









RT5707/A

Ultra-Low Quiescent Current HCOT Buck Converter

1 General Description

The RT5707/A is a high-efficiency synchronous buck converter featuring a typical quiescent current of 360nA. It provides high efficiency at light loads, down to 10mA with the input voltage range from 2.2V to 5.5V.

The RT5707 provides eight programmable output voltages ranging from 1.2V to 3.3V, while delivering an output current of up to 600mA, peaking at 1A. The RT5707A provides eight programmable output voltages ranging from 0.7V to 3.1V, and it can deliver an output current of up to 400mA, with a peak at 0.5A. The Hysteretic Constant-On-Time (HCOT) operation with internal compensation allows the transient response to be optimized across a wide range of loads and output capacitors. The RT5707/A is available in a WL-CSP-8B 0.9x1.6 (BSC) package.

The recommended junction temperature range is -40°C to 125°C, and the ambient temperature range is -40° C to 85° C.

2 Ordering Information

RT5707/A □ Package Type⁽¹⁾ WSC: WL-CSP-8B 0.9x1.6 (BSC)

Note 1.

Richtek products are Richtek Green Policy compliant and marked with (1) indicates compatible with the current requirements of IPC/JEDEC J-STD-020.

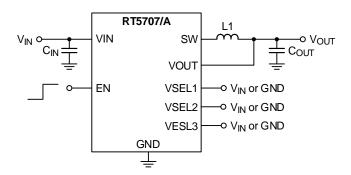
3 Features

- Input Voltage Range: 2.2V to 5.5V
- Eight-Level Programmable Output Voltage
 - RT5707 1.2V to 3.3V
 - RT5707A 0.7V to 3.1V
- Typical 360nA Quiescent Current
- PSM Operation
- Up to 94% Efficiency
- Internal Compensation
- Output Voltage Discharge
- Overcurrent Protection
- Over-Temperature Protection
- Output Current:
 - RT5707 600mA, Peak to 1A
 - RT5707A 400mA, Peak to 0.5A
- Automatic Transition to 100% Duty Cycle Operation

4 Applications

- Hand-Held Devices
- Portable Information
- **Battery Powered Equipment**
- Wearable Devices
- Internet of Things
- Smart Watches

5 Simplified Application Circuit



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6 Marking Information

RT5707WSC



6E: Product Code W: Date Code RT5707AWSC



6Z: Product Code W: Date Code



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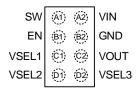
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7 Pin Configuration

(TOP VIEW)



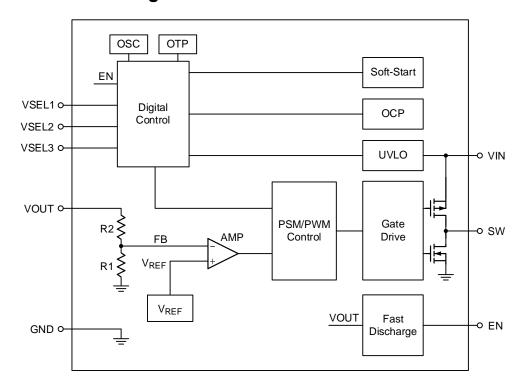
WL-CSP-8B 0.9x1.6 (BSC)

8 Functional Pin Description

Pin No.	Pin Name	Pin Function
A1	SW	This pin serves as the interface for the internal switches integrated within the IC and must be connected to an external inductor. It is imperative to connect the inductor to this pin using the shortest possible trace length to minimize parasitic inductances.
A2	VIN	Supply input. A minimum of a $10\mu F$ (for the RT5707) and a $4.7\mu F$ (for the RT5707A) ceramic capacitor must be connected to this pin using the shortest path.
B1	EN	Chip enable input pin. A high voltage level enables the device, while a low voltage level turns the device off. This pin must be properly terminated.
B2	GND	Device ground pin. This pin should be connected to input and output capacitors using the shortest path.
C1	VSEL1	Output voltage selection pin. This pin requires terminated.
C2	VOUT	Output voltage feedback pin. This pin should be connected close to the output capacitor terminal to achieve better voltage regulation. A minimum of a $10\mu F$ ceramic capacitor must be connected to this pin using the shortest path.
D1	VSEL2	Output voltage selection pin. This pin requires terminated.
D2	VSEL3	Output voltage selection pin. This pin requires terminated.



