

December 2009 Rev. 1.0.0

#### **GENERAL DESCRIPTION**

The XRP6840 is a controlled-current dual-cell supercapacitor charger and high power LED driver. Operating from a standard lithium-ion battery, the XRP6840 provides up to 4.3A of programmable Flash LED current and up to 600mA and 5.6V of programmable supercapacitor charging current and voltage.

Architectured around a 2.4MHz tri-mode 1x, 1.5x and 2x charge pump, the XRP6840 charges the stacked supercapacitor to the programmed output voltage with no more than 600mA drawn from the battery. While charging, an internal active balance circuitry insures matching of the stacked capacitors' voltages.

A standard 2-line I<sup>2</sup>C serial interface allows the dynamic programming of LED currents in torch and flash modes, flash timeout, channel enable, gain control, capacitor charge voltage, and enable. The XRP6840 comes in a 2 and 3 channel version supporting respectively 440mA/2.15A per channel and 300mA/1.45A in torch/flash modes.

The flexibility designed into the XRP6840 can also allow it to reverse the power flow back to the input to prevent unintended system resets as the battery voltage drops.

The XRP6840 is available in a lead-free, "green"/halogen free 20-pin TQFN package.

#### **APPLICATIONS**

- High Power Torch/Strobe/Flash LED
- High Resolution Cameras
- Generic High Power Lighting
- High Power White LED Backlighting

#### **FEATURES**

- Programmable 4.3A Flash LED Driver
  - Torch and Flash Modes
  - 2 and 3 Channels Versions
- Programmable Supercapacitor Charger
  - 600mA Adjustable Charging Current
  - Programmable Supercapacitor Voltage
  - In-rush Current Control
  - Active Voltage Balance Control
- Tri-mode Charge Pump Architecture
  - 1x, 1.5x, 2x Operation Modes
  - 2.4MHz Switching Frequency
- I<sup>2</sup>C Serial Interface
- Thermal, Over Current and Output Short Protection
- LED Short Detection
- RoHS Compliant "Green"/Halogen Free 20-pin 4mmx4mm TQFN package

#### TYPICAL APPLICATION DIAGRAM

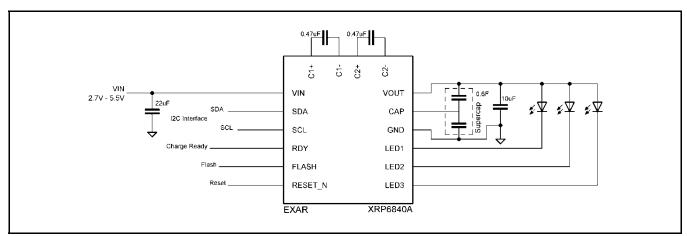


Fig. 1: XRP6840 – 3 Channels Application Diagram



#### **ABSOLUTE MAXIMUM RATINGS**

#### These are stress ratings only and functional operation of the device at these ratings or any other above those indicated in the operation sections of the specifications below is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

V <sub>IN</sub> , V <sub>OUT</sub> , LED <sub>1</sub> , LED <sub>2</sub> , LED <sub>3</sub> 0.3V to 6.0V
SCL, SDA, RDY, RESET_N, CAP0.3V to $V_{\text{IN}}$ +0.3V
C1P, C2P, C1N, C2N0.3V to $V_{\text{IN}}$ +0.3V
Storage Temperature65°C to 150°C
Power Dissipation Internally Limited (Note 1)
Lead Temperature (Soldering, 10 sec)260°C
ESD Rating (HBM - Human Body Model) All Pins 2KV

#### **OPERATING RATINGS**

Input Voltage Range V <sub>IN</sub>	2.7V to 5.5V
Junction Temperature Range	40°C to 125°C
Thermal Resistance $\theta_{JA}$	30.8°C/W

#### **ELECTRICAL SPECIFICATIONS**

Specifications with standard type are for  $T_J = 25^{\circ}\text{C}$  only; limits in applying over the full Operating Junction Temperature ( $T_J$ ) range are denoted by a "•". Minimum and Maximum limits are guaranteed through test, design, or statistical correlation. Typical values represent the most likely parametric norm at  $T_J = 25^{\circ}\text{C}$ , and are provided for reference purposes only. Unless otherwise indicated,  $V_{IN} = 3.4V$ ,  $V_{LED} = 0.8V$ ,  $C_{IN} = 47\mu\text{F}$ ,  $C_{F1} = C_{F2} = 0.47\mu\text{F}$ ,  $C_{OUT1}$  ( $V_{OUT}$  to CAP) =  $47\mu\text{F}$  and  $C_{OUT2}$  (CAP to GND) =  $47\mu\text{F}$ .  $T_A = -40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ ,  $T_J = -40^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ .

Parameter	Min.	Тур.	Max.	Units		Conditions
Operating Input Voltage Range	2.7		5.5	V	•	
UVLO Turn-On Threshold	2.2	2.4	2.6	V	•	V <sub>IN</sub> rising
UVLO Hysteresis		100		mV		V <sub>IN</sub> falling
Operating Input Current Including In-rush Current		600	725	mA	•	1x, 1.5x or 2x Mode
Torch Mode Input Current		202	228	mA		1x Mode, I <sub>OUT</sub> = 200mA
Torch Mode Input Current		302	340	mA		$1.5x$ Mode, $I_{OUT} = 200mA$
Standby Input Current		30	40	μΑ		All LEDs are Off, $T_{J(max)} = 85^{\circ}C$
Shutdown Supply Current		1	3	μΑ		$R_{ESET_N} = 0V$
Shutdown I <sub>LED</sub> Current		0		mA		STATUS1 Register [B7 B6] =00 Measure LED1 - LED3
Torch Mode Incremental current		20		mA		XRP6840A, Table 6, averaged from all channels
Step		30		mA		XRP6840B, Table 6, averaged from all channels
Flash Mode Incremental current		200		mA		XRP6840A, Table 5, averaged from all channels
Step		300		mA		XRP6840B, Table 5, averaged from all channels
Maximum Current in Torch Mode		300		mA		XRP6840A, Note 2, averaged from all channels
per Channel		440		mA		XRP6840B, Note 2, averaged from all channels
Maximum Current in Flash Mode		4.3		А		Flash Mode, Measured current into all LED pins
Outrot Coment DAC Besslution		3		Bit		Flash Mode
Output Current DAC Resolution		4		Bit		Torch Mode
DAG Command Assume to	-10		10	%	•	$120 \text{mA} < I_{\text{LED}} < 1.6 \text{A for XRP6840A} \\ 180 \text{mA} < I_{\text{LED}} < 1.6 \text{A for XRP6840B}$
DAC Current Accuracy	-12		12	mA	•	$I_{LED} \le 120$ mA, XRP6840A
	-18		18	mA	•	$I_{LED} \le 180 \text{mA}, XRP6840B$

