

The **ES9016K2M SABRE**<sup>32®</sup> **Ultra DAC** is a high-performance 32-bit, 2-channel audio D/A converter targeted for portable high-fidelity audio power sensitive applications such as digital music players, consumer applications such as Blu-ray players, audio pre-amplifiers and A/V receivers, as well as professional applications such as recording systems, mixer consoles and digital audio workstations.

Using the critically acclaimed ESS patented 32-bit HyperStream® DAC architecture and Time Domain Jitter Eliminator, the *ES9016K2M SABRE*<sup>32®</sup> *Ultra DAC* delivers a DNR of up to 122dB and THD+N of –110dB, a performance level that will satisfy the most demanding audio enthusiasts.

The **ES9016K2M SABRE**<sup>32®</sup> **Ultra DAC**'s 32-bit HyperStream® architecture can handle up to 32-bit, 384kHz PCM data via I<sup>2</sup>S, DSD-11.2MHz data as well as mono mode for highest performance applications. Both synchronous and ASRC (asynchronous sample rate conversion) modes are supported.

The **ES9016K2M SABRE® Ultra DAC** is powered by a +1.8V to +3.3V supply for both the digital and analog sections, with internal regulators generating the core supply. The DAC comes in a 28-QFN package, supports 1.8V logic levels and consumes less than 40mW in normal operating mode (< 1mW in standby mode)

The **SABRE**® **Ultra DAC** sets a new standard for high quality audio performance, **SABRE SOUND**®, in a cost effective, easy-to-use form factor for today's most demanding digital audio applications.

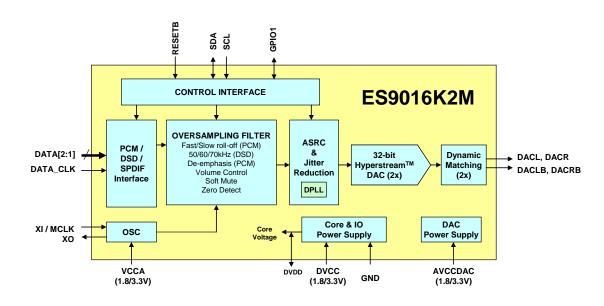
FEATURE		DESCRIPTION
Patented 32-bit HyperStream® DAC    +122dB DNR   -110dB THD+N	0	32-bit audio DAC powered by Sabre <sup>32®</sup> DAC architecture with unmatched dynamic range and ultra low distortion Supports both synchronous and ASRC (asynchronous sample rate converter) modes
Patented Time Domain Jitter Eliminator	0	Unmatched audio clarity free from input clock jitter
64-bit accumulator and 32-bit processing	0	Distortion free signal processing
Integrated DSP Functions	0 0	Click-free soft mute and volume control Programmable Zero detect De-emphasis for 32kHz, 44.1kHz, and 48kHz sampling
Customizable output configuration	0	Mono or stereo output in current or voltage mode based on performance criterion
I <sup>2</sup> C control	0	Allows software control of DAC features
28-QFN (5mm x 5mm) package	0	Minimizes PCB footprint
< 40mW operating, < 1mW standby power	0	Maximizes battery life
1.8 to 3.3V analog & digital power supplies	0	Reduces power and simplifies power supply design
1.8V digital logic supported	0	Connects to Application Processor without level shifter
Versatile digital input	0	Supports SPDIF, PCM (I2S, LJ 16-32-bit) or DSD input
Customizable filter characteristics	0	User programmable filter allowing custom roll-off response Bypassable oversampling filter

## **APPLICATIONS**

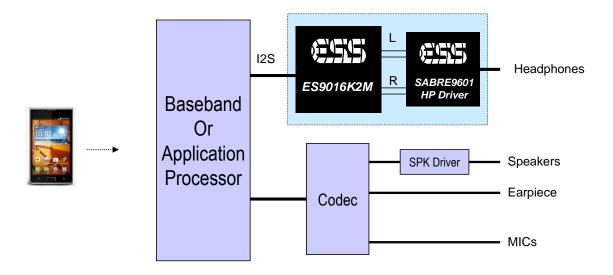
- Mobile phones / Tablets / Digital music players / Portable multimedia players
- Blu-ray / SACD / DVD-Audio player
- Audio preamplifier and A/V receiver
- Professional audio recording systems / Mixing consoles / Digital audio workstation



## **FUNCTIONAL BLOCK DIAGRAM**

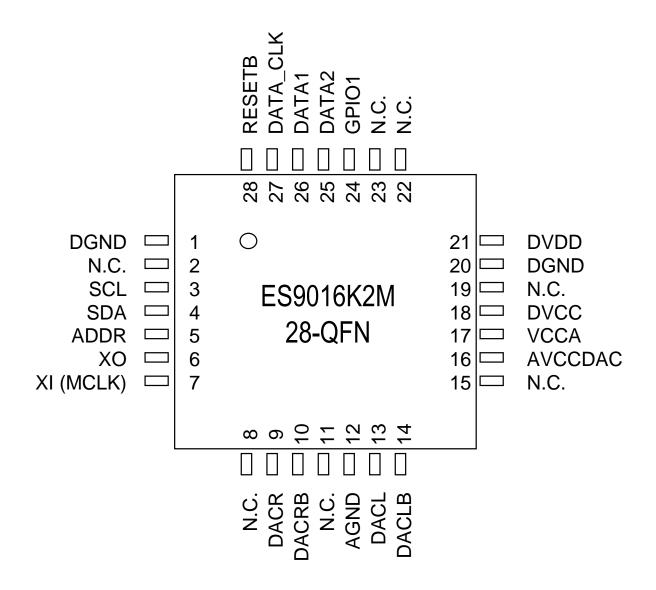


## TYPICAL APPLICATION DIAGRAM





# **PIN LAYOUT**



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# **ES9016K2M Datasheet**

# **PIN DESCRIPTIONS**

Pin	Name	Pin Type	Reset State	Pin Description		
1	DGND	Ground	Ground	Digital Ground		
2	N.C.	-	-	No internal connection. Pin may be grounded if desired.		
3	SCL	I	Tri-stated	I <sup>2</sup> C Serial Clock Input		
4	SDA	I/O	Tri-stated	I <sup>2</sup> C Serial Data Input/Output		
5	ADDR	ı	Tri-stated	I <sup>2</sup> C Address Select		
6	XO	AO	Floating	XTAL Out		
7	XI (MCLK)	Al	Floating	XTAL / MCLK In		
8	N.C.	-	-	No internal connection. Pin may be grounded if desired.		
9	DACR	AO	Driven to ground	Differential Positive Analog Output Right		
10	DACRB	AO	Driven to ground	Differential Negative Analog Output Right		
11	N.C.	-	-	No internal connection. Pin may be grounded if desired.		
12	AGND	Ground	Ground	Analog Ground		
13	DACL	AO	Driven to ground	Differential Positive Analog Output Left		
14	DACLB	AO	Driven to ground	Differential Negative Analog Output Left		
15	N.C.	-	-	No internal connection. Pin may be grounded if desired.		
16	AVCCDAC	Power	Power	Analog AVCC for DAC		
17	VCCA	Power	Power	Analog +1.8V or +3.3V for OSC		
18	DVCC	Power	Power	Digital +1.8V to +3.3V		
19	N.C.	-	-	No internal connection. Pin may be grounded if desired.		
20	DGND	Ground	Ground	Digital Ground		
21	DVDD	Power	Power	Digital Core Voltage, nominally +1.2V, is supplied by a regulator from DVCC. DVDD must be decoupled with a minimum $4.7\mu F$ capacitor to DGND for stable operation. DVDD needs to be externally supplied for high XI / MCLK frequency. Please refer to the section about the DVDD supply on page 7 for additional information.		
22	N.C.	-	-	No internal connection. Pin may be grounded if desired.		
23	N.C.	-	-	No internal connection. Pin may be grounded if desired.		
24	GPIO1	I/O	Tri-stated	GPIO 1		
25	DATA2	ı	Tri-stated	DSD Data2 (R) or PCM Data CH1/CH2 or SPDIF Input 2		
26	DATA1	I/O	Tri-stated	Master mode off: Input for DSD Data1 (L) or PCM Frame Clock or SPDIF Input 3 Master mode on: Output for PCM Frame Clock		
27	DATA_CLK	I/O	Tri-stated	Master mode off: Input for PCM Bit Clock or DSD Bit Clock or SPDIF Input 1 Master mode on: Output for PCM Bit Clock		
28	RESETB	I	Tri-stated	Master Reset / Power Down (active low)		
Exposed Pad	DGND	Ground	Ground	Digital Ground. Connect the Exposed Pad to DGND		

#### Notes:

- There are 7 N.C. (No Connect) pins. If desired, these pins can be connected to ground on the PCB to strengthen the otherwise isolated pin pads.
- The exposed pad must be connected to digital ground.



#### **FUNCTIONAL DESCRIPTION**

#### **NOTATATIONS for Sampling Rates**

Mode	fs (target sample rate)	FSR (raw sample rate)		
DSD	DATA_CLK / 64	DSD data rate		
Serial (PCM) Normal Mode	Frame Clock Rate	Frame Clock Rate		
Serial (PCM) OSF Bypass Mode	Frame Clock Rate / 8	Frame Clock Rate		
SPDIF	SPDIF Sampling Rate	SPDIF Sampling Rate		

## PCM, SPDIF and DSD Pin Connections

#### **PCM Audio Format**

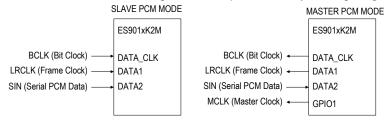
Notes:

XI clock (MCLK) must be > 192 x FSR when using PCM input (normal mode), or 128 x FSR (synchronous MCLK). XI clock (MCLK) must be > 24 x FSR when using PCM input (OSF bypass mode).

Pin Name	Description
DATA1	Frame clock
DATA2	2-channel PCM serial data
DATA_CLK	Bit clock for PCM audio format

#### Master Mode (32-bit data only)

When Register #1 'input\_select' is set to 2'd0 (I2S) and 'i2s\_length' is set to 2'd2 (32-bit), the DAC can become a master for Bit Clock and Frame Clock by setting Register #9 'master clock enable' to 1'b1. The Bit Clock frequency can be configured to MCLK / 4, MCLK / 8 or MCLK / 16 by setting Register #9 'clock divider select' to 2'b00, 2'b01 or 2'b10. GPIO 1 can be configured to output MCLK by setting Register #8 gpio1 cfg to 4'd3.



