-0/0	R/W-0/0	R-0/0	R/W-0/0	U-0	U-0	U-0	U-0
PWMxEN	PWMxOE	PWMxOUT	PWMxPOL	—	—	—	—
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	l as '0'	
u = Bit is unch	anged	x = Bit is unkr	nown	-n/n = Value a	at POR and BO	R/Value at all o	other Resets
'1' = Bit is set		'0' = Bit is clea	ared				
bit 7	PWMxEN: PV	NM Module En	able bit				
		dule is enable	-				
	0 = PWM mc	odule is disable	d				
bit 6	PWMxOE: P	WM Module Ou	itput Enable bi	t			
		PWMx pin is e					
	0 = Output to	PWMx pin is o	disabled				
bit 5	PWMxOUT: F	PWM Module C	output Value bi	t			
bit 4	PWMxPOL: F	PWMx Output F	Polarity Select	bit			
	1 = PWM out	tput is active-lo	W.				
	0 = PWM out	tput is active-hi	gh.				
bit 3-0	Unimplemen	ted: Read as '	0'				

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
			PWMx[	DCH<7:0>			
bit 7							bit 0
Legend:							
R = Readable bit W = Writable bit		U = Unimplen	nented bit, read	l as '0'			
u = Bit is uncha	anged	x = Bit is unkno	own	-n/n = Value a	at POR and BO	R/Value at all o	other Resets
'1' = Bit is set		'0' = Bit is clear	red				

bit 7-0 **PWMxDCH<7:0>:** PWM Duty Cycle Most Significant bits These bits are the MSbs of the PWM duty cycle. The two LSbs are found in the PWMxDCL Register.

### REGISTER 18-3: PWMxDCL: PWM DUTY CYCLE LOW BITS

R/W-x/u	R/W-x/u	U-0	U-0	U-0	U-0	U-0	U-0
PWMxDCL<7:6>		—	—	—	—	—	—
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-6 **PWMxDCL<7:6>:** PWM Duty Cycle Least Significant bits These bits are the LSbs of the PWM duty cycle. The MSbs are found in the PWMxDCH Register.

bit 5-0 Unimplemented: Read as '0'

### TABLE 18-3: SUMMARY OF REGISTERS ASSOCIATED WITH PWM

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	_	—	_	-	ANSA2	ANSA1	ANSA0	70
LATA	_	_	_	_	_	LATA2	LATA1	LATA0	70
PORTA	—		_		RA3	RA2	RA1	RA0	69
PR2	Timer2 module Period Register							96	
PWM1CON	PWM1EN	PWM10E	PWM10UT	PWM1POL	_	_	_	—	102
PWM1DCH	PWM1DCH<7:0>							103	
PWM1DCL	PWM1DCL<7:6>		—	_	_	_	_	—	103
PWM2CON	PWM2EN	PWM2OE	PWM2OUT	PWM2POL	_	_	_	—	102
PWM2DCH	PWM2DCH<7:0>							103	
PWM2DCL	PWM2DCL<7:6>		—	_	_	_	_	—	103
T2CON	— TOUTPS<3:0> TMR2ON T2CKPS<1:0>						97		
TMR2	Timer2 module Register						96		
TRISA	_	_	_	_	_	TRISA2	TRISA1	TRISA0	69

Legend: - = Unimplemented locations, read as '0', u = unchanged, x = unknown. Shaded cells are not used by the PWM.

# 19.0 CONFIGURABLE LOGIC CELL (CLC)

The Configurable Logic Cell (CLCx) provides programmable logic that operates outside the speed limitations of software execution. The logic cell selects any combination of the eight input signals and through the use of configurable gates reduces the selected inputs to four logic lines that drive one of eight selectable single-output logic functions.

Input sources are a combination of the following:

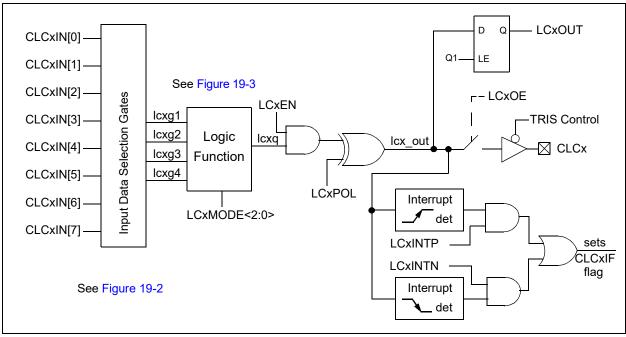
- Two I/O pins
- Internal clocks
- Peripherals
- Register bits

The output can be directed internally to peripherals and to an output pin.

Refer to Figure 19-1 for a simplified diagram showing signal flow through the CLCx.

Possible configurations include:

- Combinatorial Logic
- AND
- NAND
- AND-OR
- AND-OR-INVERT
- OR-XOR
- OR-XNOR
- Latches
  - S-R
  - Clocked D with Set and Reset
  - Transparent D with Set and Reset
  - Clocked J-K with Reset



## FIGURE 19-1: CLCx SIMPLIFIED BLOCK DIAGRAM

# 19.1 CLCx Setup

Programming the CLCx module is performed by configuring the four stages in the logic signal flow. The four stages are:

- · Data selection
- · Data gating
- Logic function selection
- Output polarity

Each stage is setup at run time by writing to the corresponding CLCx Special Function Registers. This has the added advantage of permitting logic reconfiguration on-the-fly during program execution.

### 19.1.1 DATA SELECTION

There are eight signals available as inputs to the configurable logic. Four 8-input multiplexers are used to select the inputs to pass on to the next stage.

Data inputs are selected with the CLCxSEL0 and CLCxSEL1 registers (Register 19-3 and Register 19-4, respectively).

Data selection is through four multiplexers as indicated on the left side of Figure 19-2. Data inputs in the figure are identified by a generic numbered input name.

Table 19-1 correlates the generic input name to the actual signal for each CLC module. The columns labeled lcxd1 through lcxd4 indicate the MUX output for the selected data input. D1S through D4S are abbreviations for the MUX select input codes: LCxD1S<2:0> through LCxD4S<2:0>, respectively. Selecting a data input in a column excludes all other inputs in that column.

**Note:** Data selections are undefined at power-up.

# TABLE 19-1: CLCx DATA INPUT SELECTION

Data Input	lcxd1 D1S	lcxd2 D2S	lcxd3 D3S	lcxd4 D4S	CLC 1
CLCxIN[0]	000	000	000	000	CLCx
CLCxIN[1]	001	001	001	001	CLCxIN1
CLCxIN[2]	010	010	010	010	CLCxIN2
CLCxIN[3]	011	011	011	011	PWM1
CLCxIN[4]	100	100	100	100	PWM2
CLCxIN[5]	101	101	101	101	NCOx
CLCxIN[6]	110	110	110	110	Fosc
CLCxIN[7]	111	111	111	111	LFINTOSC

### 19.1.2 DATA GATING

Outputs from the input multiplexers are directed to the desired logic function input through the data gating stage. Each data gate can direct any combination of the four selected inputs.

The gate stage is more than just signal direction. The gate can be configured to direct each input signal as inverted or noninverted data. Directed signals are ANDed together in each gate. The output of each gate can be inverted before going on to the logic function stage.

The gating is in essence a 1-to-4 input AND/NAND/OR/ NOR gate. When every input is inverted and the output is inverted, the gate is an OR of all enabled data inputs. When the inputs and output are not inverted, the gate is an AND or all enabled inputs.

Table 19-2 summarizes the basic logic that can be obtained in gate 1 by using the gate logic select bits. The table shows the logic of four input variables, but each gate can be configured to use less than four. If no inputs are selected, the output will be zero or one, depending on the gate output polarity bit.

CLCxGLS0	LCxGyPOL	Gate Logic
0x55	1	AND
0x55	0	NAND
0xAA	1	NOR
0xAA	0	OR
0x00	0	Logic 0
0x00	1	Logic 1

### TABLE 19-2: DATA GATING LOGIC

It is possible (but not recommended) to select both the true and negated values of an input. When this is done, the gate output is zero, regardless of the other inputs, but may emit logic glitches (transient-induced pulses). If the output of the channel must be zero or one, the recommended method is to set all gate bits to zero and use the gate polarity bit to set the desired level.

Data gating is configured with the logic gate select registers as follows:

- Gate 1: CLCxGLS0 (Register 19-5)
- Gate 2: CLCxGLS1 (Register 19-6)
- Gate 3: CLCxGLS2 (Register 19-7)
- Gate 4: CLCxGLS3 (Register 19-8)

Register number suffixes are different than the gate numbers because other variations of this module have multiple gate selections in the same register.

Data gating is indicated in the right side of Figure 19-2. Only one gate is shown in detail. The remaining three gates are configured identically with the exception that the data enables correspond to the enables for that gate.

# 19.1.3 LOGIC FUNCTION

There are eight available logic functions including:

- AND-OR
- OR-XOR
- AND
- S-R Latch
- D Flip-Flop with Set and Reset
- D Flip-Flop with Reset
- J-K Flip-Flop with Reset
- Transparent Latch with Set and Reset

Logic functions are shown in Figure 19-3. Each logic function has four inputs and one output. The four inputs are the four data gate outputs of the previous stage. The output is fed to the inversion stage and from there to other peripherals, an output pin, and back to the CLCx itself.

### 19.1.4 OUTPUT POLARITY

The last stage in the configurable logic cell is the output polarity. Setting the LCxPOL bit of the CLCxCON register inverts the output signal from the logic stage. Changing the polarity while the interrupts are enabled will cause an interrupt for the resulting output transition.

### 19.1.5 CLCX SETUP STEPS

The following steps must be followed when setting up the CLCx:

- Disable CLCx by clearing the LCxEN bit.
- Select desired inputs using CLCxSEL0 and CLCxSEL1 registers (See Table 19-1).
- · Clear any associated ANSEL bits.
- Set all TRIS bits associated with inputs.
- · Clear all TRIS bits associated with outputs.
- Enable the chosen inputs through the four gates using CLCxGLS0, CLCxGLS1, CLCxGLS2, and CLCxGLS3 registers.
- Select the gate output polarities with the LCxPOLy bits of the CLCxPOL register.
- Select the desired logic function with the LCxMODE<2:0> bits of the CLCxCON register.
- Select the desired polarity of the logic output with the LCxPOL bit of the CLCxPOL register. (This step may be combined with the previous gate output polarity step).
- If driving the CLCx pin, set the LCxOE bit of the CLCxCON register and also clear the TRIS bit corresponding to that output.
- If interrupts are desired, configure the following bits:
  - Set the LCxINTP bit in the CLCxCON register for rising event.
  - Set the LCxINTN bit in the CLCxCON register or falling event.
  - Set the CLCxIE bit of the associated PIE registers.
  - Set the GIE and PEIE bits of the INTCON register.
- Enable the CLCx by setting the LCxEN bit of the CLCxCON register.

# 19.2 CLCx Interrupts

An interrupt will be generated upon a change in the output value of the CLCx when the appropriate interrupt enables are set. A rising edge detector and a falling edge detector are present in each CLC for this purpose.

The CLCxIF bit of the associated PIR registers will be set when either edge detector is triggered and its associated enable bit is set. The LCxINTP enables rising edge interrupts and the LCxINTN bit enables falling edge interrupts. Both are located in the CLCxCON register.

To fully enable the interrupt, set the following bits:

- LCxON bit of the CLCxCON register
- · CLCxIE bit of the associated PIE registers
- LCxINTP bit of the CLCxCON register (for a rising edge detection)
- LCxINTN bit of the CLCxCON register (for a falling edge detection)
- PEIE and GIE bits of the INTCON register

The CLCxIF bit of the associated PIR registers must be cleared in software as part of the interrupt service. If another edge is detected while this flag is being cleared, the flag will still be set at the end of the sequence.

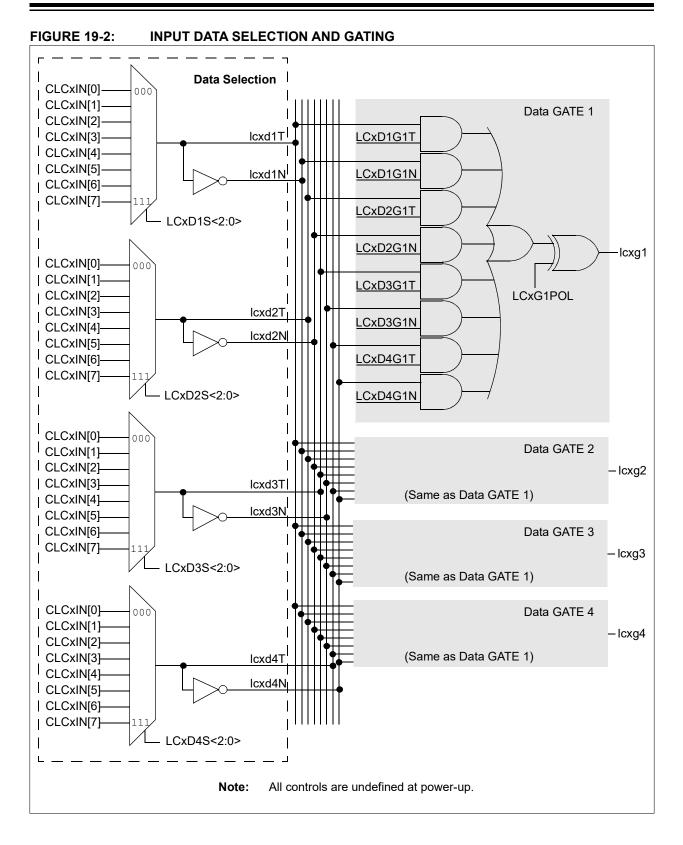
# **19.3 Effects of a Reset**

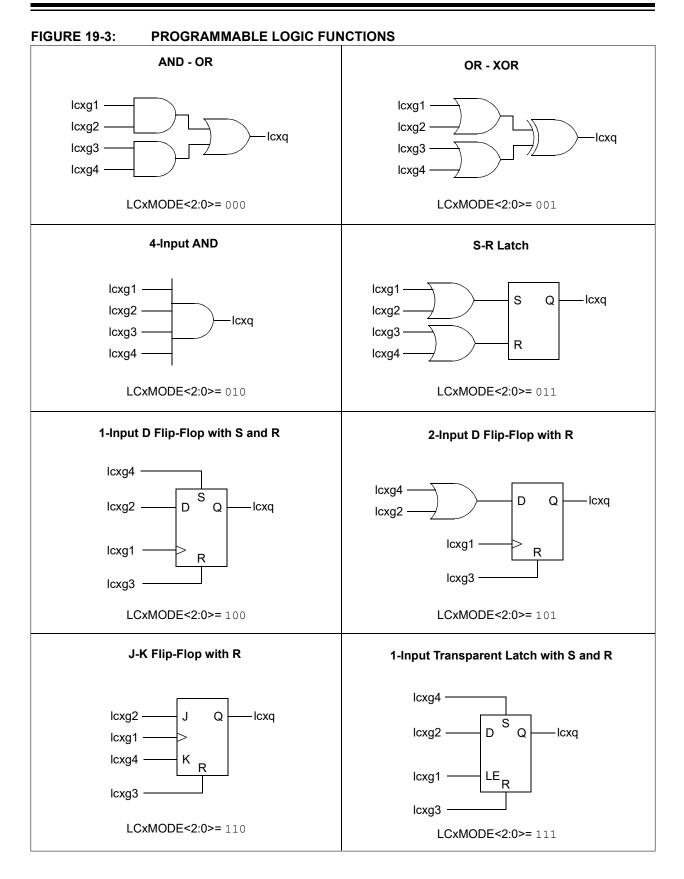
The CLCxCON register is cleared to zero as the result of a Reset. All other selection and gating values remain unchanged.

# **19.4** Operation During Sleep

The selection, gating, and logic functions are not affected by Sleep. Operation will continue provided that the source signals are also not affected by Sleep.

# PIC10(L)F320/322





# 19.5 CLC Control Registers

# REGISTER 19-1: CLCxCON: CONFIGURABLE LOGIC CELL CONTROL REGISTER

R/W-0/0	R/W-0/0	R-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
LCxEN	LCxOE	LCxOUT	LCxINTP	LCxINTN	L	CxMODE<2:0>	>
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable	bit	•	mented bit, read		
u = Bit is unch	0	x = Bit is unkr	nown	-n/n = Value a	at POR and BO	R/Value at all c	other Reset
'1' = Bit is set		'0' = Bit is clea	ared				
bit 7	LCxEN: Con	figurable Logic	Cell Enable b	it			
		able Logic Cell able Logic Cell					
bit 6	1 = Configur	figurable Logic able Logic Cell able Logic Cell	port pin outpu	ut enabled			
bit 5	LCxOUT: Co	nfigurable Logi	c Cell Data O	utput bit			
	Read-only: lo	gic cell output	data, after LC	xPOL; sampled	d from lcx_out w	/ire.	
bit 4		•		• •	nterrupt Enable	bit	
		will be set wher will not be set	n a rising edge	e occurs on lcx	_out		
bit 3	LCxINTN: Co	onfigurable Log	ic Cell Negativ	ve Edge Going	Interrupt Enabl	e bit	
		will be set wher will not be set	n a falling edg	e occurs on lcx	_out		
bit 2-0	LCxMODE<2	2:0>: Configura	ble Logic Cell	Functional Mo	de bits		
	111 = Cell is 110 = Cell is 101 = Cell is	I-input transpa J-K Flip-Flop v 2-input D Flip- I-input D Flip- S-R latch 4-input AND 0 R-XOR	arent latch with vith R Flop with R	h S and R			

R/W-x/u	U-0	U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
LCxPOL	—	—		LCxG4POL	LCxG3POL	LCxG2POL	LCxG1POL	
bit 7							bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, reac	l as '0'		
u = Bit is unch	anged	x = Bit is unkn	own	-n/n = Value a	at POR and BO	R/Value at all o	other Reset	
'1' = Bit is set		'0' = Bit is cleared						
bit 7	LCxPOL: LCO	CxPOL: LCOUT Polarity Control bit						
		1 = The output of the logic cell is inverted						
		ut of the logic o		rted				
bit 6-4	Unimplemen	ted: Read as '	)'					
bit 3		Gate 4 Output I	,					
		ut of gate 4 is in		applied to the	logic cell			
bit 2	•	ut of gate 4 is r Gate 3 Output l		ol hit				
DILZ		ut of gate 3 is i	5					
	•	ut of gate 3 is r			logic cell			
bit 1	LCxG2POL:	<b>_CxG2POL:</b> Gate 2 Output Polarity Control bit						
		<ul> <li>The output of gate 2 is inverted when applied to the logic cell</li> <li>The output of gate 2 is not inverted</li> </ul>						
bit 0	LCxG1POL:	<b>CxG1POL:</b> Gate 1 Output Polarity Control bit						
		ut of gate 1 is i ut of gate 1 is r		applied to the	logic cell			

### REGISTER 19-2: CLCxPOL: SIGNAL POLARITY CONTROL REGISTER

U-0	R/W-x/u		R/W-x/u	U-0	R/W-x/u	R/W-x/u	R/W-x/u		
—		LCxD2S<2:0>(1)		<u> </u>	L	_CxD1S<2:0> <sup>(1)</sup>			
bit 7							bit (		
Legend:									
R = Readab	ole bit	W = Writable b	bit	U = Unimple	mented bit, read	d as '0'			
u = Bit is un	ichanged	x = Bit is unkn	own	-n/n = Value	at POR and BO	R/Value at all o	ther Resets		
'1' = Bit is s	et	'0' = Bit is clea	red						
bit 7	Unimplem	ented: Read as '0	,						
bit 6-4	LCxD2S<2	LCxD2S<2:0>: Input Data 2 Selection Control bits <sup>(1)</sup>							
	111 = CLCxIN[7] is selected for lcxd2.								
	110 = CLC	CxIN[6] is selected	for lcxd2.						
	101 = CLC	CxIN[5] is selected	for lcxd2.						
		CxIN[4] is selected							
		CxIN[3] is selected							
		CxIN[2] is selected							
		CxIN[1] is selected CxIN[0] is selected							
bit 3		ented: Read as '0							
bit 2-0	•	2:0>: Input Data 1		ntrol hits(1)					
		-							
		CxIN[7] is selected CxIN[6] is selected							
		CxIN[5] is selected							
		CxIN[4] is selected							
		CxIN[3] is selected							
		CxIN[2] is selected							
		CxIN[1] is selected							
		CxIN[0] is selected							

#### REGISTER 19-3: CLCxSEL0: MULTIPLEXER DATA 1 AND 2 SELECT REGISTER

**Note 1:** See Table 19-1 for signal names associated with inputs.

U-0	R/W-x/u	R/W-x/u	R/W-x/u	U-0	R/W-x/u	R/W-x/u	R/W-x/u	
_		LCxD4S<2:0>(1)	)	_		LCxD3S<2:0>(1)		
bit 7							bit 0	
Legend:								
R = Readable	e bit	W = Writable ł	oit	U = Unimpler	mented bit, rea	d as '0'		
u = Bit is unc	hanged	x = Bit is unkn	own	-n/n = Value :	at POR and BC	DR/Value at all o	ther Resets	
'1' = Bit is set								
bit 7	Unimpleme	nted: Read as '	)'					
bit 6-4	LCxD4S<2:	0>: Input Data 4	Selection Co	ntrol bits <sup>(1)</sup>				
	111 = CLCxIN[7] is selected for lcxd4.							
	110 = CLC>	xIN[6] is selected	for lcxd4.					
	101 = CLC>	xIN[5] is selected	for lcxd4					
		xIN[4] is selected						
		xIN[3] is selected						
		xIN[2] is selected						
		xIN[1] is selected						
	000 = CLC>	xIN[0] is selected	for lcxd4.					
bit 3	Unimpleme	nted: Read as 'o	)'					
bit 2-0	LCxD3S<2:	0>: Input Data 3	Selection Co	ntrol bits <sup>(1)</sup>				
	111 = CLC>	xIN[7] is selected	for lcxd3.					
		xIN[6] is selected						
	101 = CLC>	xIN[5] is selected	for lcxd3.					
		xIN[4] is selected						
	011 = CLC>	xIN[3] is selected	for lcxd3.					
	010 = CLC>	xIN[2] is selected	for lcxd3.					
	001 = CLC>	xIN[1] is selected	for lcxd3.					
	000 = CLC>	xIN[0] is selected	l for lcxd3.					