### **XRP6840**

# 4.3A Supercapacitor Flash LED Driver with I<sup>2</sup>C

Parameter	Min.	Тур.	Max.	Units		Conditions
Channel to Channel Current Matching	-3		3	%	•	0mA < I <sub>LED</sub> < 1.6A Per Channel, Note 3
Switching Frequency	2.0	2.4	2.8	MHz	•	
Equivalent Resistance, 1X Mode		0.5		Ω	•	$V_{IN} = 3.7V$ , $I_{OUT} = 200mA$
Equivalent Resistance, 1.5X Mode		4	7	Ω	•	$\begin{split} &V_{\text{IN}} = 3.2 \text{V, } V_{\text{OUT}} < V_{\text{OUT\_LIMIT,}} \\ &I_{\text{OUT}} = 200 \text{mA, Equivalent resistance} = \\ &\left[ \left( V_{\text{IN}} \times 1.5 \right) - V_{\text{OUT}} \right] / I_{\text{OUT}} \end{split}$
LEDx Pin Dropout Voltage			0.36	V	•	Flash mode, $V_{\rm IN}=3.4V$ , $I_{\rm LED}=1A$ , Measure LED current at 95% $I_{\rm LEDX-NOMINAL}$ at $V_{\rm LEDX}{=}0.8V$
LEDx Pin Dropout Voltage			0.1	V	•	Torch mode, $V_{\rm IN}=3.4V,~I_{\rm LED}=80$ mA, Measure LED current at 95% $I_{\rm LEDX-NOMINAL}$ at $V_{\rm LEDX}{=}0.8V$
LED Short Detect Threshold	0.1	0.2	0.4	V	•	$V_{OUT}$ – $V_{LED}$ , LED = 0mA for all DAC code
Thermal Regulation			0.01	%/°C		
Thermal Shutdown Temperature		150		°C		Driver turn OFF
Thermal Shutdown Hysteresis		10		°C		Driver turn ON
Output Voltage Regulation (OVR)	5.2	5.3	5.4	V	•	Flash Mode, $I_{LED}=0$ mA, $V_{OUT}$ Rising STATUS2 Register: [B7 B6 B5] = 101 OVR value is set by STATUS2 register between 4.55V – 5.6V, all setting have +/- 100mV tolerance
Output Voltage Regulation Hysteresis		50		mV		Flash Mode, I <sub>LED</sub> = 0mA
RDY Pin Output Trip Threshold		V <sub>0</sub> - 0.1		V		STATUS2 Register: [B7 B6 B5] = 101 100mV below actual OVR value.
RDY Pin Sink Current	1			mA	•	
SDA, SCL, RESET_N, FLASH Input Logic Low Voltage			0.4	V	•	
SDA, SCL, RESET_N, FLASH Input Logic High Voltage	1.6			V	•	
Turn-Off Time (T <sub>OFF</sub> ) Into Shutdown				50	μs	RDY goes low when RESET_N goes from high top low
	0.09	0.11	0.13	S	•	STATUS1 register [B1 B0]=00
Flash Time Duration Before	0.18	0.22	0.26	S	•	STATUS1 register [B1 B0]=01
Standby	0.43	0.53	0.63	S	•	STATUS1 register [B1 B0]=10
	0.9	1.1	1.3	S	•	STATUS1 register [B1 B0]=11

### I<sup>2</sup>C SPECIFICATIONS

## I<sup>2</sup>C Addressing Format

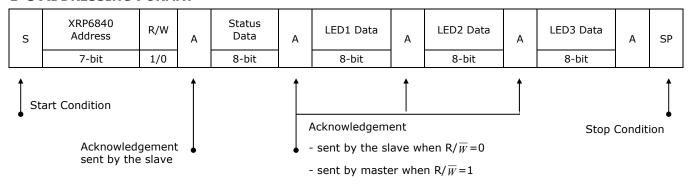


Fig. 2: I<sup>2</sup>C Data Input Format



Default I<sup>2</sup>C slave 7-bit address for XRP6840 is 0101000b

#### I<sup>2</sup>C TIMING CHARACTERISTICS

 $V_{IN}$  = 3.3V,  $T_A$  = -40°C to 85°C,  $T_J$  = -40°C to 125°C. Unless otherwise noted.

Parameter	Symbol	Min.	Тур.	Max.	Units	Conditions
Serial Clock Frequency	f <sub>SCL</sub>			400	KHz	
Bus Free Time between a STOP and a START	t <sub>BUF</sub>	1.3			μs	
Hold Time, Repeated START Condition	t <sub>HD_STA</sub>	0.6			μs	
STOP Condition Setup Time	t <sub>su,sto</sub>	0.6			ms	
Data Hold Time	t <sub>HD,DAT(OUT)</sub>	225		900	ns	
Input Data Hold Time	t <sub>HD,DAT(IN)</sub>	0		900	ns	
Data Setup Time	t <sub>su,dat</sub>	100			ns	
SCL Clock Low Period	t <sub>LOW</sub>	1.3			ms	
SCL Clock High Period	t <sub>HIGH</sub>	0.6			ms	
Rise Time of Both SDA and SCL Signals, receiving	t <sub>R</sub>	20 + 0.1Cb		300	ns	Note 4,5
Fall Time of Both SDA and SCL Signals, Receiving	t <sub>F</sub>	20 + 0.1Cb		300	ns	Note 4,5
Fall Time of SDA Transmitting	t <sub>F.TX</sub>	20 + 0.1Cb		250	ns	Note 4,5,6
Pulse Width of Spike Suppressed	$t_{SP}$	0		50	ns	Note 7
Capacitive Load for each Bus Line	Cb			400	pF	Note 4
I <sup>2</sup> C Startup Time after UVLO clears	t <sub>SRT</sub>			1	μs	Note 4

