

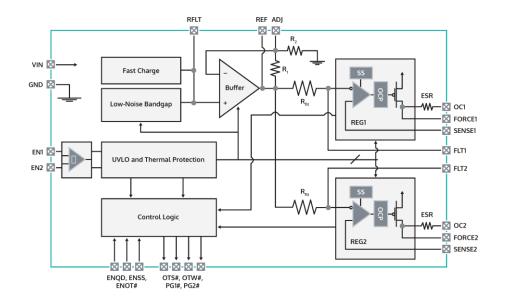
Datasheet

The **ES9311Q Dual, Ultra-Low Noise +3.3V Regulator** is the industry's highest-performance, ultra-low noise regulator optimized to supply the reference voltages for ESS' 32-Bit SABRE DACs. Audio DACs require a very low-noise reference voltage to achieve high dynamic range (DNR). The ES9311's output noise is typically under 1μVrms from 100Hz to 100kHz, and less than 1.3μVrms from 10Hz to 100kHz. The temperature coefficient of each voltage output is typically 50ppm/°C. The dual regulator replaces three active devices and several discrete components used previously to produce a high-performance, low-noise regulator. The ES9311Q enables ESS' Flagship 8-Channel DAC, the ES9038PRO, to achieve its full 140dB DNR and is an ideal low-noise regulator for use with all ESS' 32-Bit SABRE DACs.

The ES9311Q provides several unique features not available in commodity regulators: the two outputs require no decoupling capacitors, remote load sensing ensures output voltage accuracy right at the load, there are two output enable inputs, the outputs have a soft-start function, plus power-good, over-temperature warning, and over-temperature shutdown outputs.

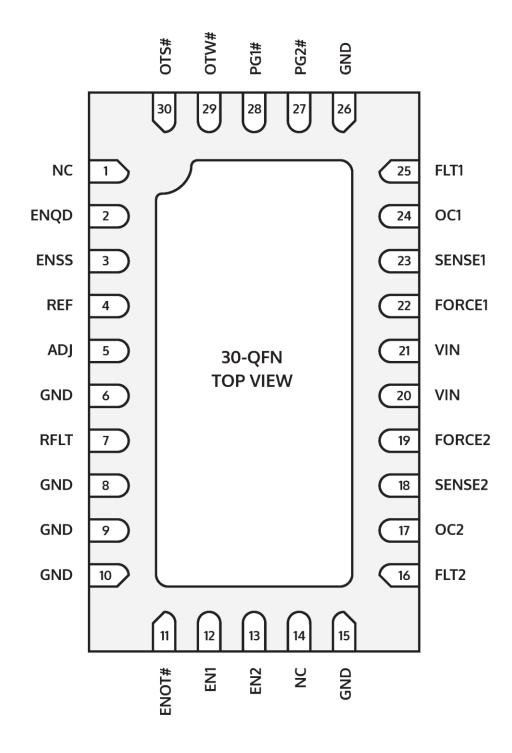
The ES9311Q Dual, Ultra-Low Noise +3.3V Regulator is available in a 30-pin, 3mm x 5mm QFN package.

FEATURE BENEFIT Dual +3.3V outputs with 300mA total load Dual regulator minimizes board space requirements < 1µVrms output noise (100Hz to 100kHz) </p> Low noise is required to achieve high converter DNR Operation from +3.5V to +5.5V Wide supply range operation 0 o Excellent line and load regulation Maintains accurate Vout with line/load variations 0 Remote sensing for each output Enables excellent load regulation, typically 16mV/A 0 Five enable inputs and four status outputs Provides design flexibility and host communications 0 Over-current & over-temperature protection Protects regulator against system fault conditions 0 3.5mA total guiescent current and 57μA typ. Suitable for USB and battery powered designs 0 standby current (both supplies disabled) o 30-QFN package, 3mm x 5mm Minimizes PCB footprint 0





PIN LAYOUT





PIN DESCRIPTIONS

| Pin | Name | Pin Type | Pin Description | |
|-----|-------------|----------|--|--|
| 1 | NC | - | No internal connection, pin may be grounded or left floating | |
| 2 | ENQD | Input | Enable for Quick Discharge function, internal pull-up | |
| 3 | ENSS | Input | Enable for Soft-Start function, internal pull-up | |
| 4 | REF | Output | Low-noise bandgap reference output voltage | |
| 5 | ADJ | Input | Input for fine adjustment of bandgap voltage output | |
| 6 | GND | Ground | Ground | |
| 7 | RFLT | Output | Connection for noise-bypass capacitor on bandgap reference | |
| 8 | GND | Ground | Ground | |
| 9 | GND | Ground | Ground | |
| 10 | GND | Ground | Ground | |
| 11 | ENOT# | Input | Enable for Over Temperature functions. Active low | |
| 12 | EN1 | Input | Enable for Output 1. Connect to VIN for automatic startup | |
| 13 | EN2 | Input | Enable for Output 2. Connect to VIN for automatic startup | |
| 14 | NC | - | No internal connection, pin may be grounded or left floating | |
| 15 | GND | Ground | Ground | |
| 16 | FLT2 | Output | Connection for noise-bypass capacitor on output 2 | |
| 17 | OC2 | Output | Optional output capacitor connection | |
| 18 | SENSE2 | Input | Output 2 Sense | |
| 19 | FORCE2 | Output | Output 2 Force | |
| 20 | VIN | Power | Regulator supply | |
| 21 | VIN | Power | Regulator supply | |
| 22 | FORCE1 | Output | Output 1 Force | |
| 23 | SENSE1 | Input | Output 1 Sense | |
| 24 | OC1 | Output | Optional output capacitor connection | |
| 25 | FLT1 | Output | Connection for noise-bypass capacitor on output 1 | |
| 26 | GND | Ground | Ground | |
| 27 | PG2# | Output | Power Good Output 2. Active low, open-drain | |
| 28 | PG1# | Output | Power Good Output 1. Active low, open-drain | |
| 29 | OTW# | Output | Over-Temperature Warning output. Active low, open-drain | |
| 30 | OTS# | Output | Over-Temperature Shutdown output. Active low, open-drain | |
| - | Exposed Pad | Ground | Connect exposed pad to Ground | |