#### REGISTER 19-4: CLCxSEL1: MULTIPLEXER DATA 3 AND 4 SELECT REGISTER

**Note 1:** See Table 19-1 for signal names associated with inputs.

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u		
LCxG1D4T	LCxG1D4N	LCxG1D3T	LCxG1D3N	LCxG1D2T	LCxG1D2N	LCxG1D1T	LCxG1D1N		
bit 7							bit (		
Legend: R = Readable	hit	\// = \//ritabla	hit.		nantad hit raad	l aa 'O'			
		W = Writable			nented bit, read at POR and BO		than Deceta		
u = Bit is uncha '1' = Bit is set	angeo	x = Bit is unki '0' = Bit is cle			at POR and BOI	R/Value at all C	liner Resels		
I – DILIS SEL			areu						
bit 7	LCxG1D4T:	Gate 1 Data 4 1	Frue (noninver	ted) bit					
		gated into lcxg		/					
	0 = Icxd4T is	Icxd4T is not gated into Icxg1							
bit 6	LCxG1D4N:	G1D4N: Gate 1 Data 4 Negated (inverted) bit							
		= lcxd4N is gated into lcxg1							
		= lcxd4N is not gated into lcxg1							
bit 5		Sate 1 Data 3 1		ted) bit					
		gated into lcxg not gated into	•						
bit 4		Gate 1 Data 3	•	rted) bit					
		gated into lcx	•						
		not gated into							
bit 3	LCxG1D2T: 0	Gate 1 Data 2 T	True (noninvert	ted) bit					
		gated into lcxg							
	0 = Icxd2T is	not gated into	lcxg1						
bit 2		Gate 1 Data 2	•	rted) bit					
		gated into lcx							
1.11.4		D = lcxd2N is not gated into lcxg1 LCxG1D1T: Gate 1 Data 1 True (noninverted) bit							
bit 1									
		gated into lcxg not gated into							
bit 0		0	0	rted) bit					
		<b>xG1D1N:</b> Gate 1 Data 1 Negated (inverted) bit lcxd1N is gated into lcxg1							
	0 = lcxd1N is								

# REGISTER 19-5: CLCxGLS0: GATE 1 LOGIC SELECT REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
LCxG2D4T	LCxG2D4N	LCxG2D3T	LCxG2D3N	LCxG2D2T	LCxG2D2N	LCxG2D1T	LCxG2D1N	
bit 7	•					•	bit 0	
Legend:								
R = Readable	bit	W = Writable	hit	II – Unimplor	nented bit, read			
					at POR and BO		thar Deceta	
u = Bit is uncha	angeo	x = Bit is unkn			at POR and BO	R/Value at all 0	iner Reseis	
'1' = Bit is set		'0' = Bit is clea	ared					
bit 7	I CxG2D4T (	Gate 2 Data 4 T	rue (noninvert	ed) bit				
		gated into lcxg	•	.ou) bit				
		not gated into						
bit 6		<b>D4N:</b> Gate 2 Data 4 Negated (inverted) bit						
		lcxd4N is gated into lcxg2						
		Icxd4N is not gated into Icxg2						
bit 5	LCxG2D3T: O	Gate 2 Data 3 T	rue (noninvert	ed) bit				
	1 = Icxd3T is	gated into lcxg	2					
	0 = Icxd3T is	not gated into	lcxg2					
bit 4	LCxG2D3N:	Gate 2 Data 3 I	Negated (inver	ted) bit				
		gated into lcxg						
		not gated into	•					
bit 3	LCxG2D2T: O	Gate 2 Data 2 T	rue (noninvert	ed) bit				
		gated into lcxg						
		not gated into	•					
bit 2		Gate 2 Data 2 I	•	ted) bit				
		gated into loxo	·					
		= lcxd2N is not gated into lcxg2						
bit 1		Gate 2 Data 1 T						
		gated into lcxg						
		not gated into	•					
bit 0		Gate 2 Data 1	•	ted) bit				
		gated into long	•					
	0 - 1000  mm is	not gated into	icxyz					

# REGISTER 19-6: CLCxGLS1: GATE 2 LOGIC SELECT REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u		
LCxG3D4T	LCxG3D4N	LCxG3D3T	LCxG3D3N	LCxG3D2T	LCxG3D2N	LCxG3D1T	LCxG3D1N		
bit 7							bit (		
Legend:									
R = Readable	bit	W = Writable	hit	II – Unimplor	nented bit, read	l ac '0'			
u = Bit is uncha		x = Bit is unkr			at POR and BO		thar Pasata		
'1' = Bit is set	angeu	0' = Bit is clear				N value at all C			
i – Dit is set			areu						
bit 7	LCxG3D4T:	Gate 3 Data 4 1	Frue (noninvert	ted) bit					
		gated into lcxo							
		= lcxd4T is not gated into lcxg3							
bit 6	LCxG3D4N:	G3D4N: Gate 3 Data 4 Negated (inverted) bit							
		Icxd4N is gated into Icxg3							
		= lcxd4N is not gated into lcxg3							
bit 5		Gate 3 Data 3	<b>`</b>	ted) bit					
		gated into lcxg not gated into							
bit 4		Gate 3 Data 3	•	rted) bit					
DIL 4		gated into lcx	•	neu) bit					
		not gated into							
bit 3	LCxG3D2T: G	Gate 3 Data 2	Frue (noninver	ted) bit					
		gated into lcxg							
	0 = Icxd2T is	not gated into	lcxg3						
bit 2	LCxG3D2N:	Gate 3 Data 2	Negated (inve	rted) bit					
		gated into lcx							
		D = lcxd2N is not gated into lcxg3 LCxG3D1T: Gate 3 Data 1 True (noninverted) bit							
bit 1									
		gated into lcxg not gated into							
bit 0		0	0	rted) hit					
		xG3D1N: Gate 3 Data 1 Negated (inverted) bit lcxd1N is gated into lcxg3							
	0 = lcxd1N is								

# REGISTER 19-7: CLCxGLS2: GATE 3 LOGIC SELECT REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	
LCxG4D4T	LCxG4D4N	LCxG4D3T	LCxG4D3N	LCxG4D2T	LCxG4D2N	LCxG4D1T	LCxG4D1N	
bit 7							bit C	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, read	l as '0'		
u = Bit is uncha		x = Bit is unkr		•	at POR and BO		ther Resets	
'1' = Bit is set	anged	'0' = Bit is clea						
		e Bitle clot						
bit 7	LCxG4D4T:	Gate 4 Data 4 T	rue (noninvert	ed) bit				
		gated into lcxg		,				
		not gated into						
bit 6	LCxG4D4N:	<b>D4N:</b> Gate 4 Data 4 Negated (inverted) bit						
		Icxd4N is gated into Icxg4						
	0 = Icxd4N is	lcxd4N is not gated into lcxg4						
bit 5		Gate 4 Data 3 T	•	ed) bit				
		gated into lcxg						
		not gated into	•					
bit 4		Gate 4 Data 3 I	•	ted) bit				
		gated into lcxg not gated into						
bit 3		Gate 4 Data 2 T		od) bit				
bit 5		gated into lcxg	· ·	eu) bit				
		not gated into						
bit 2		Gate 4 Data 2 I	•	ted) bit				
		gated into lcxc	•					
		not gated into						
bit 1	LCxG4D1T: 0	CxG4D1T: Gate 4 Data 1 True (noninverted) bit						
	1 = Icxd1T is	gated into lcxg	4					
	0 = Icxd1T is	not gated into	lcxg4					
bit 0		Gate 4 Data 1 I	•	ted) bit				
		gated into lcxg	•					
	0 = Icxd1N is	not gated into	Icxg4					

# REGISTER 19-8: CLCxGLS3: GATE 4 LOGIC SELECT REGISTER

Name	Bit7	Bit6	Bit5	Bit4	Blt3	Bit2	Bit1	Bit0	Register on Page
CLC1CON	LC1EN	LC10E	LC10UT	LC1INTP	LC1INTN	L	C1MODE<2:0	>	110
CLC1GLS0	LC1G1D4T	LC1G1D4N	LC1G1D3T	LC1G1D3N	LC1G1D2T	LC1G1D2N	LC1G1D1T	LC1G1D1N	114
CLC1GLS1	LC1G2D4T	LC1G2D4N	LC1G2D3T	LC1G2D3N	LC1G2D2T	LC1G2D2N	LC1G2D1T	LC1G2D1N	115
CLC1GLS2	LC1G3D4T	LC1G3D4N	LC1G3D3T	LC1G3D3N	LC1G3D2T	LC1G3D2N	LC1G3D1T	LC1G3D1N	116
CLC1GLS3	LC1G4D4T	LC1G4D4N	LC1G4D3T	LC1G4D3N	LC1G4D2T	LC1G4D2N	LC1G4D1T	LC1G4D1N	117
CLC1POL	LC1POL	_	_	_	LC1G4POL	LC1G3POL	LC1G2POL	LC1G1POL	111
CLC1SEL0	—		LC1D2S<2:0>		_		LC1D1S<2:0>		112
CLC1SEL1	—		LC1D4S<2:0>		_		LC1D3S<2:0>		113
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	40
PIE1	_	ADIE		NCO1IE	CLC1IE	_	TMR2IE	_	41
PIR1	_	ADIF	_	NCO1IF	CLC1IF	_	TMR2IF	_	42
TRISA	_	_	_	_	_	TRISA2	TRISA1	TRISA0	69

#### TABLE 19-3: SUMMARY OF REGISTERS ASSOCIATED WITH CLCx

Legend: — = unimplemented read as '0'. Shaded cells are not used for CLC module.

# 20.0 NUMERICALLY CONTROLLED OSCILLATOR (NCO) MODULE

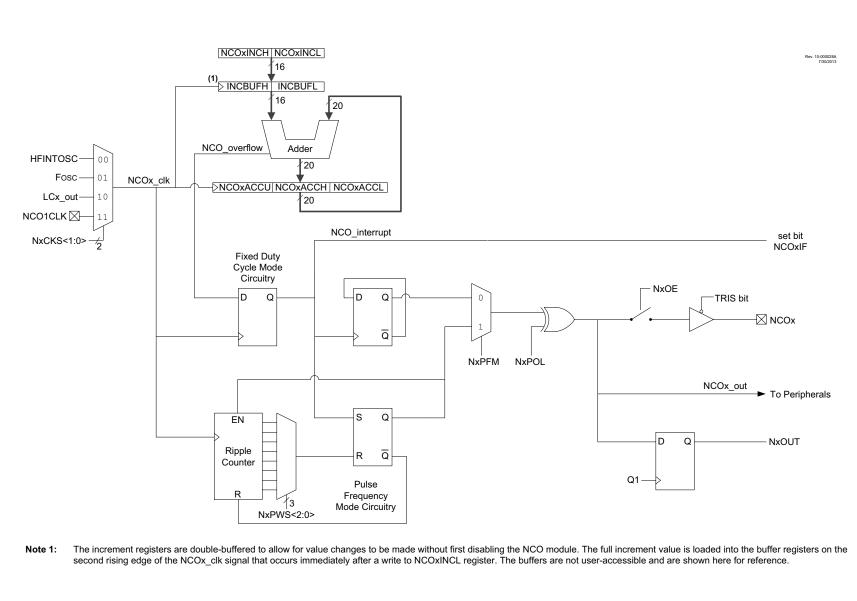
The Numerically Controlled Oscillator (NCOx) module is a timer that uses the overflow from the addition of an increment value to divide the input frequency. The advantage of the addition method over simple counter driven timer is that the resolution of division does not vary with the divider value. The NCOx is most useful for applications that requires frequency accuracy and fine resolution at a fixed duty cycle.

Features of the NCOx include:

- 16-bit increment function
- Fixed Duty Cycle (FDC) mode
- Pulse Frequency (PF) mode
- Output pulse width control
- Multiple clock input sources
- Output polarity control
- · Interrupt capability

Figure 20-1 is a simplified block diagram of the NCOx module.





# 20.1 NCOx OPERATION

The NCOx operates by repeatedly adding a fixed value to an accumulator. Additions occur at the input clock rate. The accumulator will overflow with a carry periodically, which is the raw NCOx output. This effectively reduces the input clock by the ratio of the addition value to the maximum accumulator value. See Equation 20-1.

The NCOx output can be further modified by stretching the pulse or toggling a flip-flop. The modified NCOx output is then distributed internally to other peripherals and optionally output to a pin. The accumulator overflow also generates an interrupt.

The NCOx output creates an instantaneous frequency, which may cause uncertainty. This output depends on the ability of the receiving circuit (i.e., CWG or external resonant converter circuitry) to average the instantaneous frequency to reduce uncertainty.

#### 20.1.1 NCOx CLOCK SOURCES

Clock sources available to the NCOx include:

- HFINTOSC
- Fosc
- LC1OUT
- NCO1CLK pin

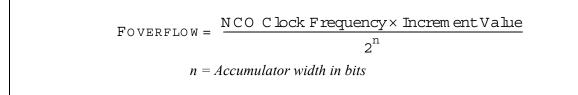
The NCOx clock source is selected by configuring the NxCKS<1:0> bits in the NCOxCLK register.

#### 20.1.2 ACCUMULATOR

The Accumulator is a 20-bit register. Read and write access to the Accumulator is available through three registers:

- NCOxACCL
- NCOxACCH
- NCOxACCU

#### EQUATION 20-1:



#### ock asyncl

20.1.3

The NCOx Adder is a full adder, which operates asynchronously to the clock source selected. The addition of the previous result and the increment value replaces the accumulator value on the rising edge of each input clock.

#### 20.1.4 INCREMENT REGISTERS

ADDER

The Increment value is stored in two 8-bit registers making up a 16-bit increment. In order of LSB to MSB they are:

- NCOxINCL
- NCOxINCH

Both of the registers are readable and writable. The Increment registers are double-buffered to allow for value changes to be made without first disabling the NCOx module.

The buffer loads are immediate when the module is disabled. Writing to the MS register first is necessary because then the buffer is loaded synchronously with the NCOx operation after the write is executed on the lower increment register.

**Note:** The increment buffer registers are not useraccessible.

# 20.2 FIXED DUTY CYCLE (FDC) MODE

In Fixed Duty Cycle (FDC) mode, every time the Accumulator overflows, the output is toggled. This provides a 50% duty cycle, provided that the increment value remains constant. For more information, see Figure 20-2.

The FDC mode is selected by clearing the NxPFM bit in the NCOxCON register.

# 20.3 PULSE FREQUENCY (PF) MODE

In Pulse Frequency (PF) mode, every time the Accumulator overflows, the output becomes active for one or more clock periods. See Section 20.3.1 "OUTPUT PULSE WIDTH CONTROL" for more information. Once the clock period expires, the output returns to an Inactive state. This provides a pulsed output.

The output becomes active on the rising clock edge immediately following the overflow event. For more information, see Figure 20-2.

The value of the Active and Inactive states depends on the Polarity bit, NxPOL in the NCOxCON register.

The PF mode is selected by setting the NxPFM bit in the NCOxCON register.

20.3.1 OUTPUT PULSE WIDTH CONTROL

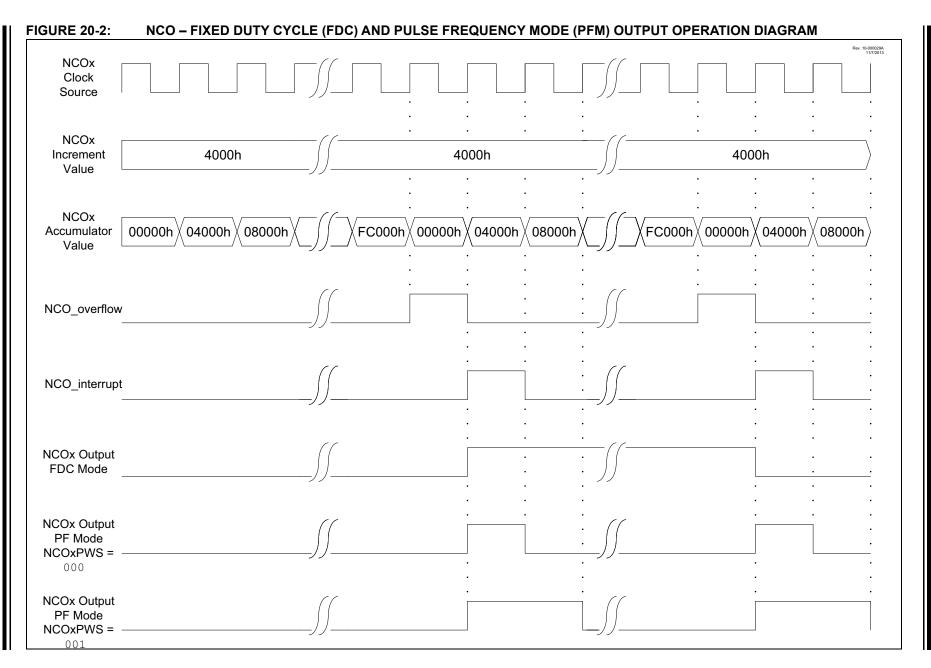
When operating in PF mode, the Active state of the output can vary in width by multiple clock periods. Various pulse widths are selected with the NxPWS<2:0> bits in the NCOxCLK register.

When the selected pulse width is greater than the Accumulator overflow time frame, then NCOx operation is undefined.

# 20.4 OUTPUT POLARITY CONTROL

The last stage in the NCOx module is the output polarity. The NxPOL bit in the NCOxCON register selects the output polarity. Changing the polarity while the interrupts are enabled will cause an interrupt for the resulting output transition.

