

Ultra-Low-Noise, High PSRR, Low-Dropout, 150mA Linear Regulators in µDFN

General Description

The MAX8840/MAX8841/MAX8842 ultra-low-noise, lowdropout (LDO) linear regulators are designed to deliver up to 150mA continuous output current. These regulators achieve a low 120mV dropout for 120mA load current. The MAX8840 uses an advanced architecture to achieve ultra-low output voltage noise of 11 μ V_{RMS} and PSRR of 54dB at 100kHz.

The MAX8841 does not require a bypass capacitor, hence achieving the smallest PC board area. The MAX8842 output voltage can be adjusted with an external divider.

The MAX8840/MAX8841 are preset to a variety of voltages in the 1.5V to 4.5V range. Designed with a p-channel MOSFET series pass transistor, the MAX8840/MAX8841/MAX8842 maintain very low ground current (40µA).

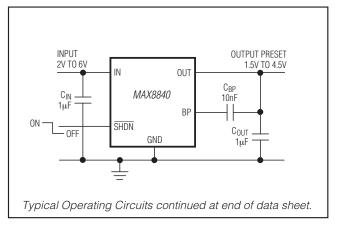
The regulators are designed and optimized to work with low-value, low-cost ceramic capacitors. The MAX8840 requires only 1μ F (typ) of output capacitance for stability with any load. When disabled, current consumption drops to below 1μ A.

The MAX8840/MAX8841/MAX8842 are available in a tiny 1.0mm x 1.5mm x 0.8mm $\mu DFN.$

Applications

Cellular and Cordless Phones PDA and Palmtop Computers Base Stations Bluetooth Portable Radios and Accessories Wireless LANs Digital Cameras Personal Stereos Portable and Battery-Powered Equipment

Typical Operating Circuits



Features

- Space-Saving 1.0mm x 1.5mm x 0.8mm µDFN
- 11µV_{RMS} Output Noise at 100Hz to 100kHz Bandwidth (MAX8840)
- 78dB PSRR at 1kHz (MAX8840)
- 120mV Dropout at 120mA Load
- Stable with 1µF Ceramic Capacitor for Any Load
- Guaranteed 150mA Output
- Only Need Input and Output Capacitors (MAX8841)
- Output Voltages: 1.5V, 1.8V, 2.5V, 2.6V, 2.7V, 2.8V, 2.85V, 3V, 3.3V, 4.5V (MAX8840/MAX8841) and Adjustable (MAX8842)
- ◆ Low 40µA Ground Current
- Excellent Load/Line Transient
- Overcurrent and Thermal Protection

Ordering Information

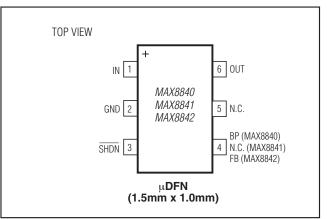
PART*	TEMP RANGE	PIN-PACKAGE
MAX8840ELTxy+T	-40°C to +85°C	6 µDFN-6 1.5mm x 1.0mm
MAX8841ELTxy+T	-40°C to +85°C	6 µDFN-6 1.5mm x 1.0mm
MAX8842ELT+T	-40°C to +85°C	6 µDFN-6 1.5mm x 1.0mm

*xy is the output voltage code (see the Output Voltage Selector Guide). Other versions between 1.5V and 4.5V are available in 100mV increments. Contact factory for other versions.

+Denotes lead-free package.

Output Voltage Selector Guide appears at end of data sheet.

Pin Configuration



For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maximintegrated.com.

ABSOLUTE MAXIMUM RATINGS

IN to GND	-0.3V to +7V
Output Short-Circuit Duration	Infinite
OUT, SHDN to GND	-0.3V to (IN + 0.3V)
FB, BP, N.C. to GND	0.3V to (OUT + 0.3V)
Continuous Power Dissipation (T _A =	
6-Pin µDFN (derate 2.1mW/°C ab	ove +70°C)0.168W