9 Functional Block Diagram





10 Absolute Maximum Ratings

(Note 2)

- Power Dissipation, PD @ TA = 25°C
 - WL-CSP-8B 0.9x1.6 (BSC) ----- 0.84W
- Package Thermal Resistance (Note 3)
 - WL-CSP-8B 0.9x1.6 (BSC), θJA ------ 118.5°C/W
- Lead Temperature (Soldering, 10 sec.) ------ 260°C
- Junction Temperature ----- -40°C to 150°C
- ESD Susceptibility (Note 4)
 - HBM (Human Body Model)----- 2kV
- Note 2. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 3. θJA is measured under natural convection (still air) at TA = 25°C with the component mounted on a high effectivethermal-conductivity four-layer test board on a JEDEC 51-7 thermal measurement standard.
- Note 4. Devices are ESD sensitive. Handling precautions are recommended.

11 Recommended Operating Conditions

(Note 5)

- Supply Input Voltage ----- 2.2V to 5.5V • RT5707 Output Current $(5.5V \ge VIN \ge (Vout_Nom + 0.7V) \ge 3V)$ ------ 0mA to 600mA
- RT5707A Output Current (5.5V \geq VIN \geq (VOUT_NOM + 0.7V) \geq 3V) ------ 0mA to 400mA

Note 5. The device is not guaranteed to function outside its operating conditions.

12 Electrical Characteristics

 $(V_{IN} = 3.6V, C_{IN} = C_{OUT} = 10\mu F, L1 = 2.2\mu H, T_A = 25^{\circ}C$, unless otherwise specified.)

Parameter	Symbol	mbol Test Conditions		Тур	Max	Unit				
BUCK Regulator										
Undervoltage-Lockout Rising Threshold	Vuvlo_r	VIN rising	1	2	2.15	V				
Undervoltage-Lockout Hysteresis	Vuvlo_HYS			0.1	0.4	V				
VOLIT Voltage Acquire ov	VOUT_ACC10	Vout = 1.8V, Iout = 10mA	-2.5		2.5	%				
VOUT Voltage Accuracy	VOUT_ACC100	Vout = 1.8V, Iout = 100mA	-2		2	70				



Parameter	Symbol	Test Condition	Min	Тур	Max	Unit	
lanut Ouissaant Current	IQ_NSW VOUT = 1.8V, IOUT = 0A, EN = VIN, non-switching			360	800	A	
Input Quiescent Current	IQ_SW	Vout = 1.8V, lout = 0/ switching	A, EN = VIN,		460	1200	nA
Shutdown Current	ISHDN	EN = GND			0.2	1	μΑ
Switching Frequency	fsw	Vout = 1.8V, CCM mo	de		1.2		MHz
UGATE Current Limit	1	3V ≤ VIN ≤ 5.5V	RT5707	1	1.2	1.4	٨
OGATE Current Limit	ILIM_H	$3V \leq VIN \leq 5.5V$	RT5707A	0.68	0.78	0.88	A
LGATE Current Limit	li ina i	$3V \le VIN \le 5.5V$	RT5707	1	1.2	1.4	Α
LGATE Current Limit	ILIM_L	$3V \leq VIN \leq 5.5V$	RT5707A	0.55	0.68	0.8	A
On-Resistance of High- Side MOSFET	RDSON_H	IOUT = 50mA			350		mΩ
On-Resistance of Low-Side MOSFET	RDSON_L	IOUT = 50mA			250		mΩ
Output Discharge Resistor	RDISCHG	EN = GND, $IOUT = -10$	mA		10		Ω
VOUT Pin Input Leakage	ILK	Vout = 2V, EN = Vin			100		nA
VOUT Minimum Off-Time	toff_MIN				80		ns
VOUT Minimum On-Time	ton_min	Vout = 1.8V, Vin = 3.6	V		420		ns
Line Regulation	VLINE_REG	VOUT = 1.8V, IOUT = 100mA, VIN = 2.2V to 5.5V			0.1		%/V
Load Regulation	VLOAD REG	Vout = 1.8V, including PFM operation			0.001		%/mA
3		Vout = 1.8V, only CCN	/I operation		0.0005		
Over-Temperature Protection	Тотр				150		°C
Over-Temperature Protection Hysteresis	Totp_HYS				20		°C
Auto 100% Duty Cycle Leave Detection Threshold	VTH_100+	Rising VIN, 100% mod VIN = VOUT + VTH_100+		150	250	350	mV
Auto 100% Duty Cycle Enter Detection Threshold	VTH_100-	Falling V _{IN} , 100% mod with V _{IN} = V _{OUT} + V _{TH}		85	200	290	mV
Timing	•			l.	l.		
Regulator Start-Up Delay Time	tDLY	IOUT = 0mA, EN = GND to VIN, Vout starts rising			0.1		ms
Regulator Soft-Start Time	tss	VOUT = 1.8V, IOUT = 10mA, EN = VIN			0.7		ms
Logic Input (EN, VSEL1, V	SEL2, and VSE	L3)		•			•
Input High Threshold	VIH	VIN = 2.2V to 5.5V		1.2			V
Input Low Threshold	VIL	V _{IN} = 2.2V to 5.5V				0.4	V
Input Pin Bias Current	IBIAS				10		nA

