

Technical Specification

PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013
Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

Key Features

- Industry standard POLA™ compatible
- 38.61 x 25.91 x 9.64 mm (1.52 x 1.02 x 0.38 in)
- High efficiency, up to. 96%
- Auto Track™ sequencing pin
 Turbo Trans™ Technology for Ultra-Fast Transient
- Parallel Operation

General Characteristics

- Operating temperature: -40°C to 85°C
- Output over current protection (Non-latching, Auto-
- Output short-circuit protection
- Input under voltage protection
- Over temperature protection
- Wide output voltage adjust function
- ±1.5% total output voltage variation
- Remote sense
- On/Off inhibit control
- Start up into a pre-biased output
- Highly automated manufacturing ensures quality
- ISO 9001/14001 certified supplier



Safety Approvals



E210157

Design for Environment





Meets requirements in hightemperature lead-free soldering processes.

Contents

Ordering Information	2
General Information	2
Safety Specification	2
Absolute Maximum Ratings	4
Electrical Specification	
0.7V, 50A / 35W	PMR 5118UW P5
1.0V, 50A / 50W	PMR 5118UW P8
1.2V, 50A / 60W	PMR 5118UW P 11
1.5V, 50A / 75W	PMR 5118UW P 14
1.8V, 50A / 90W	PMR 5118UW P 17
2.5V, 50A / 125W	PMR 5118UW P20
3.3V, 50A / 165W	PMR 5118UW P23
EMC Specification	26
Operating Information	26
Thermal Consideration	34
Connections	34
Mechanical Information	37
Soldering Information	39
Delivery Information	39
Product Qualification Specification	41



PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013
Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

Ordering Information

Product program	Output
PMR 5118UW	0.7-3.6 V, 50 A / 180 W

Product number and Packaging

PMR 5118UW n ₁ n ₂				
Options	n ₁	n_2		
Mounting	0			
Delivery package information		0		

Options	Desc	ription
n_1	P S SR	Through hole Surface mount (SMD pin) SMD pin with lead-free surface
n_2	/B /C	Tray Tape and Reel

Example: a surface mounted, lead-free SMD pin surface, tray packaged product would be PMR 5118UW SR /B.

General Information Reliability

The Mean Time Between Failure (MTBF) is calculated at full output power and an operating ambient temperature (T_A) of +40°C, which is a typical condition in Information and Communication Technology (ICT) equipment. Different methods could be used to calculate the predicted MTBF and failure rate which may give different results. Ericsson Power Modules currently uses Telcordia SR332.

Predicted MTBF for the series is:

 1.5 million hours according to Telcordia SR332, issue 1, Black box technique.

Telcordia SR332 is a commonly used standard method intended for reliability calculations in ICT equipment. The parts count procedure used in this method was originally modelled on the methods from MIL-HDBK-217F, Reliability Predictions of Electronic Equipment. It assumes that no reliability data is available on the actual units and devices for which the predictions are to be made, i.e. all predictions are based on generic reliability parameters.

Compatibility with RoHS requirements

The products are compatible with the relevant clauses and requirements of the RoHS directive 2002/95/EC and have a maximum concentration value of 0.1% by weight in homogeneous materials for lead, mercury, hexavalent chromium, PBB and PBDE and of 0.01% by weight in homogeneous materials for cadmium.

Exemptions in the RoHS directive utilized in Ericsson Power Modules products include:

- Lead in high melting temperature type solder (used to

- solder the die in semiconductor packages)
- Lead in glass of electronics components and in electronic ceramic parts (e.g. fill material in chip resistors)
- Lead as an alloying element in copper alloy containing up to 4% lead by weight (used in connection pins made of Brass)

Quality Statement

The products are designed and manufactured in an industrial environment where quality systems and methods like ISO 9000, 6σ (sigma), and SPC are intensively in use to boost the continuous improvements strategy. Infant mortality or early failures in the products are screened out and they are subjected to an ATE-based final test. Conservative design rules, design reviews and product qualifications, plus the high competence of an engaged work force, contribute to the high quality of our products.

Warranty

Warranty period and conditions are defined in Ericsson Power Modules General Terms and Conditions of Sale.

Limitation of Liability

Ericsson Power Modules does not make any other warranties, expressed or implied including any warranty of merchantability or fitness for a particular purpose (including, but not limited to, use in life support applications, where malfunctions of product can cause injury to a person's health or life).

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Safety Specification

General information

Ericsson Power Modules DC/DC converters and DC/DC regulators are designed in accordance with safety standards IEC/EN/UL60950, Safety of Information Technology Equipment.

IEC/EN/UL60950 contains requirements to prevent injury or damage due to the following hazards:

- Electrical shock
- · Energy hazards
- Fire
- Mechanical and heat hazards
- · Radiation hazards
- Chemical hazards



EN/LZT 146 410 R1B March 2013

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On-board DC-DC converters and DC/DC regulators are defined as component power supplies. As components they cannot fully comply with the provisions of any Safety requirements without "Conditions of Acceptability". Clearance between conductors and between conductive parts of the component power supply and conductors on the board in the final product must meet the applicable Safety requirements. Certain conditions of acceptability apply for component power supplies with limited stand-off (see Mechanical Information for further information). It is the responsibility of the installer to ensure that the final product housing these components complies with the requirements of all applicable Safety standards and Directives for the final product.

Component power supplies for general use should comply with the requirements in IEC60950, EN60950 and UL60950 "Safety of information technology equipment".

There are other more product related standards, e.g. IEEE802.3af "Ethernet LAN/MAN Data terminal equipment power", and ETS300132-2 "Power supply interface at the input to telecommunications equipment; part 2: DC", but all of these standards are based on IEC/EN/UL60950 with regards to safety.

Ericsson Power Modules DC/DC converters and DC/DC regulators are UL60950 recognized and certified in accordance with EN60950.

The flammability rating for all construction parts of the products meets requirements for V-0 class material according to IEC 60695-11-10.

The products should be installed in the end-use equipment, in accordance with the requirements of the ultimate application. Normally the output of the DC/DC converter is considered as SELV (Safety Extra Low Voltage) and the input source must be isolated by minimum Double or Reinforced Insulation from the primary circuit (AC mains) in accordance with IEC/EN/UL60950.

Isolated DC/DC converters

It is recommended that a slow blow fuse with a rating twice the maximum input current per selected product be used at the input of each DC/DC converter. If an input filter is used in the circuit the fuse should be placed in front of the input filter.

In the rare event of a component problem in the input filter or in the DC/DC converter that imposes a short circuit on the input source, this fuse will provide the following functions:

- Isolate the faulty DC/DC converter from the input power source so as not to affect the operation of other parts of the system.
- Protect the distribution wiring from excessive current and power loss thus preventing hazardous overheating.

The galvanic isolation is verified in an electric strength test. The test voltage ($V_{\rm iso}$) between input and output is 1500 Vdc or 2250 Vdc for 60 seconds (refer to product specification).

Leakage current is less than 1 µA at nominal input voltage.

24 V DC systems

The input voltage to the DC/DC converter is SELV (Safety Extra Low Voltage) and the output remains SELV under normal and abnormal operating conditions.

48 and 60 V DC systems

If the input voltage to the DC/DC converter is 75 Vdc or less, then the output remains SELV (Safety Extra Low Voltage) under normal and abnormal operating conditions.

Single fault testing in the input power supply circuit should be performed with the DC/DC converter connected to demonstrate that the input voltage does not exceed 75 Vdc.

If the input power source circuit is a DC power system, the source may be treated as a TNV2 circuit and testing has demonstrated compliance with SELV limits and isolation requirements equivalent to Basic Insulation in accordance with IEC/EN/UL60950.

Non-isolated DC/DC regulators

The input voltage to the DC/DC regulator is SELV (Safety Extra Low Voltage) and the output remains SELV under normal and abnormal operating conditions.



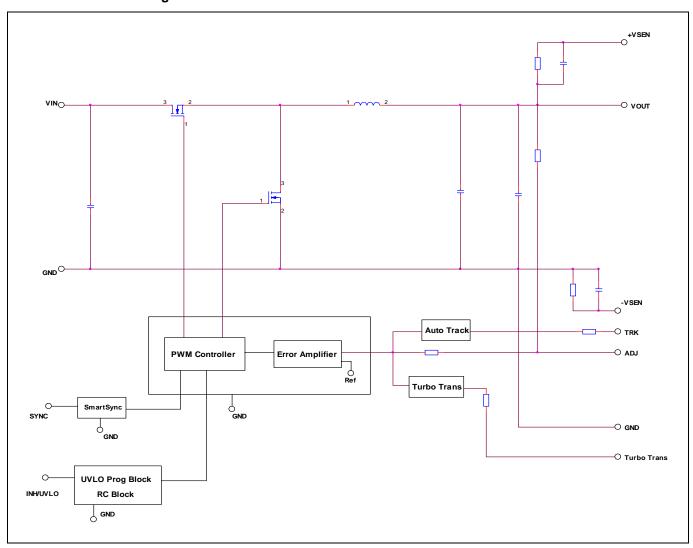
	ENW == 440 440 D4D 14 40040
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Absolute Maximum Ratings

Chara	Characteristics				max	Unit
T _{ref}	T _{ref} Operating Temperature (see Thermal Consideration section)				85	°C
Ts	Storage temperature				125	°C
VI	Input voltage			5/12	14	V
V_{RC}	Remote Control pin voltage	Positive logic option	-0.2		Open	V
V RC	(see Operating Information section)	Negative logic option	N/A		N/A	V
V_{adj}	V _{adj} Adjust pin voltage (see Operating Information section)				N/A	V

Stress in excess of Absolute Maximum Ratings may cause permanent damage. Absolute Maximum Ratings, sometimes referred to as no destruction limits, are normally tested with one parameter at a time exceeding the limits in the Electrical Specification. If exposed to stress above these limits, function and performance may degrade in an unspecified manner.

Fundamental Circuit Diagram





PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013	
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0.7V, 50A / 35.0W Electrical Specification

PMR 5118UW

 T_{P1} = -40 to +85°C, V_{I} = 4.5 to 14 V, R_{SET} = OPEN, unless otherwise specified under Conditions. Typical values given at: T_{P1} = +25°C, V_{I} = 5/12 V, max I_{O} , unless otherwise specified under Conditions. Additional C_{in} = 1000+22 μF and C_{out} = 660 μF. See Operating Information section for selection of capacitor types. Connect the sense pin, where available, to the output pin.

Charac	teristics		Conditions	min	typ	max	Unit
Vı	Input voltage rar	nge		4.5	5/12	14	V
V _{loff}	Turn-off input vo	ltage	Decreasing input voltage	4.0	4.2		V
V _{Ion}	Turn-on input voltage		Increasing input voltage		4.3	4.45	V
Cı	Internal input ca	pacitance			44		μF
Po	Output power			0		35	W
		.,,	$V_1 = 5 \text{ V}, 50 \% \text{ of max } I_0$		86.8		
		$V_1 = 5 V$	$V_I = 5 \text{ V, max } I_O$		80.7		- ·
η	Efficiency		V _I = 12 V, 50 % of max I _O		83.4		- %
		V _I = 12 V	$V_I = 12 \text{ V, max } I_O$		79.4		1
_		l	$V_I = 5 \text{ V}, \text{ max } I_O$		8.34	8.50	W
P_d	Power Dissipation		$V_I = 12 \text{ V, max } I_O$		9.05	9.50	W
_			$V_1 = 5 \text{ V}, I_0 = 0 \text{ A}$		0.45		W
P_{li}	Input idling power	er	V _I = 12 V, I _O = 0 A		1.05		W
_			V _I = 5 V (turned off with RC)		46.0		mW
P_{RC}	Input standby po	ower	V _I = 12 V (turned off with RC)		102		mW
_			$V_1 = 5 \text{ V}, \text{ max } I_0$		8.49		Α
Is	Static Input curre	ent	$V_I = 12 \text{ V, max } I_O$		4.86		Α
f _s	Switching freque	ency	0-100 % of max I _O , see Note 1		600		kHz
V _{Oi}	Output voltage in accuracy	<u> </u>	$T_{P1} = +25^{\circ}C, V_{I} = 5/12 \text{ V, max } I_{O}$	0.693	0.700	0.707	V
	Output voltage to	olerance band	10-100 % of max I _O	0.689		0.711	V
	Idling voltage Line regulation Load regulation		$V_1 = 5 \text{ V}, I_0 = 0 \text{ A}$		0.701		V
Vo			V _I = 12 V, I _O = 0 A	0.701			
			max I _O		±5		mV
			$V_1 = 5/12 \text{ V}, 0-100 \text{ % of max } I_0$		±5		mV
V_{tr}	Load transient voltage deviation		V_1 = 12 V, Load step 25-75-25 % of max I_0 , di/dt = 2.5 A/ μ s		±160		mV
t _{tr}	Load transient re	ecovery time	Without Turbo Trans C _o =660 μF Type C, see Note 2		100		μS
V_{tr}	Load transient voltage deviation		V_1 = 12 V, Load step 25-75-25 % of max I ₀ , di/dt = 2.5 A/µs		±45		mV
t _{tr}	Load transient re	ecovery time	With Turbo Trans C_o =3300 μ F Type C; R_{TT} =SHORT, see Note 2		100		μS
t _r	Ramp-up time (from 10-90 % of \	V _{Oi})	V _I = 5 V, 100 % of max I _O		7.5		ms
ts	Start-up time (from V _I connection	n to 90 % of V _{Oi})	7, 0 7, 100 % oax 10		22.2		ms
t _r	Ramp-up time (from 10-90 % of	V _{Oi})	V _I = 12 V, 100 % of max I _O	6.9			ms
ts	Start-up time (from V _I connection	on to 90 % of V _{Oi})	. ,		13.7		ms
	V _I shut-down	V _I = 5 V	Max I ₀		0.07		ms
t_{f}	fall time.		I _O = 0.1 A		17.1		ms
٦	(From V _I off to 10 % of V _O)		Max I _O		1.1		ms
	70 70 01 40)		I _O = 0.1 A		96.4		ms
$t_{\text{RC}}\;t_{\text{Inh}}$	RC start-up time		$V_i = 5 \text{ V}$, Max I_O		21.9		ms
			V _I = 12 V , Max I _O		12.1		ms
	RC shut-down $V_1 = 5 \text{ V}$		Max I _O		0.037		ms