θJA	477°C/W
Operating Temperature Range	
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{IN} = V_{OUT} + 0.5V, T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. $C_{IN} = 1\mu$ F, $C_{OUT} = 1\mu$ F, $C_{BP} = 10$ nF. Typical values are at $+25^{\circ}C$; the MAX8842 is tested with 2.45V output, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CON	MIN	ТҮР	MAX	UNITS			
Input Voltage Range	VIN		2		6	V			
		$I_{OUT} = 1$ mA, $T_A = +25$ °C	-1		+1				
Output Voltage Accuracy		$I_{OUT} = 100 \mu A$ to 80mA, T_A	-2		+2	%			
		$I_{OUT} = 100\mu A$ to 80mA	-3		+3				
Maximum Output Current	IOUT			150			mA		
Current Limit	ILIM	OUT = 90% of nominal value	ue	150	200	300	mA		
		$V_{OUT} \ge 3V$, $I_{OUT} = 80mA$			80	170			
		V _{OUT} ≥ 3V, I _{OUT} = 120mA			120				
Dropout Valtago (Nato 2)		$2.5V \le V_{OUT} < 3V, I_{OUT} = 8$	80mA		90	200			
Dropout Voltage (Note 2)		$2.5V \le V_{OUT} < 3V$, $I_{OUT} = 7$		135		mV			
		$2V \le V_{OUT} < 2.5V, I_{OUT} = 8$		120	250				
		$2V \le V_{OUT} < 2.5V$, $I_{OUT} = T$		180					
Ground Current		$I_{OUT} = 0.05 \text{mA}$		40	90	μA			
Ground Current	lq	$V_{IN} = V_{OUT(nom)} - 0.1V$, IO		220	500	μΑ			
Line Regulation	V _{LNR}	$V_{IN} = (V_{OUT} + 0.5V)$ to 6V,		0.001		%/V			
Load Regulation	VLDR	$I_{OUT} = 1$ mA to 80mA			0.003		%/mA		
Shutdown Supply Current	laurau	$\overline{SHDN} = 0V$	$T_A = +25^{\circ}C$		0.003	1			
Shuldown Supply Current	ISHDN		$T_A = +85^{\circ}C$		0.05		μA		
		f 1kHz lour 10mA	MAX8840		78				
Ripple Rejection		$f = 1 kHz, I_{OUT} = 10 mA$	MAX8841/MAX8842		72				
	PSRR		MAX8840		75		dB		
		$f = 10kHz$, $I_{OUT} = 10mA$	MAX8841/MAX8842		65		UD		
		f = 100kHz lour = 10mA	MAX8840	54]		
		$f = 100$ kHz, $I_{OUT} = 10$ mA	MAX8841/ MAX8842		46				

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ELECTRICAL CHARACTERISTICS (continued)

 $(V_{IN} = V_{OUT} + 0.5V, T_A = -40^{\circ}C$ to +85°C, unless otherwise noted. $C_{IN} = 1\mu$ F, $C_{OUT} = 1\mu$ F, $C_{BP} = 10$ nF. Typical values are at +25°C; the MAX8842 is tested with 2.45V output, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CON	MIN	ТҮР	MAX	UNITS		
		f = 100Hz to 100kHz,	MAX8840		11			
Output Noise Voltage		$I_{LOAD} = 10 mA$	MAX8841/MAX8842		230			
(RMS)		f = 100Hz to 100kHz,	MAX8840		13		μV	
		$I_{LOAD} = 80 mA$	MAX8841/MAX8842		230			
Shutdown Exit Delay		$R_{LOAD} = 50\Omega$ (Note 3)				300	μs	
SHDN Logic-Low Level		$V_{IN} = 2V$ to $6V$			0.4	V		
SHDN Logic-High Level		$V_{IN} = 2V$ to $6V$	1.5			V		
		$V_{IN} = 6V, \overline{SHDN} = 0V \text{ or}$	$T_A = +25^{\circ}C$					
SHDN Input Bias Current		6V	$T_A = +85^{\circ}C$		0.01		μA	
FB Input Bias Current			$T_A = +25^{\circ}C$		0.006	0.1		
(MAX8842)		$V_{\rm IN} = 6V, V_{\rm FB} = 1.3V$		0.01		μΑ		
Thermal Shutdown					160		°C	
Thermal-Shutdown Hysteresis					10		°C	

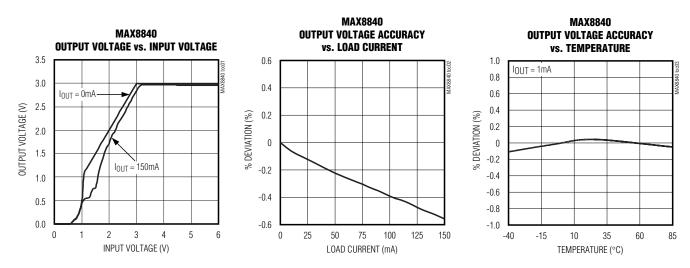
Note 1: Limits are 100% tested at +25°C. Limits over operating temperature range are guaranteed by design.