#### **SPDIF Audio Formant**

Note: XI clock (MCLK) must be > 386 x FSR when using SPDIF input.

Up to four SPDIF inputs can be connected to the 4-to-1 mux, selectable via register "spdif\_sel". SPDIF can also be sourced from a GPIO pin configured as input.

Pin Name	Description
GPIO1	SPDIF input 4
DATA1	SPDIF input 3
DATA2	SPDIF input 2
DATA_CLK	SPDIF input 1

#### **DSD Audio Format**

Note: XI clock (MCLK) must be > 3 x FSR when using DSD input.

Pin Name	Description
DATA[1:2]	2-channel DSD data input
DATA_CLK	Bit clock for DSD data input

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#### ES9016K2M Datasheet

#### FEATURE DESCRIPTION

#### Soft Mute

When Mute is asserted the output signal will ramp to the  $-\infty$  level. When Mute is reset the attenuation level will ramp back up to the previous level set by the volume control register. Asserting Mute will not change the value of the volume control register. The ramp rate is  $0.0078125 \times fs / 2^{(vol\_rate-5)}$  dB/s.

#### **Automute**

During an automute condition the ramping of the volume of each DAC to -∞ can now be programmatically enabled or disabled.

- o In PCM serial mode, "AUTOMUTE" will become active once the audio data is continuously below the threshold set by <Register Automute\_lev>, for a length of time defined by 2096896 / (<Register#4> x 64 x fs) seconds.
- o In SPDIF mode, "AUTOMUTE" will become active once the audio data is continuously below the threshold set by <Register Automute\_lev>, for a length of time defined by 2096896 / (<Register#4> x (64 x fs) seconds.
- o In the DSD Mode, "AUTOMUTE" will become active when any 8 consecutive values in the DSD stream have as many 1's and 0's for a length of time defined by 2096896 / (<Register Automute\_time> x DATA\_CLK) seconds. The following table summarizes the conditions.

Mode	<b>Detection Condition</b>	Time
PCM	Data is continuously lower than <register automute_lev=""></register>	2096896 / ( <register automute_time=""> x 64 x fs)</register>
SPDIF	Data is continuously lower than <register automute_lev=""></register>	2096896 / ( <register automute_time=""> x (64 x fs))</register>
DSD	Equal number of 1s and 0s in every 8 bits of data	2096896 / ( <register automute_time=""> x DATA_CLK)</register>

#### **Volume Control**

Each output channel has its own attenuation circuit. The attenuation for each channel is controlled independently. Each channel can be attenuated from 0dB to -127dB in 0.5dB steps.

Each 0.5dB step transition takes up to 64 intermediate levels, depending on the vol\_rate register setting. The result being that the level changes are done using small enough steps so that no switching noise occurs during the transition of the volume control. When a new volume level is set, the attenuation circuit will ramp softly to the new level.

#### **Master Trim**

The master trim sets the 0dB reference level for the volume control of each DAC. The master trim is programmable via registers 17-20 and is a 32bit signed number. Therefore it should never exceed 32'h7FFFFFFF (as this is full-scale signed).

#### All Mono Mode

An all mono mode where all DACs are driven from the same source is supported. This can be useful for high-end audio applications. The source data for all DACs can be programmatically configured to be either CH1 or CH2.

#### De-emphasis

The de-emphasis feature is included for audio data that has utilized the  $50/15\mu s$  pre-emphasis for noise reduction. There are three de-emphasis filters, one for 32kHz, one for 44.1kHz, and one for 48kHz.

#### **SPDIF Data Select**

An SPDIF source multiplexer allows for up to four SPDIF sources to be connected to the data pins. An internal programmable register (spdif\_sel) is used to select the appropriate data or GPIO pin to decode. SPDIF can also be sourced from GPIO pin configured as input.

#### ES9016K2M Datasheet



#### System Clock (XI / MCLK)

A system clock is required for proper operation of the digital filters and modulation circuitry. See p.28, Note 2 for the maximum MCLK frequencies supported. The minimum system clock frequency must also satisfy:

Data Type	Minimum MCLK Frequency	Note
DSD Data	MCLK > 3 x FSR , FSR = 2.8224MHz (x 1, 2 or 4)	The maximum FSR
Serial Normal Mode	MCLK > 192 x FSR, FSR $\le$ 384kHz or MCLK = 128 x FSR (synchronous MCLK) with FSR $\le$ 384kHz	frequency is further limited by the maximum MCLK
Serial OSF Bypass Mode	MCLK > 24 x FSR, FSR ≤ 1.536MHz	frequencies supported as shown
SPDIF Data	MCLK > 386 x FSR, FSR ≤ 200kHz	p.28, Note 2.

#### **Data Clock**

DATA\_CLOCK must be (2 x i2s\_length) x FSR for SERIAL, and FSR for DSD modes. For SPDIF mode, this pin is used for SPDIF input. This pin should be pulled low if not used.

#### **Built-in Digital Filters**

Three digital filters (fast roll-off, slow roll-off and minimum phase filters) are included for PCM data. See 'PCM Filter Characteristics' for more information.

#### Standby Mode

For lowest power consumption, the following sequence should be performed to enter stand-by mode:

- Set the soft start bit in register 14 to 1'b0 to ramp the DAC outputs (DACL, DACLB, DACR, DACRB) to ground.
- RESETB pin should be brought to low digital level to:
  - Shut off the DACs, Oscillator and internal regulator.
  - Force digital I/O pins (DATA\_CLK, DATA1, GPIO1, SDA) into tri-state mode
  - o Reset all registers to default states
- If XI/MCLK is supplied externally, it should be stopped at logic low level
- If DVDD is supplied by an external regulator, it should be shutdown during standby

To resume from standby mode, bring RESETB to high digital level and reinitialize all registers.

#### **DVDD Supply**

The ES9016K2M is equipped with an internal, regulated DVDD supply powered from DVCC. The internal DVDD regulator must be decoupled to DGND with a  $4.7\mu F$  minimum capacitor for stable operation. Recommended capacitor for decoupling DVDD is a  $4.7\mu F$  ±20%, X5R 6.3V 0402, e.g. TDK part number C1005X5R0J475M050BC or similar.

- The internal DVDD should be used except under the following conditions:
  - 1. PCM (SPDIF, I2S with OSF Bypass off or on): MCLK > 50MHz or FSR > 192kHz
  - 2. DSD: MCLK > 50MHz or FSR > 11.2MHz
- Internal DVDD may be used up to the maximum supported MCLK frequencies specified on p.28, Note 2. An External DVDD (+1.3V) supply must be used above those frequencies. The external supply voltage must be greater than the internal supply of +1.2V so the internal supply is disabled.

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#### ES9016K2M Datasheet

#### Programmable FIR filter

A two stage interpolating FIR design is used. The interpolating FIR filter is generated using MATLAB, and can then be downloaded using a custom C code.

#### Example Source Code for Loading a Filter

```
// only accept 128 or 16 coefficients
// Note: The coefficients must be quantized to 24 bits for this method!
// Note: Stage 1 consists of 128 values (0-127 being the coefficients)
// Note: Stage 2 consists of 16 values (0-13 being the coefficients, 14-15 are zeros)
// Note: Stage 2 is symmetric about coefficient 13. See the example filters for more information.
byte reg26 = (byte)(coeffs.Count == 128 ? 0 : 128);
for (int i = 0; i < coeffs.Count; i++)</pre>
    // stage 1 contains 128 coefficients, while stage 2 contains 16 coefficients
   registers.WriteRegister(26, (byte)(reg26 + i));
   // write the coefficient data
   registers.WriteRegister(27, (byte)(coeffs[i] & 0xff));
   registers.WriteRegister(28, (byte)((coeffs[i] >> 8) & 0xff));
   registers.WriteRegister(29, (byte)((coeffs[i] >> 16) & 0xff));
   registers.WriteRegister(30, 0x02); // set the write enable bit
// disable the write enable bit when we're done
registers.WriteRegister(30, (byte)(setEvenBit ? 0x04 : 0x00));
```

#### **OSF Bypass**

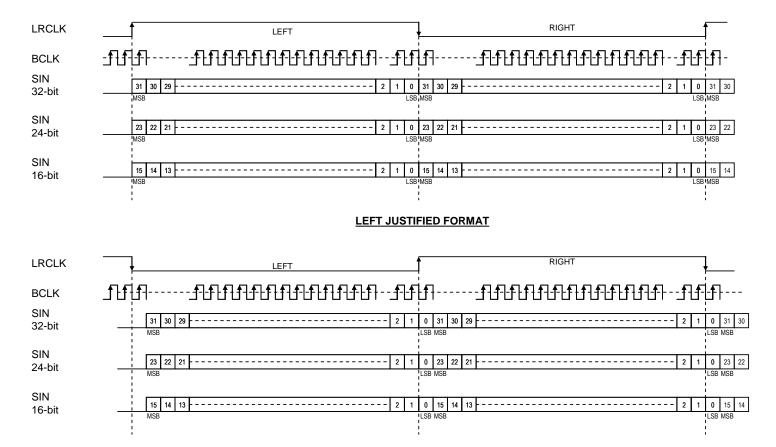
The oversampling FIR filter can be bypassed, sourcing data directly into the IIR filter. ESS recommends using 8 x FSR as the input. For example, an external signal at 44.1 kHz can be oversampled externally to 8 x 44.1 kHz = 352.8 kHz and then applied to the serial decoder in either I<sup>2</sup>S or LJ format. The maximum sample rate that can be applied is 1.536 MHz (8 x 192 kHz).

#### ES9016K2M Datasheet



## **Audio Interface Formats**

Several interface formats are provided so that direct connection to common audio processors is possible. The available formats and their accompanying diagrams are listed in the following table. The audio interface format can be set by programming the registers.