Technical Specification

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	fall time		I _o = 0.1 A		15.4		ms
	(From RC off to 10 % of V _O)	V _I = 12 V	Max I _O		0.121		ms
	10 10 70 01 10,	V = 12 V	I _o = 0.1 A		16.2		ms
Io	Output current			0		50	Α
I _{lim}	Current limit thre	eshold	$T_{P1} < max T_{P1}$		100		Α
I _{sc}	Short circuit curr	ent	$T_{P1} = 25^{\circ}C$, see Note 3		110		Α
C _{out}	Recommended (Capacitive Load	T _{P1} = 25°C, see Note 4	1000		10000	μF
V_{Oac}	Output ripple & r	noise V _I = 5 V	See ripple & noise section, max I _O	5.6		mVp-p	
V _{Oac}	Output ripple & r	noise V _I = 12 V	See ripple & noise section, max I _O	36		mVp-p	

Note 1: Frequency may be adjusted with SmartSync pin. See Operating Information section

Note 4: 1000 µF of external non-ceramic output capacitance is required for basic operation. Adding additional capacitance at the load further improves transient response. Up to 1000 µF of ceramic capacitance may be added in addition to the required non-ceramic capacitance. When not using TurboTrans technology, 8000 µF capacitance is allowed; When using TurboTrans technology, up to 10000 µF capacitance is allowed. For more information, see Operating Information Section.

Note 2: See Operating Information section for TurboTrans technology

Note 3: Describe short circuit current characteristic, i.e. fold-back, hiccup or RMS, in one short sentence or type only "See Operating Information section.")



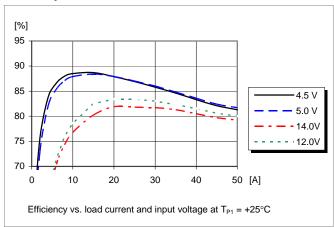
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Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

0.7V, 50A / 35.0W Typical Characteristics

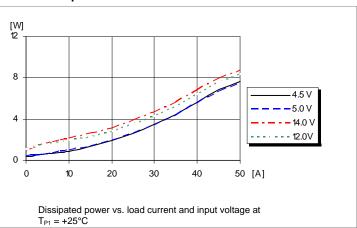
PMR 5118UW

7

Efficiency



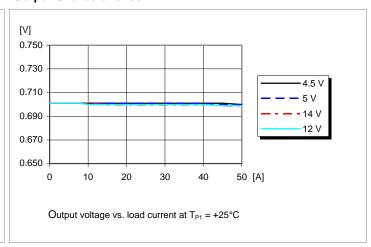
Power Dissipation



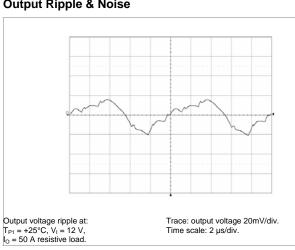
Output Current Derating

TBD Output current derating information is available only for 1.2V and 3.3V output.

Output Characteristics



Output Ripple & Noise



Output Voltage Adjust (see operating information)

Passive adjust

The resistor value for an adjusted output voltage is calculated by using the equations in the operating information.

$$R_{\mbox{SET}} = 30.1 \, k\Omega \times \frac{0.7}{\mbox{$V_0 - 0.7$}} - 6.49 \, k\Omega$$



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1.0V, 50A / 50.0W Electrical Specification

PMR 5118UW

 T_{P1} = -40 to +85°C, V_I = 4.5 to 14 V, R_{SET} = 63.4 kΩ, unless otherwise specified under Conditions. Typical values given at: T_{P1} = +25°C, V_I = 5/12 V, max I_O , unless otherwise specified under Conditions. Additional C_{in} = 1000+22 μF and C_{out} = 660 μF. See Operating Information section for selection of capacitor types. Connect the sense pin, where available, to the output pin.

Charac	teristics	<u> </u>	Conditions	min	typ	max	Unit
Vı	Input voltage rar	nge		4.5	5/12	14	V
V _{loff}	Turn-off input vo		Decreasing input voltage	4.0	4.2		V
V _{Ion}	Turn-on input vo		Increasing input voltage	4.0	4.3	4.45	V
C _I	Internal input capacitance		moreasing input voltage		44	7.70	μF
Po	Output power	pacitariec		0		50	W
10	Output power		$V_1 = 5 \text{ V}, 50 \% \text{ of max } I_0$		90.1		1 **
		V _I = 5 V	$V_1 = 5 \text{ V}, \text{ max } I_0$		85.4		+
η	Efficiency		$V_1 = 3 \text{ V}, \text{ flax } I_0$ $V_1 = 12 \text{ V}, 50 \text{ % of max } I_0$		86.5		%
		V _I = 12 V	$V_1 = 12 \text{ V}, \text{ 30 \% of max } I_0$ $V_1 = 12 \text{ V}, \text{ max } I_0$		83.5		-
			$V_1 = 12 \text{ V}, \text{ max } I_0$ $V_1 = 5 \text{ V}, \text{ max } I_0$			8.7	W
P_{d}	Power Dissipation	on	· · ·		8.4		
			$V_1 = 12 \text{ V}, \text{ max } I_0$		9.85	10.3	W
P_{li}	Input idling power	er	$V_1 = 5 \text{ V}, I_0 = 0 \text{ A}$		0.45		W
			$V_1 = 12 \text{ V}, I_0 = 0 \text{ A}$		1.19		W
P_{RC}	Input standby po	ower	V _I = 5 V (turned off with RC)		45.7		mW
	, , , ,		V _I = 12 V (turned off with RC)		102		mW
Is	Static Input curre	ent	$V_I = 5 \text{ V}, \text{ max } I_O$		11.68		A
	,		$V_I = 12 \text{ V, max } I_O$		4.98		Α
fs	Switching freque	ency	0-100 % of max I ₀		600		kHz
V_{Oi}	Output voltage in accuracy	nitial setting and	$T_{P1} = +25^{\circ}C$, $V_{I} = 5/12 \text{ V}$, max I_{O}	0.990	1.000	1.010	V
	Output voltage to	olerance band	10-100 % of max I _O	0.985		1.015	V
	Idling voltage		$V_1 = 5 \text{ V}, I_0 = 0 \text{ A}$		1.003		V
V_{O}			V _I = 12 V, I _O = 0 A		1.003		
	Line regulation		max I _O		±5		mV
	Load regulation		$V_I = 5/12 \text{ V}, 0-100 \text{ % of max } I_O$		±5		mV
V_{tr}	Load transient voltage deviation Load transient recovery time		V _I = 12 V, Load step 25-75-25 % of max I _O , di/dt = 2.5 A/μs		±160		mV
t _{tr}			Without Turbo Trans C _o =660 μF Type C		100		μS
V _{tr}	Load transient voltage deviation		V ₁ = 12 V, Load step 25-75-25 % of max I _o , di/dt = 2.5 A/μs		±45		mV
t _{tr}	Load transient re		With Turbo Trans C _o =3300 μF Type C; R _{TT} =SHORT		100		μS
t _r	Ramp-up time (from 10-90 % of	V _{Oi})	V _I = 5 V, 100 % of max I _O		8.4		ms
ts	Start-up time (from V _I connection	n to 90 % of V _{Oi})	1, 0., 100 / 0		22.1		ms
t _r	Ramp-up time (from 10-90 % of V _{Oi})		V _I = 12 V, 100 % of max I _o		7.5		ms
ts	Start-up time (from V _I connection	on to 90 % of V _{Oi})			14.0		ms
	V _I shut-down	V _I = 5 V	Max I ₀		0.071		ms
t _f	fall time.		I _O = 0.1 A		23.9		ms
	(From V _I off to 10 % of V _O)	V _I = 12 V	Max I _o		1.32		ms
	10 70 01 00/		I _O = 0.1 A		96.6		ms
$t_{RC} \; t_{Inh}$	RC start-up time	:	$V_I = 5 \text{ V}$, Max I_O		21.6		ms
			$V_I = 12 \text{ V}$, Max I_O		12.4		ms
	RC shut-down fall time	V _I = 5 V	Max I _O		0.039		ms
	ian anio		$I_o = 0.1 \text{ A}$		26.5		ms



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	(From RC off to 10 % of V _o) V _I = 12 V	Max I _O	0.120		ms
	to 10 % of V_0) $V_1 = 12 V$	I _o = 0.1 A	23.1		ms
Io	Output current		0	50	Α
I _{lim}	Current limit threshold	$T_{P1} < max T_{P1}$	100		Α
I _{sc}	Short circuit current	$T_{P1} = 25^{\circ}C$, see Note 3	110		Α
C _{out}	Recommended Capacitive Load	T _{P1} = 25°C, see Note 4	1000	10000	μF
V _{Oac}	Output ripple & noise V _I = 5 V	See ripple & noise section, max I _O	10		mVp-p
V _{Oac}	Output ripple & noise V _I = 12 V	See ripple & noise section, max I _O	10		mVp-p

Note 1: Frequency may be adjusted with SmartSync pin. See Operating Information section

Note 4: 1000 µF of external non-ceramic output capacitance is required for basic operation. Adding additional capacitance at the load further improves transient response. Up to 1000 µF of ceramic capacitance may be added in addition to the required non-ceramic capacitance. When not using TurboTrans technology, 8000 µF capacitance is allowed; When using TurboTrans technology, up to 10000 µF capacitance is allowed. For more information, see Operating Information Section.

9

Note 2: See Operating Information section for TurboTrans technology

Note 3: Describe short circuit current characteristic, i.e. fold-back, hiccup or RMS, in one short sentence or type only "See Operating Information section.")

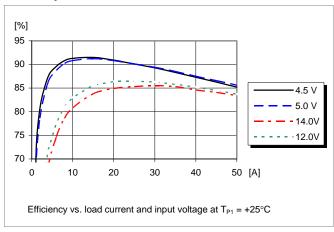


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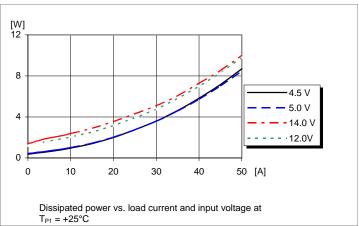
1.0V, 50A / 50.0W Typical Characteristics

PMR 5118UW

Efficiency



Power Dissipation

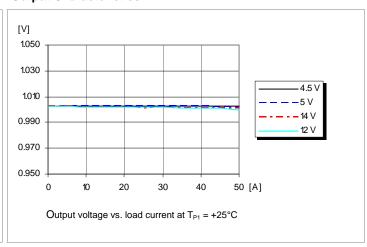


Output Current Derating

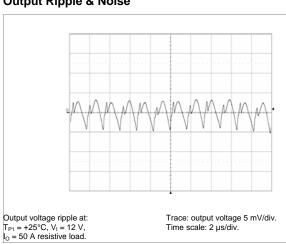


Output current derating information is available only for 1.2V and 3.3V output.

Output Characteristics



Output Ripple & Noise



Time scale: 2 µs/div.

 $R_{SET} = 30.1 \, k\Omega \times \frac{0.7}{V_0 - 0.7} - 6.49 \, k\Omega$

Passive adjust

The resistor value for an adjusted output voltage is calculated by using the equations in the operating information.

Output Voltage Adjust (see operating information)



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1.2V, 50A / 60.0W Electrical Specification

PMR 5118UW

 T_{P1} = -40 to +85°C, V_I = 4.5 to 14 V, R_{SET} = 35.7 kΩ, unless otherwise specified under Conditions. Typical values given at: T_{P1} = +25°C, V_I = 5/12 V, max I_O , unless otherwise specified under Conditions. Additional C_{in} = 1000+22 μF and C_{out} = 660 μF. See Operating Information section for selection of capacitor types. Connect the sense pin, where available, to the output pin.

