

# 8-Bit

# SAA-XC866

8-Bit Single-Chip Microcontroller

Data Sheet V1.5 2010-09

# Microcontrollers

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# Microcontrollers



SAA-XC8 Revision	66 Data Sheet History: 2010-09	V1.5
Previous \	/ersion: V1.4 2010-08	
Page	Subjects (major changes since last revision)	
Changes f	rom V1.4 2010-08 to V1.5 2010-09	
All	3.3 V device variant is added to the data sheet. Parame variant are also added.	eters specific to the

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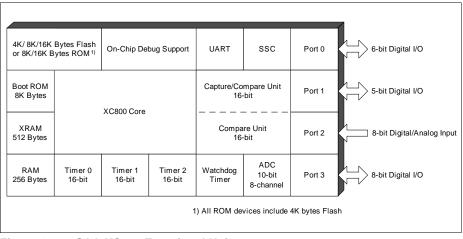
# 8-Bit Single-Chip Microcontroller XC800 Family

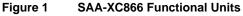
SAA-XC866

### 1 Summary of Features

- High-performance XC800 Core
  - compatible with standard 8051 processor
  - two clocks per machine cycle architecture (for memory access without wait state)
  - two data pointers
- On-chip memory
  - 8 Kbytes of Boot ROM
  - 256 bytes of RAM
  - 512 bytes of XRAM
  - 4/8/16 Kbytes of Flash; or
     8/16 Kbytes of ROM, with additional 4 Kbytes of Flash (includes memory protection strategy)
- I/O port supply at 3.3 V/5.0 V and core logic supply at 2.5 V (generated by embedded voltage regulator)

(further features are on next page)







#### Summary of Features

Features (continued):

- Reset generation
  - Power-On reset
  - Hardware reset
  - Brownout reset for core logic supply
  - Watchdog timer reset
  - Power-Down Wake-up reset
- On-chip OSC and PLL for clock generation
  - PLL loss-of-lock detection
- Power saving modes
  - slow-down mode
  - idle mode
  - power-down mode with wake-up capability via RXD or EXINT0
  - clock gating control to each peripheral
- Programmable 16-bit Watchdog Timer (WDT)
- Four ports
  - 19 pins as digital I/O
  - 8 pins as digital/analog input
- 8-channel, 10-bit ADC
- Three 16-bit timers
  - Timer 0 and Timer 1 (T0 and T1)
  - Timer 2
- Capture/compare unit for PWM signal generation (CCU6)
- Full-duplex serial interface (UART)
- Synchronous serial channel (SSC)
- On-chip debug support
  - 1 Kbyte of monitor ROM (part of the 8-Kbyte Boot ROM)
- 64 bytes of monitor RAM
- PG-TSSOP-38 pin package
- Temperature range  $T_A$ :
  - SAA (-40 to 140 °C)



#### Summary of Features

#### SAA-XC866 Variant Devices

The SAA-XC866 product family features devices with different configurations and program memory sizes, offering cost-effective solution for different application requirements.

The list of SAA-XC866 devices and their differences are summarized in Table 1.

Device Type	Device Name	Power Supply (V)	P-Flash Size (Kbytes)	D-Flash Size (Kbytes)	ROM Size (Kbytes)	LIN BSL Support
Flash <sup>1)</sup>	SAA-XC866L-4FRA	5.0	12	4	-	Yes
	SAA-XC866-4FRA	5.0	12	4	-	No
	SAA-XC866L-2FRA	5.0	4	4	-	Yes
	SAA-XC866-2FRA	5.0	4	4	-	No
	SAA-XC866L-1FRA	5.0	-	4	-	Yes
	SAA-XC866-1FRA	5.0	-	4	-	No
	SAA-XC866L-4FRA	3.3	12	4	-	Yes
ROM	SAA-XC866L-4RRA	5.0	-	4	16	Yes
	SAA-XC866-4RRA	5.0	-	4	16	No
	SAA-XC866L-2RRA	5.0	-	4	8	Yes
	SAA-XC866-2RRA	5.0	-	4	8	No

#### Table 1Device Summary

<sup>1)</sup> The flash memory (P-Flash and D-Flash) can be used for code or data.

#### **Ordering Information**

The ordering code for Infineon Technologies microcontrollers provides an exact reference to the required product. This ordering code identifies:

- The derivative itself, i.e. its function set, the temperature range, and the supply voltage
- the package and the type of delivery

For the available ordering codes for the SAA-XC866, please refer to your responsible sales representative or your local distributor.

As this document refers to all the derivatives, some descriptions may not apply to a specific product. For simplicity all versions are referred to by the term SAA-XC866 throughout this document.





### 2 General Device Information

#### 2.1 Block Diagram

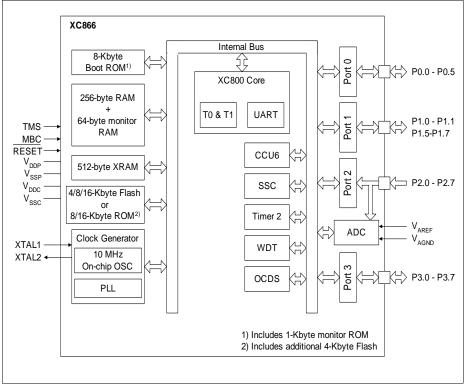


Figure 2 SAA-XC866 Block Diagram



### 2.2 Logic Symbol

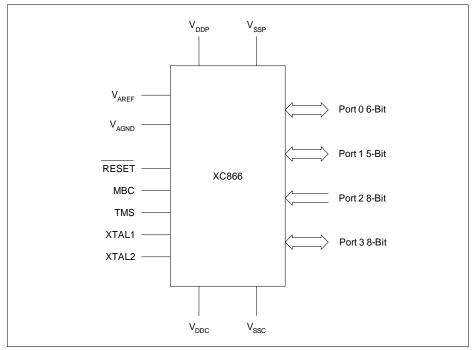


Figure 3 SAA-XC866 Logic Symbol



#### 2.3 Pin Configuration

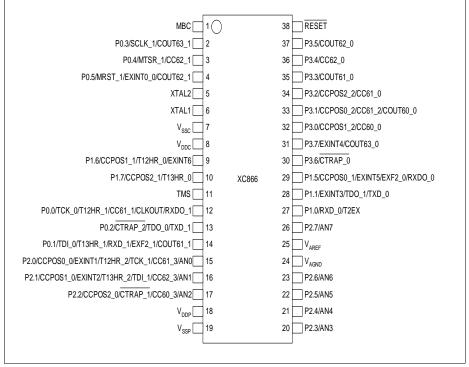


Figure 4 SAA-XC866 Pin Configuration, PG-TSSOP-38 Package (top view)



### 2.4 Pin Definitions and Functions

### Table 2 Pin Definitions and Functions

Symbol	Pin Number	Туре	Reset State	Function	
P0		I/O		port. It can b	-bit bidirectional general purpose I/O be used as alternate functions for the 6, UART, and the SSC.
P0.0	12		Hi-Z	TCK_0 T12HR_1 CC61_1	JTAG Clock Input CCU6 Timer 12 Hardware Run Input Input/Output of Capture/Compare
				CLKOUT RXDO_1	channel 1 Clock Output UART Transmit Data Output
P0.1	14		Hi-Z	TDI_0 T13HR_1	JTAG Serial Data Input CCU6 Timer 13 Hardware Run Input
				RXD_1 COUT61_1	UART Receive Data Input Output of Capture/Compare channel 1
P0.2	13		PU	EXF2_1 CTRAP_2 TDO_0 TXD_1	Timer 2 External Flag Output CCU6 Trap Input JTAG Serial Data Output UART Transmit Data Output/ Clock Output
P0.3	2		Hi-Z	SCK_1 COUT63_1	SSC Clock Input/Output Output of Capture/Compare channel 3
P0.4	3		Hi-Z	MTSR_1 CC62_1	SSC Master Transmit Output/ Slave Receive Input Input/Output of Capture/Compare channel 2
P0.5	4		Hi-Z	MRST_1 EXINT0_0 COUT62_1	SSC Master Receive Input/ Slave Transmit Output External Interrupt Input 0 Output of Capture/Compare channel 2



Symbol	Pin Number	Туре	Reset State	Function	
P1		I/O		port. It can b	bit bidirectional general purpose I/O e used as alternate functions for the 5, UART, and the SSC.
P1.0	27		PU	RXD_0 T2EX	UART Receive Data Input Timer 2 External Trigger Input
P1.1	28		PU	EXINT3 TDO_1 TXD_0	External Interrupt Input 3 JTAG Serial Data Output UART Transmit Data Output/ Clock Output
P1.5	29		PU	CCPOS0_1 EXINT5 EXF2_0 RXDO_0	CCU6 Hall Input 0 External Interrupt Input 5 TImer 2 External Flag Output UART Transmit Data Output
P1.6	9		PU	CCPOS1_1 T12HR_0 EXINT6	CCU6 Hall Input 1 CCU6 Timer 12 Hardware Run Input External Interrupt Input 6
P1.7	10		PU	CCPOS2_1 T13HR_0	CCU6 Hall Input 2 CCU6 Timer 13 Hardware Run Input
					.6 can be used as a software chip t for the SSC.

#### Table 2 Pin Definitions and Functions (cont'd)



#### Symbol Pin Type Reset Function Number State **P2** I Port 2 Port 2 is an 8-bit general purpose input-only port. It can be used as alternate functions for the digital inputs of the JTAG and CCU6. It is also used as the analog inputs for the ADC. P2.0 15 Hi-Z CCPOS0 0 CCU6 Hall Input 0 EXINT1 External Interrupt Input 1 T12HR 2 CCU6 Timer 12 Hardware Run Input TCK\_1 JTAG Clock Input CC61 3 Input of Capture/Compare channel 1 AN0 Analog Input 0 P2.1 16 Hi-Z CCPOS1 0 CCU6 Hall Input 1 EXINT2 External Interrupt Input 2 CCU6 Timer 13 Hardware Run T13HR\_2 Input TDI\_1 JTAG Serial Data Input CC62 3 Input of Capture/Compare channel 2 AN1 Analog Input 1 P2.2 17 Hi-Z CCPOS2\_0 CCU6 Hall Input 2 CTRAP\_1 CCU6 Trap Input CC60 3 Input of Capture/Compare channel 0 AN2 Analog Input 2 AN3 P2.3 20 Hi-Z Analog Input 3 P2.4 21 Hi-Z AN4 Analog Input 4 P<sub>2.5</sub> 22 Hi-Z AN5 Analog Input 5 23 P2.6 Hi-Z AN6 Analog Input 6 P2.7 26 Hi-Z AN7 Analog Input 7



Symbol	Pin	Туре	Reset	Function		
e ye.	Number	.,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,	State			
P3		I			directional general purpose I/O port. It as alternate functions for the CCU6.	
P3.0	32		Hi-Z	CCPOS1_2 CC60_0	CCU6 Hall Input 1 Input/Output of Capture/Compare channel 0	
P3.1	33		Hi-Z	CCPOS0_2 CC61_2	CCU6 Hall Input 0 Input/Output of Capture/Compare channel 1	
				COUT60_0	Output of Capture/Compare channel 0	
P3.2	34		Hi-Z	CCPOS2_2 CC61_0	CCU6 Hall Input 2 Input/Output of Capture/Compare channel 1	
P3.3	35		Hi-Z	COUT61_0	Output of Capture/Compare channel 1	
P3.4	36		Hi-Z	CC62_0	Input/Output of Capture/Compare channel 2	
P3.5	37		Hi-Z	COUT62_0	Output of Capture/Compare channel 2	
P3.6	30		PD	CTRAP_0	CCU6 Trap Input	
P3.7	31		Hi-Z	EXINT4 COUT63_0	External Interrupt Input 4 Output of Capture/Compare channel 3	

#### Table 2 Pin Definitions and Functions (cont'd)



Symbol	Pin Number	Туре	Reset State	Function
V <sub>DDP</sub>	18	-	_	I/O Port Supply (3.3 V/5.0 V) Also used by EVR and analog modules.
V <sub>SSP</sub>	19	-	-	I/O Port Ground
V <sub>DDC</sub>	8	-	-	Core Supply Monitor (2.5 V)
V <sub>SSC</sub>	7	-	-	Core Supply Ground
V <sub>AREF</sub>	25	-	-	ADC Reference Voltage
V <sub>AGND</sub>	24	-	-	ADC Reference Ground
XTAL1	6	I	Hi-Z	External Oscillator Input (NC if not needed)
XTAL2	5	0	Hi-Z	External Oscillator Output (NC if not needed)
TMS	11	I	PD	Test Mode Select
RESET	38	I	PU	Reset Input
MBC <sup>1)</sup>	1	I	PU	Monitor & BootStrap Loader Control

#### Table 2 Pin Definitions and Functions (cont'd)

<sup>1)</sup> An external pull-up device in the range of 4.7 k $\Omega$  to 100 k $\Omega$  is required to enter user mode. Alternatively MBC can be tied to high if alternate functions (for debugging) of the pin are not utilized.



### 3 Functional Description

#### 3.1 Processor Architecture

The SAA-XC866 is based on a high-performance 8-bit Central Processing Unit (CPU) that is compatible with the standard 8051 processor. While the standard 8051 processor is designed around a 12-clock machine cycle, the SAA-XC866 CPU uses a 2-clock machine cycle. This allows fast access to ROM or RAM memories without wait state. Access to the Flash memory, however, requires an additional wait state (one machine cycle). The instruction set consists of 45% one-byte, 41% two-byte and 14% three-byte instructions.

The SAA-XC866 CPU provides a range of debugging features, including basic stop/start, single-step execution, breakpoint support and read/write access to the data memory, program memory and SFRs. Figure 5 shows the CPU functional blocks.

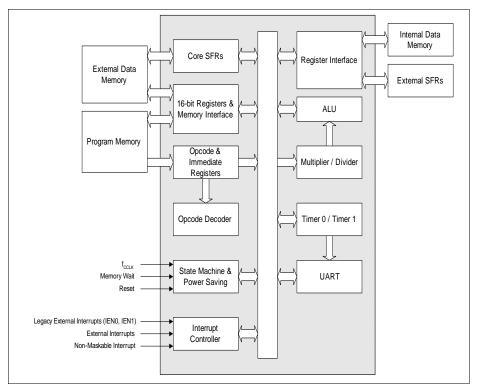


Figure 5 CPU Block Diagram



#### 3.2 Memory Organization

The SAA-XC866 CPU operates in the following five address spaces:

- 8 Kbytes of Boot ROM program memory
- 256 bytes of internal RAM data memory
- 512 bytes of XRAM memory
- a 128-byte Special Function Register area
- 4/8/16 Kbytes of Flash program memory (Flash devices); or 8/16 Kbytes of ROM program memory, with additional 4 Kbytes of Flash (ROM devices)



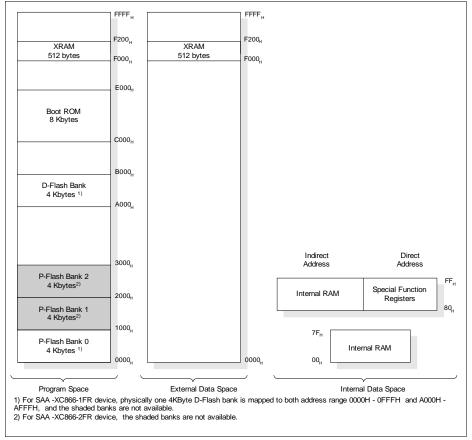
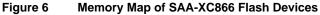
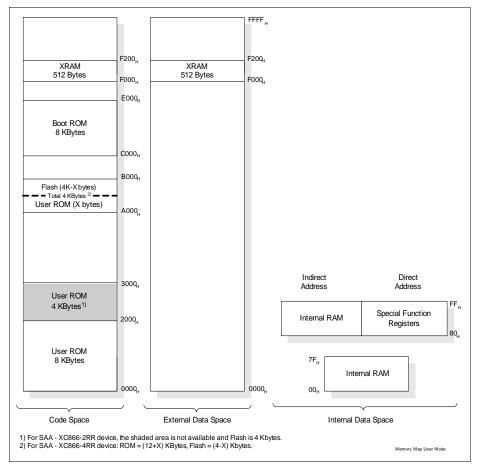


Figure 6 illustrates the memory address spaces of the SAA-XC866-4FR devices.







#### Figure 7 illustrates the memory address spaces of the SAA-XC866-4RR device.

Figure 7 Memory Map of SAA-XC866 ROM Devices



#### 3.2.1 Memory Protection Strategy

The SAA-XC866 memory protection strategy includes:

- Read-out protection: The Flash Memory can be enabled for read-out protection and ROM memory is always protected.
- Program and erase protection: The Flash memory in all devices can be enabled for program and erase protection.

Flash memory protection is available in two modes:

- Mode 0: Only the P-Flash is protected; the D-Flash is unprotected
- Mode 1: Both the P-Flash and D-Flash are protected

The selection of each protection mode and the restrictions imposed are summarized in **Table 3**.

Mode	0	1
Activation	Program a valid password via BSL m	node 6
Selection	MSB of password = 0	MSB of password = 1
P-Flash contents can be read by	Read instructions in the P-Flash	Read instructions in the P-Flash or D-Flash
P-Flash program and erase	Not possible	Not possible
D-Flash contents can be read by	Read instructions in any program memory	Read instructions in the P-Flash or D-Flash
D-Flash program	Possible	Not possible
D-Flash erase	Possible, on the condition that bit DFLASHEN in register MISC_CON is set to 1 prior to each erase operation	Not possible

#### Table 3 Flash Protection Modes

BSL mode 6, which is used for enabling Flash protection, can also be used for disabling Flash protection. Here, the programmed password must be provided by the user. A password match triggers an automatic erase of the read-protected Flash contents, see **Table 4** and **Table 5**, and the programmed password is erased. The Flash protection is then disabled upon the next reset.

#### For XC866-2FR and XC866-4FR devices:

The selection of protection type is summarized in Table 4.



#### Table 4 Flash Protection Type for XC866-2FR and XC866-4FR devices

PASSWORD	Type of Protection	Flash Banks to Erase when Unprotected
1XXXXXXX <sub>B</sub>	Flash Protection Mode 1	All Banks
0XXXXXXAB	Flash Protection Mode 0	P-Flash Bank

#### For XC866-1FR device and ROM devices:

The selection of protection type is summarized in Table 5.

#### Table 5 Flash Protection Type for XC866-1FR device and ROM devices

PASSWORD	Type of Protection (Applicable to the whole Flash)	Sectors to Erase when Unprotected	Comments
1XXXXXXX <sub>B</sub>	Read/Program/Erase	All Sectors	Compatible to Protection mode 1
00001XXX <sub>B</sub>	Erase	Sector 0	
00010XXX <sub>B</sub>	Erase	Sector 0 and 1	
00011XXX <sub>B</sub>	Erase	Sector 0 to 2	
00100XXX <sub>B</sub>	Erase	Sector 0 to 3	
00101XXX <sub>B</sub>	Erase	Sector 0 to 4	
00110XXX <sub>B</sub>	Erase	Sector 0 to 5	
00111XXX <sub>B</sub>	Erase	Sector 0 to 6	
01000XXX <sub>B</sub>	Erase	Sector 0 to 7	
01001XXX <sub>B</sub>	Erase	Sector 0 to 8	
01010XXX <sub>B</sub>	Erase	All Sectors	
Others	Erase	None	

Although no protection scheme can be considered infallible, the SAA-XC866 memory protection strategy provides a very high level of protection for a general purpose microcontroller.



#### 3.2.2 Special Function Register

The Special Function Registers (SFRs) occupy direct internal data memory space in the range  $80_H$  to FF<sub>H</sub>. All registers, except the program counter, reside in the SFR area. The SFRs include pointers and registers that provide an interface between the CPU and the on-chip peripherals. As the 128-SFR range is less than the total number of registers required, address extension mechanisms are required to increase the number of addressable SFRs. The address extension mechanisms include:

- Mapping
- Paging

#### 3.2.2.1 Address Extension by Mapping

Address extension is performed at the system level by mapping. The SFR area is extended into two portions: the standard (non-mapped) SFR area and the mapped SFR area. Each portion supports the same address range  $80_H$  to FF<sub>H</sub>, bringing the number of addressable SFRs to 256. The extended address range is not directly controlled by the CPU instruction itself, but is derived from bit RMAP in the system control register SYSCON0 at address  $8F_H$ . To access SFRs in the mapped area, bit RMAP in SFR SYSCON0 must be set. Alternatively, the SFRs in the standard area can be accessed by clearing bit RMAP. The SFR area can be selected as shown in Figure 8.

#### SYSCON0 System Control Register 0 Reset Value: 00<sub>H</sub> 2 1 7 6 5 4 3 0 0 1 0 RMAP r rw r rw

Field	Bits	Туре	Description	
RMAP	0	rw	<ul> <li>Special Function Register Map Control</li> <li>The access to the standard SFR area is enabled.</li> <li>The access to the mapped SFR area is enabled.</li> </ul>	
1	2	rw	<b>Reserved</b> Returns the last value if read; should be writter with 1.	
0	1,[7:3]	r	<b>Reserved</b> Returns 0 if read; should be written with 0.	

## Note: The RMAP bit must be cleared/set by ANL or ORL instructions. The rest bits of SYSCON0 should not be modified.

As long as bit RMAP is set, the mapped SFR area can be accessed. This bit is not cleared automatically by hardware. Thus, before standard/mapped registers are accessed, bit RMAP must be cleared/set, respectively, by software.