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13 Typical Application Circuit

13.1 For the RT5707

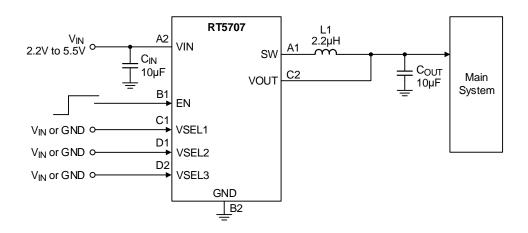


Table 1. Recommended Components Information

Reference Part Number		Description	Package	Manufacturer
CIN, COUT	GRM155R60J106ME15	10μF/6.3V/X5R	0402	Murata
L1	1239AS-H-2R2M	2.2μΗ	2520	Murata

13.2 For the RT5707A

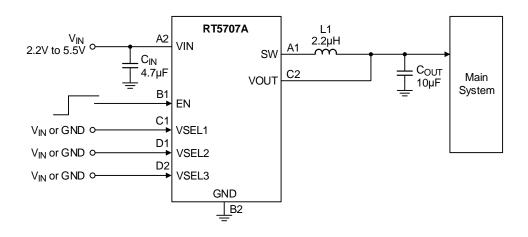
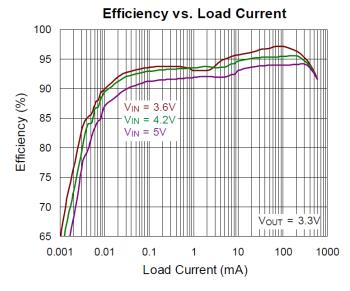


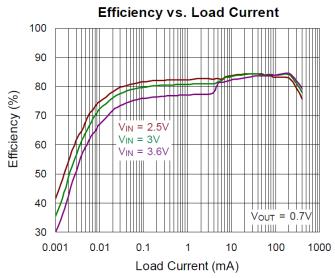
Table 2. Recommended Components Information

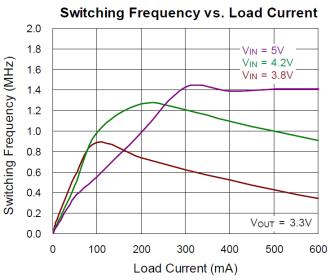
Reference Part Number		Description	Package	Manufacturer
Cin	GRM155R60J475ME47	4.7μF/6.3V/X5R	0402	Murata
Соит	GRM155R60J106ME15	10μF/6.3V/X5R	0402	Murata
L1	DFE201610E-2R2M=P2	2.2μΗ	2016	Murata