Charac	teristics	<u> </u>	Conditions	min	Тур	max	Unit
Vı	Input voltage rar	nge		4.5	5/12	14	V
V _{loff}	Turn-off input vo		Decreasing input voltage	4.0	4.2	14	V
V _{Ion}	Turn-on input vo		Increasing input voltage	4.0	4.3	4.45	V
C _I	· · · · · · · · · · · · · · · · · · ·		increasing input voltage		4.3	4.45	μF
P _o	Internal input capacitance			0	44	50	W
10	Output power	<u> </u>	$V_1 = 5 \text{ V}, 50 \% \text{ of max } I_0$		91.5	30	VV
		V _I = 5 V	$V_1 = 5 \text{ V}, \text{ max } I_0$		87.5		4
η	Efficiency		$V_1 = 5 \text{ V}, \text{ flax } I_0$ $V_1 = 12 \text{ V}, 50 \text{ % of max } I_0$		88.2		%
		V _I = 12 V					4
			$V_1 = 12 \text{ V}, \text{ max } I_0$		85.7		14/
P_{d}	Power Dissipation	on	$V_1 = 5 \text{ V}, \text{ max } I_0$		8.6	9.0	W
			$V_1 = 12 \text{ V}, \text{ max } I_0$		9.96	10.3	W
Pli	Input idling power	er	V _I = 5 V, I _O = 0 A		0.47		W
			$V_1 = 12 \text{ V}, I_0 = 0 \text{ A}$		1.21		W
P_{RC}	Input standby po	wer	V _I = 5 V (turned off with RC)		42.4		mW
	, , , ,		V _I = 12 V (turned off with RC)		101.8		mW
Is	Static Input curre	ent	$V_I = 5 \text{ V}, \text{ max } I_O$		13.71		Α
<u> </u>	,		$V_I = 12 \text{ V, max } I_O$		5.83		Α
fs	Switching freque	ency	0-100 % of max I ₀		600		kHz
V_{Oi}	Output voltage in accuracy	nitial setting and	$T_{P1} = +25^{\circ}C, V_{I} = 5/12 \text{ V, max I}_{O}$	1.188	1.200	1.212	V
	Output voltage to	olerance band	10-100 % of max I _O	1.182		1.218	V
	Idling voltage		$V_1 = 5 \text{ V}, I_0 = 0 \text{ A}$		1.201		V
V_{o}			V _I = 12 V, I _O = 0 A		1.201		
	Line regulation		max I _O		±5		mV
	Load regulation		$V_I = 5/12 \text{ V}, 0-100 \text{ % of max } I_O$		±5		mV
V_{tr}	Load transient voltage deviation Load transient recovery time		V _I = 12 V, Load step 25-75-25 % of max I _O , di/dt = 2.5 A/μs		±160		mV
t _{tr}			Without Turbo Trans C _o =660 μF Type C		100		μS
V _{tr}	Load transient voltage deviation		V ₁ = 12 V, Load step 25-75-25 % of max I _o , di/dt = 2.5 A/μs		±45		mV
t _{tr}	Load transient re		With Turbo Trans C _o =3300 μF Type C; R _{TT} =SHORT		100		μs
t _r	Ramp-up time (from 10-90 % of	/ _{Oi})	V _I = 5 V, 100 % of max I _O		8.6		ms
ts	Start-up time (from V _I connection	n to 90 % of V _{Oi})	1, 0., 100 / 0		21.3		ms
t _r	Ramp-up time (from 10-90 % of V _{Oi})		V _I = 12 V, 100 % of max I _o		8.2		ms
ts	Start-up time (from V _I connection	n to 90 % of V _{Oi})			13.9		ms
	V _I shut-down	V _I = 5 V	Max I _O		0.077		ms
t _f	fall time. (From V _I off to		I _O = 0.1 A		29.5		ms
	10 % of V _O)	V _I = 12 V	Max I _o		0.91		ms
	,		I _O = 0.1 A		99.4		ms
$t_{RC} t_{Inh}$	RC start-up time		$V_1 = 5 \text{ V}$, Max I_0		21.7		ms
	·	T	$V_I = 12 \text{ V}$, Max I_O		12.4		ms
	RC shut-down fall time	V _I = 5 V	Max I ₀		0.044		ms
	ian anno		$I_0 = 0.1 \text{ A}$		28.4		ms



Technical Specification

	ENW == 440 440 D4D 14 40040
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Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

	(From RC off to 10 % of V _o) V _I = 12 V	Max I _O	0.125		ms
	to 10 % of V_0) $V_1 = 12 V$	I _o = 0.1 A	28.1		ms
Io	Output current		0	50	Α
I _{lim}	Current limit threshold	$T_{P1} < max T_{P1}$	100		Α
I _{sc}	Short circuit current	$T_{P1} = 25^{\circ}C$, see Note 3	110		Α
C _{out}	Recommended Capacitive Load	T _{P1} = 25°C, see Note 4	1000	10000	μF
V _{Oac}	Output ripple & noise V _I = 5 V	See ripple & noise section, max I _O	10		mVp-p
V _{Oac}	Output ripple & noise V _I = 12 V	See ripple & noise section, max I ₀	10		mVp-p

Note 1: Frequency may be adjusted with SmartSync pin. See Operating Information section

Note 2: See Operating Information section for TurboTrans technology

Note 3: Describe short circuit current characteristic, i.e. fold-back, hiccup or RMS, in one short sentence or type only "See Operating Information section.")

Note 4: 1000 µF of external non-ceramic output capacitance is required for basic operation. Adding additional capacitance at the load further improves transient response. Up to 1000 µF of ceramic capacitance may be added in addition to the required non-ceramic capacitance. When not using TurboTrans technology, 8000 µF capacitance is allowed; When using TurboTrans technology, up to 10000 µF capacitance is allowed. For more information, see Operating Information Section.

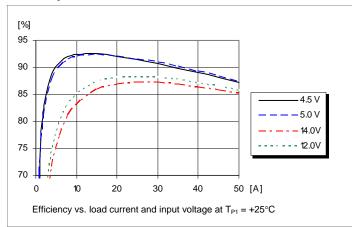


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PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013
Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

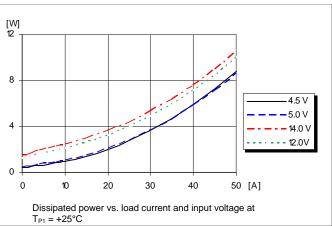
1.2V, 50A / 60.0W Typical Characteristics

PMR 5118UW

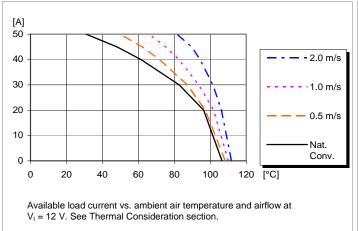
Efficiency



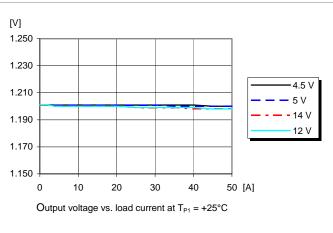
Power Dissipation



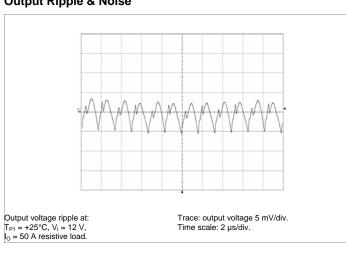
Output Current Derating



Output Characteristics



Output Ripple & Noise



Output Voltage Adjust (see operating information)

Passive adjust

The resistor value for an adjusted output voltage is calculated by using the equations in the operating information.

$$R_{SET} = 30.1 \,k\Omega \times \frac{0.7}{V_0 - 0.7} - 6.49 \,k\Omega$$



PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013
Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

1.5V, 50A / 75.0W Electrical Specification

PMR 5118UW

 T_{P1} = -40 to +85°C, V_I = 4.5 to 14 V, R_{SET} = 19.6 kΩ, unless otherwise specified under Conditions. Typical values given at: T_{P1} = +25°C, V_I = 5/12 V, max I_O , unless otherwise specified under Conditions. Additional C_{in} = 1000+22 μF and C_{out} = 660 μF. See Operating Information section for selection of capacitor types. Connect the sense pin, where available, to the output pin.

Charac	naracteristics		Conditions	min	Тур	max	Unit
V_{I}	Input voltage range			4.5	5/12	14	V
V_{loff}	Turn-off input vo	Itage	Decreasing input voltage	4.0	4.2		V
V_{lon}	Turn-on input voltage		Increasing input voltage		4.3	4.45	V
Cı	Internal input car	pacitance			44		μF
Po	Output power			0		50	W
		V _I = 5 V	$V_1 = 5 \text{ V}, 50 \text{ % of max } I_0$		92.8		
n	Efficiency	V ₁ = 5 V	V _I = 5 V, max I _O		89.5		%
η	Efficiency	V ₁ = 12 V	V _I = 12 V, 50 % of max I _O		90.0		70
		V ₁ = 12 V	V _I = 12 V, max I _O		87.9		
P _d	Power Dissipation	20	$V_I = 5 \text{ V, max } I_O$		8.81	8.95	W
Γd	Fower Dissipation	л	V _I = 12 V, max I _O		10.27	10.55	W
Pli	Input idling powe)r	V _I = 5 V, I _O = 0 A		0.50		W
Γli	Imput failing powe	51	V _I = 12 V, I _O = 0 A		1.26		W
P _{RC}	Input standby po	NA/OF	V _I = 5 V (turned off with RC)		42.4		mW
FRC	Imput standby po	wei	V _I = 12 V (turned off with RC)		101.8		mW
I.	Static Input curre	ont	V _I = 5 V, max I _O		16.76		А
Is	Static input curre	5111	V _I = 12 V, max I _O		7.11		Α
f_s	Switching freque	ency	0-100 % of max I _O		600		kHz
V _{Oi}	Output voltage in accuracy	nitial setting and	$T_{P1} = +25^{\circ}C, V_{I} = 5/12 \text{ V, max } I_{O}$	1.485	1.500	1.515	V
	Output voltage tolerance band		10-100 % of max I _o	1.477		1.523	V
	Idling voltage		V _I = 5 V, I _O = 0 A		1.498		V
V_{O}			V _I = 12 V, I _O = 0 A		1.498		
	Line regulation		max I _O		±5		mV
	Load regulation		$V_1 = 5/12 \text{ V}, 0-100 \text{ % of max } I_0$		±5		mV
V_{tr}	Load transient voltage deviation	1	V_1 = 12 V, Load step 25-75-25 % of max I_0 , di/dt = 2.5 A/ μ s		±160		mV
\mathbf{t}_{tr}	Load transient re	ecovery time	Without Turbo Trans C₀=660 μF Type C		100		μS
V _{tr}	Load transient voltage deviation	1	V _I = 12 V, Load step 25-75-25 % of max I _O , di/dt = 2.5 A/μs		±45		mV
t _{tr}	Load transient re	ecovery time	With Turbo Trans C₀=3300 µF Type C; R _{TT} =SHORT		100		μS
t _r	Ramp-up time (from 10-90 % of \	/ _{Oi})	V _I = 5 V, 100 % of max I _O		8.5		ms
ts	Start-up time (from V _I connection	n to 90 % of V _{Oi})	7, 5 7, 100 % 5		21.0		ms
t _r	Ramp-up time (from 10-90 % of	V _{Oi})	V _I = 12 V, 100 % of max I _O		8.2		ms
t _s	Start-up time (from V _I connection	on to 90 % of Voi)			13.3		ms
	V _I shut-down	V ₁ = 5 V	Max I _O		0.090		ms
t _f	fall time.		I _O = 0.1 A		37.2		ms
-1	(From V _I off to 10 % of V _O)	V _I = 12 V	Max I _O		0.892		ms
	10 70 01 00)		I _O = 0.1 A		94.3		ms
$t_{\text{RC}}\;t_{\text{Inh}}$	RC start-up time		$V_I = 5 \text{ V}$, Max I_O		21.8		ms
	No start-up time		$V_I = 12 \text{ V}$, Max I_O		12.2		ms
	RC shut-down $V_1 = 5 \text{ V}$		Max I _O		0.063		ms







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Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

	fall time (From RC off to 10 % of V _O)		I _o = 0.1 A		35.0		ms	
			V _I = 12 V	Max I _O		0.142		ms
			I _o = 0.1 A		34.7		ms	
Io	Output current			0		50	А	
I _{lim}	Current limit thre	eshold	$T_{P1} < max T_{P1}$		100		Α	
I _{sc}	Short circuit curr	rent	$T_{P1} = 25^{\circ}C$		110		А	
C _{out}	Recommended	Capacitive Load	T _{P1} = 25°C, see Note 4	1000		10000	μF	
V _{Oac}	Output ripple & noise V _I = 5 V		See ripple & noise section, max I _O		ì		mVp-p	
V _{Oac}	Output ripple & r	noise V _I = 12 V	See ripple & noise section, max I _O		10		mVp-p	

Note 1: Frequency may be adjusted with SmartSync pin. See Operating Information section

Note 2: See Operating Information section for TurboTrans technology

Note 3: Describe short circuit current characteristic, i.e. fold-back, hiccup or RMS, in one short sentence or type only "See Operating Information section.")

Note 4: 1000 µF of external non-ceramic output capacitance is required for basic operation. Adding additional capacitance at the load further improves transient response. Up to 1000 µF of ceramic capacitance may be added in addition to the required non-ceramic capacitance. When not using TurboTrans technology, 8000 µF capacitance is allowed; When using TurboTrans technology, up to 10000 µF capacitance is allowed. For more information, see Operating Information Section.