FUNCTIONAL DESCRIPTION

The ES9311 is a dual, ultra-low noise voltage, low dropout (LDO) regulator designed for applications that require two stable 3.3V outputs at up to 150mA. Output voltage noise is typically <1µVrms from 100Hz to 100kHz and excellent power supply rejection of 90dB from 100Hz to 25kHz make this dual reference ideal for audio data converters and PLL supplies. The voltage reference design combines the advantages of both series voltage and shunt voltage references in one. That means the design has a low quiescent current, as in the series voltage reference, combined with a low output impedance that is characteristic of the shunt voltage reference. The output impedance is low over the whole output current range. In addition, the circuit is designed to operate with no output decoupling capacitor hence reducing BOM cost and PCB area. High-performance regulator features such as, soft-start, under-voltage lockout, fast transient response, over-temperature shutdown, over-current protection, and power-good indicators are included for ease-of-use and device protection.

Dual Regulator Functional Blocks

The dual LDO regulator consists of: a low-noise bandgap reference with fast charge for the filter capacitor, reference buffer amplifier, two high-performance 150mA low-dropout regulators, protection circuitry against fault conditions, plus control logic and signaling. There are five Enable inputs, one for each regulated supply, plus ones for enabling soft-start, over-temperature protection, and quick-discharge of the output voltage. All the Enable inputs include Schmitt triggers to prevent signal "chatter" when the input is noisy. There are four, active-low status outputs: power-good1, power-good2, over-temperature warning, and over-temperature shutdown. Power-good signals are activated when the input voltage is above 3.0V and when the output exceeds 90% of final value. If any overload condition occurs, e.g. over-current or over-temperature, the power-good signal will disappear. The regulator outputs use Kelvin sensing, i.e. force and sense connections, to maximize the voltage accuracy and minimize voltage perturbations right at the DAC AVCC connections.

Input Supply

The E9311 operates from a single +3.5V to +5.5V supply and provides two +3.3V outputs each capable of providing up to 150mA. Power Supply Rejection (PSR) is typically 90dB that extends beyond the audio frequency range, up to 25kHz. When powering the ES9311 from a switching power supply it is advisable to employ HF decoupling on the input to minimize noise that may mix down into the audio band. High frequency PSR can also be improved by increasing the value of the noise filtering capacitors from 1µF to 10µF and this is recommended for Audio DAC and ADC applications. The ES9311 includes an Under-Voltage Lockout (UVLO) circuit that ensures the output voltage rises gracefully without

Outputs & Remote Sensing

glitches or false starts that might introduce audible clicks or noise.

Output noise from each +3.3V supply is typically under $1\mu Vrms$ from 100Hz to 100kHz, and less than $1.3\mu Vrms$ from 10Hz to 100kHz. The temperature coefficient of each +3.3V output is typically 50ppm/°C with a drift of $\pm 0.5\%$ over temperature which corresponds to a DAC gain drift of ± 0.05 dB over temperature. Line and load regulation are typically 0.7mV/V and 16mV/A ($16m\Omega$ output resistance) respectively. Isolation between the two +3.3V outputs, which might affect crosstalk between left and right channels, is over 120dB at 1kHz so the effect is negligible.

Remote sensing maintains an accurate voltage at the load, even at high load currents, despite voltage drops in the connecting copper track. Bench power supply remote sensing uses a 4-wire connection, or Kelvin connection, that separates the current path to and from the load from the voltage signal developed across the load. The 4-wire measurement bypasses the voltage drop in the connecting cables by providing a voltage signal that is fed back to high-impedance ground and output sense inputs. Slightly different from a bench power supply, the ES9311 dual regulator relies on a low-impedance ground plane for the current return path and only uses remote sensing for the +3.3V output connections as these tracks drop the most voltage. This 3-wire scheme maximizes load regulation in conjunction with a low-impedance ground plane.