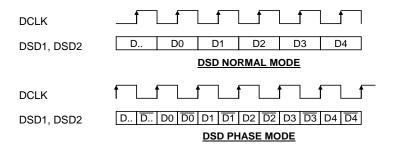


**I2S FORMAT** 

**Note:** for Left-Justified and I<sup>2</sup>S formats, the following number of BCLKs is present per LRCLK frame (left plus right channels):

16-bit mode: 32 BCLKs24-bit mode: 48 BCLKs

32-bit mode: 64 BCLKs

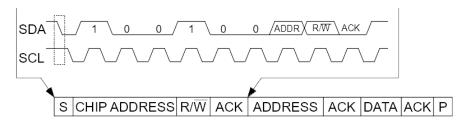




## **SERIAL CONTROL INTERFACE**

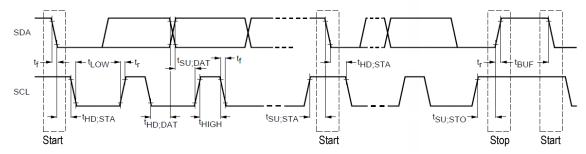
The registers inside the chip are programmed via an I<sup>2</sup>C interface. The diagram below shows the timing for this interface. The chip address can be set to 2 different settings via the "ADDR" pin. The table below summarizes this.

ADDR	CHIP ADDRESS
0	0x90
1	0x92



#### Notes:

- 1. The "ADDR" pin is used to create the CHIP ADDRESS. (0x90, 0x92)
- 2. The first byte after the chip address is the "ADDRESS" this is the register address.
- 3. The second byte after the CHIP ADDRESS is the "DATA" this is the data to be programmed into the register at the previous "ADDRESS".



Parameter	Symbol	Standa	rd-Mode	Fast-	Unit	
Parameter	Symbol	MIN	MAX	MIN	MAX	Offic
SCL Clock Frequency	fscL	0	100	0	400	kHz
START condition hold time	t <sub>HD,STA</sub>	4.0	-	0.6	-	μS
LOW period of SCL	t <sub>LOW</sub>	4.7	1	1.3	-	μS
HIGH period of SCL	tніgн	4.0	-	0.6	-	μS
START condition setup time (repeat)	tsu,sta	4.7	-	0.6	-	μS
SDA hold time from SCL falling	t <sub>HD,DAT</sub>	0.3	-	0.3	-	μS
SDA setup time from SCL rising	t <sub>SU,DAT</sub>	250	1	100	-	ns
Rise time of SDA and SCL	t <sub>r</sub>	-	1000		300	ns
Fall time of SDA and SCL	t <sub>f</sub>	-	300		300	ns
STOP condition setup time	t <sub>SU,STO</sub>	4	-	0.6	-	μS
Bus free-time between transmissions	t <sub>BUF</sub>	4.7	-	1.3	-	μS
Capacitive load for each bus line	Сь	-	400	-	400	pF

# **ES9016K2M Datasheet**



# **REGISTER MAP**

	Address	Register	D7 (MSB)	D6	D5	D4	D3	D2	D1	D0 (LSB)		
0 0 0	(Dec/Hex)		` '									
1/0x01		SVSTEM SETTINGS		09C D	OSC DRV BESEDVED SOFT DESC							
17.000												
3.7 (0.03)		CONFIGURATION	I2S_LE	NGTH								
4/0.04  AUTOMUTE TIME												
S												
10   10   10   10   10   10   10   10	4 / UXU4		ALITOMLITE	ı		AUTO	VIUTE_TIME					
6 (1006   CONTROL 3 & JOHN JULY   DERMPH   BYPASS   DEEMPH   SEL   RESERVED   VOL, RATE	5 / 0x05	_LEVEL					AUTOMUTE_LEV	EL				
	6 / 0x06	CONTROL 3 & DE-EMPHASIS			DEEMF	PH_SEL	RESERVED		VOL_RATE			
9 (0.009	7 / 0x07		RESERVED	FILTER_	SHAPE	RESERVED	IIR_	WR	V	IUTE		
9/0x09   RESERVED	8 / 0x08			RESER\	/ED			GPIO1_	_CFG			
11   10   10   10   10   10   10   10	9 / 0x09	RESERVED				RESERVED	FOR REVISION V					
11   1   1   1   1   1   1   1   1	10 / 0x0A			CLOCK_DIVID	ER_SELECT			STOP	_DIV			
12 / 0x0C	11 / 0x0B	CHANNEL			SPDIF_SEL				CH2_SEL	CH1_SEL		
13   0x0D	12 / 0x0C	DPLL/ASRC		DPLL_BW	/_I2S			_	V_DSD			
15 / 000F	13 / 0x0D	THD	RESERVED	BYPASS_THD			RES	ERVED				
16   0x10	14 / 0x0E		SOFT_START				;	SOFT_START_TIMI	<b>=</b>			
17/0x11	15 / 0x0F					VC	LUME 1					
18 / 0x12	16 / 0x10	VOLUME 2				VC	LUME 2					
19 / 0x13												
19   19   19   19   19   19   19   19		MASTER TRIM				MΔS	TER TRIM					
21 / 0x15   SELECTION & OSF   SPASS   GPIO_INPUT_SEL2   GPIO_INPUT_SEL1   RESERVED   BYPASS_IIR   RESERVED   BYPASS_OSF		WASTER TRIVI				IVIAG	ILIX_TIXIIVI					
21 / 0x16	20 / 0x14				•		,	,				
22 / 0x16	21 / 0x15	SELECTION & OSF	GPIO_INP	UT_SEL2	GPIO_INF	PUT_SEL1	RESERVED	BYPASS_IIR	RESERVED	BYPASS_OSF		
23 / 0x17	22 / 0x16						l .			l .		
24/0x18   3RD HARMONIC   COMPENSATION   COMPENSATION   COEFFICIENTS	23 / 0x17					THD_	COMP_C2					
COEFFICIENTS   PROGRAMMABLE   PROG_COEFF   STAGE   PROG_COEFF_ADDR	24 / 0x18	3RD HARMONIC				THD	COMP C3					
26 / 0x1	25 / 0x19	COEFFICIENTS	2200 00555	I		1110_						
28 / 0x1C         FILTER         PROG_COEFF           29 / 0x1D         COEFFICIENT         RESERVED         EVEN_STAGE2 COEFF         PROG_ COEFF_EN           30 / 0x1E         PROGRAMMABLE FILTER CONTROL         RESERVED         EVEN_STAGE2 COEFF_WE         COEFF_EN           Read Only         64 / 0x40         CHIP STATUS         RESERVED         CHIP_ID         AUTOMUTE STATUS         LOCK_STATUS           65 / 0x41         GPIO STATUS         RESERVED         GPIO_[[0]         GPIO_[[0]           66 / 0x42 67 / 0x43 68 / 0x445         DPLL_RATIO         DPLL_NUM         DPLL_NUM         SPDIE CHANNIEL STATUS         SPDIE CHANNIEL STATUS		FILTER ADDRESS					PROG_COEFF_AD	DDR				
RESERVED   RESERVED   REVISION   RESERVED   REVISION   CHIP_ID   AUTOMUTE   STATUS   CHANNEL STATUS   CHANNEL STATUS   SPDIE CHANNEL STATUS   STATUS   STATUS   SPDIE CHANNEL STATUS   STATUS   SPDIE CHANNEL STATUS   SPDIE CHANNE						PRO	G_COEFF					
RESERVED   COEFF_WE   COEFF_EN	29 / 0x1D											
Read Only           64 / 0x40         CHIP STATUS         RESERVED         REVISION         CHIP_ID         AUTOMUTE _STATUS         LOCK_STATUS           65 / 0x41         GPIO STATUS         RESERVED         GPIO_I[0]           66 / 0x42         67 / 0x43         DPLL_RATIO         DPLL_NUM           68 / 0x44         69 / 0x45         SPDIE CHANNIEL STATUS	30 / 0x1E											
65 / 0x41	Read Only											
65 / 0x41	64 / 0x40	CHIP STATUS	RESE	RVED								
66 / 0x42 67 / 0x43 68 / 0x44 68 / 0x44 70-93 / CHANNEL STATUS	65 / 0x41	GPIO STATUS										
67 / 0x43 68 / 0x44 69 / 0x45 70-93 / CHANNIEL STATUS				1 010_[0]								
70-93 / CHANNIEL STATUS	68 / 0x44	DPLL RATIO	DPLL_NUM									
1 11/46-11/511		CHANNEL STATUS		SPDIF CHANNEL STATUS								



# **REGISTER SETTINGS**

## **Register #0: System Settings**

8 bit, Read-Write Register, Default = 0x00

o bit, i toda i i i ito giotor, p ordant oxos								
Bits	[7] [6] [5] [4]		[3] [2] [1]			[0]		
Mnemonic	osc_drv				res	serve	d *	soft_reset
Default	0	0	0	0	0	0	0	0

Bit	Mnemonic	Description
[7:4]	osc_drv	Oscillator drive specifies the bias current to the oscillator pad.  4'b0000: full bias (default)  4'b1000: 3/4 bias  4'b1100: 1/2 bias  4'b1110: 1/4 bias  4'b1111: shut down the oscillator  Other settings: reserved  It is recommended to use the default setting.
[3:1]	reserved *	
[0]	soft_reset	1'b0 is normal operation (default) 1'b1 resets chip

<sup>\*</sup> All Reserved Bits in Register #0 must be set to the indicated logic level to ensure correct device operation.

