SP34063A

Rev. 2.1.1



1.5A Buck/Boost/Inverting DC-DC Regulator

February 2014

GENERAL DESCRIPTION

The SP34063A is a monolithic switching regulator control circuit containing the primary functions required for DC-DC converters.

This device consists of an internal temperature compensated reference, voltage comparator, controlled duty cycle oscillator with active current limit circuit, driver and high current output switch. This device was specifically designed to be used in buck, boost, and Voltage-Inverting applications with a minimum number of external components.

ithe Broduct B The SP34063A is available in the 8 pin NSOIC package.

APPLICATIONS

- Battery Charger Circuit
- NICs/Switches/Hubs
- ADSL Modems
- Negative Voltage Power Supply

FEATURES

- Supply Voltage: 3V 36V
- Current Limiting
- Output Switch Current to 1.5A
- Adjustable Output Voltage
- Operation frequency up to 180KHz
- Low Quiescent Current
- **Precision 2% Reference**
- Available in 8 pin NSOIC Package

TYPICAL APPLICATION DIAGRAM

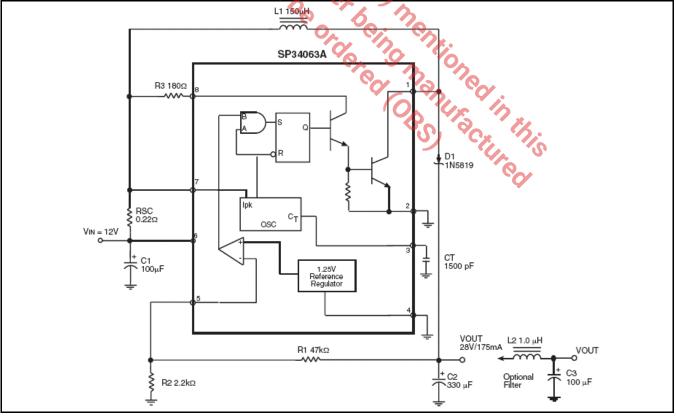


Fig. 1: SP34063A Application Diagram



ABSOLUTE MAXIMUM RATINGS

These are stress ratings only and functional operation of the device at these ratings or any other above those indicated in the operation sections of the specifications below is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

Power Supply Voltage V _{CC} 40V
Comparator Input Voltage0.3V to 40V
Switch Collector Voltage 40V
Switch Emitter Voltage (V _{PIN1} =40V)
Switch Collector to Emitter Voltage 40V
Driver Collector Voltage 40V
Driver Collector Current (Note 2)
Switch Current 1.5A
Storage Temperature
ESD Rating (HBM - Human Body Model) 2kV

OPERATING RATINGS

Input Voltage Range V _{IN}	3.0V to 36V
Power Dissipation (T _A =25°C - NSOIC) .	780mW
Junction Temperature Range	40°C to 150°C
Thermal Resistance θ_{JA}	160°C/W

ELECTRICAL SPECIFICATIONS

Specifications with standard type are for an Operating Temperature of $T_A = 25$ °C only. Minimum and Maximum limits are guaranteed through test, design, or statistical correlation. Typical values represent the most likely parametric norm at $T_A = 25$ °C, and are provided for reference purposes only. Unless otherwise indicated, $V_{CC} = 5.0V$, $T_A = -40$ °C to 85 °C.