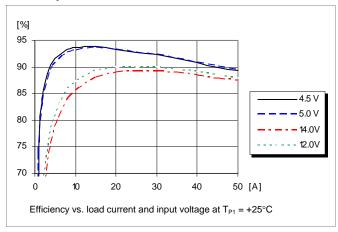


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PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013
Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

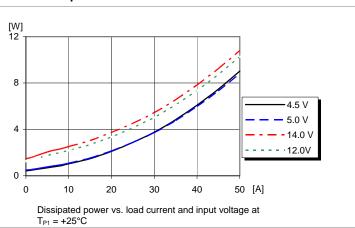
1.5V, 50A / 75.0W Typical Characteristics

PMR 5118UW

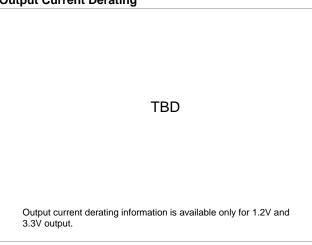
Efficiency



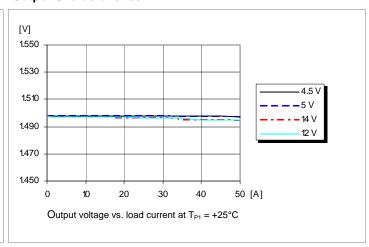
Power Dissipation



Output Current Derating



Output Characteristics



Output Ripple & Noise

Output voltage ripple at: Trace: output voltage 5 mV/div. Time scale: 2 μ s/div. μ 0 = 50 A resistive load.

Output Voltage Adjust (see operating information)

Passive adjust

The resistor value for an adjusted output voltage is calculated by using the equations in the operating information.

$$R_{SET} = 30.1 \, k\Omega \times \frac{0.7}{V_0 - 0.7} - 6.49 \, k\Omega$$



PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013	
Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB	

1.8V, 50A / 90.0W Electrical Specification

PMR 5118UW

 T_{P1} = -40 to +85°C, V_I = 4.5 to 14 V, R_{SET} = 12.7 kΩ, unless otherwise specified under Conditions. Typical values given at: T_{P1} = +25°C, V_I = 5/12 V, max I_O , unless otherwise specified under Conditions. Additional C_{in} = 1000+22 μF and C_{out} = 660 μF. See Operating Information section for selection of capacitor types. Connect the sense pin, where available, to the output pin.

	cteristics	nore available, te	Conditions	min	Тур	max	Unit
Vı	Input voltage rar	nge		4.5	5/12	14	V
V _{loff}	Turn-off input vo		Decreasing input voltage	4.0	4.2		V
V _{Ion}	Turn-on input vo		Increasing input voltage	4.0	4.3	4.45	V
C _I	Internal input car		moreasing input voltage		44	7.70	μF
Po	Output power	pacitariec		0		50	W
10	Output power	<u> </u>	$V_1 = 5 \text{ V}, 50 \% \text{ of max } I_0$		93.8		**
		V _I = 5 V	$V_1 = 5 \text{ V}, \text{ max } I_0$		90.9		+
η	Efficiency		$V_1 = 3 \text{ V}, \text{ flax } I_0$ $V_1 = 12 \text{ V}, 50 \text{ % of max } I_0$		91.4		%
		V _I = 12 V					-
	1		$V_I = 12 \text{ V, max } I_O$ $V_I = 5 \text{ V, max } I_O$		89.5	9.25	W
P_{d}	Power Dissipation	on			9.00		
			$V_1 = 12 \text{ V}, \text{ max } I_0$		10.55	10.75	W
Pli	Input idling power	er	$V_1 = 5 \text{ V}, I_0 = 0 \text{ A}$		0.53		W
			$V_1 = 12 \text{ V}, I_0 = 0 \text{ A}$		1.33		W
P_{RC}	Input standby po	wer	V _I = 5 V (turned off with RC)		42.4		mW
	1		V _I = 12 V (turned off with RC)		101.8		mW
Is	Static Input curre	ent	$V_I = 5 \text{ V}, \text{ max } I_O$		19.80		A
			$V_I = 12 \text{ V, max } I_O$		7.45		Α
fs	Switching freque	ency	0-100 % of max I ₀		600		kHz
V_{Oi}	Output voltage in accuracy	nitial setting and	$T_{P1} = +25^{\circ}C$, $V_{I} = 5/12 \text{ V}$, max I_{O}	1.782	1.800	1.818	V
	Output voltage tolerance band		10-100 % of max I _O	1.773		1.827	V
	Idling voltage		V _I = 5 V, I _O = 0 A		1.802		V
V_{O}			V _I = 12 V, I _O = 0 A		1.802		
	Line regulation		max I _O		±5		mV
	Load regulation		$V_I = 5/12 \text{ V}, 0-100 \text{ % of max } I_O$		±5		mV
V_{tr}	Load transient voltage deviation	1	V _I = 12 V, Load step 25-75-25 % of max I _O , di/dt = 2.5 A/µs		±160		mV
t _{tr}	Load transient re	ecovery time	Without Turbo Trans		100		μS
V _{tr}	Load transient voltage deviation	•	C _o =660 µF Type C V ₁ = 12 V, Load step 25-75-25 % of max I _o , di/dt = 2.5 A/µs		±45		mV
t _{tr}	Load transient re		With Turbo Trans C _o =3300 µF Type C; R _{TT} =SHORT		100		μS
t _r	Ramp-up time (from 10-90 % of \	/ _{Oi})			8.5		ms
ts	Start-up time (from V _I connection	•	$-V_1 = 5 \text{ V}, 100 \% \text{ of max } I_0$		21.0		ms
t _r	Ramp-up time (from 10-90 % of	V _{Oi})	V _I = 12 V, 100 % of max I _O		8.3		ms
ts	Start-up time (from V _I connection	n to 90 % of V _{Oi})	V ₁ = 12 V, 100 % of max 10	13.4			ms
	V _I shut-down	V _I = 5 V	Max I ₀		0.073		ms
t _f	fall time.	-	I _O = 0.1 A		44.1		ms
71	(From V _I off to 10 % of V _O)	V _I = 12 V	Max I _O		0.627		ms
	10 % OI V _O) V _I = 12 V		I _O = 0.1 A		95.1		ms
$t_{\text{RC}} t_{\text{Inh}}$	RC start-up time		$V_I = 5 \text{ V}$, Max I_O		22.0		ms
	No start-up tillle		V _I = 12 V , Max I _O		12.3		ms
	RC shut-down	V 5 V	Max I _O		0.060		ms
	fall time	V ₁ = 5 V	I _o = 0.1 A		41.8		ms
t _{RC} t _{Inh}	RC start-up time		$V_I = 5 \text{ V}$, Max I_O $V_I = 12 \text{ V}$, Max I_O Max I_O		22.0 12.3 0.060		







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Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

	(From RC off to 10 % of V_0) $V_1 = 12 \text{ V}$	Max I _O	0.143		ms
	to 10 % of V_0) $V_1 = 12 V$	I _o = 0.1 A	42.2		ms
Io	Output current		0	50	Α
I _{lim}	Current limit threshold	$T_{P1} < max T_{P1}$	100		Α
I _{sc}	Short circuit current	T _{P1} = 25°C, see Note 3	110		Α
C _{out}	Recommended Capacitive Load	T _{P1} = 25°C, see Note 4	1000	10000	μF
V _{Oac}	Output ripple & noise V _I = 5 V	See ripple & noise section, max I _O	8		mVp-p
V _{Oac}	Output ripple & noise V _I = 12 V	See ripple & noise section, max I ₀	12		mVp-p

Note 1: Frequency may be adjusted with SmartSync pin. See Operating Information section

Note 2: See Operating Information section for TurboTrans technology

Note 3: Describe short circuit current characteristic, i.e. fold-back, hiccup or RMS, in one short sentence or type only "See Operating Information section.")

Note 4: 1000 µF of external non-ceramic output capacitance is required for basic operation. Adding additional capacitance at the load further improves transient response. Up to 1000 µF of ceramic capacitance may be added in addition to the required non-ceramic capacitance. When not using TurboTrans technology, 8000 µF capacitance is allowed; When using TurboTrans technology, up to 10000 µF capacitance is allowed. For more information, see Operating Information Section.

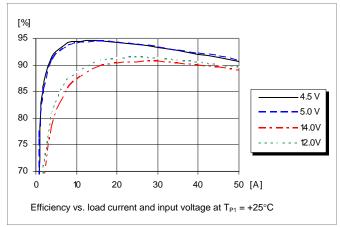


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PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013
Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

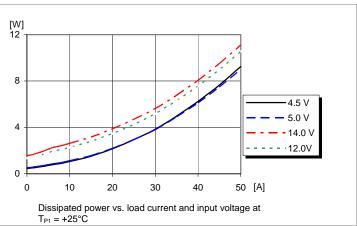
1.8V, 50A / 90.0W Typical Characteristics

PMR 5118UW

Efficiency



Power Dissipation

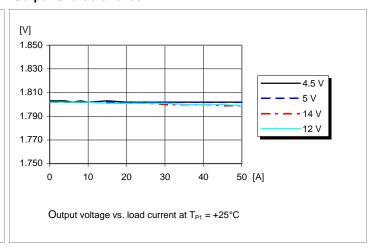


Output Current Derating

TBD

Output current derating information is available only for 1.2V and 3.3V output.

Output Characteristics



Output Ripple & Noise

Output voltage ripple at: $T_{P1} = +25^{\circ}C$, $V_{I} = 12 V$, $I_{O} = 50 A$ resistive load.

Trace: output voltage 5 mV/div. Time scale: 2 μs/div.

Output Voltage Adjust (see operating information)

Passive adjust

The resistor value for an adjusted output voltage is calculated by using the equations in the operating information.

$$R_{SET} = 30.1 \, k\Omega \times \frac{0.7}{V_0 - 0.7} - 6.49 \, k\Omega$$



PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013	
Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB	

2.5V, 50A / 125.0W Electrical Specification

PMR 5118UW

 T_{ref} = -40 to +85°C, V_I = 4.5 to 14 V, R_{SET} = 5.23 kΩ, unless otherwise specified under Conditions. Typical values given at: T_{P1} = +25°C, V_I = 5/12 V, max I_O , unless otherwise specified under Conditions. Additional C_{in} = 1000+22 μF and C_{out} = 660 μF. See Operating Information section for selection of capacitor types. Connect the sense pin, where available, to the output pin.

Charac	aracteristics		Conditions	min	Тур	max	Unit
V_{I}	Input voltage rar	nge		4.5	5/12	14	V
V_{loff}	Turn-off input vo	ltage	Decreasing input voltage	4.0	4.2		V
V_{lon}	Turn-on input voltage		Increasing input voltage		4.3	4.45	V
Cı	Internal input ca	pacitance			44		μF
Po	Output power			0		50	W
		V _I = 5 V	V _I = 5 V, 50 % of max I _O		95.3		
n	Efficiency	V ₁ = 3 V	V _I = 5 V, max I _O		93.0		%
η	Efficiency	V ₁ = 12 V	V _I = 12 V, 50 % of max I _O		93.2		70
		V ₁ = 12 V	V _I = 12 V, max I _O		91.8		
P _d	Power Dissipation	n .	$V_I = 5 \text{ V, max } I_O$		9.46	9.75	W
Γd	Power Dissipation	л	V _I = 12 V, max I _O		11.3	11.62	W
Pli	Input idling power	or.	V _I = 5 V, I _O = 0 A		0.56		W
Γli	Imput failing powe	3 1	V _I = 12 V, I _O = 0 A		1.57		W
P _{RC}	Input standby po	nwer.	V _I = 5 V (turned off with RC)		47.1		mW
I RC	input standby po	, wei	V _I = 12 V (turned off with RC)		109.3		mW
I.	Static Input curre	ont	$V_I = 5 \text{ V, max } I_O$		26.88		Α
Is	Static input curre	51 IL	V _I = 12 V, max I _O		11.35		Α
f_s	Switching freque	ency	0-100 % of max I _O		600		kHz
V_{Oi}	Output voltage in accuracy	nitial setting and	$T_{P1} = +25^{\circ}C, V_{I} = 5/12 \text{ V, max } I_{O}$	2.475	2.500	2.525	V
	Output voltage tolerance band		10-100 % of max I _o	2.462		2.538	V
	Idling voltage		V _I = 5 V, I _O = 0 A		2.500		V
V_{O}			V _I = 12 V, I _O = 0 A		2.500		
	Line regulation		max I _O		±5		mV
	Load regulation		$V_1 = 5/12 \text{ V}, 0-100 \text{ % of max } I_0$		±5		mV
V_{tr}	Load transient voltage deviation	1	V_1 = 12 V, Load step 25-75-25 % of max I_0 , di/dt = 2.5 A/ μ s		±160		mV
\mathbf{t}_{tr}	Load transient re	ecovery time	Without Turbo Trans C₀=660 μF Type C		100		μS
V _{tr}	Load transient voltage deviation	1	V _I = 12 V, Load step 25-75-25 % of max I _O , di/dt = 2.5 A/μs		±45		mV
t _{tr}	Load transient re	ecovery time	With Turbo Trans C₀=3300 µF Type C; R _{TT} =SHORT		100		μS
t _r	Ramp-up time (from 10-90 % of \	V _{Oi})	V _I = 5 V, 100 % of max I _O		8.7		ms
ts	Start-up time (from V _I connection	n to 90 % of V _{Oi})	, , , , , , , , , , , , , , , , , , , ,		19.2		ms
t _r	Ramp-up time (from 10-90 % of	V _{Oi})	V _I = 12 V, 100 % of max I _O		8.5		ms
ts	Start-up time (from V _I connection	on to 90 % of V _{Oi})			13.0		ms
	V _I shut-down	V _I = 5 V	Max I _O		0.139		ms
t _f	fall time.	V - 5 V	I _O = 0.1 A		70.8		ms
•	(From V _I off to 10 % of V _O)		Max I _O		0.587		ms
	12 /2 3. 10,	·	I _O = 0.1 A		102.4		ms
$t_{\text{RC}}\;t_{\text{Inh}}$	RC start-up time		$V_I = 5 \text{ V}$, Max I_O		21.4		ms
			V _I = 12 V , Max I _O		12.7		ms
	RC shut-down $V_i = 5 V$		Max Io		0.115		ms





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	fall time		I _o = 0.1 A		60.3		ms
	(From RC off to 10 % of V _O)	V _I = 12 V	Max I _O		0.196		ms
	(V ₁ = 12 V	I _o = 0.1 A		61.6		ms
Io	Output current			0		50	Α
I _{lim}	Current limit thre	eshold	$T_{P1} < max T_{P1}$	100		А	
I _{sc}	Short circuit curr	rent	T _{P1} = 25°C, see Note 3	110		Α	
C _{out}	Recommended	Capacitive Load	T _{P1} = 25°C, see Note 4	1000 10000		μF	
V _{Oac}	Output ripple & r	noise V _I = 5 V	See ripple & noise section, max I _O	7		mVp-p	
V _{Oac}	Output ripple & r	noise V _I = 12 V	See ripple & noise section, max I _O	12		mVp-p	

Note 1: Frequency may be adjusted with SmartSync pin. See Operating Information section

Note 2: See Operating Information section for TurboTrans technology

Note 3: Describe short circuit current characteristic, i.e. fold-back, hiccup or RMS, in one short sentence or type only "See Operating Information section.")