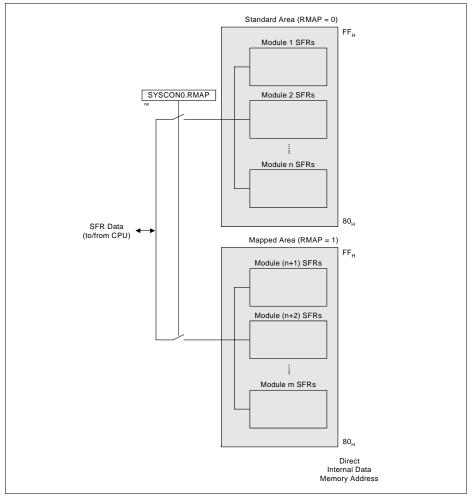
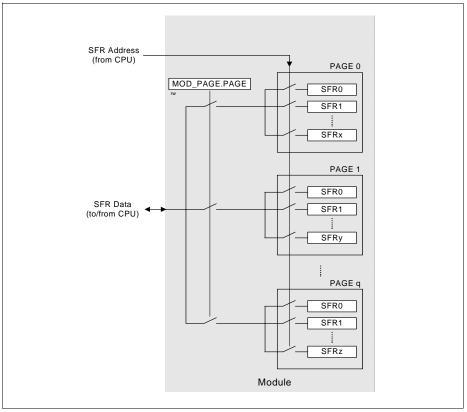


Figure 8 Address Extension by Mapping



### 3.2.2.2 Address Extension by Paging

Address extension is further performed at the module level by paging. With the address extension by mapping, the SAA-XC866 has a 256-SFR address range. However, this is still less than the total number of SFRs needed by the on-chip peripherals. To meet this requirement, some peripherals have a built-in local address extension mechanism for increasing the number of addressable SFRs. The extended address range is not directly controlled by the CPU instruction itself, but is derived from bit field PAGE in the module page register MOD\_PAGE. Hence, the bit field PAGE must be programmed before accessing the SFR of the target module. Each module may contain a different number of pages and a different number of SFRs per page, depending on the specific requirement. Besides setting the correct RMAP bit value to select the SFR area, the user must also ensure that a valid PAGE is selected to target the desired SFR. A page inside the extended address range can be selected as shown in Figure 9.



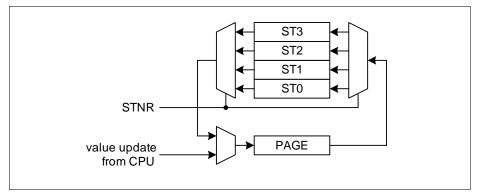


In order to access a register located in a page different from the actual one, the current page must be left. This is done by reprogramming the bit field PAGE in the page register. Only then can the desired access be performed.

If an interrupt routine is initiated between the page register access and the module register access, and the interrupt needs to access a register located in another page, the current page setting can be saved, the new one programmed and finally, the old page setting restored. This is possible with the storage fields STx (x = 0 - 3) for the save and restore action of the current page setting. By indicating which storage bit field should be used in parallel with the new page value, a single write operation can:

- Save the contents of PAGE in STx before overwriting with the new value (this is done in the beginning of the interrupt routine to save the current page setting and program the new page number); or
- Overwrite the contents of PAGE with the contents of STx, ignoring the value written to the bit positions of PAGE

(this is done at the end of the interrupt routine to restore the previous page setting before the interrupt occurred)



#### Figure 10 Storage Elements for Paging

With this mechanism, a certain number of interrupt routines (or other routines) can perform page changes without reading and storing the previously used page information. The use of only write operations makes the system simpler and faster. Consequently, this mechanism significantly improves the performance of short interrupt routines.

The SAA-XC866 supports local address extension for:

- Parallel Ports
- Analog-to-Digital Converter (ADC)
- Capture/Compare Unit 6 (CCU6)
- System Control Registers



The page register has the following definition:

## MOD\_PAGE

#### Page Register for module MOD

#### Reset Value: 00<sub>H</sub>

7	6	5	4	3	2	1	0
0	Р	ST	NR	0		PAGE	
v	V	v	V	r		rw	

Field	Bits	Туре	Description
PAGE	[2:0]	rw	Page Bits When written, the value indicates the new page. When read, the value indicates the currently active page.
STNR	[5:4]	w	<b>Storage Number</b> This number indicates which storage bit field is the target of the operation defined by bit field OP. If $OP = 10_B$ , the contents of PAGE are saved in STx before being overwritten with the new value. If $OP = 11_B$ , the contents of PAGE are overwritten by the contents of STx. The value written to the bit positions of PAGE is ignored.
			<ul> <li>00 ST0 is selected.</li> <li>01 ST1 is selected.</li> <li>10 ST2 is selected.</li> <li>11 ST3 is selected.</li> </ul>



### SAA-XC866

#### **Functional Description**

Field	Bits	Туре	Description
OP	[7:6]	w	<ul> <li>Operation</li> <li>OX Manual page mode. The value of STNR is ignored and PAGE is directly written.</li> <li>10 New page programming with automatic page saving. The value written to the bit positions of PAGE is stored. In parallel, the previous contents of PAGE are saved in the storage bit field STx indicated by STNR.</li> <li>11 Automatic restore page action. The value written to the bit positions PAGE is ignored and instead, PAGE is overwritten by the contents of the storage bit field STx indicated by STNR.</li> </ul>
0	3	r	<b>Reserved</b> Returns 0 if read; should be written with 0.



#### 3.2.3 Bit Protection Scheme

The bit protection scheme prevents direct software writing of selected bits (i.e., protected bits) using the PASSWD register. When the bit field MODE is  $11_B$ , writing  $10011_B$  to the bit field PASS opens access to writing of all protected bits, and writing  $10101_B$  to the bit field PASS closes access to writing of all protected bits. Note that access is opened for maximum 32 CCLKs if the "close access" password is not written. If "open access" password is written again before the end of 32 CCLK cycles, there will be a recount of 32 CCLK cycles. The protected bits include NDIV, WDTEN, PD, and SD.

#### PASSWD Password Register

Reset Value: 07<sub>H</sub>

7	6 5 4				2	1	0
		PASS		1	МС	DE	
		W			rh	r	W

Field	Bits	Туре	Description
MODE	[1:0]	rw	Bit Protection Scheme Control bits $00$ Scheme Disabled $11$ Scheme Enabled (default)Others: Scheme EnabledThese two bits cannot be written directly. To changethe value between $11_B$ and $00_B$ , the bit field PASSmust be written with $11000_B$ ; only then, will theMODE[1:0] be registered.
PROTECT_S	2	rh	<ul> <li>Bit Protection Signal Status bit</li> <li>This bit shows the status of the protection.</li> <li>0 Software is able to write to all protected bits.</li> <li>1 Software is unable to write to any protected bits.</li> </ul>
PASS	[7:3]	w	Password bitsThe Bit Protection Scheme only recognizes three patterns. $11000_B$ Enables writing of the bit field MODE. $10011_B$ Opens access to writing of all protected bits. $10101_B$ Closes access to writing of all protected bits.



#### 3.2.4 SAA-XC866 Register Overview

The SFRs of the SAA-XC866 are organized into groups according to their functional units. The contents (bits) of the SFRs are summarized in **Table 6** to **Table 14**, with the addresses of the bitaddressable SFRs appearing in bold typeface.

The CPU SFRs can be accessed in both the standard and mapped memory areas (RMAP = 0 or 1).

Addr	Register Name		Bit	7	6	5	4	3	2	1	0			
RMAP =	0 or 1				1	1	1	1		1				
81 <sub>H</sub>		eset: 07 <sub>H</sub>	Bit Field				S	P						
	Stack Pointer Register		Туре				r	w						
82 <sub>H</sub>	DPL Re	eset: 00 <sub>H</sub>	Bit Field	DPL7	DPL6	DPL5	DPL4	DPL3	DPL2	DPL1	DPL0			
	Data Pointer Register Low		Туре	rw	rw	rw	rw	rw	rw	rw	rw			
83 <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field	DPH7	DPH6	DPH5	DPH4	DPH3	DPH2	DPH1	DPH0			
	Data Pointer Register High		Туре	rw	rw	rw	rw	rw	rw	rw	rw			
87 <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field	SMOD		0		GF1	GF0	0	IDLE			
	Power Control Register		Туре	rw		r		rw	rw	r	rw			
88 <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0			
	Timer Control Register		Туре	rwh	rw	rwh	rw	rwh	rw	rwh	rw			
89 <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field	GATE1	0	T1	M	GATE0	0	T	M			
	Timer Mode Register		Туре	rw	r	r	w	rw	r	r	N			
8A <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field				V	۹L						
	Timer 0 Register Low		Туре				rv	vh						
8B <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field				V	۹L						
	Timer 1 Register Low		Туре				rv	vh						
8C <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field				V	۹L						
	Timer 0 Register High		Туре				rv	vh						
8D <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field				V	۹L						
	Timer 1 Register High		Туре				rv	vh						
98 <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field	SM0	SM1	SM2	REN	TB8	RB8	TI	RI			
	Serial Channel Control Regi	ster	Туре	rw	rw	rw	rw	rw	rwh	rwh	rwh			
99 <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field				V	۹L	rw         rw           DPH2         DPH1           rw         rw           GF0         0           rw         r           IT1         IE0           rw         rwh           0         T           r         -           0         T           r         -           RB8         TI           rwh         rwh           0         - <tr tr=""></tr>					
	Serial Data Buffer Register		Туре				rv	vh		IT1     IE0       rw     rwh       0     TO       r     rwh       0     TO       r     rwh       0     TI       rwh     rwh       0     rwh       0     rwh       PX11     PTO       rw     rw       PX1H     PTOH       rw     rw       OV     F1				
A2 <sub>H</sub>		eset: 00 <sub>H</sub>	Bit Field		0		TRAP_		0	DPSEL				
	Extended Operation Registe	er					EN				0			
			Туре		r		rw				rw			
A8 <sub>H</sub>	IEN0 Register 0	eset: 00 <sub>H</sub>	Bit Field	EA	0	ET2	ES	ET1			EX0			
			Туре	rw	r	rw	rw	rw			rw			
B8 <sub>H</sub>	IP Register	eset: 00 <sub>H</sub>	Bit Field		0	PT2	PS	PT1		-	PX0			
			Туре		r	rw	rw	rw			rw			
B9 <sub>H</sub>	IPH Register Hi	eset: 00 <sub>H</sub>	Bit Field		0	PT2H	PSH	PT1H			PX0H			
		•	Туре		r	rw	rw	rw			rw			
D0 <sub>H</sub>	PSW Reprogram Status Word Regis	eset: 00 <sub>H</sub>	Bit Field	CY	AC	F0	RS1	RS0			P			
	0		Туре	rw	rwh	rwh	rw	rw			rh			
E0 <sub>H</sub>	ACC Register	eset: 00 <sub>H</sub>	Bit Field	ACC7	ACC6	ACC5	ACC4	ACC3			ACC0			
	ç		Туре	rw	rw	rw	rw	rw			rw			
E8 <sub>H</sub>	IEN1 Register 1	eset: 00 <sub>H</sub>	Bit Field	ECCIP 3	ECCIP 2	ECCIP 1	ECCIP 0	EXM	EX2	ESSC	EADC			
			Туре	rw	rw	rw	rw	rw	rw	rw	rw			

#### Table 6 CPU Register Overview



#### Table 6 CPU Register Overview (cont'd)

Addr	Register Name	l.	Bit	7	6	5	4	3	2	1	0
F0 <sub>H</sub>	В	Reset: 00 <sub>H</sub>	Bit Field	B7	B6	B5	B4	B3	B2	B1	B0
	B Register		Туре	rw	rw	rw	rw	rw	rw	rw	rw
F8 <sub>H</sub>	IP1 Interrupt Priority Regi	Reset: 00 <sub>H</sub> ister 1	Bit Field	PCCIP 3	PCCIP 2	PCCIP 1	PCCIP 0	PXM	PX2	PSSC	PADC
			Туре	rw	rw	rw	rw	rw	rw	rw	rw
F9 <sub>H</sub>	IPH1 Interrupt Priority Regi	Reset: 00 <sub>H</sub> ister 1 High	Bit Field	PCCIP 3H	PCCIP 2H	PCCIP 1H	PCCIP 0H	PXMH	PX2H	PSSCH	PADC H
			Туре	rw	rw	rw	rw	rw	rw	rw	rw

The system control SFRs can be accessed in the standard memory area (RMAP = 0).

#### Table 7 System Control Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =	0 or 1					1	1			
8F <sub>H</sub>	SYSCON0 Reset: 00 <sub>H</sub>	Bit Field				0				RMAP
	System Control Register 0	Туре				r				rw
RMAP =	0									
BF <sub>H</sub>	SCU_PAGE Reset: 00 <sub>H</sub>	Bit Field	C	)P	STNR		0		PAGE	
	Page Register for System Control	Туре	,	w	w		r		rwh	
RMAP =	0, Page 0		.4							
B3 <sub>H</sub>	MODPISEL Reset: 00 <sub>H</sub> Peripheral Input Select Register	Bit Field		0	JTAG TDIS	JTAG TCKS		0	EXINT 0IS	URRIS
		Туре		r	rw	rw		r	rw	rw
B4 <sub>H</sub>	IRCON0 Reset: 00 <sub>H</sub> Interrupt Request Register 0	Bit Field	0	EXINT 6	EXINT 5	EXINT 4	EXINT 3	EXINT 2	EXINT 1	EXINT 0
		Туре	r	rwh	rwh	rwh	rwh	rwh	rwh	rwh
B5 <sub>H</sub>	IRCON1 Reset: 00 <sub>H</sub> Interrupt Request Register 1	Bit Field		0		ADCS RC1	ADCS RC0	RIR	TIR	EIR
		Туре		r		rwh	rwh	rwh rwh		rwh
B7 <sub>H</sub>	EXICON0 Reset: 00 <sub>H</sub>	Bit Field	EXI	NT3	EXI	NT2	EXI	NT1	EXI	NT0
	External Interrupt Control Register 0	Туре	r	w	r	w		rw		w
BA <sub>H</sub>	EXICON1 Reset: 00 <sub>H</sub>	Bit Field		0	EXI	NT6	EXI	INT5 EX		NT4
	External Interrupt Control Register 1	Туре		r	r	w	r	w	r	w
BB <sub>H</sub>	NMICON Reset: 00 <sub>H</sub> NMI Control Register	Bit Field	0	NMI ECC	NMI VDDP	NMI VDD	NMI OCDS	NMI FLASH	NMI PLL	NMI WDT
		Туре	r	rw	rw	rw	rw	rw	rw	rw
BC <sub>H</sub>	NMISR Reset: 00 <sub>H</sub> NMI Status Register	Bit Field	0	FNMI ECC	FNMI VDDP	FNMI VDD	FNMI OCDS	FNMI FLASH	FNMI PLL	FNMI WDT
		Туре	r	rwh	rwh	rwh	rwh	rwh	rwh	rwh
BD <sub>H</sub>	BCON Reset: 00 <sub>H</sub>	Bit Field	BG	SEL	0	BREN		BRPRE		R
	Baud Rate Control Register	Туре	r	w	r	rw		rw		rw
BE <sub>H</sub>	BG Reset: 00 <sub>H</sub>	Bit Field				BR_V	ALUE			
	Baud Rate Timer/Reload Register	Туре					w			
E9 <sub>H</sub>	FDCON Reset: 00 <sub>H</sub> Fractional Divider Control Register	Bit Field	BGS	SYNEN	ERRSY N	EOFSY N	BRK	NDOV	FDM	FDEN
		Туре	rw	rw	rwh	rwh	rwh	rwh	rw	rw
EA <sub>H</sub>	FDSTEP Reset: 00 <sub>H</sub>	Bit Field				ST	EP			
	Fractional Divider Reload Register	Туре					w			
EB <sub>H</sub>	FDRES Reset: 00 <sub>H</sub>	Bit Field				RES	ULT		-	
	Fractional Divider Result Register	Туре				r	h			
RMAP =	0, Page 1		-		-	-				



#### Table 7System Control Register Overview (cont'd)

Addr	Register Name		Bit	7	6	5	4	3	2	1	0	
B3 <sub>H</sub>	ID	Reset: 01 <sub>H</sub>	Bit Field			PRODID		1		VERID		
	Identity Register		Туре			r				r		
B4 <sub>H</sub>	PMCON0 Power Mode Control Re	Reset: 00 <sub>H</sub> gister 0	Bit Field	0	WDT RST	WKRS	WK SEL	SD	PD	W	/S	
			Туре	r	rwh	rwh	rw	rw	rwh	r	w	
B5 <sub>H</sub>	PMCON1 Power Mode Control Re	Reset: 00 <sub>H</sub> gister 1	Bit Field			0		T2_DIS	CCU _DIS	SSC _DIS	ADC _DIS	
			Туре			r		rw	rw	rw	rw	
B6 <sub>H</sub>	OSC_CON OSC Control Register	Reset: 08 <sub>H</sub>	Bit Field		0		OSC PD	XPD	OSC SS	ORD RES	OSCR	
			Туре		r		rw	rw	rw	rwh	rh	
B7 <sub>H</sub>	PLL_CON PLL Control Register	Reset: 20 <sub>H</sub>	Bit Field		N	VIC		VCO BYP	OSC DISC	RESLD	LOCK	
			Туре		r	w		rw	rw	rwh	rh	
BA <sub>H</sub>	CMCON Clock Control Register	Reset: 00 <sub>H</sub>	Bit Field	VCO SEL		0			CLK	CLKREL		
			Туре	rw		r			r	w		
BB <sub>H</sub>	PASSWD Password Register	Reset: 07 <sub>H</sub>	Bit Field			PASS			PROTE CT_S	МС	DE	
			Туре			w			rh	r	rw	
BC <sub>H</sub>	FEAL	Reset: 00 <sub>H</sub>	Bit Field			E	CCERR	ADDR[7:	0]			
	Flash Error Address Re	gister Low	Туре					'n				
BD <sub>H</sub>	FEAH	Reset: 00 <sub>H</sub>	Bit Field			E	CCERR	ADDR[15	:8]			
	Flash Error Address Re	gister High	Туре				I	'n				
BE <sub>H</sub>	COCON Clock Output Control Re	Reset: 00 <sub>H</sub> egister	Bit Field	(	)	TLEN	COUT S		CO	REL		
			Туре		r	rw	rw		r	w		
E9 <sub>H</sub>	MISC_CON Miscellaneous Control F	Reset: 00 <sub>H</sub> Register	Bit Field				0			DFLAS HEN		
			Туре				r				rwh	
RMAP =	0, Page 3											
B3 <sub>H</sub>	XADDRH	Reset: F0 <sub>H</sub>	Bit Field				AD	DRH				
	On-Chip XRAM Address	Ligher Order	Туре	rw								

The WDT SFRs can be accessed in the mapped memory area (RMAP = 1).

#### Table 8 WDT Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =	1		1		1					
BB <sub>H</sub>	WDTCON Reset: 0 Watchdog Timer Control Register	0 <sub>H</sub> Bit Field	(	0	WINB EN	WDT PR	0	WDT EN	WDT RS	WDT IN
		Туре		r	rw	rh	r	rw	rwh	rw
BC <sub>H</sub>	WDTREL Reset: 0	0 <sub>H</sub> Bit Field				WD1	REL			
	Watchdog Timer Reload Register	Туре				r	w		DT WDT N RS	
BD <sub>H</sub>	WDTWINB Reset: 0 Watchdog Window-Boundary Cour					WDT	WINB			
	Register	Туре				r	w			
BEH	WDTL Reset: 0	0 <sub>H</sub> Bit Field				WDT	[7:0]			
	Watchdog Timer Register Low	Туре				r	h			
BF <sub>H</sub>	WDTH Reset: 0	0 <sub>H</sub> Bit Field				WDT	[15:8]			
	Watchdog Timer Register High	Туре	rh							



The Port SFRs can be accessed in the standard memory area (RMAP = 0).

#### Table 9 Port Register Overview

Addr	Register Name		Bit	7	6	5	4	3	2	1	0
RMAP =	•				-	-		_			-
B2 <sub>H</sub>	PORT PAGE	Reset: 00 <sub>H</sub>	Bit Field		)P	ST	NR	0		PAGE	
H	Page Register for PORT	Hoosen oog	Туре	-	w	-	w	r		rwh	
RMAP -	0, Page 0		турс								
80 <sub>H</sub>	P0 DATA	Reset: 00 <sub>H</sub>	Bit Field		0	P5	P4	P3	P2	P1	P0
U U	P0 Data Register	Hoosen oog	Туре		r	rwh	rwh	rwh	rwh	rwh	rwh
86 <sub>H</sub>	P0 DIR	Reset: 00 <sub>H</sub>	Bit Field		0	P5	P4	P3	P2	P1	P0
UUH	P0 Direction Register	Hoosen oog	Туре		r	rw	rw	rw	rw	rw	rw
90 <sub>H</sub>	P1 DATA	Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5		0		P1	P0
	P1 Data Register		Туре	rwh	rwh	rwh		r		rwh	rwh
91 <sub>H</sub>	P1 DIR	Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5		0		P1	P0
- · n	P1 Direction Register		Туре	rw	rw	rw		r		rw	rw
A0 <sub>H</sub>	P2 DATA	Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
- 11	P2 Data Register		Type	rwh	rwh						
A1 <sub>H</sub>	P2_DIR	Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P2 Direction Register	· · · n	Туре	rw	rw						
B0 <sub>н</sub>	P3_DATA	Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Data Register		Type	rwh	rwh						
B1 <sub>H</sub>	P3_DIR	Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Direction Register		Туре	rw	rw						
RMAP =	0, Page 1			1							1
80 <sub>H</sub>	P0_PUDSEL	Reset: FF <sub>H</sub>	Bit Field		0	P5	P4	P3	P2	P1	P0
	P0 Pull-Up/Pull-Down Se	P0 Pull-Up/Pull-Down Select Register P0_PUDEN Reset: C4 <sub>H</sub>			r	rw	rw	rw	rw	rw	rw
86 <sub>H</sub>	P0_PUDEN				0	P5	P4	P3	P2	P1	P0
	P0 Pull-Up/Pull-Down Er	able Register	Туре		r	rw	rw	rw	rw	rw	rw
90 <sub>H</sub>	P1_PUDSEL	Reset: FF <sub>H</sub>	Bit Field	P7	P6	P5		0		P1	P0
	P1 Pull-Up/Pull-Down Se	lect Register	Туре	rw	rw	rw		r		rw	rw
91 <sub>H</sub>	P1_PUDEN	Reset: FF <sub>H</sub>	Bit Field	P7	P6	P5		0		P1	P0
	P1 Pull-Up/Pull-Down Er	able Register	Туре	rw	rw	rw		r		rw	rw
A0 <sub>H</sub>	P2_PUDSEL	Reset: FF <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P2 Pull-Up/Pull-Down Se	lect Register	Туре	rw	rw						
A1 <sub>H</sub>	P2_PUDEN	Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P2 Pull-Up/Pull-Down Er	able Register	Туре	rw	rw						
В0 <sub>Н</sub>	P3_PUDSEL	Reset: BF <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Pull-Up/Pull-Down Se	lect Register	Туре	rw	rw						
B1 <sub>H</sub>	P3_PUDEN	Reset: 40 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Pull-Up/Pull-Down En	able Register	Туре	rw	rw						
RMAP =	0, Page 2		•								
80 <sub>H</sub>	P0_ALTSEL0	Reset: 00 <sub>H</sub>	Bit Field	1	0	P5	P4	P3	P2	P1	P0
	P0 Alternate Select 0 Re	gister	Туре		r	rw	rw	rw	rw	rw	rw
86 <sub>H</sub>	P0_ALTSEL1	Reset: 00 <sub>H</sub>	Bit Field		0	P5	P4	P3	P2	P1	P0
	P0 Alternate Select 1 Re	gister	Туре		r	rw	rw	rw	rw	rw	rw
90 <sub>H</sub>	P1_ALTSEL0	Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5		0		P1	P0
	P1 Alternate Select 0 Re	gister	Туре	rw	rw	rw		r		rw	rw
91 <sub>H</sub>	P1_ALTSEL1	Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5		0		P1	P0
	P1 Alternate Select 1 Re	gister	Туре	rw	rw	rw		r		rw	rw
во <sub>н</sub>	P3_ALTSEL0	Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Alternate Select 0 Re	gister	Туре	rw	rw						



#### Table 9 Port Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
B1 <sub>H</sub>	P3_ALTSEL1 Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Alternate Select 1 Register	Туре	rw							
RMAP =	0, Page 3									
80 <sub>H</sub>			(	C	P5	P4	P3	P2	P1	P0
	P0 Open Drain Control Register	Туре		r	rw	rw	rw	rw	rw	rw
90 <sub>H</sub>	P1_OD Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5		0		P1	P0
	P1 Open Drain Control Register	Туре	rw	rw	rw		r		rw	rw
B0 <sub>H</sub>	P3_OD Reset: 00 <sub>H</sub>	Bit Field	P7	P6	P5	P4	P3	P2	P1	P0
	P3 Open Drain Control Register	Туре	rw							

The ADC SFRs can be accessed in the standard memory area (RMAP = 0).