## **Register #1: Input Configuration**

8 bit, Read-Write Register, Default = 0x8C

, ,								
Bits	[7]	[6]	[5]	[4]	[3]	[3] [2]		[0]
Mnemonic	i2s_le	ength	i2s_r	node	auto_input_select		input_	select
Default	1	0	0	0	1 1		0	0

Bit	Mnemonic	Description
		2'd0 = 16bit
[7:6]	i2s_length	2'd1 = 24bit
		2'd2 or 2'd3 = 32bit (default)
		2'd0 = I <sup>2</sup> S (default)
[5:4]	i2s_mode	2'd1 = LJ mode
[3.4]		$2'd2 = I^2S$
		2'd3 = LJ mode
		2'd0 = 'input select',
[3:2]	auto_input_select	$2'd1 = I^2S$ or DSD,
[3.2]	auto_iriput_select	$2'd2 = I^2S$ or SPDIF,
		2'd3 = I <sup>2</sup> S, SPDIF or DSD (default)
		$2'd0 = I^2S$ (default)
[4.0]	input coloct	2'd1 = SPDIF
[1:0]	input_select	2'd2 = reserved
		2'd3 = DSD



## Register #2: Reserved

8 bit, Read-Write Register, Default = 0x18

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]
Mnemonic		Reserved						
Default	0	0	0	1	1	0	0	0

# Register #3: Reserved

8 bit, Read-Write Register, Default = 0x10

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]
Mnemonic				Rese	erved			
Default	0	0	0	1	0	0	0	0

## Register #4: Soft Volume Control 1 (Automute Time)

8 bit, Read-Write Register, Default = 0x00

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]
Mnemonic		automute_time						
Default	0 0 0 0 0 0 0 0							0

Bit	Mnemonic	Description
[7:0] a	automute time	Default of 8'd0 (Automute Disabled)
[7.0]	automute_time	Time in Seconds = 2096896 / (automute_time x DATA_CLK) with DATA_CLK in Hz

# Register #5: Soft Volume Control 2 (Automute Level)

8 bit, Read-Write Register, Default = 0x68

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Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]
Mnemonic	automute_loopback		;	autor	nute_	leve		
Default	0	1	1	0	1	0	0	0

Bit	Mnemonic	Description
[7]	automute_loopback	1'b0 disables automute_loopback (default) 1'b1 ramps to -infinity on automute
[6:0]	automute_level	The level (in 1dB increments) of the automute, default of 7'd104



## Register #6: Soft Volume Control 3 and De-emphasis

8 bit, Read-Write Register, Default = 0x4A

Bits	[7]	[6]		[4] [3]		[2]	[1]	[0]
Mnemonic	spdif_auto_deemph	deemph_bypass	deemph_sel reserved *		V	ol_rat	te	
Default	0	1	0	0	1	0	1	0

Bit	Mnemonic	Description
[7]	spdif_auto_deemph	1'b0 disables automatic de-emphasis select in SPDIF mode (default)
[,]	opan_aato_accmpn	1'b1 enables automatic de-emphasis select in SPDIF mode
[6]	deemph_bypass	1'b0 enables de-emphasis filters
[O]	ueempn_bypass	1'b1 disabled de-emphasis filters (default)
	doomah ool	2'b00 = 32kHz (default)
[5:4]		2'b01 = 44.1kHz
[5.4]	deemph_sel	2'b10 = 48kHz
		2'b11 = RESERVED
[3]	reserved *	Must be left as 1'b1 for normal operation
[0:0]	vol roto	3'd2 by default
[2:0]	vol_rate	Sets the volume ramp rate to 0.0078125 x fs / 2(vol_rate-5) dB/s

<sup>\*</sup> All Reserved Bits in Register #6 must be set to the indicated logic level to ensure correct device operation.

## **Register #7: General Settings**

8 bit. Read-Write Register. Default = 0x80

o bit, itoua iii	nto rtogiotor,	Doiagi	. 0,10	<u> </u>				
Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]
Mnemonic	reserved	filter_shape		reserved *	iir_bw		mı	ıte
Default	1	0	0	0	0	0	0	0

Bit	Mnemonic	Description
[7]	reserved *	
[6:5]	filter_shape	2'd0 = fast rolloff (default) 2'd1 = slow rolloff 2'd2 = minimum phase 2'd3 = reserved
[4]	reserved	
[3:2]	iir_bw	2'd0 = 1.0757 x fs or 47.44kHz (fs = 44.1kHz) - Normal mode (default) 2'd1 = 1.1338 x fs or 50kHz (fs = 44.1kHz) 2'd2 = 1.3605 x fs or 60kHz (fs = 44.1kHz) 2'd3 = 1.5873 x fs or 70kHz (fs = 44.1kHz)
[1:0]	mute	This is a soft mute, which uses the ramping volume control. mute[0]  1'b0: Channel 1 (default of left channel) unmuted (default)  1'b1: Channel 1 (default of left channel) muted mute[1]  1'b0: Channel 2 (default of right channel) unmuted (default)  1'b1: Channel 2 (default of right channel) muted

<sup>\*</sup> All Reserved Bits in Register #7 must be set to the indicated logic level to ensure correct device operation.



# **Register #8: GPIO Configuration**

8 bit, Read-Write Register, Default = 0x10

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]	
Mnemonic		rese	rved	gpio1_cfg					
Default	0	0	0	1	0	0	0	0	

Bit	Mnemonic	Description
[7:4]	reserved	
[3:0]	gpio1_cfg	Set GPIO 1 configuration Default to 4'd0 (Automute Status). See GPIO Configuration Table below for meaning of all settings.

#### **GPIO Configuration Table**

Setting	Direction	GPIO Function
4'd0	Output	Automute status (active high)
4 00	Output	asserted when Automute condition is met
4'd1	Output	DPLL Lock status (active high)
441	Output	– asserted when DPLL is in lock
	_	Minimum Volume (active high)
4'd2	Output	- asserted when volume of both the left and right channels has ramped to its
		minimum value (–127.5dB).
4'd3	Output	MCLK
		DPLL Lock interrupt (active high)
4'd4	Output	- asserted when DPLL Lock status changes state
		- reading register 64 clears the interrupt
		Automute Interrupt (active high)
4'd5	Output	- asserted when Automute status changes state
		- reading register 64 clears the interrupt
		DPLL Lock or Automute interrupt (active high)
4'd6	Output	- asserted when DPLL Lock or Automute status changes state
		- reading register 64 clears the interrupt
4'd7	Output	Output low
4'd8	Input	Use as input pin - pin status can be read from register 65.
4'd9	Input	Input Selection - uses the GPIO as an input select based on register 21
4'd15	Output	Output high

## Register #9: Reserved

8 bit, Read-Write Register, Default = 0x0

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]	
Mnemonic	Reserved for Revision V								
Default	0	0	0	0	0	0	0	0	



## **Register #10: Master Mode Control**

8 bit, Read-Write Register, Default = 0x5

Bits	[7]	[6] [5]		[4]	[3]	[2]	[1]	[0]
Mnemonic	master_clock_enable	clock_divider_select		ock_divider_select sync_mode		stop	_div	
Default	0	0	0	0	0	1	0	1

Bit	Mnemonic	Description
[7]	master_clock_enable	1'b0 disables master mode (default)
[']	master_clock_chable	1'b1 enables master mode (driving Bit clock and Frame Clock)
		2'b00: Bit Clock frequency = MCLK / 4 (default)
		2'b01: Bit Clock frequency = MCLK / 8
[6:5]	clock_divider_select	2b10: Bit Clock frequency = MCLK / 16
[0.5]	Clock_divider_select	2'b11: Bit Clock frequency = MCLK / 16
		Frame Clock frequency = Bit Clock frequency / 64
		1'b1 to enable quick lock if the fs and MCLK are synchronous and MCLK is 128 x FSR.
[4]	sync_mode	1'b0 for normal operation of the DPLL and ASRC.
[,,]	dyno_mode	Note: swink look oon only be used in DCM normal mode
		Note: quick lock can only be used in PCM normal mode.
		Sets the number of FSR edges that must occur before the DPLL and
		ASRC can lock on to the incoming signal.
		4'd0 = 16384 FSR edges
		4'd1 = 8192 FSR edges
		4'd2 = 5461 FSR edges
		4'd3 = 4096 FSR edges
		4'd4 = 3276 FSR edges
		4'd5 = 2730 FSR edges (default)
[3:0]	stop_div	4'd6 = 2340 FSR edges
	. –	4'd7 = 2048 FSR edges
		4'd8 = 1820 FSR edges
		4'd9 = 1638 FSR edges
		4'd10 = 1489 FSR edges
		4'd11 = 1365 FSR edges
		4'd12 = 1260 FSR edges
		4'd13 = 1170 FSR edges
		4'd14 = 1092 FSR edges
		4'd15 = 1024 FSR edges

For correct operation, master mode should only be enabled when the DAC's input mode is set to I<sup>2</sup>S, and when i2s\_length is set to 32-bit and i2s\_mode is set to I<sup>2</sup>S in register 1.

When master mode is enabled, the DATA\_CLK pin will output Bit Clock and the DATA1 pin will output Frame Clock at frequencies specified by clock divider select.

For compatibility with Rev. W, or when PCM data with FSR > 96kHz is used, stop\_div should be set to 4'd0 (16384 FSR edges).

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# Register #11: Channel Mapping

8 bit, Read-Write Register, Default = 0x02

Bits	[7]	[6] [5] [4]		[4]	[3]	[2]	[1]	[0]
Mnemonic	reserved *	spdif_sel			ch2_analog_swap	ch1_analog_swap	ch2_sel	ch1_sel
Default	0	0	0 0 0		0	0	1	0

Bit	Mnemonic	Description
[7]	reserved *	
[6:4]	spdif_sel	select the spdif data source 3'd0 = DATA_CLK (default) 3'd1 = DATA2 3'd2 = DATA1 3'd3 = GPIO1 3'd4-7: reserved
[3]	ch2_analog_swap	1'b0 = normal operation (default) 1'b1 = swap dac and dacb
[2]	ch1_analog_swap	1'b0 = normal operation (default) 1'b1 = swap dac and dacb
[1]	ch2_sel	1'b0 = left 1'b1 = right (default)
[0]	ch1_sel	1'b0 = left (default) 1'b1 = right

<sup>\*</sup> All Reserved Bits in Register #11 must be set to the indicated logic level to ensure correct device operation.

Left and Right channels can be reversed using Register #11.