- Note 1: All parameter tested at  $T_A = 25$ °C. Specifications over temperature are guaranteed by design.
- Note 2: Current into all LED pins is up to 400mA continuously in Torch 1.5x mode.
- Note 3: LED current matching is calculated by this equation:
  - $\frac{I_{LED}-I_{AVG}}{I_{....}}\times 100\%$  where  $I_{AVG}$  is the average current of all channels.
- Note 4: Guaranteed by design.
- Note 5: Cb = total capacitance of one bus line in pF. tR and tF measured between  $0.3 \times V_{DD}$  and  $0.7 \times V_{DD}$ .
- Note 6:  $I_{SINK} \le 6mA$ . Cb =total capacitance of one bus line in pF.  $t_R$  and  $t_F$  measured between 0.3 x  $V_{DD}$  and 0.7  $V_{DD}$ .
- Note 7: Input filters on the SDA and SCL inputs suppress noise spikes less than 50ns.



#### **BLOCK DIAGRAM**

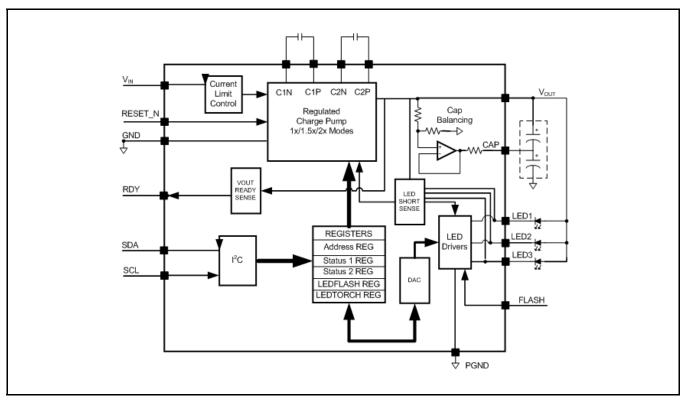


Fig. 3: XRP6840 Block Diagram (XRP6840A Shown)

#### **PIN ASSIGNEMENT**

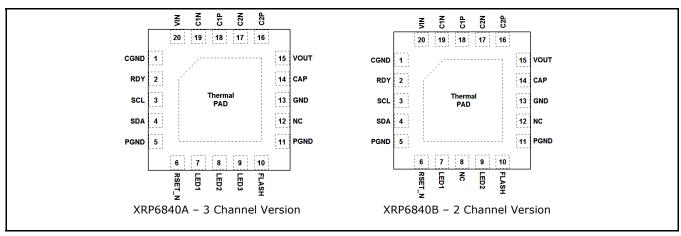


Fig. 4: XRP6840 Pin Assignment



### **PIN DESCRIPTION**

Name	Pin Number	Description
$C_GND$	1	Charge pump ground pin.
RDY	2	Active high push-pull output. RDY is high when $V_{\text{OUT}}$ reached to 100mV below its $V_{\text{OUT\_LIMIT}}$ voltage. The $V_{\text{OUT\_LIMIT}}$ for Flash mode is determined by STATUS2 register [B7 B6 B5]. The $V_{\text{OUT\_LIMIT}}$ for Torch mode is 4.50V.
SCL, SDA	3, 4	The SDA and SCL pins connect to the $\rm I^2C$ bus. Multiple functions can be programmed through his interface. They can also be used for read-back.
$P_{GND}$	5, 11	Power ground pin. The Source of internal NMOS is connected to this pin.
RESET_N	6	Active Low input pin.  If RESET_N = 0, then XRP6840 is in Shut-down mode  If RESET_N = 0 and STATUS1 register [B5] = 0, then reset all registers to logic low.  If RESET_N = 0 and STATUS1 register [B5] = 1, then all bits of all registers will be saved.
LED <sub>1</sub> , LED <sub>2</sub> , LED <sub>3</sub> (XRP6840A)	7, 8, 9	LED <sub>1</sub> , LED <sub>2</sub> , LED <sub>3</sub> connect to the drain of the internal NMOS which are current sources for LED current. These current sources are controlled by LEDFLASH or LEDTORCH registers which is programmed through I <sup>2</sup> C to provide the Torch and Flash current for
LED <sub>1</sub> , NC, LED <sub>2</sub> (XRP6840B)	7, 8, 9	the LEDs. $LED_1$ , $LED_2$ , $LED_3$ pins can be connected together to provide higher LED current. If a pin is not used connect it to $V_{OUT}$ . The XRP6840 incorporates a short LED protection circuit which shut-down LED current if LED voltage approaches to $V_{OUT\_LIMIT}$ .
FLASH	10	Digital Input pin. Active high. If STATUS1 register [B7 B6] = 11 and FLASH = 1 then LEDs are ON for one Flash timeout duration. Flash Timeout duration is controlled by STATUS1 register [B1 B0].
NC	12	No connection. This pin can be connected to P <sub>GND</sub> pin for heat sink.
GND	13	Ground pin. This ground pin doesn't carry high internal current.
САР	14	This pin is the output of an internal Op-Amp. This internal Op-Amp is powered by $V_{\text{OUT}}$ . The output voltage is half of $V_{\text{OUT}}$ , and output resistance is $470\Omega$ . The sink and source current is limited by $470\Omega$ output resistance. This provides active balancing between two internal sections of the super capacitor.
V <sub>оит</sub>	15	Output voltage. Connect positive terminal of SuperCap here. Connect the LEDs between this pin and the corresponding internal current source. Decouple with $10\mu F$ ceramic capacitor close to the pins of the IC.
$C_{2P}$ , $C_{2N}$	16, 17	Connect C <sub>2</sub> external flying capacitor between these pins.
C <sub>1P</sub> , C <sub>2P</sub>	18, 19	Connect C <sub>1</sub> external flying capacitor between these pins.
$V_{IN}$	20	Power supply input. Decouple with 10µF ceramic capacitor close to the pins of the IC.
Thermal Pad	-	Connect thermal pad to P <sub>GND</sub> pins.