The NCOx output can be used internally by source code or other peripherals. This is done by reading the NxOUT (read-only) bit of the NCOxCON register.



# 20.5 Interrupts

When the Accumulator overflows, the NCOx Interrupt Flag bit, NCOxIF, of the PIR1 register is set. To enable this interrupt event, the following bits must be set:

- NxEN bit of the NCOxCON register
- NCOxIE bit of the PIE1 register
- PEIE bit of the INTCON register
- · GIE bit of the INTCON register

The interrupt must be cleared by software by clearing the NCOxIF bit in the Interrupt Service Routine.

# 20.6 Effects of a Reset

All of the NCOx registers are cleared to zero as the result of a Reset.

# 20.7 Operation In Sleep

The NCO module operates independently from the system clock and will continue to run during Sleep, provided that the clock source selected remains active.

The HFINTOSC remains active during Sleep when the NCO module is enabled and the HFINTOSC is selected as the clock source, regardless of the system clock source selected.

In other words, if the HFINTOSC is simultaneously selected as the system clock and the NCO clock source, when the NCO is enabled, the CPU will go idle during Sleep, but the NCO will continue to operate and the HFINTOSC will remain active.

This will have a direct effect on the Sleep mode current.

# 20.8 NCOx Control Registers

R/W-0/0	R/W-0/0	R-0/0	R/W-0/0	U-0	U-0	U-0	R/W-0/0
NxEN	NxOE	NxOUT	NxPOL		_		NxPFM
bit 7					•		bit 0
r							
Legend:							
R = Readable b	it	W = Writable bi	t	U = Unimpleme	ented bit, read as	ʻ0'	
u = Bit is unchai	nged	x = Bit is unkno	wn	-n/n = Value at	POR and BOR/V	alue at all other/	Resets
'1' = Bit is set		'0' = Bit is clear	ed				
bit 7	NXEN: NCOX E	Enable bit					
	1 = NCOx mod	ule is enabled					
	0 = NCOx mod	ule is disabled					
bit 6	NxOE: NCOx (	Output Enable bit					
		ut pin is enabled					
	0 = NCOx outp	ut pin is disabled					
bit 5	NxOUT: NCOx	•					
	1 = NCOx outp	0					
	0 = NCOx outp						
bit 4	NxPOL: NCOx	,	Lever Concentral				
		out signal is active out signal is active	```	(be			
bit 3-1	•	ed: Read as '0'.					
bit 0	•		v modo bit				
DILU		Pulse Frequenc rates in Pulse Fre					
		rates in Fixed Du					

#### REGISTER 20-1: NCOxCON: NCOx CONTROL REGISTER

0 = NCOx operates in Fixed Duty Cycle mode

# REGISTER 20-2: NCOxCLK: NCOx INPUT CLOCK CONTROL REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	U-0	U-0	U-0	R/W-0/0	R/W-0/0
	NxPWS<2:0>(1,2)		—	—		NxCKS	S<1:0>
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-5	NxPWS<2:0>: NCOx Output Pulse Width Select bits <sup>(1, 2)</sup> 111 = 128 NCOx clock periods 110 = 64 NCOx clock periods 101 = 32 NCOx clock periods 100 = 16 NCOx clock periods 011 = 8 NCOx clock periods 010 = 4 NCOx clock periods 001 = 2 NCOx clock periods 000 = 1 NCOx clock periods
bit 4-2	Unimplemented: Read as '0'
bit 1-0	NxCKS<1:0>: NCOx Clock Source Select bits 11 = LC1OUT 10 = HFINTOSC (16 MHz) 01 = FOSC 00 = NCO1CLK pin
Note 1. NxP	WS applies only when operating in Pulse Frequency mode

**Note 1:** NxPWS applies only when operating in Pulse Frequency mode.

2: If NCOx pulse width is greater than NCOx overflow period, operation is undefined.

### REGISTER 20-3: NCOxACCL: NCOx ACCUMULATOR REGISTER – LOW BYTE

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0		
NCOxACC<7:0>									
bit 7							bit 0		
Legend:									
			••						

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

#### bit 7-0 NCOxACC<7:0>: NCOx Accumulator, low byte

Note 1: NxPWS applies only when operating in Pulse Frequency mode.

2: If NCOx pulse width is greater than NCOx overflow period, operation is undefined.

#### REGISTER 20-4: NCOxACCH: NCOx ACCUMULATOR REGISTER – HIGH BYTE

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0		
NCOxACC<15:8>									
bit 7 bit 0									

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

# bit 7-0 NCOxACC<15:8>: NCOx Accumulator, high byte

#### REGISTER 20-5: NCOxACCU: NCOx ACCUMULATOR REGISTER – UPPER BYTE

U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	
—	—	—	—	NCOxACC<19:16>				
bit 7							bit 0	

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-4 Unimplemented: Read as '0'

bit 3-0 NCOxACC<19:16>: NCOx Accumulator, upper byte

#### **REGISTER 20-6:** NCOxINCL: NCOx INCREMENT REGISTER – LOW BYTE

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-1/1		
NCOxINC<7:0>									
bit 7							bit 0		
Legend:									

Legena:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 NCOxINC<7:0>: NCOx Increment, low byte

#### REGISTER 20-7: NCOxINCH: NCOx INCREMENT REGISTER – HIGH BYTE

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0		
NCOxINC<15:8>									
bit 7 bit 0									

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 NCOxINC<15:8>: NCOx Increment, high byte

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
CLC1SEL0		LC1D2S2	LC1D2S1	LC1D2S0	-	LC1D1S2	LC1D1S1	LC1D1S0	112
CLC1SEL1	_	LC1D4S2	LC1D4S1	LC1D4S0	_	LC1D3S2	LC1D3S1	LC1D3S0	113
CWG1CON1	G1ASDLB<1:0> G1ASDLA<1:0>				_	—	G1IS	<1:0>	140
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	40
NCO1ACCH	NCO1ACCH<15:8>							126	
NCO1ACCL	NCO1ACCL<7:0>							126	
NCO1ACCU		-	_		NCO1ACCU<19:16				126
NCO1CLK	I	N1PWS<2:0>	>	—	_	— N1CKS<1:0>			125
NCO1CON	N1EN	N10E	N1OUT	N1POL	_	—	—	N1PFM	125
NCO1INCH				NCO1IN0	CH<15:8>				127
NCO1INCL				NCO1IN	CL<7:0>				127
PIE1	_	ADIE	_	NCO1IE	CLC1IE	_	TMR2IE	_	41
PIR1	_	ADIF		NCO1IF	CLC1IF		TMR2IF	—	42
TRISA	_	_	_	_	_	TRISA2	TRISA1	TRISA0	<b>69</b>

#### TABLE 20-1: SUMMARY OF REGISTERS ASSOCIATED WITH NCOx

Legend: x = unknown, u = unchanged, - = unimplemented read as '0', q = value depends on condition. Shaded cells are not used for NCO module.

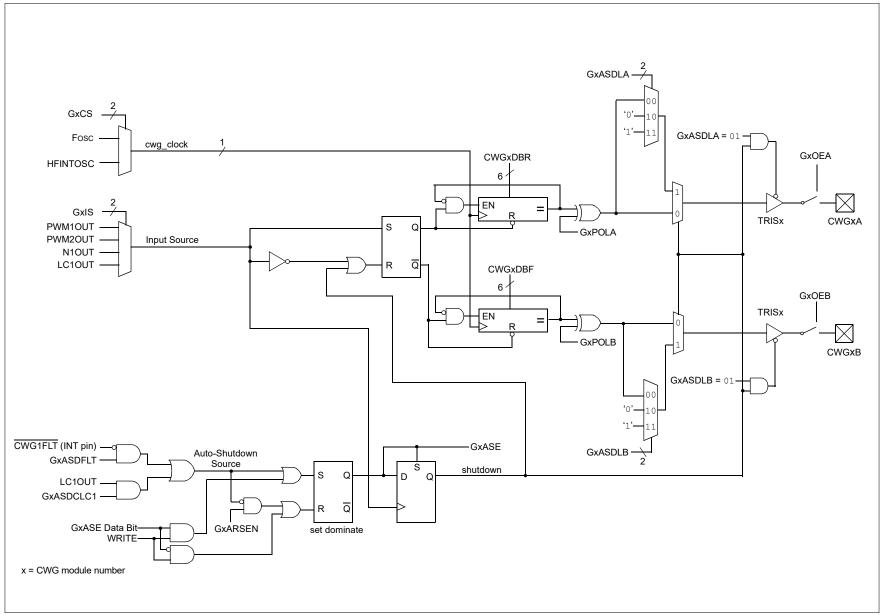
# 21.0 COMPLEMENTARY WAVEFORM GENERATOR (CWG) MODULE

The Complementary Waveform Generator (CWG) produces a complementary waveform with dead-band delay from a selection of input sources.

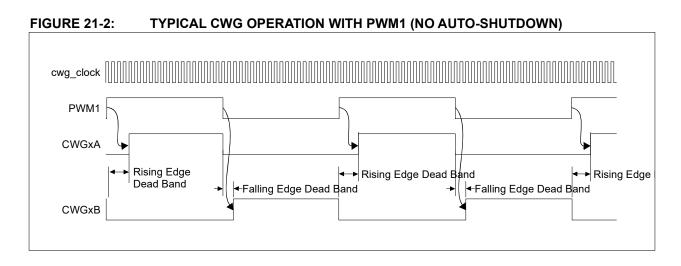
The CWG module has the following features:

- · Selectable dead-band clock source control
- Selectable input sources
- Output enable control
- Output polarity control
- Dead-band control with Independent 6-bit rising and falling edge dead-band counters
- Auto-shutdown control with:
  - Selectable shutdown sources
  - Auto-restart enable
  - Auto-shutdown pin override control

#### FIGURE 21-1: CWG BLOCK DIAGRAM



PIC10(L)F320/322



### 21.1 Fundamental Operation

The CWG generates a two output complementary waveform from one of four selectable input sources.

The off-to-on transition of each output can be delayed from the on-to-off transition of the other output, thereby, creating a time delay immediately where neither output is driven. This is referred to as dead time and is covered in **Section 21.5 "Dead-Band Control"**. A typical operating waveform, with dead band, generated from a single input signal is shown in Figure 21-2.

It may be necessary to guard against the possibility of circuit faults or a feedback event arriving too late or not at all. In this case, the active drive must be terminated before the Fault condition causes damage. This is referred to as auto-shutdown and is covered in **Section 21.9 "Auto-Shutdown Control"**.

# 21.2 Clock Source

The CWG module allows the following clock sources to be selected:

- Fosc (system clock)
- HFINTOSC (16 MHz only)

The clock sources are selected using the G1CS0 bit of the CWGxCON0 register (Register 21-1).

# 21.3 Selectable Input Sources

The CWG can generate the complementary waveform for the following input sources:

- PWM1
- PWM2
- N1OUT
- LC1OUT

The input sources are selected using the GxIS<1:0> bits in the CWGxCON1 register (Register 21-2).

# 21.4 Output Control

Immediately after the CWG module is enabled, the complementary drive is configured with both CWGxA and CWGxB drives cleared.

#### 21.4.1 OUTPUT ENABLES

Each CWG output pin has individual output enable control. Output enables are selected with the GxOEA and GxOEB bits of the CWGxCON0 register. When an output enable control is cleared, the module asserts no control over the pin. When an output enable is set, the override value or active PWM waveform is applied to the pin per the port priority selection. The output pin enables are dependent on the module enable bit, GxEN. When GxEN is cleared, CWG output enables and CWG drive levels have no effect.

#### 21.4.2 POLARITY CONTROL

The polarity of each CWG output can be selected independently. When the output polarity bit is set, the corresponding output is active-high. Clearing the output polarity bit configures the corresponding output as active-low. However, polarity does not affect the override levels. Output polarity is selected with the GxPOLA and GxPOLB bits of the CWGxCON0 register.

# 21.5 Dead-Band Control

Dead-band control provides for non-overlapping output signals to prevent shoot-through current in power switches. The CWG contains two 6-bit dead-band counters. One dead-band counter is used for the rising edge of the input source control. The other is used for the falling edge of the input source control.

Dead band is timed by counting CWG clock periods from zero up to the value in the rising or falling deadband counter registers. See CWGxDBR and CWGxDBF registers (Register 21-4 and Register 21-5, respectively).

# 21.6 Rising Edge Dead Band

The rising edge dead band delays the turn-on of the CWGxA output from when the CWGxB output is turned off. The rising edge dead-band time starts when the rising edge of the input source signal goes true. When this happens, the CWGxB output is immediately turned off and the rising edge dead-band delay time starts. When the rising edge dead-band delay time is reached, the CWGxA output is turned on.

The CWGxDBR register sets the duration of the deadband interval on the rising edge of the input source signal. This duration is from 0 to 64 counts of dead band.

Dead band is always counted off the edge on the input source signal. A count of 0 (zero), indicates that no dead band is present.

If the input source signal is not present for enough time for the count to be completed, no output will be seen on the respective output.

# 21.7 Falling Edge Dead Band

The falling edge dead band delays the turn-on of the CWGxB output from when the CWGxA output is turned off. The falling edge dead-band time starts when the falling edge of the input source goes true. When this happens, the CWGxA output is immediately turned off and the falling edge dead-band delay time starts. When the falling edge dead-band delay time is reached, the CWGxB output is turned on.

The CWGxDBF register sets the duration of the deadband interval on the falling edge of the input source signal. This duration is from 0 to 64 counts of dead band.

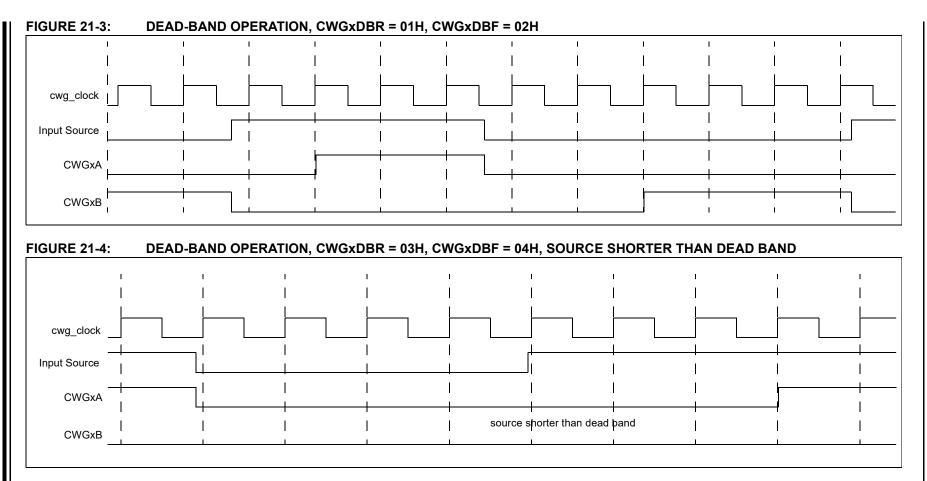
Dead band is always counted off the edge on the input source signal. A count of 0 (zero), indicates that no dead band is present.

If the input source signal is not present for enough time for the count to be completed, no output will be seen on the respective output.

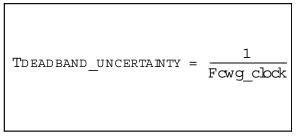
Refer to Figure 21-3 and Figure 21-4 for examples.

# 21.8 Dead-Band Uncertainty

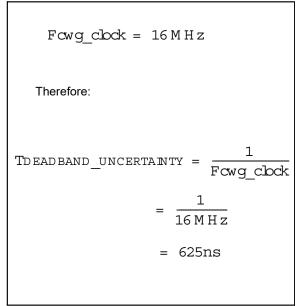
When the rising and falling edges of the input source triggers the dead-band counters, the input may be asynchronous. This will create some uncertainty in the dead-band time delay. The maximum uncertainty is equal to one CWG clock period. Refer to Equation 21-1 for more details.



# EQUATION 21-1: DEAD-BAND DELAY TIME UNCERTAINTY



#### EXAMPLE 21-1: DEAD-BAND DELAY TIME UNCERTAINTY



### 21.9 Auto-Shutdown Control

Auto-shutdown is a method to immediately override the CWG output levels with specific overrides that allow for safe shutdown of the circuit. The Shutdown state can be either cleared automatically or held until cleared by software.

#### 21.9.1 SHUTDOWN

The Shutdown state can be entered by either of the following two methods:

- Software generated
- External Input

#### 21.9.1.1 Software Generated Shutdown

Setting the GxASE bit of the CWGxCON2 register will force the CWG into the Shutdown state.

When auto-restart is disabled, the Shutdown state will persist as long as the GxASE bit is set.

When auto-restart is enabled, the GxASE bit will clear automatically and resume operation on the next rising edge event. See Figure 21-6.

#### 21.9.1.2 External Input Source

External shutdown inputs provide the fastest way to safely suspend CWG operation in the event of a Fault condition. When any of the selected shutdown inputs goes high, the CWG outputs will immediately go to the selected override levels without software delay. Any combination of two input sources can be selected to cause a Shutdown condition. The two sources are:

- LC10UT
- CWG1FLT

Shutdown inputs are selected using the GxASDS0 and GxASDS1 bits of the CWGxCON2 register. (Register 21-3).

Note:	Shutdown inputs are level sensitive, not
	edge sensitive. The Shutdown state can-
	not be cleared, except by disabling auto-
	shutdown, as long as the shutdown input
	level persists.

### 21.10 Operation During Sleep

The CWG module operates independently from the system clock and will continue to run during Sleep, provided that the clock and input sources selected remain active.

The HFINTOSC remains active during Sleep, provided that the CWG module is enabled, the input source is active, and the HFINTOSC is selected as the clock source, regardless of the system clock source selected.

In other words, if the HFINTOSC is simultaneously selected as the system clock and the CWG clock source, when the CWG is enabled and the input source is active, the CPU will go idle during Sleep, but the CWG will continue to operate and the HFINTOSC will remain active.

This will have a direct effect on the Sleep mode current.

# 21.11 Configuring the CWG

The following steps illustrate how to properly configure the CWG to ensure a synchronous start:

- 1. Ensure that the TRIS control bits corresponding to CWGxA and CWGxB are set so that both are configured as inputs.
- 2. Clear the GxEN bit, if not already cleared.
- 3. Set desired dead-band times with the CWGxDBR and CWGxDBF registers.
- 4. Setup the following controls in CWGxCON2 auto-shutdown register:
  - · Select desired shutdown source.
  - Select both output overrides to the desired levels (this is necessary even if not using auto-shutdown because start-up will be from a Shutdown state).
  - Set the GxASE bit and clear the GxARSEN bit.
- 5. Select the desired input source using the CWGxCON1 register.
- 6. Configure the following controls in CWGxCON0 register:
  - · Select desired clock source.
  - · Select the desired output polarities.
  - Set the output enables for the outputs to be used.
- 7. Set the GxEN bit.
- Clear TRIS control bits corresponding to CWGxA and CWGxB to be used to configure those pins as outputs.
- If auto-restart is to be used, set the GxARSEN bit and the GxASE bit will be cleared automatically. Otherwise, clear the GxASE bit to start the CWG.

### 21.11.1 PIN OVERRIDE LEVELS

The levels driven to the output pins, while the shutdown input is true, are controlled by the GxASDLA and GxASDLB bits of the CWGxCON1 register (Register 21-2). GxASDLA controls the CWG1A override level and GxASDLB controls the CWG1B override level. The control bit logic level corresponds to the output logic drive level while in the Shutdown state. The polarity control does not apply to the override level.

#### 21.11.2 AUTO-SHUTDOWN RESTART

After an auto-shutdown event has occurred, there are two ways to have resume operation:

- Software controlled
- Auto-restart

The restart method is selected with the GxARSEN bit of the CWGxCON2 register. Waveforms of software controlled and automatic restarts are shown in Figure 21-5 and Figure 21-6.