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## **ES9016K2M Datasheet**

#### Register #12: DPLL/ASRC Settings

8 bit, Read-Write Register, Default = 0x5A

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]		
Mnemonic	C	lpll_b	w_i2	S	dpll_bw_dsd					
Default	0	1	0	1	1	0	1	0		

Bit	Mnemonic	Description
		DPLL bandwidth setting for I <sup>2</sup> S and SPDIF modes (16 settings) 4'b0000 : OFF 4'b0001 : Lowest Bandwidth
[7:4]	dpll_bw_i2s	4'b0101 : (default)
		4'b1010 :
		▼ 4'b1111 : Highest Bandwidth
		DPLL bandwidth setting for DSD mode (16 settings) 4'b0000 : OFF
		4'b0001 : Lowest Bandwidth
[3:0]	dpll_bw_dsd	4'b0101 :
		4'b1010 : (default)
		4'b1111 : Highest Bandwidth

## **Register #13: THD Compensation**

8 bit, Read-Write Register, Default = 0x40

Bits	[7]	[6]	[5] [4] [3] [2] [1]			[1]	[0]
Mnemonic	reserved *	bypass_thd	reserved *				
Default	0	1	0 0 0 0 0		0	0	

Bit	Mnemonic	Description
[7]	reserved *	
[6]	bypass_thd	1'b0: enable THD compensation  output = input + (input²) x thd_comp_c2 + (input³) x thd_comp_c3  thd_comp_c2 is stored in registers 23-22 (16 bits signed) (register 23 stores MSBs)  thd_comp_c3 is stored in registers 25-24 (16 bits signed) (register 25 stores MSBs)  1'b1: disable THD compensation (default)  PCM mode: output = input; DSD mode: output = input / 2
[5:0]	reserved	

<sup>\*</sup> All Reserved Bits in Register #13 must be set to the indicated logic level to ensure correct device operation. THD compensation can be used to reduce the 2<sup>nd</sup> and 3<sup>rd</sup> harmonic distortion introduced by external output drivers. A system level tuning is required to arrive at the optimum coefficients for thd\_comp\_c2 and thd\_comp\_c3.

#### Notes:

- To get the same gain (output = input) for PCM and DSD modes without THD compensation, bypass\_thd should be set to 1'b0 with thd\_comp\_c2 and thd\_comp\_c3 set to 16'd0 (default)
- Erroneous compensation can lead to higher distortion than the one without compensation. If accurate tuning cannot be performed, thd comp c2 and thd comp c3 should be set to 16'd0 (default) if bypass thd is set to 1'b0.



## Register #14: Soft Start Settings

8 bit, Read-Write Register, Default = 0x8A

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]
Mnemonic	soft_start	soft_start_on_lock	mute_on_lock		soft_	start_	_time	
Default	1	0	0	0	1	0	1	0

Bit	Mnemonic	Description
[7]	coft ctart	1'b0: Ramp the output stream to ground
[/]	[7] soft_start	1'b1: Normal operation (default) - ramp the output stream to ½ x AVCC_L/R
[6]	soft start on lock	1'b0: Do not force output low when lock is lost (default)
[6]	SUIT_STAIT_UIT_TOCK	1'b1: Force output low when lock is lost
[5]	mute_on_lock	1'b0: Do not force a mute when lock is lost (default)
[5]	IIIule_oii_lock	1'b1: Force a mute when lock is lost
		Time for soft start ramp
[4:0]	soft start time	= 4096 x 2 <sup>(soft_start_time+1)</sup> / MCLK seconds (where MCLK is measured in Hz).
[4.0]	SUIL_SIGIT_IIIIE	
		The valid range of soft-start_time is from 0 to 20.

## Register #15: Volume 1 (usually selected for the Left Channel, but can be reversed using Register #11)

8 bit. Read-Write Register, Default = 0x00

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]
Mnemonic	volume1							
Default	0	0	0	0	0	0	0	0

Bit	Mnemonic	Description
[7:0]	volume1	Default to 8'd0
	volume i	0dB to -127.5dB in 0.5dB steps

#### Register #16: Volume 2 (usually selected for the Right Channel, but can be reversed using Register #11)

8 bit, Read-Write Register, Default = 0x00

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]
Mnemonic				volu	me2			
Default	0	0	0	0	0	0	0	0

Bit	Mnemonic	Description
[7:0]	volume2	Default to 8'd0 0dB to -127.5dB in 0.5dB steps

## Register #20-17: Master Trim

32 bit, Read-Write Register, Default = 32'h7ffffff. Reg 20 are the MSB's, Reg 17 are the LSB's.

Bits	[31:0]
Mnemonic	master_trim
Default	32'h7fffffff

This is a 32 bit value that sets the 0dB level for all volume controls. This is a signed number, so it should never exceed 32'h7fffffff (which is  $2^{31} - 1$ ).



#### Register #21: GPIO Input Selection and OSF Bypass

8 bit, Read-Write Register, Default = 0x00

Bits	[7	:6]	[5:4]		[3]	[2]	[1]	[0]
Mnemonic	gpio_in	put_sel2	gpio_input_sel1		reserved *	bypass_iir	reserved *	bypass_osf
Default	0	0	0	0	0	0	0	0

Bit	Mnemonic	Description
[7:6]	gpio_input_sel2	Selects which input will be selected when GPIOX = 1'b1 2'd0 = I <sup>2</sup> S data (default) 2'd1 = SPDIF data 2'd2 = reserved 2'd3 = DSD data
[5:4]	gpio_input_sel1	Selects which input will be selected when GPIOX = 1'b0 2'd0 = I <sup>2</sup> S data (default) 2'd1 = SPDIF data 2'd2 = reserved 2'd3 = DSD data
[3]	reserved *	
[2]	bypass_iir	1'b0 = Use the IIR filter (default) 1'b1 = Bypass the IIR filter.
[3]	reserved *	
[0]	bypass_osf	1'b0 = Use the interpolating 8x FIR filter (default) 1'b1 = Bypass the interpolating 8x FIR filter.  Note: Bypassing the interpolating filter requires that the input data be oversampled at 8x fs by an external oversampling filter.

<sup>\*</sup> All Reserved Bits in Register #21 must be set to the indicated logic level to ensure correct device operation.

Note: Any of the GPIO can be configured to be used as an input select. This allows an external MCU or controller to set the input type by setting the GPIO to either logic high (1'b1) or logic low (1'b0). To set this feature, the first step is to enable one of the GPIO as an input select by setting gpio\_cfg to 4'd9. Once a GPIO is configured as an input select it has the ability to select between two different inputs. The first input (logic low) is set via register 21[5:4]. The second input (logic high) is set via register 21[7:6].

# Register #23-22: 2<sup>nd</sup> Harmonic Compensation Coefficients

16 bit, Read-Write Register, Default = 0x0000 (no compensation). Register #23 is MSB. See Register #13 for more details.

Bits	[15:0]
Mnemonic	Thd_comp_c2
Default	16'd0

# Register #25-24: 3<sup>rd</sup> Harmonic Compensation Coefficients

16 bit, Read-Write Register, Default = 0x0000 (no compensation). Register #25 is MSB. See Register #13 for more details.

Bits	[15:0]
Mnemonic	Thd_comp_c3
Default	16'd0



## Register #26: Programmable Filter Address

8 bit, Read-Write Register, Default = 0x00

Bits	[7]	[6:0]						
Mnemonic	prog_coeff_stage		pro	og_c	coef	f_ac	ddr	
Default	0	0 0 0 0 0 0 0					0	

Bit	Mnemonic	Description
		Selects which stage of the filter to write.
[7]	prog_coeff_stage	1'b0 = Stage 1 of the oversampling filter (128 coefficients)
		1'b1 = Stage 2 of the oversampling filter (16 coefficients)
[6:0]	prog coeff oddr	Selects the coefficient address when writing custom coefficients
[6:0]	prog_coeff_addr	for the oversampling filter.

## Register #29-27: Programmable Filter Coefficient

8 bit, Read-Write Register, Default = 0x000000

Bits	[23:0]				
Mnemonic	prog_coeff				
Default	24'd0				

Bit	Mnemonic	Description
[23:0]	prog_coeff	A 24bit filter coefficients that will be written to address 'prog_coeff_addr'.

#### Register #30: Programmable Filter Control

8 bit. Read-Write Register, Default = 0x00

Bits	[7:3]			]		[2]	[1]	[0]	
Mnemonic	reserved *			even_stage2_coeff	prog_coeff_we	prog_coeff_en			
Default	0 0 0 0 0		0	0	0	0			

Bit	Mnemonic	Description
[7:3]	reserved *	
[2]	even_stage2_coeff	Sets the type of symmetry of the stage 2 programmable filter.  1'b0 = Uses a sine symmetric filter (27 coefficients).  1'b1 = Uses a cosine symmetric filter (28 coefficients).
[1]	prog_coeff_we	1'b0 = Disable writing to the custom filter coefficients.  1'b1 = Enable writing to the custom filter coefficients.  Note: When set to 1'b1 the custom filter will be bypassed regardless of the state of register 21[0].
[0]	prog_coeff_en	1'b0 = Use one of the built-in oversampling filters. 1'b1 = Use the custom oversampling filter. Note: The custom filter is not programmed to anything on reset, valid coefficients must be written to the filter before enabling.

<sup>\*</sup> All Reserved Bits in Register #30 must be set to the indicated logic level to ensure correct device operation.

Note: even\_stage2\_coeff sets the type of symmetry used by the second stage filter. The actual RAM is 16 coefficients, but only the first 14 coefficients are used when applying the oversampling filter. The first 14 coefficients are mirrored using either sine or cosine symmetry, resulting in a filter length of either 27 or 28 taps. This means that the second stage RAM should only contain half of the impulse response of the second stage filter, and the impulse peak value will be contained in the 14th coefficient. Also note that, due to the symmetry of the filter, only linear phase filters may be used in the second stage.



#### Register #64: Chip Status

8 bit, Read-Only Register

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]
Mnemonic	Rese	erved	revision	С	hip_i	d	automute_status	lock_status

Bit	Mnemonic	Description
[7:6]	Reserved	
[5]	revision	1'b0 => revision W. 1'b1 => revision V.
[4:2]	chip_id	3'd6 => ES9016K2M
[1]	automute_status	1'b0 => Automute condition is inactive. 1'b1 => Automute condition is active.
[0]	lock_status	1'b0 => The Jitter Eliminator is not locked to an incoming signal. 1'b1 => The Jitter Eliminator is locked to an incoming signal.

#### Register #65

8 bit, Read-Only Register

Bits	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]
Mnemonic				gpio_l[1:0]				

Bit	Mnemonic	Description
[7:2]	reserved	
[0]	gpio_I[0]	Status of pin GPIO1

#### Register #69-66: DPLL Ratio

32 bit, Read-Only Register. Reg 69 are the MSB's, Reg 66 are the LSB's

Bits	[31:0]
Mnemonic	dpll num

This is a read-only 32bit value that can be used to calculate the sample rate. The raw sample rate (FSR) can be calculated using: FSR = (DPLL\_NUM x  $F_{MCLK}$ ) /  $2^{32}$ .