#### Table 10 ADC Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0	
RMAP =	0					1					
D1 <sub>H</sub>	ADC_PAGE Reset: 0	00 <sub>H</sub> Bit Field	C	P	STNR w		0	0 PAGE r rwh			
	Page Register for ADC	Туре	,	N			r			rwh	
RMAP =	0, Page 0										
CA <sub>H</sub>	ADC_GLOBCTR Reset: 3	Bit Field	ANON	NON DW CTC			0				
	Global Control Register	Туре	rw	rw	r	w		r			
CB <sub>H</sub>	ADC_GLOBSTR Reset: 0 Global Status Register	00 <sub>H</sub> Bit Field	0		CHNR		0		SAM PLE	BUSY	
		Туре	r rh		rh		r	rh	rh		
CCH	ADC_PRAR Reset: 0	00 <sub>H</sub> Bit Field	ASEN1	ASEN0	0	ARBM	CSM1	PRIO1	CSM0	PRIOC	
	Priority and Arbitration Register	Туре	rw	rw	r	rw	rw	rw	rw	rw	
CD <sub>H</sub>	ADC_LCBR Reset: E	87 <sub>H</sub> Bit Field	BOUND1			BOUND0					
	Limit Check Boundary Register	Туре	rw					rw			
CEH	ADC_INPCR0 Reset: 0	00 <sub>H</sub> Bit Field				S	TC				
	Input Class Register 0	Туре	rw								
CF <sub>H</sub>	ADC_ETRCR Reset: 0 External Trigger Control Register	00 <sub>H</sub> Bit Field	SYNEN 1	SYNEN SYNEN ETRSEL			1	ETRSEL0			
		Туре	rw	rw rw			rw				
RMAP =	0, Page 1										
CA <sub>H</sub>	ADC_CHCTR0 Reset: 00 <sub>H</sub> Channel Control Register 0	00 <sub>H</sub> Bit Field	0	0 LCC		0		RESRSEL			
		Туре	r	r rw		r		rw			
CBH	ADC_CHCTR1 Reset: 0	00 <sub>H</sub> Bit Field	0		LCC		0		RESRSEL		
	Channel Control Register 1	Туре	r	rw		r		rw			
CCH	ADC_CHCTR2 Reset: 0	00 <sub>H</sub> Bit Field	0	LCC		0		RESRSEL			
	Channel Control Register 2	Туре	r	r rw		r		rw			
CD <sub>H</sub>	ADC_CHCTR3 Reset: 0	00 <sub>H</sub> Bit Field	0	0 LCC		0 R		RES	RSEL		
	Channel Control Register 3	Туре	r	r rw		r		rw			
CEH	ADC_CHCTR4 Reset: 0	00 <sub>H</sub> Bit Field	0	0 LCC		0		RESRSEL			
	Channel Control Register 4	Туре	r rw		r		rw				
CF <sub>H</sub>	ADC_CHCTR5 Reset: 0	00 <sub>H</sub> Bit Field	0	0 LCC		0		RESRSEL			
	Channel Control Register 5	Туре	r	r rw		r		rw			
D2 <sub>H</sub>	ADC_CHCTR6 Reset: 0	00 <sub>H</sub> Bit Field	0	0 LCC		0		RESRSEL			
	Channel Control Register 6	Туре	r rw		r		rw				
D3 <sub>H</sub>	ADC_CHCTR7 Reset: 0	00 <sub>H</sub> Bit Field	0	0 LCC		0		RESRSEL			
	Channel Control Register 7	Type	r	r rw			r	r	w		



#### Table 10 ADC Register Overview (cont'd)

Addr	Register Name		Bit	7	6	5	4	3	2	1	0	
CA <sub>H</sub>	ADC_RESR0L	Reset: 00 <sub>H</sub>	Bit Field	RESULT[1:0]		0	VF	DRC CHN		CHNR		
	Result Register 0 Low		Туре	rh		r	rh	rh	rh rh			
CB <sub>H</sub>	ADC_RESR0H	Reset: 00 <sub>H</sub>	Bit Field	RESULT[9:2]								
	Result Register 0 High		Туре	rh								
CCH	ADC_RESR1L	Reset: 00 <sub>H</sub>	Bit Field	RESULT[1:0] 0			VF	DRC CHNF				
	Result Register 1 Low		Туре			r	rh	rh	rh rh			
CD <sub>H</sub>	ADC RESR1H	Reset: 00 <sub>H</sub>	Bit Field				RESULT[9:2]					
	Result Register 1 High		Туре	rh								
CEH	ADC_RESR2L Result Register 2 Low	Reset: 00 <sub>H</sub>	Bit Field	RESULT[1:0] 0		VF	DRC		CHNR			
			Туре	rh r		r	rh	rh rh				
CF <sub>H</sub>	ADC_RESR2H Result Register 2 High	Reset: 00 <sub>H</sub>	Bit Field	RESULT[9:2]								
			Туре	rh								
D2 <sub>H</sub>	ADC RESR3L	Reset: 00 <sub>H</sub>	Bit Field	RESU	LT[1:0]	0	VF	DRC		CHNR		
5-H	Result Register 3 Low		Type	r		r	rh	rh		rh		
D3 <sub>H</sub>	ADC RESR3H	Reset: 00 <sub>H</sub>	Bit Field				RESULT[9:2]					
	Result Register 3 High		Туре					rh				
RMAP =	0, Page 3		.,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,									
CA <sub>H</sub>				RESULT[2:0]			VF	DRC		CHNR		
O, H	Result Register 0, View A		Bit Field Type	rh			rh	rh		rh		
CB <sub>H</sub>	ADC_RESRA0H Result Register 0, View /	Reset: 00 <sub>H</sub> A High	Bit Field	RESULT[1								
			Type	rh								
ССн	ADC RESRA1L	Reset: 00 <sub>H</sub>	Bit Field	RESULT[2:0]			VF .	DRC CHNR				
OOH	Result Register 1, View /		Type	rh			rh	rh		rh		
CD <sub>H</sub>	ADC RESRA1H	Reset: 00 <sub>H</sub>	Bit Field					ESULT[10:3]				
CDH	Result Register 1, View		Type	rh								
CEH	ADC_RESRA2L Result Register 2, View A	Reset: 00 <sub>H</sub>	Bit Field	RESULT[2:0] VI				DRC CHNR				
			Type	rh			rh	rh rh				
CF <sub>H</sub>	ADC_RESRA2H Result Register 2, View /	Reset: 00 <sub>H</sub>	Bit Field	IN								
			Type	RESULT[10:3]								
D2 <sub>H</sub>	ADC_RESRA3L Result Register 3, View /	Reset: 00 <sub>H</sub>	Bit Field	RESULT[2:0]			VF	rh DRC CHNF				
DZH			Type	rh			rh	rh rh				
DO	<b>e</b> .		Bit Field	m						m		
D3 <sub>H</sub>	ADC_RESRA3H Reset: 00 <sub>H</sub> Result Register 3, View A High			RESULT[10:3] rh								
		g.i	Туре				I	n				
	ADC RCR0	Desets 00	Bit Field	VFCTR		0	IEN		0		DRCT	
CA <sub>H</sub>	Result Control Register (	Reset: 00 <sub>H</sub>	BIT FIEID	VFCIR	WFR	0	IEN		0		R	
	Result Control Register C		Туре	rw	rw	r	rw		r		rw	
СВ <sub>Н</sub>	ADC RCR1	Reset: 00 <sub>H</sub>	Bit Field	VFCTR	WFR	0	IEN	0			DRCT	
CDH	Result Control Register 1		DILFIEIU	VFUIK	WER	0	IEIN		0		R	
			Туре	rw	rw	r	rw	r		rw		
CCH	ADC_RCR2 Reset: 00 <sub>H</sub>		Bit Field	VFCTR	WFR	0	IEN	0			DRCT	
004	Result Control Register 2					-			-		R	
			Туре	rw	rw	r	rw		r		rw	
CDH	ADC_RCR3	Reset: 00 <sub>H</sub>	Bit Field	VFCTR	WFR	0	IEN	0			DRCT	
	Result Control Register 3										R	
			Туре	rw	rw	r	rw		r		rw	
CEH	ADC_VFCR	Reset: 00 <sub>H</sub>	Bit Field	0				VFC3	VFC2	VFC1	VFC0	
	Valid Flag Clear Register		Туре	r				w	w	w	w	
RMAP =	0, Page 5											



### Table 10ADC Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
CA <sub>H</sub>	ADC_CHINFR Reset: 00 <sub>H</sub> Channel Interrupt Flag Register	Bit Field	CHINF 7	CHINF 6	CHINF 5	CHINF 4	CHINF 3	CHINF 2	CHINF 1	CHINF 0
		Туре	rh	rh	rh	rh	rh	rh	rh	rh
CB <sub>H</sub>	ADC_CHINCR Reset: 00 <sub>H</sub> Channel Interrupt Clear Register	Bit Field	CHINC 7	CHINC 6	CHINC 5	CHINC 4	CHINC 3	CHINC 2	CHINC 1	CHINC 0
		Туре	w	w	w	w	w	w	w	w
CCH	ADC_CHINSR Reset: 00 <sub>H</sub> Channel Interrupt Set Register	Bit Field	CHINS 7	CHINS 6	CHINS 5	CHINS 4	CHINS 3	CHINS 2	CHINS 1	CHINS 0
		Туре	w	w	w	w	w	w	w	w
CD <sub>H</sub>	ADC_CHINPR Reset: 00 <sub>H</sub> Channel Interrupt Node Pointer	Bit Field	CHINP 7	CHINP 6	CHINP 5	CHINP 4	CHINP 3	CHINP 2	CHINP 1	CHINP 0
	Register	Туре	rw	rw	rw	rw	rw	rw	rw	rw
CEH	ADC_EVINFR Reset: 00 <sub>H</sub> Event Interrupt Flag Register	Bit Field	EVINF 7	EVINF 6	EVINF 5	EVINF 4	(	0	EVINF 1	EVINF 0
		Туре	rh	rh	rh	rh		r	rh	rh
CF <sub>H</sub>	ADC_EVINCR Reset: 00 <sub>H</sub> Event Interrupt Clear Flag Register	Bit Field	EVINC 7	EVINC 6	EVINC 5	EVINC 4		0	EVINC 1	EVINC 0
		Туре	w	w	w	w		r	w	w
D2 <sub>H</sub>	ADC_EVINSR Reset: 00 <sub>H</sub> Event Interrupt Set Flag Register	Bit Field	EVINS 7	EVINS 6	EVINS 5	EVINS 4	(	0	EVINS 1	EVINS 0
		Туре	w	w	w	w	r		w	w
D3 <sub>H</sub>	ADC_EVINPR Reset: 00 <sub>H</sub> Event Interrupt Node Pointer Register	Bit Field	EVINP 7	EVINP 6	EVINP 5	EVINP 4	(	0	EVINP 1	EVINP 0
		Туре	rw	rw	rw	rw		r	rw	rw
RMAP =	0, Page 6									
CA <sub>H</sub>	ADC_CRCR1 Reset: 00 <sub>H</sub> Conversion Request Control Register 1	Bit Field	CH7	CH6	CH5	CH4		(	)	
		Туре	rwh	rwh	rwh	rwh			r	
CB <sub>H</sub>	ADC_CRPR1 Reset: 00 <sub>H</sub> Conversion Request Pending	Bit Field	CHP7	CHP6	CHP5	CHP4		(	D	
	Register 1	Туре	rwh	rwh	rwh	rwh			r	
CCH	ADC_CRMR1 Reset: 00 <sub>H</sub> Conversion Request Mode Register 1	Bit Field	Rsv	LDEV	CLR PND	SCAN	ENSI	ENTR	EN	GT
		Туре	r	w	w	rw	rw	rw	r	w
CD <sub>H</sub>	ADC_QMR0 Reset: 00 <sub>H</sub>	Bit Field	CEV	TREV	FLUSH	CLRV	TRMD	ENTR	EN	GT
	Queue Mode Register 0	Туре	w	w	w	w	rw	rw	r	w
CEH	ADC_QSR0 Reset: 20 <sub>H</sub> Queue Status Register 0	Bit Field Type	Rsv r	0 r	EMPTY rh	EV rh			) r	
CF <sub>H</sub>	ADC_Q0R0 Reset: 00 <sub>H</sub>	Bit Field	EXTR	ENSI	RF	V	0			R
-· H	Queue 0 Register 0	Type	rh	rh	rh	rh	r		rh	••
D2 <sub>H</sub>	ADC QBUR0 Reset: 00 <sub>H</sub>	Bit Field	EXTR	ENSI	RF	V	0	R	EQCHN	R
H	Queue Backup Register 0	Туре	rh	rh	rh	rh	r		rh	
		.,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,								
D2 <sub>H</sub>	ADC_QINR0 Reset: 00 <sub>H</sub>	Bit Field	EXTR	ENSI	RF	(	)	R	EQCHN	R

The Timer 2 SFRs can be accessed in the standard memory area (RMAP = 0).

### Table 11Timer 2 Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
C0 <sub>H</sub>	T2_T2CON Timer 2 Control Register	Bit Field	TF2	EXF2	(	C	EXEN2	TR2	0	CP/ RL2
		Туре	rwh	rwh		r	rw	rwh	r	rw



# Table 11 Timer 2 Register Overview (cont'd)

C1 <sub>H</sub>	T2_T2MOD Timer 2 Mode Register	Reset: 00 <sub>H</sub>	Bit Field	T2 REGS	T2 RHEN	EDGE SEL	PREN	T2PRE	DCEN
			Туре	rw	rw	rw	rw	rw	rw
C2 <sub>H</sub>	T2_RC2L	Reset: 00 <sub>H</sub>	Bit Field				RC2	[7:0]	
	Timer 2 Reload/Capture	Register Low	Туре				rv	/h	
C3 <sub>H</sub>	T2_RC2H	Reset: 00 <sub>H</sub>	Bit Field				RC2[	15:8]	
	Timer 2 Reload/Capture	Register High	Туре				rv	/h	
C4 <sub>H</sub>	T2_T2L	Reset: 00 <sub>H</sub>	Bit Field				THL2	2[7:0]	
	Timer 2 Register Low		Туре				rv	/h	
C5 <sub>H</sub>	T2_T2H	Reset: 00 <sub>H</sub>	Bit Field				THL2	[15:8]	
	Timer 2 Register High		Туре				rv	/h	

The CCU6 SFRs can be accessed in the standard memory area (RMAP = 0).

# Table 12 CCU6 Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0	
RMAP =	0		1		1						
A3 <sub>H</sub>	CCU6_PAGE Reset: 00 <sub>H</sub>	Bit Field	C	P	ST	NR	0		PAGE		
	Page Register for CCU6	Туре	١	N	١	N	r		rwh		
RMAP =	0, Page 0										
9A <sub>H</sub>	CCU6_CC63SRL Reset: 00 <sub>H</sub> Capture/Compare Shadow Register for	Bit Field				CC	3SL				
	Channel CC63 Low	Туре	rw								
9B <sub>H</sub>	CCU6_CC63SRH Reset: 00 <sub>H</sub> Capture/Compare Shadow Register for	Bit Field				CC6	3SH				
	Channel CC63 High	Туре				r	w				
9C <sub>H</sub>	CCU6_TCTR4L Reset: 00 <sub>H</sub> Timer Control Register 4 Low	Bit Field	T12 STD	T12 STR	(	0	DTRES	T12 RES	T12RS	T12RR	
		Туре	w	w		r	w	w	w	w	
9D <sub>H</sub>	CCU6_TCTR4H Reset: 00 <sub>H</sub> Timer Control Register 4 High	Bit Field	T13 STD	T13 STR			T13RS	T13RR			
		Туре	w	w	r w		w	w	w		
9E <sub>H</sub>	CCU6_MCMOUTSL Reset: 00 <sub>H</sub> Multi-Channel Mode Output Shadow	Bit Field	STRM CM	0	MCMPS						
	Register Low	Туре	w	r			r	w			
9F <sub>H</sub>	CCU6_MCMOUTSH Reset: 00 <sub>H</sub>	Bit Field	STRHP	0		CURHS		EXPHS			
	Multi-Channel Mode Output Shadow Register High	Туре	w	r		rw			rw		
A4 <sub>H</sub>	CCU6_ISRL Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status	Bit Field	RT12P M	RT12O M	RCC62 F	RCC62 R	RCC61 F	RCC61 R	RCC60 F	RCC60 R	
	Reset Register Low	Туре	w	w	w	w	w	w	w	w	
A5 <sub>H</sub>	CCU6_ISRH Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status	Bit Field	RSTR	RIDLE	RWHE	RCHE	0	RTRPF	RT13 PM	RT13 CM	
	Reset Register High	Туре	w	w	w	w	r	w	w	w	
A6 <sub>H</sub>	CCU6_CMPMODIFL Reset: 00 <sub>H</sub> Compare State Modification Register	Bit Field	0	MCC63 S		0		MCC62 S	MCC61 S	MCC60 S	
	Low	Туре	r	w		r		w	w	w	
A7 <sub>H</sub>	CCU6_CMPMODIFH Reset: 00 <sub>H</sub> Compare State Modification Register	Bit Field	0	MCC63 R		0		MCC62 R	MCC61 R	MCC60 R	
	High	Туре	r	w		r		w	w	w	
FA <sub>H</sub>	CCU6_CC60SRL Reset: 00 <sub>H</sub> Capture/Compare Shadow Register for										
	Channel CC60 Low	Туре				n	vh				



### Table 12 CCU6 Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
FB <sub>H</sub>	CCU6_CC60SRH Reset: 00 <sub>H</sub> Capture/Compare Shadow Register for	Bit Field				CC6	0SH			
	Channel CC60 High	Туре				rv	vh			
FC <sub>H</sub>	CCU6_CC61SRL Reset: 00 <sub>H</sub> Capture/Compare Shadow Register for	Bit Field				CC6	51SL			
	Channel CC61 Low	Туре				rv	vh			
FD <sub>H</sub>	CCU6_CC61SRH Reset: 00 <sub>H</sub> Capture/Compare Shadow Register for	Bit Field				CC6	1SH			
	Channel CC61 High	Туре				rv	vh			
FE <sub>H</sub>	CCU6_CC62SRL Reset: 00 <sub>H</sub> Capture/Compare Shadow Register for Channel CC62 Low	Bit Field					2SL			
		Туре					vh			
FF <sub>H</sub>	CCU6_CC62SRH Reset: 00 <sub>H</sub> Capture/Compare Shadow Register for Channel CC62 High	Bit Field					2SH			
DMAD	0, Page 1	Туре				rv	vh			
9A <sub>H</sub>	CCU6_CC63RL Reset: 00 <sub>H</sub> Capture/Compare Register for Channel	Bit Field				CC6	3VL			
	CC63 Low	Туре				r	h			
9B <sub>H</sub>	CCU6_CC63RH Reset: 00 <sub>H</sub> Capture/Compare Register for Channel	Bit Field					3VH			
	CC63 High	Туре				r	h			
9C <sub>H</sub>	CCU6_T12PRL Reset: 00 <sub>H</sub>	Bit Field				T12	PVL			
	Timer T12 Period Register Low	Туре					vh			
9D <sub>H</sub>	CCU6_T12PRH Reset: 00 <sub>H</sub>	Bit Field					PVH			
	Timer T12 Period Register High	Туре					vh			
9E <sub>H</sub>	CCU6_T13PRL Reset: 00 <sub>H</sub> Timer T13 Period Register Low	Bit Field					PVL			
	÷	Туре					vh			
9F <sub>H</sub>	CCU6_T13PRH Reset: 00 <sub>H</sub> Timer T13 Period Register High	Bit Field				-	PVH			
	CCU6 T12DTCL Reset: 00 <sub>H</sub>	Type Bit Field					vh FM			
A4 <sub>H</sub>	Dead-Time Control Register for Timer	Туре				r				
A5 <sub>H</sub>	CCU6_T12DTCH Reset: 00 <sub>H</sub> Dead-Time Control Register for Timer	Bit Field	0	DTR2	DTR1	DTR0	0	DTE2	DTE1	DTE0
	T12 High	Туре	r	rh	rh	rh	r	rw	rw	rw
A6 <sub>H</sub>	CCU6_TCTR0L Reset: 00 <sub>H</sub> Timer Control Register 0 Low	Bit Field	СТМ	CDIR	STE12	T12R	T12 PRE		T12CLK	
		Туре	rw	rh	rh	rh	rw		rw	
A7 <sub>H</sub>	CCU6_TCTR0H Reset: 00 <sub>H</sub> Timer Control Register 0 High	Bit Field	(	0	STE13	T13R	T13 PRE		T13CLK	
		Туре		r	rh	rh	rw		rw	
FA <sub>H</sub>	CCU6_CC60RL Reset: 00 <sub>H</sub> Capture/Compare Register for Channel CC60 Low	Bit Field					60VL			
		Туре				r		_	_	_
FB <sub>H</sub>	CCU6_CC60RH Reset: 00 <sub>H</sub> Capture/Compare Register for Channel CC60 High	Bit Field					ioVH			
FC	CCU6 CC61RL Reset: 00 <sub>H</sub>	Type Bit Field				r CCG	n 61VL			
FC <sub>H</sub>	Capture/Compare Register for Channel									
	0001 200	Туре				r	n			