#### 21.11.2.1 Software controlled restart

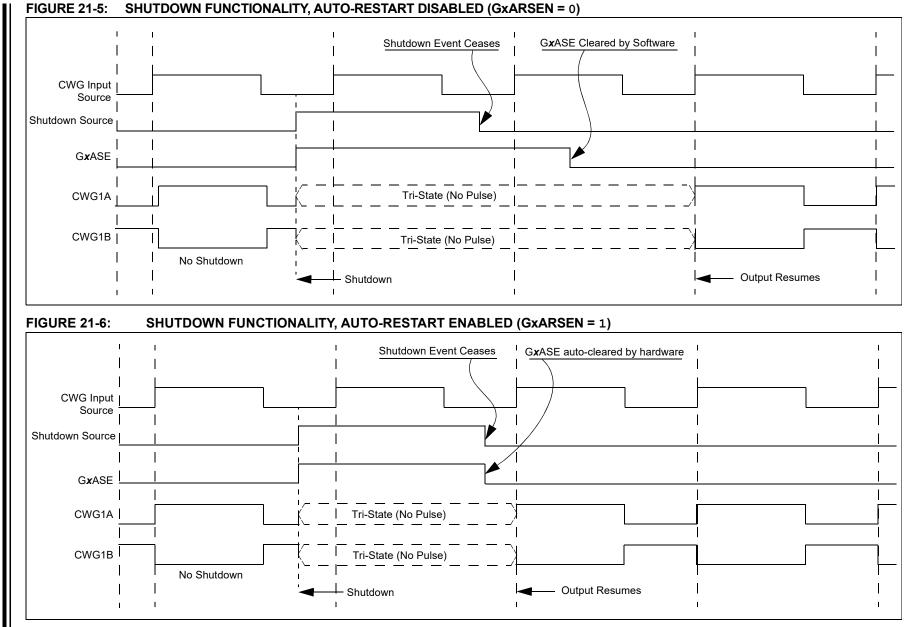
When the GxARSEN bit of the CWGxCON2 register is cleared, the CWG must be restarted after an auto-shut-down event by software.

The CWG will resume operation on the first rising edge event after the GxASE bit is cleared. Clearing the Shutdown state requires all selected shutdown inputs to be low, otherwise the GxASE bit will remain set.

#### 21.11.2.2 Auto-Restart

When the GxARSEN bit of the CWGxCON2 register is set, the CWG will restart from the Auto-shutdown state automatically.

After the shutdown event clears, the GxASE bit will clear automatically and the CWG will resume operation on the first rising edge event.



DS40001585E-page 138

PIC10(L)F320/322

# 21.12 CWG Control Registers

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0				
GxEN	GxOEB	GxOEA	GxPOLB	GxPOLA	_	—	GxCS0				
bit 7		•					bit 0				
Legend:											
R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit, read	as '0'					
u = Bit is unc	hanged	x = Bit is unkr	nown	-n/n = Value a	at POR and BOF	R/Value at all	other Resets				
'1' = Bit is set		'0' = Bit is cle	ared	q = Value dep	pends on condition	on					
bit 7	GxEN: CWG	x Enable bit									
	1 = Module i										
	0 = Module i	s disabled									
bit 6		GxB Output Enable bit									
	<ul> <li>1 = CWGxB is available on appropriate I/O pin</li> <li>0 = CWGxB is not available on appropriate I/O pin</li> </ul>										
L:4 F				te i/O pin							
bit 5			SxA Output Enable bit								
	<ul> <li>1 = CWGxA is available on appropriate I/O pin</li> <li>0 = CWGxA is not available on appropriate I/O pin</li> </ul>										
bit 4											
		<b>GxPOLB:</b> CWGxB Output Polarity bit 1 = Output is inverted polarity									
	0 = Output is normal polarity										
bit 3	GxPOLA: CV	NGxA Output P	olarity bit								
		1 = Output is inverted polarity									
	0 = Output is	s normal polarit	У								
bit 2-1	Unimplemen	Unimplemented: Read as '0'									
bit 0			e Select bit								
		GxCS0: CWGx Clock Source Select bit 1 = HFINTOSC									

#### REGISTER 21-1: CWGxCON0: CWG CONTROL REGISTER 0

#### REGISTER 21-2: CWGxCON1: CWG CONTROL REGISTER 1

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	U-0	U-0	R/W-0/0	R/W-0/0	
GxASE	)LB<1:0>	GxASDL	_A<1:0>		GxIS<		<1:0>	
bit 7							bit 0	
Legend:								
R = Readable	e bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'		
u = Bit is uncl	nanged	x = Bit is unkr	nown	-n/n = Value a	at POR and BO	R/Value at all c	other Resets	
'1' = Bit is set		'0' = Bit is clea	ared	q = Value dep	pends on condit	ion		
bit 7-6 bit 5-4	When an auto 11 = CWGxB 10 = CWGxB 01 = CWGxB 00 = CWGxB control f When an auto 00 = CWGxA control f 01 = CWGxA	<ul> <li>pin is driven to</li> <li>pin is tri-stated</li> <li>pin is driven to</li> <li>pin is driven to</li> <li>the polarity of the</li> <li>c): CWGx Ship</li> <li>c) shutdown even</li> <li>pin is driven to</li> <li>the polarity of the</li> <li>pin is tri-stated</li> </ul>	ont is present o '1', regardle o '0', regardle d o its Inactive s ne output. utdown State ent is present o its Inactive s ne output.	(GxASE = 1): ss of the setting state after the set for CWGxA (GxASE = 1): tate after the set	g of the GxPOLI g of the GxPOLI elected dead-ba elected dead-ba	B bit. and interval. G and interval. G		
					g of the GxPOL			
bit 3-2	Unimplemen	ted: Read as '	כי					
bit 1-0	GxIS<1:0>: ( 11 = LC1OU 10 = N1OUT 01 = PWM2C 00 = PWM1C	DUT	nd Source Se	elect bits				

### **REGISTER 21-3:** CWGxCON2: CWG CONTROL REGISTER 2

R/W/HC/HS-0/0	R/W-0/0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	
GxASE	GxARSEN	—	—	—	_	GxASDCLC1	GxASDFLT	
bit 7 bit 0								

Legend:		
HC = Bit is cleared by hard	ware	HS = Bit is set by hardware
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7	<ul> <li>GxASE: Auto-Shutdown Event Status bit</li> <li>1 = An Auto-Shutdown event has occurred. GxOEB/GxOEA Output Controls overridden, Outputs disabled.</li> <li>0 = No Auto-Shutdown event has occurred, or an Auto-restart has occurred. GxOEB/GxOEA Output Controls enabled.</li> </ul>
bit 6	<b>GxARSEN:</b> Auto-Restart Enable bit 1 = Auto-restart is enabled 0 = Auto-restart is disabled
bit 5-2	Unimplemented: Read as '0'
bit 1	<b>GxASDCLC1:</b> CWG Auto-shutdown Source Enable bit 1 1 = Shutdown when LC1OUT is high 0 = LC1OUT has no effect on shutdown
bit 0	<b>GxASDFLT:</b> CWG Auto-shutdown Source Enable bit 0 1 = <u>Shutdown</u> when CWG1FLT input is low 0 = CWG1FLT input has no effect on shutdown

# REGISTER 21-4: CWGxDBR: COMPLEMENTARY WAVEFORM GENERATOR (CWGx) RISING DEAD-BAND COUNT REGISTER

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u			
—	—		CWGxDBR<5:0>							
bit 7 bit 0										
Legend:										
R = Readable I	bit	W = Writable	bit	U = Unimplen	nented bit, read	l as '0'				
u = Bit is uncha	anged	x = Bit is unkr	x = Bit is unknown -n/n = Value at POR and BOR/Value at all other Resets							
'1' = Bit is set		'0' = Bit is clea	'0' = Bit is cleared q = Value depends on condition							

bit 7-6 Unimplemented: Read as '0'

bit 5-0	CWGxDBR<5:0>: Complementary Waveform Generator (CWGx) Rising Counts bits
	11 1111 = 63-64 counts of dead band
	11 1110 = 62-63 counts of dead band
	•
	•

- •
- 00 0010 = 2-3 counts of dead band
- 00 0001 = 1-2 counts of dead band
- 00 0000 = 0 counts of dead band

# REGISTER 21-5: CWGxDBF: COMPLEMENTARY WAVEFORM GENERATOR (CWGx) FALLING DEAD-BAND COUNT REGISTER

U-0	U-0	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
—	—	CWGxDBF<5:0>					
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	q = Value depends on condition

bit 7-6 Unimplemented: Read as '0'

bit 5-0 CWGxDBF<5:0>: Complementary Waveform Generator (CWGx) Falling Counts bits

- 11 1111 = 63-64 counts of dead band
- 11 1110 = 62-63 counts of dead band
- •
- •
- 00 0010 = 2-3 counts of dead band
- 00 0001 = 1-2 counts of dead band
- 00 0000 = 0 counts of dead band. Dead-band generation is bypassed.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	—	_	—	—	ANSA2	ANSA1	ANSA0	70
CWG1CON0	G1EN	G10EB	G10EA	G1POLB	G1POLA	_	—	G1CS0	139
CWG1CON1	G1ASD	LB<1:0>	G1ASDLA<1:0> G1IS<1:0>			140			
CWG1CON2	G1ASE	G1ARSEN	_	—	—	_	G1ASDCLC1	G1ASDFLT	141
CWG1DBF	_	—			CWG1	DBF<5:0>	•		142
CWG1DBR	_	_		CWG1DBR<5:0>					142
LATA	_	_	_	_	_	LATA2	LATA1	LATA0	70
TRISA	_	_	_	—	_	TRISA2	TRISA1	TRISA0	69

TABLE 21-1: SUMMARY OF REGISTERS ASSOCIATED WITH CWG

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by CWG.

# 22.0 IN-CIRCUIT SERIAL PROGRAMMING™ (ICSP™)

ICSP<sup>™</sup> programming allows customers to manufacture circuit boards with unprogrammed devices. Programming can be done after the assembly process allowing the device to be programmed with the most recent firmware or a custom firmware. Five pins are needed for ICSP<sup>™</sup> programming:

- ICSPCLK
- ICSPDAT
- MCLR/VPP
- VDD
- Vss

In Program/Verify mode the Program Memory, User IDs and the Configuration Words are programmed through serial communications. The ICSPDAT pin is a bidirectional I/O used for transferring the serial data and the ICSPCLK pin is the clock input. For more information on ICSP<sup>TM</sup> refer to the "*PIC10(L)F320/322 Flash Memory Programming Specification*" (DS41572).

# 22.1 High-Voltage Programming Entry Mode

The device is placed into High-Voltage Programming Entry mode by holding the ICSPCLK and ICSPDAT pins low then raising the voltage on MCLR/VPP to VIHH.

# 22.2 Low-Voltage Programming Entry Mode

The Low-Voltage Programming Entry mode allows the PIC<sup>®</sup> Flash MCUs to be programmed using VDD only, without high voltage. When the LVP bit of Configuration Word is set to '1', the low-voltage ICSP programming entry is enabled. To disable the Low-Voltage ICSP mode, the LVP bit must be programmed to '0'.

Entry into the Low-Voltage Programming Entry mode requires the following steps:

- 1. MCLR is brought to VIL.
- 2. A 32-bit key sequence is presented on ICSPDAT, while clocking ICSPCLK.

Once the key sequence is complete, MCLR must be held at VIL for as long as Program/Verify mode is to be maintained.

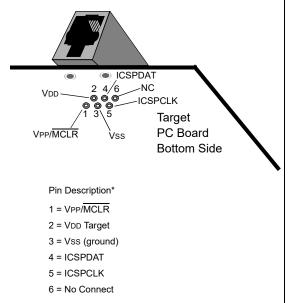
If low-voltage programming is enabled (LVP = 1), the  $\overline{MCLR}$  Reset function is automatically enabled and cannot be disabled. See **Section 5.4** "Low-Power **Brown-out Reset (LPBOR)**" for more information.

The LVP bit can only be reprogrammed to '0' by using the High-Voltage Programming mode.

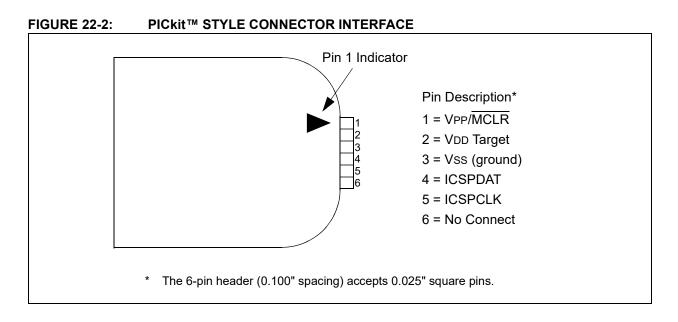
# 22.3 Common Programming Interfaces

Connection to a target device is typically done through an ICSP<sup>™</sup> header. A commonly found connector on development tools is the RJ-11 in the 6P6C (6-pin, 6 connector) configuration. See Figure 22-1.





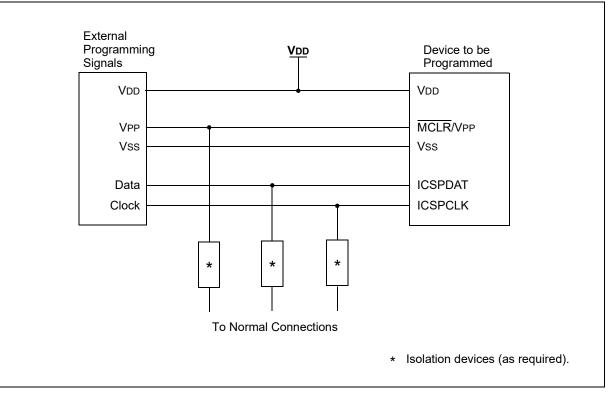
Another connector often found in use with the PICkit<sup>™</sup> programmers is a standard 6-pin header with 0.1 inch spacing. Refer to Figure 22-2.



For additional interface recommendations, refer to your specific device programmer manual prior to PCB design.

It is recommended that isolation devices be used to separate the programming pins from other circuitry. The type of isolation is highly dependent on the specific application and may include devices such as resistors, diodes, or even jumpers. See Figure 22-3 for more information.

#### FIGURE 22-3: TYPICAL CONNECTION FOR ICSP™ PROGRAMMING



### 23.0 INSTRUCTION SET SUMMARY

The PIC10(L)F320/322 instruction set is highly orthogonal and is comprised of three basic categories:

- Byte-oriented operations
- Bit-oriented operations
- · Literal and control operations

Each PIC16 instruction is a 14-bit word divided into an **opcode**, which specifies the instruction type and one or more **operands**, which further specify the operation of the instruction. The formats for each of the categories is presented in Figure 23-1, while the various opcode fields are summarized in Table 23-1.

Table 23-2 lists the instructions recognized by the MPASM<sup>TM</sup> assembler.

For **byte-oriented** instructions, 'f' represents a file register designator and 'd' represents a destination designator. The file register designator specifies which file register is to be used by the instruction.

The destination designator specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the W register. If 'd' is one, the result is placed in the file register specified in the instruction.

For **bit-oriented** instructions, 'b' represents a bit field designator, which selects the bit affected by the operation, while 'f' represents the address of the file in which the bit is located.

For **literal and control** operations, 'k' represents an 8-bit or 11-bit constant, or literal value.

One instruction cycle consists of four oscillator periods; for an oscillator frequency of 4 MHz, this gives a normal instruction execution time of 1  $\mu$ s. All instructions are executed within a single instruction cycle, unless a conditional test is true, or the program counter is changed as a result of an instruction. When this occurs, the execution takes two instruction cycles, with the second cycle executed as a NOP.

All instruction examples use the format '0xhh' to represent a hexadecimal number, where 'h' signifies a hexadecimal digit.

#### 23.1 Read-Modify-Write Operations

Any instruction that specifies a file register as part of the instruction performs a Read-Modify-Write (RMW) operation. The register is read, the data is modified, and the result is stored according to either the instruction or the destination designator 'd'. A read operation is performed on a register even if the instruction writes to that register.

For example, a CLRF PORTA instruction will read PORTA, clear all the data bits, then write the result back to PORTA. This example would have the unintended consequence of clearing the condition that set the IOCIF flag.

#### TABLE 23-1: OPCODE FIELD DESCRIPTIONS

Field	Description
f	Register file address (0x00 to 0x7F)
W	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
x	Don't care location (= $0$ or 1). The assembler will generate code with x = $0$ . It is the recommended form of use for compatibility with all Microchip software tools.
d	Destination select; d = 0: store result in W, d = 1: store result in file register f. Default is d = 1.
PC	Program Counter
TO	Time-out bit
С	Carry bit
DC	Digit carry bit
Z	Zero bit
PD	Power-down bit

### FIGURE 23-1: GENERAL FORMAT FOR INSTRUCTIONS

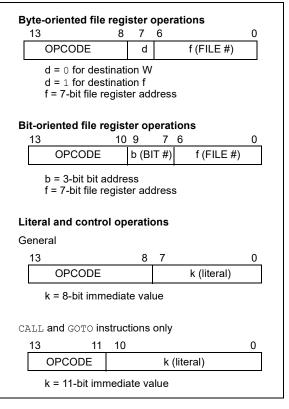


TABLE 23-2:	INSTRUCTION SET
-------------	-----------------

Mnemonic, Operands		Description	Cycles		14-Bit	Opcode	Status	Notes	
		Description		MSb			LSb	Affected	Notes
	BYTE-ORIENTED FILE REGISTER OPERATIONS								
ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C, DC, Z	1, 2
ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	1, 2
CLRF	f	Clear f	1	00	0001	lfff	ffff	Z	2
CLRW	_	Clear W	1	00	0001	0xxx	XXXX	Z	
COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	1, 2
DECF	f, d	Decrement f	1	00	0011	dfff	ffff	Z	1, 2
DECFSZ	f, d	Decrement f, Skip if 0	1 <b>(2)</b>	00	1011	dfff	ffff		1, 2, 3
INCF	f, d	Increment f	1	00	1010	dfff	ffff	Z	1, 2
INCFSZ	f, d	Increment f, Skip if 0	1 <b>(2)</b>	00	1111	dfff	ffff		1, 2, 3
IORWF	f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	1, 2
MOVF	f, d	Move f	1	00	1000	dfff	ffff	Z	1, 2
MOVWF	f	Move W to f	1	00	0000	lfff	ffff		
NOP	_	No Operation	1	00	0000	0xx0	0000		
RLF	f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	С	1, 2
RRF	f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	С	1, 2
SUBWF	f, d	Subtract W from f	1	00	0010	dfff	ffff	C, DC, Z	1, 2
SWAPF	f, d	Swap nibbles in f	1	00	1110	dfff	ffff		1, 2
XORWF	f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	1, 2
		BIT-ORIENTED FILE REGIS		RATION	NS				
BCF	f, b	Bit Clear f	1	01	00bb	bfff	ffff		1, 2
BSF	f, b	Bit Set f	1	01	01bb	bfff	ffff		1, 2
BTFSC	f, b	Bit Test f, Skip if Clear	1 (2)	01	10bb	bfff	ffff		3
BTFSS	f, b	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff		3
		LITERAL AND CONTRO	L OPERAT	IONS					
ADDLW	k	Add literal and W	1	11	111x	kkkk	kkkk	C, DC, Z	
ANDLW	k	AND literal with W	1	11	1001	kkkk	kkkk	Z	
CALL	k	Call Subroutine	2	10	0 k k k	kkkk	kkkk		
CLRWDT	_	Clear Watchdog Timer	1	00	0000	0110	0100	TO, PD	
GOTO	k	Go to address	2	10	1kkk	kkkk	kkkk		
IORLW	k	Inclusive OR literal with W	1	11	1000	kkkk	kkkk	Z	
MOVLW	k	Move literal to W	1	11	00xx	kkkk	kkkk		
RETFIE	_	Return from interrupt	2	00	0000	0000	1001		
RETLW	k	Return with literal in W	2	11	01xx	kkkk	kkkk		
RETURN	_	Return from Subroutine	2	00	0000	0000	1000		
SLEEP	_	Go into Standby mode	1	00	0000	0110	0011	TO, PD	
SUBLW	k	Subtract W from literal	1	11	110x	kkkk	kkkk	C, DC, Z	
XORLW	k	Exclusive OR literal with W	1	11	1010	kkkk	kkkk	Z	

**Note 1:** When an I/O register is modified as a function of itself (e.g., MOVF PORTA, 1), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 module.