### **ORDERING INFORMATION**

Part Number	Junction Temperature Range	Marking	Package	Package Packing Quantity		Note 2
XRP6840AILB-F	-40°C≤T <sub>A</sub> ≤+125°C	6840AI YYWWX	20-pin TQFN	Bulk	RoHS Compliant/ Halogen Free	3 Channels
XRP6840AILBTR-F	-40°C≤T <sub>A</sub> ≤+125°C	6840AI YYWWX	20-pin TQFN	3K/Tape & Reel	RoHS Compliant/ Halogen Free	3 Channels
XRP6840BILB-F	-40°C≤T <sub>A</sub> ≤+125°C	6840BI YYWWX	20-pin TQFN	Bulk	RoHS Compliant/ Halogen Free	2 Channels
XRP6840BILBTR-F	-40°C≤T <sub>A</sub> ≤+125°C	6840BI YYWWX	20-pin TQFN	3K/Tape & Reel	RoHS Compliant/ Halogen Free	2 Channels
XRP6840EVB	XRP6840 Evaluation	Board – XRP6	840A based.			

<sup>&</sup>quot;YY" = Year - "WW" = Work Week - "X" = Lot Number



### **TYPICAL PERFORMANCE CHARACTERISTICS**

All data taken at  $V_{IN}=2.7V$  to 5.5V,  $T_J=T_A=25^{\circ}C$ , unless otherwise specified - Schematic and BOM from Application Information section of this datasheet.

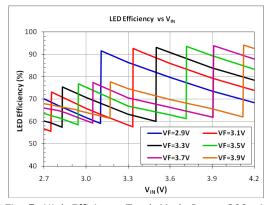


Fig. 5: High Efficiency Torch Mode  $I_{\text{LED}}$  at 200mA no supercap

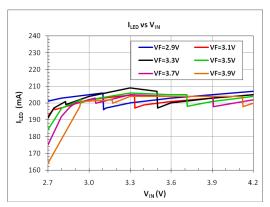


Fig. 6: High Efficiency Torch Mode, LED Current versus  $V_{\rm IN}$  no supercap

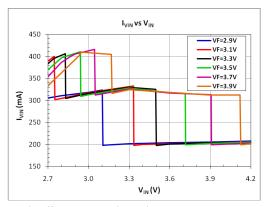


Fig. 7: High Efficiency Torch Mode, Input Current versus  $V_{\text{IN}}$   $I_{\text{LED}}$  at 200mA, no supercap

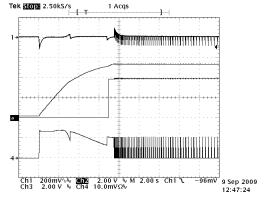


Fig. 8: VOUT RDY: 0.6F Supercap, CH1 =  $V_{IN}$ , CH2 =  $V_{OUT}$ , CH3 = RDY, CH4 =  $I_{VIN}$  = 0.5A/div

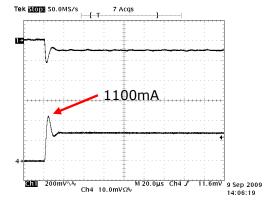


Fig. 9: In-rush Current with Li-ion Battery 0.6F Supercap,  $C_{IN}$ =10 $\mu$ F CH1 =  $V_{IN}$ , CH4 =  $I_{VIN}$  =0.5A/div

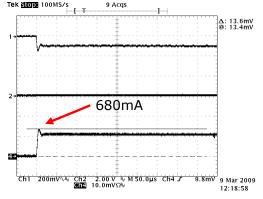


Fig. 10: In-rush Current, 0.6F Supercap,  $C_{IN}=22\mu F$ ,  $CH1=V_{IN}, CH4=I_{VIN}=0.5A/div$ 



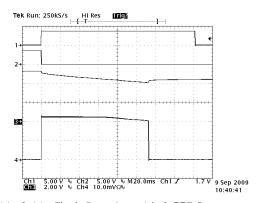


Fig. 11: 0.11s Flash Duration with 0.55F Supercap at 4.3A, LED  $V_F < 3.8V$ , CH1 = FLASH, CH2 = RDY, CH3 = $V_{OUT}$ , CH4 =  $I_{LED}$  =2A/div

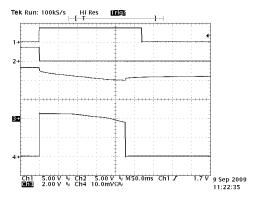


Fig. 12: 0.22s Flash Duration with 0.9F Supercap at 4.3A, LED  $V_F < 3.8V$ , CH1 = FLASH, CH2 = RDY, CH3 = $V_{OUT}$ , CH4 =  $I_{LED}$  =2A/div

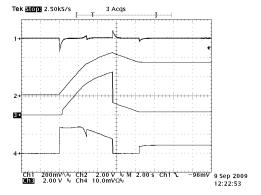


Fig. 13: 200mA High Efficiency Torch Mode with 0.55F Supercap,  $CH1=V_{IN}$ ,  $CH2=V_{OUT}$ ,  $CH3=LED_X$ ,  $CH4=I_{VIN}=0.5A/div$ 

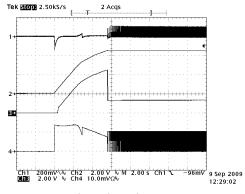


Fig. 14: 200mA Torch Mode with 0.55F Supercap, CH1=V $_{\rm IN}$ , CH2=V $_{\rm OUT}$ , CH3=LED $_{\rm X}$ , CH4=I $_{\rm VIN}$  =0.5A/div

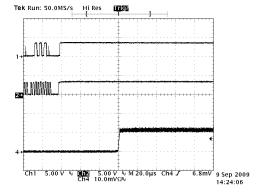


Fig. 15: LED Current Settling Time: 0mA to 100mA. CH1=SDA, CH2=SCL, CH4= $I_{LED}$ =0.1A/div