Note 4: 1000 µF of external non-ceramic output capacitance is required for basic operation. Adding additional capacitance at the load further improves transient response. Up to 1000 µF of ceramic capacitance may be added in addition to the required non-ceramic capacitance. When not using TurboTrans technology, 8000 µF capacitance is allowed; When using TurboTrans technology, up to 10000 µF capacitance is allowed. For more information, see Operating Information Section.

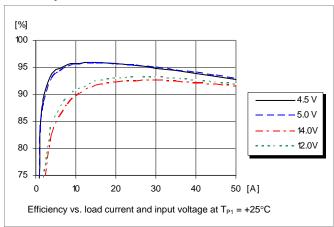


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PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013
Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

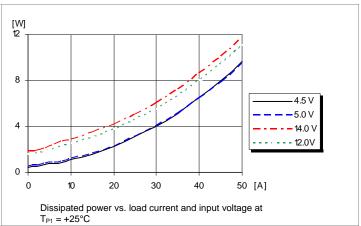
2.5V, 50A / 125.0W Typical Characteristics

PMR 5118UW

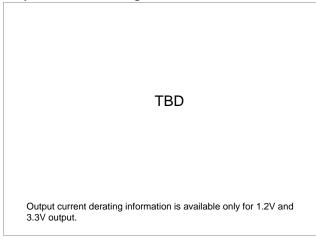
Efficiency



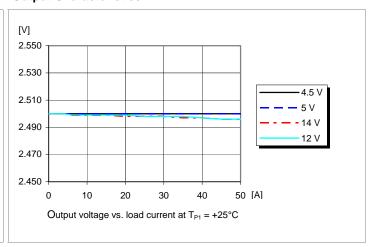
Power Dissipation

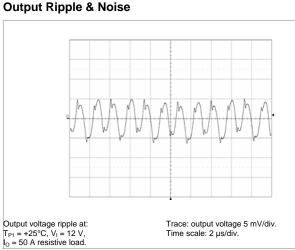


Output Current Derating



Output Characteristics





Output Voltage Adjust (see operating information)

Passive adjust

The resistor value for an adjusted output voltage is calculated by using the equations in the operating information.

$$R_{SET} = 30.1 \, k\Omega \times \frac{0.7}{V_0 - 0.7} - 6.49 \, k\Omega$$



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Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

3.3V, 50A / 165.0W Electrical Specification

PMR 5118UW

 T_{P1} = -40 to +85°C, V_I = 4.5 to 14 V, R_{SET} = 1.62 kΩ, unless otherwise specified under Conditions. Typical values given at: T_{P1} = +25°C, V_I = 5/12 V, max I_O , unless otherwise specified under Conditions. Additional C_{in} = 1000+22 μ F and C_{out} = 660 μ F. See Operating Information section for selection of capacitor types. Connect the sense pin, where available, to the output pin.

Charac	cteristics		Conditions	min	Тур	max	Unit
Vı	Input voltage range			4.5	5/12	14	V
V _{loff}	Turn-off input voltage		Decreasing input voltage	4.0	4.2		V
V _{Ion}	Turn-on input voltage		Increasing input voltage		4.3	4.45	V
Cı	Internal input cap	pacitance			44		μF
Po	Output power			0		50	W
		V _I = 5 V	V _I = 5 V, 50 % of max I _O		96.3		
n	Efficiency	V ₁ = 5 V	V _I = 5 V, max I _O		94.4		0/
η	Efficiency	V _I = 12 V	V _I = 12 V, 50 % of max I _O		94.4		%
		V ₁ = 12 V	V _I = 12 V, max I _O		93.4		1
D	Dawer Dissipatio		$V_I = 5 \text{ V, max } I_O$		9.80	10.2	W
P _d	Power Dissipation	ori	V _I = 12 V, max I _O		11.57	11.8	W
Б	Lancat Calling and a second		$V_1 = 5 \text{ V}, I_0 = 0 \text{ A}$		0.51		W
Pli	Input idling power	er	V _I = 12 V, I _O = 0 A		1.86		W
Б	1 1 -1 11		V _I = 5 V (turned off with RC)		42.4		mW
P _{RC}	Input standby po	ower	V _I = 12 V (turned off with RC)		94.3		mW
	Otaria la santa santa		$V_I = 5 \text{ V, max } I_O$		34.96		Α
I _S	Static Input curre	ent	V _I = 12 V, max I _O		14.72		А
fs	Switching freque	ency	0-100 % of max I _O		600		kHz
V _{Oi}	Output voltage initial setting and accuracy		T _{P1} = +25°C, V _I = 5/12 V, max I _O	3.267	3.300	3.333	V
	Output voltage to	olerance band	10-100 % of max I _O	3.250		3.350	V
	Idling voltage		$V_1 = 5 \text{ V}, I_0 = 0 \text{ A}$		3.299		V
Vo			V _I = 12 V, I _O = 0 A		3.299		
	Line regulation		max I _O		±5		mV
	Load regulation		$V_1 = 5/12 \text{ V}, 0-100 \% \text{ of max } I_0$		±5		mV
V _{tr}	Load transient voltage deviation		V_1 = 12 V, Load step 25-75-25 % of max I_0 , di/dt = 2.5 A/ μ s		±160		mV
t _{tr}	Load transient re	ecovery time	Without Turbo Trans C₀=TBD μF Type C		100		μS
V _{tr}	Load transient voltage deviation	1	V ₁ = 12 V, Load step 25-75-25 % of max I _O , di/dt = 2.5 A/μs		±45		mV
t _{tr}	Load transient re	ecovery time	With Turbo Trans $C_o = TBD \mu F$ Type $C; R_{TT} = TBD k\Omega$		100		μS
t _r	Ramp-up time (from 10-90 % of \	/ _{Oi})	V _I = 5 V, 100 % of max I _O		8.7		ms
ts	Start-up time (from V ₁ connection	n to 90 % of V _{Oi})			17.7		ms
t _r	Ramp-up time (from 10-90 % of V _{Oi})		V _I = 12 V, 100 % of max I _O		8.7		ms
t _s	Start-up time (from V ₁ connection to 90 % of V _{Oi})				13.1		ms
	V _I shut-down fall time. (From V _I off to 10 % of V _O)	V _I = 5 V	Max I _O		0.141		ms
t _f		V _I = 12 V	I _O = 0.1 A		81.2		ms
1			Max I _O		0.414		ms
	10 /0 01 10/	''	I _O = 0.1 A		124.1		ms
$t_{RC} \; t_{Inh}$	RC start-up time RC shut-down $V_1 = 5 V$		$V_I = 5 \text{ V}$, Max I_O		21.8		ms
			$V_I = 12 \text{ V}$, Max I_O		12.1		ms
			Max I _O		0.126		ms



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	fall time		I _o = 0.1 A		81.8		ms
	(From RC off to 10 % of V _O)	V _I = 12 V	Max I _O		0.206		ms
		V - 12 V	$I_0 = 0.1 \text{ A}$		79.3		ms
Io	Output current			0		50	Α
I _{lim}	Current limit thre	eshold	$T_{P1} < max T_{P1}$	100		Α	
I _{sc}	Short circuit curr	rent	$T_{P1} = 25^{\circ}C$, see Note 3	110		Α	
Cout	Recommended	Capacitive Load	T _{P1} = 25°C, see Note 4	1000 10000		μF	
V _{Oac}	Output ripple & r	noise V _I = 5 V	See ripple & noise section, max I _O	7		mVp-p	
V _{Oac}	Output ripple & r	noise V _I = 12 V	See ripple & noise section, max I _O	12		mVp-p	

Note 1: Frequency may be adjusted with SmartSync pin. See Operating Information section

Note 2: See Operating Information section for TurboTrans technology

Note 3: Describe short circuit current characteristic, i.e. fold-back, hiccup or RMS, in one short sentence or type only "See Operating Information section.")

Note 4: 1000 µF of external non-ceramic output capacitance is required for basic operation. Adding additional capacitance at the load further improves transient response. Up to 1000 µF of ceramic capacitance may be added in addition to the required non-ceramic capacitance. When not using TurboTrans technology, 8000 µF capacitance is allowed; When using TurboTrans technology, up to 10000 µF capacitance is allowed. For more information, see Operating Information Section.

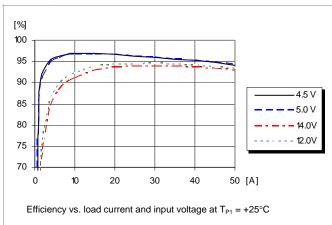


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Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB

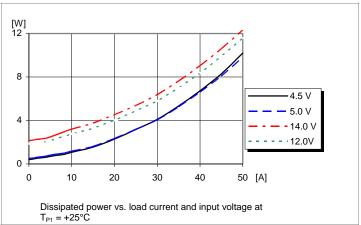
3.3V, 50A / 165.0W Typical Characteristics

PMR 5118UW

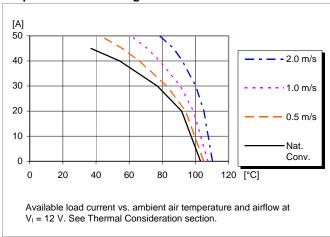
Efficiency



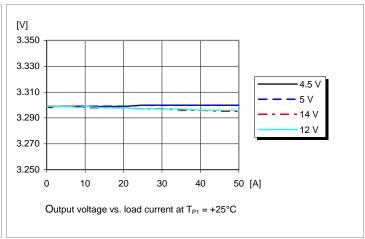
Power Dissipation

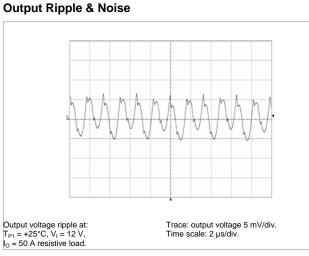


Output Current Derating



Output Characteristics





Output Voltage Adjust (see operating information)

The resistor value for an adjusted output voltage is calculated by using the equations in the operating information.

$$R_{SET} = 30.1 \,\text{k}\Omega \times \frac{0.7}{V_0 - 0.7} - 6.49 \,\text{k}\Omega$$

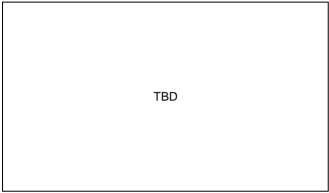


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PMR 5000 series PoL Regulator	EN/LZT 146 410 R1B March 2013	
Input 4.5 - 14 V, Output up to 50 A / 180 W	© Ericsson AB	

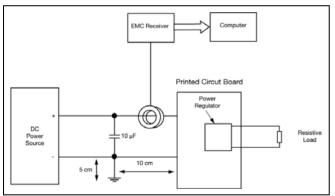
EMC Specification

Conducted EMI measured according to test set-up. The fundamental switching frequency is 600 kHz for PMR 5118UW@ $V_I = 12 \text{ V}$, max I_O .

Conducted EMI Input terminal value (typ)



EMI without filter



Test set-up

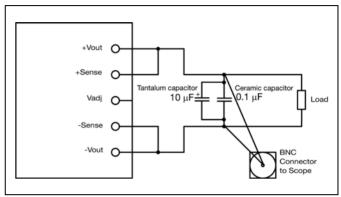
Layout recommendations

The radiated EMI performance of the product will depend on the PCB layout and ground layer design. It is also important to consider the stand-off of the product. If a ground layer is used, it should be connected to the output of the product and the equipment ground or chassis.

A ground layer will increase the stray capacitance in the PCB and improve the high frequency EMC performance.

Output ripple and noise

Output ripple and noise measured according to figure below. See Design Note 022 for detailed information.



Output ripple and noise test setup

Operating information

Extended information for POLA products is found in Application Note 205.

Input Voltage

The input voltage range 4.5 to 14 Vdc makes the product easy to use in intermediate bus applications when powered by a non-regulated bus converter or a regulated bus converter.