**3:** If the Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

ADDLW	Add literal and W		
Syntax:	[ <i>label</i> ] ADDLW k		
Operands:	$0 \leq k \leq 255$		
Operation:	$(W) + k \to (W)$		
Status Affected:	C, DC, Z		
Description:	The contents of the W register are added to the 8-bit literal 'k' and the result is placed in the W register.		

BCF	Bit Clear f
Syntax:	[ <i>label</i> ]BCF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	0 → (f <b>)</b>
Status Affected:	None
Description:	Bit 'b' in register 'f' is cleared.

ADDWF	Add W and f				
Syntax:	[ <i>label</i> ]ADDWF f,d				
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$				
Operation:	(W) + (f) $\rightarrow$ (destination)				
Status Affected:	C, DC, Z				
Description:	Add the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.				

BSF	Bit Set f
Syntax:	[ <i>label</i> ] BSF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	1 → (f <b>)</b>
Status Affected:	None
Description:	Bit 'b' in register 'f' is set.

ANDLW	AND literal with W
Syntax:	[ <i>label</i> ]ANDLW k
Operands:	$0 \leq k \leq 255$
Operation:	(W) .AND. (k) $\rightarrow$ (W)
Status Affected:	Z
Description:	The contents of W register are AND'ed with the 8-bit literal 'k'. The result is placed in the W register.

BTFSC	Bit Test f, Skip if Clear
Syntax:	[ <i>label</i> ]BTFSC f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	skip if (f <b>) = 0</b>
Status Affected:	None
Description:	If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b' in register 'f' is '0', the next instruction is discarded, and a NOP is executed instead, making this a 2-cycle instruction.

ANDWF	AND W with f
Syntax:	[ label ] ANDWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(W) .AND. (f) $\rightarrow$ (destination)
Status Affected:	Z
Description:	AND the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

BTFSS	Bit Test f, Skip if Set
Syntax:	[ <i>label</i> ]BTFSS f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b < 7 \end{array}$
Operation:	skip if (f <b>) = 1</b>
Status Affected:	None
Description:	If bit 'b' in register 'f' is '0', the next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.

CLRWDT	Clear Watchdog Timer
Syntax:	[label] CLRWDT
Operands:	None
Operation:	$\begin{array}{l} \text{00h} \rightarrow \text{WDT} \\ 0 \rightarrow \text{WDT prescaler,} \\ 1 \rightarrow \overline{\text{TO}} \\ 1 \rightarrow \overline{\text{PD}} \end{array}$
Status Affected:	TO, PD
Description:	CLRWDT instruction resets the Watchdog Timer. It also resets the prescaler of the WDT. Status bits TO and PD are set.

CALL	Call Subroutine
Syntax:	[ <i>label</i> ] CALL k
Operands:	$0 \leq k \leq 2047$
Operation:	(PC)+ 1 $\rightarrow$ TOS, k $\rightarrow$ PC<10:0>, (PCLATH<4:3>) $\rightarrow$ PC<12:11>
Status Affected:	None
Description:	Call Subroutine. First, return address (PC + 1) is pushed onto the stack. The 11-bit immediate address is loaded into PC bits <10:0>. The upper bits of the PC are loaded from PCLATH. CALL is a 2-cycle instruction.

COMF	Complement f
Syntax:	[ <i>label</i> ] COMF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	$(\overline{f}) \rightarrow (destination)$
Status Affected:	Z
Description:	The contents of register 'f are complemented. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

CLRF	Clear f
Syntax:	[ label ] CLRF f
Operands:	$0 \leq f \leq 127$
Operation:	$\begin{array}{l} 00h \rightarrow (f) \\ 1 \rightarrow Z \end{array}$
Status Affected:	Z
Description:	The contents of register 'f' are cleared and the Z bit is set.

DECF	Decrement f
Syntax:	[ <i>label</i> ] DECF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d  \in  [0,1] \end{array}$
Operation:	(f) - 1 $\rightarrow$ (destination)
Status Affected:	Z
Description:	Decrement register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

CLRW	Clear W
Syntax:	[label] CLRW
Operands:	None
Operation:	$\begin{array}{l} 00h \rightarrow (W) \\ 1 \rightarrow Z \end{array}$
Status Affected:	Z
Description:	W register is cleared. Zero bit (Z) is set.

DECFSZ	Decrement f, Skip if 0
Syntax:	[ <i>label</i> ] DECFSZ f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d  \in  [0,1] \end{array}$
Operation:	(f) - 1 $\rightarrow$ (destination); skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are decremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', then a NOP is executed instead, making it a 2-cycle instruction.

INCFSZ	Increment f, Skip if 0
Syntax:	[label] INCFSZ f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) + 1 $\rightarrow$ (destination), skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', a NOP is executed instead, making it a 2-cycle instruction.

GOTO	Unconditional Branch
Syntax:	[ <i>label</i> ] GOTO k
Operands:	$0 \leq k \leq 2047$
Operation:	$k \rightarrow PC<10:0>$ PCLATH<4:3> $\rightarrow$ PC<12:11>
Status Affected:	None
Description:	GOTO is an unconditional branch. The 11-bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<4:3>. GOTO is a 2-cycle instruction.

IORLW	Inclusive OR literal with W
Syntax:	[ <i>label</i> ] IORLW k
Operands:	$0 \leq k \leq 255$
Operation:	(W) .OR. $k \rightarrow$ (W)
Status Affected:	Z
Description:	The contents of the W register are OR'ed with the 8-bit literal 'k'. The result is placed in the W register.

INCF	Increment f
Syntax:	[ <i>label</i> ] INCF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(f) + 1 $\rightarrow$ (destination)
Status Affected:	Z
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

IORWF	Inclusive OR W with f
Syntax:	[ <i>label</i> ] IORWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in [0,1] \end{array}$
Operation:	(W) .OR. (f) $\rightarrow$ (destination)
Status Affected:	Z
Description:	Inclusive OR the W register with register 'f'. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

MOVF	Move f		
Syntax:	[ <i>label</i> ] MOVF f,d		
Operands:	$0 \le f \le 127$ $d \in [0,1]$		
Operation:	$(f) \rightarrow (dest)$		
Status Affected:	Z		
Description:	The contents of register 'f' is moved to a destination dependent upon the status of 'd'. If $d = 0$ , destination is W register. If $d = 1$ , the destination is file register 'f' itself. $d = 1$ is useful to test a file register since Status flag Z is affected.		
Words:	1		
Cycles:	1		
<u>Example:</u>	MOVF FSR, 0		
	After Instruction W = value in FSR register Z = 1		

MOVWF	Move W to f	
Syntax:	[ <i>label</i> ] MOVWF f	
Operands:	$0 \leq f \leq 127$	
Operation:	$(W) \rightarrow (f)$	
Status Affected:	None	
Description:	Move data from W register to register 'f'.	
Words:	1	
Cycles:	1	
Example:	MOVW OPTION_REG F	
	Before Instruction $OPTION_REG = 0xFF$ W = 0x4F After Instruction $OPTION_REG = 0x4F$ W = 0x4F	

MOVLW	Move literal to W		
Syntax:	[ <i>label</i> ] MOVLW k		
Operands:	$0 \le k \le 255$		
Operation:	$k \rightarrow (W)$		
Status Affected:	None		
Description:	The 8-bit literal 'k' is loaded into W register. The "don't cares" will assemble as '0's.		
Words:	1		
Cycles:	1		
Example:	MOVLW 0x5A		
	After Instruction		
	W = 0x5A		

NOP	No Operation
Syntax:	[label] NOP
Operands:	None
Operation:	No operation
Status Affected:	None
Description:	No operation.
Words:	1
Cycles:	1
Example:	NOP

RETFIE	Return from Interrupt		
Syntax:	[ <i>label</i> ] RETFIE		
Operands:	None		
Operation:	$\begin{array}{l} TOS \to PC, \\ 1 \to GIE \end{array}$		
Status Affected:	None		
Description:	Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE (INT- CON<7>). This is a 2-cycle instruction.		
Words:	1		
Cycles:	2		
<u>Example:</u>	RETFIE		
	After Interrupt PC = TOS GIE = 1		

RETLW	Return with literal in W	
Syntax:	[ <i>label</i> ] RETLW k	
Operands:	$0 \le k \le 255$	
Operation:	$k \rightarrow (W);$	
	$TOS \rightarrow PC$	
Status Affected:	None	
Description:	The W register is loaded with the 8-bit literal 'k'. The program counter is loaded from the top of the stack (the return address). This is a 2-cycle instruction.	
Words:	1	
Cycles:	2	
<u>Example:</u>	CALL TABLE;W contains ;table offset ;value	
TABLE	GOTO DONE •	
	•	
	ADDWF PC ;W = offset RETLW k1 ;Begin table RETLW k2 ; •	
	• RETLW kn ;End of table	
DONE	REILW KII , ENG OI CADIE	
	Before Instruction W = 0x07	
	After Instruction W = value of k8	
RETURN	Return from Subroutine	
Syntax:	[label] RETURN	
Operands:	None	
Operation:	$TOS \rightarrow PC$	
Status Affected:	None	
Description:	Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a 2-cycle instruc- tion.	

RLF	Rotate Left f through Carry		
Syntax:	[ <i>label</i> ] RLF f,d		
Operands:	$0 \le f \le 127$ $d \in [0,1]$		
Operation:	See description below		
Status Affected:	С		
Description:	The contents of register 'f' are rotated one bit to the left through the Carry flag. If 'd' is '0', the result is placed in the W register.		
	If 'd' is '1', the result is stored back in register 'f'. ▲ C ▲ Register f ▲		
Words:	If 'd' is '1', the result is stored back in register 'f'.		
Words: Cycles:	If 'd' is '1', the result is stored back in register 'f'. ▲ C ← Register f ←		
	If 'd' is '1', the result is stored back in register 'f'. ▲ C ★ Register f ★ 1		
Cycles:	If 'd' is '1', the result is stored back in register 'f'.		
Cycles:	If 'd' is '1', the result is stored back in register 'f'. C - Register f 1 1 RLF REG1, 0		
Cycles:	If 'd' is '1', the result is stored back in register 'f'. C  Register f 1 1 RLF REG1,0 Before Instruction REG1 = 1110 0110 C = 0		
Cycles:	If 'd' is '1', the result is stored back in register 'f'.		
Cycles:	If 'd' is '1', the result is stored back in register 'f'. C  Register f 1 1 RLF REG1,0 Before Instruction REG1 = 1110 0110 C = 0		

RRF	Rotate Right f through Carry		
Syntax:	[ <i>label</i> ] RRF f,d		
Operands:	$\begin{array}{l} 0\leq f\leq 127\\ d\in [0,1] \end{array}$		
Operation:	See description below		
Status Affected:	С		
Description:	C The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.		
	C→ Register f		

SLEEP	Enter Sleep mode
Syntax:	[label] SLEEP
Operands:	None
Operation:	$\begin{array}{l} \text{00h} \rightarrow \text{WDT,} \\ 0 \rightarrow \underline{\text{WDT}} \text{ prescaler,} \\ 1 \rightarrow \overline{\underline{\text{TO}}}, \\ 0 \rightarrow \overline{\text{PD}} \end{array}$
Status Affected:	TO, PD
Description:	The power-down Status bit, $\overline{\text{PD}}$ is cleared. Time-out Status bit, $\overline{\text{TO}}$ is set. Watchdog Timer and its prescaler are cleared. The processor is put into Sleep mode with the oscillator stopped.

SUBLW	Subtract W from literal		
Syntax:	[label] SU	JBLW k	
Operands:	$0 \leq k \leq 255$		
Operation:	$k - (W) \to (W)$		
Status Affected:	C, DC, Z		
Description:	The W register is subtracted (2's complement method) from the 8-bit literal 'k'. The result is placed in the W register.		
	Result	Condition	
	$\mathbf{C} = 0$	W > k	

Result	Condition	
<b>C</b> = 0	W > k	
<b>C =</b> 1	$W \leq k$	
DC = 0	W<3:0> > k<3:0>	
<b>DC =</b> 1	$W<3:0> \le k<3:0>$	

SUBWF	Subtract W from f		
Syntax:	[label] SU	JBWF f,d	
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d  \in  [0,1] \end{array}$		
Operation:	(f) - (W) $\rightarrow$ (destination)		
Status Affected:	C, DC, Z		
Description:	Subtract (2's complement method) W register from register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.		
	<b>C =</b> 0	W > f	
	<b>C =</b> 1	$W \leq f$	
		W<3:0> > f<3:0>	

**DC =** 1

 $W<3:0> \le f<3:0>$ 

XORWF	Exclusive OR W with f
Syntax:	[ <i>label</i> ] XORWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) .XOR. (f) $\rightarrow$ (destination)
Status Affected:	Z
Description:	Exclusive OR the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

SWAPF	Swap Nibbles in f
Syntax:	[ <i>label</i> ] SWAPF f,d
Operands:	$\begin{array}{l} 0\leq f\leq 127\\ d\in [0,1] \end{array}$
Operation:	$(f<3:0>) \rightarrow (destination<7:4>),$ $(f<7:4>) \rightarrow (destination<3:0>)$
Status Affected:	None
Description:	The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed in register 'f'.

XORLW	Exclusive OR literal with W
Syntax:	[ <i>label</i> ] XORLW k
Operands:	$0 \leq k \leq 255$
Operation:	(W) .XOR. $k \rightarrow (W)$
Status Affected:	Z
Description:	The contents of the W register are XOR'ed with the 8-bit literal 'k'. The result is placed in the W register.

### 24.0 ELECTRICAL SPECIFICATIONS

### 24.1 Absolute Maximum Ratings<sup>(†)</sup>

Ambient temperature under bias	°C to +125°C
-65 Storage temperature	
Voltage on pins with respect to Vss	
on VDD pin	
PIC10F320/322	).3V to +6.5V
PIC10LF320/322	).3V to +4.0V
on MCLR pin	).3V to +9.0V
on all other pins	(VDD + 0.3V)
Maximum current	
on Vss pin <sup>(1)</sup>	
-40°C $\leq$ TA $\leq$ +85°C	250 mA
+85°C $\leq$ Ta $\leq$ +125°C	85 mA
on VDD pin <sup>(1)</sup>	
$-40^{\circ}C \le TA \le +85^{\circ}C$	250 mA
+85°C $\leq$ Ta $\leq$ +125°C	85 mA
Sunk by any I/O pin	50 mA
Sourced by any I/O pin	
Clamp current, Iк (VPIN < 0 or VPIN > VDD)	
Total power dissipation <sup>(2)</sup>	800 mW

Note 1: Maximum current rating requires even load distribution across I/O pins. Maximum current rating may be limited by the device package power dissipation characterizations, see Section 24.4 "Thermal Considerations" to calculate device specifications.

2: Power dissipation is calculated as follows: PDIS = VDD x {IDD  $-\sum$  IOH} +  $\sum$  {(VDD - VOH) x IOH} +  $\sum$ (VOI x IOL).

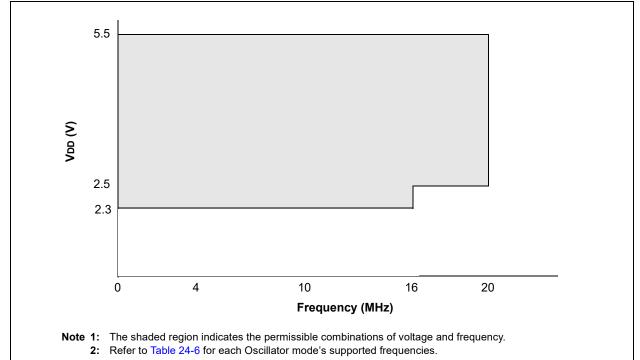
† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure above maximum rating conditions for extended periods may affect device reliability.

### 24.2 Standard Operating Conditions

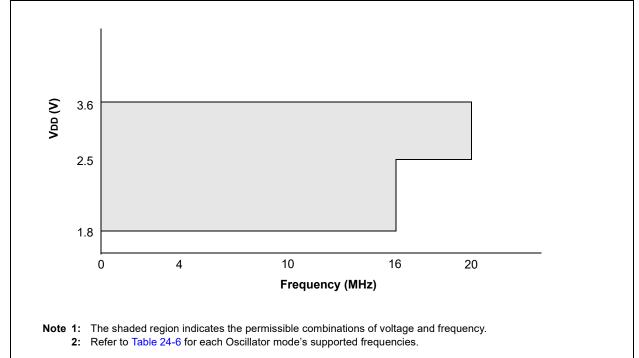
The standard operating conditions for any device are defined as:	
Operating Voltage:VDDMIN $\leq$ VDD $\leq$ VDDMAXOperating Temperature:TA_MIN $\leq$ TA $\leq$ TA_MAX	
/DD — Operating Supply Voltage <sup>(1)</sup>	
PIC10LF320/322	
VDDMIN (Fosc $\leq$ 16 MHz)	V
VDDMIN (16 MHz < Fosc $\leq$ 20 MHz)	V
VDDMAX	V
PIC10F320/322	
VDDMIN (Fosc $\leq$ 16 MHz)	V
VDDMIN (16 MHz < Fosc $\leq$ 20 MHz)	V
VDDMAX	V
A — Operating Ambient Temperature Range	
Industrial Temperature	
TA_MIN	С
TA_MAX	С
Extended Temperature	
TA_MIN40°0	С
TA_MAX +125°	С

Note 1: See Parameter D001, DC Characteristics: Supply Voltage.









#### 24.3 DC Characteristics

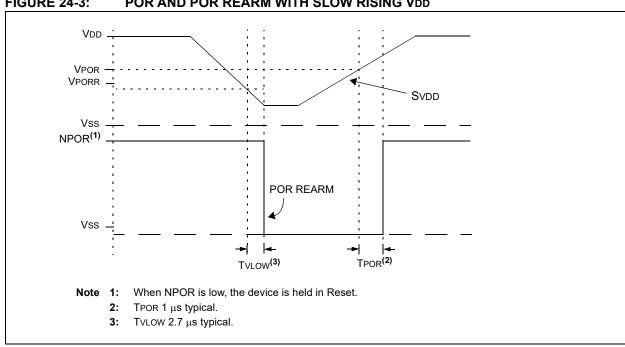
### TABLE 24-1:SUPPLY VOLTAGE

PIC10L	F320/322		Standar	d Opera	ting Cond	ditions (	unless otherwise stated)
PIC10F	320/322						
Param. No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions
D001	Vdd	Supply Voltage					
			1.8 2.5		3.6 3.6	V V	Fosc ≤ 16 MHz: Fosc ≤ 20 MHz
D001			2.3 2.5	_	5.5 5.5	V V	Fosc ≤ 16 MHz: Fosc ≤ 20 MHz
D002*	Vdr	RAM Data Retention Voltage <sup>(1)</sup>					
			1.5		_	V	Device in Sleep mode
D002*			1.7	_	_	V	Device in Sleep mode
	VPOR*	Power-on Reset Release Voltage	—	1.6	—	V	
	VPORR*	Power-on Reset Rearm Voltage	—	0.8	—	V	Device in Sleep mode
			_	1.7	_	V	Device in Sleep mode
D003	VFVR	Fixed Voltage Reference Voltage					
		1x gain (1.024V nominal) 2x gain (2.048V nominal) 4x gain (4.096V nominal)	-8	_	+6	%	$ \begin{array}{l} V{\rm DD} \geq 2.5{\rm V}, \ -40^{\circ}{\rm C} \leq {\rm TA} \leq +85^{\circ}{\rm C} \\ V{\rm DD} \geq 2.5{\rm V}, \ -40^{\circ}{\rm C} \leq {\rm TA} \leq +85^{\circ}{\rm C} \\ V{\rm DD} \geq 4.75{\rm V}, \ -40^{\circ}{\rm C} \leq {\rm TA} \leq +85^{\circ}{\rm C} \end{array} $
D004*	SVDD	VDD Rise Rate to ensure internal Power-on Reset signal	0.05	—	_	V/ms	See Section 5.1 "Power-On Reset (POR)" for details.