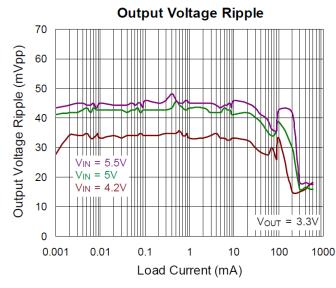


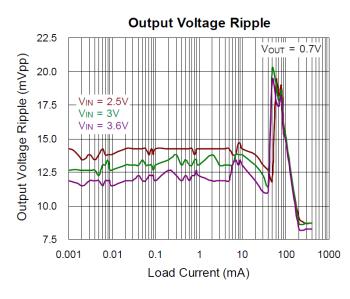
14 Typical Operating Characteristics

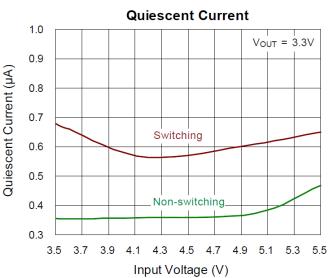










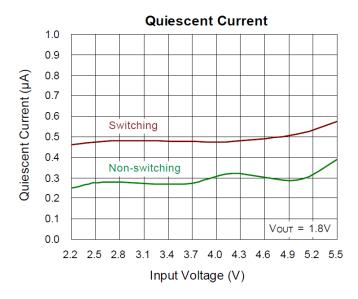


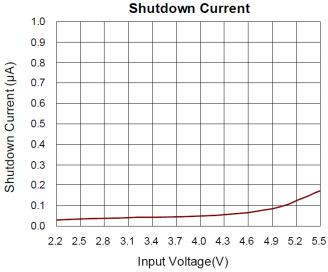
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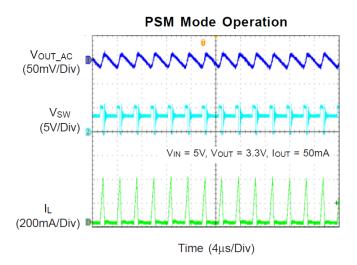
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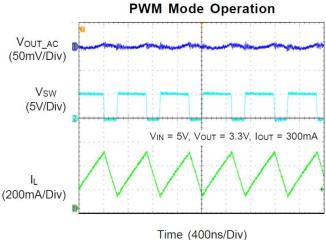
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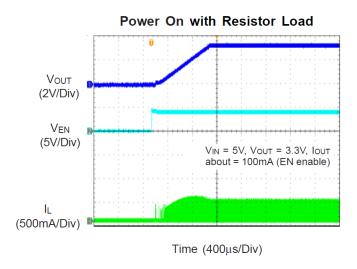


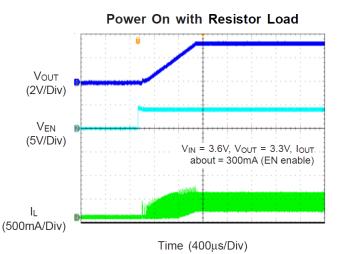




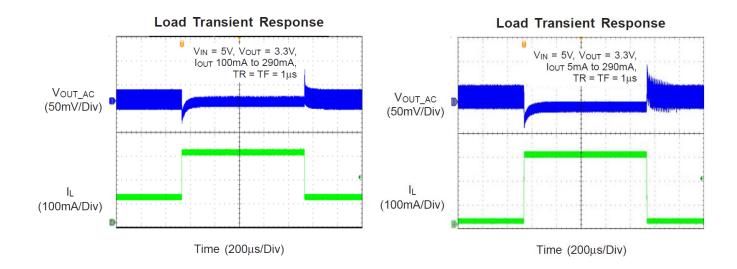


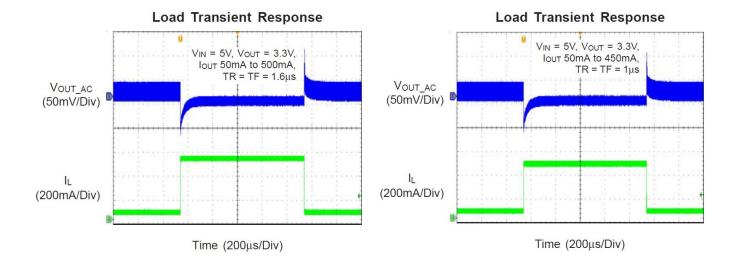


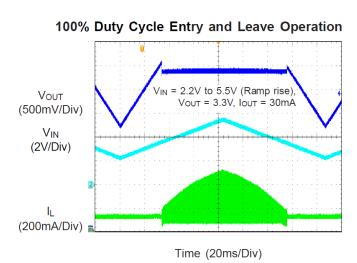












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15 Operation

The RT5707/A is a hysteretic constant on-time (HCOT) switching buck converter. It features Over-Temperature Protection (OTP) and Overcurrent Protection (OCP) mechanisms to prevent the device against damage from abnormal operating conditions. When the EN pin voltage is at a logic low level, the IC enters shutdown mode, drawing a low input supply current of less than $1\mu A$.