### Table 12 CCU6 Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0	
FD <sub>H</sub>	CCU6_CC61RH Reset: 00 <sub>H</sub>	Bit Field			1	CCe	51VH				
	Capture/Compare Register for Channel CC61 High	Туре				r	h				
FE <sub>H</sub>	CCU6_CC62RL Reset: 00 <sub>H</sub> Capture/Compare Register for Channel	Bit Field				CC	32VL				
	CC62 Low	Туре				r	h				
FF <sub>H</sub>	CCU6_CC62RH Reset: 00 <sub>H</sub> Capture/Compare Register for Channel	Bit Field	CC62VH								
	CC62 High	Туре				r	h				
	0, Page 2										
9A <sub>H</sub>	CCU6_T12MSELL Reset: 00 <sub>H</sub> T12 Capture/Compare Mode Select Register Low	Bit Field	MSEL61 MSEL				EL60				
	•	Туре	0010	r		10)//10					
9B <sub>H</sub>	CCU6_T12MSELH Reset: 00 <sub>H</sub> T12 Capture/Compare Mode Select Register High	Bit Field	DBYP		HSYNC				EL62		
~~	0 0	Type	TW ENTER	ENIT40	rw	FNOC	ENIOC	1	W	ENOS	
9C <sub>H</sub>	CCU6_IENL Reset: 00 <sub>H</sub> Capture/Compare Interrupt Enable Register Low	Bit Field	ENT12 PM	ENT12 OM	ENCC 62F	ENCC 62R	ENCC 61F	ENCC 61R	ENCC 60F	ENCC 60R	
	•	Туре	rw	rw	rw	rw	rw	rw	rw	rw	
9D <sub>H</sub>	CCU6_IENH Reset: 00 <sub>H</sub> Capture/Compare Interrupt Enable Register High	Bit Field	ENSTR	EN IDLE	EN WHE	EN CHE	0	EN TRPF	ENT13 PM	CM	
	0 0	Туре	rw	rw	rw rw r		rw	rw	rw		
9E <sub>H</sub>	CCU6_INPL Reset: 40 <sub>H</sub> Capture/Compare Interrupt Node Pointer Register Low	Bit Field	INPCHE INPCC62			CC61		CC60			
	÷	Туре		w	rw INPT13		rw INPT12		rw		
9F <sub>H</sub>	CCU6_INPH Reset: 39 <sub>H</sub> Capture/Compare Interrupt Node Pointer Register High	Bit Field		C r		v113 w		v112	INPERR		
A 4	CCU6_ISSL Reset: 00 <sub>H</sub>	Type Bit Field	ST12P	ST120		SCC62		SCC61		SCC60	
A4 <sub>H</sub>	Capture/Compare Interrupt Status Set Register Low	Туре	M	M W	F	R 8	F	R	F	R	
<u>۸</u> ۲	*	Bit Field	SSTR	SIDLE	SWHE	SCHE		STRPF	W ST13	ST13	
A5 <sub>H</sub>	CCU6_ISSH Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status Set Register High			-					PM	CM	
4.0	5 5	Type Dit Field	w PSL63	w 0	w	w	W	w SL	w	w	
A6 <sub>H</sub>	CCU6_PSLR Reset: 00 <sub>H</sub> Passive State Level Register	Bit Field		-				-			
A7 <sub>H</sub>	CCU6 MCMCTR Reset: 00 <sub>H</sub>	Type Bit Field	rwh	r D	S/W	SYN	0	wh I	SWSEL		
Λ'H	Multi-Channel Mode Control Register	Type		r	-	W	r		rw		
FA <sub>H</sub>	CCU6_TCTR2L Reset: 00 <sub>H</sub> Timer Control Register 2 Low	Bit Field	0		TED	v	T13TEC	;	T13 SSC	T12 SSC	
		Туре	r	r	w		rw		rw	rw	
FB <sub>H</sub>	CCU6_TCTR2H Reset: 00 <sub>H</sub>	Bit Field			0	T13RSEL T12R		RSEL			
	Timer Control Register 2 High	Туре			r	rw rv		w			
FC <sub>H</sub>	CCU6_MODCTRL Reset: 00 <sub>H</sub> Modulation Control Register Low	Bit Field	MC MEN	0		T12MODEN					
		Туре	rw	r	rw						
FD <sub>H</sub>	CCU6_MODCTRH Reset: 00 <sub>H</sub> Modulation Control Register High	Bit Field	ECT13 O	0	T13MODEN						
		Туре	rw	r	r rw						
FE <sub>H</sub>	CCU6_TRPCTRL Reset: 00 <sub>H</sub>	Bit Field			0			TRPM2	TRPM1	TRPM	
	Trap Control Register Low	Туре			r			rw	rw	rw	



### Table 12 CCU6 Register Overview (cont'd)

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
FF <sub>H</sub>	CCU6_TRPCTRH Reset: 00 <sub>H</sub>	Bit Field		TRPEN			TRF	PEN		
	Trap Control Register High		N	13						
		Туре	rw	rw			r	w		
	0, Page 3									
9A <sub>H</sub>	CCU6_MCMOUTL Reset: 00 <sub>H</sub> Multi-Channel Mode Output Register	Bit Field	0	R			MC	MP		
	Low	Туре	r	rh			r	h		
9B <sub>H</sub>	CCU6_MCMOUTH Reset: 00 <sub>H</sub> Multi-Channel Mode Output Register	Bit Field		0		CURH			EXPH	
	High	Туре		r		rh			rh	
9C <sub>H</sub>	CCU6_ISL Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status	Bit Field	T12PM	T12OM	ICC62F	ICC62 R	ICC61F	ICC61 R	ICC60F	ICC60 R
	Register Low	Туре	rh	rh	rh	rh	rh	rh	rh	rh
9D <sub>H</sub>	CCU6_ISH Reset: 00 <sub>H</sub> Capture/Compare Interrupt Status	Bit Field	STR	IDLE	WHE	CHE	TRPS	TRPF	T13PM	T13CM
	Register High	Туре	rh	rh	rh	rh	rh	rh	rh	rh
9E <sub>H</sub>	CCU6_PISEL0L Reset: 00 <sub>H</sub>	Bit Field	ISTRP		ISC	C62	ISCC61		ISC	C60
	Port Input Select Register 0 Low	Туре	rw rw r		w	r	w			
9F <sub>H</sub>	CCU6_PISEL0HReset: 00 <sub>H</sub> Port Input Select Register 0 High	Bit Field	IST1	2HR	ISP	OS2	ISP	OS1	ISPOS0	
		Туре	r	w	r	w	n	w	r	w
A4 <sub>H</sub>	CCU6_PISEL2 Reset: 00 <sub>H</sub>	Bit Field			(	C			IST1	3HR
	Port Input Select Register 2	Туре				r			r	w
FA <sub>H</sub>	CCU6_T12L Reset: 00 <sub>H</sub>	Bit Field				T12	CVL			
	Timer T12 Counter Register Low	Туре				rv	vh			
FB <sub>H</sub>	CCU6_T12H Reset: 00 <sub>H</sub>	Bit Field				T12	CVH			
	Timer T12 Counter Register High	Туре				rv	vh			
FC <sub>H</sub>	CCU6_T13L Reset: 00 <sub>H</sub>	Bit Field				T13	CVL			
	Timer T13 Counter Register Low	Туре				rv	vh			
FD <sub>H</sub>	CCU6_T13H Reset: 00 <sub>H</sub>	Bit Field	Field T13CVH							
	Timer T13 Counter Register High	Туре								
FE <sub>H</sub>	CCU6_CMPSTATL Reset: 00 <sub>H</sub> Compare State Register Low	Bit Field	0	CC63 ST	CCPO S2	CCPO S1	CCPO S0	CC62 ST	CC61 ST	CC60 ST
		Туре	r	rh	rh	rh	rh	rh	rh	rh
FF <sub>H</sub>	CCU6_CMPSTATH Reset: 00 <sub>H</sub> Compare State Register High	Bit Field	T13IM	COUT 63PS	COUT 62PS	CC62 PS	COUT 61PS	CC61 PS	COUT 60PS	CC60 PS
		Туре	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh

The SSC SFRs can be accessed in the standard memory area (RMAP = 0).

# Table 13 SSC Register Overview

Addr	Register Name		Bit	7	6	5	4	3	2	1	0
RMAP =	0										
A9 <sub>H</sub>	SSC_PISEL	Reset: 00 <sub>H</sub>	Bit Field			0			CIS	SIS	MIS
	Port Input Select Regist	er	Туре			r			rw	rw	rw
AA <sub>H</sub>	SSC_CONL	Reset: 00 <sub>H</sub>	Bit Field	LB	PO	PH	HB		В	М	
	Control Register Low Programming Mode		Туре	rw	rw	rw	rw		r	w	
	Operating Mode		Bit Field			0			В	С	
			Туре			r			r	h	



AB <sub>H</sub>	SSC_CONH Reset: 00 Control Register High	Bit Field	EN	EN MS 0 AREN BEN PEN REN						
	Programming Mode	Туре	rw	rw	r	rw	rw	rw	rw	rw
	Operating Mode	Bit Field	EN	MS	0	BSY	BE	PE	RE	TE
		Туре	rw	rw	r	rh	rwh	rwh	rwh	rwh
ACH	SSC_TBL Reset: 00	Bit Field				TB_V	ALUE			
	Transmitter Buffer Register Low	Туре				r	w			
AD <sub>H</sub>	SSC_RBL Reset: 00	Bit Field				RB_V	ALUE			
	Receiver Buffer Register Low	Туре				r	h			
AE <sub>H</sub>	SSC_BRL Reset: 00					BR_VAI	_UE[7:0]			
	Baudrate Timer Reload Register Low	Туре				r	w			
AF <sub>H</sub>	SSC_BRH Reset: 00					BR_VAL	UE[15:8]	]		
	Baudrate Timer Reload Register Hig	п Туре				r	w			

# Table 13 SSC Register Overview

The OCDS SFRs can be accessed in the mapped memory area (RMAP = 1).

# Table 14 OCDS Register Overview

Addr	Register Name	Bit	7	6	5	4	3	2	1	0
RMAP =	1			1	1	1	1	1		
E9 <sub>H</sub>	MMCR2 Reset: 0U <sub>H</sub> Monitor Mode Control Register 2	Bit Field	EXBC_ P	EXBC	MBCO N_P	MBCO N	MMEP _P	MMEP	MMOD E	JENA
		Туре	w	rw	w	rwh	w	rwh	rh	rh
F1 <sub>H</sub>	MMCR Reset: 00 <sub>H</sub> Monitor Mode Control Register	Bit Field	MEXIT _P	MEXIT	MSTEP _P	MSTEP	MRAM S_P	MRAM S	TRF	RRF
		Туре	w	rwh	w	rw	w	rwh	rh	rh
F2 <sub>H</sub>	MMSR Reset: 00 <sub>H</sub> Monitor Mode Status Register	Bit Field	MBCA M	MBCIN	EXBF	SWBF	HWB3 F	HWB2 F	HWB1 F	HWB0 F
		Туре	rw	rh	rwh	rwh	rwh	rwh	rwh	rwh
F3 <sub>H</sub>	MMBPCR Reset: 00 <sub>H</sub> BreakPoints Control Register	Bit Field	SWBC	HW	B3C	HWB2C		HWB1 C	HW	B0C
		Туре	rw	r	w n		w	rw	r	w
F4 <sub>H</sub>	MMICR Reset: 00 <sub>H</sub> Monitor Mode Interrupt Control Register	Bit Field	DVECT	DRETR	(	Ċ	MMUIE _P	MMUIE	RRIE_ P	RRIE
		Туре	rwh	rwh		r	w	rw	w	rw
F5 <sub>H</sub>	MMDR Reset: 00 <sub>H</sub> Monitor Mode Data Register	Bit Field				MN	IRR			
	Receive	Туре				r	h			
	Transmit	Bit Field				MN	ITR			
		Туре				١	v			
F6 <sub>H</sub>	HWBPSR Reset: 00 <sub>H</sub> Hardware Breakpoints Select Register	Bit Field		0		BPSEL _P		BP	SEL	
		Туре		r		w		r	w	
F7 <sub>H</sub>	HWBPDR Reset: 00 <sub>H</sub>	Bit Field				HWE	BPxx			
	Hardware Breakpoints Data Register	Туре				r	w			



# 3.3 Flash Memory

The Flash memory provides an embedded user-programmable non-volatile memory, allowing fast and reliable storage of user code and data. It is operated from a single 2.5 V supply from the Embedded Voltage Regulator (EVR) and does not require additional programming or erasing voltage. The sectorization of the Flash memory allows each sector to be erased independently.

#### Features

- In-System Programming (ISP) via UART
- In-Application Programming (IAP)
- Error Correction Code (ECC) for dynamic correction of single-bit errors
- Background program and erase operations for CPU load minimization
- Support for aborting erase operation
- Minimum program width<sup>1)</sup> of 32-byte for D-Flash and 32-byte for P-Flash
- 1-sector minimum erase width
- 1-byte read access
- Flash is delivered in erased state (read all zeros)
- Operating supply voltage: 2.5 V ± 7.5 %
- Read access time: 3 × t<sub>CCLK</sub> = 112.5 ns<sup>2)</sup>
- Program time: 209440 /  $f_{SYS} = 2.6 \text{ ms}^{3}$
- Erase time:  $8175360 / f_{SYS} = 102 \text{ ms}^{3}$

P-Flash: 32-byte wordline can only be programmed once, i.e., one gate disturb allowed. D-Flash: 32-byte wordline can be programmed twice, i.e., two gate disturbs allowed.

 $f_{\text{SVS}} = 80 \text{ MHz} \pm 7.5\%$  ( $f_{\text{CCLK}} = 26.7 \text{ MHz} \pm 7.5\%$ ) is the maximum frequency range for Flash read access.

<sup>&</sup>lt;sup>3)</sup>  $f_{sys} = 80 \text{ MHz} \pm 7.5\%$  is the only frequency range for Flash programming and erasing.  $f_{sysmin}$  is used for obtaining the worst case timing.



Table 15 shows the Flash data retention and endurance targets.

	Thash Bata Retention			
Retention	Endurance <sup>1)</sup>		Size	Remarks
		<i>T</i> <sub>A</sub> =- 40 to 125 °C	<i>T</i> <sub>A</sub> = 125 to 140 °C	

#### Table 15 Flash Data Retention and Endurance

### **Program Flash**

20 years	1,000 cycles	up to 16 Kbytes <sup>2)</sup>	for 16-Kbyte Variant
20 years	1,000 cycles	up to 8 Kbytes <sup>2)</sup>	for 8-Kbyte Variant
20 years	1,000 cycles	up to 4 Kbytes <sup>2)</sup>	for 4-Kbyte Variant

#### Data Flash

20 years	1,000 cycles <sup>3)</sup>	4 Kbytes	1 Kbytes	
5 years	10,000 cycles <sup>3)</sup>	1 Kbyte	256 bytes	
2 years	70,000 cycles <sup>3)</sup>	512 bytes	128 bytes	
2 years	100,000 cycles <sup>3)</sup>	128 bytes	32 bytes	

<sup>1)</sup> One cycle refers to the programming of all wordlines in a sector and erasing of sector. The Flash endurance data specified in Table 15 is valid only if the following conditions are fulfilled:

- the maximum number of erase cycles per Flash sector must not exceed 100,000 cycles.

- the maximum number of erase cycles per Flash bank must not exceed 300,000 cycles.

- the maximum number of program cycles per Flash bank must not exceed 2,500,000 cycles.

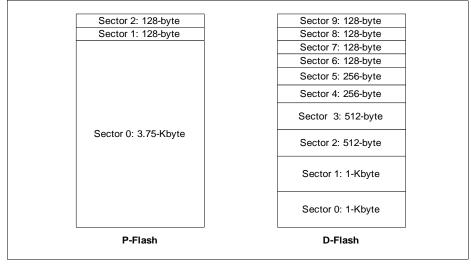
<sup>2)</sup> If no Flash is used for data, the Program Flash size can be up to the maximum Flash size available in the device variant. Having more Data Flash will mean less Flash is available for Program Flash.

 $^{3)}$  For T\_A=125 to 140°C, refers to programming of second 8 bytes (bytes 8 to 15) per WL



# 3.3.1 Flash Bank Sectorization

The SAA-XC866 product family offers four Flash devices with either 8 Kbytes or 16 Kbytes of embedded Flash memory. These Flash memory sizes are made up of two or four 4-Kbyte Flash banks, respectively. Each Flash device consists of Program Flash (P-Flash) bank(s) and a single Data Flash (D-Flash) bank with different sectorization shown in **Figure 11**. Both types can be used for code and data storage. The label "Data" neither implies that the D-Flash is mapped to the data memory region, nor that it can only be used for data storage. It is used to distinguish the different Flash bank sectorizations.



#### Figure 11 Flash Bank Sectorization

The internal structure of each Flash bank represents a sector architecture for flexible erase capability. The minimum erase width is always a complete sector, and sectors can be erased separately or in parallel. Contrary to standard EPROMs, erased Flash memory cells contain 0s.

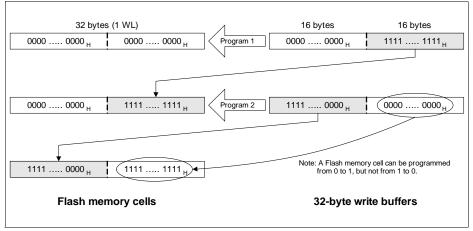
The D-Flash bank is divided into more physical sectors for extended erasing and reprogramming capability; even numbers for each sector size are provided to allow greater flexibility and the ability to adapt to a wide range of application requirements.



# 3.3.2 Flash Programming Width

For the P-Flash banks, a programmed wordline (WL) must be erased before it can be reprogrammed as the Flash cells can only withstand one gate disturb. This means that the entire sector containing the WL must be erased since it is impossible to erase a single WL.

For the D-Flash bank, the same WL can be programmed twice before erasing is required as the Flash cells are able to withstand two gate disturbs. Hence, it is possible to program the same WL, for example, with 16 bytes of data in two times (see Figure 12).



#### Figure 12 D-Flash Programming

Note: When programming a D-Flash WL the second time, the previously programmed Flash memory cells (whether 0s or 1s) should be reprogrammed with 0s to retain its original contents and to prevent "over-programming".

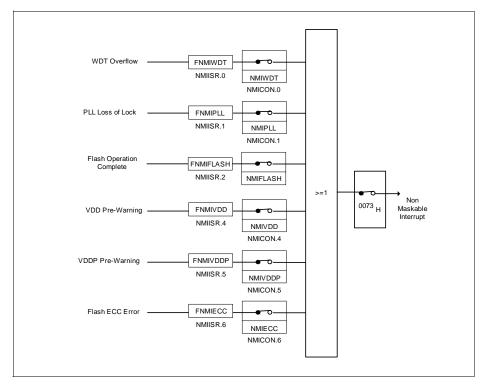


# 3.4 Interrupt System

The XC800 Core supports one non-maskable interrupt (NMI) and 14 maskable interrupt requests. In addition to the standard interrupt functions supported by the core, e.g., configurable interrupt priority and interrupt masking, the XC866 interrupt system provides extended interrupt support capabilities such as the mapping of each interrupt vector to several interrupt sources to increase the number of interrupt sources supported, and additional status registers for detecting and determining the interrupt source.

# 3.4.1 Interrupt Source

Figure 13 to Figure 17 give a general overview of the interrupt sources and illustrates the request and control flags.



#### Figure 13 Non-Maskable Interrupt Request Sources



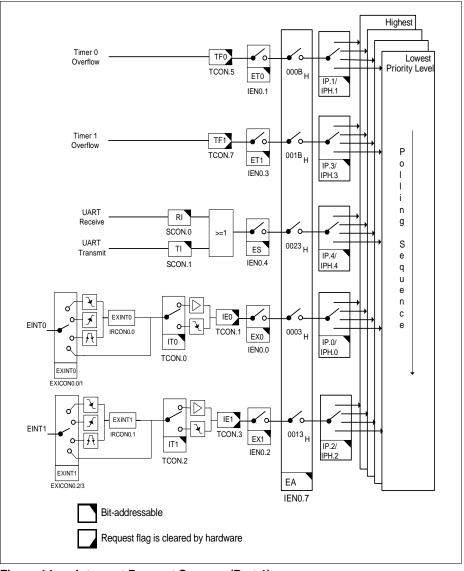


Figure 14 Interrupt Request Sources (Part 1)



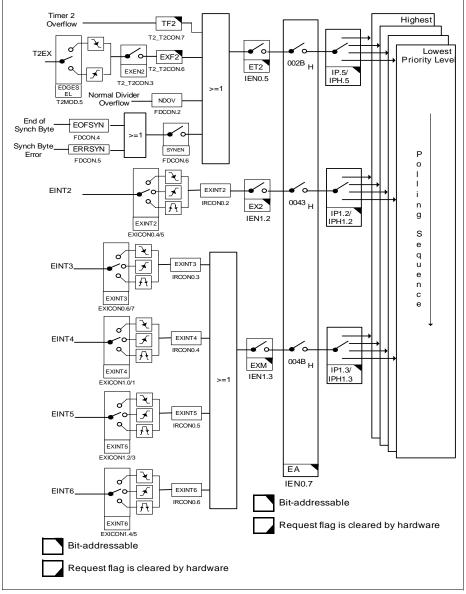


Figure 15 Interrupt Request Sources (Part 2)



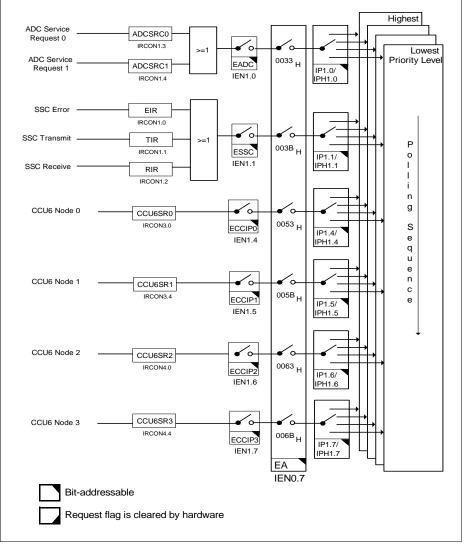


Figure 16 Interrupt Request Sources (Part 3)



# SAA-XC866

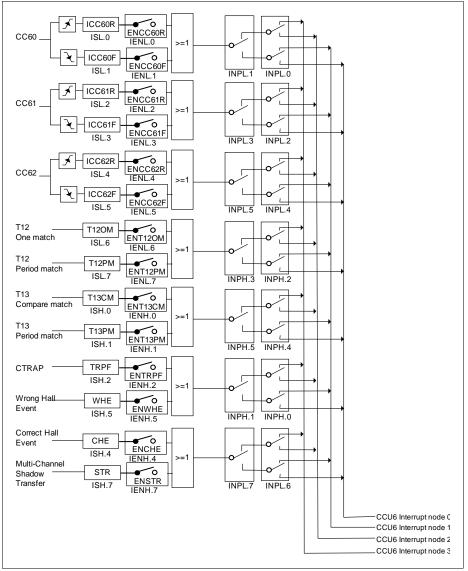


Figure 17 Interrupt Request Sources (Part 4)



# 3.4.2 Interrupt Source and Vector

Each interrupt source has an associated interrupt vector address. This vector is accessed to service the corresponding interrupt source request. The interrupt service of each interrupt source can be individually enabled or disabled via an enable bit. The assignment of the SAA-XC866 interrupt sources to the interrupt vector addresses and the corresponding interrupt source enable bits are summarized in Table 16.