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.3V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in Sleep mode without losing RAM data.



#### FIGURE 24-3: POR AND POR REARM WITH SLOW RISING VDD

PIC10LF3	20/322	Standard Operating Conditions (unless otherwise stated)					
PIC10F32							
Param	Device	Min.	Trent	Max	Units		Conditions
No.	Characteristics	with.	Тур†	Max.	Units	VDD	Note
D013		—	34	45	μΑ	1.8	Fosc = 500 kHz
		—	60	105	μA	3.0	EC mode
D013		—	76	101	μA	2.3	Fosc = 500 kHz
		—	110	148	μA	3.0	EC mode
		—	153	211	μA	5.0	
D014		—	190	290	μA	1.8	Fosc = 8 MHz
		—	350	500	μA	3.0	EC mode
D014		_	290	430	μA	2.3	Fosc = 8 MHz
		_	395	600	μA	3.0	EC mode
		—	480	775	μA	5.0	
D015		—	0.8	1.3	mA	3.0	Fosc = 20 MHz
		—	1.1	1.8	mA	3.6	EC mode
D015			0.8	1.4	mA	3.0	Fosc = 20 MHz
		—	1.1	1.8	mA	5.0	EC mode
D016			2.2	4.1	μA	1.8	Fosc = 32 kHz
		—	3.9	6.5	μA	3.0	LFINTOSC mode, 85°C
D016			31	44	μA	2.3	Fosc = 32 kHz
			40	57	μA	3.0	LFINTOSC mode, 85°C
		—	71	117	μA	5.0	
D016A			3.2	4.5	μA	1.8	Fosc = 32 kHz
		—	4.8	7.0	μA	3.0	LFINTOSC mode, 125°C
D016A		_	31	44	μA	2.3	Fosc = 32 kHz
			40	57	μA	3.0	LFINTOSC mode,125°C
		—	71	117	μA	5.0	

#### TABLE 24-2: SUPPLY VOLTAGE (IDD)<sup>(1,2)</sup>

**Note 1:** The test conditions for all IDD measurements in Active Operation mode are: CLKIN = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; MCLR = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

### TABLE 24-2: SUPPLY VOLTAGE (IDD)<sup>(1,2)</sup> (CONTINUED)

PIC10LF3	20/322	22 Standard Operating Conditions (unless otherwise stated)			less otherwise stated)		
PIC10F320/322							
Param	Device	Min.	Typt	Max.	Units		Conditions
No.	Characteristics	WIIII.	Тур†	IVIAX.	Units	VDD	Note
D017		—	213	290	μA	1.8	Fosc = 500 kHz
		—	264	360	μA	3.0	HFINTOSC mode
D017		—	272	368	μA	2.3	Fosc = 500 kHz
		_	310	422	μA	3.0	HFINTOSC mode
			372	515	μA	5.0	
D018		—	0.33	0.50	mA	1.8	Fosc = 8 MHz
		_	0.43	0.70	mA	3.0	HFINTOSC mode
D018		_	0.45	1.0	mA	2.3	Fosc = 8 MHz
			0.56	1.1	mA	3.0	HFINTOSC mode
			0.64	1.2	mA	5.0	
D019		—	0.46	1.1	mA	1.8	Fosc = 16 MHz
		_	0.73	1.2	mA	3.0	HFINTOSC mode
D019			0.60	1.1	mA	2.3	Fosc = 16 MHz
			0.76	1.2	mA	3.0	HFINTOSC mode
		—	0.85	1.3	mA	5.0	

**Note 1:** The test conditions for all IDD measurements in Active Operation mode are: CLKIN = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; MCLR = VDD; WDT disabled.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

PIC10LF3	20/322		Standa	rd Opera	ting Cond	itions (u	nless ot	herwise stated)
PIC10F320	)/322							
Param			<u> </u>	Max.	Max.			Conditions
No.	Device Characteristics	Min.	Тур†	+85°C	+125°C	Units	VDD	Note
D023		—	0.06	1.1	2	μA	1.8	WDT, BOR, and FVR disabled,
		—	0.08	1.3	2	μA	3.0	all Peripherals Inactive
D023		_	0.20	1.1	2	μA	2.3	WDT, BOR, and FVR disabled,
		—	0.30	1.4	2	μA	3.0	all Peripherals Inactive
		—	0.40	2.4	2.4	μA	5.0	
D024		_	0.5	9	11	μA	1.8	WDT Current (Note 1)
		—	0.8	11	13	μA	3.0	
D024		—	4.0	10	12	μA	2.3	WDT Current (Note 1)
		_	4.2	12	14	μA	3.0	
		—	4.3	14	16	μA	5.0	7
D025		_	30	96	120	μA	1.8	FVR current
			39	106	123	μA	3.0	
D025			32	96	120	μA	2.3	FVR current
			39	106	133	μA	3.0	7
			70	136	170	μA	5.0	7
D026			7.5	16	18	μA	3.0	BOR Current (Note 1)
D026			8	18	20	μA	3.0	BOR Current (Note 1)
			9	20	20.2	μA	5.0	7
D026A			2.7	10	15	μA	3.0	LPBOR Current
D026A			3.0	10	15	μA	3.0	LPBOR Current
			3.2	15	20	μA	5.0	7
D028			0.1	4	5	μA	1.8	A/D Current (Note 1, Note 3), no
			0.1	5	6	μA	3.0	conversion in progress
D028		_	3.4	6	7	μA	2.3	A/D Current (Note 1, Note 3), no
		_	3.6	7	8	μA	3.0	conversion in progress
		_	3.8	8	9	μA	5.0	
D029		_	250	_	_	μA	1.8	A/D Current (Note 1, Note 3),
		_	250	—	—	μA	3.0	conversion in progress
D029		—	280	—	—	μA	2.3	A/D Current (Note 1, Note 3),
		_	280	—	—	μA	3.0	conversion in progress
			280	_	_	μA	5.0	

IABLE 24-3: POWER-DOWN CURRENTS (IPD)	<b>TABLE 24-3</b> :	POWER-DOWN CURRENTS (IPD) <sup>(1,2)</sup>
---------------------------------------	---------------------	--

These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The peripheral current is the sum of the base IDD or IPD and the additional current consumed when this peripheral is enabled. The peripheral ∆ current can be determined by subtracting the base IDD or IPD current from this limit. Max values must be used when calculating total current consumption.

2: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in High-Impedance state and tied to VDD.

**3:** A/D oscillator source is FRC.

\*

#### **TABLE 24-4: I/O PORTS**

Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions
	VIL	Input Low Voltage					
		I/O PORT:					
D032		with TTL buffer	—		0.8	V	$4.5V \leq V\text{DD} \leq 5.5V$
D032A			—	_	0.15 VDD	V	$1.8V \le V\text{DD} \le 4.5V$
D033		with Schmitt Trigger buffer	—	_	0.2 Vdd	V	$2.0V \le V\text{DD} \le 5.5V$
D034		MCLR		_	0.2 Vdd	V	
	VIH	Input High Voltage					
		I/O ports:					
D040		with TTL buffer	2.0			V	$4.5V \leq V\text{DD} \leq 5.5V$
D040A			0.25 VDD + 0.8	—	_	V	$1.8V \leq V\text{DD} \leq 4.5V$
D041		with Schmitt Trigger buffer	0.8 VDD	_	_	V	$2.0V \le VDD \le 5.5V$
D042		MCLR	0.8 VDD	_	_	V	
	lı∟	Input Leakage Current <sup>(2)</sup>	11			l	
D060		I/O ports		± 5	± 125	nA	VSS $\leq$ VPIN $\leq$ VDD, Pin at
							high-impedance @ 85°C
				± 5	± 1000	nA	125°C
D061		MCLR		± 50	± 200	nA	$Vss \le VPIN \le VDD @ 85^{\circ}C$
	IPUR	Weak Pull-up Current					
D070*			25	100	200		VDD = 3.3V, VPIN = VSS
			25	140	300	μA	VDD = 5.0V, VPIN = VSS
	Vol	Output Low Voltage				1	ſ
D080		I/O ports					IOL = 8mA, VDD = 5V
			—	—	0.6	V	IOL = 6mA, VDD = 3.3V
	1/211	Outrast Illingh Malta and					IOL = 1.8mA, VDD = 1.8V
Daga	Vон	Output High Voltage	<u> </u>				
D090		I/O ports	Vdd - 0.7			v	ІОН = 3.5mA, VDD = 5V ІОН = 3mA, VDD = 3.3V
			VDD - 0.7	_	-	v	10H = 3MA, VDD = 3.3V 10H = 1mA, VDD = 1.8V

These parameters are characterized but not tested.

Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are t not tested.

Note 1: Negative current is defined as current sourced by the pin.

2: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

Stanua	iu Opera	ating Conditions (unless otherwis	e stateu)	1	1	1	1
Param No.	Sym.	Characteristic	Min.	Typ†	Max.	Units	Conditions
		Program Memory Programming Specifications					
D110	VIHH	Voltage on MCLR/VPP pin	8.0	_	9.0	V	(Note 2)
D111	IDDP	Supply Current during Programming	—	_	10	mA	
D112		VDD for Bulk Erase	2.7	—	V <sub>DD</sub> max.	V	
D113	VPEW	VDD for Write or Row Erase	Vdd min.	—	V <sub>DD</sub> max.	V	
D114	IPPPGM	Current on MCLR/VPP during Erase/Write	_	—	1.0	mA	
D115	IDDPGM	Current on VDD during Erase/Write	_		5.0	mA	
		Program Flash Memory					
D121	Eр	Cell Endurance	10K			E/W	-40°C to +85°C (Note 1)
D122	Vpr	VDD for Read	Vdd min.	—	V <sub>DD</sub> max.	V	
D123	Tiw	Self-timed Write Cycle Time	_	2	2.5	ms	
D124	TRETD	Characteristic Retention	40	—	_	Year	Provided no other specifications are violated
D125	EHEFC	High-Endurance Flash Cell	100K	—	—	E/W	$0^{\circ}C \leq TA \leq 60,$ lower byte last 128 addresses

### Standard Operating Conditions (unless otherwise stated)

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Self-write and Block Erase.

**2:** Required only if single-supply programming is disabled.

#### 24.4 Thermal Considerations

#### Standard Operating Conditions (unless otherwise stated)

	a operating	Conditions (diffess other wise stated)			
Param No.	Sym.	Characteristic	Тур.	Units	Conditions
TH01	θJA	Thermal Resistance Junction to Ambient	60	°C/W	6-pin SOT-23 package
			80	°C/W	8-pin PDIP package
			90	°C/W	8-pin DFN package
TH02	θJC	Thermal Resistance Junction to Case	31.4	°C/W	6-pin SOT-23 package
			24	°C/W	8-pin PDIP package
			24	°C/W	8-pin DFN package
TH03	TJMAX	Maximum Junction Temperature	150	°C	
TH04	PD	Power Dissipation	_	W	PD = PINTERNAL + PI/O
TH05	PINTERNAL	Internal Power Dissipation	_	W	PINTERNAL = IDD x VDD <sup>(1)</sup>
TH06	Pi/o	I/O Power Dissipation	_	W	$PI/O = \Sigma (IOL * VOL) + \Sigma (IOH * (VDD - VOH))$
TH07	Pder	Derated Power	_	W	Pder = PDmax (Τj - Τa)/θja <sup>(2)</sup>

Note 1: IDD is current to run the chip alone without driving any load on the output pins.

2: TA = Ambient Temperature

**3:** T<sub>J</sub> = Junction Temperature

### 24.5 AC Characteristics

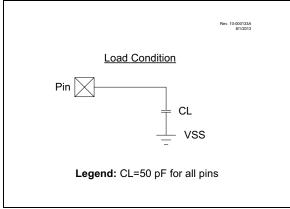
Timing Parameter Symbology has been created with one of the following formats:

1. TppS2ppS

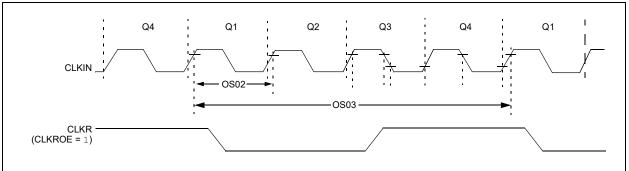
2. TppS

2. Tpp3					
т					
F	Frequency	Т	Time		
Lowercase letters (pp) and their meanings:					
рр					
сс	CCP1	OSC	CLKIN		
ck	CLKR	rd	RD		
cs	CS	rw	RD or WR		
di	SDI	sc	SCK		
do	SDO	SS	SS		
dt	Data in	tO	TOCKI		
io	I/O PORT	t1	T1CKI		
mc	MCLR	wr	WR		
Upperc	case letters and their meanings:				
S					
F	Fall	Р	Period		
н	High	R	Rise		
I	Invalid (High-impedance)	V	Valid		
L	Low	Z	High-impedance		

#### FIGURE 24-4: LOAD CONDITIONS



#### FIGURE 24-5: CLOCK TIMING



#### TABLE 24-6: CLOCK OSCILLATOR TIMING REQUIREMENTS

Standard	Standard Operating Conditions (unless otherwise stated)									
Param No.Sym.CharacteristicMin.Typ†Max.UnitsConditions										
OS01	Fosc	External CLKIN Frequency <sup>(1)</sup>	DC	_	20	MHz	EC mode			
OS02	Tosc	External CLKIN Period <sup>(1)</sup>	31.25	—	×	ns	EC Oscillator mode			
OS03	Тсү	Instruction Cycle Time <sup>(1)</sup>	200	TCY	DC	ns	Tcy = 4/Fosc			

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (TcY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to CLKIN pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

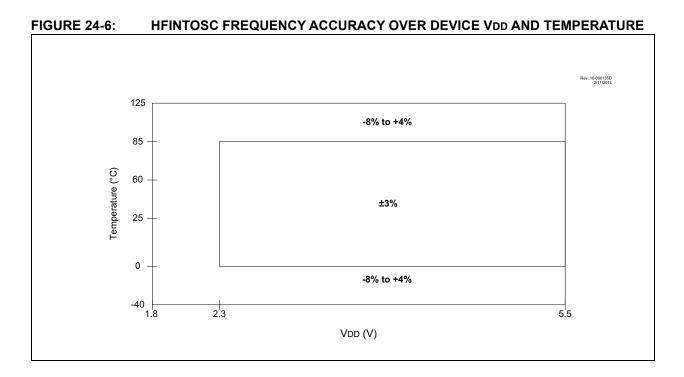
#### TABLE 24-7: OSCILLATOR PARAMETERS

Standar	Standard Operating Conditions (unless otherwise stated)									
Param No.	Sym.	Characteristic	Freq. Tolerance	Min.	Тур†	Max.	Units	Conditions		
OS08	HFosc	Internal Calibrated HFINTOSC Frequency <sup>(1)</sup>	±3% -8 to +4%	_	16.0 16.0	—	MHz MHz	$\begin{array}{l} 0^{\circ}C \leq TA \leq +85^{\circ}C, \ VDD \geq 2.3V \\ -40^{\circ}C \leq TA \leq 125^{\circ}C \end{array}$		
OS09	LFosc	Internal LFINTOSC Frequency	±25%	_	31	_	kHz			
OS10*	TWARM	HFINTOSC Wake-up from Sleep Start-up Time	—	—	5	8	μS			

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** To ensure these oscillator frequency tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1  $\mu$ F and 0.01  $\mu$ F values in parallel are recommended.





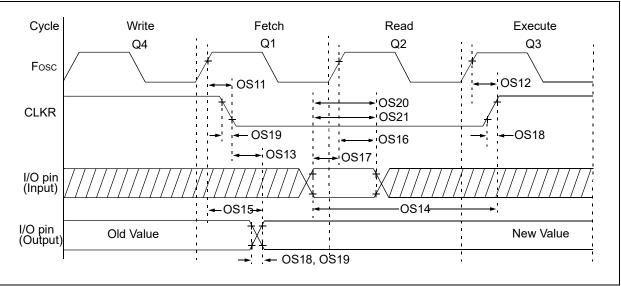


TABLE 24-8: CL	KR AND I/O T	<b>FIMING PARAMETERS</b>
----------------	--------------	--------------------------

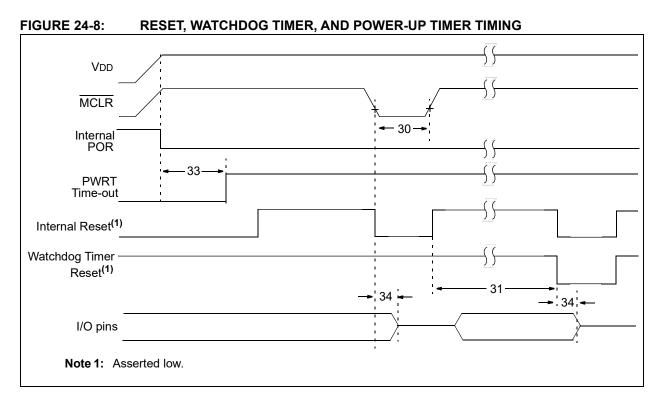
Standar	d Operating	Conditions (unless otherwise stated)					_
Param. No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions
OS11	TosH2ckL	Fosc↑ to CLKOUT↓ <sup>(1)</sup>	—	_	70	ns	$3.3V \leq V\text{DD} \leq 5.0V$
OS12	TosH2ckH	Fosc↑ to CLKOUT↑ <sup>(1)</sup>	_	—	72	ns	$3.3V \le V\text{DD} \le 5.0V$
OS13	TckL2ioV	CLKOUT↓ to Port out valid <sup>(1)</sup>	_	—	20	ns	
OS14	TioV2ckH	Port input valid before CLKOUT <sup>(1)</sup>	Tosc + 200 ns	—	_	ns	
OS15	TosH2ioV	Fosc↑ (Q1 cycle) to Port out valid	—	50	70*	ns	$3.3V \le V\text{DD} \le 5.0V$
OS16	TosH2iol	Fosc↑ (Q2 cycle) to Port input invalid (I/O in setup time)	50	_	_	ns	$3.3V \le V\text{DD} \le 5.0V$
OS17	TioV2osH	Port input valid to Fosc↑ (Q2 cycle) (I/O in setup time)	20	—		ns	
OS18*	TioR	Port output rise time		40 15	72 32	ns	$\begin{array}{l} VDD = 1.8V \\ 3.3V \leq VDD \leq 5.0V \end{array}$
OS19*	TioF	Port output fall time	—	28 15	55 30	ns	$VDD = 1.8V$ $3.3V \le VDD \le 5.0V$
OS20*	Tinp	INT pin input high or low time	25	—	—	ns	
OS21*	Tioc	Interrupt-on-change new input level time	25	—	_	ns	

	-				
Standard	Operating	Conditions	(unless	otherwise stat	ed)

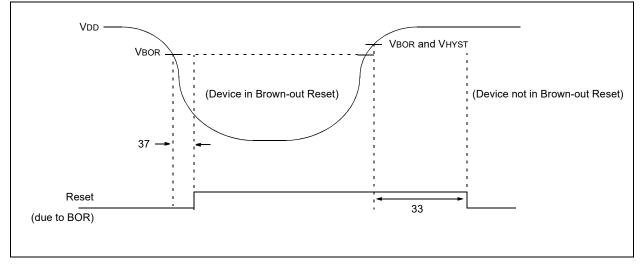
\* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated.