Output Noise Reduction

The ES9311 requires three $1\mu F$ noise filter capacitors, CRF, CF1, and CF2 to achieve its low-noise performance. These capacitors should be low-leakage ceramic types with low-ESR and temperature rated for operation up to the maximum internal operating temperature of the end product. Dielectric types X7R or X5R are highly recommended for these capacitors. Higher values of noise filter capacitor can be used to increase high frequency PSR, e.g. $10\mu F$, and this value and is recommended for the highest performance Audio applications. A fast-charge circuit that rapidly charges the noise bypass capacitor CRF minimizes startup time.



Input & Output Decoupling

The ES9311 requires one 4.7µF input decoupling capacitor, but NO output decoupling capacitors are required. Adding large decoupling capacitors directly to the outputs may cause instability and will degrade the performance of the regulators. A small amount of load capacitance is unavoidable, but if this exceeds 1nF then a 1µF should be connected between COUT and ground (see Figure 2). Also, if high load step-currents are anticipated, e.g. 1mA to 50mA, then 1µF capacitors should be connected between COUT1 and COUT2 and ground to improve load transient response. These capacitors are unnecessary in audio DAC and ADC applications as the load current is relatively constant, that is unless load capacitance exceeds 1nF.

Output Adjustment

The ES9311Q was designed to provide a precision 3.3 V at its two outputs. External resistors or a potentiometer may be used to adjust the output voltage to exactly 3.3 V. The regulated output voltage, however, can be changed if desired. These adjustments are optional, and are usually unnecessary for audio DAC and ADC applications.

Voltage adjustment is made using the ADJ input. To change the output voltage, one or two resistors, R_5 , R_7 , can be added as shown in Figure 12. Adding R_5 and R_7 will create parallel resistances with the internal gain setting resistors of 3.6045 k Ω and 10 k Ω respectively. The equation for setting output voltage is as follows:

$$V_{OUT} = 2.42 (1 + \left(\frac{3.6045 \text{ k}\Omega || \text{ R}_5}{10 \text{ k}\Omega || \text{ R}_7}\right))$$

While it is possible to use this adjustment to operate the ES9311Q with an output voltage below its design point of 3.3 V, the "power good" outputs are designed to activate when the output voltage exceeds a level of approximately 3.0 V, and will not indicate correctly if an output at or below that level is selected. Furthermore, the performance of the device is only specified with a nominal 3.3 V output.

Overload Protection

Each +3.3V output has over-current protection, set at ~250mA, and the regulator includes an over-temperature warning output, and an over-temperature shutdown output. The regulator outputs can be shorted to ground indefinitely but this is not recommended as the temperature cycling may reduce product longevity. Thermal-overload protection limits the total power dissipation in the ES9311Q. The over-temperature shutdown function can be disabled for high-availability systems using the ENOT# enable input. The over-temperature warning (OTW#) output is activated when the junction temperature reaches +105°C and turns off when the junction temperature drops below +100°C. The OTW output may be used to activate a system cooling-fan or notify the host processor of an impending thermal shutdown. Over-Temperature Shutdown (OTS) is activated at +130°C and is designed to protect the ES9311Q in the event of a fault condition. OTS has 15°C of hysteresis preventing the regulator from restarting automatically until the junction temperature drops below +115°C. Under continuous output fault conditions, the regulator outputs will switch ON and OFF as the device alternately cools and then reheats. Disabling the OTS function permits the junction temperature to exceed +130°C so be careful not to exceed the absolute maximum junction temperature rating of +150°C as this can impact the device's reliability.

Power Dissipation

The ES9311's maximum power dissipation depends on the thermal resistance of the case and PCB, the temperature difference between the die's junction temperature and the ambient temperature, and the airflow around the regulator. The power dissipated in the regulator is calculated from:

 $P = \text{lavcc1} \times (\text{Vin} - \text{AVCC1}) + \text{lavcc2} \times (\text{Vin} - \text{AVCC2}) + \text{Vin} \times \text{Iq}$

The maximum power dissipation is: $Pmax = (Ti - Ta) / (\Thetai-c + \Thetac-a)$

Where Tj - Ta is the temperature difference between the ES9311 junction temperature and the ambient temperature, Ojc is the thermal resistance from junction to case, and Oca is the thermal resistance from the case through the PC board, copper traces, and other materials to ambient.



ES9311Q Block Diagram

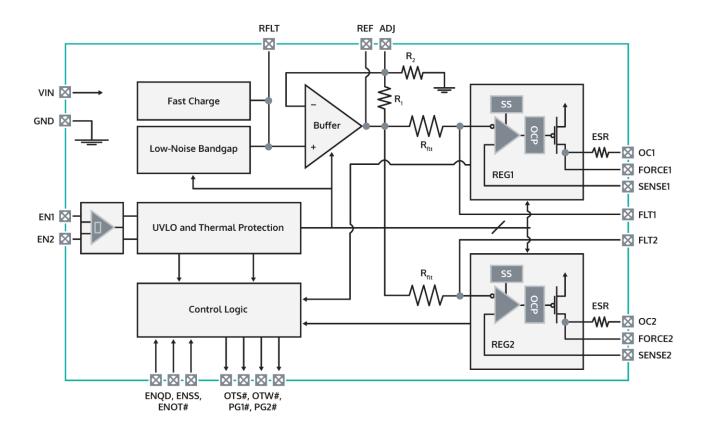


Figure 1. Block Diagram of the ES9311Q Dual, Ultra-Low Noise +3.3V Regulator.