Interrupt Source	Vector Address	Assignment for SAA- XC866	Enable Bit	SFR
NMI	0073 <sub>H</sub>	Watchdog Timer NMI	NMIWDT	NMICON
		PLL NMI	NMIPLL	
		Flash NMI	NMIFLASH	
		VDDC Prewarning NMI	NMIVDD	
		VDDP Prewarning NMI	NMIVDDP	
		Flash ECC NMI	NMIECC	
XINTR0	0003 <sub>H</sub>	External Interrupt 0	EX0	IEN0
XINTR1	000B <sub>H</sub>	Timer 0	ET0	
XINTR2	0013 <sub>H</sub>	External Interrupt 1	EX1	
XINTR3	001B <sub>H</sub>	Timer 1	ET1	
XINTR4	0023 <sub>H</sub>	UART	ES	
XINTR5	002B <sub>H</sub>	T2	ET2	
		Fractional Divider (Normal Divider Overflow)		
		LIN		

#### Table 16 Interrupt Vector Addresses



XINTR6	0033 <sub>H</sub>	ADC	EADC	IEN1	
XINTR7	003B <sub>H</sub>	SSC	ESSC		
XINTR8	0043 <sub>H</sub>	External Interrupt 2	EX2		
XINTR9	004B <sub>H</sub>	External Interrupt 3	EXM		
		External Interrupt 4			
		External Interrupt 5			
		External Interrupt 6			
XINTR10	0053 <sub>H</sub>	CCU6 INP0	ECCIP0		
XINTR11	005B <sub>H</sub>	CCU6 INP1	ECCIP1		
XINTR12	0063 <sub>H</sub>	CCU6 INP2	ECCIP2		
XINTR13	006B <sub>H</sub>	CCU6 INP3	ECCIP3		

# Table 16 Interrupt Vector Addresses (cont'd)



# 3.4.3 Interrupt Priority

Each interrupt source, except for NMI, can be individually programmed to one of the four possible priority levels. The NMI has the highest priority and supersedes all other interrupts. Two pairs of interrupt priority registers (IP and IPH, IP1 and IPH1) are available to program the priority level of each non-NMI interrupt vector.

A low-priority interrupt can be interrupted by a high-priority interrupt, but not by another interrupt of the same or lower priority. Further, an interrupt of the highest priority cannot be interrupted by any other interrupt source.

If two or more requests of different priority levels are received simultaneously, the request of the highest priority is serviced first. If requests of the same priority are received simultaneously, then an internal polling sequence determines which request is serviced first. Thus, within each priority level, there is a second priority structure determined by the polling sequence shown in Table 17.

Source	Level
Non-Maskable Interrupt (NMI)	(highest)
External Interrupt 0	1
Timer 0 Interrupt	2
External Interrupt 1	3
Timer 1 Interrupt	4
UART Interrupt	5
Timer 2, Fractional Divider, LIN Interrupts	6
ADC Interrupt	7
SSC Interrupt	8
External Interrupt 2	9
External Interrupt [6:3]	10
CCU6 Interrupt Node Pointer 0	11
CCU6 Interrupt Node Pointer 1	12
CCU6 Interrupt Node Pointer 2	13
CCU6 Interrupt Node Pointer 3	14

Table 17	Priority Structure within Interrupt Level
----------	---



# 3.5 Parallel Ports

The SAA-XC866 has 27 port pins organized into four parallel ports, Port 0 (P0) to Port 3 (P3). Each pin has a pair of internal pull-up and pull-down devices that can be individually enabled or disabled. Ports P0, P1 and P3 are bidirectional and can be used as general purpose input/output (GPIO) or to perform alternate input/output functions for the on-chip peripherals. When configured as an output, the open drain mode can be selected. Port P2 is an input-only port, providing general purpose input functions, alternate input functions for the on-chip peripherals, and also analog inputs for the Analog-to-Digital Converter (ADC).

#### **Bidirectional Port Features:**

- Configurable pin direction
- Configurable pull-up/pull-down devices
- Configurable open drain mode
- Transfer of data through digital inputs and outputs (general purpose I/O)
- Alternate input/output for on-chip peripherals

#### Input Port Features:

- Configurable input driver
- Configurable pull-up/pull-down devices
- Receive of data through digital input (general purpose input)
- Alternate input for on-chip peripherals
- Analog input for ADC module



# SAA-XC866

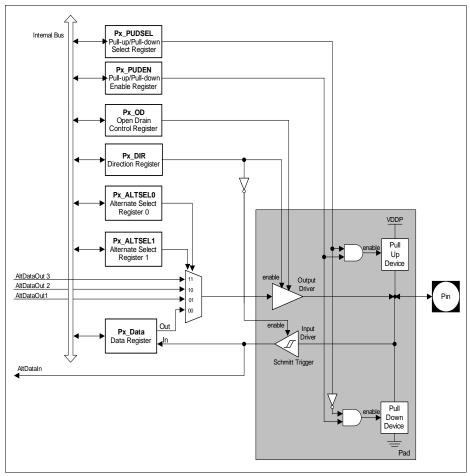


Figure 18 General Structure of Bidirectional Port





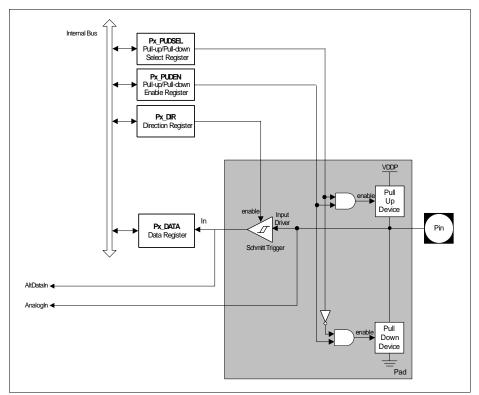


Figure 19 General Structure of Input Port



# 3.6 Power Supply System with Embedded Voltage Regulator

The SAA-XC866 microcontroller requires two different levels of power supply:

- 3.3 V or 5.0 V for the Embedded Voltage Regulator (EVR) and Ports
- 2.5 V for the core, memory, on-chip oscillator, and peripherals

**Figure 20** shows the SAA-XC866 power supply system. A power supply of 3.3 V or 5.0 V must be provided from the external power supply pin. The 2.5 V power supply for the logic is generated by the EVR. The EVR helps to reduce the power consumption of the whole chip and the complexity of the application board design.

The EVR consists of a main voltage regulator and a low power voltage regulator. In active mode, both voltage regulators are enabled. In power-down mode, the main voltage regulator is switched off, while the low power voltage regulator continues to function and provide power supply to the system with low power consumption.

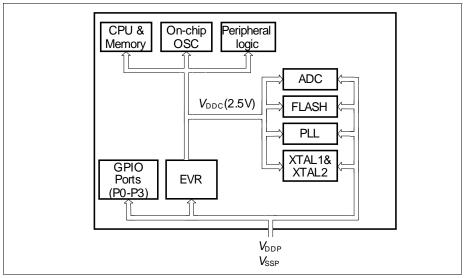


Figure 20 SAA-XC866 Power Supply System

#### **EVR Features:**

- Input voltage (V<sub>DDP</sub>): 3.3 V/5.0 V
- Output voltage (V<sub>DDC</sub>): 2.5 V ± 7.5%
- · Low power voltage regulator provided in power-down mode
- $V_{\text{DDC}}$  and  $V_{\text{DDP}}$  prewarning detection
- V<sub>DDC</sub> brownout detection



# 3.7 Reset Control

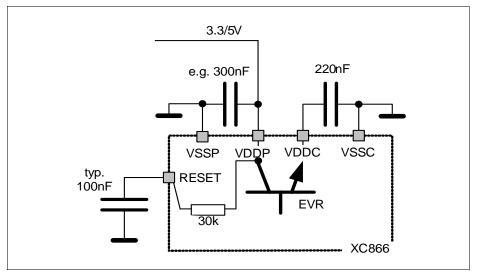
The SAA-XC866 has five types of reset: power-on reset, hardware reset, watchdog timer reset, power-down wake-up reset, and brownout reset.

When the SAA-XC866 is first powered up, the status of certain pins (see **Table 19**) must be defined to ensure proper start operation of the device. At the end of a reset sequence, the sampled values are latched to select the desired boot option, which cannot be modified until the next power-on reset or hardware reset. This guarantees stable conditions during the normal operation of the device.

In order to power up the system properly, the external reset pin RESET must be asserted until  $V_{DDC}$  reaches 0.9\*  $V_{DDC}$ . The delay of external reset can be realized by an external capacitor at RESET pin. This capacitor value must be selected so that  $V_{RESET}$  reaches 0.4 V, but not before  $V_{DDC}$  reaches 0.9\*  $V_{DDC}$ .

A typical application example is shown in Figure 21.  $V_{DDP}$  capacitor value is 300 nF.  $V_{DDC}$  capacitor value is 220 nF. The capacitor connected to RESET pin is 100 nF.

Typically, the time taken for  $V_{DDC}$  to reach  $0.9^*V_{DDC}$  is less than 50 µs once  $V_{DDP}$  reaches 2.3V. Hence, based on the condition that 10% to 90%  $V_{DDP}$  (slew rate) is less than 500 µs, the RESET pin should be held low for 500 µs typically. See Figure 22.







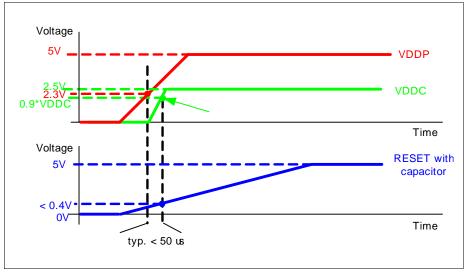


Figure 22 V<sub>DDP</sub>, V<sub>DDC</sub> and V<sub>RESET</sub> during Power-on Reset

The second type of reset in SAA-XC866 is the hardware reset. This reset function can be used during normal operation or when the chip is in power-down mode. A reset input pin RESET is provided for the hardware reset. To ensure the recognition of the hardware reset, pin RESET must be held low for at least 100 ns.

The Watchdog Timer (WDT) module is also capable of resetting the device if it detects a malfunction in the system.

Another type of reset that needs to be detected is a reset while the device is in power-down mode (wake-up reset). While the contents of the static RAM are undefined after a power-on reset, they are well defined after a wake-up reset from power-down mode.



# 3.7.1 Module Reset Behavior

**Table 18** shows how the functions of the SAA-XC866 are affected by the various reset types. A "∎" means that this function is reset to its default state.

Module/ Function	Wake-Up Reset	Watchdog Reset	Hardware Reset	Power-On Reset	Brownout Reset
CPU Core					
Peripherals					
On-Chip Static RAM	Not affected, reliable	Not affected, reliable	Not affected, reliable	Affected, un- reliable	Affected, un- reliable
Oscillator, PLL		Not affected			
Port Pins					
EVR	The voltage regulator is switched on	Not affected			
FLASH					
NMI	Disabled	Disabled			

Table 18 Effect of Reset on Device Functions

# 3.7.2 Booting Scheme

When the SAA-XC866 is reset, it must identify the type of configuration with which to start the different modes once the reset sequence is complete. Thus, boot configuration information that is required for activation of special modes and conditions needs to be applied by the external world through input pins. After power-on reset or hardware reset, the pins MBC, TMS and P0.0 collectively select the different boot options. Table 19 shows the available boot options in the SAA-XC866.

MBC	TMS	P0.0	Type of Mode	PC Start Value
1	0	х	User Mode; on-chip OSC/PLL non-bypassed	0000 <sub>H</sub>
0	0	х	BSL Mode; on-chip OSC/PLL non-bypassed	0000 <sub>H</sub>
0	1	0	OCDS Mode <sup>1)</sup> ; on-chip OSC/PLL non- bypassed	0000 <sub>H</sub>
1	1	0	Standalone User (JTAG) Mode <sup>2)</sup> ; on-chip OSC/PLL non-bypassed (normal)	0000 <sub>H</sub>

Table 19 SAA-XC866 Boot Selection

<sup>1)</sup> The OCDS mode is not accessible if Flash is protected.

<sup>2)</sup> Normal user mode with standard JTAG (TCK,TDI,TDO) pins for hot-attach purpose.



# 3.8 Clock Generation Unit

The Clock Generation Unit (CGU) allows great flexibility in the clock generation for the SAA-XC866. The power consumption is indirectly proportional to the frequency, whereas the performance of the microcontroller is directly proportional to the frequency. During user program execution, the frequency can be programmed for an optimal ratio between performance and power consumption. Therefore the power consumption can be adapted to the actual application state.

#### Features:

- Phase-Locked Loop (PLL) for multiplying clock source by different factors
- PLL Base Mode
- Prescaler Mode
- PLL Mode
- Power-down mode support

The CGU consists of an oscillator circuit and a PLL.In the SAA-XC866, the oscillator can be from either of these two sources: the on-chip oscillator (10 MHz) or the external oscillator (4 MHz to 12 MHz). The term "oscillator" is used to refer to both on-chip oscillator and external oscillator, unless otherwise stated. After the reset, the on-chip oscillator will be used by default. The external oscillator can be selected via software. In addition, the PLL provides a fail-safe logic to perform oscillator run and loss-of-lock detection. This allows emergency routines to be executed for system recovery or to perform system shut down.

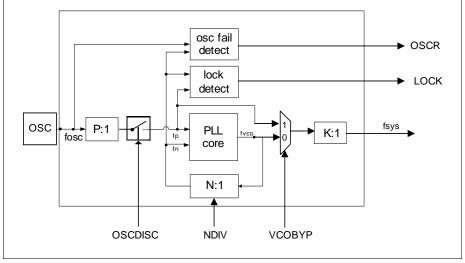


Figure 23 CGU Block Diagram



The clock system provides three ways to generate the system clock:

#### PLL Base Mode

The system clock is derived from the VCO base (free running) frequency clock divided by the K factor.

$$f_{SYS} = f_{VCObase} \times \frac{1}{K}$$

# Prescaler Mode (VCO Bypass Operation)

In VCO bypass operation, the system clock is derived from the oscillator clock, divided by the P and K factors.

$$f_{SYS} = f_{OSC} \times \frac{1}{P \times K}$$

#### PLL Mode

The system clock is derived from the oscillator clock, multiplied by the N factor, and divided by the P and K factors. Both VCO bypass and PLL bypass must be inactive for this PLL mode. The PLL mode is used during normal system operation.

$$f_{SYS} = f_{OSC} \times \frac{N}{P \times K}$$

 Table 20 shows the settings of bits OSCDISC and VCOBYP for different clock mode selection.

OSCDISC	VCOBYP	Clock Working Modes
0	0	PLL Mode
0	1	Prescaler Mode
1	0	PLL Base Mode
1	1	PLL Base Mode

### Table 20 Clock Mode Selection

Note: When oscillator clock is disconnected from PLL, the clock mode is PLL Base mode regardless of the setting of VCOBYP bit.

# **System Frequency Selection**

For the SAA-XC866, the values of P and K are fixed to "1" and "2", respectively. In order to obtain the required system frequency,  $f_{sys}$ , the value of N can be selected by bit NDIV for different oscillator inputs. Table 21 provides examples on how  $f_{sys} = 80$  MHz can be obtained for the different oscillator sources.



rasio 21 Oyston nequency (isys = 50 mile)					
Oscillator	f <sub>osc</sub>	Ν	Ρ	κ	f <sub>sys</sub>
On-chip	10 MHz	16	1	2	80 MHz
External	10 MHz	16	1	2	80 MHz
	8 MHz	20	1	2	80 MHz
	5 MHz	32	1	2	80 MHz

### Table 21 System frequency (f<sub>svs</sub> = 80 MHz)

Table 22 shows the VCO range for the SAA-XC866.

Table 22 VCO	Range
--------------	-------

f <sub>VCOmin</sub>	f <sub>VCOmax</sub>	f <sub>VCOFREEmin</sub>	f <sub>VCOFREEmax</sub>	Unit
150	200	20	80	MHz
100	150	10	80	MHz

# 3.8.1 Recommended External Oscillator Circuits

The oscillator circuit, a Pierce oscillator, is designed to work with both, an external crystal oscillator or an external stable clock source. It basically consists of an inverting amplifier and a feedback element with XTAL1 as input, and XTAL2 as output.

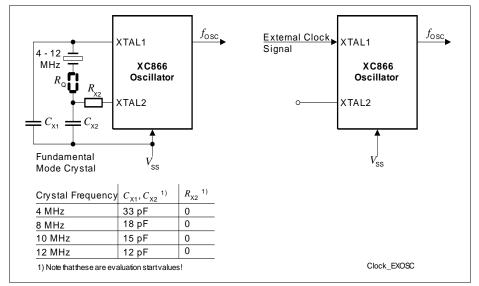
When using a crystal, a proper external oscillator circuitry must be connected to both pins, XTAL1 and XTAL2. The crystal frequency can be within the range of 4 MHz to 12 MHz. Additionally, it is necessary to have two load capacitances  $C_{X1}$  and  $C_{X2}$ , and depending on the crystal type, a series resistor  $R_{X2}$ , to limit the current. A test resistor  $R_Q$  may be temporarily inserted to measure the oscillation allowance (negative resistance) of the oscillator circuitry.  $R_Q$  values are typically specified by the crystal vendor. The  $C_{X1}$  and  $C_{X2}$  values shown in Figure 24 can be used as starting points for the negative resistance evaluation and for non-productive systems. The exact values and related operating range are dependent on the crystal frequency and have to be determined and optimized together with the final target system is strongly recommended to verify the input amplitude at XTAL1 and to determine the actual oscillation allowance (margin negative resistance) for the oscillator-crystal system.

When using an external clock signal, the signal must be connected to XTAL1. XTAL2 is left open (unconnected).

The oscillator can also be used in combination with a ceramic resonator. The final circuitry must also be verified by the resonator vendor.

Figure 24 shows the recommended external oscillator circuitries for both operating modes, external crystal mode and external input clock mode.





#### Figure 24 External Oscillator Circuitries

Note: For crystal operation, it is strongly recommended to measure the negative resistance in the final target system (layout) to determine the optimum parameters for the oscillator operation. Please refer to the minimum and maximum values of the negative resistance specified by the crystal supplier.



# 3.8.2 Clock Management

The CGU generates all clock signals required within the microcontroller from a single clock,  $f_{sys}$ . During normal system operation, the typical frequencies of the different modules are as follow:

- CPU clock: CCLK, SCLK = 26.7 MHz
- CCU6 clock: FCLK = 26.7 MHz
- Other peripherals: PCLK = 26.7 MHz
- Flash Interface clock: CCLK3 = 80 MHz and CCLK = 26.7 MHz

In addition, different clock frequency can output to pin CLKOUT(P0.0). The clock output frequency can further be divided by 2 using toggle latch (bit TLEN is set to 1), the resulting output frequency has 50% duty cycle. **Figure 25** shows the clock distribution of the SAA-XC866.

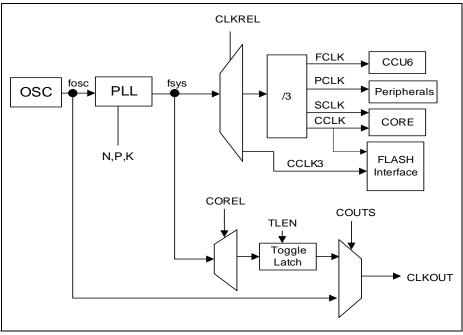


Figure 25 Clock Generation from f<sub>svs</sub>



For power saving purposes, the clocks may be disabled or slowed down according to **Table 23**.

# Table 23System frequency (f<sub>sys</sub> = 80 MHz)

Power Saving Mode	Action
Idle	Clock to the CPU is disabled.
Slow-down	Clocks to the CPU and all the peripherals, including CCU6, are divided by a common programmable factor defined by bit field CMCON.CLKREL.
Power-down	Oscillator and PLL are switched off.