Parameter	Min.	Тур.	Max.	Units		Conditions	
Oscillator Cx							
Frequency F _{osc}	30	38	45	KHz		$V_{PIN5} = 0V, CT = 1.0nF, T_A = 25^{\circ}C$	
Charge Current I_{CHG}	30	38	45	Αų	2	$V_{CC} = 5.0V$ to 36V, $T_A = 25^{\circ}C$	
Discharge Current I_{DISCHG}	180	240	290	μA		Vcc = 5.0V to 36V, T _A = 25°C	
Discharge to Charge Current Ration I_{DISCHG}/I_{CHG}	5.2	6.5	7.5	20	3	Pin 7 to V _{cc} , T _A = 25°C	
Current Limit Sense Voltage V _{IPK(sense)}	250	300	350	mV	20	$I_{\text{DISCHG}} = I_{\text{CHG}}, T_{\text{A}} = 25^{\circ}\text{C}$	
Output Switch (Note 1)					07		
Saturation Voltage, Darlington Connection V _{CE(sat)}		1.0	1.3	V		I _{sw} =1A, pin1,8 connected	
Saturation Voltage (note 2)		0.45	0.7	V		I_{sw} =1A, R pin8 = 820hms to V _{cc} , forced B=20	
DC Current Gain	50	75				$I_{SW}=1A, V_{CE} = 5V, T_A = 25^{\circ}C$	
Collector Off-State Current		0.01	100	μA		V _{CE} =36V	
Comparator							
Threshold Valtage V	1.225	1.250	1.275	V		T _A = 25°C	
Threshold Voltage V_{TH}	1.210	1.250	1.290	V		$T_A = -40^{\circ}C \text{ to } 85^{\circ}C$	
Threshold Voltage Line Regulation R_{EGLINE}		1.4	5	mV		$V_{CC} = 3.0V$ to 36V	
Input Bias Current I _{IB}		-20	-400	nA		V _{IN} = 0V	
Total Device							
Supply Current I _{cc}			4	mA		V_{CC} =5.0V to 36V, CT=1nF, pin 7 = V_{CC} Vpin5 >V _{TH}	

Note 1: Low duty cycle pulse techniques are used during the test program to maintain junction temperature as close to ambient temperature as possible.

Note 2: If the output switch is driven into hard saturation (non-Darlington configuration) at low switch currents (\leq 300mA), and high driver currents (\geq 30mA), it may take up to 2.0µs for it to come out of saturation. This condition will shorten the

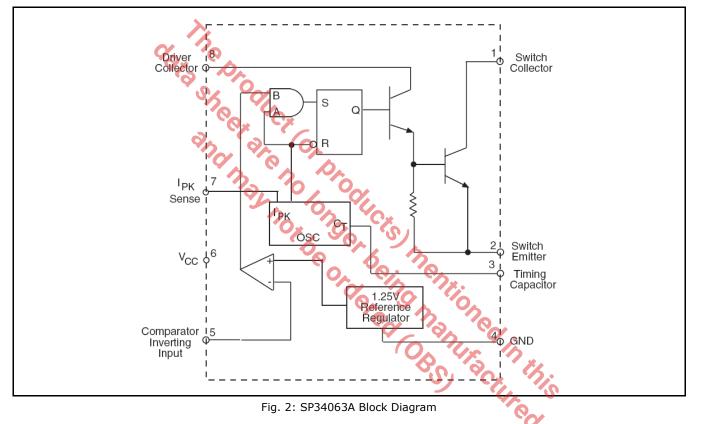


off time at frequencies above 30KHz, and is magnified at high temperatures. This condition does not occur with a Darlington configuration, since the output switch cannot saturate. If a non-Darlington configuration is used, the following output drive condition is recommended.

Forced ß of output switch =
$$\frac{I_C Output}{I_C Driver - 7mA*} \ge 10$$

* The 100 Ω resistor in the emitter of the driver device requires about 7.0mA before the output switch conducts.