Note 2: Dropout is defined as $V_{IN} - V_{OUT}$ when V_{OUT} is 100mV below the value of V_{OUT} for $V_{IN} = V_{OUT} + 0.5V$.

Note 3: Time needed for V_{OUT} to reach 90% of final value.

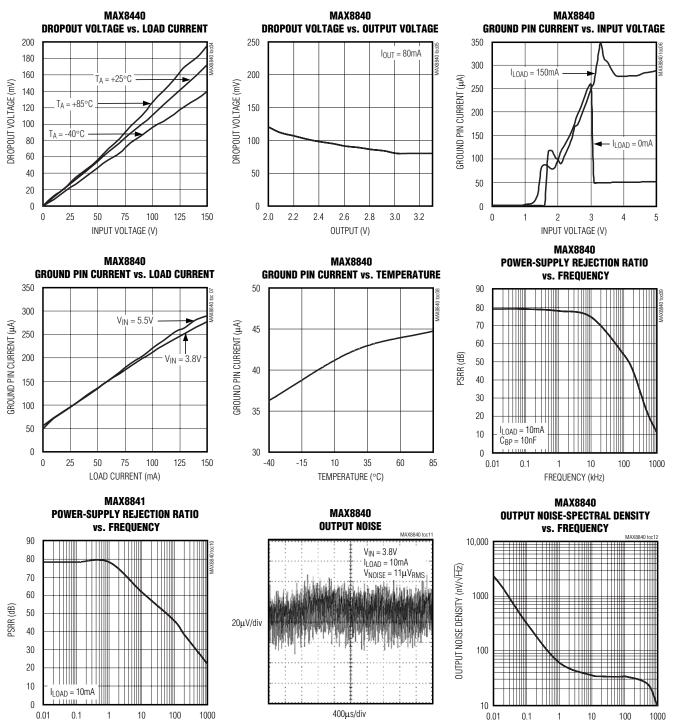
Typical Operating Characteristics

(VIN = VOUT + 0.5V, CIN = 1 μ F, COUT = 1 μ F, CBP = 10nF, TA = +25°C, unless otherwise noted.)



_Typical Operating Characteristics (continued)

 $(V_{IN} = V_{OUT} + 0.5V, C_{IN} = 1\mu F, C_{OUT} = 1\mu F, C_{BP} = 10nF, T_A = +25^{\circ}C$, unless otherwise noted.)



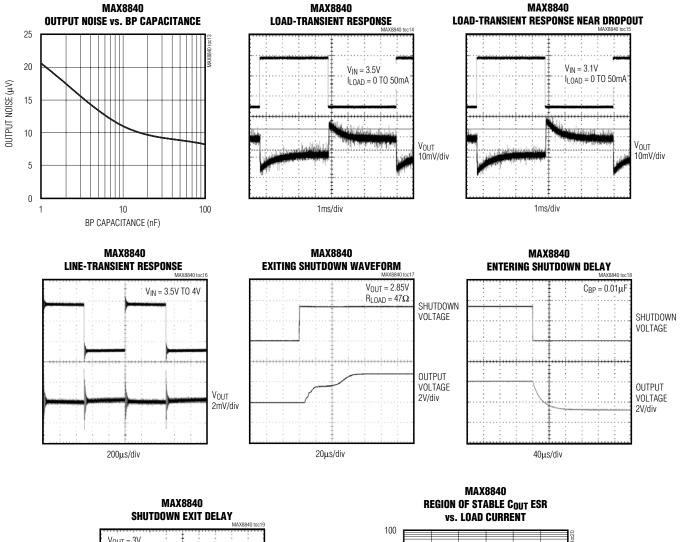
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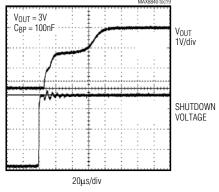
FREQUENCY (kHz)

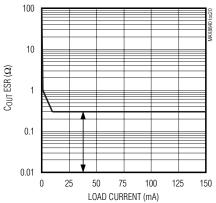
FREQUENCY (kHz)

_Typical Operating Characteristics (continued)

 $(V_{IN} = V_{OUT} + 0.5V, C_{IN} = 1\mu F, C_{OUT} = 1\mu F, C_{BP} = 10nF, T_A = +25^{\circ}C$, unless otherwise noted.)