Turn-off Input Voltage

The products monitor the input voltage and will turn on and turn off at predetermined levels.

The typical hysteresis between turn on and turn off input voltage is 0.1V.

Turn on/off voltage can be adjusted by using UVLO (Undervoltage lockout) function. The UVLO character is defined by the ON threshold (V_{THD}) voltage. Below the ON threshold, the Inhibit control is overridden, and the module does not produce an output.

The UVLO feature allows for limited adjustment of the ON threshold voltage. It is made by using a single resistor between the Inhibit/UVLO pin (pin 21) and ground pins (pin 8,9,12,13). The $V_{\mbox{THD}}$ value can be adjusted from 5.5V to 11V. Default value of $V_{\mbox{THD}}$ is 4.3V.

Below equation determines the value of resistor required to adjust $V_{\mbox{\scriptsize THD}}$ to a new value.

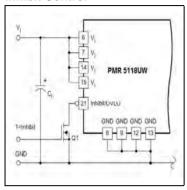
$$R_{\text{UVLO}} = \frac{230}{V_{\text{THD}} - 4.6} (k\Omega)$$



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Inhibit Control



The products are fitted with a remote control function by using the Inhibit/UVLO pin. The Inhibit control function allows the product to be turned on/off by an external device like a semiconductor or mechanical switch. The RC pin has an internal pull up. An external pull-up resistor should never be used with the inhibit pin.

When the Inhibit pin is left open, the regulator will turn on when the input voltage is applied. Turn off is achieved by connecting the Inhibit pin to the GND.

The reference figure above shows the typical application of the inhibit function. The input is not compatible with TTL logic device. An open-collector (or open-drain) discrete transistor is recommended for control. Turning the discrete transistor on applies a low voltage to the Inhibit control pin and disables the output of the module. If this device is then turned off, the module executes a soft-start power-up sequence. A regulated output voltage is produced within 20 ms.

External Decoupling Capacitors

Input capacitors:

The PMR 5118UW requires a minimum input capacitance of e 1000 μ F. The ripple current rating of the input capacitor must be at least 600 mA rms. An optional 22 μ F X5R/X7R ceramic capacitor is recommended to reduce RMS ripple current.

The size and value of the input capacitor is determined by the converter's transient performance capability. This minimum value assumes that the converter is supplied with a responsive, low inductance input source. This source should have ample capacitive decoupling, and be distributed to the converter via PCB power and ground planes.

Ceramic capacitors should be located as close as possible to the module's input pins, within 0.5 inch (1,3 cm). Adding ceramic capacitance is necessary to reduce the high-frequency ripple voltage at the module's input. This reduces the magnitude of the ripple current through the electrolytic capacitor, as well as the amount of ripple current reflected back to the input source. Additional ceramic capacitors can be added to further reduce the RMS ripple current requirement for the electrolytic capacitor.

The main considerations when selecting input capacitors are the RMS ripple current rating, temperature stability, and less than 100 m Ω of equivalent series resistance (ESR).

Regular tantalum capacitors are not recommended for the input bus. These capacitors require a recommended minimum

voltage rating of 2x (maximum dc voltage + ac ripple). This is standard practice to ensure reliability. No tantalum capacitors were found with a sufficient voltage rating to meet this requirement.

Output capacitors:

The PMR 5118UW module requires a minimum output capacitance of 660µF of polymer-aluminum, tantulum, or polymer-tantalum type.

The required capacitance above the minimum is determined by actual transient deviation requirements. See "TurboTrans Technology" information below.

For both input and output capacitors, when the operating temperature is below 0°C, the ESR of aluminium electrolytic capacitors increases. For these applications, OS-CON, polyaluminium, and polymer-tantalum types should be considered.

If the TurboTrans feature is not used, minimum ESR and maximum capacitor limits must be followed. System stability may be effected and increased output capacitance may be required without TurboTrans.

When using the PMR 5118UW, observe the minimum ESR of the entire output capacitor bank. The minimum ESR limit of the output capacitor bank is $7m\Omega$.

When using PMR 5118UW without the TurboTrans feature, the maximum amount of capacitance is 1000 µF of ceramic type. Large amounts of capacitance may reduce system stability.

Utilizing the TurboTrans feature improves system stability, improves transient response, and reduces the amount of output capacitance required to meet system transient design requirements. For detaile information, see "TurboTrans Technology" information below.

Output Voltage Adjust (Vadj)

The product has an Output Voltage Adjust function. The function can be used to adjust the output voltage in the range from 0.7V to 3.6V.

The Vo Adjust control sets the output voltage of the PMR 5118UW. The adjustment method requires the addition of a single external resistor, RSET, that must be connected directly between pins Vo Adjust (pin 18) and AGND (pin 4). The value of the required resistor can be calculated using the following formula.

$$R_{SET} = 30.1(k\Omega) \times \frac{0.7}{V_{O} - 0.7} - 6.49(k\Omega)$$

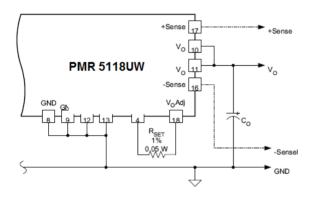
Note:

(1) $R_{\overline{SET}}$: Use a 0.05 W resistor with a tolerance of 1% and temperature stability of 100 ppm/°C (or better). Connect the



resistor directly between pins 18 and 4, as close to the regulator as possible, using dedicated PCB traces.

- (2) Never connect capacitors from Vo Adjust to either + Sense, GND, or Vo. Any capacitance added to the Vo Adjust pin affects the stability of the regulator.
- (3) For output voltages less than 1.2 V, the output ripple may increase (up to $2\times$) when operating at input voltages greater than (Vo \times 12). Adjusting the switching frequency using the SmartSync feature may increase or decrease this ratio.



Parallel Operation

The PMR 5818UW module is capable of being configured in parallel with another PMR 5118UW module to share load current. To parallel the two modules, it is necessary to configure one module as the Master and one module as the Slave. To configure a module as the Master, connect the CONFIG pin (pin 1) to GND. The CONFIG pin of the Slave must be connected to V_I. In order to share current, pins 2 through 5 of both the Master and Slave must be connected between the two modules. The module that is configured as

between the two modules. The module that is configured as the MASTER is used to control all of the functions of the two modules including Inhibit, ON/OFF control, AutoTrack sequencing, TurboTrans, SmartSync, +/- Remote Sense, and Output Voltage Adjust. The MASTER and the SLAVE must be powered from the same input voltage supply.

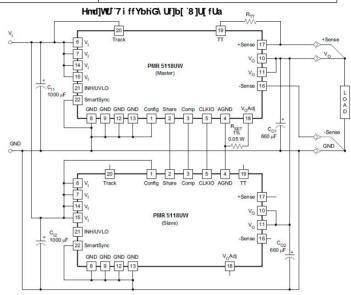
When using TurboTrans while paralleling two modules, the TurboTrans resistor, R_{TT}, must be connected from the TurboTrans pin (pin 19) of the Master module to the +Sense

pin (pin 17) of the Master module. When paralleling modules the procedure to calculate the proper value of output capacitance and R_{TT} is similar to that explained in the

TurboTrans Selection section, however the values must be calculated for a single module. Therefore, the total output current load step must be halved before determining the required output capacitance and the R_{TT} value as explained

in the TurboTrans Selection section. The value of output capacitance calculated is the minimum required output capacitance per module and the value of RTT must be calculated using this value of output capacitance. The TurboTrans pin of the Slave module must be left open.





Required Connections for Current Sharing

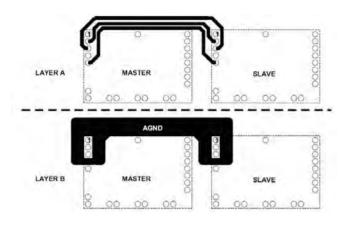
TERMINAL NAME NO.		MASTER	SLAVE		
		MASIER	SLAVE		
VI	6,7,14,15	Connect to the Input Bus.	Connect to the Input Bus.		
Vo	10,11	Connect to the Output Bus.	Connect to the Output Bus.		
GND	8,9,12,13	Connect to Common Power GND.	Connect to Common Power GND.		
Inhibit and UVLO	21	Use for Inhibit control & UVLO adjustments. If unused leave open-circuit.	No Connection. Leave open-circuit.		
V _o Adjust	18	Use to set the output voltage. Connect R _{SET} resistor between this pin and AGND (pin 4).	No Connection. Leave open-circuit.		
+Sense	17	Connect to the output voltage either at the load or at the module (pin 11).	No Connection. Leave open-circuit.		
-Sense	16	Connect to the output GND either at the load or at the module (pin 13).	No Connection. Leave open-circuit.		
Track	20	Connect to Track control or to V _I (pin 15).	No Connection. Leave open-circuit.		
TurboTrans™	19	Connect TurboTrans resistor, R _{TT} , between this pin and +Sense (pin 17).	No Connection. Leave open-circuit.		
SmartSync	22	Connect to an external clock. If unused connect to GND.	Connect to Common Power GND.		
CONFIG	1	Connect to GND.Á	Connect to the Input Bus.		
Share	2	Connect to pin 2 of Slave.	Connect to pin 2 of Master.		
Comp	3	Connect to pin 3 of Slave.	Connect to pin 3 of Master.		
AGND	4	Connect to pin 4 of Slave.	Connect to pin 4 of Master.		
CLKIO	5	Connect to pin 5 of Slave.	Connect to pin 5 of Master.		

Current sharing layout

In current sharing applications the V_I pins of both modules must be connected to the same input bus. The V_O pins of both modules are connected together to power the load. The GND pins of both modules are connected via the GND plane. Four other inter-connection pins are connected between the modules. Below figure shows the required layout of the inter-connection pins for two modules configured to share current. Notice that the Share (pin 2) connection is routed between the Comp (pin 3) and CLKIO (pin 5) connections. AGND (pin 4) should be connected as a thicker trace on an adjacent layer, running parallel to pins 2, 3 and 5. AGND must not be connected to the GND plane.



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Remote Sense

The products have remote sense that can be used to compensate for voltage drops between the output and the point of load. The sense traces should be located close to the PCB ground layer to reduce noise susceptibility. The remote sense circuitry will compensate for up to 0.3 voltage drop between output pins and the point of load. If the remote sense is not needed +Sense should be

connected to +Out and -Sense should be connected to -Out.

Over Temperature Protection (OTP)

The regulators are protected from thermal overload by an internal over temperature shutdown circuit. If the internal temperature exceeds the OTP threshold, the module's inhibit control is internally pulled low. This turns the output off. The voltage drops as the external output capacitors are discharged by the load circuit. The product will make continuous attempts to start up (non-latching mode) and resume normal operation automatically when the temperature has dropped >10°C below the temperature threshold.

Over Current Protection (OCP)

The regulators include current limiting circuitry for protection at continuous overload. The output voltage will decrease towards zero for output currents in excess of max output current (max l_0). The regulator will resume normal operation after removal of the overload. The load distribution should be designed for the maximum output short circuit current specified.

Soft-start Power Up

From the moment a valid input voltage is applied, the soft-start control introduces a short time-delay (typically 5-15 ms) before allowing the output voltage to rise. The initial rise in input current when the input voltage first starts to rise is the charge current drawn by the input capacitors.

Auto-Track™ Function

Auto-Track was designed to simplify the amount of circuitry required to make the output voltage from each unit power up and power down in sequence. The sequencing of two or more supply voltages during power up is a common requirement for complex mixed-signal applications, that use dual-voltage VLSI

ICs such as DSPs, micro-processors and ASICs.

Notes on Use of Auto-Track™

- 1. The Track pin voltage must be allowed to rise above the module set-point voltage before the module regulates at its adjusted set-point voltage.
- 2. The Auto-Track function tracks almost any voltage ramp during power up, and is compatible with ramp speeds of up to 1 V/ms.
- 3. The absolute maximum voltage that may be applied to the Track pin is the input voltage V_I .
- 4. The module cannot follow a voltage at its track control input until it has completed its soft-start initialization.

 This takes about 20 ms from the time that a valid voltage has been applied to its input. During this period, it is recommended that the Track pin be held at ground potential.

 5. The Auto-Track function is disabled by connecting the Track pin to the input voltage (V_I). When Auto-Track is disabled, the output voltage rises according to its softstart rate after input power has been applied.
- The Auto-Track pin should never be used to regulate the module's output voltage for long-term, steady-state operation.

Smart Sync

Smart Sync is a feature that allows multiple power modules to be synchronized to a common frequency. When not used, this pin must be connect to GND. Driving the Smart Sync pins with an external oscillator set to the desired frequency, synchronizes all connected modules to the selected requency. The synchronization frequency can be higher or lower than the nominal switching frequency of the modules within the range of 240 kHz to 400 kHz.

Synchronizing modules powered from the same bus eliminates beat frequencies reflected back to the input supply, and also reduces EMI filtering requirements. Eliminating the low beat frequencies (usually<10kHz) allows the EMI filter to be designed to attenuate only the synchronization frequency. Power modules can also be synchronized out of phase to minimize ripple current and reduce input capacitance requirements.

The PMR 5118UW requires that the external synchronization frequency be present before a valid input voltage is present or before release of the inhibit control.