BLOCK DIAGRAM



PIN ASSIGNEMENT

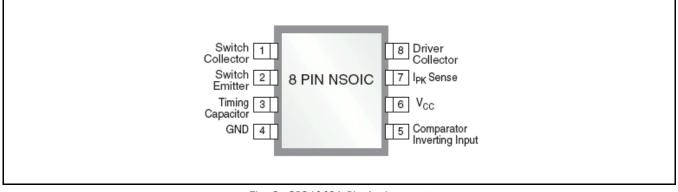


Fig. 3: SP34063A Pin Assignment



PIN DESCRIPTION

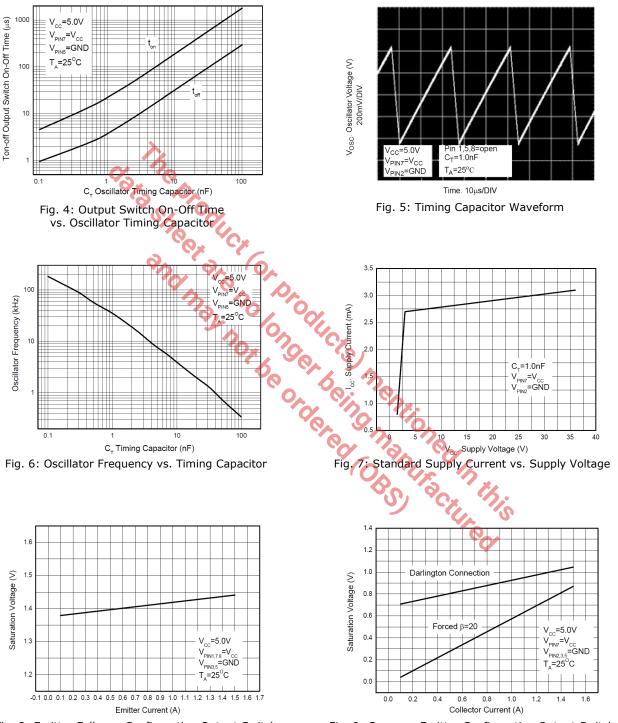
Name	Pin Number	Description			
Switch Collector	1	Internal switch transistor collector			
Switch Emitter	2	nternal switch transistor emitter			
Timing Capacitor	3	Timing capacitor to control the switching frequency			
GND	4	Ground pin for all internal circuit			
Comparator Inverting Input	5	Inverting input pin for internal comparator			
V _{cc}	6	Voltage supply			
I_{PK} Sense	7	Peak Current Sense Input by monitoring the voltage drop across an external I sense resistor to limit the peak current through the switch			
Driver Collector	8	Voltage driver collector			
ORDERING INFORMATION					

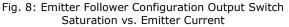
ORDERING INFORMATION

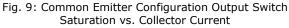
Part Number	Temperature Range	Package	Packing Quantity	Note 1	Note 2
SP34063AEN-L	-40°C≤T _A ≤+85°C	NSOIC-8	Bulk	Lead Free	
SP34063AEN-L/TR	-40°C≤T _A ≤+85°C	NSOIC-8	2.5K/Tape & Reel	Lead Free	
			dered manuel (OBS)	ned in this ractured	



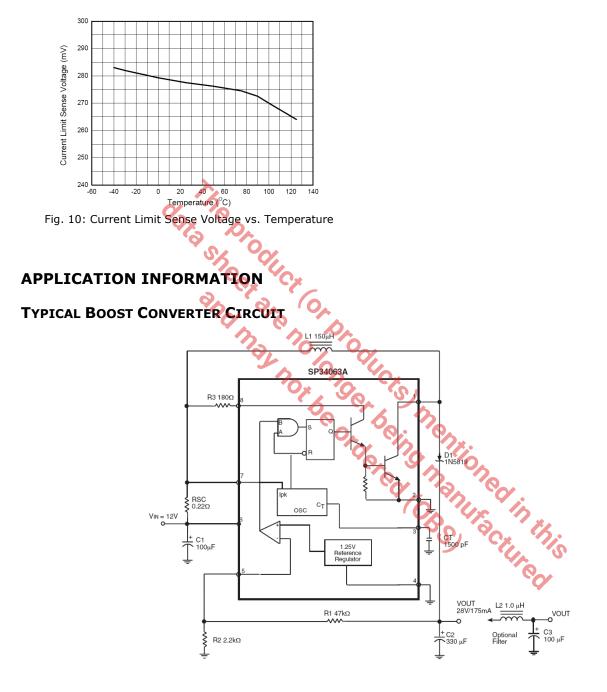
TYPICAL PERFORMANCE CHARACTERISTICS







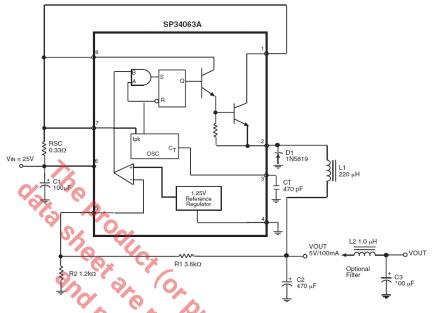




This is a typical boost converter configuration. In the steady state, if the resistor divider voltage at pin 5 is greater than the voltage in the non-inverting input, which is 1.25V determined by the internal reference, the output of the comparator will go low. At the next switching period, the output switch will not conduct and the output voltage will eventually drop below its nominal voltage until the divider voltage at pin 5 is lower than 1.25. Then the output of the comparator will go high, the output switch will be allowed to conduct. Since $V_{PIN5} = V_{OUT}*R2/(R1+R2) = 1.25(V)$, The output voltage can be decided by $V_{OUT} = 1.25V*(R1+R2)/R2(V)$.