APPLICATION DIAGRAM

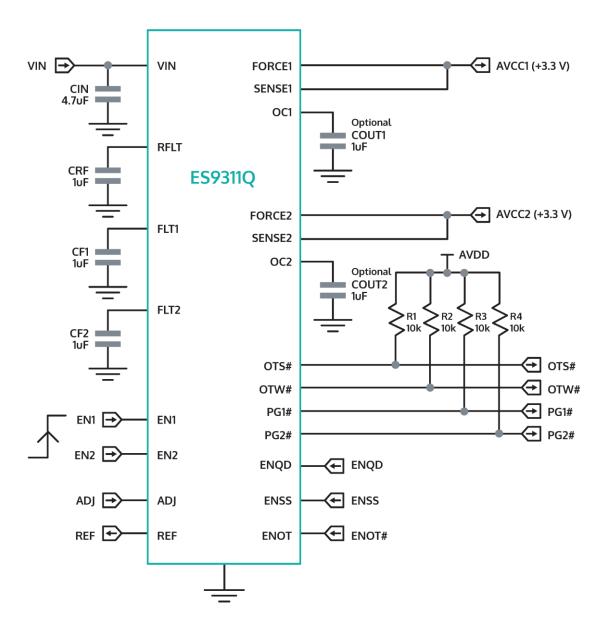


Figure 2. Simplified ES9311Q Application Circuit.



ABSOLUTE MAXIMUM RATINGS

| PARAMETER | RATING |
|--------------------------------------|---------------------------|
| Power Supply Voltage | +6.0V with respect to GND |
| Storage temperature | −65°C to +150°C |
| Operating Junction Temperature | +125°C |
| Thermal Impedance, Θ _{J-A} | 46°C/W |
| Thermal Impedance, O _{J-C} | 5°C/W |
| Voltage range for digital input pins | -0.3V to VIN + 0.3V |
| ESD Protection | |
| Human Body Model (HBM) | 2000V |
| Machine Model (MM) | 200V |

WARNING: Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute—maximum—rated conditions for extended periods may affect device reliability.

WARNING: Electrostatic Discharge (ESD) can damage this device. Proper procedures must be followed to avoid ESD when handling this device.

RECOMMENDED OPERATING CONDITIONS

| PARAMETER | SYMBOL | CONDITIONS |
|--|--------|----------------|
| Operating Temperature Range T _a | | −25°C to +85°C |

| Power Supply | Symbol | Voltage | Quiescent Current (Note 1) | Standby Current (Note 2) |
|-----------------------------|--------|-------------------|-------------------------------|--------------------------|
| Analog power supply voltage | VIN | +3.5V to +5.5V | 3.5mA typical | 57μA typical |

Notes

- 1) Quiescent current with both channels enabled, no output load, nominal +5V supply
- 2) Both +3.3V outputs are disabled

DC ELECTRICAL CHARACTERISTICS

| PARAMETER | SYMBOL | MIN | TYP | MAX | UNIT |
|--------------------------|--------|-----|-----|-----|------|
| High-level input voltage | ENH | 0.8 | | | V |
| Low-level input voltage | ENL | | | 0.4 | V |
| Hysteresis | | | 200 | | mV |



ANALOG PERFORMANCE

Test Conditions

- 1. VIN = Vout + 0.5V, EN = VIN; Iload = 10mA; CIN = $4.7\mu F$; CRF = CF1 = CF2 = $1\mu F$; Tj = Ta = $25^{\circ}C$ for typical specs
- 2. Tj = −25°C to +85°C for minimum and maximum specifications, unless otherwise noted.