**Note 1:** Measurements are taken in EXTRC mode where CLKOUT output is 4 x Tosc.



#### FIGURE 24-9: BROWN-OUT RESET TIMING AND CHARACTERISTICS



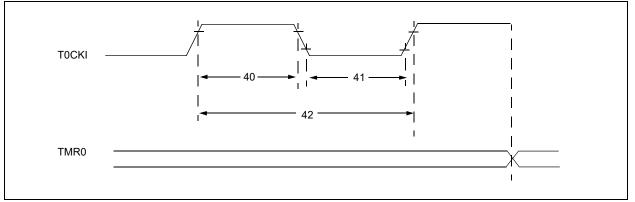
### TABLE 24-9:RESET, WATCHDOG TIMER, POWER-UP TIMER AND BROWN-OUT RESET<br/>PARAMETERS

Standa	rd Opera	ting Conditions (unless otherwise sta	ated)				
Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions
30	ТмсL	MCLR Pulse Width (low)	2 5	_	_	μs μs	VDD = 3.3-5V, -40°C to +85°C VDD = 3.3-5V
31	TWDTLP	Low-Power Watchdog Timer Time-out Period	10	16	27	ms	VDD = 3.3V-5V 1:16 Prescaler used
33*	TPWRT	Power-up Timer Period, PWRTE = 0	40	64	140	ms	
34*	Tioz	I/O high-impedance from MCLR Low or Watchdog Timer Reset		—	2.0	μS	
35	VBOR	Brown-out Reset Voltage <sup>(1)</sup>	2.55	2.70	2.85	V	BORV = 0
			2.30 1.80	2.40 1.90	2.55 2.05	V V	BORV = 1 (PIC10F320/322) BORV = 1 (PIC10LF320/322)
36*	VHYST	Brown-out Reset Hysteresis	0	25	50	mV	-40°C to +85°C
37*	TBORDC	Brown-out Reset DC Response Time	1	3	5	μs	$VDD \leq VBOR$
38	Vlpbor	Low-Power Brown-Out Reset Voltage	1.8	2.1	2.5	V	LPBOR = 1

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

#### FIGURE 24-10: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS



#### TABLE 24-10: TIMER0 EXTERNAL CLOCK REQUIREMENTS

Standard Operating Conditions (unless otherwise stated)									
Param No.	Sym.	Characteris	tic	Min.	Тур†	Max.	Units	Conditions	
40*	Тт0Н	T0CKI High Pulse Width	No Prescaler	0.5 Tcy + 20			ns		
			With Prescaler	10			ns		
41*	TT0L	T0CKI Low Pulse Width	No Prescaler	0.5 Tcy + 20			ns		
			With Prescaler	10			ns		
42*	TT0P	T0CKI Period	•	Greater of:		_	ns	N = prescale value	
				20 or <u>Tcy + 40</u> N				(2, 4,, 256)	

These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: To ensure these voltage tolerances, VDD and VSS must be capacitively decoupled as close to the device as possible. 0.1  $\mu$ F and 0.01  $\mu$ F values in parallel are recommended.

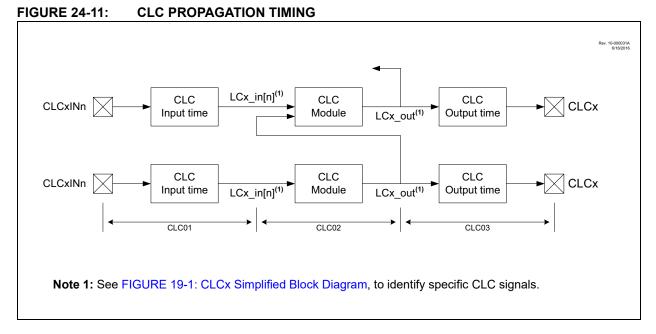


TABLE 24-11:	CONFIGURATION LOGIC CELL	(CLC) CHARACTERISTICS
--------------	--------------------------	-----------------------

Standard Operating Conditions (unless otherwise stated)									
Param. No.	Sym.	Characteristic		Тур†	Max.	Units	Conditions		
CLC01*	TCLCIN	CLC input time		7	_	ns			
CLC02*	TCLC	CLC module input to output propagation time	_	24 12		ns ns	VDD = 1.8V VDD > 3.6V		
CLC03*	TCLCOUT	CLC output time Rise Time		OS18		—	(Note 1)		
		Fall Time		OS19		_	(Note 1)		
CLC04*	FCLCMAX	CLC maximum switching frequency		45	_	MHz			

\* These parameters are characterized but not tested.

Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance t only and are not tested.

Note 1:See Table 24-8 for OS18 and OS19 rise and fall times.

#### TABLE 24-12: A/D CONVERTER (ADC) CHARACTERISTICS:

### Standard Operating Conditions (unless otherwise stated)

Standard Operating Conditions (unless otherwise stated)											
Param No.	Sym.	Characteristic	Min.	Тур†	Max.	Units	Conditions				
AD01	NR	Resolution	—	_	8	bit					
AD02	EIL	Integral Error	_		±1.7	LSb	VREF = 3.0V				
AD03	Edl	Differential Error	_		±1	LSb	No missing codes VREF = 3.0V				
AD04	EOFF	Offset Error		_	±2.5	LSb	VREF = 3.0V				
AD05	Egn	Gain Error		_	±2.0	LSb	VREF = 3.0V				
AD06	Vref	Reference Voltage	1.8	_	Vdd	V	VREF = (VREF+ minus VREF-)				
AD07	VAIN	Full-Scale Range	Vss	_	VREF	V					
AD08	ZAIN	Recommended Impedance of Analog Voltage Source			10	kΩ	Can go higher if external 0.01µF capacitor is present on input pin.				

\* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Total Absolute Error includes integral, differential, offset and gain errors.

2: The A/D conversion result never decreases with an increase in the input voltage and has no missing codes.

**3:** When ADC is off, it will not consume any current other than leakage current. The power-down current specification includes any such leakage from the ADC module.

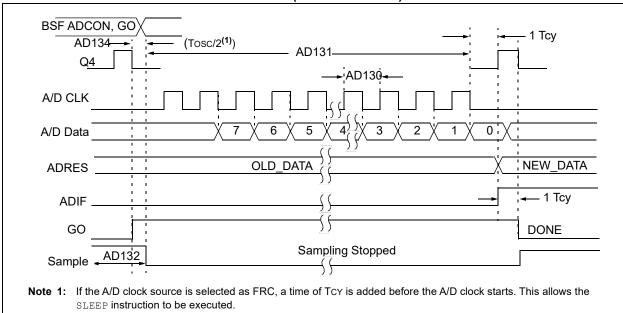
#### TABLE 24-13: A/D CONVERSION REQUIREMENTS

Standard Operating Conditions (unless otherwise stated)										
Param No.	Sym.	. Characteristic Min. Typ†		Тур†	Max.	Units	Conditions			
AD130*	TAD	A/D Clock Period	1.0	—	6.0	μs	Tosc-based			
		A/D Internal FRC Oscillator Period	1.0	1.6	6.0	μs	ADCS<1:0> = 11 (ADRC mode)			
AD131	TCNV	Conversion Time (not including Acquisition Time) <sup>(1)</sup>	_	9.5	_	TAD	Set GO/DONE bit to conversion complete			
AD132*	TACQ	Acquisition Time		5.0		μs				
AD133*	THCD	Holding Capacitor Disconnect Time	_	1/2 Tad	_		Fosc-based			
			—	1/2 TAD + 1TCY	_		ADCS<2:0> = x11 (ADC FRC mode)			

\* These parameters are characterized but not tested.

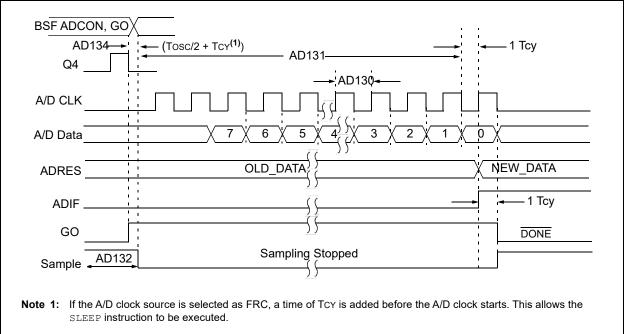
† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The ADRES register may be read on the following TCY cycle.



#### FIGURE 24-12: A/D CONVERSION TIMING (NORMAL MODE)

#### FIGURE 24-13: A/D CONVERSION TIMING (SLEEP MODE)



### 25.0 DC AND AC CHARACTERISTICS GRAPHS AND CHARTS

Graphs and charts are not available at this time.

### 26.0 DEVELOPMENT SUPPORT

The PIC<sup>®</sup> microcontrollers (MCU) and dsPIC<sup>®</sup> digital signal controllers (DSC) are supported with a full range of software and hardware development tools:

- Integrated Development Environment
- MPLAB<sup>®</sup> X IDE Software
- Compilers/Assemblers/Linkers
  - MPLAB XC Compiler
  - MPASM<sup>™</sup> Assembler
  - MPLINK<sup>™</sup> Object Linker/ MPLIB<sup>™</sup> Object Librarian
  - MPLAB Assembler/Linker/Librarian for Various Device Families
- · Simulators
  - MPLAB X SIM Software Simulator
- · Emulators
  - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers/Programmers
  - MPLAB ICD 3
  - PICkit™ 3
- Device Programmers
  - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards, Evaluation Kits and Starter Kits
- Third-party development tools

#### 26.1 MPLAB X Integrated Development Environment Software

The MPLAB X IDE is a single, unified graphical user interface for Microchip and third-party software, and hardware development tool that runs on Windows<sup>®</sup>, Linux and Mac OS<sup>®</sup> X. Based on the NetBeans IDE, MPLAB X IDE is an entirely new IDE with a host of free software components and plug-ins for high-performance application development and debugging. Moving between tools and upgrading from software simulators to hardware debugging and programming tools is simple with the seamless user interface.

With complete project management, visual call graphs, a configurable watch window and a feature-rich editor that includes code completion and context menus, MPLAB X IDE is flexible and friendly enough for new users. With the ability to support multiple tools on multiple projects with simultaneous debugging, MPLAB X IDE is also suitable for the needs of experienced users.

Feature-Rich Editor:

- Color syntax highlighting
- Smart code completion makes suggestions and provides hints as you type
- Automatic code formatting based on user-defined rules
- · Live parsing

User-Friendly, Customizable Interface:

- Fully customizable interface: toolbars, toolbar buttons, windows, window placement, etc.
- · Call graph window
- Project-Based Workspaces:
- Multiple projects
- Multiple tools
- · Multiple configurations
- · Simultaneous debugging sessions

File History and Bug Tracking:

- · Local file history feature
- Built-in support for Bugzilla issue tracker

#### 26.2 MPLAB XC Compilers

The MPLAB XC Compilers are complete ANSI C compilers for all of Microchip's 8, 16, and 32-bit MCU and DSC devices. These compilers provide powerful integration capabilities, superior code optimization and ease of use. MPLAB XC Compilers run on Windows, Linux or MAC OS X.

For easy source level debugging, the compilers provide debug information that is optimized to the MPLAB X IDE.

The free MPLAB XC Compiler editions support all devices and commands, with no time or memory restrictions, and offer sufficient code optimization for most applications.

MPLAB XC Compilers include an assembler, linker and utilities. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. MPLAB XC Compiler uses the assembler to produce its object file. Notable features of the assembler include:

- · Support for the entire device instruction set
- Support for fixed-point and floating-point data
- Command-line interface
- · Rich directive set
- Flexible macro language
- · MPLAB X IDE compatibility

#### 26.3 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel<sup>®</sup> standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code, and COFF files for debugging.

The MPASM Assembler features include:

- Integration into MPLAB X IDE projects
- User-defined macros to streamline assembly code
- Conditional assembly for multipurpose source files
- Directives that allow complete control over the assembly process

#### 26.4 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

#### 26.5 MPLAB Assembler, Linker and Librarian for Various Device Families

MPLAB Assembler produces relocatable machine code from symbolic assembly language for PIC24, PIC32 and dsPIC DSC devices. MPLAB XC Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- · Support for the entire device instruction set
- · Support for fixed-point and floating-point data
- Command-line interface
- · Rich directive set
- Flexible macro language
- · MPLAB X IDE compatibility

#### 26.6 MPLAB X SIM Software Simulator

The MPLAB X SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB X SIM Software Simulator fully supports symbolic debugging using the MPLAB XC Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

#### 26.7 MPLAB REAL ICE In-Circuit Emulator System

The MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs all 8, 16 and 32-bit MCU, and DSC devices with the easy-to-use, powerful graphical user interface of the MPLAB X IDE.

The emulator is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with in-circuit debugger systems (RJ-11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradable through future firmware downloads in MPLAB X IDE. MPLAB REAL ICE offers significant advantages over competitive emulators including full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, logic probes, a ruggedized probe interface and long (up to three meters) interconnection cables.

#### 26.8 MPLAB ICD 3 In-Circuit Debugger System

The MPLAB ICD 3 In-Circuit Debugger System is Microchip's most cost-effective, high-speed hardware debugger/programmer for Microchip Flash DSC and MCU devices. It debugs and programs PIC Flash microcontrollers and dsPIC DSCs with the powerful, yet easy-to-use graphical user interface of the MPLAB IDE.

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer's PC using a highspeed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

#### 26.9 PICkit 3 In-Circuit Debugger/ Programmer

The MPLAB PICkit 3 allows debugging and programming of PIC and dsPIC Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB IDE. The MPLAB PICkit 3 is connected to the design engineer's PC using a fullspeed USB interface and can be connected to the target via a Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the Reset line to implement in-circuit debugging and In-Circuit Serial Programming<sup>™</sup> (ICSP<sup>™</sup>).

#### 26.10 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages, and a modular, detachable socket assembly to support various package types. The ICSP cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices, and incorporates an MMC card for file storage and data applications.

#### 26.11 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM<sup>™</sup> and dsPICDEM<sup>™</sup> demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ<sup>®</sup> security ICs, CAN, IrDA<sup>®</sup>, PowerSmart battery management, SEEVAL<sup>®</sup> evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

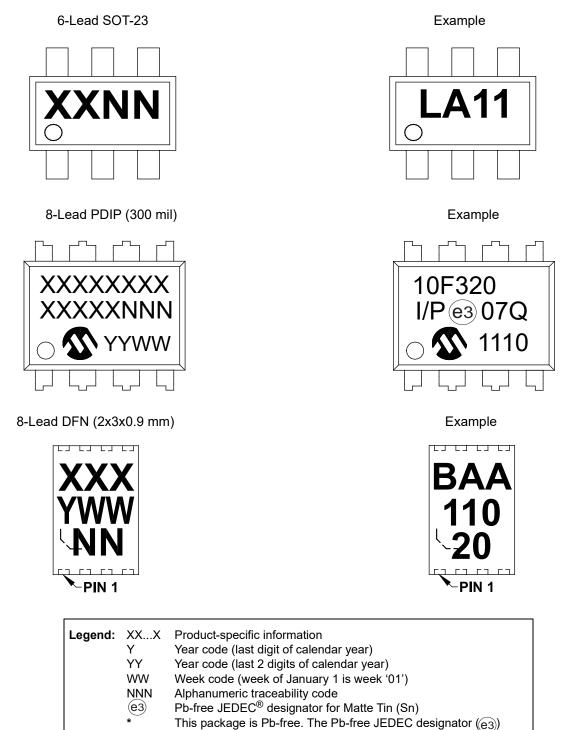
#### 26.12 Third-Party Development Tools

Microchip also offers a great collection of tools from third-party vendors. These tools are carefully selected to offer good value and unique functionality.

- Device Programmers and Gang Programmers from companies, such as SoftLog and CCS
- Software Tools from companies, such as Gimpel and Trace Systems
- Protocol Analyzers from companies, such as Saleae and Total Phase
- Demonstration Boards from companies, such as MikroElektronika, Digilent<sup>®</sup> and Olimex
- Embedded Ethernet Solutions from companies, such as EZ Web Lynx, WIZnet and IPLogika<sup>®</sup>

# 27.0 PACKAGING INFORMATION

# 27.1 Package Marking Information



**Note**: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

can be found on the outer packaging for this package.

# TABLE 27-1:8-LEAD 2x3 DFN (MC) TOP<br/>MARKING

Part Number	Marking
PIC10F322(T)-I/MC	BAA
PIC10F322(T)-E/MC	BAB
PIC10F320(T)-I/MC	BAC
PIC10F320(T)-E/MC	BAD
PIC10LF322(T)-I/MC	BAF
PIC10LF322(T)-E/MC	BAG
PIC10LF320(T)-I/MC	BAH
PIC10LF320(T)-E/MC	BAJ

# TABLE 27-2:6-LEAD SOT-23 (OT)PACKAGE TOP MARKING

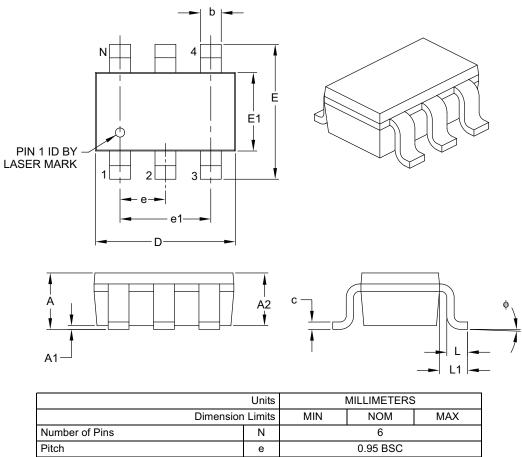
Part Number	Marking
PIC10F322(T)-I/OT	LA/LJ
PIC10F322(T)-E/OT	LB/LK
PIC10F320(T)-I/OT	LC
PIC10F320(T)-E/OT	LD
PIC10LF322(T)-I/OT	LE
PIC10LF322(T)-E/OT	LF
PIC10LF320(T)-I/OT	LG
PIC10LF320(T)-E/OT	LH

# 27.2 Package Details

The following sections give the technical details of the packages.

# 6-Lead Plastic Small Outline Transistor (OT) [SOT-23]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Child				
	Dimension Limits	MIN	NOM	MAX
Number of Pins	N		6	
Pitch	e		0.95 BSC	
Outside Lead Pitch	e1		1.90 BSC	
Overall Height	A	0.90	-	1.45
Molded Package Thickness	A2	0.89	-	1.30
Standoff	A1	0.00	-	0.15
Overall Width	E	2.20	-	3.20
Molded Package Width	E1	1.30	-	1.80
Overall Length	D	2.70	-	3.10
Foot Length	L	0.10	-	0.60
Footprint	L1	0.35	-	0.80
Foot Angle	φ	0°	-	30°
Lead Thickness	С	0.08	_	0.26
Lead Width	b	0.20	_	0.51

#### Notes:

1. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.127 mm per side.