Note that the DPLL number (register 66-69) should be read from LSB to MSB as it is latched on the LSBs (register 66).

#### Register #93-70: Channel Status

Reg 93 are the MSB's, Reg 70 are the LSB's Format is [191:0]

These registers allow read back of the SPDIF channel status. The status definition is different for the consumer configuration and professional configuration. Please refer to the following two tables for details.



SPDIF CHANNEL STATUS - Consumer configuration											
Address Offset	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]			
0	Reserved	Reserved	0:2Channel 1:4Channel	Reserved	0:No-Preemph 1:Preemph	0:CopyRight 1:Non-CopyRight	0:Audio 1:Data	0:Consumer 1:Professional			
1	0x05: Musion 0x06: Preson 0x08: Solid	eral r-Optical Converter netic al Broadcast cal Instrumer ent A/D Conv State Memo re A/D Conve	erter								
2	Channel Nic 0x0: Don't ( 0x1: A (Lefi 0x2: B (Rig 0x3: C 0x4: D 0x5: E 0x6: F 0x7: G 0x8: H 0x9: I 0xA: J 0xB: K 0xC: L 0xD: M 0xE: N 0xF: O	umber Care t)			Source Number 0x0: Don't Care 0x1: 1 0x2: 2 0x3: 3 0x4: 4 0x5: 5 0x6: 6 0x7: G 0x8: 8 0x9: 9 0xA: 10 0xB: 11 0xC: 12 0xD: 13 0xE: 14 0xF: 15						
3	Reserved	Reserved	Clock Accuracy 0x0:Level 2 +-10 0x1:Level 1 +-50 0x2:Level 3 varia	ppm	Sample Frequer 0x0: 44.1k 0x2: 48k 0x3: 32k 0x4: 22.05k 0x6: 24k 0x8: 88.2k 0xA: 96k 0xC: 176.4k 0xE: 192k	icy					
4	Reserved	Reserved	Reserved	Reserved	Word Length:	te=0  If Word Field S ed  000=Not indicat  100 = 19bits  010 = 18bits  110 = 17bits  001 = 16bits  101 = 20bits		Word Field Siz 0:Max 20bits 1:Max 24bits			

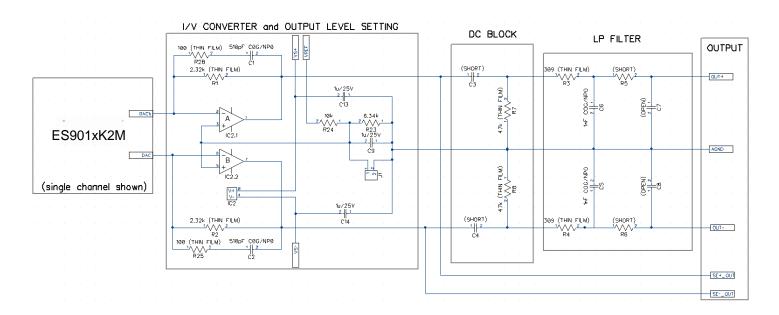


	SPDIF CHANNEL STATUS - Professional configuration								
Address Offset	[7]	[6]	[5]	[4]	[3]	[2]	[1]	[0]	
0	sampling frequency: 00: not indicated (or see byte 4) 10: 48kHz 01: 44.1kHz 11: 32kHz		lock: 0: locked 1: unlocked	emphasis: 000: Emphasis not in 001: No emphasis 011: CD-type emphasis 111: J-17 emphasis		asis	0:Audio 1:Non-audio	0:Consumer 1:Professional	
1	User bit management: 0000: no indication 1000: 192-bit block as channel status 0100: As defined in AES18 1100: user-defined 0010: As in IEC60958-3 (consumer)					Channel mode: 0000: not indicated (default to 2 ch) 1000: 2 channel 0100: 1 channel (monophonic) 1100: primary / secondary 0010: stereo 1010: reserved for user applications 0110: reserved for user applications 1110: SCDSR (see byte 3 for ID) 0001: SCDSR (stereo left) 1001: SCDSR (stereo right) 1111: Multichannel (see byte 3 for ID)			
2	alignment level:  00: not indicated  10: -20dB FS  01: -18.06dB FS  010 = 22bits  100 = 22bits  100 = 22bits  100 = 20bits  100 = 24bits  101 = 24bits			If max =	: 24bits indicated bits bits bits bits	Use of aux sample word: 000: not defined, audio max 20 bits 100: used for main audio, max 24 bits 010: used for coord, audio max 20 bits 110: reserved			
3	Channel identification:  if bit 7 = 0 then channel nu  if bit 7 = 1 then bits 4–6 de		s 1 plus the numeric v	alue of bits	s 0-6 (bit re		onel number within	that mode	
4	fs scaling: 0: no scaling 1: apply factor of 1 / 1.001 to value	ing 0000: not indicated 00: not a DARS actor of 0001: 24kHz 00: not a DARS oracle 2 (±10 p						udio reference signal): e 2 (±10 ppm)	
5	Reserved								
6-9	alphanumerical channel o	rigin: fo	ur-character label usi	ng 7-bit AS	CII with no	parity. Bits 55, 6	63, 71, 79 = 0.		
10-13	alphanumerical channel d	estinati	on: four-character lab	el using 7-	oit ASCII w	ith no parity. Bits	8 87, 95, 103, 111 =	= 0.	
14-17	local sample address code: 32-bit binary number representing the sample count of the first sample of the channel status block.								
18-21	time of day code: 32-bit bi	nary nu	mber representing tin	ne of sourc	e encoding	in samples sinc	e midnight		
22	reliability flags 0: data in byte range is rel 1: data in byte range is un								
23	CRCC 00000000: not implement X: error check code for bit		3						

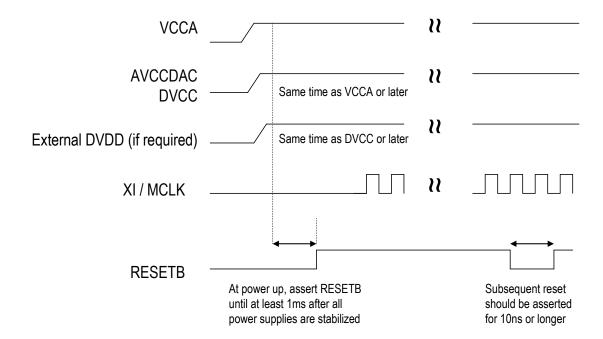
# **ES9016K2M Datasheet**



## **APPLICATION DIAGRAM**



## RECOMMENDED POWER-UP SEQUENCE





# **ABSOLUTE MAXIMUM RATINGS**

PARAMETER	RATING
Storage temperature	–65°C to +105°C
Voltage range for digital input pins	-0.3V to DVCC+ 0.3V
ESD Protection	
Human Body Model (HBM)	2000V
Machine Model (MM)	200V

**WARNING:** Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute—maximum—rated conditions for extended periods may affect device reliability.

WARNING: Electrostatic Discharge (ESD) can damage this device. Proper procedures must be followed to avoid ESD when handling this device.

#### RECOMMENDED OPERATING CONDITIONS

PARAMETER	SYMBOL	CONDITIONS
Operating temperature	T <sub>A</sub>	−20°C to +70°C

Power Supply		Voltage	Current nominal (Note 1)	Current standby (Notes 1, 2)
Digital Power Supply Voltage	DVCC	+1.8V ± 5% +3.3V ± 5%	13.0mA 14.2mA	0mA 0mA
Internal Digital Core supply	DVDD	+1.2V (typical)		
External Digital Core Supply	DVDD	+1.3V ± 5% (Note 3)	50mA	
Analog Core Supply Voltage	VCCA (Note 4)	+3.3V ± 5% +1.8V ± 5%	0.8mA	0mA
Analog Power Supply Voltage	AVCCDAC (Note 4)	+3.3V ± 5% +1.8V ± 5%	8.0mA	0mA
Total Power		DVCC = +1.8V DVCC = +3.3V	≤ 36mW ≤ 59mW	< 1mW < 1mW

#### **Notes:**

- (1) fs = 44.1kHz, external MCLK = 22MHz,  $I^2S$  input, DAC output connected to current-to-voltage converter, internal DVDD, all external supply voltages at nominal center values
- (2) With RESETB held low after setting the soft\_start bit in register 14 to 1'b0 to fully ramp the DAC outputs to ground
- (3) Internal DVDD should be used except under the conditions described on page 7. External DVDD current measured at 192kHz sample rate and MCLK = 80MHz.
- (4) For correct operation, VCCA  $\geq$  AVCCDAC.

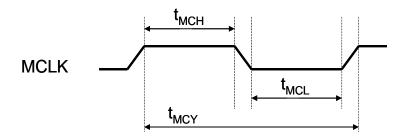
## DC ELECTRICAL CHARACTERISTICS

Symbol	Parameter	arameter Minimum N		Unit	Comments
VIH	High-level input voltage	DVCC / 2 + 0.4		V	
VIL	Low-level input voltage		0.4	V	
VOH	High-level output voltage	DVCC - 0.2		V	IOH = 100μA
VOL	Low-level output voltage		0.2	V	IOL = 100μA

## **ES9016K2M Datasheet**

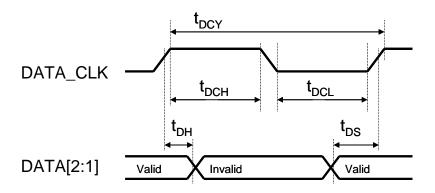


# XI / MCLK Timing



Parameter	Symbol	Min	Max	Unit
MCLK pulse width high	Тмсн	4.5		ns
MCLK pulse width low	T <sub>MCL</sub>	4.5		ns
MCLK cycle time	T <sub>MCY</sub>	10		ns
MCLK duty cycle		45:55	55:45	

# **Audio Interface Timing**



Parameter	Symbol	Min	Max	Unit
DATA_CLK pulse width high	t <sub>DCH</sub>	4.5		ns
DATA_CLK pulse width low	tDCL	4.5		ns
DATA_CLK cycle time	tDCY	10		ns
DATA_CLK duty cycle		45:55	55:45	
DATA set-up time to DATA_CLK rising edge	t <sub>DS</sub>	4.1		ns
DATA hold time to DATA_CLK rising edge	t <sub>DH</sub>	2		ns

#### Notes

- Audio data on DATA[2:1] are sampled at the rising edges of DATA\_CLK and must satisfy the setup and hold time requirements relative to the rising edge of DATA\_CLK
- For DSD Phase mode, the normal data (D0, D1, D2.. on p.10) must satisfy the setup and hold time requirements relative to the rising edge of DATA\_CLK. The complimentary data (D0, D1, etc.) will be ignored.