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|------------------------------|---------------|--|------|-------|------|--------|
| Supply | | | | | | |
| Input Voltage Range | VIN | | 3.5 | | 5.5 | V |
| Quiescent Current | Iq | lload = 0μA. Iq is for both regulators | | 3.5 | 4 | mA |
| | | Iload = 150mA. Iq is for both regulators | | 4.5 | | mA |
| Shutdown Current | Isd | EN1 = EN2 = GND | | 57 | 100 | μΑ |
| Power Supply Rejection | PSR | 100Hz to 25kHz, lout = 1mA to 150mA | | 90 | | dB |
| Noise | | | | | | |
| Output Noise | en | 100Hz to 100kHz | | 0.725 | | μVrms |
| | en | 10Hz to 100kHz | | 1.3 | | μVrms |
| Noise Spectral Density | | 100Hz to 100kHz | | | | nV/√Hz |
| Output | | | | | | |
| Output Voltage | Vout | Iload = 10 mA, Tj = +25°C | 3.27 | 3.30 | 3.33 | V |
| Temperature coefficient | VoutTC | Tj = −25°C to +85°C | | 50 | | ppm/°C |
| Load Current | lload | Maximum output for each supply | 0 | | 150 | mA |
| Short-circuit current | Isc | Current for each supply | | 270 | | mA |
| Start-up Time from Enable | Tsu | Soft-Start enabled | | | | ms |
| Start-up Time from Enable | Tsu | Soft-Start disabled | | | | ms |
| Shut-down Time for Vout | Tsd | Cload = 1µF, Quick-Discharge disabled | | | | ms |
| Shut-down Time for Vout | Tsd | Cload = 1µF, Quick-Discharge enabled | | | | ms |
| Regulation | | | | | | |
| Line Regulation | ΔVout / ΔVIN | VIN = Vout + 0.5V to 5.5V | | 0.7 | | mV/V |
| Load Regulation | ΔVout / Δlout | lout = 1mA to 150mA | | 16 | | mV/A |
| Dropout Voltage | Vdropout | lout = 10mA | | 50 | | mV |
| | Vdropout | lout = 150mA | | | | mV |
| Thermal Protection | | | | | | |
| Over-temperature Warning | OTW | Over-Temperature Enabled, Tj rising | | 105 | | °C |
| Hysteresis | | Over-Temperature Enabled | | 5 | | °C |
| Over-temperature Shutdown | OTS | Over-Temperature Enabled, Tj rising | | 130 | | °C |
| Hysteresis | | Over-Temperature Enabled | | 15 | | °C |
| Under-Voltage Lock-Out | | | | | | |
| VIN Rising | UVLO | | | 3.00 | | V |
| VIN Falling | | | | | | V |
| Hysteresis UVLOhys | | | | | | mV |



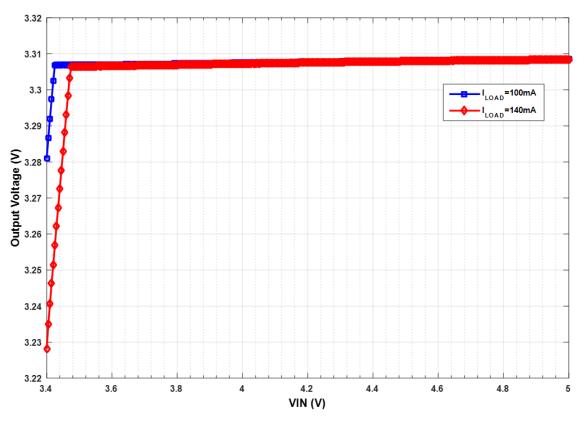


Figure 3. Output Voltage vs. Input Voltage (Dropout)

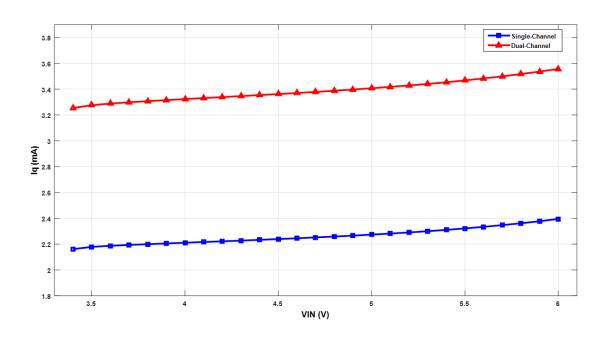




Figure 4. Quiescent Current vs. Input Voltage

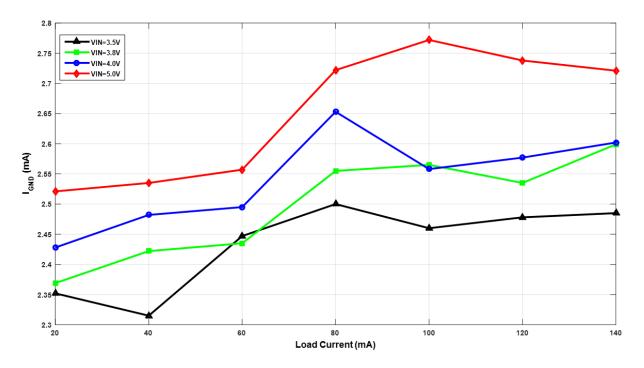


Figure 5. GND Pin Current (single channel) vs. Load Current

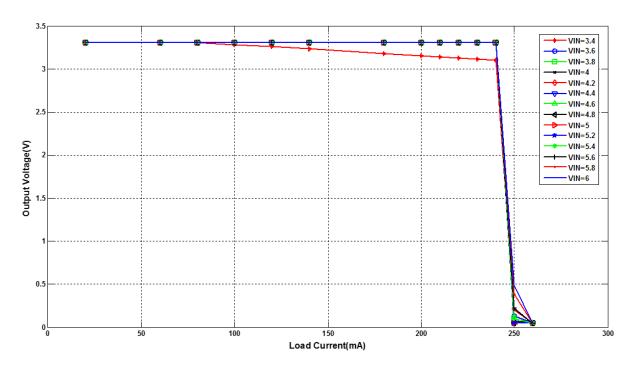


Figure 6. Output Voltage vs. Load Current (Current Limit)