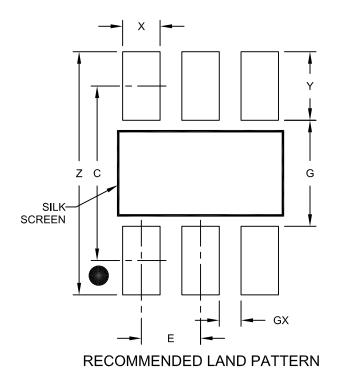
2. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-028B

# 6-Lead Plastic Small Outline Transistor (OT) [SOT-23]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		0.95 BSC	
Contact Pad Spacing	С		2.80	
Contact Pad Width (X6)	Х			0.60
Contact Pad Length (X6)	Y			1.10
Distance Between Pads	G	1.70		
Distance Between Pads	GX	0.35		
Overall Width	Z			3.90

#### Notes:

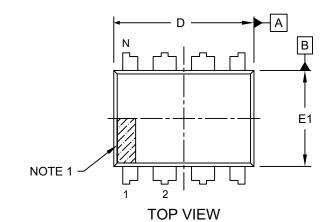
1. Dimensioning and tolerancing per ASME Y14.5M

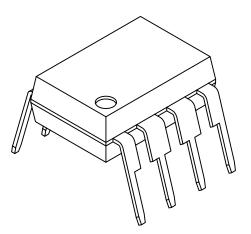
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

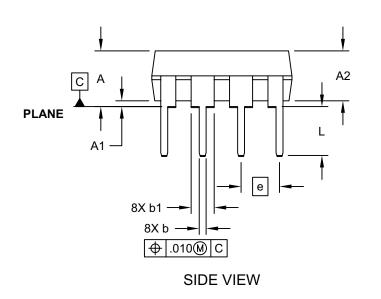
Microchip Technology Drawing No. C04-2028A

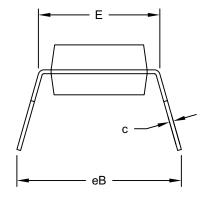
# 8-Lead Plastic Dual In-Line (P) - 300 mil Body [PDIP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging







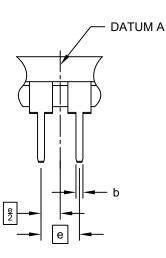


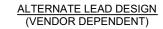
**END VIEW** 

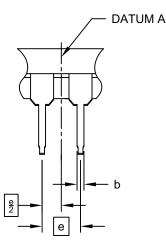
Microchip Technology Drawing No. C04-018D Sheet 1 of 2

# 8-Lead Plastic Dual In-Line (P) - 300 mil Body [PDIP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging







Units			INCHES		
Dimension	Dimension Limits		NOM	MAX	
Number of Pins	N		8		
Pitch	е		.100 BSC		
Top to Seating Plane	A	-	-	.210	
Molded Package Thickness	A2	.115	.130	.195	
Base to Seating Plane	A1	.015	-	-	
Shoulder to Shoulder Width	E	.290	.310	.325	
Molded Package Width	E1	.240	.250	.280	
Overall Length	D	.348	.365	.400	
Tip to Seating Plane	L	.115	.130	.150	
Lead Thickness	С	.008	.010	.015	
Upper Lead Width	b1	.040	.060	.070	
Lower Lead Width	b	.014	.018	.022	
Overall Row Spacing §	eВ	-	-	.430	

Notes:

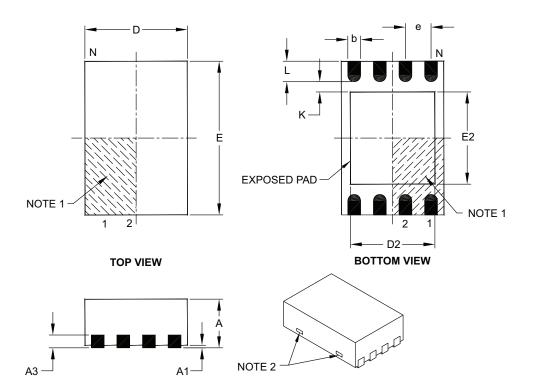
- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-018D Sheet 2 of 2

## 8-Lead Plastic Dual Flat, No Lead Package (MC) – 2x3x0.9 mm Body [DFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units		MILLIMETERS		
	Dimension Limits	MIN	NOM	MAX	
Number of Pins	N		8		
Pitch	е		0.50 BSC		
Overall Height	A	0.80	0.90	1.00	
Standoff	A1	0.00	0.02	0.05	
Contact Thickness	A3	0.20 REF			
Overall Length	D		2.00 BSC		
Overall Width	E		3.00 BSC		
Exposed Pad Length	D2	1.30	-	1.55	
Exposed Pad Width	E2	1.50	-	1.75	
Contact Width	b	0.20	0.25	0.30	
Contact Length	L	0.30	0.40	0.50	
Contact-to-Exposed Pad	К	0.20	-	_	

#### Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

- 2. Package may have one or more exposed tie bars at ends.
- 3. Package is saw singulated.
- 4. Dimensioning and tolerancing per ASME Y14.5M.

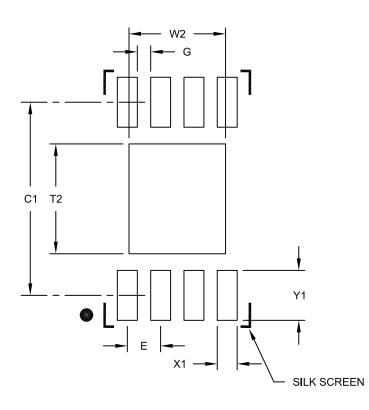
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-123C

8-Lead Plastic Dual Flat, No Lead Package (MC) - 2x3x0.9mm Body [DFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



# RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E	0.50 BSC		
Optional Center Pad Width	W2			1.45
Optional Center Pad Length	T2			1.75
Contact Pad Spacing	C1		2.90	
Contact Pad Width (X8)	X1			0.30
Contact Pad Length (X8)	Y1			0.75
Distance Between Pads	G	0.20		

#### Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2123B

# APPENDIX A: DATA SHEET REVISION HISTORY

# **Revision E (04/2021)**

Updated Product Identification System chapter. Other minor editorial updates.

# **Revision D (11/2015)**

Updated the "eXtreme Low-Power (XLP) Features" section; added "Memory" section. Updated "Family Types" table; Updated Table 2-1, 24-5, 24-7, 24-9, 24-12 and 24-13; Updated Figure 7-1, 24-6 and section 15.2.5; Other minor corrections.

### **Revision C (05/2015)**

Updated Figures 7-1 and 11-1. Update Sections 5.4.1, 24.1, and 24.3. Updated Tables 24-2 and 24-9.

## **Revision B (02/2014)**

Electrical Specifications update and new formats; Minor edits.

## Revision A (07/2011)

Original release.

# THE MICROCHIP WEBSITE

Microchip provides online support via our WWW site at www.microchip.com. This website is used as a means to make files and information easily available to customers. Accessible by using your favorite Internet browser, the website contains the following information:

- Product Support Data sheets and errata, application notes and sample programs, design resources, user's guides and hardware support documents, latest software releases and archived software
- General Technical Support Frequently Asked Questions (FAQ), technical support requests, online discussion groups, Microchip consultant program member listing
- Business of Microchip Product selector and ordering guides, latest Microchip press releases, listing of seminars and events, listings of Microchip sales offices, distributors and factory representatives

# CUSTOMER CHANGE NOTIFICATION SERVICE

Microchip's customer notification service helps keep customers current on Microchip products. Subscribers will receive e-mail notification whenever there are changes, updates, revisions or errata related to a specified product family or development tool of interest.

To register, access the Microchip website at www.microchip.com. Under "Support", click on "Customer Change Notification" and follow the registration instructions.

# **CUSTOMER SUPPORT**

Users of Microchip products can receive assistance through several channels:

- Distributor or Representative
- Local Sales Office
- Field Application Engineer (FAE)
- Technical Support

Customers should contact their distributor, representative or Field Application Engineer (FAE) for support. Local sales offices are also available to help customers. A listing of sales offices and locations is included in the back of this document.

Technical support is available through the website at: http://microchip.com/support

# **PRODUCT IDENTIFICATION SYSTEM**

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	$\begin{array}{cccccccccccccccccccccccccccccccccccc$	Examples:
Device	Tape and Reel Temperature Package Pattern Option Range	a) PIC10LF320T - I/OT Tape and Reel, Industrial temperature, SOT-23 package
Device:	PIC10F320, PIC10LF320, PIC10F322, PIC10LF322	b) PIC10F322 - I/P Industrial temperature PDIP package
Tape and Reel Option:	Blank = Standard packaging (bag, tube, or tray) T = Tape and Reel <sup>(1)</sup>	c) PIC10F322 - E/MC Extended temperature, DFN package
Temperature Range:	I = $-40^{\circ}$ C to $+85^{\circ}$ C (Industrial) E = $-40^{\circ}$ C to $+125^{\circ}$ C (Extended)	
Package:	OT = SOT-23 P = PDIP MC = DFN	<b>Note 1:</b> Tape and Reel identifier only appears in the catalog part number description. This
Pattern:	QTP, SQTP, Code or Special Requirements (blank otherwise)	identifier is used for ordering purposes and is not printed on the device package. Check with your Microchip Sales Office for package availability with the Tape and Reel option.

#### Note the following details of the code protection feature on Microchip devices:

- Microchip products meet the specifications contained in their particular Microchip Data Sheet.
- · Microchip believes that its family of products is secure when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods being used in attempts to breach the code protection features of the Microchip devices. We believe that these methods require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Attempts to breach these code protection features, most likely, cannot be accomplished without violating Microchip's intellectual property rights.
- Microchip is willing to work with any customer who is concerned about the integrity of its code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of its code. Code protection does not
  mean that we are guaranteeing the product is "unbreakable." Code protection is constantly evolving. We at Microchip are
  committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection
  feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or
  other copyrighted work, you may have a right to sue for relief under that Act.

Information contained in this publication is provided for the sole purpose of designing with and using Microchip products. Information regarding device applications and the like is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications.

THIS INFORMATION IS PROVIDED BY MICROCHIP "AS IS". MICROCHIP MAKES NO REPRESENTATIONS OR WAR-RANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION INCLUDING BUT NOT LIMITED TO ANY IMPLIED WARRANTIES OF NON-INFRINGEMENT, MERCHANTABILITY, AND FITNESS FOR A PARTICULAR PURPOSE OR WARRANTIES RELATED TO ITS CONDITION, QUALITY, OR PERFORMANCE.

IN NO EVENT WILL MICROCHIP BE LIABLE FOR ANY INDI-RECT, SPECIAL, PUNITIVE, INCIDENTAL OR CONSEQUEN-TIAL LOSS, DAMAGE, COST OR EXPENSE OF ANY KIND WHATSOEVER RELATED TO THE INFORMATION OR ITS USE, HOWEVER CAUSED, EVEN IF MICROCHIP HAS BEEN ADVISED OF THE POSSIBILITY OR THE DAMAGES ARE FORESEEABLE. TO THE FULLEST EXTENT ALLOWED BY LAW, MICROCHIP'S TOTAL LIABILITY ON ALL CLAIMS IN ANY WAY RELATED TO THE INFORMATION OR ITS USE WILL NOT EXCEED THE AMOUNT OF FEES, IF ANY, THAT YOU HAVE PAID DIRECTLY TO MICROCHIP FOR THE INFORMATION. Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights unless otherwise stated.

# For information regarding Microchip's Quality Management Systems, please visit www.microchip.com/quality.

#### Trademarks

The Microchip name and logo, the Microchip logo, Adaptec, AnyRate, AVR, AVR logo, AVR Freaks, BesTime, BitCloud, chipKIT, chipKIT logo, CryptoMemory, CryptoRF, dsPIC, FlashFlex, flexPWR, HELDO, IGLOO, JukeBlox, KeeLoq, Kleer, LANCheck, LinkMD, maXStylus, maXTouch, MediaLB, megaAVR, Microsemi, Microsemi logo, MOST, MOST logo, MPLAB, OptoLyzer, PackeTime, PIC, picoPower, PICSTART, PIC32 logo, PolarFire, Prochip Designer, QTouch, SAM-BA, SenGenuity, SpyNIC, SST, SST Logo, SuperFlash, Symmetricom, SyncServer, Tachyon, TimeSource, tinyAVR, UNI/O, Vectron, and XMEGA are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

AgileSwitch, APT, ClockWorks, The Embedded Control Solutions Company, EtherSynch, FlashTec, Hyper Speed Control, HyperLight Load, IntelliMOS, Libero, motorBench, mTouch, Powermite 3, Precision Edge, ProASIC, ProASIC Plus, ProASIC Plus logo, Quiet-Wire, SmartFusion, SyncWorld, Temux, TimeCesium, TimeHub, TimePictra, TimeProvider, WinPath, and ZL are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Adjacent Key Suppression, AKS, Analog-for-the-Digital Age, Any Capacitor, AnyIn, AnyOut, Augmented Switching, BlueSky, BodyCom, CodeGuard, CryptoAuthentication, CryptoAutomotive, CryptoCompanion, CryptoController, dsPICDEM, dsPICDEM.net, Dynamic Average Matching, DAM, ECAN, Espresso T1S, EtherGREEN, IdealBridge, In-Circuit Serial Programming, ICSP, INICnet, Intelligent Paralleling, Inter-Chip Connectivity, JitterBlocker, maxCrypto, maxView, memBrain, Mindi, MiWi, MPASM, MPF, MPLAB Certified logo, MPLIB, MPLINK, MultiTRAK, NetDetach, Omniscient Code Generation, PICDEM, PICDEM.net, PICkit, PICtail, PowerSmart, PureSilicon, QMatrix, REAL ICE, Ripple Blocker, RTAX, RTG4, SAM-ICE, Serial Quad I/O, simpleMAP, SimpliPHY, SmartBuffer, SMART-I.S., storClad, SQI, SuperSwitcher, SuperSwitcher II, Switchtec, SynchroPHY, Total Endurance, TSHARC, USBCheck, VariSense, VectorBlox, VeriPHY, ViewSpan, WiperLock, XpressConnect, and ZENA are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries

 $\ensuremath{\mathsf{SQTP}}$  is a service mark of Microchip Technology Incorporated in the U.S.A.

The Adaptec logo, Frequency on Demand, Silicon Storage Technology, and Symmcom are registered trademarks of Microchip Technology Inc. in other countries.

GestIC is a registered trademark of Microchip Technology Germany II GmbH & Co. KG, a subsidiary of Microchip Technology Inc., in other countries.

All other trademarks mentioned herein are property of their respective companies.

 $\ensuremath{\textcircled{\sc 0}}$  2011-2021, Microchip Technology Incorporated, All Rights Reserved.

ISBN: 978-1-5224-7960-4



# Worldwide Sales and Service

#### AMERICAS

**Corporate Office** 2355 West Chandler Blvd. Chandler, AZ 85224-6199 Tel: 480-792-7200 Fax: 480-792-7277 **Technical Support:** http://www.microchip.com/ support

Web Address: www.microchip.com

Atlanta Duluth, GA Tel: 678-957-9614 Fax: 678-957-1455

Austin, TX Tel: 512-257-3370

**Boston** Westborough, MA Tel: 774-760-0087 Fax: 774-760-0088

Chicago Itasca, IL Tel: 630-285-0071 Fax: 630-285-0075

Dallas Addison, TX Tel: 972-818-7423 Fax: 972-818-2924

Detroit Novi, MI Tel: 248-848-4000

Houston, TX Tel: 281-894-5983

Indianapolis Noblesville, IN Tel: 317-773-8323 Fax: 317-773-5453 Tel: 317-536-2380

Los Angeles Mission Viejo, CA Tel: 949-462-9523 Fax: 949-462-9608 Tel: 951-273-7800

Raleigh, NC Tel: 919-844-7510

New York, NY Tel: 631-435-6000

San Jose, CA Tel: 408-735-9110 Tel: 408-436-4270

Canada - Toronto Tel: 905-695-1980 Fax: 905-695-2078

#### ASIA/PACIFIC

Australia - Sydney Tel: 61-2-9868-6733

China - Beijing Tel: 86-10-8569-7000 China - Chengdu

Tel: 86-28-8665-5511 China - Chongqing Tel: 86-23-8980-9588

China - Dongguan Tel: 86-769-8702-9880

China - Guangzhou Tel: 86-20-8755-8029

China - Hangzhou Tel: 86-571-8792-8115

China - Hong Kong SAR Tel: 852-2943-5100

China - Nanjing Tel: 86-25-8473-2460

China - Qingdao Tel: 86-532-8502-7355

China - Shanghai Tel: 86-21-3326-8000

China - Shenyang Tel: 86-24-2334-2829

China - Shenzhen Tel: 86-755-8864-2200

China - Suzhou Tel: 86-186-6233-1526

China - Wuhan Tel: 86-27-5980-5300

China - Xian Tel: 86-29-8833-7252

China - Xiamen Tel: 86-592-2388138

Tel: 86-756-3210040

#### ASIA/PACIFIC

India - Bangalore Tel: 91-80-3090-4444

India - New Delhi Tel: 91-11-4160-8631 India - Pune

Tel: 91-20-4121-0141 Japan - Osaka

Tel: 81-6-6152-7160

Japan - Tokyo Tel: 81-3-6880- 3770 Korea - Daegu

Tel: 82-53-744-4301 Korea - Seoul

Tel: 82-2-554-7200

Malaysia - Kuala Lumpur Tel: 60-3-7651-7906

Malaysia - Penang Tel: 60-4-227-8870

Philippines - Manila Tel: 63-2-634-9065

Singapore Tel: 65-6334-8870

Taiwan - Hsin Chu

Taiwan - Kaohsiung

Tel: 886-2-2508-8600

Thailand - Bangkok

Vietnam - Ho Chi Minh Tel: 84-28-5448-2100

Fax: 39-0331-466781 Italy - Padova

> Tel: 39-049-7625286 **Netherlands - Drunen**

EUROPE

Austria - Wels

Tel: 43-7242-2244-39

Tel: 45-4485-5910

Fax: 45-4485-2829

Tel: 358-9-4520-820

Tel: 33-1-69-53-63-20

Fax: 33-1-69-30-90-79

Germany - Garching

Tel: 49-2129-3766400

Germany - Heilbronn

Germany - Karlsruhe

Tel: 49-7131-72400

Tel: 49-721-625370

Germany - Munich

Tel: 49-89-627-144-0

Fax: 49-89-627-144-44

Germany - Rosenheim

Tel: 49-8031-354-560

Israel - Ra'anana

Italy - Milan

Tel: 972-9-744-7705

Tel: 39-0331-742611

Tel: 49-8931-9700

Germany - Haan

Finland - Espoo

France - Paris

Fax: 43-7242-2244-393

Denmark - Copenhagen

Tel: 31-416-690399 Fax: 31-416-690340

Norway - Trondheim Tel: 47-7288-4388

Poland - Warsaw Tel: 48-22-3325737

Romania - Bucharest Tel: 40-21-407-87-50

Spain - Madrid Tel: 34-91-708-08-90 Fax: 34-91-708-08-91

Sweden - Gothenberg Tel: 46-31-704-60-40

Sweden - Stockholm Tel: 46-8-5090-4654

**UK - Wokingham** Tel: 44-118-921-5800 Fax: 44-118-921-5820

Tel: 886-3-577-8366

Tel: 886-7-213-7830

Taiwan - Taipei

Tel: 66-2-694-1351

China - Zhuhai

# **Mouser Electronics**

Authorized Distributor

Click to View Pricing, Inventory, Delivery & Lifecycle Information:

# Microchip:

PIC10F320-E/MC PIC10F320-E/OT PIC10F320-E/P PIC10F320-I/MC PIC10F320-I/OT PIC10F320-I/P PIC10F320T-I/MC PIC10F320T-I/OT PIC10F322-E/MC PIC10F322-E/OT PIC10F322-E/P PIC10F322-I/MC PIC10F322-I/OT PIC10F322-I/P PIC10F322T-I/MC PIC10F322T-I/OT PIC10LF320-E/MC PIC10LF320-E/OT PIC10LF320-E/P PIC10LF320-I/MC PIC10LF320-I/OT PIC10LF320-I/P PIC10LF320T-I/MC PIC10LF320T-I/OT PIC10LF322-E/MC PIC10LF322-E/OT PIC10LF322-E/P PIC10LF322-I/MC PIC10LF322-I/OT PIC10LF322-I/P PIC10LF322-E/MC PIC10LF322-E/OT PIC10LF322-E/P PIC10LF322-I/MC PIC10LF322-I/OT PIC10LF322-I/P PIC10LF322T-I/MC PIC10LF322T-I/OT PIC10F320T-E/OT PIC10F322T-E/OT PIC10LF322T-E/OT PIC10LF320T-I/OT PIC10LF322T-I/MC PIC10LF322T-I/OT PIC10F320T-E/OT PIC10F322T-E/OT PIC10LF322T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF320T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF320T-E/OT PIC10LF322T-E/OT PIC10LF320T-E/OT PIC10LF30D-E/OT