

Used for Consumer and Outdoor Power Tools, Robotics, Drones, Ebikes

About this document

Scope and purpose

The purpose of this document is to serve as an user guide for evaluating the new 160 V three phase gate driver from Infineon Technologies – 6ED2742S01Q. EVAL-6ED2742S01QM1 for Battery Powered Applications (BPA) evaluation kit consists of a three-phase inverter power board with the 160 V rated 6ED2742S01Q (in 5 x 5 QFN-32 package) three phase gate driver driving six 150 V rated OptiMOS™ MOSFETs BSC074N15NS5 (in 5 x 6 Super SO8 package). The power board has a M1 connector that is used to interface with iMotion™ Modular Application Design Kit (MADK) control card – EVAL-M1-101T. This control card is powered by IMC101T-T038 iMOTION™ Motor Control IC featuring advanced Motor Control Engine (MCE), that enables state-of-the-art Sensorless Field Oriented Control (FOC) without need for any coding. IMC101T-T038 is a part of the iMOTION™ IMC100 series, which is the next generation of well established iMOTION™ motor control IC platform dedicated to field-oriented control of PMSM (BLDC) motors. The evaluation kit comes with both EVAL-6ED2742S01QM1 and EVAL-M1-101T including a USB connector to interface with a laptop having MCE Designer software.

Intended audience

Direct Current (DC) based Motor (BLDC, Servo, Stepper) control system engineers designing battery operated consumer power tools, outdoor power tools, robotics systems, drones, electric bikes and Small Home Appliances such as hair dryers, vacuum cleaners.

Evaluation Board

This board is to be used during the design-in process for evaluating and measuring characteristic curves, and for checking datasheet specifications. This board must be used in laboratory environment by trained power electronic engineers adhering to safety requirements.

Note: PCB and auxiliary circuits are NOT optimized for final customer design.

Note: Boards do not necessarily meet safety, EMI, quality standards (for example UL, CE) requirements.

Ordering information

| Base part number | Package | Standard pack | | Orderable part number | |
|---------------------------------|-------------|---------------|----------|-----------------------|--|
| | | Form | Quantity | | |
| EVAL-6ED274201Q BPA EVAL Kit | MADK EVAL | Boxed | 1 | EVAL6ED2742S01Q | |
| 6ED2742S01Q | PG-VQFN-32 | Tape and Reel | 3000 | 6ED2742S01QPXUMA1 | |
| EVAL-M1-101T | MADK EVAL | Boxed | 1 | EVALM1101TTOBO2 | |
| BSC074N15NS5 | PG-TSON-8 | Tape and Reel | 5000 | BSC074N15NS5ATMA1 | |
| IFX25001 V33 | PG-SOT223-4 | Tape and Reel | 1000 | IFX25001 ME V33 | |



Used for Consumer and Outdoor Power Tools, Robotics, Drones, Ebikes Important notice

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Used for Consumer and Outdoor Power Tools, Robotics, Drones, Ebikes Safety precautions

Safety precautions

Please note the following warnings regarding the hazards associated with development systems Note:

| Table 1 | Safety precautions |
|---------|--------------------|
|---------|--------------------|

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|----------|---|--|--|--|
| | 7 | | | |

Warning: The DC link potential of this board is up to 160 VDC. When measuring voltage waveforms by oscilloscope, high voltage differential probes must be used. Failure to do so may result in personal injury or death.



Warning: The evaluation or reference board contains DC bus capacitors which take time to discharge after removal of the main supply. Before working on the drive system, wait five minutes for capacitors to discharge to safe voltage levels. Failure to do so may result in personal injury or death. Darkened display LEDs are not an indication that capacitors have discharged to safe voltage levels.



Warning: Remove or disconnect power from the drive before you disconnect or reconnect wires, or perform maintenance work. Wait five minutes after removing power to discharge the bus capacitors. Do not attempt to service the drive until the bus capacitors have discharged to zero. Failure to do so may result in personal injury or death.



Caution: The heat sink and device surfaces of the evaluation or reference board may become hot during testing. Hence, necessary precautions are required while handling the board. Failure to comply may cause injury.



Caution: Only personnel familiar with the drive, power electronics and associated machinery should plan, install, commission and subsequently service the system. Failure to comply may result in personal injury and/or equipment damage.



Caution: The evaluation or reference board contains parts and assemblies sensitive to electrostatic discharge (ESD). Electrostatic control precautions are required when installing, testing, servicing or repairing the assembly. Component damage may result if ESD control procedures are not followed. If you are not familiar with electrostatic control procedures, refer to the applicable ESD protection handbooks and guidelines.



Caution: A drive that is incorrectly applied or installed can lead to component damage or reduction in product lifetime. Wiring or application errors such as undersizing the motor, supplying an incorrect or inadequate AC supply, or excessive ambient temperatures may result in system malfunction.



Caution: The evaluation or reference board is shipped with packing materials that need to be removed prior to installation. Failure to remove all packing materials that are unnecessary for system installation may result in overheating or abnormal operating conditions.



Used for Consumer and Outdoor Power Tools, Robotics, Drones, Ebikes

Table of contents

| Abou | t this document | 1 |
|---------------------|--|----|
| Impo | rtant notice | 2 |
| Safet | ty precautions | 3 |
| Table | e of contents | 4 |
| 1 | The board at a glance | |
| - 1.1 | Scope of delivery | |
| 1.2 | Block diagram | |
| 1.3 | Main features | |
| 1.4 | Board parameters and technical data | |
| 2 | System and functional description | |
| - 2.1 | Getting started with EVAL-6ED2742S01QM1 | |
| 2.1.1 | The iMOTION™ control board | |
| 2.1.2 | iMOTION™ development tools and software | |
| 2.1.3 | MCEWizard setup overview | |
| 2.1.4 | MCEDesigner setup overview | 13 |
| 2.2 | M1 Connector Pin details | 13 |
| 2.3 | Description of functional blocks | 15 |
| 2.3.1 | Overview of 6ED2742S01Q | 15 |
| 2.3.2 | 6ED2742S01Q PMU in Config #2 - Pre-regulator / charge pump: VBATT connected to VIN | 17 |
| 2.3.3 | 6ED2742S01Q PMU in Config #3: VBATT connected to VIN and external BJT | 17 |
| 2.4 | Shoot-through protection | 18 |
| 2.5 | Enable, Fault reporting and programmable fault clear timer | 19 |
| 2.6 | Over-current protection | |
| 2.7 | Integrated Current Sense Operational amplifier – unused in current EVAL-6ED2742S01QM1 b | |
| 2.8 | Gain settings for the over current protection (ITRIP_Conf) and current sense operational amp | |
| | (CSO) | |
| 2.9 | Advanced input filter | |
| 2.10 | Short-Pulse / Noise rejection filters | |
| 2.10. | | |
| 3 | System design | |
| 3.1.1 | DC-link voltage measurement and inrush current limiter | |
| 3.1.2 | Inverter section using OptiMOS™ 5 BSC074N15NS5 | |
| 3.1.3 | Drive circuit with 6ED2742S01Q | |
| 3.1.4 | Overcurrent protection | |
| 3.1.5 | Auxiliary power supply | |
| 3.2 | Schematics | |
| 3.3 | Layout | |
| 3.3.1 3.3.2 | Layout details | |
| | Layout guidelines | |
| 3.4 3.5 | Bill of material | |
| | | |
| 4 | System performance | |
| 4.1 | System setup for running a BLDC motor | |
| 4.2 | The output phase current | 41 |



Used for Consumer and Outdoor Power Tools, Robotics, Drones, Ebikes Table of contents

| 4.3 | Test result for overcurrent protection | 42 |
|-----|--|----|
| 5 | References and appendices | 43 |
| 5.1 | Abbreviations and definitions | |
| 5.2 | References | 43 |



The board at a glance

The board at a glance 1

The evaluation board EVAL-6ED2742S01QM1 is powered by a DC source in the range of 10.8 V to 120 V and provides a three-phase output to power upto 250 W (without heat sink) to 500 W (with heat sink) BLDC motors operating at a nominal 24 V DC bus voltage. It contains the new 160 V, three phase gate driver IC 6ED2742S01Q, the OptiMOS™ BSC074N15NS5, bus capacitors, a single shunt for current sensing, a voltage divider for DC-link voltage measurement, and other peripheral circuits. The board can also be powered up from a 10.8 V to 96 V Liion battery pack. Before connecting the battery pack, ensure reverse polarity protection and inrush protection is provided to EVAL-6ED2742S01QM1 board.

1.1 Scope of delivery

The EVAL-6ED2742S01QM1 evaluation board is delivered along with an iMOTION™ 2.0 M1 interface connector and EVAL-M1-101T control card.

The package also contains a DC-DC buck converter with integrated MOSFET to provide auxiliary power supply of 12 V and 3.3 V; a single shunt for current sensing and overcurrent protection; and a voltage divider for DC-link voltage measurement. The board can be operated directly with the DC power supply without any additional components. An inrush current limiter is recommended if operating directly from a battery source.

1.2 **Block diagram**

Figure 1 shows a typical application diagram of EVAL-6ED2742S01QM1 for driving a BLDC motor. The system adopts a single-shunt configuration, which is prevalent for current BLDC motor drive applications.