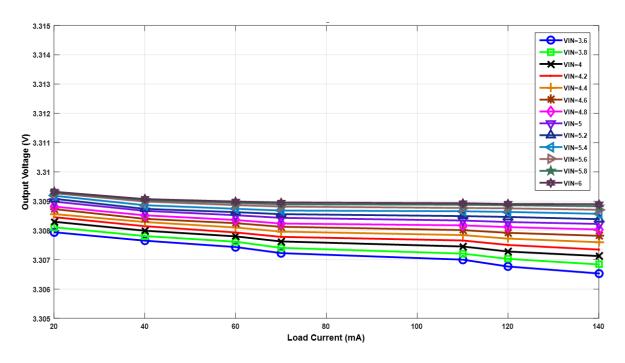


Figure 7. Output Voltage vs. Load Current (Load Regulation)

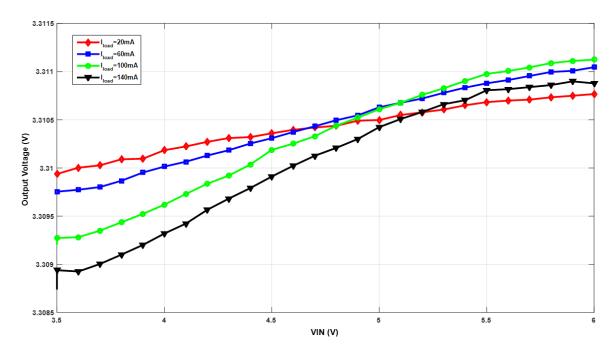


Figure 8. Output Voltage vs. Input Voltage (Line Regulation)



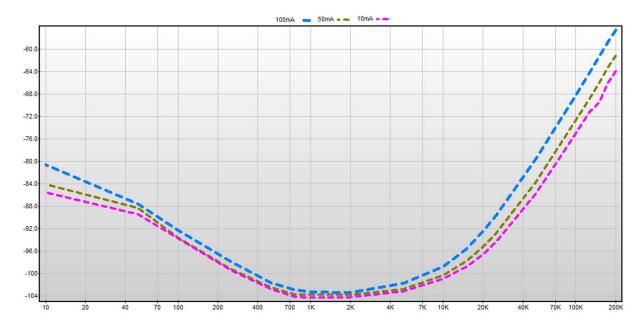


Figure 9. Power Supply Rejection (dB) vs. Frequency (Hz) with VIN = 3.8V, 10µF Filter Caps

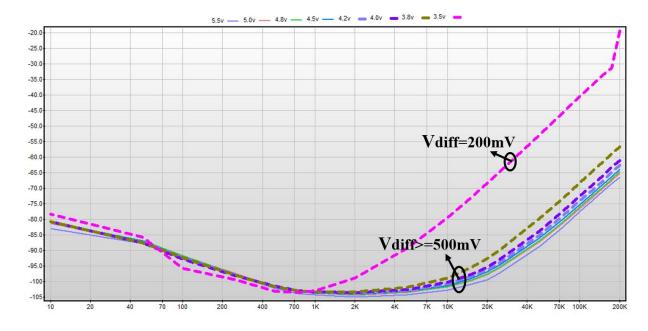


Figure 10. Power Supply Rejection (dB) vs. Frequency (Hz) with Iload = 100mA, 10µF Filter Caps



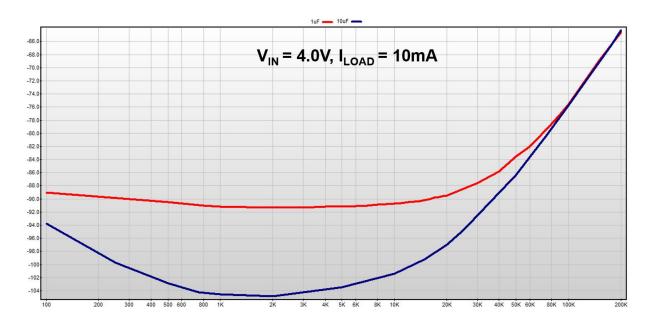


Figure 11. Power Supply Rejection (dB) vs. Frequency (Hz) with 1μF and 10μF filter Capacitors