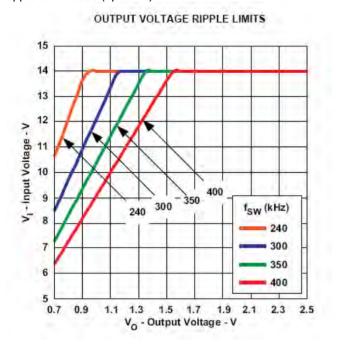
Operating the PMR 5118UW with a low duty cycle may increase the output voltage ripple. When operating at the nominal switching frequency, input voltages greater than $(V_O \times 12)$ may cause the output voltage ripple to increase (up to $2\times$).

When using Smart Sync, the minimum duty cycle varies as a function of output voltage and switching frequency. Synchronizing to a higher frequency causes greater restrictions on the duty cycle range. For a given switching frequency, below figure shows the operating region where the output voltage ripple meets the electrical specifications.



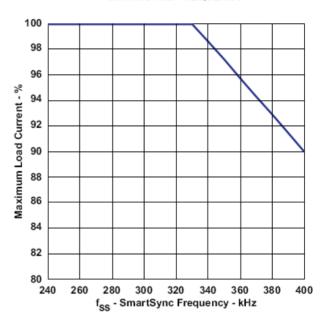
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Operation above a given curve may cause the output voltage ripple to increase (up to 2x).



The maximum output current that a single module can deliver may also be affected by the sychronization frequency. See Figure below for load current derating when sychronizing at frequencies greater than 330 kHz. First consult the temperature derating graphs in the Typical Characteristics section to determine the maximum output current based on operating conditions. Any derating due to the SmartSync frequency is in addition to the temperature derating.

MAXIMUM LOAD CURRENT SMARTSYNC FREQUENCY



Pre-Bias Startup Capability

A prebias startup condition occurs as a result of an external voltage being present at the output of a power module prior to its output becoming active. This often occurs in complex digital systems when current from another power source is backfed through a dual-supply logic component, such as FPGA or ASIC.

The PMR family of regulators incorporate synchronous rectifiers, but will not sink current during startup, or whenever the Inhibit pin is held low. However, to ensure satisfactory operation of this function, certain conditions must be maintained.

For more inforamtion, please refer to Application Note 205.

Turbo Trans[™] Technology
Turbo Trans[™] optimizes the transient response of the regulator with added external capacitance using a single external resistor. The benefits of this technology include: reduced output capacitance, minimized output voltage deviation following a load transient, and enhanced stability when using ultra-low ESR output capacitors. The amout of output capacitance required to meet a target output voltage deviation, is reduded with Turbo TransTM activated. Likewise, for a given amout of output capacitance, with Turbo TransTM engaged, the amplitude of the voltage deviation following a load transient is reduced. Applications requiring tight transient voltage tolerances and minimized capacitor footprint area benefit from this technology.

Utilizing Turbo TransTM requires connecting a resistor, R_{TT},



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Input 4.5 - 14 V, Output up to 50 A / 180 W

EN/LZT 146 410 R1B March 2013

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between the +Sense pin (pin 17) and the Turbo TransTM pin (pin 19), The value of the resistor directly corresponds to the amount of output capacitance required. For the PMR 5118UW, the minimum required capacitance is $1000\mu F$. When using Turbo TransTM, capacitors with a capacitance×ESR product below $10,000~\mu F\times m\Omega$ are required.

To have a better understanding of the required capacitors with Turbo TransTM, three types of capacitors are defined as below.

- a. TypeA = $(100 < \text{capacitance} \times \text{ESR} \le 1,000)$
- b. TypeB = $(1,000 < \text{capacitance} \times \text{ESR} \leq 5,000)$
- c. TypeC = $(5,000 < \text{capacitance} \times \text{ESR} \leq 10,000)$

As an example, let's look at a 12-V application requiring a 60 mv deviation during an 15A load transient. A majority of 470 μ F, 10m Ω output capacitors are used. Use the 12 V, Type B capacitor chart. Dividing 60mV by 15A gives 4mV/A transient voltage deviation per amp of transient load setp. Select 4mV/A on the Y-axis and read across to the "With TurboTrans" plot. Following this point down to the X-axis gives us a minimum required output capacitance of approximately 1500μ F. The required R_{TT} resistor value for 1500μ F can then be calculated or selected from the below table. The required R_{TT} resistor is approximately $17.4K\Omega$.

To see the benefit of Turbo Trans TM , follow the 4mV/A marking across to the "Without TurboTrans" plot. Following that point down shows that you would need a minimum of $7500\mu F$ of output capacitance to meet the same transient deviation limit. This is the benefit of Turbo Trans TM .

A typical Turbo TransTM application schematic is also shown.

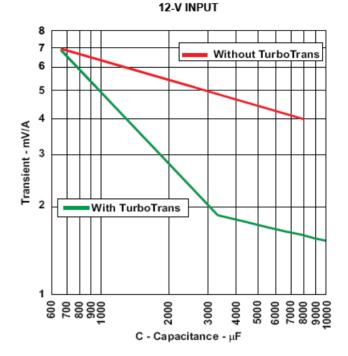


Figure 12. Capacitor Type B, 1000 < C(μF)×ESR(mΩ) ≤ 5000 (e.g. Polymer-Tantalum)

5-V INPUT

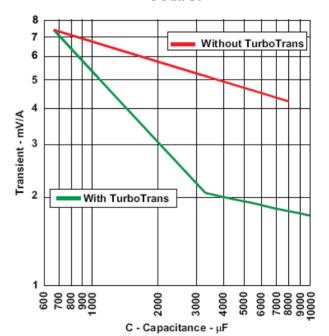


Figure 13. Capacitor Type B, 1000 < C(μF)×ESR(mΩ) ≤ 5000 (e.g. Polymer-Tantalum)



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Table: Type B TurboTrans Co Values and Required RTT Selection Table

Transient Voltage Deviation (mV)		12 Volt Input		5 Volt Input		
25% load step (12.5 A)	50% load step (25 A)	75% load step (37.5 A)	C _O Minimum Required Output Capacitance (μF)	R _{TT} Required TurboTrans Resistor (kΩ)	C _O Minimum Required Output Capacitance (μF)	R _{TT} Required TurboTrans Resistor (kΩ)
100	200	300	660	open	660	open
85	170	255	660	open	750	226
75	150	225	800	143	870	93.1
60	120	180	1050	46.4	1150	34.8
50	100	150	1300	24.9	1450	18.7
40	70	105	1750	11.3	1950	8.45
30	60	90	2500	3.48	2800	1.87
25	50	75	3100	0.649	4000	short



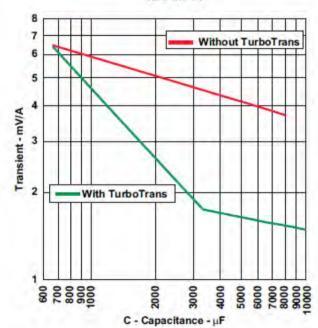


Figure 14. Capacitor Type C, $5000 < C(\mu F) \times ESR(m\Omega) \le 10,000(e.g. OS-CON)$

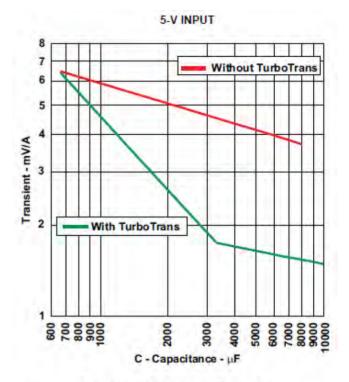


Figure 15. Capacitor Type C, $5000 < C(\mu F) \times ESR(m\Omega) \le 10,000(e.g. OS-CON)$

Table; Type B TurboTrans Co Values and Required RT Selection Table

Transient Voltage Deviation (mV)		12 Volt Input		5 Volt Input		
25% load step (12.5 A)	50% load step (25 A)	75% load step (37.5 A)	C _O Minimum Required Output Capacitance (µF)	R _{TT} Required TurboTrans Resistor (kΩ)	C _O Minimum Required Output Capacitance (µF)	R _{TT} Required TurboTrans Resistor (kΩ)
100	200	300	660	open	660	open
85	170	255	660	open	750	226
75	150	225	800	143	870	93.1
60	120	180	1050	46.4	1150	34.8
50	100	150	1300	24.9	1450	18.7
40	70	105	1750	11.3	1950	8.45
30	60	90	2500	3.48	2800	1.87
25	50	75	3100	0.649	4000	short

R_{TT} Resistor Selection

The Turbo TransTM resistor value, R_{TT} can be determined from the Turbo TransTM programming equation, see the equation below.

$$R_{TT} = 40 \times \frac{1 - (\frac{C_o}{3300})}{5 \times (\frac{C_o}{3300}) - 1} (k\Omega)$$

Where C_0 is the total output capacitance in μF . C_0 values greater than or equal to 3300 μF require R_{TT} to be a short,





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 $0\Omega.$ (The above equation results in a negative value for R_{TT} when $C_o \geq 3300~\mu F)$

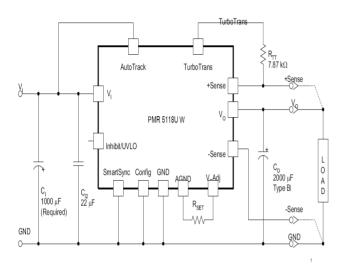


Figure: Typical TurboTrans™ Application



EN/LZT 146 410 R1B March 2013

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Thermal Consideration

General

The regulators are designed to operate in different thermal environments and sufficient cooling must be provided to ensure reliable operation.

Cooling is achieved mainly by conduction, from the pins to the host board, and convection, which is dependant on the airflow across the regulator. Increased airflow enhances the cooling of the regulator.

The typical Output Current Derating graph can be found in the Output section for each model provides the available output current vs. ambient air temperature and air velocity at $V_i = 12$ V.

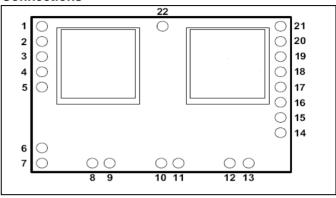
The product is tested on a 100 x 100 mm double-sided PCB with 2 oz. copper and the direction of airfow fro pin 10 to pin 22. For surface mount packages, multiple vias must be utilized.

Definition of product operating temperature

The product operating temperatures is used to monitor the temperature of the product, and proper thermal conditions can be verified by measuring the temperature at positions P1, P2, and P3. The temperature at these positions $(T_{P1}, T_{P2}, T_{P3},)$ should not exceed the maximum temperatures in the table below. The number of measurement points may vary with different thermal design and topology. Temperatures above maximum T_{P1} , measured at the reference point P1 are not allowed and may cause permanent damage.

Position	Description	Max Temp.
P1	Reference point (Q501)	T _{P1} =130° C
P2	Inductor (L501)	T _{P2} =130° C
P3	Inductor (L502)	T _{P3} =130° C
CONFIG (98) VSHARE (98) COMP (98) ADND (98) CLXID (98) VIN (98) VIN (98) CS01	P2 PXXXXXXX	MH TRK TRK MI TT ADJ ADJ VS+ VS- VIN VIN NO OSO4 MG CSO2 NO

Connections



Pin	Designation	Function
1	CONFIG	When two modules are connected together to share load current one must be configured as the MASTER and the other as the SLAVE. This pin is used to configure the module as either MASTER or SLAVE. To configure the module as the MASTER, connect this pin to GND. To condigure the module as the SLAVE, connect this pin to V _I (pin 6). When not sharing current, this pin should be connected to GND.
2	Share	This pin is used when connecting two modules together to share load current. When two modules are sharing the current the Share pin of both modules must be connected together. When not sharing current, this pin MUST be left open (floating).
3	Comp	This pin is used when connecting two modules together to share load current. When two modules are sharing current the Comp pin of bothe modules must be connected together. When not sharing current, this pin MUST be left open (floating).



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4	AGND	This pip is the internal analas
7	AGNU	This pin is the internal analog ground of the module. This pin provides the return path for the V _o Adjust resistor (R _{SET}). When two modules are sharing current the AGND pin of both modules must be connected together. Also, when two modules are connected, R _{SET} must be connected only on the MASTER module.
5	CLKIO	This pin is used when connecting two modules together to share load current. When two modules are sharing current the CLKO pin of both modules must be connected togethe. When not sharing current, this pin MUST be left open (floating).
6	V _I	The positive input voltage power node to the module, which is referenced to common GND.
7	V _I	See pin 6
8	GND	This is the common ground connection for the V_I and V_o power connections. It is also the 0 V_{dc} reference for the control inputs.
9	GND	See pin 8
10	V _o	This regulated positive power output with respect to GND.
11	V_{o}	See pin 10
12	GND	See pin 8
13	GND	See pin 8
14	V _I	See pin 6
15	V ₁	See pin 6
16	-Sense	The sense input allows the regulation circuit to compensate for voltage drop between the module and the load. The –Sense pin should always be connected to GND, either at the load for optimal voltage accuracy, or at the module (pin 13).

17	+Sense	The sense input allows the regulation circuit to compensate for voltage drop between the module and the load. The +Sense pin should always be connected to Vo, either at the load for optimal voltage accuracy, or at the module (pin 11).
18	V _o Adjust	A 0.05 W 1% resistor must be directly connected between this pin and pin4 (AGND) to set the output voltage to a value higher than 0.7 V. The temperature stability of the resistor should be 100 ppm/°C (or better). The setpoint range for the output voltage is from 0.7V to 3.6V. If left open circuit, the output voltage defaults to its lowest value. For further information, refer to the information for each output voltage sector.
19	Turbo Trans ^{1M}	This input pin adjusts the transient response of the regulator. To activate the Turbo Trans [™] feature, a 1%, 50mW resistor must be connected between this pin and pin 17 (+Sense) very close to the module. For a given value of output capacitance, a reduction in peak output voltage deviation is achieved by using this feature. If unused, this pin must be left open-circuit. External capacitance must never be connected to this pin. The resistance requirement can be selected from the Turbo Trans [™] resistor table which is shown above.