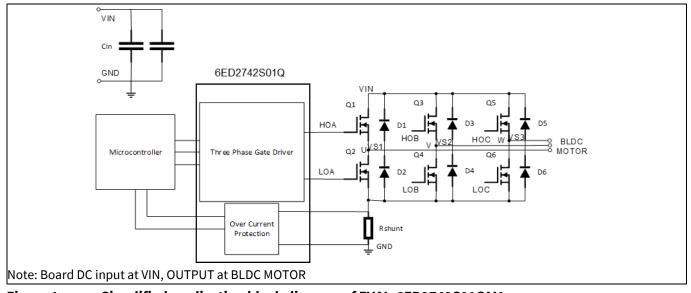


Figure 1 Simplified application block diagram of EVAL-6ED2742S01QM1

The 6ED2742S01Q is a 160 V SOI based gate driver designed for three phase BLDC motor drive applications. Integrated bootstrap diodes are used to supply the three external high sides charging bootstrap capacitors and supports 100% duty cycle operation by a trickle charge pump. Protection features include under voltage lockout, over current protection with configurable threshold, fault communication and automatic fault clear. The output drivers feature a high-pulse current buffer stage designed for minimum driver cross-conduction. A current sense operational amplifier (CSA) with selectable gain is integrated between the VSS and COM



The board at a glance

The functional blocks of EVAL-6ED2742S01QM1 BPA Evaluation Kit are presented in Figure 2

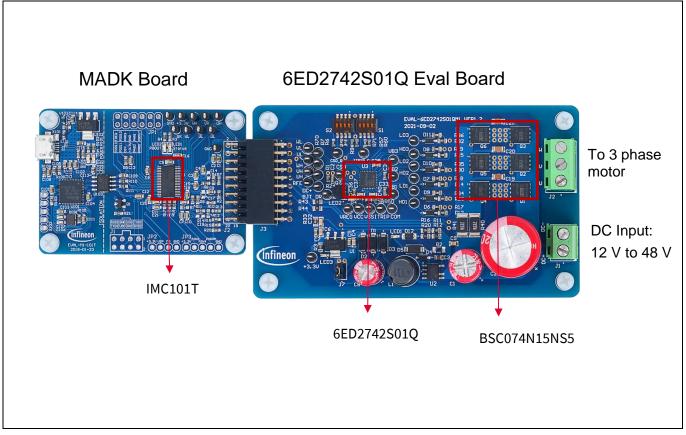


Figure 2 Functional blocks of EVAL-6ED2742S01QM1 BPA Evaluation Kit (top view)

1.3 **Main features**

The main features of EVAL-6ED2742S01QM1 include:

- Nominal DC input voltage of 24 V
- Design for maximum 250 W motor power output (500 W with added heatsink / forced air cooling)
- Single shunt for current sensing
- Sensing of DC-link voltage
- +12 V and +3.3 V auxiliary power supplies on the board
- M1 interface compatible with the iMOTION™ control board
- Integrated power management unit (PMU) inside 6ED2742S01Q with
 - Linear pre-regulator to enable wide input VIN range
 - Integrated charge pump for stable V_{cc}
- Integrated current sense amplifier (CSA) inside 6ED2742S01Q with selectable gain
- Integrated over-current protection inside 6ED2742S01Q with selectable V_{REF} threshold
- Integrated shoot-through protection with built-in dead time
- Multi-function RFE pin (Enable, Fault, and automatic Fault clear) to enable / disable motor drive
- Integrated short pulse / noise rejection input filter



The board at a glance

1.4 Board parameters and technical data

The key specifications of EVAL-6ED2742S01QM1 are listed in Table 2.

Table 2 Specifications of EVAL-6ED2742S01QM1

| Damana atau | Complete of | Sumbal Canditions | | Value | | |
|----------------------------|------------------|---|------|-------|------|-----|
| Parameter | Symbol | Conditions | Min. | Nom. | Max. | |
| Input | | | | | | |
| Input voltage | V _{IN} | DC voltage | 10 | 24 | 48 | V |
| Input current | I _{IN} | Input current | - | - | 12 | Α |
| Output | | | | | | |
| Output current | I _{OUT} | RMS phase-out current | - | - | 10 | Α |
| Output power | Роит | Additional heat dissipation is needed (e.g. heat sink , forced air cooling) | 150 | 250 | 500 | W |
| Overcurrent protection | I _{OCP} | Peak phase-out current | _ | _ | 11 | Α |
| Switching frequency | | | | | | |
| Inverter carrier frequency | F _{sw} | | - | 23 | _ | kHz |
| Auxiliary power supply | | | | | | |
| Fan power supply | | | - | 12 | _ | V |
| Controller power supply | | | _ | 3.3 | _ | ٧ |
| System environment | | | | | | |
| Ambient temperature | | With adequate cooling method | - | 25 | _ | °C |
| PCB characteristics | | | | | | |
| | L | Length | _ | 120 | _ | mm |
| Dimensions | W | Width | _ | 60 | _ | mm |
| | Н | Height | _ | 25 | _ | mm |
| Layer | | | _ | 4 | _ | |
| PCB thickness | | | - | 1.6 | _ | mm |
| Copper thickness | | | - | 4 | _ | oz. |
| Material | | FR-4, RoHS-compliant | | | | |

Note: The evauation board is designed to evaluate the parameters of 6ED2742S01Q three phase gate driver IC. The board does not have over temperature protection and care should be taken to set the correct OCP – over current protection level with ITRIP config setting before starting the board test with a given BLDC motor. The board is design limits with and without heatsink are as below:

Table 3 EVAL-6ED2742S01QM1 board thermal / overcurrent limits:

| With heatsink | With air cooling | Typical Speed | Board temperature after 15 minutes | Phase current |
|------------------|---------------------|------------------|--|--|
| No | No | 52 kRPM | 80°C | 8 A _{rms} , 16 A _{peak} |
| Yes | No | 58 kRPM | 75°C | 10 A _{rms} , 20 A _{peak} |
| No | Yes | 60 kRPM | 75°C | 11 A _{rms} , 21 A _{peak} |
| Yes | Yes | 62 kRPM | 82°C | 12 A _{rms} , 22 A _{neak} |



System and functional description

2 System and functional description

2.1 Getting started with EVAL-6ED2742S01QM1

To run the motor system, a combination of the power board EVAL-6ED2742S01QM1 and a matching iMOTION™ control board (e.g., EVAL-M1-101T) are required. This chapter explains how to set up the system and get started with the iMOTION™ development platform.

The power board EVAL-6ED2742S01QM1 can run a BLDC motor by connecting to the iMOTION™ 2.0 control board EVAL-M1-101T through an M1 interface. The system connection is shown in Figure 3, and the test results are listed in Section 4.1.

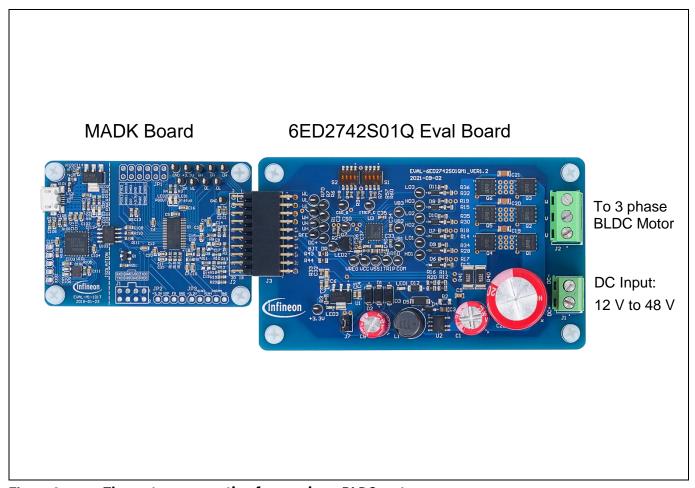


Figure 3 The system connection for running a BLDC motor.

2.1.1 The iMOTION™ control board

The latest IMC101T-T038 MCE software package can be downloaded from www.infineon.com/imotion-software.

With EVAL-M1-101T and the latest software, connect your PC to the control board via a USB cable to program and tune it.

- 1. Connect EVAL-M1-101T's M1 20-pin interface connector J2 to the EVAL-6ED2742S01QM1 board connector J3.
- 2. In the MCEWizard, enter the system and operating parameters of the target motor and the hardware parameters of the evaluation board. This data is used for calculating the digital parameter set of the controller, representing the complete motor drive system.



System and functional description

- Go to the Verify & Save page and click Calculate Parameters. Then, save the drive parameter set into your project directory by clicking Export to MCEDesigner file (.txt) (see Figure 6). This saved drive system parameter file will be used later by the MCEDesigner. Refer to Section 2.1.4 or the MCEWizard user guide for more information.
- Connect motor phase outputs (J2 on the EVAL-6ED2742S01QM1 board) to the motor. 4.
- Connect DC input to the power input connector (J1 on the EVAL-6ED2742S01QM1 board), then power on the 5. system. Note the
- 6. Start the MCEDesigner tool.
- 7. Click **File > Open** to open the MCEDesigner default configuration file (.irc) for the IMC101T-T038 controller (IMC101T_Vxxx.irc). The IMC101T_Vxxx.irc file is included in the IMC101T-T038 MCE software package.
- The MCEDesigner should automatically connect to the EVAL-M1-101T control board using the default COM 8. port (indicated by a green circle next to the "COMx Up" status in the bottom frame of the MCEDesigner GUI). If the connection cannot be established, change the COM port as follows:
 - a) Open the System Page window.
 - b) Click Preferences > Connection > Connect using.
 - c) Choose one of the other available COM ports from the drop-down list.
- 9. Program the system parameters into the internal SRAM of the iMOTION™ IC using the following steps:
 - a. Click Tools > Programmer and select Program Parameters.
 - b. Browse and select the System Drive Parameters file created in step 3. See Section 2.1.4 for more information.
- 10. Start the motor by clicking the green traffic light button in the control bar. To stop the motor, click the red traffic light button in the control bar.

2.1.2 iMOTION™ development tools and software

The EVAL-6ED2742S01QM1 evaluation board can run a BLDC motor when connected to an external control board EVAL-M1-101T. Users must configure the iMOTION™ development tool and software according to the system and BLDC motor parameters.

The MCEDesigner and MCEWizard are based on MCEDesigner v2.3.1 and MCEWizard v2.3.1 and can be downloaded together with supported files from http://www.infineon.com/imotion-software.

A review of references [10] to [16], on page 45 will provide a deeper understanding of the overall iMotion™ ecosystem.

2.1.3 MCEWizard setup overview

Use the MCEWizard to configure the parameters for the evaluation board or motor. Figure 4 shows the Welcome page of the MCEWizard. Here, users can select the control board or power board from a drop-down list. Infineon releases new control and power boards regularly. Therefore, it is possible that some of the latest power boards have not been pre-configured in the MCEWizard tool and cannot be selected from the drop-down menu. In that case, users can select another power board (as similar as possible) and follow the setup steps in the MCEWizard by entering parameters specific to the power board chosen. Please refer to the application note of the corresponding power board for additional information.

After selecting the modular application design kit (MADK) control board and the power board, click **Next** in the right-hand bottom corner of the Welcome page to start the MCEWizard system setup process.



System and functional description



Figure 4 Welcome page of the MCEWizard

The iMOTION™ system enables users to easily test different combinations of control and power boards with their motors. Users should be familiar with system-level parameters related to the motor used. There are a limited number of parameters specific to the control board or power board hardware. Table 3 lists the hardware parameters specific to the EVAL-6ED2742S01QM1 power board for MCEWizard setup. Similar tables are available for each control board in its application note. A combination of data in this table and the corresponding table of the control board provides sufficient information to set up the motor drive system quickly.