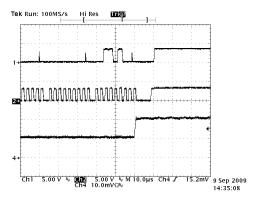


Fig. 16: LED Current Settling Time: 100mA to 200mA. CH1=SDA, CH2=SCL, CH4= $I_{\rm LED}$ =0.1A/div



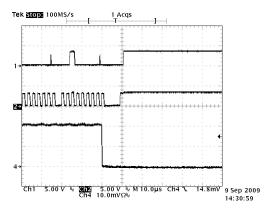


Fig. 17: LED Current Settling Time: 200mA to 0mA. CH1=SDA, CH2=SCL, CH4= $I_{LED}$ =0.1A/div

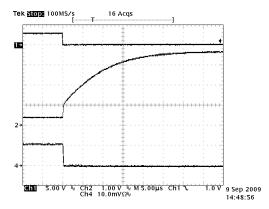


Fig. 18: LED Current Settling Time 200mA to Shutdown. CH1=RESET\_N, CH2=LEDx, CH4= $I_{\rm LED}$ =0.2A/div

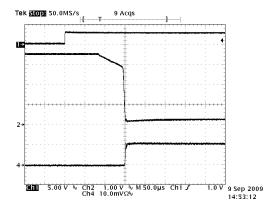


Fig. 19: LED Current Settling Time From Shutdown to 200mA. CH1=RESET\_N, CH2=LEDx, CH4= $I_{LED}$ =0.2A/div

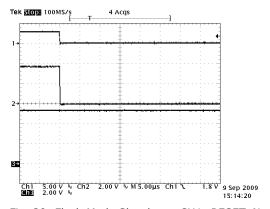


Fig. 20: Flash Mode Shutdown. CH1=RESET\_N, CH2=RDY, CH3= $V_{\text{OUT}}$ 

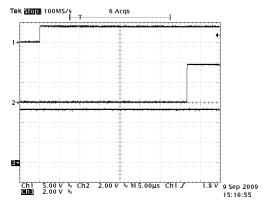


Fig. 21: Figure 21: Flash Mode Enable From Shutdown CH1=RESET\_N, CH2=RDY, CH3= $V_{\text{OUT}}$ 



#### APPLICATION INFORMATION

### I<sup>2</sup>C SERIAL INTERFACE

#### **REGISTERS**

The XRP6840 has five serially programmable data registers via the I<sup>2</sup>C interface. These registers can be reset to '0' through power on reset or setting RESET\_N, pin 6, to '0'. The first register is for Device Address as shown in Figure 3; it is activated by '28' HEX (7-bit format). The second register, STATUS1, contains the control word for programming operating modes, shutdown control, charge pump modes and flash timeout. The third register, STATUS2, contains the flash voltage level, read back of the Flash Ready, and fault conditions UVLO and over temperature. The fourth register, LEDFLASH, controls individual LED channels and current level in Flash Mode. fifth register, LEDTORCH, controls individual LED channels and current level in Torch Mode.

Register	В7	В6	В5	В4	В3	В2	В1	во
Address	0	1	0	1	0	0	0	R/W
STATUS1	Flash Torch Mode	Flash Torch Mode	Shut Down Control	Gain	Gain	-	Flash Timeout	Flash Timeout
STATUS2	Flash V	Flash V	Flash V	Flash Ready	UVLO Fault	Temp Fault	0	0
LEDFLASH	LED1	LED2	LED3*	D2	D1	D0	-	-
LEDTORCH	LED1	LED2	LED3*	D3	D2	D1	D0	-

Table 1: XRP6840  $I^2C$  Registers Bit Map

#### **STATUS1** Register

#### STATUS1 Register Bits B7, B6, B4 and B3

В7	В6	В4	В3	Operation Mode
0	1	0	0	High Efficiency Torch Mode
0	1	0	1	1X Torch Mode
0	1	1	0	1.5X Torch Mode
0	1	1	1	2X Torch Mode
1	0	0	0	Auto Gain Torch Mode
1	0	0	1	1X Torch Mode
1	0	1	0	1.5X Torch Mode
1	0	1	1	2X Torch Mode
1	1	0	0	Auto Gain Flash Mode
1	1	0	1	1X Flash Mode
1	1	1	0	1.5X Flash Mode
1	1	1	1	2X Flash Mode

Table 2: STATUS1 Register Operation Modes

#### STATUS1 Register Bit B5

This bit is used to save or reset the contents of STATUS1, STATUS2, LEDFLASH and LEDTORCH registers for the next I<sup>2</sup>C command when RESET\_N, pin 6, is '1'. If B5 is '1' then all bits in these registers will be saved. Otherwise if B5 is '0' then all bits will be reset to '0'.

### STATUS1 Register Bits B1 and B0

These two bits program the flash timeout duration as follows:

B1	во	Flash Duration
0	0	0.11s
0	1	0.22s
1	0	0.53s
1	1	1.10s

Table 3: Charge Pump Mode Selection

#### **STATUS2** Register

#### Flash Mode Voltage Programming

V<sub>OUT\_LIMIT</sub> in Flash Mode is regulated with 50mV of hysteresis and is programmed through bits B7, B6 and B5 of STATUS2 register as follows:

В7	В6	В5	Comment
0	0	0	4.55V
0	0	1	4.70V
0	1	0	4.85V
0	1	1	5.00V
1	0	0	5.15V
1	0	1	5.30V
1	1	0	5.45V
1	1	1	5.60V

Table 4: Flash Mode  $V_{\text{OUT\_LIMIT}}$  Settings

#### Bit B4

RDY bit is available for  $I^2C$  read-back. This bit is set to '1' when  $V_{\text{OUT}} > V_{\text{OUT\_LIMIT}}$ , and set to '0' otherwise.

#### Fault mode read-back bits B3, B2

These bits are designed for 2 Fault Mode flags and are also available for  $I^2C$  read-back. Bits 3 and 2 are set to '1' when UVLO and over temperature conditions are detected, as shown in Table 1.