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## **ES9016K2M Datasheet**

## **ANALOG PERFORMANCE**

#### **Test Conditions (unless otherwise stated)**

- $1.\ T_A=25^{\circ}C,\ AVCCDAC=VCCA=DVCC=3.3V,\ internal\ DVDD\ with\ 4.7\mu F\pm 20\%\ decoupling,\ fs=44.1kHz,\ MCLK=27MHz\ \&\ 32-bit\ data=25^{\circ}C$
- 2. SNR/DNR: A-weighted over 20Hz-20kHz in averaging mode

THD+N: un-weighted over 20Hz-20kHz bandwidth

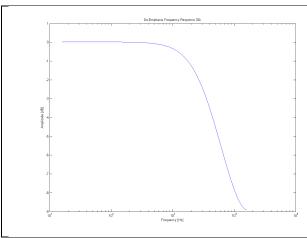
PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
Resolution			32		Bits
MCLK (PCM normal mode)	Note *3	192FSR			
MCLK (PCM OSF bypass mode)		24FSR		Note *2	LI-
MCLK (DSD mode)		3FSR		Note *2	Hz
MCLK (SPDIF mode)		386FSR			
DYNAMIC PERFORMANCE					•
DNR (differential current mode)	-60dBFS		122		dB-A
THD+N (differential current mode)	0dBFS		-110		dB
ANALOG OUTPUT					
Differential (+ or –) voltage out range (Note *4)	Full-scale out		3.05 (0.924 x AVCCDAC)		Vp-p
Differential (+ or –) voltage out offset (Note *4)	Bipolar zero out		1.65 (AVCCDAC / 2)		V
Differential (+ or –) current out range (Notes *1, *4)	Full-scale out		3.784		mAp-p
Differential (+ or –) current out offset (Notes *1, *4)	Bipolar zero out to virtual ground at voltage Vg (V)		2.047 – (1000 x Vg) / 806		mA
Digital Filter Performance					
De-emphasis error				±0.2	dB
Mute Attenuation			127		dB
PCM Filter Characteristics (Sharp Roll C	Off)				
Doorboard	±0.003dB			0.454fs	Hz
Pass band	-3dB			0.49fs	Hz
Stop band	< -115dB	0.546fs			Hz
Group Delay			35 / fs		S
PCM Filter Characteristics (Slow Roll O	ff)	•		•	•
	±0.05dB			0.308fs	Hz
Pass band	-3dB			0.454fs	Hz
Stop band	< -100dB	0.814fs			Hz
Group Delay			6.25 / fs		S
PCM Filter Characteristics (Minimum Pr	nase)	•	•		•
Daga hand	±0.003dB			0.454fs	Hz
Pass band	-3dB			0.49fs	Hz
Stop band	< -115dB	0.546fs			Hz

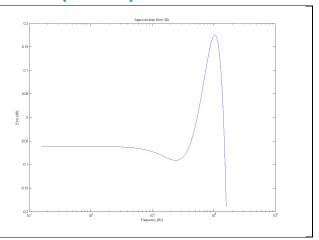
#### Notes:

- \*1. Differential (+ or –) current output is equivalent to a differential (+ or –) voltage source in series with an  $806\Omega \pm 11\%$  resistor. The differential (+ or –) voltage source has a peak-to-peak output range of 0.924 x AVCCDAC = 3.05V and an output offset of AVCCDAC / 2 = 1.65V.
- \*2. With internal DVDD, maximum MCLK frequency is 50MHz (DVCC = 1.8V), or 100MHz (DVCC = 3.3V) with an external +1.3V DVDD supply.
- \*3. Synchronous MCLK at 128 x FSR is also supported.
- \*4. Values are valid for AVCCDAC = 3.3V. For AVCCDAC = 1.8V, formulas should be used.

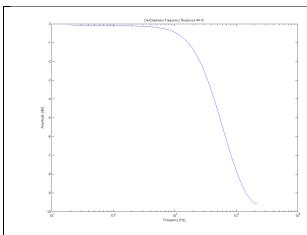


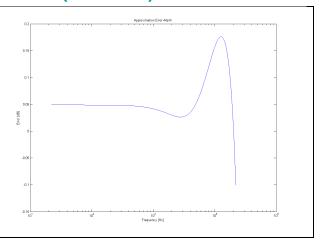
# PCM DE-EMPHASIS FILTER RESPONSE (32kHz)



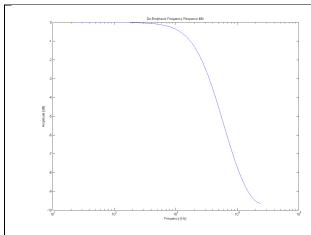


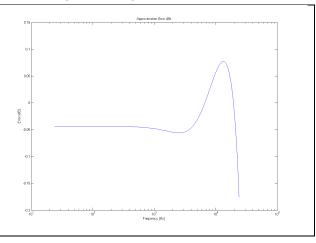
# PCM DE-EMPHASIS FILTER RESPONSE (44.1kHz)





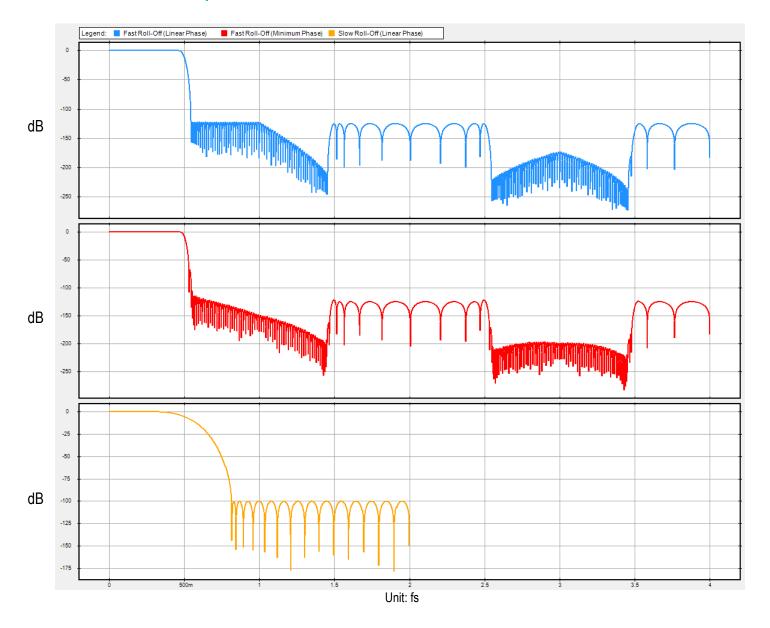
# PCM DE-EMPHASIS FILTER RESPONSE (48kHz)







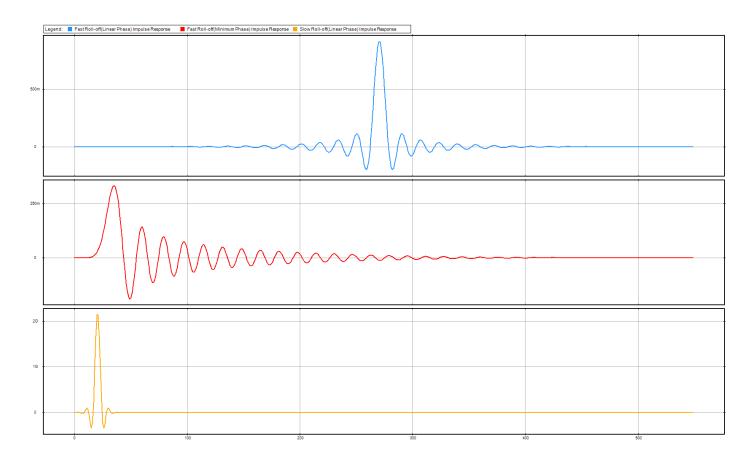
# **PCM FILTER FREQUENCY RESPONSE**



## ES9016K2M Datasheet



# **PCM FILTER IMPULSE RESPONSE**

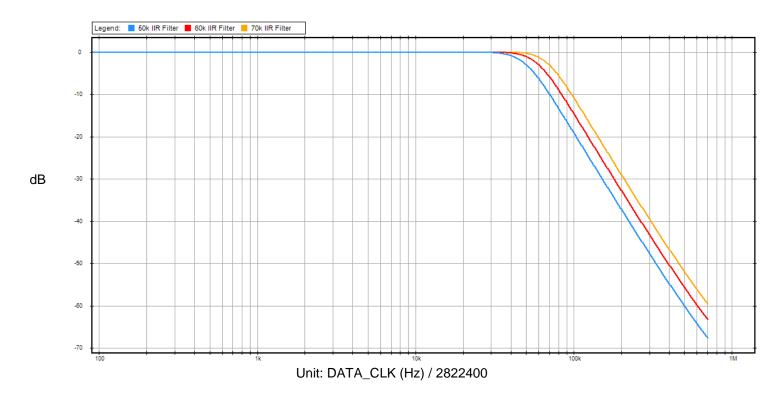


Unit: 1/fs (s)





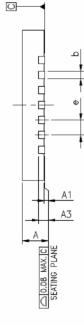
# **DSD FILTER RESPONSE**



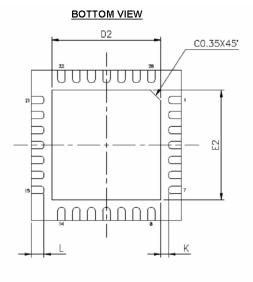


# 28-Pin QFN Mechanical Dimensions

# **TOP VIEW** PIN 1 21 15 8 14



0	<b>─</b> ¶
_	
	SEATING PLANE



- 1. ALL DIMENSIONS ARE IN MILLIMETERS.
- 2. DIMENSION b APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN 0.15mm AND 0.30mm FROM THE TERMINAL TIP. IF THE TERMINAL HAS THE OPTIONAL RADIUS ON THE OTHER END OF THE TERMINAL, THE DIMENSION b SHOULD NOT BE MEASURED IN THAT RADIUS AREA.
- 3. BILATERAL COPLANARITY ZONE APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS.