Table 4 MCEWizard setup overview table

| Parameter | Value |
|--------------------------------|---|
| Control board selecting | EVAL-M1-101T for example |
| Motor 1 shunt configuration | Single shunt |
| Controller supply voltage | +3.3 V |
| Max DC bus voltage | 36 V |
| DC bus sensing high resistor | 300 kΩ |
| DC bus sensing low resistor | 4.7 kΩ |
| NTC temperature shutdown value | Refer to the control board's application note [6] |
| Gate sense low-side devices | High is true |
| Gate sense high-side devices | High is true |
| Motor 1 current input scaling | Calculated in the corresponding section of the control board's application note [6] |

After all the MCEWizard questions are answered, the Verify & Save page opens (see Figure 5.)



System and functional description

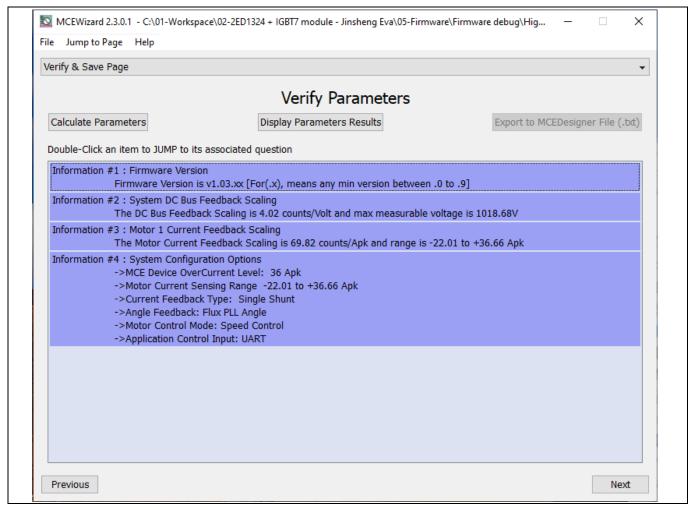


Figure 5 Verify & Save page of the MCEWizard.

Click Calculate Parameters and then Export to MCEDesigner File (.txt) to save the parameter file. This file will be used by the MCEDesigner in the next steps.



System and functional description

2.1.4 MCEDesigner setup overview

The MCEDesigner is a user interface to access or debug the control board. Open the MCEDesigner and then open the IMC101T_xx.irc file. The main display page for EVAL-M1-101T opens, as shown in Figure 6.

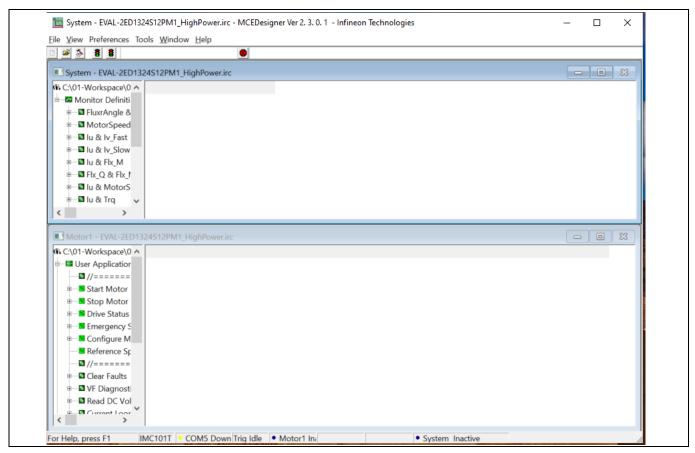


Figure 6 MCEDesigner's main display for EVAL-M1-101T

After the drive system parameter file has been programmed into the IMC101 controller, and the motor drive system powered, the MCEDesigner can be used to start/stop the motor, display motor current traces, change the motor speed, modify drive parameters, and perform many other functions. Please refer to the MCEDesigner documentation [8] for more details.

2.2 M1 Connector Pin details

In case the user prefers to use custom microcontroller, the pinout details of the 20 pin M1 connector used on the EVAL-6ED2742S01QM1 EVAL board is as shown in Table 5. The PWM signal output from the custom microcontroller can be fed to Pin 1, 2, 5, 7, 9,11 and the single shunt current feedback is available on Pin 10 and Pin 12. Over current protection / RFE output from the 6ED2742S01Q is available on Pin 13. VCC of 3.3 V (Pin 6, 8) and GND (Pin 2, 4) is also available.

User can interface their microcontroller signals as per the pinout shown in Table 5 and evaluate with their existing code of different control schemes such as trapezoidal / block commutation and Sensorless Field Oriented Control (FOC).

User can also look at <u>EVAL-M1-101TF</u> control card (available to order separately), that comes with Hall Sensor Inputs to evaluate their existing code of Hall Sensor based FOC control.



System and functional description

20-pin M1 connector setup Table 5

| Pin | Name | Pin Name Connectors |
|-----|----------|---|
| 1 | PWMUH | VDD compatible logic output for high side gate driver-Phase U |
| 2 | GND | Ground |
| 3 | PWMUL | VDD compatible logic output for low side gate driver-Phase U |
| 4 | GND | Ground |
| 5 | PWMVH | VDD compatible logic output for high side gate driver-Phase V |
| 6 | +3.3V | On board VDD supply |
| 7 | PWMVL | VDD compatible logic output for low side gate driver-Phase V |
| 8 | +3.3V | On board VDD supply |
| 9 | PWMWH | VDD compatible logic output for high side gate driver-Phase W |
| 10 | IU+ | Shunt voltage phase U |
| 11 | PWMWL | VDD compatible logic output for low side gate driver-Phase W |
| 12 | IU- | Ground |
| 13 | GK | Gate kill signal – active low when over current is detected |
| 14 | DCBSENSE | DC bus positive voltage, scaled in 0 - VDD range by a voltage divider |
| 15 | VTH | Thermistor input |
| 16 | IV+ | Shunt voltage phase V |
| 17 | IV- | Ground |
| 18 | IW+ | Shunt voltage phase W |
| 19 | IW- | Ground |
| 20 | VCC | Defined for 15 V Power Supply (not used in this board) |



System and functional description

2.3 Description of functional blocks

The motor inverter in EVAL-6ED2742S01QM1 hardware design is implemented by the new 160 V SOI three phase gate drivers 6ED2742S01Q and six OptiMOS™ 150 V N-Channel MOSFETs - BSC074N15NS5. The auxiliary power supply is derived from a high voltage DC-DC buck converter providing 12 V followed by a LDO providing the 3.3 V required to power the iMotion™ IMC101T microcontroller.

2.3.1 Overview of 6ED2742S01Q

Figure 7 shows the functional block diagram of 6ED2742S01Q. For more information such as static and dynamic electrical characteristics of the gate driver, please refer to the datasheet of 6ED2742S01Q.

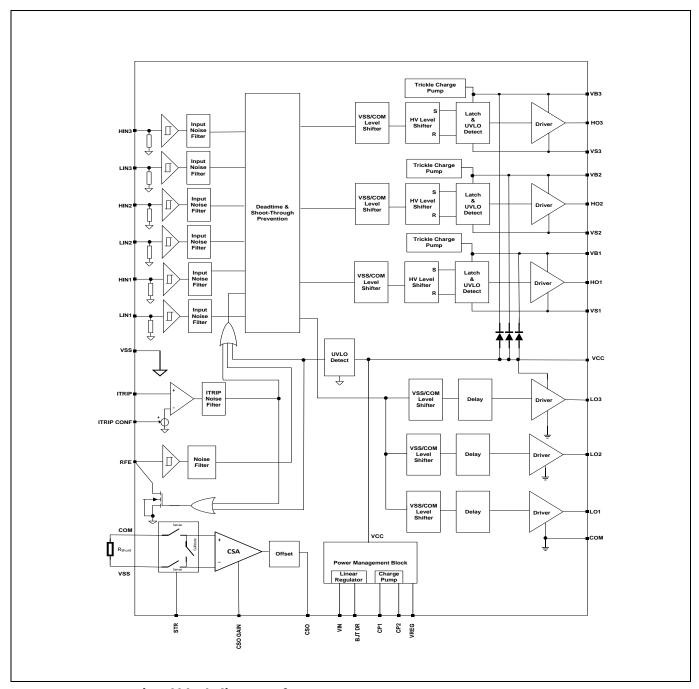


Figure 7 Functional block diagram of 6ED2742S01Q



System and functional description

The main features of 6ED2742S01Q include:

- Bootstrap voltage (VB node) of +160 V
- Floating channel designed for bootstrap operation •
- Integrated power management unit (PMU) with
 - Linear pre-regulator to enable wide input VIN range
 - Integrated charge pump for stable V_{cc}
- Integrated current sense amplifier (CSA) with selectable gain
- Integrated over-current protection with selectable V_{REF} threshold
- 100% duty cycle operation with trickle charge pump per high side
- Integrated low R_{ON}, ultra-fast bootstrap diodes •
- Independent under voltage lockout for both high and low side •
- Integrated shoot-through protection with built-in dead time •
- Multi-function RFE pin (Enable, Fault, and automatic Fault clear)
- Integrated short pulse / noise rejection input filter
- Schmitt trigger inputs with hysteresis
- 3.3 V, 5 V input logic compatible, outputs in phase with inputs
- Available in small footprint QFN32 lead, 5x5 mm package
- 2kV HBM ESD, RoHS compliant in a 5 x 5 mm VQFN-32 package

Table 6 Absolute maximum ratings of 6ED2742S01Q

| Symbol | De | efinition | Min. | Max. | Units |
|-----------------------|--|------------------------|----------------------|----------------|--------|
| V _B | High-side floating well su | pply voltage (Note 1) | -0.3 | 160 | |
| V _{IN} | Input battery supply volta | age | 5 | 140 | |
| V_{S} | High-side floating well su | pply return voltage | -0.3 | 140 | |
| V_{BS} | Bootstrap supply range | | -0.3 | 20 | |
| V_{HO} | Floating gate drive outpu | t voltage | V _S - 0.3 | $V_{B} + 0.3$ | |
| V_{REG} | Charge pump input volta | ge | -0.3 | 20 | |
| V_{cc} | Low side supply voltage | | -0.3 | 20 | |
| V_{LO} | Low-side output voltage | | -0.3 | $V_{cc} + 0.3$ | |
| V _{LOGIC IN} | Logic input voltage (HIN, | LIN) | -0.3 | 5 | |
| BJTDR | Output signal for base dri | ve of external NPN BJT | -0.3 | 20 | V |
| RFE | Enable, Fault and automa | atic fault clear pin | -0.3 | 5 | V |
| ITRIP | Over current protection in | nput pin | -0.3 | 5 | |
| ITRIP CONF | Over current protection threshold configuration pin | | -0.3 | 5 | |
| CSO | Current sense op amp output | | -0.3 | 3.6 | |
| CSO GAIN | Current sense op amp gain configuration pin | | -0.3 | 5 | |
| STR | Strobe signal for op amp offset compensation | | -0.3 | 5 | |
| CP1 | Charge pump capacitor p | in1 | -0.3 | 20 | |
| CP2 | Charge pump capacitor p | in2 | -0.3 | 20 | |
| СОМ | Low side power ground re | eturn | -5.0 | +5.0 | |
| dV _s /dt | Allowable V _s offset supply transient relative to COM | | _ | 50 | V / ns |
| P _D | Package power | 5 x 5mm QFN-32 | | 3 | W |
| . _U | dissipation @ T _A ≤+25°C | 3 x 3 | | | |
| Rth _{JA} | Thermal resistance, | 5 x 5mm QFN-32 | | 41 | °C/W |
| | junction to ambient | | | 450 | |
| T _J | Junction temperature | | _ | 150 | °C |