#### **LEDFLASH Register**

LEDFLASH register, bits B7, B6 and B5 are used to activate the LED outputs channels as summarized in Table 1; they correspond to  $LED_1$ ,  $LED_2$  and  $LED_3$ . Bits B4, B3 and B2

<sup>\*</sup> Not used for XRP6840B



represent the DAC codes D2, D1, D0. They are used to set the flash LED current levels in each channel. Table 5 provides the DAC codes and the corresponding nominal current levels for each channel:

D2-D0 Code	XRP6840A I <sub>OUT</sub> /Ch. (mA)	D2-D0 Code	XRP6840B I <sub>OUT</sub> /Ch. (mA)
000	0	000	0
001	400	001	597
0 1 0	586	0 1 0	875
0 1 1	770	0 1 1	1127
100	948	100	1400
101	1197	101	1671
1 1 0	1291	1 1 0	1910
111	1445	111	2150

Table 5: Nominal Flash Mode Output Current

#### **LEDTORCH Register**

LEDTORCH register bits B2 to B7, also summarized in Table 1; they correspond to LED1, LED2 and LED3. Bits B4, B3, B2, and B1 represent the DAC codes D3, D2, D1, D0. They are used to set the torch LED current levels in each channel. Table 6 provides the DAC codes and their corresponding nominal current levels for each channel. Remember that the total current that can be supported in torch mode is 600mA divided by the gain of the charge pump. If 2 channels are set to 440mA (a total of 880mA), even with a gain of 1X, the input current limit will clamp the total current to approximately 600mA.

D3-D0 Code	XRP6840A I <sub>OUT</sub> /Ch. (mA)	D3-D0 Code	XRP6840B I <sub>OUT</sub> /Ch. (mA)
0000	0	0000	0
0001	23	0001	34
0010	46	0010	66
0011	66	0011	97
0100	86	0100	127
0101	105	0 1 0 1	155
0110	125	0 1 1 0	185
0 1 1 1	145	0 1 1 1	215
1000	165	1000	245
1001	185	1001	275
1010	205	1010	305
1011	225	1011	330
1100	245	1100	360
1101	260	1101	385
1110	280	1110	415
1111	300	1111	440

Table 6: Torch Mode Output Current

#### **Addressing and Writing Data**

To write data to the XRP6840 one of the following two sequences is required:

Easy shutdown/startup sequence

[Slave Address with write bit][Data for Status]

Full shutdown/startup sequence

[Slave Address with write bit][Data for Status]
[Data for LEDFLASH][Data for LEDTORCH]

Slave address is '28' Hex.

#### **Addressing and Reading Data**

To read data from the XRP6840 the following sequence is required:

[Slave Address with read bit][Data for Status]
[Data for LEDFLASH][Data for LEDTORCH]



#### THEORY OF OPERATION

When  $V_{\text{IN}}$  reaches 2.7V during initial power up, a power on reset signal will be issued to reset all the registers and the internal logic, and the system will be ready for operation.

To be in operation the XRP6840 must be enabled through RESET\_N, pin 6. The LEDs can be activated through LEDFLASH or LEDTORCH registers.

#### CHARGE PUMP MODES 1x, 1.5x and 2x

The XRP6840 charge pump can operate in 1x, 1.5x and 2x modes to ensure desired current regulation. Once the output reaches  $V_{\text{OUT\_LIMIT}}$ , the charge pump will turn off.

In 1x mode, the input is simply connected to the output through an internal  $0.5\Omega$  MOSFET. An internal in-rush current limit will keep the charging current controlled to a maximum of 600mA.

In 1.5x mode, during the first clock phase the 2 flying capacitors,  $C_1$  and  $C_2$ , are series connected between  $V_{IN}$  and ground, charging each capacitor to  $\frac{1}{2}$   $V_{IN}$ . In the second clock phase, the flying capacitors are parallel connected and placed in series with  $V_{IN}$ , producing  $1.5 \text{x} V_{IN}$  to be discharged across the output capacitor. Note the maximum total output current is now  $600 \text{mA} \div 1.5$  or 400 mA.

In 2x mode, during the first clock phase, one flying capacitors is charged to  $V_{IN}$ . In the second clock phase, this capacitor is placed in series with  $V_{IN}$ , producing  $2xV_{IN}$  to be discharged across the output capacitor. Note the maximum total output current is now  $600\text{mA} \div 2$  or 300mA.

#### **CHARGE PUMP CONTROLLER**

The charge pump controller regulates the output by turning the charge pump off once the output reaches  $V_{\text{OUT\_LIMIT}}$ . For both Torch and Flash Modes, there is 50mV of output voltage hysteresis before the charge pump is re-enabled. However, in High Efficiency Torch mode, it will only turn on again when  $V_{\text{LEDX}}$  is below the drop out voltage of approximately 0.36V. Once the charge pump turns on again, it will then monitor both  $V_{\text{OUT}}$  and  $V_{\text{LEDX}}$  voltages. This operation will be explained

further in the High Efficiency Torch Mode section.

The XRP6840 can operate in either "Auto Gain Mode" or "Programmable Gain Mode". However, either mode will be overridden during start-up based on the following internal control algorithm:

- 1. If  $V_{OUT} > VOUT_{LIMIT}$  then the charge pump is OFF.
- 2. If  $V_{OUT} < V_{IN} 1V$  then the charge pump can only operate in 1x mode.
- 3. If  $V_{IN}$   $1V < V_{OUT} < V_{IN}$  then the charge pump is allowed to operate in either 1x or 1.5x mode but not in 2x mode.
- 4. If  $V_{OUT} > V_{IN}$  then the charge pump can operate either in 1.5x or 2x mode but not in 1x mode (there is no reverse current limit on the 0.5 $\Omega$  bypass MOSFET).
- 5. If  $V_{OUT} > V_{OUT\_LIMIT}$ , and all LEDs are deactivated then XRP6840 will operate in standby mode with 30µA supply current.

 $V_{\text{OUT\_LIMIT}}$  is 4.5V for Torch Mode and High Efficiency Torch Mode; for Flash mode  $V_{\text{OUT\_LIMIT}}$  is shown in Table 4.