15.1 Enable

The device can be enabled or disabled via the EN pin. When the voltage on the EN pin exceeds the logic-high threshold, the IC enters normal operation. A transition of the EN pin from high to low causes the converter to enter shutdown mode, halting switching activity, powering down internal control circuitry, and initiating the discharge function.

That discharge function will deactivate after approximately 10ms.

If the system requires toggling of the EN pin, the duration of the EN pin being turned off must be greater than $100\mu s$ to allow sufficient time for the internal circuitry to reset.

15.2 UVLO Protection

To protect the IC from operating under insufficient supply voltage conditions, a UVLO feature is implemented. If the input voltage is lower than the UVLO threshold, the device enters a lockout state, preventing operation until the voltage is restored above the threshold.

15.3 100% Duty Cycle Operation

When the input voltage drops, making the difference between input and output less than V_{TH_100} , the converter operates at a 100% duty cycle. In this mode, the output voltage is the input voltage minus losses in the P-MOSFET and inductor. If the input voltage rises above V_{TH_100} , the converter switches back to its regular mode. Refer to Figure 1.

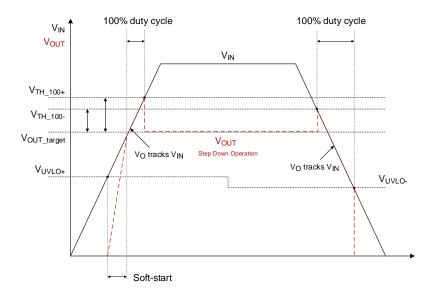


Figure 1. Automatic Transition into 100% Duty Cycle



15.4 Over-Temperature Protection

When the junction temperature exceeds the OTP threshold value, the IC will shut down the switching operation. Once the junction temperature cools down and falls below the OTP lower threshold, the converter will automatically resume switching.

15.5 Overcurrent Protection

The OCP functions are implemented by the high-side MOSFET and low-side MOSFET. When the inductor current reaches the UGATE current limit threshold, the high-side MOSFET is turned off. Subsequently, the low-side MOSFET is turned on to discharge the inductor current until it falls below the LGATE current limit threshold. Once the UGATE current limit is activated, the maximum inductor current is determined by the rate of increase of the inductor current and the response delay time of the internal circuitry.

During an OCP event, if the output voltage falls below the preset threshold (0.4V typical), the current limit value is reduced. This action decreases the device's power loss, mitigates heat generation, and prevents further damage to the device.

15.6 Output Voltage Selection

The RT5707/A provides eight levels of output voltages, which can be programmed via the voltage select pins VSEL1, VSEL2, and VSEL3. <u>Table 3</u> indicates the settings for each output voltage.

Device	Vout (V)	VSEL3	VSEL2	VSEL1
	1.2	0	0	0
	1.5	0	0	1
	1.8	0	1	0
RT5707	2.1 (<u>Note 6</u>)	0	1	1
K15707	2.5	1	0	0
	2.8	1	0	1
	3	1	1	0
	3.3 (<u>Note 6</u>)	1	1	1
	0.7	0	0	0
	1	0	0	1
	1.3	0	1	0
RT5707A	1.6 (<u>Note 6</u>)	0	1	1
K15/U/A	1.9	1	0	0
	2	1	0	1
	2.9	1	1	0
	3.1 (<u>Note 6</u>)	1	1	1

Table 3. Output Voltage Setting

Note 6. Connect a $100k\Omega$ resistor in series to the VSEL1 pin from the EN pin.