# 3.9 Power Saving Modes

The power saving modes of the SAA-XC866 provide flexible power consumption through a combination of techniques, including:

- Stopping the CPU clock
- · Stopping the clocks of individual system components
- Reducing clock speed of some peripheral components
- · Power-down of the entire system with fast restart capability

After a reset, the active mode (normal operating mode) is selected by default (see **Figure 26**) and the system runs in the main system clock frequency. From active mode, different power saving modes can be selected by software. They are:

- Idle mode
- · Slow-down mode
- Power-down mode

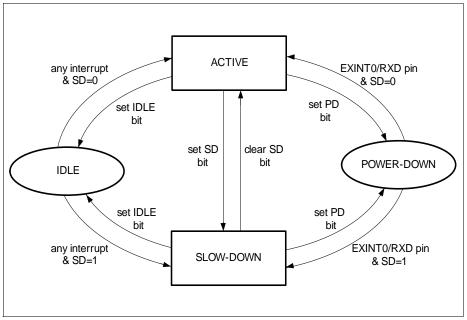


Figure 26 Transition between Power Saving Modes



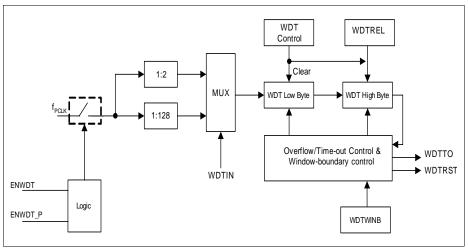
# 3.10 Watchdog Timer

The Watchdog Timer (WDT) provides a highly reliable and secure way to detect and recover from software or hardware failures. The WDT is reset at a regular interval that is predefined by the user. The CPU must service the WDT within this interval to prevent the WDT from causing an SAA-XC866 system reset. Hence, routine service of the WDT confirms that the system is functioning properly. This ensures that an accidental malfunction of the SAA-XC866 will be aborted in a user-specified time period. In debug mode, the WDT is suspended and stops counting. Therefore, there is no need to refresh the WDT during debugging.

#### Features:

- 16-bit Watchdog Timer
- Programmable reload value for upper 8 bits of timer
- Programmable window boundary
- Selectable input frequency of f<sub>PCLK</sub>/2 or f<sub>PCLK</sub>/128
- Time-out detection with NMI generation and reset prewarning activation (after which a system reset will be performed)

The WDT is a 16-bit timer incremented by a count rate of  $f_{PCLK}/2$  or  $f_{PCLK}/128$ . This 16-bit timer is realized as two concatenated 8-bit timers. The upper 8 bits of the WDT can be preset to a user-programmable value via a watchdog service access in order to modify the watchdog expire time period. The lower 8 bits are reset on each service access. Figure 27 shows the block diagram of the WDT unit.







If the WDT is not serviced before the timer overflow, a system malfunction is assumed. As a result, the WDT NMI is triggered (assert WDTTO) and the reset prewarning is entered. The prewarning period lasts for  $30_{\rm H}$  count, after which the system is reset (assert WDTRST).

The WDT has a "programmable window boundary" which disallows any refresh during the WDT's count-up. A refresh during this window boundary constitutes an invalid access to the WDT, causing the reset prewarning to be entered but without triggering the WDT NMI. The system will still be reset after the prewarning period is over. The window boundary is from  $0000_{\rm H}$  to the value obtained from the concatenation of WDTWINB and  $00_{\rm H}$ .

After being serviced, the WDT continues counting up from the value (<WDTREL>  $* 2^8$ ). The time period for an overflow of the WDT is programmable in two ways:

- the input frequency to the WDT can be selected to be either f<sub>PCLK</sub>/2 or f<sub>PCLK</sub>/128
- the reload value WDTREL for the high byte of WDT can be programmed in register WDTREL

The period,  $P_{WDT}$ , between servicing the WDT and the next overflow can be determined by the following formula:

$$P_{WDT} = \frac{2^{(1+WDTIN \times 6)} \times (2^{16} - WDTREL \times 2^8)}{f_{PCLK}}$$

If the Window-Boundary Refresh feature of the WDT is enabled, the period  $P_{WDT}$  between servicing the WDT and the next overflow is shortened if WDTWINB is greater than WDTREL, see Figure 28. This period can be calculated using the same formula by replacing WDTREL with WDTWINB. For this feature to be useful, WDTWINB should not be smaller than WDTREL.

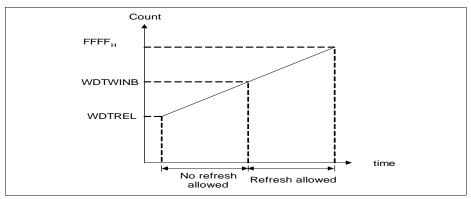


Figure 28 WDT Timing Diagram



 Table 24
 lists the possible watchdog time range that can be achieved for different module clock frequencies. Some numbers are rounded to 3 significant digits.

Reload value in WDTREL	Prescaler for f <sub>PCLK</sub>		
	2 (WDTIN = 0) 26.7 MHz	128 (WDTIN = 1) 26.7 MHz	
			FF <sub>H</sub>
7F <sub>H</sub>	2.48 ms	159 ms	
00 <sub>H</sub>	4.92 ms	315 ms	

Table 24	Watchdog Time Ranges
----------	----------------------



# 3.11 Universal Asynchronous Receiver/Transmitter

The Universal Asynchronous Receiver/Transmitter (UART) provides a full-duplex asynchronous receiver/transmitter, i.e., it can transmit and receive simultaneously. It is also receive-buffered, i.e., it can commence reception of a second byte before a previously received byte has been read from the receive register. However, if the first byte still has not been read by the time reception of the second byte is complete, one of the bytes will be lost.

#### Features:

- Full-duplex asynchronous modes
  - 8-bit or 9-bit data frames, LSB first
  - fixed or variable baud rate
- Receive buffered
- Multiprocessor communication
- Interrupt generation on the completion of a data transmission or reception

The UART can operate in four asynchronous modes as shown in **Table 25**. Data is transmitted on TXD and received on RXD.

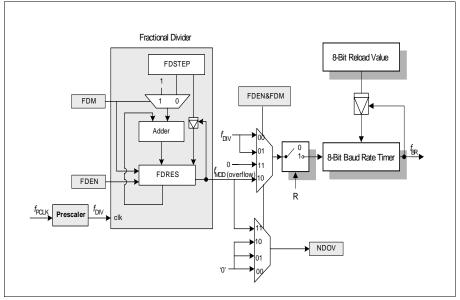
Operating Mode	Baud Rate
Mode 0: 8-bit shift register	f <sub>PCLK</sub> /2
Mode 1: 8-bit shift UART	Variable
Mode 2: 9-bit shift UART	f <sub>PCLK</sub> /32 or f <sub>PCLK</sub> /64
Mode 3: 9-bit shift UART	Variable

There are several ways to generate the baud rate clock for the serial port, depending on the mode in which it is operating. In mode 0, the baud rate for the transfer is fixed at  $f_{PCLK}/2$ . In mode 2, the baud rate is generated internally based on the UART input clock and can be configured to either  $f_{PCLK}/32$  or  $f_{PCLK}/64$ . The variable baud rate is set by either the underflow rate on the dedicated baud-rate generator, or by the overflow rate on Timer 1.



# 3.11.1 Baud-Rate Generator

The baud-rate generator is based on a programmable 8-bit reload value, and includes divider stages (i.e., prescaler and fractional divider) for generating a wide range of baud rates based on its input clock  $f_{PCLK}$ , see Figure 29.



#### Figure 29 Baud-rate Generator Circuitry

The baud rate timer is a count-down timer and is clocked by either the output of the fractional divider ( $f_{MOD}$ ) if the fractional divider is enabled (FDCON.FDEN = 1), or the output of the prescaler ( $f_{DIV}$ ) if the fractional divider is disabled (FDEN = 0). For baud rate generation, the fractional divider must be configured to fractional divider mode (FDCON.FDM = 0). This allows the baud rate control run bit BCON.R to be used to start or stop the baud rate timer. At each timer underflow, the timer is reloaded with the 8-bit reload value in register BG and one clock pulse is generated for the serial channel.

Enabling the fractional divider in normal divider mode (FDEN = 1 and FDM = 1) stops the baud rate timer and nullifies the effect of bit BCON.R. See Section 3.12.

The baud rate (f<sub>BR</sub>) value is dependent on the following parameters:

- Input clock f<sub>PCLK</sub>
- Prescaling factor (2<sup>BRPRE</sup>) defined by bit field BRPRE in register BCON
- Fractional divider (STEP/256) defined by register FDSTEP (to be considered only if fractional divider is enabled and operating in fractional divider mode)



• 8-bit reload value (BR\_VALUE) for the baud rate timer defined by register BG The following formulas calculate the final baud rate without and with the fractional divider respectively:

baud rate = 
$$\frac{f_{PCLK}}{16 \times 2^{BRPRE} \times (BR_VALUE + 1)}$$
 where  $2^{BRPRE} \times (BR_VALUE + 1) > 1$ 

baud rate = 
$$\frac{f_{PCLK}}{16 \times 2^{BRPRE} \times (BR VALUE + 1)} \times \frac{STEP}{256}$$

The maximum baud rate that can be generated is limited to  $f_{PCLK}/32$ . Hence, for a module clock of 26.7 MHz, the maximum achievable baud rate is 0.83 MBaud.

Standard LIN protocol can support a maximum baud rate of 20kHz, the baud rate accuracy is not critical and the fractional divider can be disabled. Only the prescaler is used for auto baud rate calculation. For LIN fast mode, which supports the baud rate of 20kHz to 115.2kHz, the higher baud rates require the use of the fractional divider for greater accuracy.

**Table 26** lists the various commonly used baud rates with their corresponding parameter settings and deviation errors. The fractional divider is disabled and a module clock of 26.7 MHz is used.

Baud rate	Prescaling Factor (2 <sup>BRPRE</sup> )	Reload Value (BR_VALUE + 1)	Deviation Error	
19.2 kBaud	1 (BRPRE=000 <sub>B</sub> )	87 (57 <sub>H</sub> )	-0.22 %	
9600 Baud	1 (BRPRE=000 <sub>B</sub> )	174 (AE <sub>H</sub> )	-0.22 %	
4800 Baud	2 (BRPRE=001 <sub>B</sub> )	174 (AE <sub>H</sub> )	-0.22 %	
2400 Baud	4 (BRPRE=010 <sub>B</sub> )	174 (AE <sub>H</sub> )	-0.22 %	

Table 26	Typical Baud rates for UART with Fractional Divider disabled

The fractional divider allows baud rates of higher accuracy (lower deviation error) to be generated. **Table 27** lists the resulting deviation errors from generating a baud rate of 115.2 kHz, using different module clock frequencies. The fractional divider is enabled (fractional divider mode) and the corresponding parameter settings are shown.



#### Table 27 Deviation Error for UART with Fractional Divider enabled

f <sub>PCLK</sub>	Prescaling Factor (2 <sup>BRPRE</sup> )	Reload Value (BR_VALUE + 1)	STEP	Deviation Error
26.67 MHz	1	10 (A <sub>H</sub> )	177 (B1 <sub>H</sub> )	+0.03 %
13.33 MHz	1	7 (7 <sub>H</sub> )	248 (F8 <sub>H</sub> )	+0.11 %
6.67 MHz	1	3 (3 <sub>H</sub> )	212 (D4 <sub>H</sub> )	-0.16 %



# 3.11.2 Baud Rate Generation using Timer 1

In UART modes 1 and 3, Timer 1 can be used for generating the variable baud rates. In theory, this timer could be used in any of its modes. But in practice, it should be set into auto-reload mode (Timer 1 mode 2), with its high byte set to the appropriate value for the required baud rate. The baud rate is determined by the Timer 1 overflow rate and the value of SMOD as follows:

[3.1]

Mode 1, 3 baud rate=  $\frac{2^{\text{SMOD}} \times f_{\text{PCLK}}}{32 \times 2 \times (256 - \text{TH1})}$ 

# 3.12 Normal Divider Mode (8-bit Auto-reload Timer)

Setting bit FDM in register FDCON to 1 configures the fractional divider to normal divider mode, while at the same time disables baud rate generation (see **Figure 29**). Once the fractional divider is enabled (FDEN = 1), it functions as an 8-bit auto-reload timer (with no relation to baud rate generation) and counts up from the reload value with each input clock pulse. Bit field RESULT in register FDRES represents the timer value, while bit field STEP in register FDSTEP defines the reload value. At each timer overflow, an overflow flag (FDCON.NDOV) will be set and an interrupt request generated. This gives an output clock  $f_{MOD}$  that is 1/n of the input clock  $f_{DIV}$ , where n is defined by 256 - STEP.

The output frequency in normal divider mode is derived as follows:

 $f_{MOD} = f_{DIV} \times \frac{1}{256 - STEP}$ [3.2]



# 3.13 LIN Protocol

The UART can be used to support the Local Interconnect Network (LIN) protocol for both master and slave operations. The LIN baud rate detection feature provides the capability to detect the baud rate within LIN protocol using Timer 2. This allows the UART to be synchronized to the LIN baud rate for data transmission and reception.

LIN is a holistic communication concept for local interconnected networks in vehicles. The communication is based on the SCI (UART) data format, a single-master/multipleslave concept, a clock synchronization for nodes without stabilized time base. An attractive feature of LIN is self-synchronization of the slave nodes without a crystal or ceramic resonator, which significantly reduces the cost of hardware platform. Hence, the baud rate must be calculated and returned with every message frame.

The structure of a LIN frame is shown in **Figure 30**. The frame consists of the:

- header, which comprises a Break (13-bit time low), Synch Byte (55<sub>H</sub>), and ID field
- response time
- data bytes (according to UART protocol)
- checksum

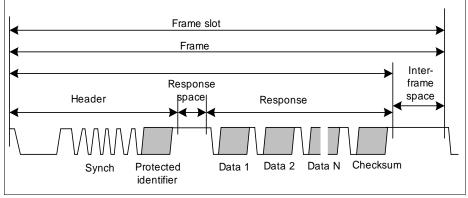


Figure 30 Structure of LIN Frame

# 3.13.1 LIN Header Transmission

LIN header transmission is only applicable in master mode. In the LIN communication, a master task decides when and which frame is to be transferred on the bus. It also identifies a slave task to provide the data transported by each frame. The information needed for the handshaking between the master and slave tasks is provided by the master task through the header portion of the frame.

The header consists of a break and synch pattern followed by an identifier. Among these three fields, only the break pattern cannot be transmitted as a normal 8-bit UART data.



The break must contain a dominant value of 13 bits or more to ensure proper synchronization of slave nodes.

In the LIN communication, a slave task is required to be synchronized at the beginning of the protected identifier field of frame. For this purpose, every frame starts with a sequence consisting of a break field followed by a synch byte field. This sequence is unique and provides enough information for any slave task to detect the beginning of a new frame and be synchronized at the start of the identifier field.

Upon entering LIN communication, a connection is established and the transfer speed (baud rate) of the serial communication partner (host) is automatically synchronized in the following steps:

STEP 1: Initialize interface for reception and timer for baud rate measurement

STEP 2: Wait for an incoming LIN frame from host

STEP 3: Synchronize the baud rate to the host

STEP 4: Enter for Master Request Frame or for Slave Response Frame

Note: Re-synchronization and setup of baud rate are always done for **every** Master Request Header or Slave Response Header LIN frame.



# 3.14 High-Speed Synchronous Serial Interface

The High-Speed Synchronous Serial Interface (SSC) supports full-duplex and half-duplex synchronous communication. The serial clock signal can be generated by the SSC internally (master mode), using its own 16-bit baud-rate generator, or can be received from an external master (slave mode). Data width, shift direction, clock polarity and phase are programmable. This allows communication with SPI-compatible devices or devices using other synchronous serial interfaces.

#### Features:

- Master and slave mode operation
  - Full-duplex or half-duplex operation
- Transmit and receive buffered
- Flexible data format
  - Programmable number of data bits: 2 to 8 bits
  - Programmable shift direction: LSB or MSB shift first
  - Programmable clock polarity: idle low or high state for the shift clock
  - Programmable clock/data phase: data shift with leading or trailing edge of the shift clock
- Variable baud rate
- · Compatible with Serial Peripheral Interface (SPI)
- Interrupt generation
  - On a transmitter empty condition
  - On a receiver full condition
  - On an error condition (receive, phase, baud rate, transmit error)



Data is transmitted or received on lines TXD and RXD, which are normally connected to the pins MTSR (Master Transmit/Slave Receive) and MRST (Master Receive/Slave Transmit). The clock signal is output via line MS\_CLK (Master Serial Shift Clock) or input via line SS\_CLK (Slave Serial Shift Clock). Both lines are normally connected to the pin SCLK. Transmission and reception of data are double-buffered.

Figure 31 shows the block diagram of the SSC.

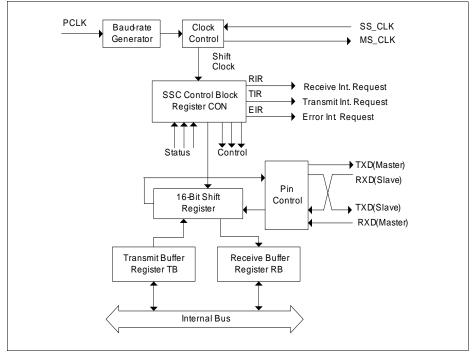


Figure 31 SSC Block Diagram



# 3.15 Timer 0 and Timer 1

Timers 0 and 1 are count-up timers which are incremented every machine cycle, or in terms of the input clock, every 2 PCLK cycles. They are fully compatible and can be configured in four different operating modes for use in a variety of applications, see **Table 28**. In modes 0, 1 and 2, the two timers operate independently, but in mode 3, their functions are specialized.

Mode	Operation
0	<b>13-bit timer</b> The timer is essentially an 8-bit counter with a divide-by-32 prescaler. This mode is included solely for compatibility with Intel 8048 devices.
1	<b>16-bit timer</b> The timer registers, TLx and THx, are concatenated to form a 16-bit counter.
2	<b>8-bit timer with auto-reload</b> The timer register TLx is reloaded with a user-defined 8-bit value in THx upon overflow.
3	Timer 0 operates as two 8-bit timersThe timer registers, TL0 and TH0, operate as two separate 8-bit counters.Timer 1 is halted and retains its count even if enabled.

#### Table 28 Timer 0 and Timer 1 Modes



# 3.16 Timer 2

Timer 2 is a 16-bit general purpose timer (THL2) that has two modes of operation, a 16-bit auto-reload mode and a 16-bit one channel capture mode. If the prescalar is disabled, Timer 2 counts with an input clock of PCLK/12. Timer 2 continues counting as long as it is enabled.

Table 29	Timer 2 Modes				
Mode	Description				
Auto-reload	<ul> <li>Up/Down Count Disabled</li> <li>Count up only</li> <li>Start counting from 16-bit reload value, overflow at FFFF<sub>H</sub></li> <li>Reload event configurable for trigger by overflow condition only, or by negative/positive edge at input pin T2EX as well</li> <li>Programmble reload value in register RC2</li> <li>Interrupt is generated with reload event</li> </ul>				
	<ul> <li>Up/Down Count Enabled</li> <li>Count up or down, direction determined by level at input pin T2EX</li> <li>No interrupt is generated</li> <li>Count up <ul> <li>Start counting from 16-bit reload value, overflow at FFFF<sub>H</sub></li> <li>Reload event triggered by overflow condition</li> <li>Programmble reload value in register RC2</li> </ul> </li> <li>Count down <ul> <li>Start counting from FFFF<sub>H</sub>, underflow at value defined in register RC2</li> <li>Reload event triggered by underflow condition <ul> <li>Reload event triggered by underflow condition</li> </ul> </li> </ul></li></ul>				
Channel capture	<ul> <li>Count up only</li> <li>Start counting from 0000<sub>H</sub>, overflow at FFFF<sub>H</sub></li> <li>Reload event triggered by overflow condition</li> <li>Reload value fixed at 0000<sub>H</sub></li> <li>Capture event triggered by falling/rising edge at pin T2EX</li> <li>Captured timer value stored in register RC2</li> <li>Interrupt is generated with reload or capture event</li> </ul>				



# 3.17 Capture/Compare Unit 6

The Capture/Compare Unit 6 (CCU6) provides two independent timers (T12, T13), which can be used for Pulse Width Modulation (PWM) generation, especially for AC-motor control. The CCU6 also supports special control modes for block commutation and multi-phase machines.

The timer T12 can function in capture and/or compare mode for its three channels. The timer T13 can work in compare mode only.

The multi-channel control unit generates output patterns, which can be modulated by T12 and/or T13. The modulation sources can be selected and combined for the signal modulation.

#### Timer T12 Features:

- Three capture/compare channels, each channel can be used either as a capture or as a compare channel
- Supports generation of a three-phase PWM (six outputs, individual signals for highside and lowside switches)
- 16-bit resolution, maximum count frequency = peripheral clock frequency
- Dead-time control for each channel to avoid short-circuits in the power stage
- Concurrent update of the required T12/13 registers
- Generation of center-aligned and edge-aligned PWM
- Supports single-shot mode
- · Supports many interrupt request sources
- Hysteresis-like control mode

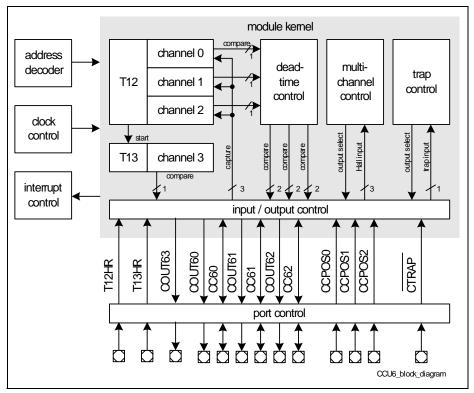
#### **Timer T13 Features:**

- · One independent compare channel with one output
- 16-bit resolution, maximum count frequency = peripheral clock frequency
- Can be synchronized to T12
- Interrupt generation at period-match and compare-match
- Supports single-shot mode

#### Additional Features:

- Implements block commutation for Brushless DC-drives
- Position detection via Hall-sensor pattern
- · Automatic rotational speed measurement for block commutation
- Integrated error handling
- Fast emergency stop without CPU load via external signal (CTRAP)
- · Control modes for multi-channel AC-drives
- Output levels can be selected and adapted to the power stage





The block diagram of the CCU6 module is shown in Figure 32.

Figure 32 CCU6 Block Diagram



# 3.18 Analog-to-Digital Converter

The SAA-XC866 includes a high-performance 10-bit Analog-to-Digital Converter (ADC) with eight multiplexed analog input channels. The ADC uses a successive approximation technique to convert the analog voltage levels from up to eight different sources. The analog input channels of the ADC are available at Port 2.