Note 1: There is an additional power dissipation in the internal bootstrap diode between pins VCC and VB for activated bootstrap diode in case VCC > VB



System and functional description

The 6ED2742S01Q internal Power Management Unit (PMU) can be configured to operate in four configurations. EVAL-6ED2742S01QM1 is designed to work in config #2 by default and can be configured to work in config #3 by removing R24 resistor on PCB board. Further information on operating the 6ED2742S01Q in config #1 and config #4 mode can be found in the device datasheet.

2.3.2 6ED2742S01Q PMU in Config #2 - Pre-regulator / charge pump: V_{BATT} connected to VIN

Configuration #2 provides wider input voltage flexibility with the smallest form factor since an external BJT is not used. Typical input voltage range operation up to 96 V is possible.

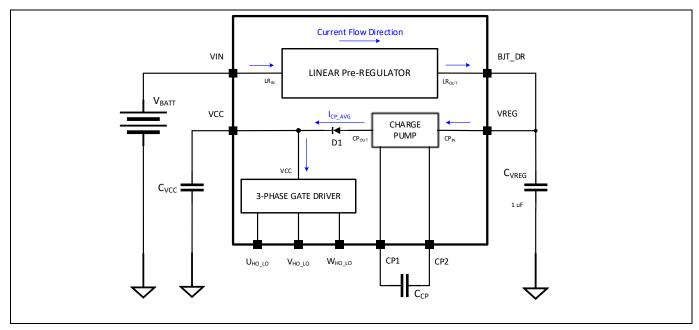


Figure 8 Typical configuration for 12 V to 90 V applications

Table 7 Charge pump only operating conditions (typical)

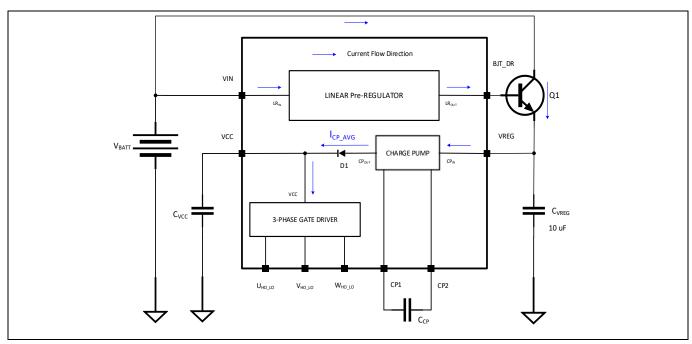
| V _{BATT} = VIN (V) | BJT_DR = VREG (V) | VCC (V) |
|--------------------------------|-------------------------|------------|
| 90 | | |
| 72 | | |
| 60 | 15.5 | 12 |
| 48 | 15.5 | |
| 36 | | 12 |
| 24 | | |
| 12 | 10.5 | |
| 9 | 7.5 | |

2.3.3 6ED2742S01Q PMU in Config #3: V_{BATT} connected to VIN and external BJT

Configuration #3 provides the widest input voltage range with the addition of an external NPN BJT to share the power dissipation between the internal PMU blocks and BJT.



System and functional description



Typical configuration for VIN > 60 V_{DC} applications. For stability reasons VREG capacitor needs Figure 9 to be 10 uF in this configuration

Pre-regulator, Charge pump, external BJT operating conditions (typ - 25°C) Table 8

| VIN (V) | BJT_DR (V) | VREG (V) | VCC (V) |
|---------|------------|----------|---------|
| 120 | | | |
| 90 | | | |
| 80 | 16.5 | 15.5 | |
| 36 | 10.3 | | 12 |
| 24 | | | |
| 18 | | 14.5 | |
| 12 | 10.5 | 9.5 | |
| 8 | 6.7 | 5.7 | 10 |

Typical input voltage range recommended is up to 120 V. I_{CP_AVG} of 15 mA for VIN operating range of 10 V - 120 V.

2.4 **Shoot-through protection**

The 6ED2742S01Q is equipped with shoot-through protection circuitry (also known as cross-conduction prevention circuitry). Figure 10 shows how this protection circuitry prevents both the high- and low-side switches from conducting at the same time.



System and functional description

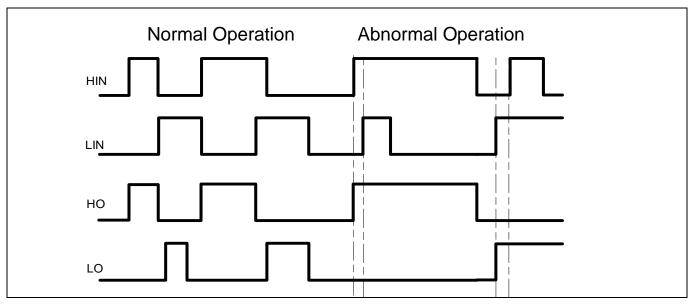
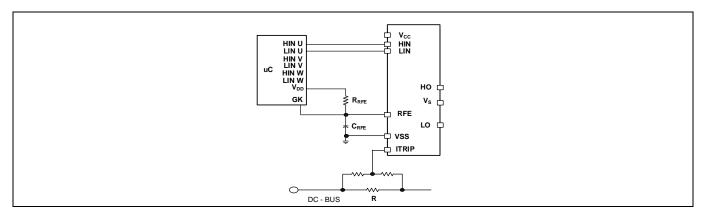


Figure 10 Illustration of shoot-through protection circuitry

2.5 Enable, Fault reporting and programmable fault clear timer

The 6ED2742S01Q provides an enable functionality that allows it to shutdown or enable the HVIC and also provides an integrated fault reporting output along with an adjustable fault clear timer. There are two situations that would cause the IC to report a fault via the RFE pin. The first is an undervoltage condition of VCC and the second is if the over-current feature has recognized a fault. Once the fault condition occurs, the RFE pin is internally pulled to VSS and the fault clear timer is activated. The RFE output stays in the low state until the fault condition has been removed and the fault clear timer expires; once the fault clear timer expires, the voltage on the RFE pin will return to its external pull-up voltage.

The length of the fault clear time period (t_{FLTCLR}) is determined by exponential charging characteristics of the capacitor where the time constant is set by R_{RFE} and C_{RFE}. Figure 15 shows that R_{RFE} is connected between the external supply (VDD) and the RFE pin, while CRFE is placed between the RFE and VSS pins.



Programming the fault clear timer Figure 11

The design guidelines for this network are shown in Table 7



System and functional description

Table 9 Fault clear timer design guidelines

| <u> </u> | ≤1 nF | | |
|------------------|--------------------------|--|--|
| C _{RFE} | Ceramic | | |
| 5 | 0.5 MΩ to 2 MΩ | | |
| R _{RFE} | >> R _{ON, RCIN} | | |

The length of the fault clear time period can be determined by using the formula below.

$$v_{c}(t) = V_{f^{*}}(1-e^{-t/RC})$$

$$t_{FLTCLR} = -(R_{RFE} \cdot C_{RFE}) \cdot ln (1 - V_{RFE} + V_{DD}) + 100ns$$

The voltage on the RFE pin should not exceed the VDD of the uC power supply.

The EVAL-6ED2742S01QM1 board has C_{RFE} of 1 nF and R_{RFE} of 1 M Ω .

2.6 Over-current protection

The 6ED2742S01Q is equipped with an over-current feature in addition to the stand-alone Current Sense Amplifier (CSA). This functionality can sense over-current events in the DC- bus. Once the IC detects an over-current event, the outputs are shutdown, and RFE is pulled to VSS.

The level of current at which the over-current protection (OCP) is initiated is determined by the shunt resistor connected between COM pin and VSS pin as shown in Figure 12, and by the threshold configured by ITRIP CONF pin (V_{ITRIP+}). The circuit designer will need to determine the maximum allowable level of current in the DC- bus and select R_0 and V_{ITRIP+} .

 $V_{ITRIP+} = R_0 \times I_{DC-}$

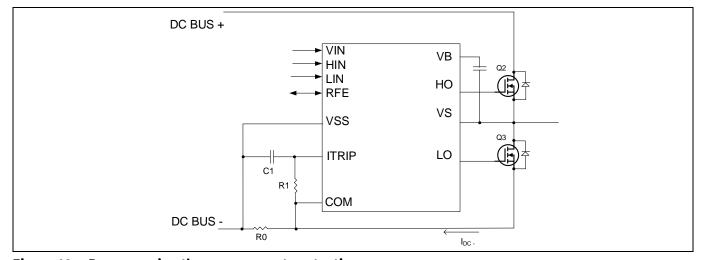
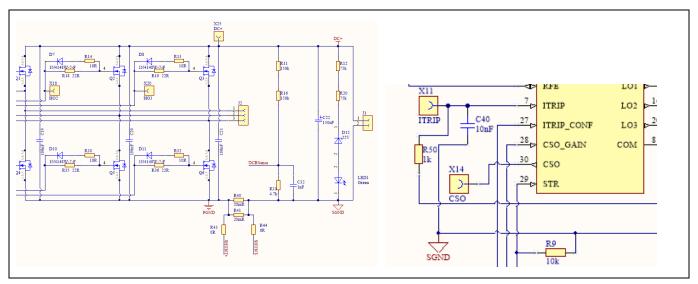


Figure 12 Programming the over-current protection

For example, a typical value for resistor R_0 could be 10 m Ω as with EVAL-6ED2742S01QM1 board providing the over-current protection on the ITRIP as 25 A with a ITRIP config threshold set to 0.25 V. See figure 14 to enable the correct switch position for this configuration on board.