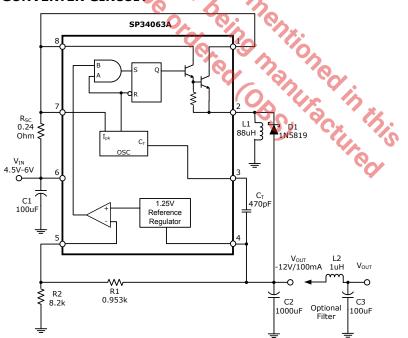


TYPICAL BUCK CONVERTER CIRCUIT



This is a typical buck converter configuration. The working process in the steady state is similar to a boost converter, $V_{PIN5} = V_{OUT}*R2/(R1+R2) = 1.25(V)$. The output voltage can be decided by $V_{OUT} = 1.25V*(R1+R2)/R2(V)$.

TYPICAL INVERTING CONVERTER CIRCUIT



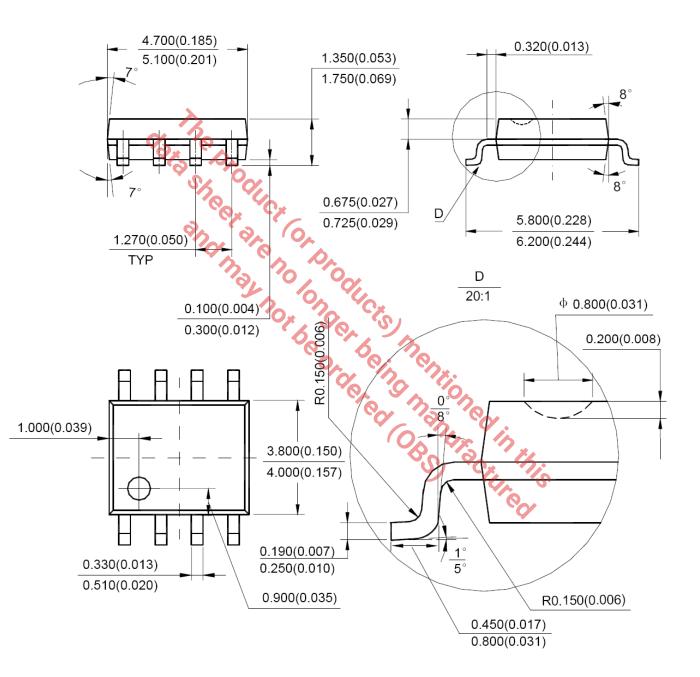
This is a typical boost converter configuration. The working process in the steady state is similar to a boost converter, the difference in this situation is that the voltage at the non-inverting pin of the comparator is equal to $1.25V + V_{OUT}$, then $V_{PIN5} = V_{OUT} * R2/(R1+R2) = 1.25V$, $+ V_{OUT}$. The output voltage can be decided by $V_{OUT} = -1.25V * (R1+R2)/R1$ (V).



PACKAGE SPECIFICATION

8-PIN NSOIC

Unit: mm (inch)





REVISION HISTORY

Revision	Date	Description			
2.0.0	01/16/2009	Reformat of Datasheet			
2.0.1	8/24/2010	Pg1, changed operation frequency from 110kHz to 180kHz Fig. 6: Changed title to: Oscillator Frequency vs. Timing Capacitor Pg7, Corrected the inverting converter circuit			
2.1.0	02/14/2011	Corrected Power Dissipation value to 780mW under Operating Ratings Added C_T =fct(T_{ON}) formula on figure 10 graph Updated package specification			
2.1.1	02/06/2014	Updated figure 4 and 6 [ECN 1407-07]			

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