Standby mode is set by either LEDFLASH or LEDTORCH registers [B7 B6 B5] to [0 0 0]. This will deactivate all the LED channels.

#### **AUTO GAIN START-UP**

The Auto Gain Start mode is the fastest way to charge the output toward VOUT\_LIMIT. It is initiated with the following conditions; as shown in Table 7 (a sub-set of Table 2), in STATUS1 Register.

В7	В6	В4	В3	3 Operation Mode	
0	1	0	0	High Efficiency Torch Mode	
1	0	0	0	Auto Gain Start Torch Mode	
1	1	0	0	Auto Gain Start Flash Mode	

Table 7: Auto Start Mode from STATUS1 Register

When in Auto Gain Start Mode, the XRP6840 will charge up  $V_{\text{OUT}}$  to  $V_{\text{OUT\_LIMIT}}$  with maximum available current within the constraints defined above for start-up.

In this mode the XRP6840 will initialize the system as follows:

 Turn off all the LEDs that were programmed through LEDTORCH or LEDFLASH registers.



 Determine if VOUT\_LIMIT is reached. If this condition is met anytime, the charge pump will turn off, release the control of the LEDTORCH or LEDFLASH registers, and will be ready for turning on the LEDs.

After system initialization, the XRP6840 will automatically switch between 1x, 1.5x, or 2x mode modes to try and keep input current to a maximum without exceeding 600mA. Auto Start Mode ends when  $V_{\text{OUT}}$  reaches  $V_{\text{OUT}}$  LIMIT.

The Auto Start Mode begins in 1x mode or 1.5x mode if  $V_{OUT} > V_{IN}$ . When the input current drops to approximately 400mA the XRP6840 is able to switch from 1x to 1.5x mode without exceeding 600mA. The switch from 1.5x mode to 2x mode occurs if  $V_{OUT}$  has not reached  $V_{OUT\_LIMIT}$  and when the input current is below 300mA. After switching to 2x mode, the charge pump continues to operate until  $V_{OUT\_LIMIT}$  is reached.

If  $V_{\text{OUT\_LIMIT}}$  could not be reached, then the XRP6840 will continuously run the charge pump in 2x mode. At this point, it is recommended to shut down through  $I^2C$ , RESET\_N to '0' or power down.

There are probably 2 main causes for the XRP6840 to remain in 2x mode:  $V_{\text{OUT}}$  overloaded or  $V_{\text{IN}}$  too low. If the output were shorted, the current in 1x mode would never drop below 600mA and the XRP6840 would never move to the higher gains. Removing the short or overload will allow the XRP6840 to recover back to normal operation. If  $V_{\text{IN}}$  is too low then it is advisable to power down the system and change the battery.

#### **PROGRAMMABLE START MODE**

Programmable start mode allows the flexibility to select the maximum gain. Start-up gains are based strictly on input and output voltage differences. The 600mA input current limit is still imposed, but typically is not triggered. This is why Auto Start Mode is recommended because it charges the output capacitor the fastest.

Even though the maximum gain is programmed, the 5 stage internal control algorithm on the previous page will still override the programmed gain to ensure optimum operation and reverse current

protection in 1x mode. Reverse current is possible in 1.5x or 2x modes and that leads to interesting applications possibilities.

#### **REVERSING THE POWER FLOW**

By forcing the charge pump into 1.5x mode once the output voltage is charged in Flash Mode, it is possible to reverse the current back to the input if the input voltage is <3.6V. This gives one the possibility of using the XRP6840 Flash system to provide power back to the input when other parts of the system require unusually high loads. A couple of examples are; a high power burst transmit, or HDD spin up.

In Flash Mode, charge the output capacitor to 5.45V (STATUS2 Register B7-B5 = 110) using Auto Start Mode. When power is required back at the input, force to 1.5x mode (STATUS1 Register B4-B3 = 10) and change  $V_{\text{OUT\_LIMIT}}$  to 5.60V (STATUS2 Register B7-B5 = 111). This will have the effect of instantly turning on the charge pump in 1.5x mode. If the input voltage is 3V, then approximately 300mA will flow from the output super capacitor to the input for 650ms. The current will drop as the super capacitor discharges, but as much as 150mA is available after 1.5seconds.

A more complete application note will be forthcoming on this topic.

#### **HIGH EFFICIENCY TORCH MODE**

High Efficiency Torch Mode is a special XRP6840 feature designed to achieve the highest torch mode power efficiency. This mode is activated only when STATUS1 [B7 B6 B4 B3] are [0 1 0 0]. LEDTORCH [B7 B6 B5] can be set according to the number of desired LED channels.

First, the charge pump will charge  $V_{\text{OUT}}$  to  $V_{\text{OUT\_LIMIT}}$  then turn off. Then when the LEDs are enabled, the control loop will adjust  $V_{\text{OUT}}$  to the minimum value required to maintain current regulation. It does this by monitoring and regulating the  $V_{\text{LEDX}}$  voltage with respect to an adaptive drop out voltage,  $V_{\text{DO}}$ .

The adaptive  $V_{DO}$  algorithm will allow  $V_{LEDX}$  to drop down to 220mV for 1x mode and 180mV for both 1.5x and 2x modes. This will ensure the XRP6840 stays in the lowest charge pump



mode to maintain good current regulation. As Figure 5 shows, very high efficiency can be obtained if the system can operate at the lowest available charge pump gain at very low  $V_{\rm IN}$  levels. Due to the XRP6840's low drop out voltage design, good current regulation can still be achieved at extremely low  $V_{\rm LED}$  voltage levels.

When  $V_{LEDX}$  drops below  $V_{DO}$ , and  $V_{OUT}$  is also below  $V_{OUT\_LIMIT}$ , the XRP6840 charge pump will turn on to charge the  $V_{OUT}$  and  $V_{LEDX}$ . If  $V_{LEDX}$  is still below  $V_{DO}$  after 64 clock cycles or 27µs (at 2.4MHz), the XRP6840 will jump to the next higher gain. If at any time,  $V_{OUT\_LIMIT}$  is reached, then charge pump will turn off.