System and functional description



Over current protection section of EVAL-6ED2742S01QM1 board Figure 13

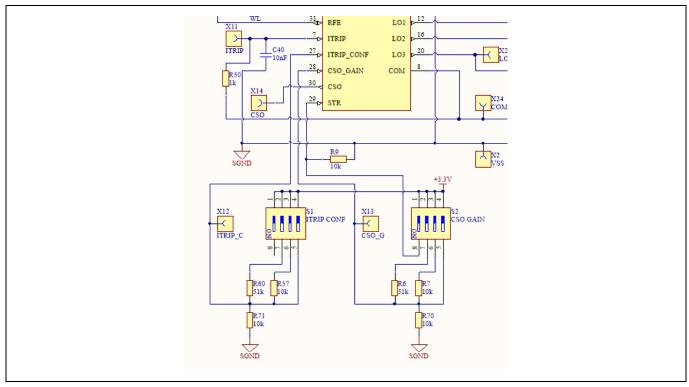


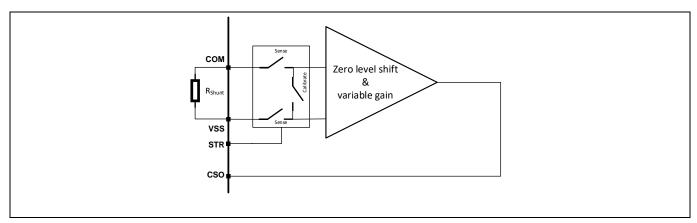
Figure 14 Over current protection ITRIP-Config section of EVAL-6ED2742S01QM1 board. Slide switch on position 2 to set 0.25 V ITRIP threshold.



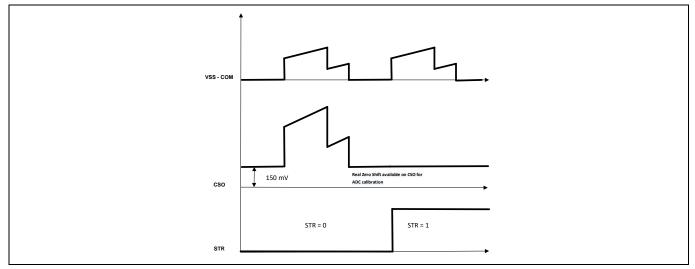
System and functional description

2.7 Integrated Current Sense Operational amplifier - unused in current EVAL-6ED2742S01QM1

A current sense operational amplifier (CSA) with configurable gain is integrated in the gate driver sensing the voltage between VSS and COM pins. The amplifier has a strobe input signal. When strobe signal is LOW the operational amplifier output signal CSO is following the VSS-COM voltage multiplied by a certain gain, when strobe signal is HIGH CSO signal is reporting the op amp offset. CSO output has an offset added above VSS of 150 mV. This offset ensures the measured current remains positive and does not go negative in any regular operating conditions.



Current sense operational amplifier Figure 15



Current sense operational amplifier timing diagram Figure 16

This section is not enabled in the EVAL-6ED2742S01QM1 board. Pages 42 and 43 in the current document show the simulated performance of the above integrated current sense amplifier.

User can tap the CSO output at test pin X14 on the EVAL-6ED2742S10Q board to verify the linearity and slew rate of the amplifier. A new EVAL-6ED2742S01QM1 Multifold Control board will demonstrate the performance of this amplifer along with XMC1302 microcontroller based MADK control card.



System and functional description

2.8 Gain settings for the over current protection (ITRIP_Conf) and current sense operational amplifier (CSO)

The gain for the over current protection via ITRIP and current sense operation amplifier output can be set using a simple potential divider network as shown in below figure 17.

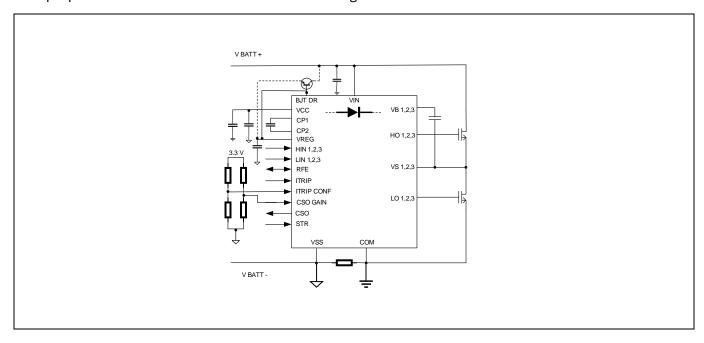


Figure 17 Setting over current protection (ITRIP) gain and current sense operational amplifier gain via the resistor divider

The voltage of the potential divider network can be scaled down from an external 3.3 V. It should be noted that when powering from external 3.3 V the supply sequencing should be such that the 6ED2742 is first powered up via VIN before applying the 3.3 V to the Gain / Config pins. Two use cases are shown in figure 20. Figure "a" shows the option where the 3.3 V input is either ramped up after or along with VIN / VCC. RFE gets asserted (IC gets enabled) after the UVLO of VCC is crossed and V_{RFE+} threshold is crossed. Figure" b" shows use case when the 3.3 V input is already available before VIN / VCC arrive. In this case, RFE gets asserted when the UVLO of VCC is crossed. It is good to note that the RFE pin is held low under two conditions:

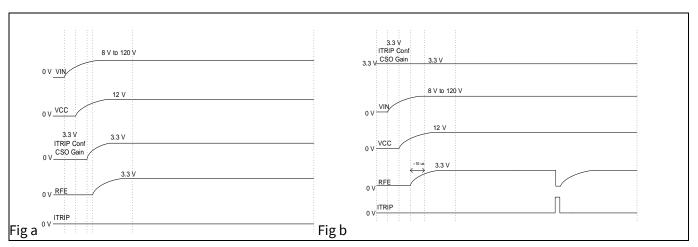
Under Voltage Lock Out (UVLO) of VCC / VRFE

ITRIP threshold as set by ITRIP Config is crossed

In case the 3.3 V arrives before VCC toggling (pulling down and the pulling up) the RFE will re-read the values of ITRIP config and CSO Gain.



System and functional description



Input voltage sequencing for setting up different gain and ITRIP config Figure 18

The voltage limits for different gains are as shown in the electrical characterisitics table 3 in page 6 and 7. They are summarized as beow in Table 8 and 9

ITRIP Config pin voltage settings for different overcurrent threshold voltages Table 10

| | Over current threshold voltage | Over current threshold | 950 | 1000 | 1050 | | ITRIP Conf > 3 V |
|--------------------|--------------------------------|------------------------|-----|------|------|------------------------|------------------------|
| V | | | 475 | 500 | 525 | | 1.5V < ITRIP Conf < 2V |
| V _{ITRIP} | | 225 | 250 | 275 | mV · | 0.5V < ITRIP Conf < 1V | |
| | | 110 | 130 | 150 | | ITRIP Conf < 0.25V | |

CSO gain pin voltage settings for different gain of CSO output Table 11

| | Current sense amplifier gain | 36 | 40 | 41 | | CSO gain > 3 V |
|-----------|------------------------------|----|----|----|-----|----------------------|
| C | | 18 | 20 | 21 | | 1.5V < CSO gain< 2V |
| G_{cso} | | 9 | 10 | 11 | V/V | 0.5V < CSO gain < 1V |
| | | 4 | 5 | 6 | | CSO gain < 0.25V |



System and functional description

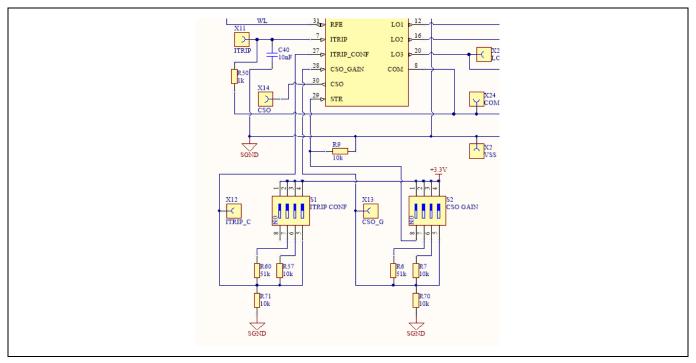


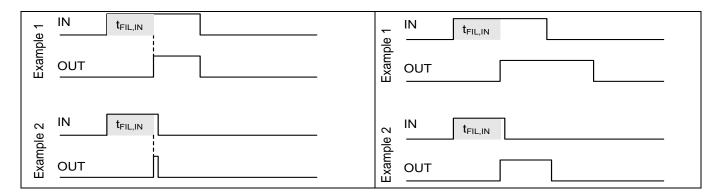
Figure 19 EVAL-6ED2742S01QM1 provides switch configurations to test different gain and threshold levels

2.9 Advanced input filter

The advanced input filter allows an improvement in the input/output pulse symmetry of the HVIC and helps to reject noise spikes and short pulses. This input filter has been applied to the HIN and LIN inputs.

Figure 19 shows a typical input filter and the asymmetry of the input and output. The upper pair of waveforms (Example 1) show an input signal with a duration much longer then $t_{\text{FIL,IN}}$; the resulting output is approximately the difference between the input signal and $t_{\text{FIL,IN}}$. The lower pair of waveforms (Example 2) show an input signal with a duration slightly longer then $t_{\text{FIL,IN}}$; the resulting output is approximately the difference between the input signal and $t_{\text{FIL,IN}}$.

Figure 22 shows the advanced input filter and the symmetry between the input and output. The upper pair of waveforms (Example 1) show an input signal with a duration much longer then $t_{\text{FIL},IN}$; the resulting output is approximately the same duration as the input signal. The lower pair of waveforms (Example 2) show an input signal with a duration slightly longer then $t_{\text{FIL},IN}$; the resulting output is approximately the same duration as the input signal.





System and functional description

Figure 19 Typical input filter Figure 20 Advanced input filter

2.10 Short-Pulse / Noise rejection filters

This device's input filter provides protection against short pulses (e.g., noise) on the input lines. If the duration of the input signal is less than $t_{\text{FIL,IN}}$, the output will not change states. Example 1 of Figure 21 shows the input and output in the low state with positive noise spikes of durations less than $t_{\text{FIL,IN}}$; the output does not change states. Example 2 of Figure 21 shows the input and output in the high state with negative noise spikes of durations less than $t_{\text{FIL,IN}}$; the output does not change states.