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16 Application Information

(Note 7)

The RT5707/A is a synchronous low voltage buck converter that supports an input voltage range of 2.2V to 5.5V. It can deliver an output current of up to 600mA with a peak capability of 1A for the RT5707, or 400mA with a peak of 0.5A for the RT5707A. Integrated internal compensation minimizes the need for external components. The device includes several protection features, such as overcurrent protection, undervoltage protection, and overtemperature protection.

16.1 Inductor Selection

The recommended power inductor is $2.2\mu H$. The inductor's saturation current rating should be selected based on overcurrent protection design considerations. For optimal performance and efficiency, it is crucial to choose an inductor with a low Direct Current Resistance (DCR).

16.2 CIN and COUT Selection

The input capacitance, CIN, is required to filter the trapezoidal current present at the source terminal of the top MOSFET. To mitigate large voltage ripples, it is advisable to use a low ESR input capacitor rated for the maximum RMS current

$$I_{RMS} = I_{OUT(MAX)} \times \frac{V_{OUT}}{V_{IN}} \times \sqrt{\frac{V_{IN}}{V_{OUT}} - 1}$$

This formula reaches its maximum when $V_{IN} = 2V_{OUT}$, at which $I_{RMS} = I_{OUT} / 2$. This worst-case scenario is commonly used for design considerations because significant deviations from this condition typically do not provide substantial improvements. It is advisable to select a capacitor with a higher temperature rating than necessary, and multiple capacitors can be connected in parallel to satisfy the size or height requirements of the design.

The choice of C_{OUT} is influenced by the Effective Series Resistance (ESR) needed to reduce voltage ripple and load step transients, as well as by the required bulk capacitance to maintain control loop stability. Loop stability can be assessed by analyzing the load transient response, which will be discussed in a subsequent section.

The output ripple, ΔV_{OUT} , is determined by:

$$\Delta V_{OUT} \leq \ \Delta I_L \Bigg[\text{ESR} + \frac{1}{8 \times f_{SW} \times C_{OUT}} \Bigg]$$

16.3 Thermal Considerations

The junction temperature should never exceed the absolute maximum junction temperature T_{J(MAX)}, listed under Absolute Maximum Ratings, to avoid permanent damage to the device. The maximum allowable power dissipation depends on the thermal resistance of the IC package, the PCB layout, the rate of surrounding airflow, and the difference between the junction and ambient temperatures. The maximum power dissipation can be calculated using the following formula:

$$PD(MAX) = (TJ(MAX) - TA) / \theta JA$$

where $T_{J(MAX)}$ is the maximum junction temperature, T_A is the ambient temperature, and θ_{JA} is the junction-to-ambient thermal resistance.

For continuous operation, the maximum operating junction temperature indicated under Recommended Operating Conditions is 125°C. The junction-to-ambient thermal resistance, θ_{JA} , is highly package dependent. For a WL-CSP-8B 0.9x1.6 (BSC) package, the thermal resistance, θ_{JA} , is 118.5°C/W on a standard JEDEC 51-7 high effective-thermal-conductivity four-layer test board. The maximum power dissipation at $T_A = 25$ °C can be



calculated as follows:

 $PD(MAX) = (125^{\circ}C - 25^{\circ}C) / (118.5^{\circ}C/W) = 0.84W$ for a WL-CSP-8B 0.9x1.6 (BSC) package.

The maximum power dissipation depends on the operating ambient temperature for the fixed $T_{J(MAX)}$ and the thermal resistance, θ_{JA} . The derating curve in Figure 2 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

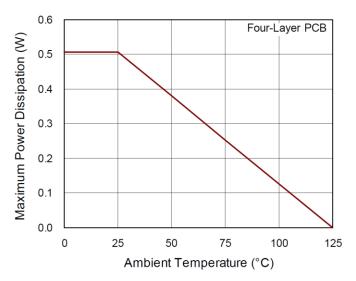


Figure 2. Derating Curve of Maximum Power Dissipation

16.4 Layout Considerations

For high-frequency switching power supplies, optimal PCB layout is crucial for achieving good regulation, high efficiency, and stability. The following descriptions are the guidelines for better PCB layout.

- To ensure good regulation, position the power components as close together as possible. Additionally, traces should be both wide and short, particularly for the high-current loop, to optimize the performance.
- Minimize the length of the SW node trace and increase its width for the optimal performance.