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Pin Description

	PIN		NAME	FUNCTION			
MAX8840	MAX8841	MAX8842		FUNCTION			
1	1	1	IN	Unregulated Input Supply			
2	2	2	GND Ground				
3	3	3	SHDN Shutdown. Pull low to disable the regulator.				
4	_		BP	Noise Bypass for Low-Noise Operation. Connect a 10nF capacitor from BP to OUT. It is shorted to OUT in shutdown mode.			
_	—	4	FB	Adjustable Output Feedback Point			
5	4, 5	5	N.C.	No Connection. Not internally connected.			
6	6	6	OUT	Regulated Output Voltage. Bypass with a capacitor to GND. See the <i>Capacitor Selection and Regulator Stability</i> section.			

Detailed Description

The MAX8840/MAX8841/MAX8842 are ultra-low-noise, low-dropout, low-quiescent current linear regulators designed for space-restricted applications. The parts are available with preset output voltages ranging from 1.5V to 4.5V. These devices can supply loads up to 150mA. As shown in the *Functional Diagram*, the MAX8840/MAX8841 consist of an innovative bandgap core and noise bypass circuit, error amplifier, p-channel pass transistor, and internal feedback voltagedivider. The MAX8842 allows for adjustable output with an external feedback network.

The 1.225V bandgap reference is connected to the error amplifier's inverting input. The error amplifier compares this reference with the feedback voltage and amplifies the difference. If the feedback voltage is lower than the reference voltage, the pass-transistor gate is pulled low. This allows more current to pass to the output and increases the output voltage. If the feedback voltage is pulled high, allowing less current to pass to the output. The output voltage is fed back through an internal resistor voltage-divider connected to the OUT pin.

An external bypass capacitor connected to BP (MAX8840) reduces noise at the output. Additional blocks include a current limiter, thermal sensor, and shutdown logic.

Internal p-Channel Pass Transistor

The MAX8840/MAX8841/MAX8842 feature a 1 Ω (typ) p-channel MOSFET pass transistor. This provides several advantages over similar designs using a pnp pass transistor, including longer battery life. The p-channel MOSFET requires no base drive, which considerably

reduces quiescent current. pnp-based regulators waste considerable current in dropout when the pass transistor saturates. They also use high base-drive current under heavy loads. The MAX8840/MAX8841/MAX8842 do not suffer from these problems and consume only 40µA of quiescent current in light load and 220µA in dropout (see the *Typical Operating Characteristics*).

Output Voltage Selection

The MAX8840/MAX8841 are supplied with factory-set output voltages from 1.5V to 4.5V, in 100mV increments (see the *Ordering Information*). The MAX8842 features a user-adjustable output through an external feedback network (see the *Typical Operating Circuits*).

To set the output of the MAX8842, use the following equation:

$$R1 = R2 X \left(\frac{V_{OUT}}{V_{REF}} - 1\right)$$

where R2 is chosen to be less than 240k Ω and VREF = 1.225V. Use 1% or better resistors.

Shutdown

The MAX8840/MAX8841/MAX8842 feature a low-power shutdown mode that reduces quiescent current less than 1µA. Driving SHDN low disables the voltage reference, error amplifier, gate-drive circuitry, and pass transistor (see the *Functional Diagram*), and the device output enters a high-impedance state. Connect SHDN to IN for normal operation.

Current Limit

The MAX8840/MAX8841/MAX8842 include a current limiter, which monitors and controls the pass transistor's gate voltage, limiting the output current to 200mA.

For design purposes, consider the current limit to be 150mA (min) to 300mA (max). The output can be shorted to ground for an indefinite amount of time without damaging the part.