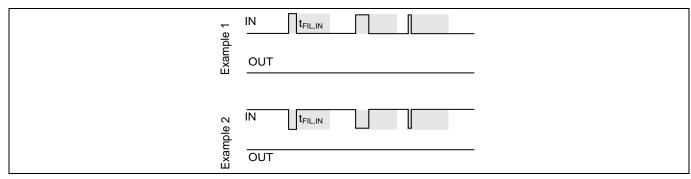


Figure 21 Noise Rejecting Input Filters

2.10.10verview of BSC074N15NS5

The BSC074N15NS5 is a 150 V rated N-Channel OptiMOSTM 5 power MOSFET. It was mainly chosen for its 5 x 6 Super SO8 package and its excellent gate charge x $R_{DS(on)}$ product Figure of Merit (FOM).

Figure 22 shows the key features of BSC074N15NS5. For more information about the MOSFET, please refer to the datasheet of BSC074N15NS5.

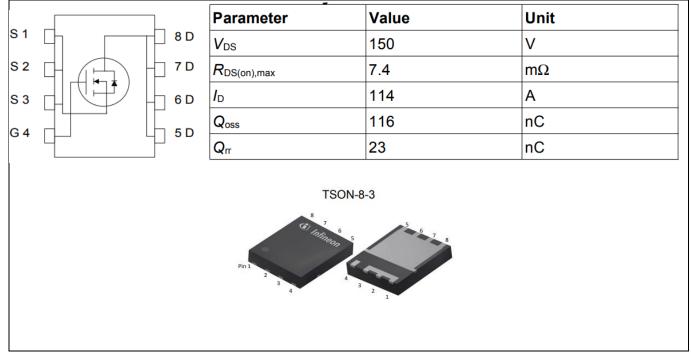


Figure 22 Key features of BSC074N15NS5



System and functional description

The typical gate charge characteristics and typical capacitances of BSC074N15NS5 are shown in Figure 23.

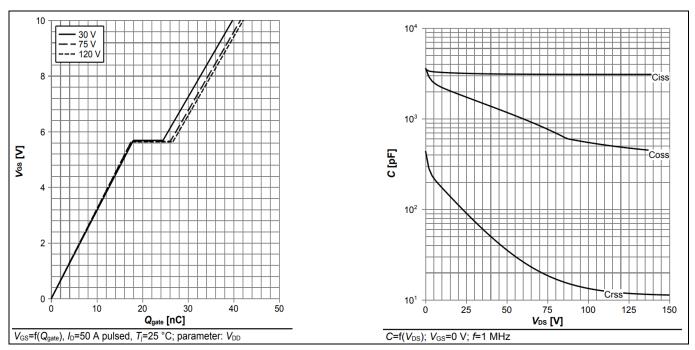


Figure 23 BSC074N15NS5 typical characteristics

A review of references [8] and [9], on <u>page 45</u> will provide a deeper understanding of the features of OptiMOS[™] 5 technology and its benefits in low voltage high power motor drive applications.

User Guide

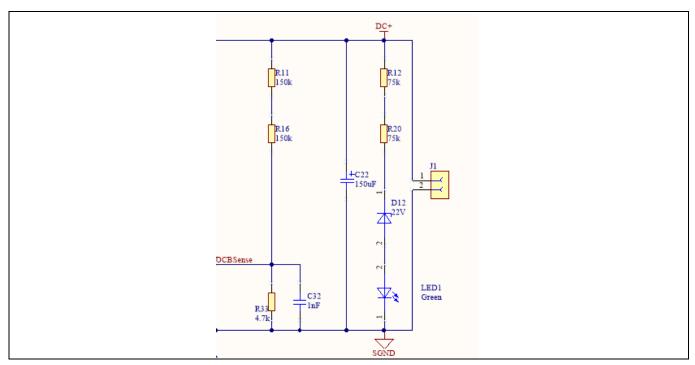
System design

System design 3

This chapter covers the hardware design of EVAL-6ED2742S01QM1 in detail. Users can modify the circuit or reselect the component values based on DC bus voltage, motor type.

3.1.1 DC-link voltage measurement and inrush current limiter

Bus capacitors should be large enough to stabilize the bus voltage.



Bus capacitor configuration and DC bus sensing Figure 24



System design

3.1.2 Inverter section using OptiMOS™ 5 BSC074N15NS5

The inverter section is implemented using BSC074N15NS5 as shown in Figure 25. It is a three-phase inverter bridge section with Infineon's OptiMOS™ 5 N-channel MOSFETs. R40/R41 (in parallel) are single shunt for current sensing.

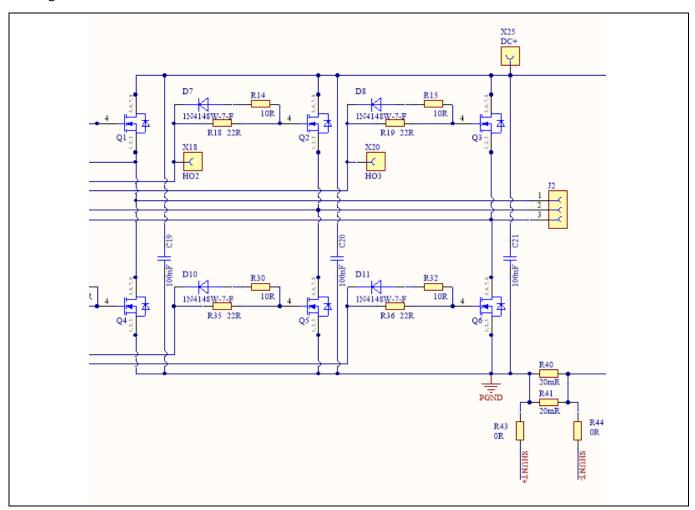


Figure 25 The diagram of the BSC074N15NS5 based inverter section with single shunt current sense



System design

3.1.3 Drive circuit with 6ED2742S01Q

The configuration of 6ED2742S01Q is shown in Figure 26.

The capacitor C33 (C34, C35) is used as bootstrap capacitor to provide the necessary floating supply voltage VBS. Thanks to the bootstrap diode integration in 6ED2742S01Q, an external bootstrap diode is not needed.

6ED2742S01Q is equipped with ITRIP input pin. This functionality can be used to detect overcurrent events in the DC bus. When the gate driver IC detects an overcurrent event through the ITRIP pin for one of the three phases, the output is shut down and the RFE is pulled to VSS to report the fault condition. The outputs of the other two phases are also shut down because of the daisy chain connection with the RFE pins. The RFE can also be used as the enable input pin and fault-clear time programmable pin.

The fault-clear time setup is based on the datasheet of 6ED2742S01Q. The main feature of 6ED2742S01Q designed with Infineon's SOI technology is its tolerance in terms of negative transient voltage in inductive load applications. 6ED2742S01Q can tolerate repetitive negative transient voltage of up to 15 V (pulse width up to 100 ns).

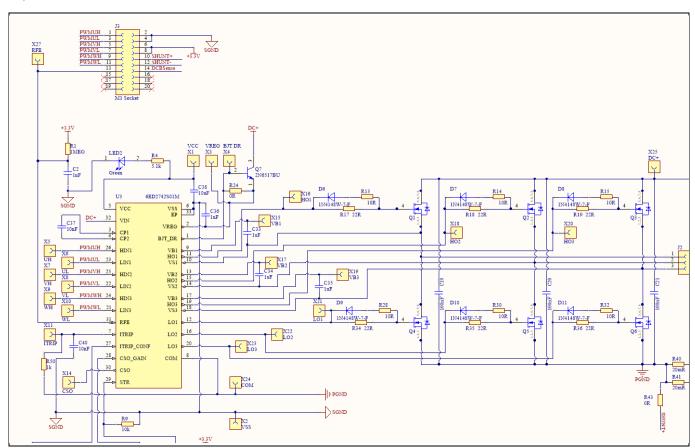
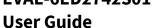


Figure 26 Drive circuit design with 6ED2742S01Q



System design

3.1.4 Overcurrent protection

The overcurrent circuit is shown in Figure 25. To save power loss of the shunt resistor, the R_{shunt} value is set as 10 $m\Omega$ by using two 20 $m\Omega$ / 2 W surface-mounted devices (SMD) resistors (R40/R41) and low equivalent series inductance (ESL) in parallel. In the design, the target current for overcurrent protection (I_{OCP}) is 25 A. The overcurrent threshold of the ITRIP pin is 0.54 V. An additional 0.19 V DC-bias voltage is needed at the ITRIP pin to meet the target current of 25 A. This is achieved by using resistor dividers R19 and R20 with 3.3 V supply.

The IOCP is calculated using:

$$locp = \frac{(V_{ITRIP+} - \frac{R19}{R19 + R20} * 3.3V) * (R19 + R20)}{R20 * Rshunt}$$
= 24.93 A

Here, V_{ITRIP} is the positive-going threshold for ITRIP and its typical value is 0.44 V, R19 = 100 Ω , and R20 = 1 k Ω .

Note: If EVAL-M1-101T is not used as the controller board, an offset circuit needs to be added.

3.1.5 Auxiliary power supply

The auxiliary power supply is derived from BM0650HV based DC-DC buck converter. The 12 V DC can be connected an external Fan (not provided with the Eval board) for forced air cooling during prolonged testing. The 3.3 V power supply required for the IMC101T based MADK control card is derived from the 12 V DC using the linear regulator IFX25001ME V33. This 3.3 V power supply is also used to bias different circuitry on board such as the ITRIP config and CSO config. The auxiliary supply stage is as shown in the schematic in figure 27.