Protection Type		Threshold Refer to Protection Electrical Spec. Method		Reset Method
DT5707	UGATE Current Limit	Isw > 1.2A (Typical)	Turn off UG MOS	Isw < 1.2A (Typical)
RT5707	LGATE Current Limit	Isw > 1.2A (Typical)	Turn on LG MOS	Isw < 1.2A (Typical)
RT5707A	UGATE Current Limit	Isw > 0.78A (Typical)	Turn off UG MOS	Isw < 0.78A (Typical)
KISTOTA	LGATE Current Limit	Isw > 0.68A (Typical)	Turn on LG MOS	Isw < 0.68A (Typical)
	UVLO	VuvLof < 1.9V (Typical)	Shutdown	VuvLor > 2V (Typical)
ОТР		Temperature > 150°C (Typical)	Shutdown	Temperature < 130°C (Typical)

Table 4. Protection Trigger Condition and Behavior

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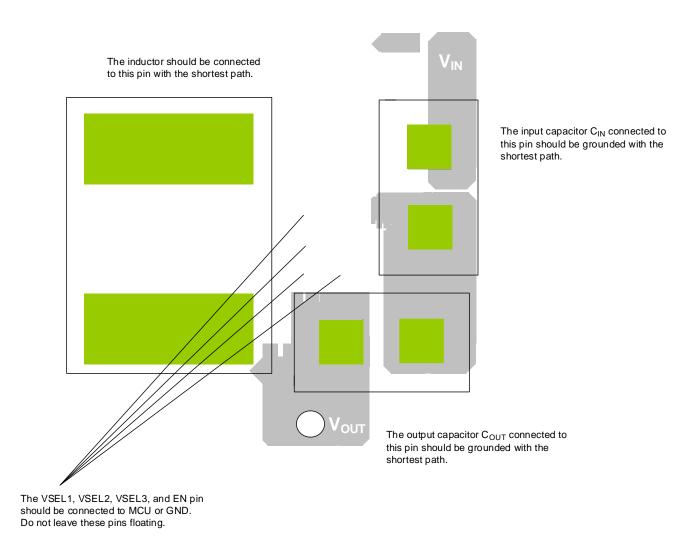


Figure 3. PCB Layout Guide for the RT5707

TOP View

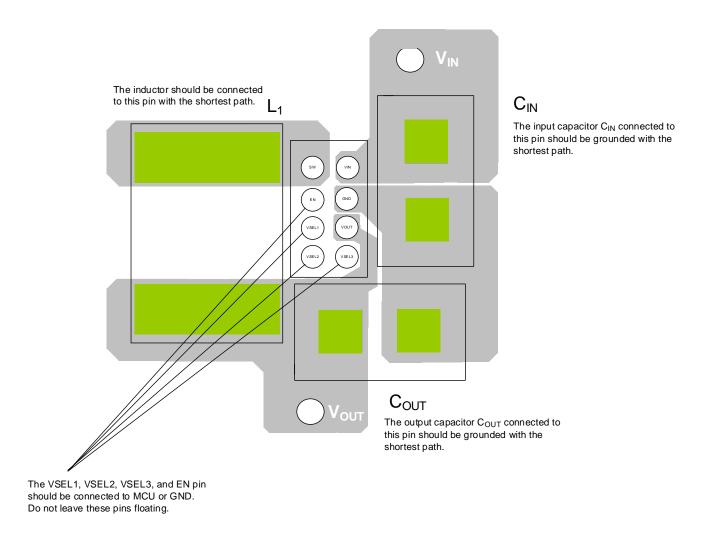


Figure 4. PCB Layout Guide for the RT5707A

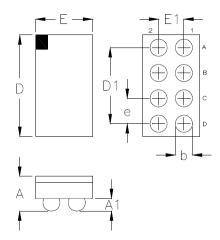
Note 7. The information provided in this section is for reference only. The customer is solely responsible for designing, validating, and testing any applications incorporating Richtek's product(s). The customer is also responsible for applicable standards and any safety, security, or other requirements.