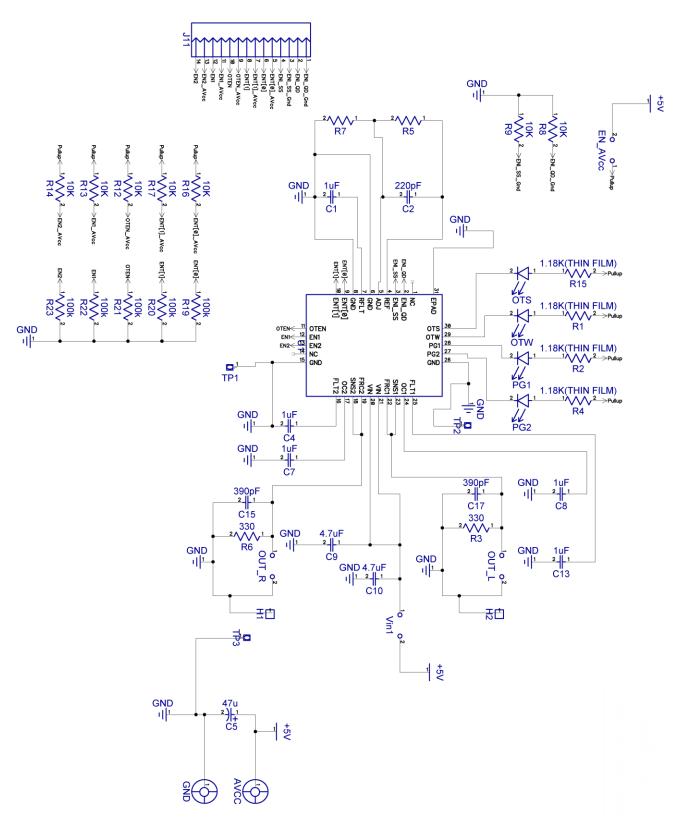


Figure 12. ES9311Q Evaluation Board Circuit



Marking Specification

Note: Marking diagram is subject to change.

Figure 13 shows the markings of the ES9113Q package. This drawing is not to scale.

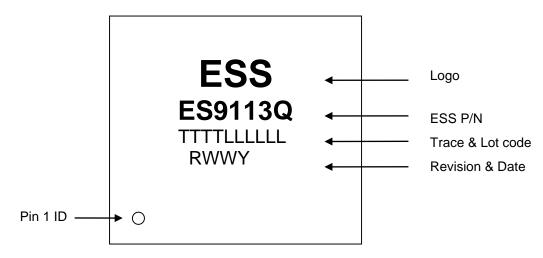
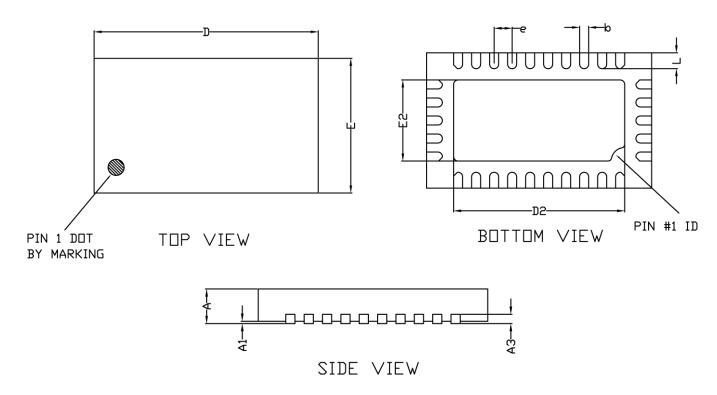


Figure 13. ES9311Q Marking Diagram



30-Pin QFN Mechanical Dimensions



| | Common Dimensions (mm) | | | |
|-----------|------------------------|---------|---------|--|
| Package | Very Very Thin | | | |
| Reference | Minimum | Nominal | Maximum | |
| Α | 0.70 | 0.75 | 0.80 | |
| A1 | 0.00 | - | 0.05 | |
| А3 | 0.2 REF | | | |
| D | 4.95 5.00 5.05 | | | |
| E | 2.95 | 3.00 | 3.05 | |
| b | 0.15 | 0.20 | 0.25 | |
| L | 0.25 | 0.35 | 0.45 | |
| D2 | 3.75 | 3.80 | 3.85 | |
| E2 | 1.75 | 1.80 | 1.85 | |
| е | 0.4 BSC | | | |

Table 1. Package Dimensions



Reflow Process Considerations

For lead-free soldering, the characterization and optimization of the reflow process is the most important factor you need to consider.

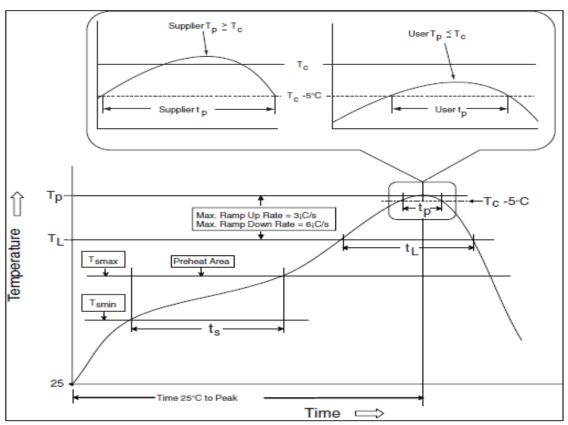
The lead-free alloy solder has a melting point of 217°C. This alloy requires a minimum reflow temperature of 235°C to ensure good wetting. The maximum reflow temperature is in the 245°C to 260°C range, depending on the package size (*Table RPC-2*). This narrows the process window for lead-free soldering to 10°C to 20°C.

The increase in peak reflow temperature in combination with the narrow process window makes the development of an optimal reflow profile a critical factor for ensuring a successful lead-free assembly process. The major factors contributing to the development of an optimal thermal profile are the size and weight of the assembly, the density of the components, the mix of large and small components, and the paste chemistry being used.