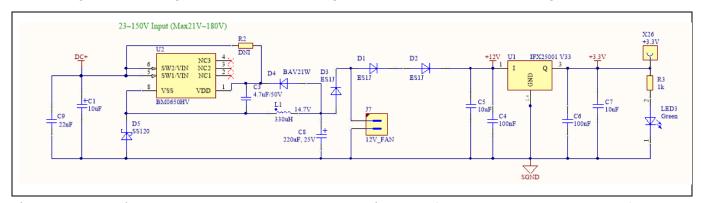


Figure 27 A high voltage DC-DC buck converter provides 12 V followed by a 3.3 V generated from a linear regualtor

infineon

System design

3.2 Schematics

The schematics of EVAL-6ED2742S01QM1 are shown in Figure 28

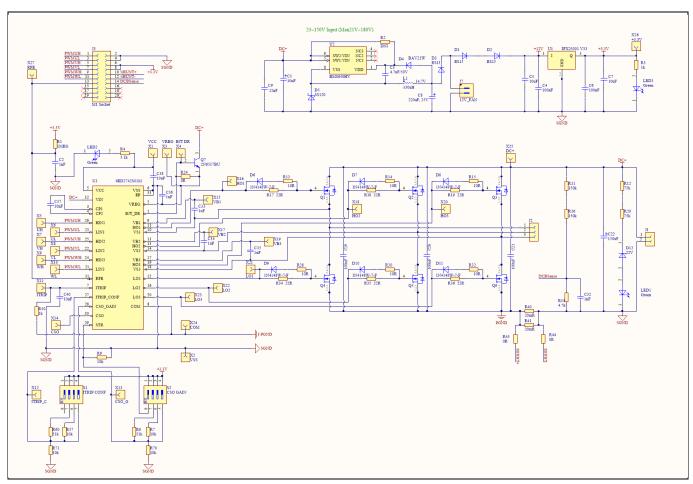


Figure 28 Circuitry of the power inverter board (main board)



System design

3.3 Layout

3.3.1 Layout details

Detailed layouts of the main inverter board are shown in Figure 29 to Figure 32.

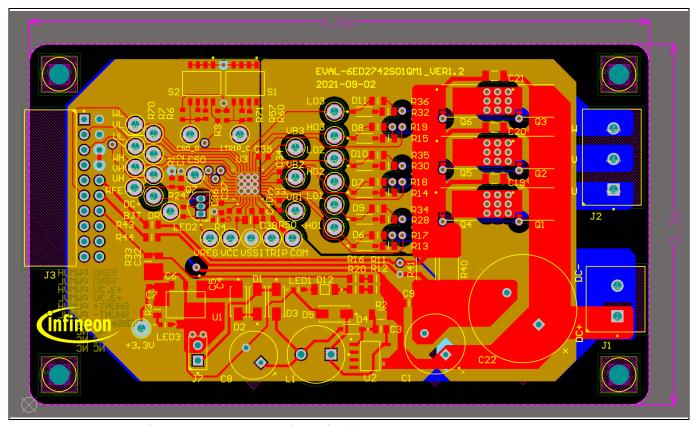


Figure 29 Power inverter board layout (top view)



System design

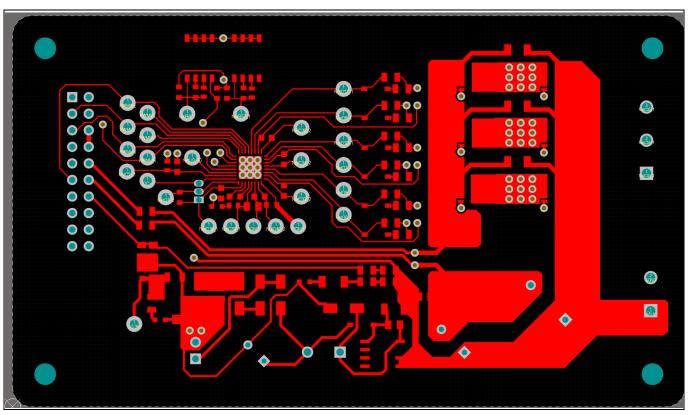


Figure 30 Power inverter board layout (bottom view)

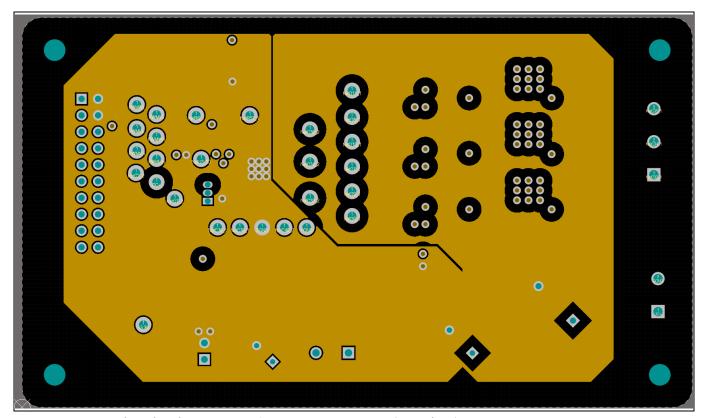


Figure 31 Drive circuitry layout of the daughter board (top view)



System design

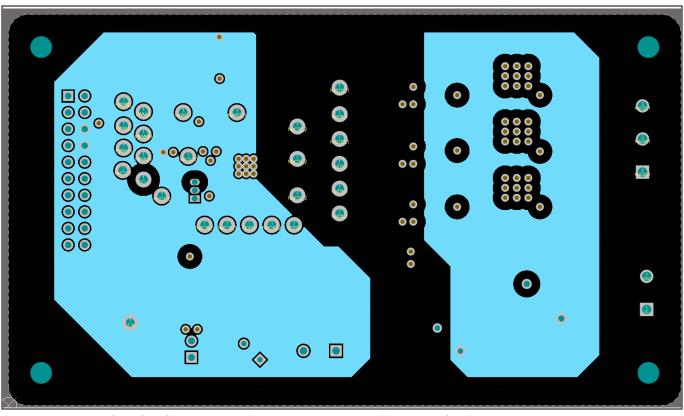


Figure 32 Drive circuitry layout of the daughter board (bottom view)

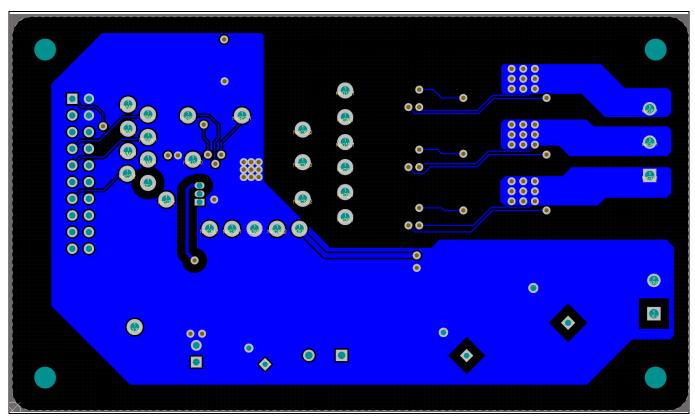


Figure 33 Drive circuitry layout of the daughter board (bottom view)



System design

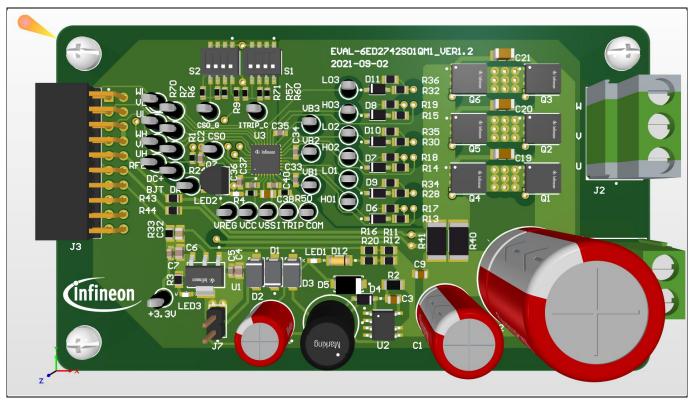


Figure 34 Drive circuitry layout of the daughter board (bottom view)

Both the schematic and layout files in Altium are provided on the 6ED2742S01Q webpage online. Users can download the same by logging in with their infineon.com credentials.

Layout guidelines 3.3.2

Some basic layout guidelines are as follows:

- The V_{CC} and V_{BS} bypass capacitors should be close to the IC
- The drive loop should be as small as possible
- The loop of VSS and COM should be made as small as possible by connecting the VSS and COM directly at the shunt-resistor terminals
- The two current-sensing traces should be started from the shunt terminals and placed close to each other

See pages 30 and 31 of 6ED2742S01Q datasheet for further layout guidelines.



System design

3.4 Bill of material

The complete bill of material (BOM) is available in the Downloads section of Infineon's homepage. A login is required to download this material.

Table 12 lists the important components used in EVAL-6ED2742S01QM1.

Table 12 BOM with Infineon components in Bold

| Designator | Component Value | Description | Quantity |
|--------------------|-----------------|---|----------|
| | | CAP / ELCO / 10uF / 200V / 20% / | |
| C1 | 10uF | Aluminiumelectrolytic | 1 |
| C2 | 1nF | CAP / - / 1nF / 25V / 10% / X7R (EIA) | 1 |
| C3 | 4.7uF/50V | Chip Monolithic Ceramic Capacitor | 1 |
| C4, C6 | 100nF | Surface Mount Ceramic Capacitor 0.1uF 25V X7R 10% | 2 |
| C5, C7 | 10uF | Chip Monolithic Ceramic Capacitor | 2 |
| | | CAP / ELCO / 220uF / 25V / 20% / | _ |
| C8 | 220uF, 25V | Aluminiumelectrolytic | 1 |
| C9 | 22nF | CAP / - / 22nF / 250V / 10% / X7R (EIA) | 1 |
| C19, C20, C21 | 100nF | Ceramic Capacitor 0.1uF 250V X7R 10% SMD 1206 | 3 |
| | | CAP / ELCO / 150uF / 200V / 20% / | _ |
| C22 | 150uF | Aluminiumelectrolytic | 1 |
| C32 | 1nF | Surface Mount Multilayer Ceramic Chip Capacitor | 1 |
| C33, C34, C35, C36 | 1uF | CAP / - / 1uF / 25V / 10% / X7R (EIA) | 4 |
| C37, C40 | 10nF | CAP / - / 10nF / 25V / 5% / X7R (EIA) | 2 |
| C38 | 10uF | Chip Monolithic Ceramic Capacitor | 1 |
| D1, D2, D3 | ES1J | 1.0A Ultra Fast Recovery Rectifier | 3 |
| D4 | BAV21W | Surface Mount Fast Switching Diode | 1 |
| D5 | SS120 | Surface Mount Schottky Barrier Rectifier, 20V | 1 |
| D6, D7, D8, D9, | | · | |
| D10, D11 | 1N4148W-7-F | Surface Mount Fast Switching Diode | 6 |
| D12 | 22V | Zener Diode, 5% 22V | 1 |
| | | IND / STD / 330uH / 1.1A / 10% / 470mR / Radial / | |
| L1 | 330uH | Inductor, Radial; | 1 |
| | | WL-SMCW SMT Mono-color Chip LED Waterclear, | |
| LED1, LED2, LED3 | Green | Green, 515nm | 3 |
| Q1, Q2, Q3, Q4, | | | |
| Q5, Q6 | BSC074N15NS5 | OptiMOS 5 Power-Transistor, 150 V, N-channel | 6 |
| | | NPN Epitaxial Silicon Transistor, High Voltage | |
| Q7 | 2N6517BU | Transistor | 1 |
| R1 | 1MEG | RES / STD / 1MEG / 100mW / 1% / 100ppm/K | 1 |
| R2 | DNI | Standard Thick Film Chip Resistor | 1 |
| R3 | 1k | Standard Thick Film Chip Resistor | 1 |
| R4 | 5.1k | RES / STD / 5.1k / 100mW / 1% / 100ppm/K | 1 |
| R6, R60 | 51k | RES / STD / 51k / 100mW / 1% / 100ppm/K | 2 |
| R7, R9, R57, R70, | | | |
| R71 | 10k | RES / STD / 10k / 100mW / 1% / 100ppm/K | 5 |
| R11, R16 | 150k | Standard Thick Film Chip Resistor | 2 |
| R12, R20 | 75k | Standard Thick Film Chip Resistor | 2 |