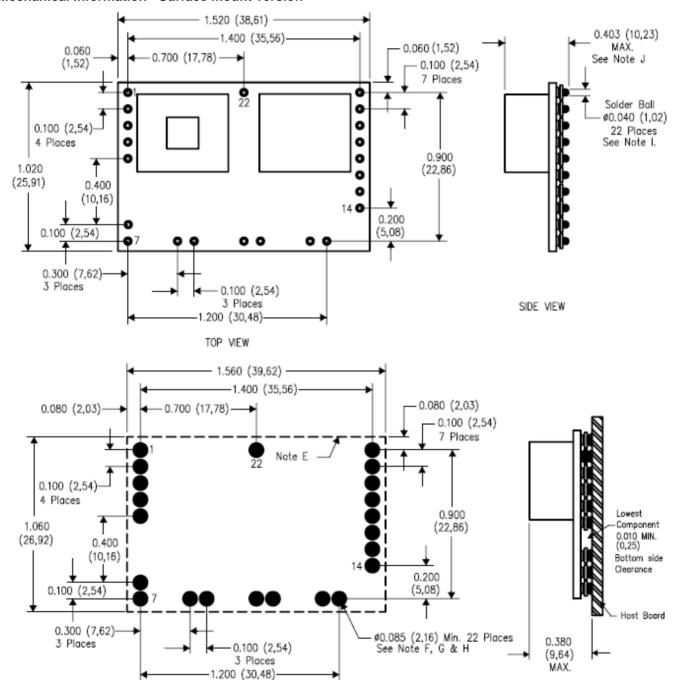
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20	Track	This is an analog control input
	Track	that enables the output
		voltage to follow an external
		voltage. This pin becomes
		active typically 25 ms after
		the input voltage has been
		applied, and allows direct
		control of the output voltage
		from 0 V up to the nominal
		set-point voltage. Within this
		range the module's output
		voltage follows the voltage at
		the Track pin on a volt-for-volt
		basis. When the control
		voltage is raised above this range, the module regulates
		at its set-point voltage. The
		features allows the output
		voltage to rise simultaneously
		with other modules powered
		from the same input bus. If
		unused, this input should be
		connected to $V_{\rm I}$.
		NOTE: Due to the
		undervoltage lockout feature,
		the output of the module
		cannot follow its own input
		voltage during power up. For
		more information, see the
		related application note.
21	Inhibit and UVLO	The Inhibit pin is an open-
		collector/drain, negative logic
		input that is referenced to
		GND. Applying a low level ground signal to this input
		disables the module's output
		and turns off the output
		voltage. When the Inhibit
		control is active, the input
		current drawn by the
		regulator is significantly
		reduced. If the Inhibit pin is
		left open-circuit, the module
		produces an output whenever
		·
		a valid input source is
		applied.
		applied. This pin is also used for input
		applied. This pin is also used for input undervoltage lockout(UVLO)
		applied. This pin is also used for input undervoltage lockout(UVLO) programming. Connecting a
		applied. This pin is also used for input undervoltage lockout(UVLO) programming. Connecting a resistor from this pin to GND
		applied. This pin is also used for input undervoltage lockout(UVLO) programming. Connecting a resistor from this pin to GND (pin 13) allows the ON
		applied. This pin is also used for input undervoltage lockout(UVLO) programming. Connecting a resistor from this pin to GND (pin 13) allows the ON threshold of the UVLO to be
		applied. This pin is also used for input undervoltage lockout(UVLO) programming. Connecting a resistor from this pin to GND (pin 13) allows the ON threshold of the UVLO to be adjusted higher than the
		applied. This pin is also used for input undervoltage lockout(UVLO) programming. Connecting a resistor from this pin to GND (pin 13) allows the ON threshold of the UVLO to be
		applied. This pin is also used for input undervoltage lockout(UVLO) programming. Connecting a resistor from this pin to GND (pin 13) allows the ON threshold of the UVLO to be adjusted higher than the default value. For more



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Mechanical Information - Surface mount version

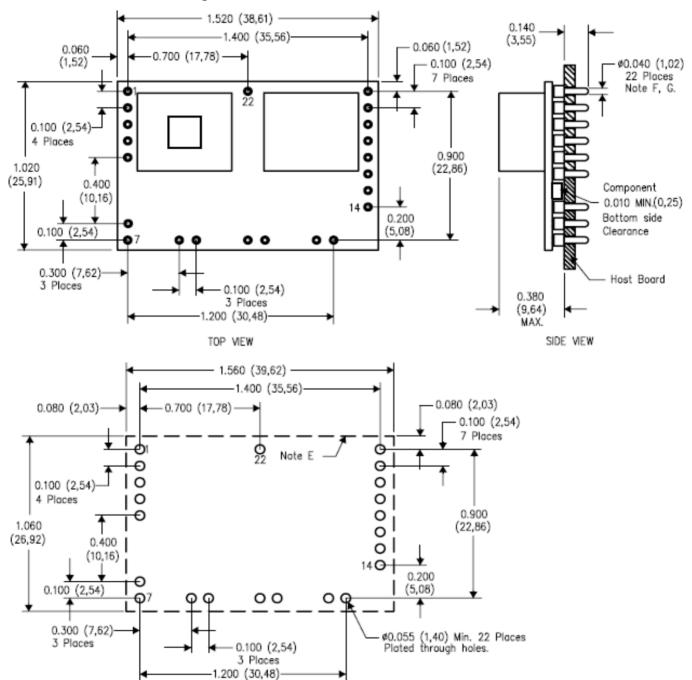


- NOTES: A. All linear dimensions are in inches (mm).
 - $\ensuremath{\mathsf{B.}}$ This drawing is subject to change without notice.
 - C. 2 place decimals are ±0.030 (±0,76mm)
 - D. 3 place deciamls are ±0.010 (±0,25mm)
 - E. Recommended keep out area for user components.
 - F. Power pin connection should utilize four or more vias to the interior power plane of 0.025 (0,63) I.D. per input, ground and output pin (or the elctrical equivalent).
- G. Paste screen opening: 0.080 (2,03) to 0.085 (2,16)
 Paste screen thickness: 0.006 (0,15)
- H. Pad type: Solder mask defined.
- l. This is a lead-free solder ball design. Finish: Tin (100%) over Nickel plating Solder ball: 96.5 Sn/3.0 Ag/0.5 Cu
- J. Dimension prior to reflow solder.



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Mechanical Information- Through hole mount version



- NOTES: A. All linear dimensions are in inches (mm).
 - B. This drawing is subject to change without notice.
 - C. 2 place decimals are ± 0.030 (± 0.76 mm).
 - D. 3 place decimals are ± 0.010 (± 0.25 mm).
 - E. Recommended keep out area for user components.
- F. Pins are 0.040" (1,02) diameter with $0.070\ensuremath{^{\circ}}$ (1,78) diameter standoff shoulder.
- G. All pins: Material Copper Alloy
 Finish Tin (100%) over Nickel Plating.



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Soldering Information - Hole Mounting

The hole mounted product is intended for plated through hole mounting by wave or manual soldering. The pin temperature is specified to maximum to 270°C for maximum 10 seconds.

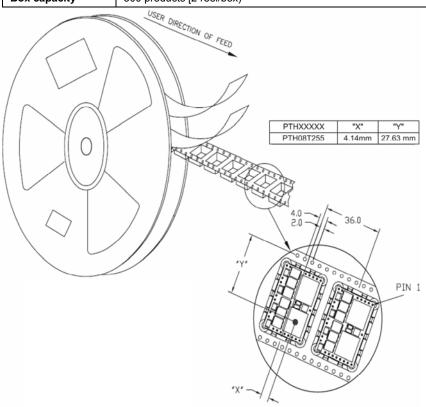
A maximum preheat rate of 4°C/s and maximum preheat temperature of 150°C is suggested. When soldering by hand, care should be taken to avoid direct contact between the hot soldering iron tip and the pins for more than a few seconds in order to prevent overheating.

A no-clean flux is recommended to avoid entrapment of cleaning fluids in cavities inside the product or between the product and the host board. The cleaning residues may affect long time reliability and isolation voltage.

Delivery Package Information

The products are delivered in tape and reel (SMD) or antistatic trays (TH & SMD)

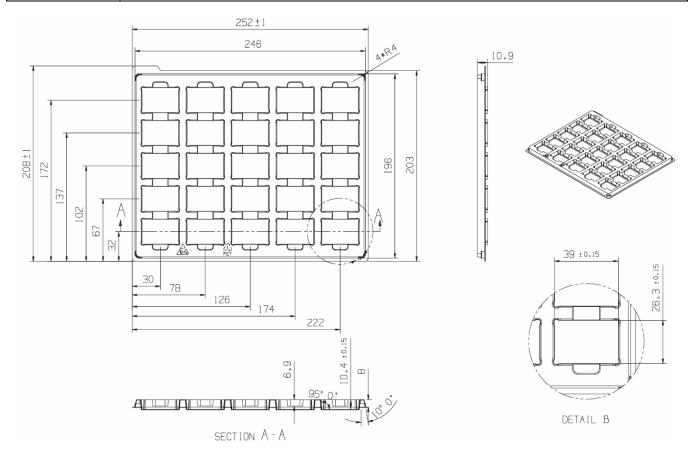
Reel Specifications		
Material	Antistatic PS	
Surface resistance	10 ⁸ < Ohm/square < 10 ¹²	
Bakeability	The reels cannot be baked	
Tape width, W	56 mm [2.205 inch]	
Pocket pitch, P ₁	36 mm [1.417 inch]	
Pocket depth, K ₀	10.4 mm [0.41 inch]	
Reel diameter	330 mm [13 inch]	
Reel capacity	150 products /reel	
Reel weight	150 g empty, 2550 g/full reel	
Carrier thickness	0.05 mm [0.002 inch]	
Box capacity	300 products [2 reel/box)	





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Tray Specifications		
Material	Antistatic PET	
Surface resistance	10 ⁸ < Ohm/square < 10 ¹²	
Bakability	The trays cannot be baked	
Tray thickness	0.8 mm [0.03 inch]	
Box capacity	125 products (5 full trays/box)	
Tray weight	40 g empty, 440 g/full tray	
Tray capacity	25 products/tray	



Dry pack information

The products are delivered in trays or tape on reel. These inner shipment containers are dry packed in standard moisture barrier bags according to IPC/JEDEC standard J-STD-033A (Handling, packing, shipping, and use of moisture/reflow sensitivity surface mount devices).

Using products in high temperature Pb-free soldering processes requires dry pack storage and handling. In case the products have been stored in an uncontrolled environment and no longer can be considered dry, the modules must be baked according to the referred IPC/JEDEC standard.







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Product Qualification Specification

Characteristics			
External visual inspection	IPC-A-610		
Change of temperature (Temperature cycling)	IEC 60068-2-14 Na	Temperature range Number of cycles Dwell/transfer time	-40 to 100°C 1000 15 min/0-1 min
Cold (in operation)	IEC 60068-2-1 Ad	Temperature T _A Duration	-45°C 72 h
Damp heat	IEC 60068-2-67 Cy	Temperature Humidity Duration	85°C 85 % RH 1000 hours
Dry heat	IEC 60068-2-2 Bd	Temperature Duration	125°C 1000 h
Electrostatic discharge susceptibility	IEC 61340-3-1, JESD 22-A114 IEC 61340-3-2, JESD 22-A115	Human body model (HBM) Machine Model (MM)	Class 2, 2000 V Class 3, 200 V
Immersion in cleaning solvents	IEC 60068-2-45 XA, method 2	Water Glycol ether Isopropyl alcohol	55°C 35°C 35°C
Mechanical shock	IEC 60068-2-27 Ea	Peak acceleration Duration	100 g 6 ms
Moisture reflow sensitivity 1	J-STD-020C	Level 1 (SnPb-eutectic) Level 3 (Pb Free)	225°C 260°C
Operational life test	MIL-STD-202G, method 108A	Duration	1000 h
Resistance to soldering heat ²	IEC 60068-2-20 Tb, method 1A	Solder temperature Duration	270°C 10-13 s
Robustness of terminations	IEC 60068-2-21 Test Ua1 IEC 60068-2-21 Test Ue1	Through hole mount products Surface mount products	All leads All leads
Solderability	IEC 60068-2-58 test Td ¹	Preconditioning Temperature, SnPb Eutectic Temperature, Pb-free	150°C dry bake 16 h 215°C 235°C
ColdClubility	IEC 60068-2-20 test Ta ²	Preconditioning Temperature, SnPb Eutectic Temperature, Pb-free	Steam ageing 235°C 245°C
Vibration, broad band random	IEC 60068-2-64 Fh, method 1	Frequency Spectral density Duration	10 to 500 Hz 0.07 g ² /Hz 10 min in each direction

Notes

¹ Only for products intended for reflow soldering (surface mount products)

² Only for products intended for wave soldering (plated through hole products)

Mouser Electronics

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