Thermal-Overload Protection

Thermal-overload protection limits total power dissipation in the MAX8840/MAX8841/MAX8842. When the junction temperature exceeds $T_J = +160^{\circ}$ C, the thermal sensor signals the shutdown logic, turning off the pass transistor and allowing the IC to cool down. The thermal sensor turns the pass transistor on again after the IC's junction temperature drops by 10°C, resulting in a pulsed output during continuous thermal-overload conditions.

Thermal-overload protection is designed to protect the MAX8840/MAX8841/MAX8842 in the event of a fault condition. For continual operation, do not exceed the absolute maximum junction temperature rating of $T_J = +150^{\circ}C$.

Operating Region and Power Dissipation The MAX8840/MAX8841/MAX8842 maximum power dissipation depends on the thermal resistance of the case and circuit board, the temperature difference between the die junction and ambient, and the rate of airflow. The power dissipation across the device is:

$$P = I_{OUT} (V_{IN} - V_{OUT})$$

The maximum power dissipation is:

$$P_{MAX} = (T_J - T_A) / (\theta_{JC} + \theta_{CA})$$

where T_J - T_A is the temperature difference between the MAX8840/MAX8841/MAX8842 die junction and the surrounding air, θ_{JC} is the thermal resistance of the package, and θ_{CA} is the thermal resistance through the PC board, copper traces, and other materials to the surrounding air.

The GND pin of the MAX8840/MAX8841/MAX8842 performs the dual function of providing an electrical connection to ground and channeling heat away. Connect the GND pin to ground using a large pad or ground plane.

Noise Reduction

For the MAX8840, an external 0.01µF bypass capacitor between BP and OUT with an innovative noise bypass scheme reduces output noises dramatically, exhibiting 11µV_{RMS} of output voltage noise with C_{BP} = 0.01µF and C_{OUT} = 1µF. Startup time is minimized by a power-on circuit that precharges the bypass capacitor.

Applications Information

Capacitor Selection and Regulator Stability

Use a 1 μ F capacitor on the MAX8840/MAX8841/ MAX8842 input and a 1 μ F capacitor on the output. Larger input capacitor values and lower ESRs provide better noise rejection and line-transient response. Reduce output noise and improve load-transient response, stability, and power-supply rejection by using large output capacitors. Note that some ceramic dielectrics exhibit large capacitance and ESR variation with temperature. With dielectrics such as Z5U and Y5V, it may be necessary to use a 2.2 μ F or larger output capacitor to ensure stability at temperatures below -10°C. With X7R or X5R dielectrics, 1 μ F is sufficient at all operating temperatures. A graph of the Region of Stable COUT ESR vs. Load Current is shown in the *Typical Operating Characteristics*.

Use a 0.01μ F bypass capacitor at BP (MAX8840) for low-output voltage noise. The leakage current going into the BP pin should be less than 10nA. Increasing the capacitance slightly decreases the output noise. Values above 0.1μ F and below 0.001μ F are not recommended.

Noise, PSRR, and Transient Response

The MAX8840/MAX8841/MAX8842 are designed to deliver ultra-low noise and high PSRR, as well as low dropout and low quiescent currents in battery-powered systems. The MAX8840 power-supply rejection is 78dB at 1kHz and 54dB at 100kHz. The MAX8841/MAX8842 PSRR is 72dB at 1kHz and 46dB at 100kHz (see the Power-Supply Rejection Ratio vs. Frequency graph in the *Typical Operating Characteristics*).

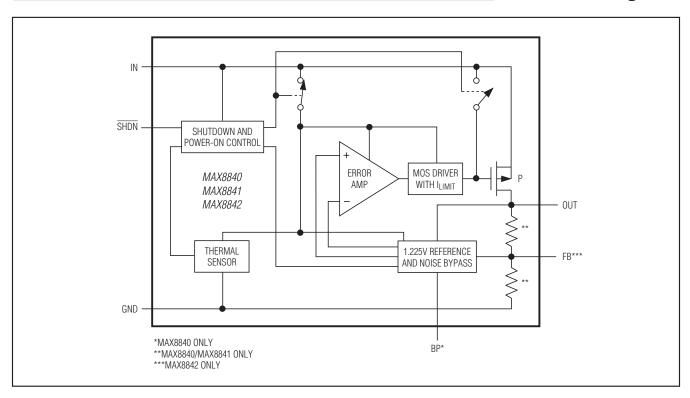
When operating from sources other than batteries, improved supply-noise rejection and transient response can be achieved by increasing the values of the input and output bypass capacitors, and through passive filtering techniques. The *Typical Operating Characteristics* show the MAX8840/MAX8841/MAX8842 line- and loadtransient responses.