System design

| R13, R14, R15, | | | |
|----------------|--------------|---|---|
| R28, R30, R32 | 10R | Standard Thick Film Chip Resistor | 6 |
| R17, R18, R19, | | | |
| R34, R35, R36 | 22R | Standard Thick Film Chip Resistor | 6 |
| R24 | OR | RES / STD / OR / 100mW / OR / 0ppm/K | 1 |
| R33 | 4.7k | 4.7k/150V/1% | 1 |
| R40, R41 | 20mR | Low Value Power Metal Strip Resistor | 2 |
| R43, R44 | OR | Standard Thick Film Chip Resistor | 2 |
| R50 | 1k | RES / STD / 1k / 100mW / 1% / 100ppm/K | 1 |
| S1 | ITRIP CONF | Slide DIP Switch with seal tape | 1 |
| S2 | CSO GAIN | Slide DIP Switch with seal tape | 1 |
| U1 | IFX25001 V33 | Low Dropout Voltage Regulator, 3.3V Output | 1 |
| | | PFM Controller and 200 V Smart power with high | |
| U2 | BM0650HV | avalanche capability MOSFET | 1 |
| U3 | 6ED2742S01Q | 160 V pre-regulated three phase SOI gate driver | 1 |

Connector details 3.5

Table 13 **Connectors**

| PIN | Label | Function |
|-------------------------------|----------|---|
| Various signal input / output | X1 – X27 | Test points for various onboard signals |
| DC+, DC- | J1 | DC power supply input |
| U/V/W | J2 | Three phase-outs (U/V/W) to the motor |
| M1 connector | 13 | M1 Socket - iMOTION™ MADK-M1 20-pin interface connector |
| 12 V Aux output | J7 | 12V_FAN Header |



System performance

System performance 4

System setup for running a BLDC motor 4.1

EVAL-6ED2742S01QM1 is tested while running a custom BLDC motor in the setup shown in Figure 35.

Test conditions:

- Input: $V_{IN} = 24 \text{ VdC}$
- Phase-out current: I_{PHASE} = 8 A_{rms},
- Room temperature
- BLDC motor: Custom, I_{PHASE} = 10 A, test at speed = 52000 r/min

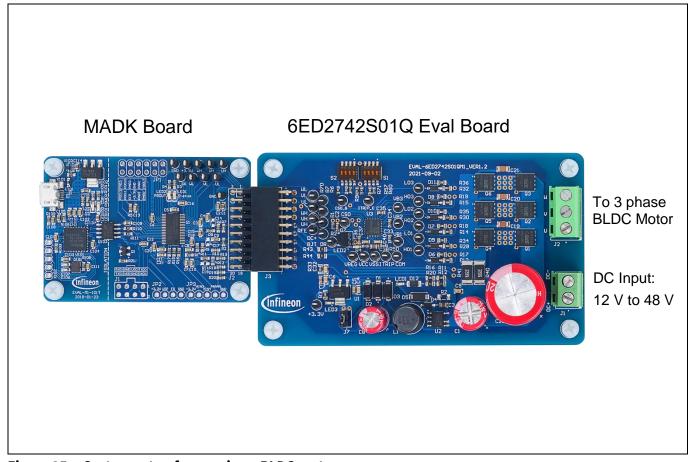


Figure 35 System setup for running a BLDC motor



System performance

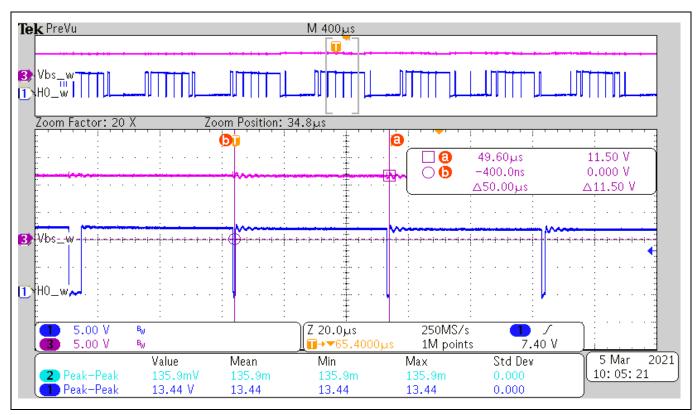


Figure 36 Trickle Charge Pump operation for large duty cycle (> 90 %) operation

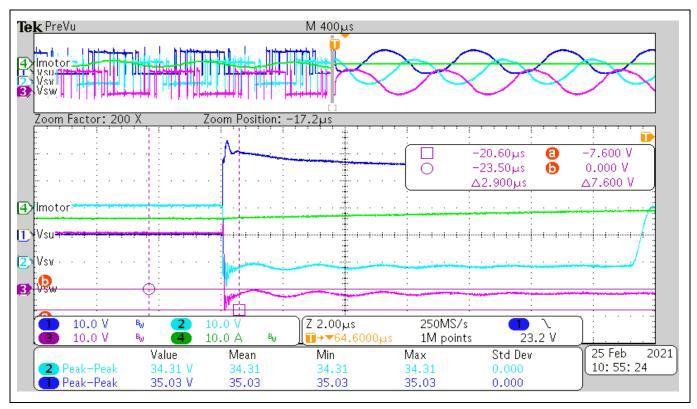


Figure 37 Motor Over Current protection



System performance

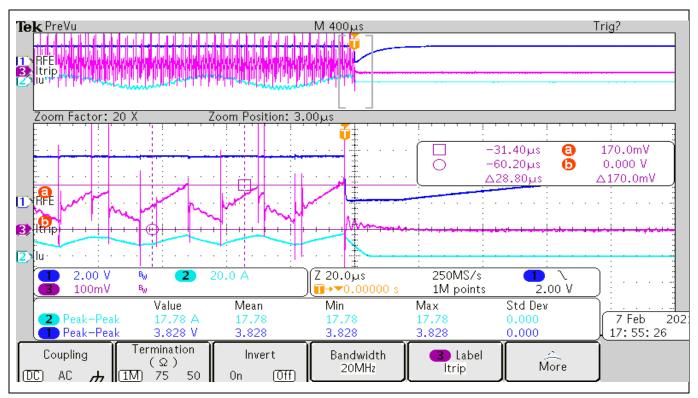


Figure 38 RFE pin in over current protection mode

4.2 The output phase current

Figure 39 shows the waveform when running a BLDC motor.

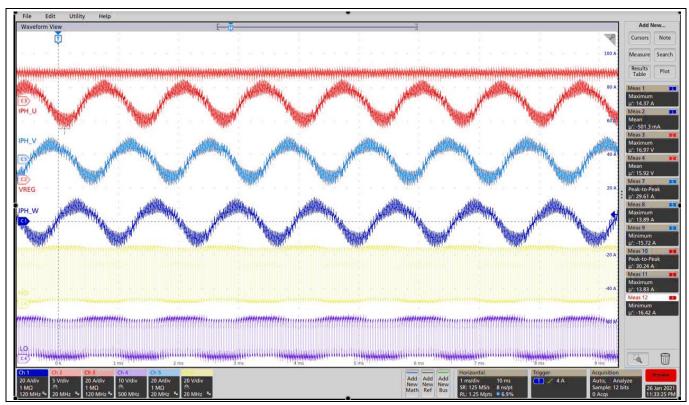


Figure 39 EVAL-6ED2742S01QM1 board test results showing three phase output currents at F_{sw} = 20 kHz



System performance

4.3 Test result for overcurrent protection

Error! Reference source not found. shows the overcurrent detection function of EVAL-6ED2742S01QM1 by ITRIP input of 6ED2742S01Q. Overcurrent detection shuts down the outputs of the gate driver if the ITRIP pin input is over 440 mV and lasts longer than 500 nanoseconds. In that case, the overcurrent protection starts when the peak current is around 24.93 A.

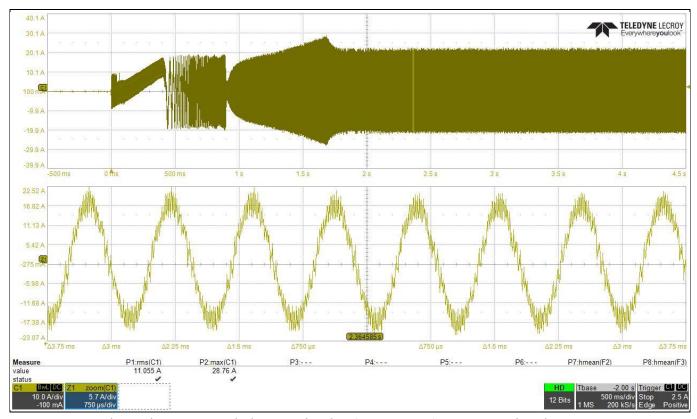


Figure 40 Typical drive characteristics at switching frequency of 23kHz, parking time = 0.4s, startup time = 0.93s

For the EVAL-6ED2742S01QM1 to run for 30 minutes at the rated speed without exceeding temperatures above 85°C (and damaging the board) a heat sink was glued with Thermal Glue onto the OptiMOS™ inverter stage.



References and appendices

5 References and appendices

5.1 Abbreviations and definitions

Table 14 Abbreviations

| Abbreviation | Meaning |
|--------------|---|
| IC | Integrated circuit |
| MOSFET | Metal Oxide Semiconductor Field Effect Transistor |
| DC | Direct Current |
| BLDC | Brushless Direct Current |
| PWM | Pulse Width Modulation |
| EVAL | Evaluation board |
| PCB | Printed Circuit Board |
| SOI | Silicon on Insulator |

5.2 References

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