Reflow profiling needs to be performed by attaching calibrated thermocouples well adhered to the device as well as other critical locations on the board to ensure that all components are heated to temperatures above the minimum reflow temperatures and that smaller components do not exceed the maximum temperature limits (*Table RPC-2*).

To ensure that all packages can be successfully and reliably assembled, the reflow profiles studied and recommended by ESS are based on the JEDEC/IPC standard J-STD-020 revision D.1.

Figure RPC-1. IR/Convection Reflow Profile (IPC/JEDEC J-STD-020D.1)



Note: Reflow is allowed 3 times. Caution must be taken to ensure time between re-flow runs does not exceed the allowed time by the moisture sensitivity label. If the time elapsed between the re-flows exceeds the moisture sensitivity time bake the board according to the moisture sensitivity label instructions.

Manual Soldering:

Allowed up to 2 times with maximum temperature of 350 degrees no longer than 3 seconds.



Table RPC-1 Classification reflow profile

| Profile Feature | Pb-Free Assembly | | |
|--|--|--|--|
| Preheat/Soak | | | |
| Temperature Min (Tsmin) | 150°C | | |
| Temperature Max (Tsmax) | 200°C | | |
| Time (ts) from (Tsmin to Tsmax) | 60-120 seconds | | |
| Ramp-up rate (TL to Tp) | 3°C / second max. | | |
| Liquidous temperature (TL) | 217°C | | |
| Time (tL) maintained above TL | 60-150 seconds | | |
| Peak package body temperature (Tp) | For users Tp must not exceed the classification temp in Table RPC-2. For suppliers Tp must equal or exceed the Classification temp in Table RPC-2. | | |
| Time (tp)* within 5°C of the specified classification temperature (Tc), see Figure RPC-1 | 30* seconds | | |
| Ramp-down rate (Tp to TL) | 6°C / second max. | | |
| Time 25°C to peak temperature | 8 minutes max. | | |
| * Tolerance for peak profile temperature (Tp) is defined as a supplier minimum and a user maximum. | | | |

Note 1: All temperatures refer to the center of the package, measured on the package body surface that is facing up during assembly reflow (e.g., live-bug). If parts are reflowed in other than the normal live-bug assembly reflow orientation (i.e., dead-bug), Tp shall be within ±2°C of the live-bug Tp and still meet the Tc requirements, otherwise, the profile shall be adjusted to achieve the latter. To accurately measure actual peak package body temperatures, refer to JEP140 for recommended thermocouple use.

Note 2: Reflow profiles in this document are for classification/preconditioning and are not meant to specify board assembly profiles. Actual board assembly profiles should be developed based on specific process needs and board designs and should not exceed the parameters in Table RPC-1. For example, if Tc is 260°C and time tp is 30 seconds, this means the following for the supplier and the user.

For a supplier: The peak temperature must be at least 260°C. The time above 255°C must be at least 30 seconds.

For a user: The peak temperature must not exceed 260°C. The time above 255°C must not exceed 30 seconds.

Note 3: All components in the test load shall meet the classification profile requirements.

Table RPC-2 Pb-Free Process – Classification Temperatures (Tc)

| Package Thickness | Volume mm3, <350 | Volume mm3, 350 to 2000 | Volume mm3, >2000 |
|-------------------|------------------|-------------------------|-------------------|
| <1.6 mm | 260°C | 260°C | 260°C |
| 1.6 mm – 2.5 mm | 260°C | 250°C | 245°C |
| >2.5 mm | 250°C | 245°C | 245°C |

Note 1: At the discretion of the device manufacturer, but not the board assembler/user, the maximum peak package body temperature (Tp) can exceed the values specified in Table RPC-2. The use of a higher Tp does not change the classification temperature (Tc).

Note 2: Package volume excludes external terminals (e.g., balls, bumps, lands, leads) and/or non-integral heat sinks.

Note 3: The maximum component temperature reached during reflow depends on package thickness and volume. The use of convection reflow processes reduces the thermal gradients between packages. However, thermal gradients due to differences in thermal mass of SMD packages may still exist.



ORDERING INFORMATION

| Part Number | Description | Package |
|-------------|---------------------------------------|------------|
| ES9311Q | Dual, Ultra-Low Noise +3.3V Regulator | 30-pin QFN |

The letter Q identifies the package type QFN.

Revision History

| Rev. | Date | Notes | |
|------|-------------------|---|--|
| 0.1 | December 15, 2015 | Initial release | |
| 0.2 | January 4, 2016 | Added characterization graphs | |
| 0.3 | February 16, 2016 | Updated values of shutdown current and quiescent current | |
| 0.31 | March 1, 2016 | Corrected package dimension reference in Table 1 | |
| 0.4 | March 14, 2016 | Added connection information for exposed pad. Package thermal impedance values added. Expanded section on input and output decoupling | |
| 0.51 | May 31, 2016 | Added information about voltage adjustment using external resistors, "Patent pending", and corrected pin count on package drawing. Added package marking diagram. | |

Patent pending.

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