#### Features:

- Successive approximation
- 8-bit or 10-bit resolution (TUE of ± 1 LSB and ± 2 LSB, respectively)
- Eight analog channels
- Four independent result registers
- Result data protection for slow CPU access
   (wait-for-read mode)
- Single conversion mode
- Autoscan functionality
- Limit checking for conversion results
- Data reduction filter (accumulation of up to 2 conversion results)
- Two independent conversion request sources with programmable priority
- Selectable conversion request trigger
- · Flexible interrupt generation with configurable service nodes
- Programmable sample time
- Programmable clock divider
- · Cancel/restart feature for running conversions
- · Integrated sample and hold circuitry
- · Compensation of offset errors
- · Low power modes



# 3.18.1 ADC Clocking Scheme

A common module clock  ${\rm f}_{\rm ADC}$  generates the various clock signals used by the analog and digital parts of the ADC module:

- f<sub>ADCA</sub> is input clock for the analog part.
- f<sub>ADCI</sub> is internal clock for the analog part (defines the time base for conversion length and the sample time). This clock is generated internally in the analog part, based on the input clock f<sub>ADCA</sub> to generate a correct duty cycle for the analog components.
- f<sub>ADCD</sub> is input clock for the digital part.

The internal clock for the analog part  $f_{ADCI}$  is limited to a maximum frequency of 10 MHz. Therefore, the ADC clock prescaler must be programmed to a value that ensures  $f_{ADCI}$  does not exceed 10 MHz. The prescaler ratio is selected by bit field CTC in register GLOBCTR. A prescaling ratio of 32 can be selected when the maximum performance of the ADC is not required.

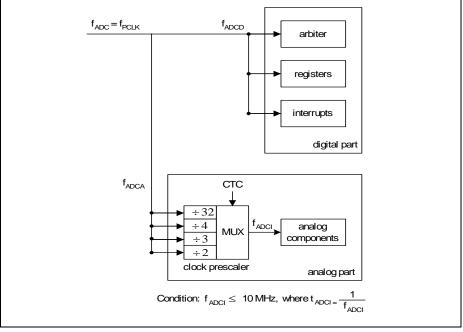


Figure 33 ADC Clocking Scheme



For module clock  $f_{ADC}$  = 26.7 MHz, the analog clock  $f_{ADCI}$  frequency can be selected as shown in **Table 30**.

Module Clock f <sub>ADC</sub>	СТС	Prescaling Ratio	Analog Clock f <sub>ADCI</sub>	
26.7 MHz	00 <sub>B</sub>	÷2	13.3 MHz (N.A)	
	01 <sub>B</sub>	÷ 3	8.9 MHz	
	10 <sub>B</sub>	÷ 4	6.7 MHz	
	11 <sub>B</sub> (default)	÷ 32	833.3 kHz	

# Table 30fFrequency Selection

As  $f_{ADCI}$  cannot exceed 10 MHz, bit field CTC should not be set to  $00_B$  when  $f_{ADC}$  is 26.7 MHz. During slow-down mode where  $f_{ADC}$  may be reduced to 13.3 MHz, 6.7 MHz etc., CTC can be set to  $00_B$  as long as the divided analog clock  $f_{ADCI}$  does not exceed 10 MHz. However, it is important to note that the conversion error could increase due to loss of charges on the capacitors, if  $f_{ADC}$  becomes too low during slow-down mode.

# 3.18.2 ADC Conversion Sequence

The analog-to-digital conversion procedure consists of the following phases:

- Synchronization phase (t<sub>SYN</sub>)
- Sample phase (t<sub>S</sub>)
- Conversion phase
- Write result phase (t<sub>WR</sub>)

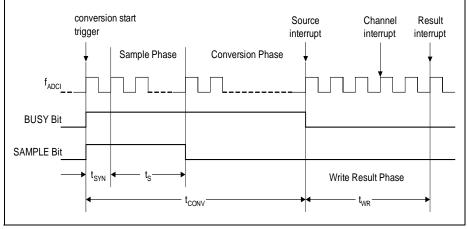


Figure 34 ADC Conversion Timing



# 3.19 On-Chip Debug Support

The On-Chip Debug Support (OCDS) provides the basic functionality required for the software development and debugging of XC800-based systems.

The OCDS design is based on these principles:

- use the built-in debug functionality of the XC800 Core
- add a minimum of hardware overhead
- provide support for most of the operations by a Monitor Program
- use standard interfaces to communicate with the Host (a Debugger)

#### Features:

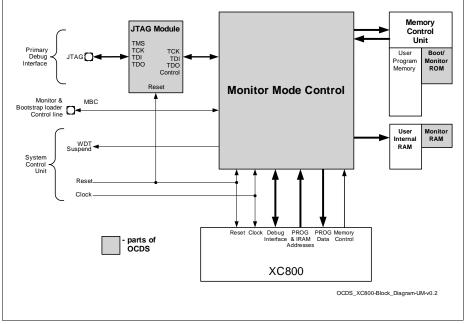
- Set breakpoints on instruction address and within a specified address range
- Set breakpoints on internal RAM address
- · Support unlimited software breakpoints in Flash/RAM code region
- Process external breaks
- Step through the program code

The OCDS functional blocks are shown in **Figure 35**. The Monitor Mode Control (MMC) block at the center of OCDS system brings together control signals and supports the overall functionality. The MMC communicates with the XC800 Core, primarily via the Debug Interface, and also receives reset and clock signals. After processing memory address and control signals from the core, the MMC provides proper access to the dedicated extra-memories: a Monitor ROM (holding the code) and a Monitor RAM (for work-data and Monitor-stack). The OCDS system is accessed through the JTAG<sup>1</sup>, which is an interface dedicated exclusively for testing and debugging activities and is not normally used in an application. The dedicated MBC pin is used for external configuration and debugging control.

Note: All the debug functionality described here can normally be used only after SAA-XC866 has been started in OCDS mode.

<sup>&</sup>lt;sup>1)</sup> The pins of the JTAG port can be assigned to either Port 0 (primary) or Ports 1 and 2 (secondary). User must set the JTAG pins (TCK and TDI) as input during connection with the OCDS system.







# 3.19.1 JTAG ID Register

This is a read-only register located inside the JTAG module, and is used to recognize the device(s) connected to the JTAG interface. Its content is shifted out when INSTRUCTION register contains the IDCODE command (opcode  $04_H$ ), and the same is also true immediately after reset.

The JTAG ID register contents for the SAA-XC866 devices are given in Table 31.

•			
Device Name	JTAG ID		
SAA-XC866L-4FRA	1010 0083 <sub>H</sub>		
SAA-XC866-4FRA	100F 5083 <sub>H</sub>		
SAA-XC866L-2FRA	1010 2083 <sub>H</sub>		
SAA-XC866-2FRA	1010 1083 <sub>H</sub>		
SAA-XC866L-1FRA	1013 8083 <sub>H</sub>		
SAA-XC866-1FRA	1013 8083 <sub>H</sub>		
	SAA-XC866L-4FRA SAA-XC866-4FRA SAA-XC866L-2FRA SAA-XC866-2FRA SAA-XC866L-1FRA		

#### Table 31JTAG ID Summary



#### Table 31JTAG ID Summary

ROM	SAA-XC866L-4RRA	1013 9083 <sub>H</sub>
	SAA-XC866-4RRA	1013 9083 <sub>H</sub>
	SAA-XC866L-2RRA	1013 9083 <sub>H</sub>
	SAA-XC866-2RRA	1013 9083 <sub>H</sub>

# 3.20 Identification Register

The SAA-XC866 identity register is located at Page 1 of address B3<sub>H</sub>.

#### ID

#### **Identity Register**

# Reset Value: 0000 0010<sub>B</sub>

7	6	5	4	3	2	1	0
	I	PRODID	I	I		VERID	I
	r					r	<u> </u> ]

Field	Bits	Туре	Description
VERID	[2:0]	r	Version ID 010 <sub>B</sub>
PRODID	[7:3]	r	Product ID 00000 <sub>B</sub>



# 4 Electrical Parameters

**Chapter 4** provides the characteristics of the electrical parameters which are implementation-specific for the SAA-XC866.

## 4.1 General Parameters

The general parameters are described here to aid the users in interpreting the parameters mainly in **Section 4.2** and **Section 4.3**.

#### 4.1.1 Parameter Interpretation

The parameters listed in this section represent partly the characteristics of the SAA-XC866 and partly its requirements on the system. To aid interpreting the parameters easily when evaluating them for a design, they are indicated by the abbreviations in the "Symbol" column:

• CC

These parameters indicate **C**ontroller **C**haracteristics, which are distinctive features of the SAA-XC866 and must be regarded for a system design.

• SR

These parameters indicate **S**ystem **R**equirements, which must be provided by the microcontroller system in which the SAA-XC866 is designed in.



## 4.1.2 Absolute Maximum Rating

Maximum ratings are the extreme limits to which the SAA-XC866 can be subjected to without permanent damage.

Table 32	Absolute Maximum Rating Parameters
----------	------------------------------------

Parameter	Symbol	Limit Values		Unit	Notes	
		min.	min. max.			
Ambient temperature	T <sub>A</sub>	-40	140	°C	under bias	
Storage temperature	T <sub>ST</sub>	-65	150	°C	1)	
Junction temperature	TJ	-40	150	°C	under bias <sup>1)</sup>	
Voltage on power supply pin with respect to $V_{\rm SS}$	$V_{DDP}$	-0.5	6	V	1)	
Voltage on core supply pin with respect to $V_{\rm SS}$	V <sub>DDC</sub>	-0.5	3.25	V	1)	
Voltage on any pin with respect to $V_{\rm SS}$	V <sub>IN</sub>	-0.5	V <sub>DDP</sub> + 0.5 or max. 6	V	Whichever is lower <sup>1)</sup>	
Input current on any pin during overload condition	I <sub>IN</sub>	-10	10	mA	1)	
Absolute sum of all input currents during overload condition	$\Sigma  I_{IN} $	-	50	mA	1)	

<sup>1)</sup> Not subjected to production test, verified by design/characterization.

Note: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. During absolute maximum rating overload conditions ( $V_{IN} > V_{DDP}$  or  $V_{IN} < V_{SS}$ ) the voltage on  $V_{DDP}$  pin with respect to ground ( $V_{SS}$ ) must not exceed the values defined by the absolute maximum ratings.



# 4.1.3 Operating Conditions

The following operating conditions must not be exceeded in order to ensure correct operation of the SAA-XC866. All parameters mentioned in the following table refer to these operating conditions, unless otherwise noted.

#### Table 33 Operating Condition Parameters

Parameter	Symbol	Lim	it Values	Unit	Notes/	
		min.	max.		Conditions	
Digital power supply voltage	V <sub>DDP</sub>	4.5	5.5	V	5V Device	
Digital power supply voltage	V <sub>DDP</sub>	3.0	3.6	V	3.3V Device	
Digital ground voltage	V <sub>SS</sub>		0	V		
Digital core supply voltage	V <sub>DDC</sub>	2.3	2.7	V		
System Clock Frequency <sup>1)</sup>	fsys	74	86	MHz		
Ambient temperature	T <sub>A</sub>	-40	140	°C	SAA-XC866	

<sup>1)</sup>  $f_{SYS}$  is the PLL output clock. During normal operating mode, CPU clock is  $f_{SYS}$  / 3. Please refer to Figure 25 for detailed description.



# 4.2 DC Parameters

#### 4.2.1 Input/Output Characteristics

# Table 34 Input/Output Characteristics (Operating Conditions apply)

Parameter	Symbol	Limit	Values	Unit	Test Conditions	
		min.	max.			
$V_{\text{DDP}}$ = 5V Range				1		
Output low voltage	V <sub>OL</sub> CC	-	1.0	V	I <sub>OL</sub> = 15 mA	
		-	0.4	V	I <sub>OL</sub> = 5 mA	
Output high voltage	V <sub>OH</sub> CC	V <sub>DDP</sub> - 1.0	-	V	I <sub>OH</sub> = -15 mA	
		V <sub>DDP</sub> - 0.4	-	V	I <sub>OH</sub> = -5 mA	
Input low voltage on port pins (all except P0.0 & P0.1)	V <sub>ILP</sub> SR	-	$0.3  imes V_{ m DDP}$	V	CMOS Mode	
Input low voltage on P0.0 & P0.1	$V_{\sf ILP0}$ SR	-0.2	$0.3  imes V_{ extsf{DDP}}$	V	CMOS Mode	
Input low voltage on RESET pin	V <sub>ILR</sub> SR	-	$0.3  imes V_{ m DDP}$	V	CMOS Mode	
Input low voltage on TMS pin	V <sub>ILT</sub> SR	-	$0.3  imes V_{ ext{DDP}}$	V	CMOS Mode	
Input high voltage on port pins (all except P0.0 & P0.1)	V <sub>IHP</sub> SR	$0.7  imes V_{ m DDP}$	-	V	CMOS Mode	
Input high voltage on P0.0 & P0.1	$V_{IHP0}SR$	$0.7  imes V_{ extsf{DDP}}$	$V_{DDP}$	V	CMOS Mode	
Input high voltage on RESET pin	V <sub>IHR</sub> SR	$0.7 \times V_{ m DDP}$	-	V	CMOS Mode	
Input high voltage on TMS pin	V <sub>IHT</sub> SR	$0.75  imes V_{ m DDP}$	-	V	CMOS Mode	
Input Hysteresis <sup>1)</sup> on Port Pins	HYS CC	$0.08 \times V_{ m DDP}$	-	V	CMOS Mode	
Input Hysteresis <sup>1)</sup> on XTAL1	HYSXCC	$0.07 \times V_{ m DDC}$	-	V		



#### Table 34 Input/Output Characteristics (Operating Conditions apply)

Parameter	Symbol		Limit	Values	Unit	Test Conditions	
				max.			
Input low voltage at XTAL1	$V_{ILX}$	SR	V <sub>SS</sub> - 0.5	$0.3  imes V_{ m DDC}$	V		
Input high voltage at XTAL1	$V_{IHX}$	SR	$0.7  imes V_{ m DDC}$	V <sub>DDC</sub> + 0.5	V		
Pull-up current	I <sub>PU</sub>	SR	-	-10	μA	V <sub>IH,min</sub>	
			-150	-	μA	V <sub>IL,max</sub>	
Pull-down current	I <sub>PD</sub>	SR	-	10	μA	V <sub>IL,max</sub>	
			150	-	μA	V <sub>IH,min</sub>	
Input leakage current <sup>2)</sup>	I <sub>OZ1</sub>	CC	-2	2	μA	$0 < V_{IN} < V_{DDP},$ $T_A \le 140^{\circ}C$	
Input current at XTAL1	I <sub>ILX</sub>	CC	-10	10	μA		
Overload current on any pin	I <sub>OV</sub>	SR	-5	5	mA	3)	
Absolute sum of overload currents	$\Sigma   I_{OV}$	 SR	-	25	mA	3)	
Voltage on any pin during $V_{\text{DDP}}$ power off	V <sub>PO</sub>	SR	-	0.3	V	4)	
Maximum current per pin (excluding $V_{\text{DDP}}$ and $V_{\text{SS}}$ )	I <sub>M</sub>	SR	-	15	mA		
Maximum current for all pins (excluding $V_{\rm DDP}$ and $V_{\rm SS}$ )	$\Sigma  I_{M} $	SR	-	60	mA		
Maximum current into $V_{\text{DDP}}$	I <sub>MVDE</sub>	SR	-	80	mA	3)	
Maximum current out of $V_{\rm SS}$	I <sub>MVSS</sub>	SR	_	80	mA	3)	



#### Table 34 Input/Output Characteristics (Operating Conditions apply)

Parameter	Symbol		Limit	Values	Unit	<b>Test Conditions</b>
			min.	max.		
V <sub>DDP</sub> = 3.3V Range						
Output low voltage	$V_{OL}$	СС	-	1.0	V	I <sub>OL</sub> = 8 mA
			-	0.4	V	I <sub>OL</sub> = 2.5 mA
Output high voltage	V <sub>OH</sub>	СС	V <sub>DDP</sub> - 1.0	-	V	<i>I</i> <sub>OH</sub> = -8 mA
			V <sub>DDP</sub> - 0.4	-	V	I <sub>OH</sub> = -2.5 mA
nput low voltage on port pins (all except P0.0 & P0.1)	V <sub>ILP</sub>	SR	-	$0.3  imes V_{ m DDP}$	V	CMOS Mode
Input low voltage on P0.0 & P0.1	V <sub>ILP0</sub>	SR	-0.2	$0.3  imes V_{ extsf{DDP}}$	V	CMOS Mode
nput low voltage on RESET pin	$V_{ILR}$	SR	-	$0.3  imes V_{ extsf{DDP}}$	V	CMOS Mode
nput low voltage on MS pin	$V_{ILT}$	SR	-	$0.3 \times V_{\text{DDP}}$	V	CMOS Mode
nput high voltage on ort pins all except P0.0 & P0.1)	V <sub>IHP</sub>	SR	$0.7 \times V_{ m DDP}$	-	V	CMOS Mode
put high voltage on 0.0 & P0.1	V <sub>IHP0</sub>	SR	$0.7 \times V_{\text{DDP}}$	$V_{DDP}$	V	CMOS Mode
iput high voltage on ESET pin	V <sub>IHR</sub>	SR	$0.7 \times V_{\text{DDP}}$	-	V	CMOS Mode
nput high voltage on MS pin	V <sub>IHT</sub>	SR	$0.75  imes V_{ m DDP}$	-	V	CMOS Mode
put Hysteresis <sup>1)</sup> on ort Pins	HYS	СС	$0.03  imes V_{ m DDP}$	-	V	CMOS Mode
put Hysteresis <sup>1)</sup> on TAL1	HYSX	CC	$0.07  imes V_{ m DDC}$	-	V	
nput low voltage at TAL1	$V_{ILX}$	SR	V <sub>SS</sub> - 0.5	$0.3  imes V_{ m DDC}$	V	
nput high voltage at (TAL1	V <sub>IHX</sub>	SR	$0.7 \times V_{ m DDC}$	V <sub>DDC</sub> + 0.5	V	



#### Table 34 Input/Output Characteristics (Operating Conditions apply)

Parameter	Symb	ool	Limit	Values	Unit	<b>Test Conditions</b>
			min.	max.		
Pull-up current	I <sub>PU</sub>	SR	-	-5	μA	V <sub>IH,min</sub>
			-50	-	μA	$V_{IL,max}$
Pull-down current	$I_{PD}$	SR	-	5	μA	$V_{IL,max}$
			50	-	μA	V <sub>IH,min</sub>
Input leakage current <sup>2)</sup>	I <sub>OZ1</sub>	CC	-2	2	μA	$0 < V_{IN} < V_{DDP},$ $T_A \le 140^{\circ}C$
Input current at XTAL1	$I_{ILX}$	CC	- 10	10	μA	
Overload current on any pin	I <sub>OV</sub>	SR	-5	5	mA	3)
Absolute sum of overload currents	$\Sigma  I_{OV} $	SR	_	25	mA	3)
Voltage on any pin during $V_{DDP}$ power off	V <sub>PO</sub>	SR	-	0.3	V	4)
Maximum current per bin (excluding $V_{\rm DDP}$ and $V_{\rm SS}$ )	I <sub>M</sub>	SR	-	15	mA	
Maximum current for all pins (excluding $V_{\text{DDP}}$ and $V_{\text{SS}}$ )	$\Sigma  I_{M} $	SR	-	60	mA	
Maximum current into $V_{DDP}$	I <sub>MVDE</sub>	SR	-	80	mA	
Maximum current out of $V_{\rm SS}$	I <sub>MVSS</sub>	SR	_	80	mA	

<sup>1)</sup> Not subjected to production test, verified by design/characterization. Hysteresis is implemented to avoid meta stable states and switching due to internal ground bounce. It cannot be guaranteed that it suppresses switching due to external system noise.

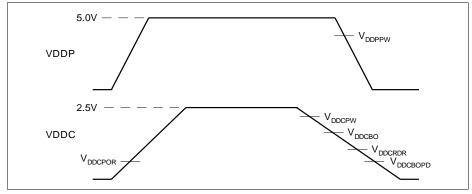
<sup>2)</sup> An additional error current ( $l_{IN,I}$ ) will flow if an overload current flows through an adjacent pin. TMS pin and RESET pin have internal pull devices and are not included in the input leakage current characteristic.

<sup>3)</sup> Not subjected to production test, verified by design/characterization.

<sup>4)</sup> Not subjected to production test, verified by design/characterization. However, for applications with strict low power-down current requirements, it is mandatory that no active voltage source is supplied at any GPIO pin when VDDP is powered off.







# 4.2.2 Supply Threshold Characteristics

Figure 36 Supply Threshold Parameters

#### Table 35 Supply Threshold Parameters (Operating Conditions apply)

Parameters	Symbol		Limit Values			
		min	n. typ.	max.		
V <sub>DDC</sub> prewarning voltage <sup>1)</sup>	V <sub>DDCPW</sub> C	C 2.2	2.3	2.4	V	
$V_{\text{DDC}}$ brownout voltage in active mode <sup>1)</sup>	V <sub>DDCBO</sub> C	C 2.0	2.1	2.2	V	
RAM data retention voltage	V <sub>DDCRDR</sub> C	C 0.9	1.0	1.1	V	
$V_{\text{DDC}}$ brownout voltage in power-down mode <sup>2)</sup>	V <sub>DDCBOPD</sub> C	C 1.3	1.5	1.7	V	
V <sub>DDP</sub> prewarning voltage <sup>3)</sup>	V <sub>DDPPW</sub> C	C 3.3	4.0	4.65	V	
Power-on reset voltage <sup>2)4)</sup>	V <sub>DDCPOR</sub> C	C 1.3	1.5	1.7	V	

<sup>1)</sup> Detection is disabled in power-down mode.

<sup>2)</sup> Detection is enabled in both active and power-down mode.

<sup>3)</sup> Detection is enabled for external power supply of 5.0V Detection must be disabled for external power supply of 3.3V.

<sup>4)</sup> The reset of EVR is extended by 300 µs typically after the VDDC reaches the power-on reset voltage.



# 4.2.3 ADC Characteristics

The values in the table below are given for an analog power supply between 4.5 V to 5.5 V. The ADC can be used with an analog power supply down to 3 V. But in this case, the analog parameters may show a reduced performance. All ground pins ( $V_{SS}$ ) must be externally connected to one single star point in the system. The voltage difference between the ground pins must not exceed 200mV.