Dropout Voltage

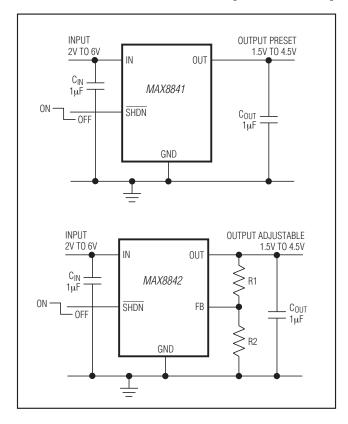
A regulator's minimum dropout voltage determines the lowest usable supply voltage. In battery-powered systems, this determines the useful end-of-life battery voltage. Because the MAX8840/MAX8841/MAX8842 use a p-channel MOSFET pass transistor, their dropout voltage is a function of drain-to-source on-resistance (RDS(ON)) multiplied by the load current (see the *Typical Operating Characteristics*).

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Functional Diagram



Ultra-Low-Noise, High PSRR, Low-Dropout, 150mA Linear Regulators in µDFN



Typical Operating Circuits (continued)

PART	Vout (V)	TOPMARK
MAX8840ELT16+T	1.6	MN
MAX8840ELT18+T	1.8	МО
MAX8840ELT19+T	1.9	LX
MAX8840ELT25+T	2.5	MP
MAX8840ELT26+T	2.6	МК
MAX8840ELT27+T	2.7	LU
MAX8840ELT28+T	2.8	LV
MAX8840ELT29+T	2.85	LW
MAX8840ELT30+T	3	LZ
MAX8840ELT31+T	3.1	ML
MAX8840ELT33+T	3.3	ММ
MAX8841ELT15+T	1.5	MQ
MAX8841ELT18+T	1.8	MR
MAX8841ELT25+T	2.5	MS
MAX8841ELT26+T	2.6	МТ
MAX8841ELT28+T	2.8	MU
MAX8841ELT29+T	2.85	MV
MAX8841ELT89+T	2.9	MW
MAX8841ELT31+T	3.1	МХ
MAX8841ELT33+T	3.3	MY
MAX8841ELT45+T	4.5	MZ
MAX8842ELT+T	Adjustable	NA

Output Voltage Selector Guide

(*Note:* Standard output voltage options, shown in *bold*, are available. Contact the factory for other output voltages between 1.5V and 4.5V. Minimum order quantity is 15,000 units.)

Chip Information

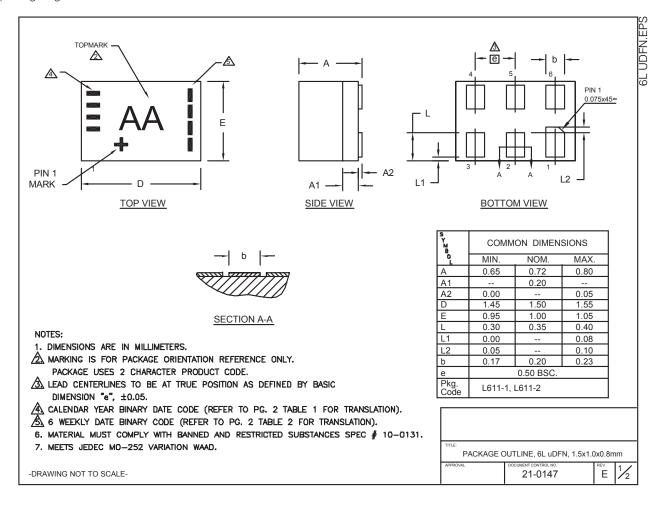
Revision History

Pages changed at Rev 1: 1

TRANSISTOR COUNT: 284 PROCESS: BICMOS

Package Information

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.



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TABLE 1 Translation	Table for	Calenda	r Year Co	ode								
Calendar Year	2005	2006	2007	2008	2009	2010	2011	2012	2013	2014		
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TABLE 2 Translation	Table for	Payweeł	Binary (Coding								
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	Π	Π	Π	X	Π	Π	X	Π	X			
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