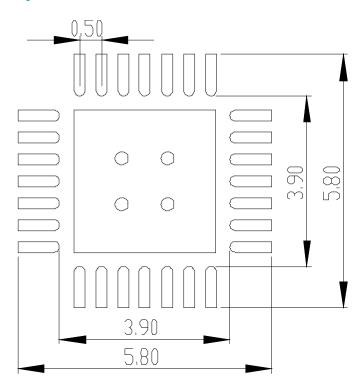
SYMBOLS	MIN.	NOM.	MAX.		
Α	0.70		0.90		
A1	0.00	0.02	0.05		
A3	0.203 REF.				
b	0.18	0.30			
D	5	.00 BS	SC OS		
E	5	5.00 BSC			
e	0.50 BSC				
K	0.20	_	_		

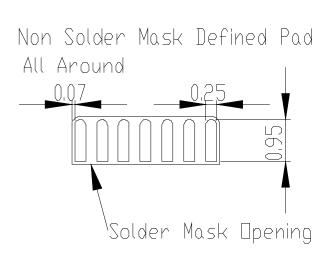
		E2			D2		L			LEAD FINISH	
PAD SIZ	E MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	Pure Tin	PPF
	2.50		3.60	2.50		3.60	0.50	0.55	0.60	V	Χ

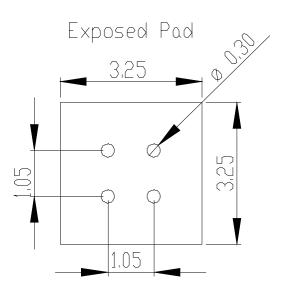
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# **Example 28-Pin QFN Land Pattern**







#### Notes:

- 1. All dimensions are in millimeters.
- 2. Thermal vias should be 0.3mm to 0.33mm in diameter, with the barrel plated to 1oz copper.
- 3. For maximum solder mask in the corners, round the inner corners of each row.
- 4. Exposed pad should be solder mask defined.
- 5. Pad width can be reduced to 0.25mm if additional pad to pad clearance is required.
- 6. For applications where solder loss through vias is a concern, plugging or tenting of the vias should be used. The solder mask diameter for each via should be 0.1mm larger than the via diameter.

#### ES9016K2M Datasheet



#### **Reflow Process Considerations**

For lead-free soldering, the characterization and optimization of the reflow process is the most important factor you need to consider.

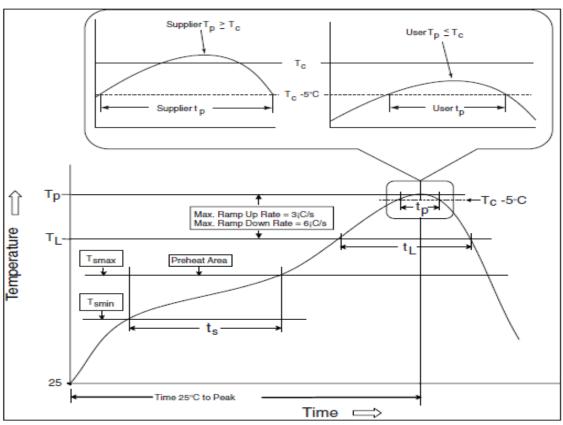
The lead-free alloy solder has a melting point of 217°C. This alloy requires a minimum reflow temperature of 235°C to ensure good wetting. The maximum reflow temperature is in the 245°C to 260°C range, depending on the package size (*Table RPC-2*). This narrows the process window for lead-free soldering to 10°C to 20°C.

The increase in peak reflow temperature in combination with the narrow process window makes the development of an optimal reflow profile a critical factor for ensuring a successful lead-free assembly process. The major factors contributing to the development of an optimal thermal profile are the size and weight of the assembly, the density of the components, the mix of large and small components, and the paste chemistry being used.

Reflow profiling needs to be performed by attaching calibrated thermocouples well adhered to the device as well as other critical locations on the board to ensure that all components are heated to temperatures above the minimum reflow temperatures and that smaller components do not exceed the maximum temperature limits (*Table RPC-2*).

To ensure that all packages can be successfully and reliably assembled, the reflow profiles studied and recommended by ESS are based on the JEDEC/IPC standard J-STD-020 revision D.1.

Figure RPC-1. IR/Convection Reflow Profile (IPC/JEDEC J-STD-020D.1)



Note: Reflow is allowed 3 times. Caution must be taken to ensure time between re-flow runs does not exceed the allowed time by the moisture sensitivity label. If the time elapsed between the re-flows exceeds the moisture sensitivity time bake the board according to the moisture sensitivity label instructions.

#### Manual Soldering:

Allowed up to 2 times with maximum temperature of 350 degrees no longer than 3 seconds.

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#### Table RPC-1 Classification reflow profile

Profile Feature	Pb-Free Assembly
Preheat/Soak	
Temperature Min (Tsmin)	150°C
Temperature Max (Tsmax)	200°C
Time (ts) from (Tsmin to Tsmax)	60-120 seconds
Ramp-up rate (TL to Tp)	3°C / second max.
Liquidous temperature (TL)	217°C
Time (tL) maintained above TL	60-150 seconds
Peak package body temperature	For users Tp must not exceed the classification temp in Table RPC-2.
(Tp)	For suppliers Tp must equal or exceed the Classification temp in Table RPC-2.
Time (tp)* within 5°C of the	
specified classification temperature	30* seconds
(Tc), see Figure RPC-1	
Ramp-down rate (Tp to TL)	6°C / second max.
Time 25°C to peak temperature	8 minutes max.
* Tolerance for peak profile temperatu	ure (Tp) is defined as a supplier minimum and a user maximum.

- Note 1: All temperatures refer to the center of the package, measured on the package body surface that is facing up during assembly reflow (e.g., live-bug). If parts are reflowed in other than the normal live-bug assembly reflow orientation (i.e., dead-bug), Tp shall be within ± 2°C of the live-bug Tp and still meet the Tc requirements, otherwise, the profile shall be adjusted to achieve the latter. To accurately measure actual peak package body temperatures refer to JEP140 for recommended thermocouple use.
- Note 2: Reflow profiles in this document are for classification/preconditioning and are not meant to specify board assembly profiles. Actual board assembly profiles should be developed based on specific process needs and board designs and should not exceed the parameters in Table RPC-1. For example, if Tc is 260°C and time tp is 30 seconds, this means the following for the supplier and the user. For a supplier: The peak temperature must be at least 260°C. The time above 255°C must be at least 30 seconds.
- For a user: The peak temperature must not exceed 260°C. The time above 255°C must not exceed 30 seconds.

Note 3: All components in the test load shall meet the classification profile requirements.

#### Table RPC-2 Pb-Free Process – Classification Temperatures (Tc)

Package Thickness	Volume mm3, <350	Volume mm3, 350 to 2000	Volume mm3, >2000
<1.6 mm	260°C	260°C	260°C
1.6 mm – 2.5 mm	260°C	250°C	245°C
>2.5 mm	250°C	245°C	245°C

- Note 1: At the discretion of the device manufacturer, but not the board assembler/user, the maximum peak package body temperature (Tp) can exceed the values specified in Table RPC-2. The use of a higher Tp does not change the classification temperature (Tc).
- Note 2: Package volume excludes external terminals (e.g., balls, bumps, lands, leads) and/or non-integral heat sinks.
- Note 3: The maximum component temperature reached during reflow depends on package thickness and volume. The use of convection reflow processes reduces the thermal gradients between packages. However, thermal gradients due to differences in thermal mass of SMD packages may still exist.

## **ES9016K2M Datasheet**



## ORDERING INFORMATION

Part Number	Description	Package
ES9016K2M	Sabre <sup>32®</sup> Ultra 32-bit Low Power Stereo Audio DAC	28-pin QFN

The letter K identifies the package type QFN.

# **Revision History**

Rev.	Date	Notes	
1.1	March 19, 2014	Update MCLK requirement	
1.2	April 15, 2014	Update sync_mode requirement	
1.3	May 28, 2014	Update DSD L/R pin assignment. Add THD compensation registers.  Update mechanical drawing and add land pattern. Update migration notes	
1.4	June 5, 2014	Added SABRE SOUND™ trademark	
1.5	July 28, 2014	Updated ESS' FAX number. Added medical usage legal disclaimer	
1.6	August 28, 2014	Added conditions when an external DVDD regulator is required	
1.7	September 8, 2014	Corrected typo on Register#7 Bit [6:5], 3'dX changed to 2'dX. Identified Left and Right channels for Registers #15 and #16 respectively. Updated DAC output impedance from $781.25\Omega$ to $806\Omega$	
1.8	September 24, 2014	Added "8 bit, Read-Write Register, Default = 0x00" heading on Register #21.  Corrected part number on revision marking diagram, page 3.  Removed reference to Right Justified data format that is not supported	
1.9	October 16, 2014	Added table to Register #65 description.	
2.0	January 8, 2015	Added details on decoupling required for the DVDD core supply.  Deleted old revision history from 0.1 to 0.91.	
2.1	April 10, 2015	Added notes on the connection of reserved Bits in the device control registers.  Added SABRE HiFi logo. Updated ESS' address and phone number.	
2.2	June 10, 2015	Increased typical value of AVCC_L plus AVCC_R from 3mA to 8mA	
2.3	December 2, 2016	Correct Recommended Operating Conditions table formatting.	
2.4	January 24, 2017	Corrected THD compensation description and Recommended Operating Conditions table formatting.	
2.5	January 31, 2017	Remove references to Revision W silicon, clarify I2C address description.	
2.6	February 14, 2017	Added description for Registers #2, #3 and #9. Register #65 labeled as GPIO Status. Added register map. Adjusted page number references as needed.	
2.7	November 14, 2018	Added Low Power Audio DAC description, removed Advanced Information	
2.8	March 13, 2019	Removed ESR capacitor requirement for DVDD. Updated SABRE®, SABRE SOUND® and Sabre <sup>32®</sup>	
2.9	December 30, 2019	Corrected SDA setup time from SCL rising units from "µs" to "ns".	
3.0	April 27. 2020	Updated analog performance	

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