#### **FAULT PROTECTION**

Although most of these modes of operation have already been previously described, they are repeated here to emphasize the robustness of the XRP6840.

The output voltage is directly monitored and controlled through the  $V_{\text{OUT}}$  pin. Should an open occur on the  $V_{\text{OUT}}$  pin, the output is disconnected from the input and no damaging voltages will be applied to the output super capacitor due to an open loop condition.

The XRP6840 also has a built-in over current protection because when the output is shorted, the XRP6840 will force the part to be in 1x mode and its output current is always limited to 600mA regardless of whether or not the part is in Auto or Programmed Start Modes.

The XRP6840 also has thermal protection. If the junction temperature rises above 150°C, the part is disabled. Once the temperature drops below 140°C the part is re-enabled.

#### **OPEN AND SHORT LED PROTECTION**

An open LED has no real effect on the operation of the XRP6840.

If an LED fails short, that channel will be disabled. The short condition is asserted when  $V_{\text{OUT}}$  to  $V_{\text{LEDX}}$  is less than 200mV. Only the driver of the shorted LED will be turned off and no current will flow. However, the other channels will continue to operate as intended, independently of the shorted channel.

#### **COMPONENT SELECTION**

The XRP6840 charge pump circuit requires the following capacitors:

Input Cap: 22μF
 Output Cap: 10μF

Charge Pump Flying caps: 2x0.47μF
 Supercapacitor: 0.6F (0.3F to 0.9F).

Input capacitances higher than  $10\mu F$  will help reduce input voltage ripple and in-rush current. Refer to Figures 9 and 10 for comparison.

The input and output capacitors should be located as close to the  $V_{\rm IN}$  and  $V_{\rm OUT}$  pins as possible to obtain best bypassing. Their returns paths should be connected directly to the  $P_{\rm GND}$  pin or to the thermal pad ground located under the XRP6840. The flying capacitors should be located as close to the  $C_{\rm 1P}$ ,  $C_{\rm 1N}$  and  $C_{\rm 2P}$ ,  $C_{\rm 2N}$  pins as possible.

All the capacitors should be surface mount ceramic types for low ESR and for low lead inductances. These capacitors can also improve bypassing. X5R or X7R temperature grades are recommended for this application.

The supercapacitor selection depends on LED current, flash duration, and LED forward voltage. The minimum super capacitor is determined as follows:

$$C = \frac{I_{OUT} \times T_{FLASH}}{V_{OUT\ LIMIT} - V_F - V_{LED} - I_{OUT} \times R_{CAP\ ESR}}$$

 $I_{OUT}$  is the target flash current,  $T_{FLASH}$  is the flash duration,  $V_{OUT\_LIMIT}$  is the initial Super Capacitor voltage programmed by STATUS2 register bits [B7 B6 B5],  $V_F$  is the LED forward voltage and  $V_{LED}$  is the maximum LED pin voltage before dropout (360mV at 125°C).

For example, for a 4.3A flash with 110ms duration, LED  $V_F$  of 3.5V and  $V_{OUT\_LIMIT}$  voltage of 5.3V, the minimum capacitance is:

$$C = \frac{4.8A \times 0.11s}{5.3V - 3.5V - 0.36V - 4.8A \times 0.05\Omega} \cong 0.44F$$



To achieve 4.3A flash pulses, we recommend using 0.55F Super Capacitor from TDK EDLC2720-501-2F-50 with voltage rating of

5.5V and  $50m\Omega$  of ESR, or the 0.6F HS206F Super Capacitor from CAP-XX with a voltage rating of 5.5V and  $80m\Omega$  of ESR.

### TYPICAL APPLICATION SCHEMATICS

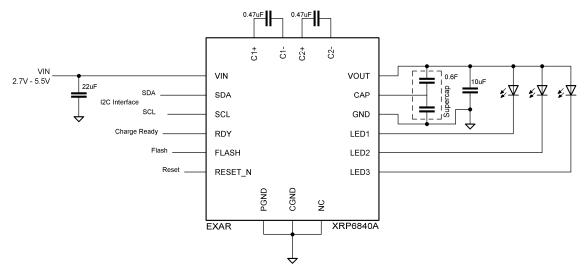
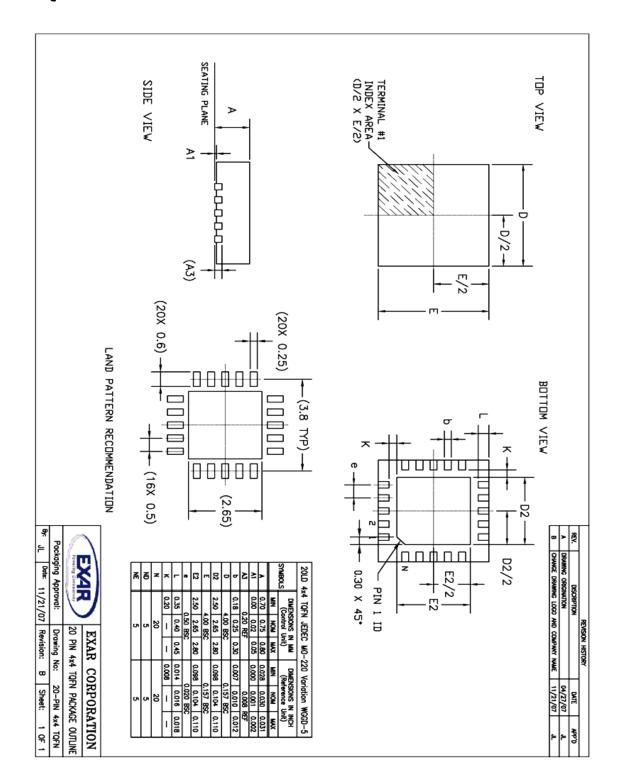


Fig. 22: XRP6840A - 3 Channel - 4.3A Total LED Flash Current



### **PACKAGE SPECIFICATION**

### **20-PIN TQFN**





#### **REVISION HISTORY**

Revision	Date	Description
1.0.0	12/05/2009	Initial Release of Data Sheet

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