Parameter	Symbol	Li	mit Valu	es	Unit	Test Conditions/ Remarks	
		min.	typ.	max.			
Analog reference voltage	V <sub>AREF</sub> SR	V <sub>AGND</sub> + 1	V <sub>DDP</sub>	V <sub>DDP</sub> + 0.05	V	1)	
Analog reference ground	V <sub>AGND</sub> SR	V <sub>SS</sub> - 0.05	V <sub>SS</sub>	V <sub>AREF</sub> - 1	V	1)	
Analog input voltage range	V <sub>AIN</sub> SR	V <sub>AGND</sub>	-	V <sub>AREF</sub>	V		
ADC clocks	f <sub>ADC</sub>	-	20	40	MHz	module clock <sup>1)</sup>	
	f <sub>ADCI</sub>	-	-	10	MHz	internal analog clock <sup>1)</sup> See <b>Figure 33</b>	
Sample time	t <sub>S</sub> CC	(2 + INI t <sub>ADCI</sub>	PCR0.S	TC) ×	μs	1)	
Conversion time	t <sub>C</sub> CC	See Se	ction 4.	2.3.1	μs	1)	
Total unadjusted	TUE CC	_	-	1	LSB	8-bit conversion. <sup>2)</sup>	
error		_	-	2	LSB	10-bit conversion. <sup>2)</sup>	
Differential Nonlinearity	EA <sub>DNL</sub>   CC	-	1	-	LSB	10-bit conversion <sup>1)</sup>	
Integral Nonlinearity	EA <sub>INL</sub>   CC	-	1	-	LSB	10-bit conversion <sup>1)</sup>	
Offset	EA <sub>OFF</sub>   CC	-	1	-	LSB	10-bit conversion <sup>1)</sup>	
Gain	EA <sub>GAIN</sub>   CC	-	1	-	LSB	10-bit conversion <sup>1)</sup>	
Overload current coupling factor for	K <sub>OVA</sub> CC	-	-	1.0 x 10 <sup>-4</sup>	-	$I_{\rm OV} > 0^{1)3)$	
analog inputs		-	-	1.5 x 10 <sup>-3</sup>	-	$I_{\rm OV} < 0^{1)3)}$	

#### Table 36ADC Characteristics (Operating Conditions apply; $V_{DDP} = 5V$ Range)



Table 36	ADC Characteristics (Operating Conditions apply; $V_{\text{DDP}}$ = 5V Range)
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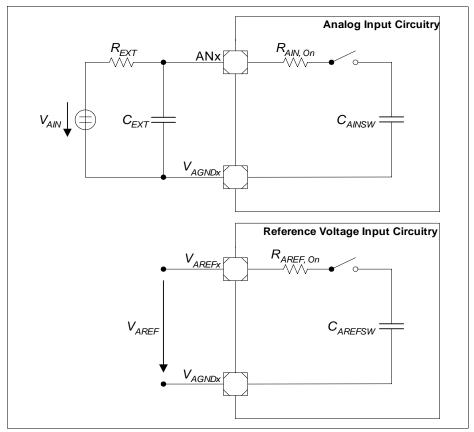
Parameter	Symbol	Li	mit Valu	es	Unit	Test Conditions/	
		min.	typ.	max.		Remarks	
Overload current coupling factor for	K <sub>OVD</sub> CC	-	-	5.0 x 10 <sup>-3</sup>	-	$I_{\rm OV} > 0^{1)3)$	
digital I/O pins		-	-	1.0 x 10 <sup>-2</sup>	-	$I_{\rm OV} < 0^{1)3)}$	
Switched capacitance at the reference voltage input	C <sub>AREFSW</sub> CC	-	10	20	pF	1)4)	
Switched capacitance at the analog voltage inputs	C <sub>AINSW</sub> CC	-	5	7	pF	1)5)	
Input resistance of the reference input	R <sub>AREF</sub> CC	-	1	2	kΩ	1)	
Input resistance of the selected analog channel	R <sub>AIN</sub> CC	-	1	1.5	kΩ	1)	

<sup>1)</sup> Not subject to production test, verified by design/characterization.

<sup>2)</sup> TUE is tested at  $V_{AREF} = 5.0 \text{ V}$ ,  $V_{AGND} = 0 \text{ V}$ ,  $V_{DDP} = 5.0 \text{ V}$ .

- <sup>3)</sup> An overload current ( $I_{OV}$ ) through a pin injects a certain error current ( $I_{INJ}$ ) into the adjacent pins. This error current adds to the respective pin's leakage current ( $I_{OZ}$ ). The amount of error current depends on the overload current and is defined by the overload coupling factor  $K_{OV}$ . The polarity of the injected error current is inverse compared to the polarity of the overload current that produces it. The total current through a pin is  $|I_{TOT}| = |I_{OZ1}| + (|I_{OV}| \times K_{OV})$ . The additional error current may distort the input voltage on analog inputs.
- <sup>4)</sup> This represents an equivalent switched capacitance. This capacitance is not switched to the reference voltage at once. Instead of this, smaller capacitances are successively switched to the reference voltage.
- <sup>5)</sup> The sampling capacity of the conversion C-Network is pre-charged to  $V_{AREF}/2$  before connecting the input to the C-Network. Because of the parasitic elements, the voltage measured at ANx is lower than  $V_{AREF}/2$ .







# 4.2.3.1 ADC Conversion Timing

Conversion time,  $t_C = t_{ADC} \times (1 + r \times (3 + n + STC))$ , where r = CTC + 2 for  $CTC = 00_B$ ,  $01_B$  or  $10_B$ , r = 32 for  $CTC = 11_B$ , CTC = Conversion Time Control (GLOBCTR.CTC), STC = Sample Time Control (INPCR0.STC), n = 8 or 10 (for 8-bit and 10-bit conversion respectively),  $t_{ADC} = 1 / f_{ADC}$ 



#### 4.2.4 Power Supply Current

Table 37	Power Supply Current Parameters (Operating Conditions and	(ylqc

Parameter	Symbol	Limit	Values	Unit	<b>Test Condition</b>	
		typ. <sup>1)</sup>	max. <sup>2)</sup>			
Active Mode	I <sub>DDP</sub>	22.6	24.5	mA	3)	
Idle Mode	I <sub>DDP</sub>	17.2	19.7	mA	4)	
Active Mode with slow-down enabled	I <sub>DDP</sub>	7.2	8.2	mA	5)	
Idle Mode with slow-down enabled	I <sub>DDP</sub>	7.1	8	mA	6)	

<sup>1)</sup> The typical  $I_{\text{DDP}}$  values are periodically measured at  $T_{\text{A}}$  = + 25 °C and  $V_{\text{DDP}}$  = 5.0 V.

- <sup>2)</sup> The maximum  $I_{\text{DDP}}$  values are measured under worst case conditions ( $T_{\text{A}}$  = + 140 °C and  $V_{\text{DDP}}$  = 5.5 V).
- <sup>3)</sup> I<sub>DDP</sub> (active mode) is measured with: CPU clock and input clock to all peripherals running at 26.7 MHz(set by on-chip oscillator of 10 MHz and NDIV in PLL\_CON to 0010<sub>B</sub>), RESET = V<sub>DDP</sub>.
- <sup>4)</sup> I<sub>DDP</sub> (idle mode) is measured with: CPU clock disabled, watchdog timer disabled, input clock to all peripherals enabled and running at 26.7 MHz, RESET = V<sub>DDP</sub>.
- <sup>5)</sup> I<sub>DDP</sub> (active mode with slow-down mode) is measured with: CPU clock and input clock to all peripherals running at 833 KHz by setting CLKREL in CMCON to 0101<sub>B</sub>, RESET = V<sub>DDP</sub>.
- <sup>6)</sup>  $I_{DDP}$  (idle mode with slow-down mode) is measured with: CPU clock disabled, watchdog timer disabled, input <u>clock to</u> all peripherals enabled and running at 833 KHz by setting CLKREL in CMCON to 0101<sub>B</sub>, <u>RESET</u> =  $V_{DDP}$ .



#### Table 38 Power Down Current (Operating Conditions apply)

Parameter	Symbol	Limit Values		Limit Values		Unit	<b>Test Condition</b>
		typ. <sup>1)</sup>	max. <sup>2)</sup>				
Power-Down Mode <sup>3)</sup>	I <sub>PDP</sub>	1	10	μA	$T_{A} = +25 \text{ °C.}^{4)}$		
		-	30	μA	$T_{A} = + 85 \text{ °C.}^{(4)5)}$		

<sup>1)</sup> The typical  $I_{PDP}$  values are measured at  $V_{DDP} = 5.0$  V.

<sup>2)</sup> The maximum  $I_{PDP}$  values are measured at  $V_{DDP}$  = 5.5 V.

<sup>3)</sup> I<sub>PDP</sub> (power-down mode) has a maximum value of 400  $\mu$ A at  $T_A$  = + 140 °C.

<sup>4)</sup> I<sub>PDP</sub> (power-down mode) is measured with: RESET = V<sub>DDP</sub>, V<sub>AGND</sub>= V<sub>SS</sub>, RXD/INT0 = V<sub>DDP</sub>; rest of the ports are programmed to be input with either internal pull devices enabled or driven externally to ensure no floating inputs.

<sup>5)</sup> Not subject to production test, verified by design/characterization.



#### 4.3 **AC Parameters**

#### **Testing Waveforms** 4.3.1

The testing waveforms for rise/fall time, output delay and output high impedance are shown in Figure 38, Figure 39 and Figure 40.

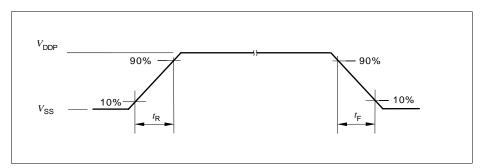
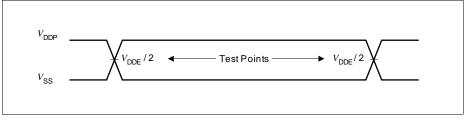


Figure 38 **Rise/Fall Time Parameters** 



**Testing Waveform, Output Delay** Figure 39

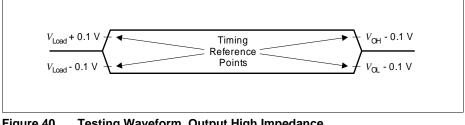


Figure 40 **Testing Waveform, Output High Impedance** 



# 4.3.2 Output Rise/Fall Times

#### Table 39 Output Rise/Fall Times Parameters (Operating Conditions apply)

Parameter	Symbol		imit alues	Unit	Test Conditions	
		min.	max.			
$V_{\text{DDP}}$ = 5V Range						
Rise/fall times 1) 2)	t <sub>R</sub> , t <sub>F</sub>	-	10	ns	20 pF. <sup>3)</sup>	
$V_{\text{DDP}}$ = 3.3V Range						
Rise/fall times 1) 2)	t <sub>R</sub> , t <sub>F</sub>	-	10	ns	20 pF. <sup>4)</sup>	

<sup>1)</sup> Rise/Fall time measurements are taken with 10% - 90% of the pad supply.

<sup>2)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

<sup>3)</sup> Additional rise/fall time valid for  $C_L = 20pF - 100pF @ 0.125 ns/pF$ .

<sup>4)</sup> Additional rise/fall time valid for  $C_L = 20pF - 100pF$  @ 0.225 ns/pF.

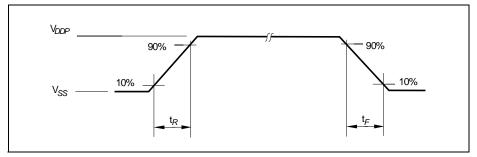


Figure 41 Rise/Fall Times Parameters



# 4.3.3 Power-on Reset and PLL Timing

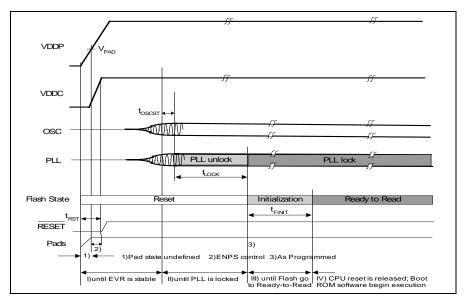
#### Table 40 Power-On Reset and PLL Timing (Operating Conditions apply)

Parameter	Symbol	Limit Values			Unit	Test Conditions	
		min.	typ.	max.			
Pad operating voltage	V <sub>PAD</sub> CC	2.3	-	-	V	1)	
On-Chip Oscillator start-up time	t <sub>OSCST</sub> CC	-	-	500	ns	1)	
Flash initialization time	t <sub>FINIT</sub> CC	-	160	-	μs	1)	
RESET hold time	t <sub>RST</sub> SR	_	500	-	μs	$V_{\rm DDP}$ rise time (10% – 90%) $\leq$ 500 $\mu$ s <sup>1)2)</sup>	
PLL lock-in in time	t <sub>LOCK</sub> CC	-	-	200	μs	1)	
PLL accumulated jitter	D <sub>P</sub>	-	-	0.7	ns	1)3)	

<sup>1)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

<sup>2)</sup> RESET signal has to be active (low) until  $V_{\text{DDC}}$  has reached 90% of its maximum value (typ. 2.5V).

<sup>3)</sup> PLL lock at 80 MHz using a 4 MHz external oscillator. The PLL Divider settings are K = 2, N = 40 and P = 1.



# Figure 42 Power-on Reset Timing



# 4.3.4 On-Chip Oscillator Characteristics

Table 41	On-chip Oscillator Characteristics (Operating Conditions apply)
----------	---

Parameter	Symbol	Lir	nit Va	lues	Unit	Test Conditions
		min.	typ.	max.		
Nominal frequency	f <sub>NOM</sub> CC	9.75	10	10.25	MHz	under nominal conditions <sup>1)</sup>
Long term frequency deviation	$\Delta f_{LT}$ CC	0	-	6.0	%	with respect to $f_{NOM}$ , over lifetime and temperature (125°C to 140°C), for one device after trimming
		-5.0	-	5.0	%	with respect to $f_{NOM}$ , over lifetime and temperature (-10°C to 125°C), for one device after trimming
		-6.0	-	0	%	with respect to $f_{NOM}$ , over lifetime and temperature (-40°C to -10°C), for one device after trimming
Short term frequency deviation	$\Delta f_{ST}$ CC	-1.0	-	1.0	%	within one LIN message (<10 ms 100 ms)

<sup>1)</sup> Nominal condition:  $V_{DDC} = 2.5$  V,  $T_A = +25^{\circ}$ C.

# 4.3.5 JTAG Timing

# Table 42TCK Clock Timing (Operating Conditions apply; $C_L = 50 \text{ pF}$ )

Parameter	Symbol	Lii	Limits	
		min	max	
TCK clock period <sup>1)</sup>	t <sub>TCK</sub> SR	50	_	ns
TCK high time <sup>1)</sup>	t <sub>1</sub> SR	20	_	ns
TCK low time <sup>1)</sup>	t <sub>2</sub> SR	20	-	ns
TCK clock rise time <sup>1)</sup>	t <sub>3</sub> SR	-	4	ns
TCK clock fall time <sup>1)</sup>	t <sub>4</sub> SR	-	4	ns

<sup>1)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

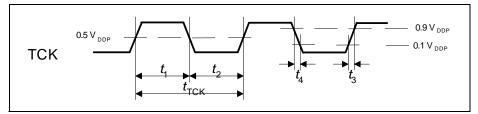


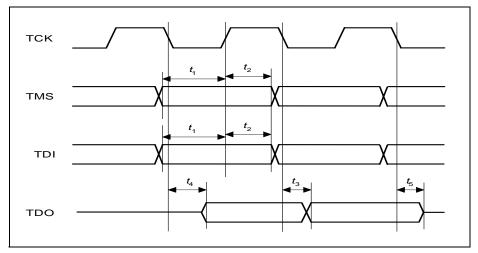
Figure 43 TCK Clock Timing



Parameter	Symbol		Limits		Unit
			min	max	
TMS setup to TCK <sup>1)</sup>	<i>t</i> <sub>1</sub>	SR	8.0	-	ns
TMS hold to TCK <sup>1)</sup>	<i>t</i> <sub>2</sub>	SR	5.0	-	ns
TDI setup to TCK <sup>1)</sup>	<i>t</i> <sub>1</sub>	SR	11.0	_	ns
TDI hold to TCK <sup>1)</sup>	<i>t</i> <sub>2</sub>	SR	6.0	_	ns
TDO valid output from TCK <sup>1)</sup>	$t_3$	СС	-	23	ns
TDO high impedance to valid output from TCK <sup>1)</sup> -	<i>t</i> <sub>4</sub>	CC	-	26	ns
TDO valid output to high impedance from TCK <sup>1)</sup> -	$t_5$	CC	-	18	ns

# Table 43 JTAG Timing (Operating Conditions apply; $C_{L} = 50 \text{ pF}$ )

<sup>1)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.







## 4.3.6 SSC Master Mode Timing

# Table 44SSC Master Mode Timing (Operating Conditions apply; $C_L = 50 \text{ pF}$ )

Parameter	Symbo	l Lim	Limit Values		
		min.	max.		
SCLK clock period <sup>1)</sup>	<i>t</i> <sub>0</sub> CC	2*T <sub>SSC</sub> <sup>2)</sup>	-	ns	
MTSR delay from SCLK <sup>1)</sup> 🦨	<i>t</i> <sub>1</sub> CC	0	8	ns	
MRST setup to SCLK <sup>1)</sup> ¬	t <sub>2</sub> SF	22	_	ns	
MRST hold from SCLK <sup>1)</sup> ¬	t <sub>3</sub> SF	R 0	-	ns	

<sup>1)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

<sup>2)</sup>  $T_{SSCmin} = T_{CPU} = 1/f_{CPU}$ . When  $f_{CPU} = 26.7$ MHz,  $t_0 = 74.9$ ns.  $T_{CPU}$  is the CPU clock period.

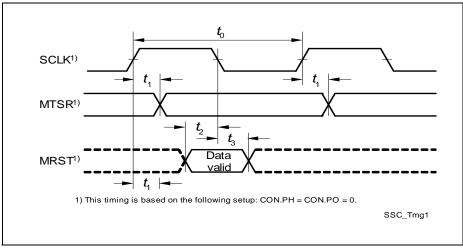


Figure 45 SSC Master Mode Timing



## Package and Reliability

# 5 Package and Reliability

# 5.1 Package Parameters (PG-TSSOP-38)

Table 45 provides the thermal characteristics of the package.

Parameter	Symbol		Limi	t Values	Unit	Notes	
			Min.	Max.			
Thermal resistance junction case <sup>1)2)</sup>	R <sub>TJC</sub>	СС	-	15.7	K/W	-	
Thermal resistance junction lead <sup>1)2)</sup>	R <sub>TJL</sub>	CC	-	39.2	K/W	-	

<sup>1)</sup> The thermal resistances between the case and the ambient (R<sub>TCA</sub>), the lead and the ambient (R<sub>TLA</sub>) are to be combined with the thermal resistances between the junction and the case (R<sub>TJC</sub>), the junction and the lead (R<sub>TJL</sub>) given above, in order to calculate the total thermal resistance between the junction and the ambient (R<sub>TLA</sub>). The thermal resistances between the case and the ambient (R<sub>TCA</sub>), the lead and the ambient (R<sub>TLA</sub>) depend on the external system (PCB, case) characteristics, and are under user responsibility.

The junction temperature can be calculated using the following equation:  $T_J=T_A + R_{TJA} \times P_D$ , where the  $R_{TJA}$  is the total thermal resistance between the junction and the ambient. This total junction ambient resistance  $R_{TJA}$  can be obtained from the upper four partial thermal resistances, by

a) simply adding only the two thermal resistances (junction lead and lead ambient), or

b) by taking all four resistances into account, depending on the precision needed.

<sup>2)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.



## Package and Reliability

# 5.2 Package Outline

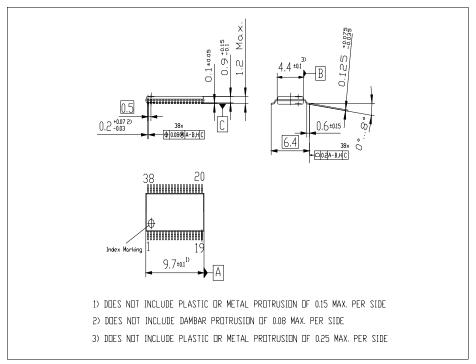


Figure 46 PG-TSSOP-38-4 Package Outline



## Package and Reliability

# 5.3 Quality Declaration

 Table 46 shows the characteristics of the quality parameters in the SAA-XC866.

#### Table 46 Quality Parameters

Parameter	Symbol	Limit Values			Unit	Notes	
		Min. Typ.		Max.			
Operation Lifetime when	t <sub>OP</sub>	-	_	1500	hours	<i>T</i> <sub>A</sub> = 140°C	
the device is used at the four stated $T_A^{(1)2)}$		-	-	2000	hours	<i>T</i> <sub>A</sub> = 125°C	
Tour stated TA		-	-	10000	hours	<i>T</i> <sub>A</sub> = 85°C	
		-	-	1500	hours	$T_{A}$ = -40°C	
Operation Lifetime when	t <sub>OP2</sub>	-	-	18000	hours	<i>T</i> <sub>A</sub> = 108°C	
the device is used at the two stated $T_A^{(1)2)}$		-	-	130000	hours	<i>T</i> <sub>A</sub> = 27°C	
Weighted Average Temperature <sup>2)3)</sup>	T <sub>WA</sub>	-	106	-	°C	for 15000 hours	
ESD susceptibility according to Human Body Model (HBM) for all pins $(except V_{DDC})^{2}$	V <sub>HBM</sub>	-	-	2000	V	Conforming to EIA/JESD22- A114-B	
ESD susceptibility according to Human Body Model (HBM) for $V_{DDC}^{(2)}$	V <sub>HBMC</sub>	-	_	600	V	Conforming to EIA/JESD22- A114-B	
ESD susceptibility according to Charged Device Model (CDM) pins <sup>2)</sup>	V <sub>CDM</sub>	-	-	750	V	Conforming to JESD22-C101-C	

<sup>1)</sup> This lifetime refers only to the time when the device is powered-on.

<sup>2)</sup> Not all parameters are 100% tested, but are verified by design/characterization and test correlation.

<sup>3)</sup> This parameter is derived based on the Arrhenius model.

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