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17 Outline Dimension

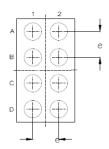


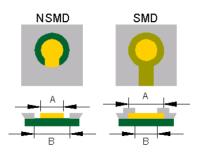
Sumb al	Dimensions I	In Millimeters	Dimension	s In Inches	
Symbol	Min	Max	Min	Max	
Α	0.500	0.600	0.020	0.024	
A1	0.170	0.230	0.007	0.009	
b	0.240	0.300	0.009	0.012	
D	1.560	1.640	0.061	0.065	
D1	1.2	200	0.047		
E	0.860	0.940	0.034	0.037	
E1	0.4	100	0.016		
е	0.4	100	0.016		

8B WL-CSP 0.9x1.6 Package (BSC)



18 Footprint Information



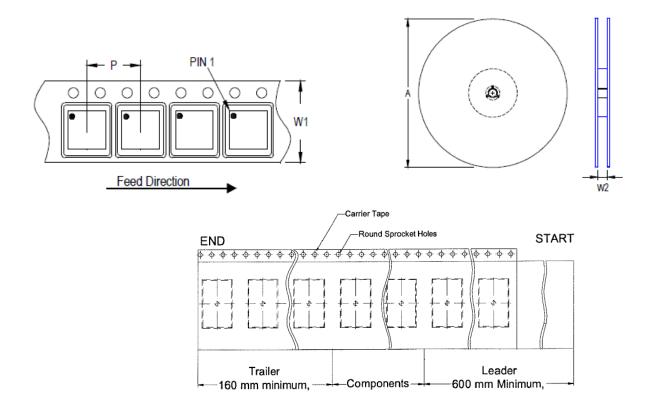


Dookogo	Number of Type		Footpri	Toloranco			
Package			е	Α	В	Tolerance	
WL-CSP0.9x1.6-8(BSC)	8	NSMD	0.400	0.240	0.340	.0.025	
WL-CSP0.9X1.0-0(DSC)	0	SMD	0.400	0.270	0.240	±0.025	

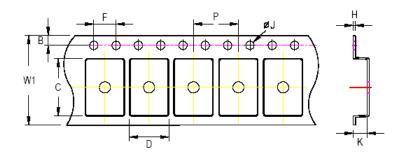


19 Packing Information

19.1 **Tape and Reel Data**



Package Type	Tape Size	Pocket Pitch	Reel Si	ze (A)	Units	Trailer	Leader	Reel Width (W2)
	(W1) (mm)	(P) (mm)	(mm)	(in)	per Reel	(mm)	(mm)	Min/Max (mm)
WL-CSP 0.9x1.6	8	4	180	7	3,000	160	600	8.4/9.9



C, D, and K are determined by component size. The clearance between the components and the cavity is as follows:

- For 8mm carrier tape: 0.5mm max.

Tape Size	W1	Р		В		F		Ø٦		Н
Tape Size	Max	Min	Max	Min	Max	Min	Max	Min	Max	Max
8mm	8.3mm	3.9mm	4.1mm	1.65mm	1.85mm	3.9mm	4.1mm	1.5mm	1.6mm	0.6mm



19.2 Tape and Reel Packing

Step	Photo/Description	Step	Photo/Description		
1	Reel 7"	4	12 inner boxes per outer box		
2	Packing by Anti-Static Bag	5	Outer box Carton A		
3	3 reels per inner box Box A	6	COLOT BOX CUITOTI A		

Container Reel				Вох		Carton			
Package	Size	Units	Item	Reels	Units	Item	Boxes	Unit	
WL-CSP	"	7"	2.000	Box A	3	9,000	Carton A	12	108,000
0.9x1.6	7	3,000	Box E	1	3,000	For Combined or Partial Reel.			

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RT5707/A



20.1 **Packing Material Anti-ESD Property**

Surface Resistance	Aluminum Bag	Reel	Cover tape	Carrier tape	Tube	Protection Band
Ω /cm ²	10 ⁴ to 10 ¹¹					

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21 Datasheet Revision History

Version	Date	Description	Item
05	2025/1/7	Modify	General Description on page 1 - Added description of temperature Ordering Information on page 1 - Added note Electrical Characteristics on page 6, 7 - Modified symbol Packing Information on page 20, 21, 22 -Added packing information

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